

# ***Operational Amplifiers and Comparators Data Book***

***Volume A***



## **IMPORTANT NOTICE**

Texas Instruments (TI) reserves the right to make changes to its products or to discontinue any semiconductor product or service without notice, and advises its customers to obtain the latest version of relevant information to verify, before placing orders, that the information being relied on is current.

TI warrants performance of its semiconductor products and related software to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

Certain applications using semiconductor products may involve potential risks of death, personal injury, or severe property or environmental damage ("Critical Applications").

**TI SEMICONDUCTOR PRODUCTS ARE NOT DESIGNED, INTENDED, AUTHORIZED, OR WARRANTED TO BE SUITABLE FOR USE IN LIFE-SUPPORT APPLICATIONS, DEVICES OR SYSTEMS OR OTHER CRITICAL APPLICATIONS.**

Inclusion of TI products in such applications is understood to be fully at the risk of the customer. Use of TI products in such applications requires the written approval of an appropriate TI officer. Questions concerning potential risk applications should be directed to TI through a local SC sales offices.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards should be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance, customer product design, software performance, or infringement of patents or services described herein. Nor does TI warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used.



## INTRODUCTION

Texas Instruments (TI) offers an extensive line of industry-standard and leadership products dedicated to operational amplifier and comparator functions. The technologies represented in this book include traditional bipolar through BIFET, BIFET, IMPACT™, LinCMOS™, Advanced LinCMOS™, and Excalibur processes. The IMPACT, Advanced LinCMOS, and Excalibur technologies feature a step-function improvement in impedance, speed, power requirements, and threshold stability.

The Operational Amplifier/Comparator Data Books (volumes A and B) provide information on an extensive listing of TI operational amplifier and comparator products:

- Commercial, Industrial, Automotive, Military, and Extended Temperature Ranges
- Noncompensated, Single, Dual
- Internally Compensated, Single, Dual, Quad
- Precision, Copper Stabilized
- Excalibur: High Speed, Low Power, Precision, JFET Input, High Output, Low Noise

## EXCALIBUR PROCESS DISCUSSION

Excalibur is a 44-V n-epi bipolar process that includes isolated high-speed PNPs, metal-nitride-poly capacitors, p-channel JFETs, as well as the common bipolar devices. In other bipolar processes, the capacitors have one plate made from silicon substrate (bottom) and the other from metal. At low levels of operating current, the leakage current from the silicon bottom plate can significantly impact the dc performance of the circuits. The ac performance is also affected by the parasitic substrate capacitance. Use of the metal-nitride-poly capacitor significantly reduces these effects, yielding higher ac performance and stable bias currents.

## FEATURES IN THIS BOOK

- Excalibur-process devices
- Selected Macromodels (Level I)
- Expanded product characterization over supply voltage and temperature
- Extensive graphs showing the characterizations
- New 3-V devices that are specifically designed, characterized, and tested for operation at 3.3 V or less

The first section of each volume contains an alphanumeric listing, a selection guide, and a cross reference for each type of device. The alphanumeric listing in the book includes all the devices contained in volumes A and B of the Operational Amplifier/Comparator Data Book. The sections in each book are numbered consecutively across volumes (sections 1, 2, and 3 are in volume A and sections 4, 5, 6, 7 are in volume B). Thus, the reader can easily find the particular volume for a given device.

Because of the great number of devices available from TI, the selection guide for the operational amplifiers is broken down into eight primary categories with an complete alphanumeric listing at the end. The comparator selection guide is a complete alphanumeric listing. The cross references in section one help to identify devices that are comparable to other manufacturers and older TI parts.

The last section in each volume contains ordering information and mechanical data for the devices in that particular volume.

While these volumes offer information only on the operational amplifier and comparator devices available now from TI, complete technical data for upcoming analog or any other TI semiconductor product is available from your nearest TI field sales office, local authorized TI distributor, or by writing directly to:

Texas Instruments Incorporated  
Literature Response Center  
P.O. Box 809066  
Dallas, Texas 75380-9066

We feel that the Operational Amplifier/Comparator Data Book (volumes A and B) will be significant additions to your library of technical literature from Texas Instruments.

<b>General Information (Volume A)</b>	<b>1</b>
<b>Operational Amplifiers</b>	<b>2</b>
<b>Mechanical Data (Volume A)</b>	<b>3</b>
<b>General Information (Volume B)</b>	<b>4</b>
<b>Operational Amplifiers (continued)</b>	<b>5</b>
<b>Comparators</b>	<b>6</b>
<b>Mechanical Data (Volume B)</b>	<b>7</b>

# Contents

1

General Information (Volume A)

	Page
Alphanumeric Index .....	1-
Operational Amplifiers – Selection Guide .....	1-
Operational Amplifiers – Cross-Reference Guide .....	1-3
Operational Amplifiers – Glossary .....	1-4
Comparators – Selection Guide .....	1-4
Comparators – Cross-Reference Guide .....	1-4
Comparators – Glossary .....	1-4

# ALPHANUMERIC INDEX

LF347	2-3	LM2903	6-27
LF347B	2-3	LM2903Q	6-27
LF351	2-5	LM2904	2-29
LF353	2-7	LM2904A	2-29
LF411C	2-9	LM2904Q	2-29
LF412C	2-11	LM3900	2-43
LM111	6-3	LP111	6-49
LM118	2-13	LP211	6-49
LM124	2-17	LP239	6-53
LM124A	2-17	LP311	6-49
LM139	6-19	LP339	6-53
LM139A	6-19	LP2901	6-53
LM148	2-25	LT1013	2-51
LM158	2-29	LT1013A	2-51
LM158A	2-29	LT1013D	2-51
LM193	6-27	LT1013Y	2-51
LM193A	6-27	MC1458	2-75
LM211	6-3	MC1558	2-75
LM218	2-13	MC3303	2-79
LM224	2-17	MC3403	2-79
LM224A	2-17	NE5532	2-85
LM239	6-19	NE5532A	2-85
LM239A	6-19	NE5534	2-89
LM248	2-25	NE5534A	2-89
LM258	2-29	OP07C	2-95
LM258A	2-29	OP07D	2-95
LM293	6-27	OP07Y	2-95
LM293A	6-27	RC4136	2-101
LM306	6-33	RC4558	2-105
LM311	6-3	RC4558Y	2-105
LM311Y	6-3	RM4136	2-101
LM318	2-13	RM4558	2-105
LM324	2-17	RV4136	2-101
LM324A	2-17	RV4558	2-105
LM324Y	2-17	SE5534	2-89
LM324x2	2-39	SE5534A	2-89
LM339	6-19	TL022	2-111
LM339A	6-19	TL031	2-115
LM339Y	6-19	TL031A	2-115
LM339x2	6-41	TL032	2-143
LM348	2-25	TL032A	2-143
LM358	2-29	TL034	2-171
LM358A	2-29	TL034A	2-171
LM358Y	2-29	TL034Y	2-171
LM393	6-27	TL051	2-199
LM393A	6-27	TL051A	2-199
LM393Y	6-27	TL051Y	2-199
LM2900	2-43	TL052	2-229
LM2901	6-19	TL052A	2-229
LM2901Q	6-19	TL052Y	2-229
LM2902	2-17	TL054	2-261
LM2902Q	2-17	TL054A	2-261

# ALPHANUMERIC INDEX

TL054Y	2-261	TLC252B	2-429
TL061	2-291	TLC252Y	2-429
TL061A	2-291	TLC25L2	2-429
TL061B	2-291	TLC25L2A	2-429
TL061Y	2-291	TLC25L2B	2-429
TL062	2-291	TLC25L2Y	2-429
TL062A	2-291	TLC25M2	2-429
TL062B	2-291	TLC25M2A	2-429
TL062Y	2-291	TLC25M2B	2-429
TL064	2-291	TLC25M2Y	2-429
TL064A	2-291	TLC254	2-449
TL064B	2-291	TLC254A	2-449
TL064Y	2-291	TLC254B	2-449
TL064x2	2-311	TLC254Y	2-449
TL070	2-321	TLC25L4	2-449
TL071	2-335	TLC25L4A	2-449
TL071A	2-335	TLC25L4B	2-449
TL071B	2-335	TLC25L4Y	2-449
TL072	2-335	TLC25M4	2-449
TL072A	2-335	TLC25M4A	2-449
TL072B	2-335	TLC25M4B	2-449
TL074	2-335	TLC25M4Y	2-449
TL074A	2-335	TLC271	2-469
TL074B	2-335	TLC271A	2-469
TL074x2	2-351	TLC271B	2-469
TL081	2-361	TLC272	2-537
TL081A	2-361	TLC272A	2-537
TL081B	2-361	TLC272B	2-537
TL082	2-361	TLC272Y	2-537
TL082A	2-361	TLC27L2	2-625
TL082B	2-361	TLC27L2A	2-625
TL082Y	2-361	TLC27L2B	2-625
TL084	2-361	TLC27M2	2-693
TL084A	2-361	TLC27M2A	2-693
TL084B	2-361	TLC27M2B	2-693
TL084Y	2-361	TLC274	2-573
TL084x2	2-381	TLC274A	2-573
TL393	6-59	TLC274B	2-573
TL393Y	6-59	TLC274Y	2-573
TL712	6-65	TLC274x2	2-609
TL714	6-69	TLC27L4	2-657
TL2828Y	2-391	TLC27L4A	2-657
TL2828Z	2-391	TLC27L4B	2-657
TL2829Y	2-397	TLC27L4Y	2-657
TL2829Z	2-397	TLC27M4	2-727
TLC139	6-73	TLC27M4A	2-727
TLC251	2-411	TLC27M4B	2-727
TLC251A	2-411	TLC277	2-537
TLC251B	2-411	TLC279	2-573
TLC251Y	2-411	TLC27L7	2-625
TLC252	2-429	TLC27L9	2-657
TLC252A	2-429	TLC27M7	2-693



# ALPHANUMERIC INDEX

TLC27M9	2-727	TLC3704	6-179
TLC339	6-73	TLC3704Y	6-179
TLC339Q	6-73	TLE2021	5-3
TLC352	6-89	TLE2021A	5-3
TLC354	6-97	TLE2021B	5-3
TLC354Y	6-97	TLE2021Y	5-3
TLC371	6-107	TLE2022	5-29
TLC371Y	6-107	TLE2022A	5-29
TLC372	6-117	TLE2022B	5-29
TLC372Q	6-117	TLE2022Y	5-29
TLC372Y	6-117	TLE2024	5-57
TLC374	6-128	TLE2024A	5-57
TLC374Q	6-128	TLE2024B	5-57
TLC374Y	6-128	TLE2024Y	5-57
TLC393	6-141	TLE2027	5-83
TLC393Y	6-141	TLE2027A	5-83
TLC1078	2-763	TLE2037	5-109
TLC1079	2-779	TLE2037A	5-109
TLC2201	2-795	TLE2037Y	5-109
TLC2201A	2-795	TLE2061	5-137
TLC2201B	2-795	TLE2061A	5-137
TLC2201Y	2-795	TLE2061B	5-137
TLC2202	2-825	TLE2061Y	5-137
TLC2202A	2-825	TLE2062	5-167
TLC2202B	2-825	TLE2062A	5-167
TLC2202Y	2-825	TLE2062B	5-167
TLC2262	2-855	TLE2062Y	5-167
TLC2262A	2-855	TLE2064	5-197
TLC2262Y	2-855	TLE2064A	5-197
TLC2264	2-889	TLE2064B	5-197
TLC2264A	2-889	TLE2064Y	5-197
TLC2264Y	2-889	TLE2071	5-227
TLC2272	2-923	TLE2071A	5-227
TLC2272A	2-923	TLE2071Y	5-227
TLC2272Y	2-923	TLE2072	5-263
TLC2274	2-957	TLE2072A	5-263
TLC2274A	2-957	TLE2072Y	5-263
TLC2274Y	2-957	TLE2074	5-299
TLC2652	2-993	TLE2074A	5-299
TLC2652A	2-993	TLE2074Y	5-299
TLC2652Y	2-993	TLE2081	5-335
TLC2654	2-1017	TLE2081A	5-335
TLC2654A	2-1017	TLE2081Y	5-335
TLC2654Y	2-1017	TLE2082	5-367
TLC2801Y	2-1041	TLE2082A	5-367
TLC2801Z	2-1041	TLE2082Y	5-367
TLC2810Y	2-1051	TLE2084	5-403
TLC2810Z	2-1051	TLE2084A	5-403
TLC2872Y	2-1073	TLE2084Y	5-403
TLC2872Z	2-1073	TLE2141	5-435
TLC3702	6-157	TLE2141A	5-435
TLC3702Y	6-157	TLE2141Y	5-435



# ALPHANUMERIC INDEX

---

TLE2142	5-463	TLV2322	5-755
TLE2142A	5-463	TLV2322Y	5-755
TLE2142Y	5-463	TLV2324	5-779
TLE2144	5-491	TLV2324Y	5-779
TLE2144A	5-491	TLV2332	5-803
TLE2144Y	5-491	TLV2332Y	5-803
TLE2161	5-519	TLV2334	5-827
TLE2161A	5-519	TLV2334Y	5-827
TLE2161B	5-519	TLV2341	5-851
TLE2227	5-547	TLV2341Y	5-851
TLE2227Y	5-547	TLV2342	5-901
TLE2237	5-569	TLV2342Y	5-901
TLE2237Y	5-569	TLV2344	5-925
TLE2301	5-589	TLV2344Y	5-925
TLE2662	5-611	TLV2352	6-217
TLE2682	5-647	TLV2352Y	6-217
TLV1393	6-203	TLV2354	6-229
TLV1393Y	6-203	TLV2354Y	6-229
TLV2262	5-695	TLV2362	5-949
TLV2262A	5-695	TLV2362Y	5-949
TLV2262Y	5-695	TLV2393	6-203
TLV2264	5-725	TLV2393Y	6-203
TLV2264A	5-725	μA741	5-957
TLV2264Y	5-725		



---

## INTRODUCTION

This selection guide is designed to help you quickly identify which operational amplifiers best suit your needs. This section includes specification tables for each operational amplifier, sorted by the primary performance category; this permits a quick comparison of key specifications, enabling a final decision on which amplifier is best for you. Also included in this section is a complete alphanumerically sorted list of all Texas Instruments advanced linear amplifiers with key specifications.

## DEFINITION OF TERMS

This selection guide is broken into eight primary-selection categories:

- DC precision
- Single supply
- Noise
- Low voltage
- High speed
- Low power
- Rail to rail
- High temperature

These categories are then subdivided into secondary and tertiary groups combining performance indices. An understanding of what is meant by each term is helpful when choosing the right amplifier for your application.

### DC Precision

Precision refers to an amplifier's inherent dc errors, the input offset voltage ( $V_{IO}$ ), its temperature coefficient ( $\alpha_{VIO}$ ), and long-term drift ( $\Delta V_{IO}$ ). In direct-coupled applications, these errors are amplified by the amplifier and carried through the system. The magnitude of the input offset voltage limits the minimum signal level that can be accurately measured. This document defines precision operational amplifiers as those having  $V_{IO} \leq 1$  mV. In the precision-operational-amplifiers specification table, these operational amplifiers are sorted in ascending order of  $V_{IOmax}$  at 25°C; the  $\alpha_{VIO}$  specification is also provided for comparison.

### Single Supply

Single-supply operational amplifiers are those that are designed to operate well with only one power-supply rail, typically 5 V. They are generally characterized as having a common-mode input voltage range ( $V_{ICR}$ ) that includes ground and outputs that can swing to or very near ground ( $V_{OL} \approx 0$  V). Most single-supply operational amplifiers are manufactured using CMOS technology, although some bipolar single-supply amplifiers are available. Single-supply operational amplifiers can be used in systems with split supplies (e.g.,  $\pm 5$  V), but care must be taken not to exceed the maximum supply voltage across the device. For example,  $V_{DDmax}$  for CMOS operational amplifiers is 16 V. No more than  $\pm 8$  V should be applied to these devices in a split-supply system. Also, some single-supply operational amplifier output stages are not designed to both source and sink current; when used with split supplies, they may exhibit some crossover distortion as the signal passes through midsupply.

### Rail to Rail

Rail-to-rail operational amplifiers feature outputs that swing close to both the positive and negative supply rails. To achieve expected results, maintain loading conditions within the specified drive capability of the amplifier; output swing decreases as load increases.

# OPERATIONAL AMPLIFIER SELECTION GUIDE

---

## Noise

Noise in operational amplifiers typically has two components: voltage noise and current noise. Current noise is primarily a function of input bias currents ( $I_{IB}$ ) and is negligible in JFET-input (BiFET) and CMOS amplifiers. Voltage noise ( $V_n$ ) is noise generated by the amplifier due to the thermal noise of the channel resistance in JFET and CMOS amplifiers or the emitter resistance in bipolar amplifiers. Bipolar technology offers the lowest voltage noise and offers the greatest advantage when interfacing to low-impedance sources. As source impedance increases to about 10 k $\Omega$ , system noise is dominated by the thermal noise of the source and feedback resistances and selection of an amplifier is usually driven by other characteristics. At higher source impedances, the noise contribution due to the high-input currents of bipolar amplifiers becomes prohibitive and either a CMOS or BiFET amplifier should be chosen. Amplifiers in the low-noise operational amplifier sections have  $V_n \leq 15 \text{ nV}/\sqrt{\text{Hz}}$ . Current noise, though not specified, can be approximated by:

$$I_n \approx \sqrt{(2 \times q \times I_{IB})}, \text{ where } q = 1.6 \times 10^{-19}$$

## Low Voltage

Low-voltage amplifiers operate with  $V_{CC}$  or  $V_{DD} \leq 3 \text{ V}$ . Some CMOS amplifiers operate with  $V_{DD} = 1.4 \text{ V}$ . When using any supply voltage, you must ensure that input signals are within the common-mode input voltage range ( $V_{ICR}$ ) of the device. To address the emerging 3-V device market, Texas Instruments has introduced a full line of 3-V operational amplifiers, the TLV series of devices.

## High Speed

Speed refers to an operational amplifier's slew rate (SR) and its bandwidth. Slew rate describes the ability of the amplifier's output to follow a large rapidly changing signal at its input, expressed in  $\text{V}/\mu\text{s}$ . Slew rate is a function of and inversely proportional to supply current ( $I_{CC}$  or  $I_{DD}$ ); increased power consumption must often be traded for faster output response. BiFET amplifiers have traditionally offered the best speed performance, although new complementary bipolar technologies are gaining ground. The high-speed operational amplifiers in this selection guide have a bandwidth  $\geq 6 \text{ MHz}$ ; the amplifiers' slew rate is included in the specification tables for reference.

## Low Power

Low power in this document refers to amplifiers whose quiescent currents are less than 500  $\mu\text{A}$ . This category is further broken down to delineate micropower amplifiers, or those with  $I_{CC}$  or  $I_{DD} \leq 250 \mu\text{A}$ . The supply current is specified under no-load conditions; the outputs neither sink nor source current. To minimize power consumption, unused amplifiers should be connected as unity-gain followers with their inputs grounded.

## High Temperature

High-temperature operational amplifiers are those manufactured using Texas Instruments patent-pending high temperature and high-reliability process. These operational amplifiers perform reliably at temperatures up to 150°C and are well suited for automotive and geophysical (down-hole) applications where temperatures often exceed the industrial or military temperature ranges.



PRECISION OPERATIONAL AMPLIFIERS  
 $V_{IOmax} \leq 1 \text{ mV}$

DEVICE	$V_{IO}$ mV (max)	$\alpha_{VIO}$ $\mu\text{V}/^\circ\text{C}$ (typ)	$I_{CC}$ mA (max)	$I_{IB}$ nA (typ)	CMRR dB (typ)	$V_n$ (1 kHz) nV/Hz (typ)	SR V/ $\mu\text{s}$ (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
TLC2652A	0.001	0.03	2.4	0.004	140	23	2.8	1.9	C, 1, M	Single, chopper stabilized	2-993
TLC2652	0.003	0.03	2.4	0.004	140	23	2.8	1.9	C, 1, M	Single, chopper stabilized	2-993
TLC2654A	0.01	0.3	2.4	0.05	125	13	2	1.9	C, 1, M	Single, chopper stabilized	2-1017
TLC2654	0.02	0.3	2.4	0.05	125	13	2	1.9	C, 1, M	Single, chopper stabilized	2-1017
TLE2027A	0.025	0.6	4.7	15	131	3.3	2.8	13	C, 1, M	Single, low noise, high speed	5-83
TLE2037A	0.025	0.2	4.7	15	131	3.3	2.8	76	C, 1, M	Single, low noise, high speed, decompensated	5-109
TLE2027	0.1	1	5.3	15	131	3.3	2.8	13	C, 1, M	Single, low noise, high speed	5-83
TLE2037	0.1	0.4	5.3	15	131	3.3	2.8	76	C, 1, M	Single, low noise, high speed, decompensated	5-109
OP07C	0.15	0.5	—	1.8	120	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
OP07D	0.15	0.5	—	2	110	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
LT1013A	0.15	0.4	0.5	12	117	22	0.4	—	C, 1, M	Dual, precision	2-51
TLE2022B	0.15	0.3	0.35	25	105	15	0.65	2.8	C, 1, M	Dual, low power, high speed	5-29
TLC2201A	0.2	0.5	1.5	0.001	110	8	2.5	1.8	C, 1, M	Single, low noise, rail-to-rail output	2-795
TLE2021A	0.2	2	0.3	25	115	15	0.65	2	C, 1, M	Single, low noise, high speed	5-3
LT1013	0.3	0.4	0.55	15	117	22	0.4	—	C, 1, M	Dual, precision	2-51
TLE2022A	0.3	2	0.35	25	102	15	0.65	2.8	C, 1, M	Dual, low power, high speed	5-29
TLE2227	0.35	0.4	5.3	15	115	2.5	2.5	13	C	Dual, low noise, high speed	5-547
TLE2237	0.35	0.4	5.3	15	115	2.5	5	50	C	Dual, low noise, high speed, decompensated	5-569
TLC1078	0.45	1.1	0.017	0.0006	95	68	0.032	0.085	C, 1, M	Dual, low voltage, micropower	2-763
TLC277	0.5	1.8	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Dual, single supply, low power	2-537
TLC27L7	0.5	1.1	0.017	0.0006	94	68	0.03	0.085	C, 1, M	Dual, single supply, micropower	2-625
TLC27M7	0.5	1.7	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Dual, single supply, low power	2-693
TLC2201	0.5	0.5	1.5	0.001	110	8	2.5	1.8	C, 1, M	Single, low noise, rail-to-rail output	2-795
TLC2202A	0.5	0.5	1.2	0.001	110	8	2.5	1.8	C, 1, M	Single, low noise, rail-to-rail output	2-825
TLE2021	0.5	2	0.3	25	115	15	0.65	2	C, 1, M	Single, low power, high speed	5-3
TLE2022	0.5	2	0.35	25	100	15	0.65	2.8	C, 1, M	Dual, low power, high speed	5-29
TLE2024B	0.5	2	0.35	40	108	15	0.7	2.8	C, 1, M	Quad, low power, high speed	5-57
TLE2141A	0.5	1.7	4.5	-700	108	10.5	45	6	C, 1, M	Single, high speed, high output drive, low noise	5-435
TLE2024A	0.75	2	0.35	45	105	15	0.7	2.8	C, 1, M	Quad, low power, high speed	5-57
TLE2142A	0.75	1.7	4.5	-700	108	10.5	45	6	C, 1, M	Dual, high speed, high output drive, low noise	5-463

<sup>†</sup> C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## PRECISION OPERATIONAL AMPLIFIERS (Continued) $V_{IO\max} \leq 1\text{ mV}$

DEVICE	$V_{IO}$ mV (max)	$\alpha_{VIO}$ $\mu\text{V}/^\circ\text{C}$ (typ)	$I_{CC}$ mA (max)	$I_{IB}$ nA (typ)	CMRR dB (typ)	$V_n$ (1 kHz) nV/√Hz (typ)	SR V/ $\mu\text{s}$ (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
LT1013D	0.8	0.7	0.55	15	114	22	0.4	—	C, I, M	Dual, precision	2-51
TL031A	0.8	5.9	0.28	0.2	94	41	5.1	1.1	C, I, M	Single, JFET input, low power	2-115
TL032A	0.8	10.8	0.28	0.2	94	41	5.1	1.1	C, I, M	Dual, JFET input, low power	2-143
TL051A	0.75	8	3.2	0.2	93	18	23.7	3.1	C, I, M	Single, JFET input, high speed, precision	2-199
TL052A	0.8	8	2.8	0.2	93	19	20.7	3	C, I, M	Dual, JFET input, high speed, precision	2-229
TLC1079	0.85	1.1	0.017	0.0006	95	68	0.032	0.085	C, I, M	Quad, precision, low voltage, micropower	2-779
TLC279	0.9	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Quad, precision, single supply, low power	2-573
TLC27L9	0.9	1.1	0.017	0.0006	94	70	0.03	0.085	C, I, M	Quad, precision, low voltage, micropower	2-657
TLC27M9	0.9	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Quad, precision, single supply, low power	2-727
TLE2141	0.9	1.7	4.5	-700	108	10.5	45	6	C, I, M	Single, high speed, high output drive, low noise	5-435
TLC2262A	0.95	2	0.25	0.001	80	12	0.55	0.82	I	Dual, precision, low power, rail-to-rail output, low noise	2-855
TLC2264A	0.95	2	0.25	0.001	83	12	0.55	0.82	I	Quad, low power, rail-to-rail output, low noise	2-889
TLC2272A	0.95	2	3	0.001	75	9	3.6	2.18	C, I, M	Dual, rail-to-rail output, precision	2-923
TLC2274A	0.95	2	3	0.001	75	9	3.6	2.18	C, M	Dual, rail-to-rail output, precision	2-957
TLV2262A	0.95	2	0.25	0.001	77	12	0.55	0.82	I	Dual, low voltage, rail-to-rail output, low noise	5-695
TLV2264A	0.95	2	0.25	0.001	80	12	0.55	0.82	I	Quad, low voltage, rail-to-rail output, low noise	5-725
TLC2202	1	0.5	1.2	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLE2024	1	2	0.95	50	102	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57

<sup>†</sup> C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

## SINGLE-SUPPLY OPERATIONAL AMPLIFIERS

DEVICE	V <sub>CC</sub> V (min-max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP† RANGE	DESCRIPTION	PAGE NO.
TLC251 (high bias)	1.4-16	10	1.6	0.0006	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251A (high bias)	1.4-16	5	1.6	0.0006	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251B (high bias)	1.4-16	2	1.6	0.0006	80	25	3.6	1.7	C	Single, low offset, programmable power	2-411
TLC251 (medium bias)	1.4-16	10	0.28	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251A (medium bias)	1.4-16	5	0.28	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251B (medium bias)	1.4-16	2	0.28	0.0006	91	32	0.43	0.525	C	Single, low offset, programmable power	2-411
TLC251 (low bias)	1.4-16	10	0.017	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251A (low bias)	1.4-16	10	0.017	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251B (low bias)	1.4-16	2	0.017	0.0006	94	68	0.03	0.085	C	Single, low offset, programmable power	2-411
TLC252	1.4-16	10	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252A	1.4-16	5	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252B	1.4-16	2	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage, low offset	2-429
TLC25L2	1.4-16	10	0.017	0.0006	94	68	0.03	0.085	C	Dual, low voltage, micropower	2-429
TLC25L2A	1.4-16	5	0.017	0.0006	80	25	3.6	1.7	C	Dual, low voltage, micropower	2-429
TLC25L2B	1.4-16	2	0.017	0.0006	94	68	0.03	0.085	C	Dual, low voltage, low offset, micropower	2-429
TLC25M2	1.4-16	10	0.28	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low power	2-429
TLC25M2A	1.4-16	5	0.017	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-429
TLC25M2B	1.4-16	2	0.28	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-429
TLC254	1.4-16	10	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254A	1.4-16	5	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254B	1.4-16	2	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage, low offset	2-449
TLC25L4	1.4-16	10	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449
TLC25L4A	1.4-16	5	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449
TLC25L4B	1.4-16	2	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, low offset, micropower	2-449
TLC25M4	1.4-16	10	0.28	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4A	1.4-16	5	0.28	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4B	1.4-16	2	0.28	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low offset, micropower	2-449
TLC1078	1.4-16	0.45	0.017	0.0006	95	68	0.032	0.085	C, I, M	Dual, precision, low voltage, micropower	2-763
TLC1079	1.4-16	0.85	0.017	0.0006	95	68	0.032	0.085	C, I, M	Quad, precision, low voltage, micropower	2-779
TLV2362	2-5	6	2.5	20	85	8	3	7	I	Dual, low voltage, low noise, high speed	5-949

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## SINGLE-SUPPLY OPERATIONAL AMPLIFIERS (Continued)

DEVICE	V <sub>CC</sub> V (min-max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
TLV2322	2-8	9	0.017	0.0006	94	68	0.03	0.085	I	Dual, low voltage, micropower	5-755
TLV2332	2-8	9	0.28	0.0006	91	32	0.43	0.525	I	Dual, low voltage, low power	5-803
TLV2342	2-8	9	1.6	0.0006	80	25	3.6	1.7	I	Dual, low voltage	5-901
TLV2324	2-8	10	0.017	0.0006	94	68	0.03	0.085	I	Quad, low voltage, micropower	5-779
TLV2334	2-8	10	0.28	0.0006	91	32	0.43	0.525	I	Quad, low voltage, low power	5-763
TLV2344	2-8	10	1.6	0.0006	80	25	3.6	1.7	I	Quad, low voltage	5-925
TLV2341 (high bias)	2-8	8	1.6	0.0006	80	25	3.6	1.7	I	Single, low voltage, programmable power	5-851
TLV2341 (medium bias)	2-8	8	0.28	0.0006	91	32	0.43	0.525	I	Single, low voltage, programmable power	5-851
TLV2341 (low bias)	2-8	8	0.017	0.0006	94	68	0.03	0.085	I	Single, low voltage, programmable power	5-851
TLV2262	2.7-8	2.5	0.25	0.001	80	12	0.55	0.82	C, I	Dual, low voltage, low noise, rail to rail	5-695
TLV2262A	2.7-8	0.95	0.25	0.001	80	12	0.55	0.82	C, I	Dual, precision, low noise, rail to rail	5-695
TLV2264	2.7-8	2.5	0.25	0.001	80	12	0.55	0.82	I	Quad, low voltage, low noise, rail to rail	5-725
TLV2264A	2.7-8	0.95	0.25	0.001	80	12	0.55	0.82	I	Quad, precision, low noise, rail to rail	5-725
TLC2262	2.7-16	25	0.25	0.001	80	12	0.55	0.82	C, I	Dual, low power, low noise, rail to rail	2-855
TLC2262A	2.7-16	0.95	0.25	0.001	80	12	0.55	0.82	C, I	Dual, precision, low noise, rail to rail	2-855
TLC2264	2.7-16	25	0.25	0.001	80	12	0.55	0.82	C, I	Quad, low power, low noise, rail to rail	2-889
TLC2264A	2.7-16	0.95	0.25	0.001	80	12	0.55	0.82	C, I	Quad, precision, low noise, rail to rail	2-889
TLV2362	3-7	6	5	20	85	9	3	7	C, I	Dual, low voltage, high speed	5-949
TLC271 (high bias)	3-16	10	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, low voltage, programmable power	2-469
TLC271A (high bias)	3-16	5	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (high bias)	3-16	2	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, low offset, programmable power	2-469
TLC271 (medium bias)	3-16	10	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, low voltage, programmable power	2-469
TLC271A (medium bias)	3-16	5	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (medium bias)	3-16	2	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, low offset, programmable power	2-469
TLC271 (low bias)	3-16	10	0.17	0.0006	94	68	0.03	0.085	C, I, M	Single, low voltage, programmable power	2-469
TLC271A (low bias)	3-16	10	0.17	0.0006	94	68	0.03	0.085	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (low bias)	3-16	2	0.17	0.0006	94	68	0.03	0.085	C, I, M	Single, low offset, programmable power	2-469
TLC272	3-16	10	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, low voltage	2-537
TLC272A	3-16	5	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, low voltage	2-537
TLC272B	3-16	2	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, low voltage, low offset	2-537

<sup>†</sup> C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C



SINGLE-SUPPLY OPERATIONAL AMPLIFIERS (Continued)

DEVICE	V <sub>CC</sub> V (min-max)	V <sub>IQ</sub> mV (max)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLC27L2	3-16	10	0.017	0.0006	94	68	0.03	0.085	C, 1, M	Dual, low voltage, micropower	2-625
TLC27L2A	3-16	5	0.017	0.0006	80	25	3.6	1.7	C, 1, M	Dual, low voltage, micropower	2-625
TLC27L2B	3-16	2	0.017	0.0006	94	68	0.03	0.085	C, 1, M	Dual, low voltage, low offset, micropower	2-625
TLC27M2	3-16	10	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Dual, low voltage, low power	2-693
TLC27M2A	3-16	5	0.017	0.0006	91	32	0.43	0.525	C, 1, M	Dual, low voltage, low offset, low power	2-693
TLC27M2B	3-16	2	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Dual, low voltage, low offset, low power	2-693
TLC274	3-16	10	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Quad, low voltage	2-573
TLC274A	3-16	5	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Quad, low voltage	2-573
TLC274B	3-16	2	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Quad, low voltage, low offset	2-573
TLC274x2	3-16	10	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Octal, low voltage	2-609
TLC27L4	3-16	10	0.017	0.0006	94	70	0.03	0.085	C, 1, M	Quad, low voltage, micropower	2-657
TLC27L4A	3-16	5	0.017	0.0006	94	70	0.03	0.085	C, 1, M	Quad, low voltage, micropower	2-657
TLC27L4B	3-16	2	0.017	0.0006	94	70	0.03	0.085	C, 1, M	Quad, low voltage, low offset, micropower	2-657
TLC27M4	3-16	10	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Quad, low voltage, low power	2-727
TLC27M4A	3-16	5	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Quad, low voltage, low power	2-727
TLC27M4B	3-16	2	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Quad, low voltage, low offset, micropower	2-727
TLC277	3-16	0.5	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Dual, precision, single supply, low power	2-537
TLC279	3-16	0.9	1.6	0.0006	80	25	3.6	1.7	C, 1, M	Quad, precision, single supply, low power	2-573
TLC27L7	3-16	0.5	0.017	0.0006	94	68	0.3	0.085	C, 1, M	Dual, precision, single supply, micropower	2-625
TLC27L9	3-16	0.9	0.017	0.0006	94	70	0.3	0.085	C, 1, M	Quad, precision, single supply, micropower	2-657
TLC27M7	3-16	0.5	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Dual, precision, single supply, low power	2-693
TLC27M9	3-16	0.9	0.28	0.0006	91	32	0.43	0.525	C, 1, M	Quad, precision, single supply, low power	2-727
LM2904	3-26	7	2	-20	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29
LM2904Q	3-26	7	2	-20	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29
LM158	3-30	5	2	-20	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29
LM158A	3-30	2	2	-15	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29
LM258	3-30	5	2	-20	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM258A	3-30	3	2	-15	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM358	3-30	7	2	-20	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29
LM358A	3-30	3	2	-15	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C, Q = -40°C to 105°C



# OPERATIONAL AMPLIFIER SELECTION GUIDE

## SINGLE-SUPPLY OPERATIONAL AMPLIFIERS (Continued)

DEVICE	V <sub>CC</sub> V (min - max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLE2021	4-40	0.5	0.3	25	115	15	0.65	2	C, I, M	Single, precision, low power, high speed	5-3
TLE2021A	4-40	0.2	0.3	25	115	15	0.65	2	C, I, M	Single, precision, low power, high speed	5-3
TLE2022	4-40	0.5	0.35	25	100	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022A	4-40	0.3	0.35	25	102	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022B	4-40	0.15	0.35	25	105	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2024	4-40	1	0.35	50	102	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024A	4-40	0.75	0.35	45	105	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024B	4-40	0.5	0.35	40	108	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2141	4-38	0.9	4.5	-700	108	10.5	45	6	C, I, M	Single, precision, high speed, high output drive, low noise	5-435
TLE2141A	4-38	0.5	4.5	-700	108	10.5	45	6	C, I, M	Single, precision, high speed, high output drive, low noise	5-435
TLE2142	4-38	1.2	4.5	-700	108	10.5	45	6	C, I, M	Dual, precision, high output drive, low noise	5-463
TLE2142A	4-38	0.75	4.5	-700	108	10.5	45	6	C, I, M	Dual, precision, high speed, high output drive, low noise	5-463
TLE2144	4-38	2.4	4.5	-700	108	10.5	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491
TLE2144A	4-38	1.5	4.5	-700	108	10.5	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491
TLC2272	4.4-16	2.5	1.5	0.001	75	9	3.6	2.18	C, I, M	Quad, low noise, rail to rail	2-923
TLC2272A	4.4-16	0.95	1.5	0.001	75	9	3.6	2.18	C, I, M	Quad, precision, low noise, rail to rail	2-923
LM2900	4.5-32	—	6.2	30	—	—	0.6	—	I	Quad, Norton amplifier, bipolar	2-43
LM3900	4.5-32	—	6.2	30	—	—	0.6	—	C	Quad, Norton amplifier, bipolar	2-43
TLC2201	4.6-16	0.5	1.5	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2201A	4.6-16	0.2	1.5	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2202	4.6-16	1	1.2	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLC2652	4.6-16	0.003	1.2	0.004	140	23	2.8	1.9	C, I, M	Single, precision, chopper stabilized	2-993
TLC2652A	4.6-16	0.001	2.4	0.004	140	23	2.8	1.9	C, I, M	Single, precision, chopper stabilized	2-993
TLC2654	4.6-16	0.02	2.4	0.05	125	13	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
TLC2654A	4.6-16	0.01	2.4	0.05	125	13	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
MC3303	5-30	8	7	-200	90	—	0.6	—	I	Quad, low power, bipolar	2-79
MC3403	5-30	10	7	-200	90	—	0.6	—	C	Quad, low power, bipolar	2-79

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C



## LOW-NOISE OPERATIONAL AMPLIFIERS $V_{n,typ} \leq 15 \text{ nV}/\sqrt{\text{Hz}}$

DEVICE	$V_n$ (1 kHz) nV/ $\sqrt{\text{Hz}}$ (typ)	$V_{IO}$ mV (max)	$I_{CC}$ mA (max)	$I_{IB}$ nA (typ)	CMRR dB (typ)	SR $V/\mu\text{s}$ (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLE2027	2.5	0.1	5.3	15	131	2.8	13	C, I, M	Single, low noise, high speed, precision	5-83
TLE2027A	2.5	0.025	4.7	15	131	2.8	13	C, I, M	Single, low noise, high speed, precision	5-83
TLE2037	2.5	0.1	5.3	15	131	7.5	76	C, I, M	Single, low noise, high speed, precision, decompensated	5-109
TLE2037A	2.5	0.025	4.7	15	131	7.5	76	C, I, M	Single, low noise, high speed, precision, decompensated	5-109
TLE2227	2.5	0.35	5.3	15	115	2.5	13	C	Dual, precision, low noise, high speed	5-547
TLE2237	2.5	0.35	5.3	15	115	5	50	C	Dual, precision, low noise, high speed, decompensated	5-569
SE5534A	3.5	2	6.5	400	100	13	—	M	Single, low noise, high performance	2-89
NE5534	4	4	8	500	100	13	—	C	Single, low noise, high performance	2-89
NE5534A	4	4	8	500	100	13	—	C	Single, low noise, high performance	2-89
SE5534	4	2	6.5	400	100	13	—	M	Single, low noise, high performance	2-89
NE5532	5	4	16	200	100	9	—	C, I	Dual, low noise	2-85
NE5532A	5	4	16	200	100	9	—	C, I	Dual, low noise	2-85
RC4558	8	6	5.6	5	90	1.7	—	C	Dual, high performance	2-105
RM4558	8	6	5.6	5	90	1.7	—	M	Dual, high performance	2-105
RV4558	8	6	5.6	5	90	1.7	—	I	Dual, high performance	2-105
TLC2201	8	0.5	1.5	0.001	110	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2201A	8	0.2	1.5	0.001	110	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2202	8	1	1.2	0.001	110	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLC2202A	8	0.5	1.2	0.001	110	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLC2801	8	0.5	1.5	0.001	110	2.5	1.8	Z	Single, precision, low noise, rail-to-rail output, high temperature	2-1041
TLC2872	9	2.5	3	0.001	75	3.6	2.18	Z	Dual, rail-to-rail output, low noise, high temperature	2-1073
TLC2272	9	2.5	3	0.001	75	3.6	2.18	C, I, M	Dual, rail-to-rail output, low noise	2-923
TLC2272A	9	0.95	3	0.001	75	3.6	2.18	C, I, M	Dual, rail-to-rail output, precision, low noise	2-923
TLC2274	9	2.5	3	0.001	75	3.6	2.18	C, I, M	Quad, rail-to-rail output, low noise	2-957
TLC2274A	9	0.95	3	0.001	75	3.6	2.18	C, I, M	Quad, rail-to-rail output, precision, low noise	2-957
TLE2141	10.5	0.9	4.5	-700	108	45	6	C, I, M	Single, precision, high speed, high output drive, low noise	5-435
TLE2141A	10.5	0.5	4.5	-700	108	45	6	C, I, M	Single, precision, high speed, high output drive, low noise	5-435
TLE2142	10.5	1.2	4.5	-700	108	45	6	C, I, M	Dual, precision, high speed, high output drive, low noise	5-463
TLE2142A	10.5	0.75	4.5	-700	108	45	6	C, I, M	Dual, precision, high speed, high output drive, low noise	5-463
TLE2144	10.5	2.4	4.5	-700	108	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491

$T_C = 0^\circ\text{C}$  to  $70^\circ\text{C}$ ,  $I = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $M = -55^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $Z = -40^\circ\text{C}$  to  $150^\circ\text{C}$

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## LOW-NOISE OPERATIONAL AMPLIFIERS (Continued) $V_n \text{ typ} \leq 15 \text{ nV}/\sqrt{\text{Hz}}$

DEVICE	$V_n$ (1 kHz) nV/ $\sqrt{\text{Hz}}$ (typ)	$V_{IO}$ mV (max)	$I_{CC}$ mA (max)	$I_{IB}$ nA (typ)	CMRR dB (typ)	SR $V/\mu\text{s}$ (typ)	GBW MHz (typ)	TEMP'T RANGE	DESCRIPTION	PAGE NO.
TLE2144A	10.5	1.5	4.5	-700	108	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491
TLE2071	11.6	4	2.2	0.02	98	40	10	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2071A	11.6	2	2.2	0.02	98	40	10	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2072	11.6	6	1.8	0.02	98	40	10	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2072A	11.6	3.5	1.8	0.02	98	40	10	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2074	11.6	5	1.875	0.025	98	40	10	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2074A	11.6	3	1.875	0.025	98	40	10	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2081	11.6	6	2.2	0.02	98	40	10	C, I, M	Single, high speed, JFET input	5-335
TLE2081A	11.6	3	2.2	0.02	98	40	10	C, I, M	Single, high speed, JFET input	5-335
TLE2082	11.8	7	1.8	0.02	98	40	10	C, I, M	Dual, high speed, JFET input	5-367
TLE2082A	11.8	4	1.8	0.02	98	40	10	C, I, M	Dual, high speed, JFET input	5-367
TLE2084	11.6	7	1.875	0.025	98	40	10	C, I, M	Quad, high speed, JFET input	5-403
TLE2084A	11.6	4	1.875	0.025	98	40	10	C, I, M	Quad, high speed, JFET input	5-403
TLC2654	13	0.020	2.4	0.05	125	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
TLC2654A	13	0.010	2.4	0.05	125	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
TLE2021	15	0.5	0.3	25	115	0.65	2	C, I, M	Single, precision, low power, high speed	5-3
TLE2021A	15	0.2	0.3	25	115	0.65	2.8	C, I, M	Single, precision, low power, high speed	5-3
TLE2022	15	0.5	0.35	25	100	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022A	15	0.3	0.35	25	102	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022B	15	0.15	0.35	25	105	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2024	15	1	0.35	50	102	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024A	15	0.75	0.35	45	105	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024B	15	0.5	0.35	40	108	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C



LOW-VOLTAGE OPERATIONAL AMPLIFIERS  
 $V_{CCmin} \leq 3V$

DEVICE	V <sub>CC</sub> V (min - max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	V <sub>i</sub> μs V/μs (typ)	SR MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
TLC251 (high bias)	1.4 - 16	10	1.6	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251A (high bias)	1.4 - 16	5	1.6	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251B (high bias)	1.4 - 16	2	1.6	80	25	3.6	1.7	C	Single, low voltage, low offset, programmable power	2-411
TLC251 (medium bias)	1.4 - 16	10	0.28	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251A (medium bias)	1.4 - 16	5	0.28	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251B (medium bias)	1.4 - 16	2	0.28	91	32	0.43	0.525	C	Single, low voltage, low offset, programmable power	2-411
TLC251 (low bias)	1.4 - 16	10	0.017	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251A (low bias)	1.4 - 16	5	0.017	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251B (low bias)	1.4 - 16	2	0.017	94	68	0.03	0.085	C	Single, low voltage, low offset, programmable power	2-411
TLC252	1.4 - 16	10	1.6	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252A	1.4 - 16	5	1.6	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252B	1.4 - 16	2	1.6	80	25	3.6	1.7	C	Dual, low voltage, low offset	2-429
TLC25L2	1.4 - 16	10	0.017	94	68	0.03	0.085	C	Dual, low voltage, micropower	2-429
TLC25L2A	1.4 - 16	5	0.017	94	68	0.03	0.085	C	Dual, low voltage, micropower	2-429
TLC25L2B	1.4 - 16	2	0.017	94	68	0.03	0.085	C	Dual, low voltage, low offset, micropower	2-429
TLC25M2	1.4 - 16	10	0.28	91	32	0.43	0.525	C	Dual, low voltage, low power	2-429
TLC25M2A	1.4 - 16	5	0.28	91	32	0.43	0.525	C	Dual, low voltage, low power	2-429
TLC25M2B	1.4 - 16	2	0.28	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-429
TLC254	1.4 - 16	10	1.6	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254A	1.4 - 16	5	1.6	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254B	1.4 - 16	2	1.6	80	25	3.6	1.7	C	Quad, low voltage, low offset	2-449
TLC25L4	1.4 - 16	10	0.017	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449
TLC25L4A	1.4 - 16	5	0.017	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449
TLC25L4B	1.4 - 16	2	0.017	94	70	0.03	0.085	C	Quad, low voltage, low offset, micropower	2-449
TLC25M4	1.4 - 16	10	0.28	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4A	1.4 - 16	5	0.28	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4B	1.4 - 16	2	0.28	91	32	0.43	0.525	C	Quad, low voltage, low offset, micropower	2-449
TLC1078	1.4 - 16	0.45	0.017	95	68	0.032	0.085	C, I, M	Dual, precision, low voltage, micropower	2-763
TLC1079	1.4 - 16	0.85	0.017	95	68	0.032	0.085	C, I, M	Quad, precision, low voltage, micropower	2-779
TLV2322	2 - 8	9	0.017	94	68	0.03	0.085	I	Dual, low voltage, micropower	5-755

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## LOW-VOLTAGE OPERATIONAL AMPLIFIERS (Continued) $V_{CCmin} \leq 3V$

DEVICE	V <sub>CC</sub> V (min - max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP† RANGE	DESCRIPTION	PAGE NO.
TLV2332	2 - 8	9	0.28	91	32	0.43	0.525	I	Dual, low voltage, low power	5-803
TLV2342	2 - 8	9	1.6	80	25	3.6	1.7	I	Dual, low voltage	5-901
TLV2324	2 - 8	10	0.017	94	68	0.03	0.085	I	Quad, low voltage, micropower	5-779
TLV2334	2 - 8	10	0.28	91	32	0.43	0.525	I	Quad, low voltage, low power	5-827
TLV2344	2 - 8	10	1.6	80	25	3.6	1.7	I	Quad, low voltage	5-925
TLV2341 (high bias)	2 - 8	8	1.6	80	25	3.6	1.7	I	Single, low voltage, programmable power	5-851
TLV2341 (medium bias)	2 - 8	8	0.28	91	32	0.43	0.525	I	Single, low voltage, programmable power	5-851
TLV2341 (low bias)	2 - 8	8	0.017	94	68	0.03	0.085	I	Single, low voltage, programmable power	5-851
TLV2262	2.7 - 8	25	0.250	80	12	0.55	0.82	C, I	Dual, low power	5-695
TLV2262A	2.7 - 8	0.95	0.25	80	12	0.55	0.82	C, I	Dual, low power, precision	5-695
TLV2264	2.7 - 8	25	0.250	80	12	0.55	0.82	C, I	Quad, low power	5-725
TLV2264A	2.7 - 8	0.95	0.25	80	12	0.55	0.82	C, I	Quad, low power, precision	5-725
TLV2362	3 - 7	6	5	85	9	3	7	C, I	Dual, high speed	5-949
TLC271 (high bias)	3 - 16	10	1.6	80	25	3.6	1.7	C, I, M	Single, programmable power	2-469
TLC271A (high bias)	3 - 16	5	1.6	80	25	3.6	1.7	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (high bias)	3 - 16	2	1.6	80	25	3.6	1.7	C, I, M	Single, low voltage, low offset, programmable power	2-469
TLC271 (medium bias)	3 - 16	10	0.28	91	32	0.43	0.525	C, I, M	Single, low voltage, programmable power	2-469
TLC271A (medium bias)	3 - 16	5	0.28	91	32	0.43	0.525	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (medium bias)	3 - 16	2	0.28	91	32	0.43	0.525	C, I, M	Single, low voltage, low offset, programmable power	2-469
TLC271 (low bias)	3 - 16	10	0.017	94	68	0.03	0.085	C, I, M	Single, low voltage, programmable power	2-469
TLC271A (lc v bias)	3 - 16	5	0.017	94	68	0.03	0.085	C, I, M	Single, low voltage, programmable power	2-469
TLC271B (low bias)	3 - 16	2	0.017	94	68	0.03	0.085	C, I, M	Single, low voltage, low offset, programmable power	2-469
TLC272	3 - 16	10	1.6	80	25	3.6	1.7	C, I, M	Dual, low voltage	2-537
TLC272A	3 - 16	5	1.6	80	25	3.6	1.7	C, I, M	Dual, low voltage	2-537
TLC272B	3 - 16	2	1.6	80	25	3.6	1.7	C, I, M	Dual, low voltage, low offset	2-537
TLC27L2	3 - 16	10	0.017	94	68	0.03	0.085	C, I, M	Dual, micropower	2-625
TLC27L2A	3 - 16	5	0.017	94	68	0.03	0.085	C, I, M	Dual, micropower	2-625
TLC27L2B	3 - 16	2	0.017	94	68	0.03	0.085	C, I, M	Dual, low offset, micropower	2-625
TLC27M2	3 - 16	10	0.28	91	32	0.43	0.525	C, I, M	Dual, low power	2-693
TLC27M2A	3 - 16	5	0.28	91	32	0.43	0.525	C, I, M	Dual, low power	2-693

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

LOW-VOLTAGE OPERATIONAL AMPLIFIERS (Continued)  
 $V_{CCmin} \leq 3V$

DEVICE	V <sub>CC</sub> V (min - max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLC27M2B	3-16	2	0.28	91	32	0.43	0.525	C, I, M	Dual, low offset, low power	2-693
TLC274	3-16	10	1.6	80	25	3.6	1.7	C, I, M	Quad, low voltage	2-573
TLC274A	3-16	5	1.6	80	25	3.6	1.7	C, I, M	Quad, low voltage	2-573
TLC274B	3-16	2	1.6	80	25	3.6	1.7	C, I, M	Quad, low voltage, low offset	2-573
TLC274x2	3-16	10	1.6	80	25	3.6	1.7	C, I, M	Octal, low voltage	2-609
TLC27L4	3-16	10	0.017	94	70	0.03	0.085	C, I, M	Quad, micropower	2-657
TLC27L4A	3-16	5	0.017	94	70	0.03	0.085	C, I, M	Quad, micropower	2-657
TLC27L4B	3-16	2	0.017	94	70	0.03	0.085	C, I, M	Quad, low offset, micropower	2-657
TLC27M4	3-16	10	0.28	91	32	0.43	0.525	C, I, M	Quad, low power	2-727
TLC27M4A	3-16	5	0.28	91	32	0.43	0.525	C, I, M	Quad, low power	2-727
TLC27M4B	3-16	2	0.28	91	32	0.43	0.525	C, I, M	Quad, low offset, micropower	2-727
TLC277	3-16	0.5	1.6	80	25	3.6	1.7	C, I, M	Dual, precision, single supply, low power	2-537
TLC279	3-16	0.9	1.6	80	25	3.6	1.7	C, I, M	Quad, precision, single supply, low power	2-573
TLC27L7	3-16	0.5	0.017	94	68	0.3	0.085	C, I, M	Dual, precision, single supply, micropower	2-625
TLC27L9	3-16	0.9	0.017	94	70	0.3	0.085	C, I, M	Quad, precision, single supply, micropower	2-657
TLC27M7	3-16	0.5	0.28	91	32	0.43	0.525	C, I, M	Dual, precision, single supply, low power	2-693
TLC27M9	3-16	0.9	0.28	91	32	0.43	0.525	C, I, M	Quad, precision, single supply, low power	2-727
NE5534	3-20	4	8	100	4	13	—	C	Single, low noise, high performance	2-89
NE5534A	3-20	4	8	100	4	13	—	C	Single, low noise, high performance	2-89
OP07C	3-20	0.15	—	120	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
OP07D	3-20	0.15	—	110	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
SE5534	3-20	2	6.5	100	3.5	13	—	M	Single, low noise, high performance	2-89
SE5534A	3-20	2	6.5	100	3.5	13	—	M	Single, low noise, high performance	2-89
LM2902	3-26	7	0.7	80	—	0.3	—	Q	Quad, general purpose, bipolar	2-17
LM2902Q	3-26	7	0.7	80	—	0.3	—	Q	Quad, general purpose, bipolar	2-17
LM2904	3-26	7	2	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29
LM2904Q	3-26	7	2	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29
LM124	3-30	5	0.7	80	—	0.13	—	M	Quad, general purpose, bipolar	2-17
LM124A	3-30	2	1.5	80	—	0.13	—	M	Quad, general purpose, bipolar	2-17
LM158	3-30	5	2	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29

T<sub>C</sub> = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C, Q = -40°C to 105°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## LOW-VOLTAGE OPERATIONAL AMPLIFIERS (Continued) $V_{CCmin} \leq 3V$

DEVICE	V <sub>CC</sub> V (min - max)	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
LM158A	3-30	2	2	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29
LM224	3-30	5	0.7	80	—	0.3	—	I	Quad, general purpose, bipolar	2-17
LM224A	3-30	3	1.5	80	—	0.3	—	I	Quad, general purpose, bipolar	2-17
LM258	3-30	5	2	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM258A	3-30	3	2	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM324	3-30	7	0.7	80	—	0.3	—	C	Quad, general purpose, bipolar	2-17
LM324A	3-30	3	1.5	80	—	0.3	—	C	Quad, general purpose, bipolar	2-17
LM324x2	3-30	7	0.7	80	—	0.3	—	C	Octal, general purpose, bipolar	2-39
LM358	3-30	7	2	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29
LM358A	3-30	3	2	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

## HIGH-TEMPERATURE OPERATIONAL AMPLIFIERS -40°C ≤ T<sub>A</sub> ≤ 150°C

DEVICE	V <sub>IO</sub> mV (max)	αV <sub>IO</sub> μV/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (max)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TL2828	7	15	1	-100	75	23	0.15	0.4	Z	Dual, high temperature, bipolar	2-391
TL2829	7	15	0.3	-100	75	23	0.25	0.4	Z	Quad, high temperature, low power, bipolar	2-397
TLC2801	0.5	0.5	1.5	0.001	110	8	2.5	1.8	Z	Single, precision, low noise, rail-to-rail output	2-1041
TLC2810	10	3.5	1.6	0.007	90	25	3.6	1.7	Z	Dual, high temperature, CMOS	2-1051
TLC2872	2.5	2	1.5	0.001	75	9	3.6	2.18	Z	Dual, low noise, high temperature	2-1073

† Z = -40°C to 150°C



HIGH-SPEED OPERATIONAL AMPLIFIERS  
GBW<sub>typ</sub> > 6 MHz

DEVICE	GBW MHz (typ)	SR V/μs (typ)	V <sub>IO</sub> mV (max)	αV <sub>IO</sub> μV/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	TEMP† RANGE	DESCRIPTION	PAGE NO.
TLE2037	76	7.5	0.1	0.4	5.3	15	131	2.5	C, I, M	Single, decompensated, precision, low noise	5-109
TLE2037A	76	7.5	0.025	0.2	4.7	15	131	2.5	C, I, M	Single, decompensated, precision, low noise	5-109
TLE2237	50	5	0.35	0.4	5.3	15	115	2.5	C, I, M	Dual, precision, low noise, decompensated	5-569
TLE2071	10	40	4	3.2	2.2	0.02	98	11.6	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2071A	10	40	2	3.2	2.2	0.02	98	11.6	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2072	10	40	6	2.4	1.8	0.02	98	11.6	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2072A	10	40	3.5	2.4	1.8	0.02	98	11.6	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2074	10	40	5	10.1	1.875	0.025	98	11.6	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2074A	10	40	3	10.1	1.875	0.025	98	11.6	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2081	10	40	6	3.2	2.2	0.02	98	11.6	C, I, M	Single, high speed, JFET input	5-335
TLE2081A	10	40	3	3.2	2.2	0.02	98	11.6	C, I, M	Single, high speed, JFET input	5-335
TLE2082	10	40	7	2.4	1.8	0.02	98	11.8	C, I, M	Dual, JFET input	5-367
TLE2082A	10	40	4	2.4	1.8	0.2	98	11.8	C, I, M	Dual, JFET input	5-367
TLE2084	10	40	7	10.1	1.875	0.025	98	11.6	C, I, M	Quad, JFET input, high speed	5-403
TLE2084A	10	40	4	10.1	1.875	0.025	98	11.6	C, I, M	Quad, JFET input, high speed	5-403
TLE2161	6.4	10	3	6	0.35	0.004	90	40	C, I, M	Single, decompensated, micropower, high output drive	5-519
TLE2161A	6.4	10	1.5	6	0.35	0.004	90	40	C, I, M	Single, decompensated, micropower, high output drive	5-519
TLE2161B	6.4	10	0.5	6	0.35	0.004	90	40	C, I, M	Single, decompensated, precision, high output drive	5-519

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## LOW-POWER OPERATIONAL AMPLIFIERS I<sub>CC</sub>max ≤ 500 μA/Channel

DEVICE	I <sub>CC</sub> μA (max)	V <sub>IO</sub> mV (max)	αV <sub>IO</sub> μV/°C (typ)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sub>T</sub> RANGE	DESCRIPTION	PAGE NO.
TLC251 (low bias)	17	10	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251A (low bias)	17	5	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251B (low bias)	17	2	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251.2	17	10	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage	2-429
TLC251.2A	17	5	1.1	0.0006	94	68	0.03	0.085	C	Dual, low voltage	2-429
TLC251.2B	17	2	1.1	0.0006	94	68	0.03	0.085	C	Dual, low voltage, low offset	2-429
TLC251.4	17	10	1.1	0.0006	94	70	0.03	0.085	C	Quad, low voltage	2-449
TLC251.4A	17	5	1.1	0.0006	94	70	0.03	0.085	C	Quad, low voltage	2-449
TLC251.4B	17	2	1.1	0.0006	94	70	0.03	0.085	C	Quad, low voltage, low offset	2-449
TLC271 (low bias)	17	10	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-469
TLC271A (low bias)	17	10	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-469
TLC271B (low bias)	17	2	1.1	0.0006	94	68	0.03	0.085	C	Single, low voltage, low offset, programmable power	2-469
TLC271.7	17	0.5	1.1	0.0006	94	68	0.03	0.085	C, I, M	Dual, precision, single supply	2-825
TLC271.9	17	0.9	1.1	0.0006	94	70	0.03	0.085	C, I, M	Quad, precision, single supply	2-657
TLC1078	17	0.45	1.1	0.0006	94	68	0.032	0.085	C, I, M	Dual, precision, low voltage	2-763
TLC1079	17	0.85	1.1	0.0006	94	68	0.032	0.085	C, I, M	Quad, precision, low voltage	2-779
TLV2322	17	9	1.1	0.0006	94	68	0.03	0.085	I	Dual, low voltage	5-755
TLV2324	17	10	1.1	0.0006	94	68	0.03	0.085	I	Dual, low voltage	5-779
TLV2341 (low bias)	17	8	1.1	0.0006	94	68	0.03	0.085	I	Single, low voltage, programmable power	5-851
TL022	250	5	—	100	72	50	0.5	—	C, M	Dual, low power	2-111
TL061	250	15	10	0.03	86	42	3.5	—	C, I, M	Single, low power, JFET input, BIFET	2-291
TL061A	250	6	10	0.03	86	42	3.5	—	C	Single, low power, JFET input, BIFET	2-291
TL061B	250	3	10	0.03	86	42	3.5	—	C	Single, low power, JFET input, BIFET	2-291
TL062	250	15	10	0.03	86	42	3.5	—	C, I, M	Dual, low power, JFET input, BIFET	2-291
TL062A	250	6	10	0.03	86	42	3.5	—	C	Dual, low power, JFET input, BIFET	2-291
TL062B	250	3	10	0.03	86	42	3.5	—	C	Dual, low power, JFET input, BIFET	2-291
TL064	250	15	10	0.03	86	42	3.5	—	C, I, M	Quad, low power, JFET input, BIFET	2-291
TL064A	250	6	10	0.03	86	42	3.5	—	C	Quad, low power, JFET input, BIFET	2-291
TL064B	250	3	10	0.03	86	42	3.5	—	C	Quad, low power, JFET input, BIFET	2-291
TL064B	250	3	10	0.03	86	42	3.5	—	C	Quad, low power, JFET input, BIFET	2-291

T<sub>C</sub> = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C





LOW-POWER OPERATIONAL AMPLIFIERS (Continued)  
 $I_{CCmax} \leq 500 \mu A/Channel$

DEVICE	$I_{CC}$ $\mu A$ (max)	$V_{IO}$ mV (max)	$\alpha_{VIO}$ $\mu V/^\circ C$ (typ)	$I_{IB}$ nA (typ)	CMRR dB (typ)	$V_n$ (1 kHz) $nV/\sqrt{Hz}$ (typ)	SR V/ $\mu s$ (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
TL064x2	250	7	10	0.03	86	42	3.5	—	C, I, M	Octal, low power, JFET input, BIFET	2-311
TL02262	250	2.5	2	0.001	85	12	0.55	0.82	C, I	Dual, low noise, low voltage	2-855
TLC2262A	250	0.95	2	0.001	85	12	0.55	0.82	C, I	Dual, precision, low noise, low voltage	2-855
TLC2264	250	2.5	2	0.001	85	12	0.55	0.82	C, I	Quad, low noise, low voltage	2-889
TLC2264A	250	0.95	2	0.001	85	12	0.55	0.82	C, I	Quad, precision, low noise, low voltage	2-889
TLV2262	250	2.5	2	0.001	85	12	0.55	0.82	C, I	Dual, precision, low noise, low voltage	5-695
TLV2262A	250	0.95	2	0.001	85	12	0.55	0.82	C, I	Dual, low noise, low voltage	5-695
TLV2264	250	2.5	2	0.001	85	12	0.55	0.82	C, I	Quad, precision, low noise, low voltage	5-725
TLV2264A	250	0.95	2	0.001	85	12	0.55	0.82	C, I	Quad, low noise, low voltage	5-725
TL031	280	1.5	5.9	0.2	94	41	5.1	1.1	C, I, M	Single, JFET input, low power, precision	2-115
TL031A	280	0.8	5.9	0.2	94	41	5.1	1.1	C, I, M	Single, JFET input, low power, precision	2-115
TL032	280	1.5	10.8	0.2	94	41	5.1	1.1	C, I, M	Dual, JFET input, low power, precision	2-143
TL032A	280	0.8	10.8	0.2	94	41	5.1	1.1	C, I, M	Dual, JFET input, low power, precision	2-143
TL034	280	4	12	0.2	94	43	5.1	1.1	C, I, M	Quad, JFET input, low power, precision	2-171
TL034A	280	1.5	12	0.2	94	43	5.1	1.1	C, I, M	Quad, JFET input, low voltage, micropower	2-171
TLC251 (medium bias)	280	10	1.7	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251A (medium bias)	280	5	1.7	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251B (medium bias)	280	2	1.7	0.0006	91	32	0.43	0.525	C	Single, low offset, programmable power	2-411
TLC25M2	280	10	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage	2-429
TLC25M2A	280	5	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset	2-429
TLC25M2B	280	2	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset	2-429
TLC25M4	280	10	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, micropower	2-449
TLC25M4A	280	5	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, micropower	2-449
TLC25M4B	280	2	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, micropower	2-449
TLC271 (medium bias)	280	10	1.7	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-469
TLC271A (medium bias)	280	5	1.7	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-469
TLC271B (medium bias)	280	2	1.7	0.0006	91	32	0.43	0.525	C	Single, low offset, programmable power	2-469
TLC27M2	280	10	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low power	2-693
TLC27M2A	280	5	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-693
TLC27M2B	280	2	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-693

<sup>†</sup> C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## LOW-POWER OPERATIONAL AMPLIFIERS (Continued) I<sub>CC</sub>max ≤ 500 μA/Channel

DEVICE	I <sub>CC</sub> μA (max)	V <sub>IO</sub> mV (max)	αV <sub>IO</sub> μV/°C (typ)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLC27M4	280	10	1.7	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-727
TLC27M4A	280	5	1.7	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-727
TLC27M4B	280	2	1.7	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-727
TL031	280	1.5	5.9	0.2	94	41	5.1	1.1	C	Single, low voltage, programmable power	2-115
TL031A	280	0.8	5.9	0.2	94	41	5.1	1.1	C	Single, low voltage, programmable power	2-115
TL032	280	1.5	10.8	0.2	94	41	5.1	1.1	C	Single, low offset, programmable power	2-143
TL032A	280	0.8	10.8	0.2	94	41	5.1	1.1	C	Single, low offset, programmable power	2-143
TLC27M7	280	0.5	1.7	0.0006	91	32	0.43	0.525	C, I, M	Dual, precision, single supply, low power	2-693
TLC27M9	280	0.9	1.7	0.0006	91	32	0.43	0.525	C, I, M	Quad, precision, single supply, low power	2-727
TLV2332	280	9	1.7	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	5-803
TLV2334	280	10	1.7	0.0006	91	32	0.43	0.525	C	Quad, low voltage	5-827
TLV2341 (medium bias)	280	8	1.7	0.0006	91	32	0.43	0.525	C	Quad, low voltage	5-851
TLE2021A	300	0.2	2	25	115	15	0.65	2	C	Quad, low voltage, micropower	5-3
TLE2062	345	4	6	0.004	90	40	3.4	2	C	Quad, low voltage, low offset, micropower	5-167
TLE2062A	345	2	6	0.004	90	40	3.4	2	C	Quad, low voltage, low power	5-167
TLE2022	350	0.5	2	25	100	15	0.65	2.8	C	Quad, low voltage, low power	5-29
TLE2022A	350	0.3	2	25	102	15	0.65	2.8	C	Quad, low voltage, low offset, micropower	5-29
TLE2022B	350	0.15	2	25	105	15	0.65	2.8	C, I, M	Dual, precision, low voltage, micropower	5-29
TLE2024	350	1	2	50	102	15	0.7	2.8	C, I, M	Quad, precision, low voltage, micropower	5-57
TLE2024A	350	0.75	2	45	105	15	0.7	2.8	I	Dual, low voltage, micropower	5-57
TLE2024B	350	0.5	2	40	108	15	0.7	2.8	I	Dual, low voltage, low power	5-57
TLE2061	350	3	6	0.004	90	40	3.4	2	I	Dual, low voltage	5-137
TLE2061A	350	1.5	6	0.004	90	40	3.4	2	I	Quad, low voltage, micropower	5-137
TLE2064	350	6	6	0.004	90	40	3.4	2	I	Quad, low voltage	5-197
TLE2064A	350	4	6	0.004	90	40	3.4	1.7	I	Single, low voltage, programmable power	5-197
TLE2161	350	3	6	0.004	90	40	10	0.525	I	Single, low voltage, programmable power	5-519
TLE2161A	350	1.5	6	0.004	90	40	10	0.085	I	Single, low voltage, programmable power	5-519
TLE2161B	350	0.5	6	0.004	90	40	10	6.4	C, I, M	Single, decompensated, high output drive	5-519
LT1013A	500	0.15	0.3	12	117	22	0.4	—	C, I, M	Dual, precision	2-51

T<sub>C</sub> = 0°C to 70°C, I<sub>L</sub> = -40°C to 85°C, M = -55°C to 125°C



RAIL-TO-RAIL OPERATIONAL AMPLIFIERS

DEVICE	V <sub>IO</sub> mV (max)	I <sub>CC</sub> mA (max)	VOL - V <sub>OH</sub> V (typ)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
TLC2201	0.2	1.5	0 - 4.8	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise	2-795
TLC2201A	0.2	1.5	0 - 4.8	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise	2-795
TLC2202	0.5	1.2	0 - 4.8	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise	2-825
TLC2202A	0.5	1.2	0 - 4.8	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise	2-825
TLC2262	2.5	0.25	0 - 4.99	0.001	80	12	0.55	0.82	C, I	Dual, low power, low noise	2-855
TLC2262A	0.95	0.25	0 - 4.99	0.001	80	12	0.55	0.82	I	Dual, precision, low power, low noise	2-855
TLC2264	2.5	0.25	0 - 4.99	0.001	80	12	0.55	0.82	C, I	Quad, low power, low noise	2-889
TLC2264A	0.95	0.25	0 - 4.99	0.001	80	12	0.55	0.82	I	Quad, precision, low power, low noise	2-889
TLC2272	2.5	1.5	0 - 4.99	0.001	75	9	3.6	2.18	C, I, M	Dual, low noise	2-923
TLC2272A	0.95	1.5	0 - 4.99	0.001	75	9	3.6	2.18	C, I, M	Dual, precision, low noise	2-923
TLC2274	2.5	1.5	0 - 4.99	0.001	75	9	3.6	2.18	C, I, M	Quad, low noise	2-957
TLC2274A	0.95	1.5	0 - 4.99	0.001	75	9	3.6	2.18	C, I, M	Quad, precision, low noise	2-957
TLV2262	2.5	0.25	0 - 4.99	0.001	80	12	0.55	0.82	C, I	Dual, low power, low voltage	5-695
TLV2262A	0.95	0.25	0 - 4.99	0.001	80	12	0.55	0.82	I	Dual, precision, low power, low noise	5-695
TLV2264	2.5	0.25	0 - 4.99	0.001	80	12	0.55	0.82	C, I	Quad, low power, low voltage	5-725
TLV2264A	0.95	0.25	0 - 4.99	0.001	80	12	0.55	0.82	I	Quad, precision, low power, low noise	5-725

<sup>†</sup> C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## OPERATIONAL AMPLIFIERS (Listed Alphabetically)

DEVICE	V <sub>IO</sub> mV (max)	$\alpha$ V <sub>IO</sub> $\mu$ V/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sup>†</sup> RANGE	DESCRIPTION	PAGE NO.
LF347	10	18	11	0.05	100	18	13	—	C	Quad, wide bandwidth, JFET input	2-3
LF347B	5	18	11	0.05	100	18	13	—	C	Quad, wide bandwidth, JFET input	2-3
LF351	10	10	3.4	0.05	100	18	13	—	C	Dual, wide bandwidth, JFET input	2-5
LF353	10	10	6.5	0.05	100	18	13	—	C	Dual, wide bandwidth, JFET input	2-7
LF411C	2	10	3.4	0.05	100	18	13	—	C	Single, JFET input	2-9
LF412C	3	10	6.8	0.05	100	18	13	—	C	Dual, JFET input	2-11
LM118	4	—	8	120	100	—	70	—	M	Single, high performance	2-13
LM124	5	—	0.7	-20	80	—	0.13	—	M	Quad, general purpose, bipolar	2-17
LM124A	2	—	1.5	—	80	—	0.13	—	M	Quad, general purpose, bipolar	2-17
LM148	5	—	2.4	30	90	—	0.5	—	M	Quad, general purpose, bipolar	2-25
LM158	5	7	2	-20	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29
LM158A	2	7	2	-15	80	—	0.2	—	M	Dual, high gain, low power, bipolar	2-29
LM218	4	—	8	120	100	—	70	—	I	Single, high performance	2-13
LM224	5	—	0.7	-20	80	—	0.3	—	I	Quad, general purpose, bipolar	2-17
LM224A	3	—	1.5	-15	80	—	0.3	—	I	Quad, general purpose, bipolar	2-17
LM248	6	—	—	30	90	—	0.5	—	I	Quad, general purpose, bipolar	2-25
LM256	5	7	2	-20	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM258A	3	7	2	-15	80	—	0.2	—	I	Dual, high gain, low power, bipolar	2-29
LM318	10	—	10	150	100	—	70	—	C	Single, high performance	2-13
LM324	7	—	0.7	-20	80	—	0.3	—	C	Quad, general purpose, bipolar	2-17
LM324A	3	—	1.5	-15	80	—	0.3	—	C	Quad, general purpose, bipolar	2-17
LM324x2	7	—	0.7	-20	80	—	0.3	—	C	Octal, general purpose, bipolar	2-39
LM348	6	—	—	30	90	—	0.5	—	C	Quad, general purpose, bipolar	2-25
LM358	7	7	2	-20	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29
LM358A	3	7	2	-15	80	—	0.2	—	C	Dual, high gain, low power, bipolar	2-29
LM2900	—	—	6.2	30	—	—	0.6	—	I	Quad, Norton amplifier, bipolar	2-43
LM2902	7	—	0.7	-20	80	—	0.3	—	Q	Quad, general purpose, bipolar	2-17
LM2902Q	7	—	0.7	-20	80	—	0.3	—	Q	Quad, general purpose, bipolar	2-17
LM2904	7	7	2	-20	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29
LM2904Q	7	7	2	-20	80	—	0.2	—	Q	Dual, high gain, low power, bipolar	2-29

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C, Q = -40°C to 105°C



OPERATIONAL AMPLIFIERS (Listed Alphanumerically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	α <sub>V</sub> /IO μV/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP'T RANGE	DESCRIPTION	PAGE NO.
LM3900	—	—	6.2	30	—	—	0.6	—	C	Quad, Norton amplifier, bipolar	2-43
LT1013	0.3	0.4	0.55	15	117	22	0.4	—	C, I, M	Dual, precision	2-51
LT1013A	0.15	0.3	0.5	12	117	22	0.4	—	C, I, M	Dual, precision	2-51
LT1013D	0.3	0.7	0.55	15	114	22	0.4	—	C, I, M	Dual, precision	2-51
MC1458	6	—	5.6	80	90	45	0.5	—	C	Dual, general purpose	2-75
MC1558	5	—	5	80	90	45	0.5	—	M	Dual, general purpose	2-75
MC3303	8	10	7	-200	90	—	0.6	—	I	Quad, low power, bipolar	2-79
MC3403	10	10	7	-200	90	—	0.6	—	C	Quad, low power, bipolar	2-79
NE5532	4	—	16	200	100	5	9	—	C, I	Dual, low noise	2-85
NE5532A	4	—	16	200	100	5	9	—	C, I	Dual, low noise	2-85
NE5534	4	—	8	500	100	4	13	—	C	Single, low noise, high performance	2-89
NE5534A	4	—	8	500	100	4	13	—	C	Single, low noise, high performance	2-89
OP07C	0.15	0.5	—	1.8	120	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
OP07D	0.15	0.5	—	2	110	9.8	0.3	—	C	Single, ultra-low offset voltage	2-95
RC4136	6	—	2.8	140	90	—	1.7	—	C	Quad, high performance	2-101
RC4558	6	—	5.6	5	90	—	1.7	—	C	Dual, high performance	2-105
RM4136	4	—	2.8	140	90	—	1.7	—	M	Quad, high performance	2-101
RM4558	6	—	5.6	5	90	—	1.7	—	M	Dual, high performance	2-105
RV4146	6	—	2.8	140	90	—	1.7	—	I	Quad, high performance	2-101
RV4558	6	—	5.6	5	90	—	1.7	—	I	Dual, high performance	2-105
SE5534	2	—	6.5	400	100	3.5	13	—	M	Single, low noise, high performance	2-89
SE5534A	2	—	6.5	400	100	3.5	13	—	M	Single, low noise, high performance	2-89
TL022	5	—	0.25	100	72	50	0.5	—	C, M	Dual, low power	2-111
TL031	1.5	5.9	0.28	0.2	94	41	5.1	1.1	C, I, M	Single, JFET input, low power, precision	2-115
TL031A	0.8	5.9	0.28	0.2	94	41	5.1	1.1	C, I, M	Single, JFET input, low power, precision	2-115
TL032	1.5	10.8	0.28	0.2	94	41	5.1	1.1	C, I, M	Dual, JFET input, low power, precision	2-143
TL032A	0.8	10.8	0.28	0.2	94	41	5.1	1.1	C, I, M	Dual, JFET input, low power, precision	2-143
TL034	4	12	0.28	0.2	94	43	5.1	1.1	C, I, M	Quad, JFET input, low power, precision	2-171
TL034A	1.5	12	0.28	0.2	94	43	5.1	1.1	C, I, M	Quad, JFET input, low voltage, micropower	2-171
TL051	1.5	8	3.2	0.2	93	18	23.7	3.1	C, I, M	Single, JFET-input, high speed, precision	2-199

T<sub>C</sub> = 0°C to 70°C; I = -40°C to 85°C; M = -55°C to 125°C



# OPERATIONAL AMPLIFIER SELECTION GUIDE

## OPERATIONAL AMPLIFIERS (Listed Alphanumerically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	α <sub>VIO</sub> μV/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP† RANGE	DESCRIPTION	PAGE NO.
TL051A	0.8	8	3.2	0.2	93	18	23.7	3.1	C, I, M	Single, JFET input, high speed, precision	2-199
TL052	1.5	8	2.8	0.2	93	19	20.7	3	C, I, M	Dual, JFET-input, high speed, precision	2-229
TL052A	0.8	8	2.8	0.2	93	19	20.7	3	C, I, M	Dual, JFET input, high speed, precision	2-229
TL054	4	23	2.8	0.2	92	21	17.8	2.7	C, I, M	Quad, JFET-input, high speed, precision	2-261
TL054A	1.5	23	2.8	0.2	92	21	17.8	2.7	C, I, M	Quad, JFET input, high speed, precision	2-261
TL081	15	10	0.25	0.03	86	42	3.5	—	C, I, M	Single, low power, JFET input, BIFET	2-291
TL061A	6	10	0.25	0.03	86	42	3.5	—	C	Single, low power, JFET input, BIFET	2-291
TL061B	3	10	0.25	0.03	86	42	3.5	—	C	Single, low power, JFET input, BIFET	2-291
TL062	15	10	0.25	0.03	86	42	3.5	—	C, I, M	Dual, low power, JFET input, BIFET	2-291
TL062A	6	10	0.25	0.03	86	42	3.5	—	C	Dual, low power, JFET input, BIFET	2-291
TL062B	3	10	0.25	0.03	86	42	3.5	—	C	Dual, low power, JFET input, BIFET	2-291
TL064	15	10	0.25	0.03	86	42	3.5	—	C, I, M	Quad, low power, JFET input, BIFET	2-291
TL064A	6	10	0.25	0.03	86	42	3.5	—	C	Quad, low power, JFET input, BIFET	2-291
TL064B	3	10	0.25	0.03	86	42	3.5	—	C	Quad, low power, JFET input, BIFET	2-291
TL064x2	7	10	0.25	0.03	86	42	3.5	—	C	Octal, low power, JFET input, BIFET	2-311
TL070	10	18	2.5	0.065	100	18	13	—	C	Single, low noise JFET input, BIFET	2-321
TL071	10	18	2.5	0.065	100	18	13	—	C, I, M	Single, low noise JFET input, BIFET	2-335
TL071A	6	18	2.5	0.065	100	18	13	—	C	Single, low noise JFET input, BIFET	2-335
TL071B	3	18	2.5	0.065	100	18	13	—	C	Single, low noise JFET input, BIFET	2-335
TL072	10	18	2.5	0.065	100	18	13	—	C, I, M	Dual, low noise JFET input, BIFET	2-335
TL072A	6	18	2.5	0.065	100	18	13	—	C	Dual, low noise JFET input, BIFET	2-335
TL072B	3	18	2.5	0.065	100	18	13	—	C	Dual, low noise JFET input, BIFET	2-335
TL074	10	18	2.5	0.065	100	18	13	—	C, I, M	Quad, low noise JFET input, BIFET	2-335
TL074A	6	18	2.5	0.065	100	18	13	—	C	Quad, low noise JFET input, BIFET	2-335
TL074B	3	18	2.5	0.065	100	18	13	—	C	Quad, low noise JFET input, BIFET	2-335
TL074x2	10	18	2.5	0.065	100	18	13	—	C	Octal, low noise, JFET input, BIFET	2-351
TL081	15	18	2.8	0.03	86	18	13	—	C, I, M	Single, general purpose, JFET input, BIFET	2-361
TL081A	6	18	2.8	0.03	86	18	13	—	C	Single, general purpose, JFET input, BIFET	2-361
TL081B	3	18	2.8	0.03	86	18	13	—	C	Single, general purpose, JFET input, BIFET	2-361
TL082	15	18	2.8	0.03	86	18	13	—	C, I, M	Dual, general purpose, JFET input, BIFET	2-361

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C





OPERATIONAL AMPLIFIERS (Listed Alphabetically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	$\alpha$ /V <sub>IO</sub> $\mu$ V/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
TLC082A	6	18	2.8	0.03	86	18	13	—	C	Dual, general purpose, JFET input, BIFET	2-361
TLC082B	3	18	2.8	0.03	86	18	13	—	C	Dual, general purpose, JFET input, BIFET	2-361
TLC084	15	18	2.8	0.03	86	18	13	—	C, I, M	Quad, general purpose, JFET input, BIFET	2-361
TLC084A	6	18	2.8	0.03	86	18	13	—	C	Quad, general purpose, JFET input, BIFET	2-361
TLC084B	3	18	2.8	0.03	86	18	13	—	C	Quad, general purpose, JFET input, BIFET	2-361
TLC084x2	15	18	2.8	0.03	86	18	13	—	C	Octal, general purpose, JFET input, BIFET	2-381
TLC2828	7	15	1	-100	75	23	0.15	0.4	Z	Dual, high temperature, bipolar	2-391
TLC2829	7	15	0.3	-100	75	23	0.25	0.4	Z	Quad, high temperature, low power, bipolar	2-397
TLC251 (high bias)	10	1.8	1.6	0.0006	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251A (high bias)	5	1.8	1.6	0.0006	80	25	3.6	1.7	C	Single, low voltage, programmable power	2-411
TLC251B (high bias)	2	1.8	1.6	0.0006	80	25	3.6	1.7	C	Single, low offset, programmable power	2-411
TLC251 (medium bias)	10	1.7	0.28	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251A (medium bias)	5	1.7	0.28	0.0006	91	32	0.43	0.525	C	Single, low voltage, programmable power	2-411
TLC251B (medium bias)	2	1.7	0.28	0.0006	91	32	0.43	0.525	C	Single, low offset, programmable power	2-411
TLC251 (low bias)	10	1.1	0.017	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251A (low bias)	5	1.1	0.017	0.0006	94	68	0.03	0.085	C	Single, low voltage, programmable power	2-411
TLC251B (low bias)	2	1.1	0.017	0.0006	94	68	0.03	0.085	C	Single, low offset, programmable power	2-411
TLC252	10	1.8	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252A	5	1.8	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage	2-429
TLC252B	2	1.8	1.6	0.0006	80	25	3.6	1.7	C	Dual, low voltage, low offset	2-429
TLC252L	10	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, low voltage, micropower	2-429
TLC252L2A	5	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, low voltage, micropower	2-429
TLC252L2B	2	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, low voltage, low offset, micropower	2-429
TLC25M2	10	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low power	2-429
TLC25M2A	5	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low power	2-429
TLC25M2B	2	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low voltage, low offset, low power	2-429
TLC254	10	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254A	5	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage	2-449
TLC254B	2	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, low voltage, low offset	2-449
TLC254L	10	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C, Z = -40°C to 150°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## OPERATIONAL AMPLIFIERS (Listed Alphabetically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	$\alpha$ V <sub>IO</sub> $\mu$ V/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>B</sub> nA	CMRR dB	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	S <sub>r</sub> V/μs (typ)	GBW MHz (typ)	TEMPT RANGE	DESCRIPTION	PAGE NO.
TLC25L4A	5	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, micropower	2-449
TLC25L4B	2	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, low voltage, low offset, micropower	2-449
TLC25M4	10	1.7	0.28	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4A	5	1.7	0.28	0.0006	91	32	0.43	0.525	C	Quad, low voltage, low power	2-449
TLC25M4B	2	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Quad, low voltage, low offset, micropower	2-449
TLC271 (high bias)	10	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, programmable power	2-469
TLC271A (high bias)	5	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, programmable power	2-469
TLC271B (high bias)	2	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Single, low offset, programmable power	2-469
TLC271 (medium bias)	10	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, programmable power	2-469
TLC271A (medium bias)	5	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, programmable power	2-469
TLC271B (medium bias)	2	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Single, low offset, programmable power	2-469
TLC271 (low bias)	10	1.1	0.017	0.0006	94	68	0.03	0.085	C, I, M	Single, programmable power	2-469
TLC271A (low bias)	5	1.1	0.017	0.0006	94	68	0.03	0.085	C, I, M	Single, programmable power	2-469
TLC271B (low bias)	2	1.1	0.017	0.0006	94	68	0.03	0.085	C, I, M	Single, low offset, programmable power	2-469
TLC272	10	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, single supply	2-537
TLC272A	5	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, single supply	2-537
TLC272B	2	1.8	1.6	0.0006	80	25	3.6	1.7	C	Dual, low offset, single supply	2-537
TLC27L2	10	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, micropower, single supply	2-625
TLC27L2A	5	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, micropower, single supply	2-625
TLC27L2B	2	1.1	0.017	0.0006	94	68	0.03	0.085	C	Dual, low offset, micropower, single supply	2-625
TLC27M2	10	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low power, single supply	2-693
TLC27M2A	5	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low offset, low power, single supply	2-693
TLC27M2B	2	1.7	0.28	0.0006	91	32	0.43	0.525	C	Dual, low offset, low power, single supply	2-693
TLC274	10	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, single supply	2-573
TLC274A	5	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, single supply	2-573
TLC274B	2	1.8	1.6	0.0006	80	25	3.6	1.7	C	Quad, low offset, single supply	2-573
TLC274A2	10	1.8	1.6	0.0006	80	25	3.6	1.7	C	Octal, single supply	2-609
TLC27L4	10	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, micropower, single supply	2-657
TLC27L4A	5	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, micropower, single supply	2-657
TLC27L4B	2	1.1	0.017	0.0006	94	70	0.03	0.085	C	Quad, low offset, micropower, single supply	2-657

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C



**OPERATIONAL AMPLIFIERS (Listed Alphanumerically) (Continued)**

DEVICE	V <sub>IO</sub> mV (max)	α <sub>VIO</sub> μV/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sup>T</sup> RANGE	DESCRIPTION	PAGE NO.
TLC27M4	10	1.7	0.28	0.0006	91	32	0.43	0.525	C	Quad, low power, single supply	2-727
TLC27M4A	5	1.7	0.28	0.0006	91	32	0.43	0.525	C	Quad, low power, single supply	2-727
TLC27M4B	2	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Quad, low offset, micropower, single supply	2-727
TLC277	0.5	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Dual, precision, single supply, low power	2-537
TLC279	0.9	1.8	1.6	0.0006	80	25	3.6	1.7	C, I, M	Quad, precision, single supply, low power	2-573
TLC27L7	0.5	1.1	0.017	0.0006	94	68	0.03	0.085	C, I, M	Dual, precision, single supply, micropower	2-625
TLC27L9	0.9	1.1	0.017	0.0006	94	70	0.03	0.085	C, I, M	Quad, precision, single supply, micropower	2-657
TLC27M7	0.5	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Dual, precision, single supply, low power	2-693
TLC27M9	0.9	1.7	0.28	0.0006	91	32	0.43	0.525	C, I, M	Quad, precision, single supply, low power	2-727
TLC1078	0.45	1.1	0.017	0.0006	95	68	0.032	0.085	C, I, M	Dual, precision, low voltage, micropower	2-763
TLC1079	0.85	1.1	0.017	0.0006	95	68	0.032	0.085	C, I, M	Quad, precision, low voltage, micropower	2-779
TLC2201	0.5	0.5	1.5	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2201A	0.2	0.5	1.5	0.001	110	8	2.5	1.8	C, I, M	Single, precision, low noise, rail-to-rail output	2-795
TLC2202	1	0.5	1.2	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLC2202A	0.5	0.5	1.2	0.001	110	8	2.5	1.9	C, I, M	Dual, precision, low noise, rail-to-rail output	2-825
TLC2262	2.5	2	0.25	0.001	80	12	0.55	0.82	C, I	Dual, low power, rail-to-rail output, low noise	2-855
TLC2262A	0.95	2	0.25	0.001	80	12	0.55	0.82	C, I	Dual, precision, low power, rail-to-rail output	2-855
TLC2264	2.5	2	0.25	0.001	80	12	0.55	0.82	C, I	Quad, low power, rail-to-rail output, low noise	2-889
TLC2264A	0.95	2	0.25	0.001	80	12	0.55	0.82	I	Quad, low power, rail-to-rail output, low noise	2-889
TLC2272	2.5	2	3	0.001	75	9	3.6	2.18	C, I, M	Dual, rail-to-rail output	2-923
TLC2272A	0.95	2	3	0.001	75	9	3.6	2.18	C, I, M	Dual, rail-to-rail output, precision	2-923
TLC2274	2.5	2	3	0.001	75	9	3.6	2.18	C, I, M	Quad, rail-to-rail output	2-957
TLC2274A	0.95	2	3	0.001	75	9	3.6	2.18	C, I, M	Quad, rail-to-rail output, precision	2-957
TLC2652	0.003	0.03	2.4	0.004	140	23	2.8	1.9	C, I, M	Single, precision, chopper stabilized	2-993
TLC2652A	0.001	0.03	2.4	0.004	140	23	2.8	1.9	C, I, M	Single, precision, chopper stabilized	2-993
TLC2654	0.02	0.3	2.4	0.05	125	13	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
TLC2654A	0.01	0.3	2.4	0.05	125	13	2	1.9	C, I, M	Single, precision, chopper stabilized	2-1017
TLC2801	0.5	0.5	1.5	0.001	110	8	2.5	2.18	Z	Single, precision, low noise, rail-to-rail output	2-1041
TLC2810	10	3.5	1.6	0.007	90	25	3.6	1.7	Z	Dual, high temperature, CMOS	2-1051
TLC2872	2.5	2	3	0.001	75	9	3.6	2	Z	Dual, low noise, high temperature	2-1073

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C, Z = -40°C to 150°C

# OPERATIONAL AMPLIFIER SELECTION GUIDE

## OPERATIONAL AMPLIFIERS (Listed Alphabetically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	$\alpha$ V <sub>IO</sub> $\mu$ V/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP <sup>t</sup> RANGE	DESCRIPTION	PAGE NO.
TLE2021	0.5	2	0.3	25	115	15	0.65	2	C, I, M	Single, precision, low power, high speed	5-3
TLE2021A	0.2	2	0.3	25	115	15	0.65	2.8	C, I, M	Single, precision, low power, high speed	5-3
TLE2022	0.5	2	0.35	25	100	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022A	0.3	2	0.35	25	102	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2022B	0.15	2	0.35	25	105	15	0.65	2.8	C, I, M	Dual, precision, low power, high speed	5-29
TLE2024	1	2	0.35	50	102	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024A	0.75	2	0.35	45	105	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2024B	0.5	2	0.35	40	108	15	0.7	2.8	C, I, M	Quad, precision, low power, high speed	5-57
TLE2027	0.1	0.4	5.3	15	131	3.3	2.8	13	C, I, M	Single, low noise, high speed, precision	5-83
TLE2027A	0.025	0.2	4.7	15	131	3.3	2.8	13	C, I, M	Single, low noise, high speed, precision	5-83
TLE2037	0.1	0.4	5.3	15	131	3.3	2.8	76	C, I, M	Single, low noise, high speed, precision, decompensated	5-109
TLE2037A	0.025	0.2	4.7	15	131	3.3	2.8	76	C, I, M	Single, low noise, high speed, precision, decompensated	5-109
TLE2061	3	6	0.35	0.004	90	40	3.4	2	C, I, M	Single, JFET input, micropower, high output drive	5-137
TLE2061A	1.5	6	0.35	0.004	90	40	3.4	2	C, I, M	Single, JFET input, micropower, high output drive	5-137
TLE2062	4	6	0.345	0.004	90	40	3.4	2	C, I, M	Dual, JFET input, micropower, high output drive	5-167
TLE2062A	2	6	0.345	0.004	90	40	3.4	2	C, I, M	Dual, JFET input, micropower, high output drive	5-167
TLE2064	6	6	0.35	0.004	90	40	3.4	2	C, I, M	Quad, JFET input, micropower, high output drive	5-197
TLE2064A	4	6	0.35	0.004	90	40	3.4	2	C, I, M	Quad, JFET input, micropower, high output drive	5-197
TLE2071	4	3.2	2.2	0.02	98	11.6	40	10	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2071A	2	3.2	2.2	0.02	98	11.6	40	10	C, I, M	Single, high speed, low noise, JFET input	5-227
TLE2072	6	2.4	1.8	0.02	98	11.6	40	10	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2072A	3.5	2.4	1.8	0.02	98	11.6	40	10	C, I, M	Dual, high speed, low noise, JFET input	5-263
TLE2074	5	10.1	1.875	0.025	98	11.6	40	10	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2074A	3	10.1	1.875	0.025	98	11.6	40	10	C, I, M	Quad, high speed, low noise, JFET input	5-299
TLE2081	6	3.2	2.2	0.02	98	11.6	40	10	C, I, M	Single, high speed, JFET input	5-335
TLE2081A	3	3.2	2.2	0.02	98	11.6	40	10	C, I, M	Single, high speed, JFET input	5-335
TLE2082	7	2.4	1.8	0.02	98	11.8	40	10	C, I, M	Dual, JFET input, high speed	5-367
TLE2082A	4	2.4	1.8	0.02	98	11.8	40	10	C, I, M	Dual, JFET input, high speed	5-367
TLE2084	7	10.1	1.875	0.025	98	11.6	40	10	C, I, M	Quad, JFET input, high speed	5-403
TLE2084A	4	10.1	1.875	0.025	98	11.6	40	10	C, I, M	Quad, JFET input, high speed	5-403

T<sub>C</sub> = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

OPERATIONAL AMPLIFIERS (Listed Alphanumerically) (Continued)

DEVICE	V <sub>IO</sub> mV (max)	$\alpha$ V <sub>IO</sub> $\mu$ V/°C (typ)	I <sub>CC</sub> mA (max)	I <sub>IB</sub> nA (typ)	CMRR dB (typ)	V <sub>n</sub> (1 kHz) nV/√Hz (typ)	SR V/μs (typ)	GBW MHz (typ)	TEMP T RANGE	DESCRIPTION	PAGE NO.
TLE2141	0.9	1.7	4.5	-700	108	10.5	45	6	C, I, M	Single, precision, high speed, high output drive	5-435
TLE2141A	0.5	1.7	4.5	-700	108	10.5	45	6	C, I, M	Single, precision, high speed, high output drive	5-435
TLE2142	1.2	1.7	4.5	-700	108	10.5	45	6	C, I, M	Dual, precision, high speed, high output drive, low noise	5-463
TLE2142A	0.75	1.7	4.5	-700	108	10.5	45	6	C, I, M	Dual, precision, high speed, high output drive, low noise	5-463
TLE2144	2.4	1.7	4.5	-700	108	10.5	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491
TLE2144A	1.5	1.7	4.5	-700	108	10.5	45	6	C, I, M	Quad, precision, high speed, high output drive, low noise	5-491
TLE2161	3	6	0.35	0.004	90	40	10	6.4	C, I, M	Single, decompensated, micropower, high output drive	5-519
TLE2161A	1.5	6	0.35	0.004	90	40	10	6.4	C, I, M	Single, decompensated, micropower, high output drive	5-519
TLE2227	0.35	0.4	5.3	15	115	2.5	2.5	13	C	Dual, precision, low noise, high speed	5-547
TLE2237	0.35	0.4	5.3	15	115	2.5	5	50	C	Dual, precision, low noise, decompensated	5-569
TLE2301	7	—	21	—	88	44	12	—	I	Single, 3-state output, wide bandwidth	5-589
TLE2662	5	6	620	3	82	43	3.4	—	I	Dual with switched-capacitor voltage converter	5-611
TLE2682	7.5	2.4	3.6	15	89	—	35	—	I	Single with switched-capacitor voltage converter	5-647
TLV2322	9	1.1	0.017	0.0006	94	68	0.03	0.085	I	Dual, low voltage, micropower	5-755
TLV2324	10	1.1	0.017	0.0006	94	68	0.03	0.085	I	Quad, low voltage, micropower	5-779
TLV2332	9	1.7	0.28	0.0006	91	32	0.43	0.525	I	Dual, low voltage, low power	5-803
TLV2334	10	1.7	0.28	0.0006	91	32	0.43	0.525	I	Quad, low voltage, low power	5-827
TLV2341 (high bias)	8	2.7	1.6	0.0006	80	25	3.6	1.7	I	Single, low voltage, programmable power	5-851
TLV2341 (medium bias)	8	1.7	0.28	0.0006	91	32	0.43	0.525	I	Single, low voltage, programmable power	5-851
TLV2341 (low bias)	8	1.1	0.017	0.0006	94	68	0.03	0.085	I	Single, low voltage, programmable power	5-851
TLV2342	9	2.7	1.6	0.0006	80	25	3.6	1.7	I	Dual, low voltage	5-901
TLV2344	10	2.7	1.6	0.0006	80	25	3.6	1.7	I	Quad, low voltage	5-925
TLV2262	2.5	2	0.25	0.001	80	12	0.55	0.82	I	Dual, precision, low voltage, low power, rail to rail	5-695
TLV2262A	2.5	2	0.25	0.001	80	12	0.55	0.82	I	Dual, precision, low voltage, low power, rail to rail	5-695
TLV2264	2.5	2	0.25	0.001	80	12	0.55	0.82	I	Quad, precision, low voltage, low power, rail to rail	5-725
TLV2264A	2.5	2	0.25	0.001	80	12	0.55	0.82	I	Quad, precision, low voltage, low power, rail to rail	5-725
TLV2362	6	—	0.5	20	85	9	3	7	C, I	Dual, low voltage, high speed	5-949
μA741	6	—	2.8	80	90	—	0.5	—	C, I, M	Single, general purpose	5-957

† C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C





# OPERATIONAL AMPLIFIER CROSS-REFERENCE GUIDE

Replacements are based on similarity of electrical and mechanical characteristics shown in currently published data. Interchangeability in particular applications is not guaranteed. Before using a device as a substitute, the user should compare the specifications of the substitute device with the specifications of the original.

Texas Instruments makes no warranty as to the information furnished and the buyer assumes all risk in the use thereof. No liability is assumed for damages resulting from the use of the information contained herein.

Manufacturers are arranged in alphabetical order.

<b>ADVANCED LINEAR DEVICES</b>			
PART NO.	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
ALD1701, ALD1702, or ALD1703		TLC271	2-469
<b>ANALOG DEVICES</b>			
AD510 or AD517		OP07	2-95
AD712J		TLE2082A	5-367
<b>FAIRCHILD</b>			
$\mu$ A714		OP07C	2-95
$\mu$ A714L		OP07D	2-95
$\mu$ A741	$\mu$ A741		5-957
$\mu$ A771		TL071	2-335
$\mu$ A771A		TL071B	2-335
		TL081B	2-361
$\mu$ A771B		TL071A	2-335
		TL081A	2-361
$\mu$ A771L		TL081	2-361
$\mu$ A772		TL072	2-335
$\mu$ A772A		TL072B	2-335
$\mu$ A772B		TL072A	2-335
		TL082A	2-361
$\mu$ A772L		TL082	2-361
$\mu$ A774		FL074	2-335
$\mu$ A774B		TL074A or TL074B	2-335
$\mu$ A774L		TL084	2-361
<b>BURR BROWN</b>			
OPA111		TLC2201	2-795
OPA211		TLC2202	2-825
<b>GENERAL ELECTRIC</b>			
ICL7611, ICL7612, or ICL7613		TLC271	2-469
ICL7621		TLC272	2-537
ICL7641		TLC274	2-573
		TLC27L9	2-657
ICL7642		TLC27M9	2-727



**OPERATIONAL AMPLIFIER  
CROSS-REFERENCE GUIDE**

<b>HARRIS</b>			
<b>PART NO.</b>	<b>DIRECT TI REPLACEMENT</b>	<b>SUGGESTED TI REPLACEMENT</b>	<b>PAGE NO.</b>
HA2515		LM318	2-13
HA5127		TLE2027	5-83
HA5135-5		OP07C	2-95
HA5137		TLE2037	5-109
<b>INTERSIL</b>			
ICL7611, ICL7612, or ICL7613		TLC271	2-469
ICL7621		TLC272	2-537
ICL7641		TLC274	2-573
		TLC27L9	2-657
ICL7642		TLC27M9	2-727
ICL7652		TLC2652	2-993
		TLC2654	2-1017
<b>LINEAR TECHNOLOGY</b>			
LT1001		OP07C or OP07D	2-95
LT1007		TLE2027	5-83
LT1037		TLE2037	5-109
LTC1052		TLC2652	2-993
		TLC2654	2-1017
<b>MAXIM</b>			
ICL7611, ICL7612, or ICL7613		TLC271	2-465
ICL7621		TLC272	2-537
ICL7641		TLC274	2-573
		TLC27L9	2-657
ICL7642		TLC27M9	2-625
ICL7652		TLC2652	2-993
		TLC2654	2-1017
<b>MOTOROLA</b>			
MC1458	MC1458		2-75
MC1558	MC1558		2-75
MC1741	μA741		5-957
MC3403		RC4136	2-101
MC4558	RC4558		2-105
MC4741	LM348		2-17
MC34001		TL071	2-335
		LF351	2-5



**OPERATIONAL AMPLIFIER  
CROSS-REFERENCE GUIDE**

<b>MOTOROLA (CONTINUED)</b>			
PART NO.	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
MC34002		TL072	2-335
		LF353	2-7
MC34004		TL074	2-335
		LF347	2-3
MC34004B		TL074A	2-335
		LF347B	2-3
MC34071		TLE2141	5-435
MC34072		TLE2142	5-463
MC34181		TLE2061	5-137
MC34182		TLE2062	5-167
MC34184		TLE2064	5-197
<b>NATIONAL</b>			
LF347	LF347		2-3
		TL074	2-335
		TL084	2-361
LF347B	LF347B		2-3
		TL074A or TL074B	2-335
		TL084A	2-361
LF351	LF351		2-5
		TL071	2-335
		TL081A	2-361
LF353	LF353		2-7
		TL072 or TL072A	2-335
		TL082A	2-361
LF411	LF411		2-9
		TL081A	2-361
LF411A		TL071A or TL071B	2-335
		TL081A or TL081B	2-361
LF412	LF412		2-11
		TL072A	2-335
		TL082A or TL082B	2-361
LF412-1A		TLE2082	5-367
LF441		TL061	2-291
		TLE2061	5-137
LF441A		TL061A or TL061B	2-291
LF442		TL062	2-291
		TLE2062	5-167



# OPERATIONAL AMPLIFIER CROSS-REFERENCE GUIDE

NATIONAL (CONTINUED)			
PART NO.	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
LF442A		TL062B	2-291
LF444		TL064	2-291
		TLE2064	5-197
LF444A		TL064A	2-291
LH0044		OP07C	2-95
LH0044B		OP07D	2-95
LM201A	LM201A		2-13
LM218	LM218		2-13
LM224	LM224		2-17
LM248	LM248		2-17
LM258	LM258		2-29
LM318	LM318		2-13
LM324	LM324		2-17
		TLE2024	5-57
LM348	LM348		2-17
LM358	LM358		2-29
		TLE2022	5-29
LM741	$\mu$ A741		5-957
LM883		RC4558	2-105
LM1458	MC1458		2-75
LM2900	LM2900		2-43
LM2902	LM2902		2-17
LM2904	LM2904		2-29
LM3900	LM3900		2-43
LMC660		TLC274	2-573
UMC662		TLC2202	2-845
<b>NEC</b>			
uPC159		LM318	2-13
uPC251		MC1458	2-75
uPC354		OP07	2-95
uPC801		TL071	2-335
		TL081A	2-361
		LF351	2-5



# OPERATIONAL AMPLIFIER CROSS-REFERENCE GUIDE

PMI			
PART NO.	DIRECT TI REPLACEMENT	SUGGESTED TI REPLACEMENT	PAGE NO.
OP-02		μA741	5-957
OP-07C	OP07C		2-95
OP-07D	OP07D		2-95
OP-07F		RC4136	2-101
OP-14C or OP-14E		MC1458	2-75
OP-14J		MC1558	2-75
OP-15F		TL071	2-335
		TL081A	2-361
		LF351	2-5
OP-215F		TL072	2-335
		TL082A	2-361
		LF353	2-7
		TLE2082	5-367
OP-215G		TLE2082A	5-367
OP-21		TLE2021	5-3
OP-27		TLE2027	5-83
OP-37		TLE2037	5-109
OP-221		TLE2022	5-29
OP-421		TLE2024	5-57
RAYTHEON			
RC4136	RC4136		2-101
RC4156		LM348	2-17
RC4157		LM348	2-17
RC4558	RC4558		2-105
RCA			
CA081A		TL081	2-361
CA081A		TL081A	2-361
CA082		TL082	2-361
CA082A		TL082A	2-361
CA084		TL084	2-361

**OPERATIONAL AMPLIFIER  
CROSS-REFERENCE GUIDE**

<b>SIGNETICS</b>			
<b>PART NO.</b>	<b>DIRECT TI REPLACEMENT</b>	<b>SUGGESTED TI REPLACEMENT</b>	<b>PAGE NO.</b>
NE532		LM358	2-29
		TL022	2-111
NE5532	NE5532		2-85
NE5532A	NE5532A		2-85
NE5534	NE5534		2-89
		TLE2037	5-109
NE5534A	NE5534A		2-89
		TLE2037A	5-109
SE5534	SE5534		2-89
SE5534A	SE5534A		2-89
<b>SGS-THOMSON</b>			
TS271		TLC271	2-469
TS271A		TLC271A	2-469
TS271B		TLC271B	2-469
TS272		TLC272	2-537
TS272A		TLC272A	2-537
TS272B		TLC272B	2-537
TS274		TLC274	2-573
TS274A		TLC274A	2-573
TS274B		TLC274B	2-573
TS27L2		TLC27L2	2-625
TS27L2A		TLC27L2A	2-625
TS27L2B		TLC27L2B	2-625
TS27L4		TLC27L4	2-657
TS27L4A		TLC27L4A	2-657
TS27L4B		TLC27L4B	2-657
TS27M2		TLC27M2	2-693
TS27M2A		TLC27M2A	2-693
TS27M2B		TLC27M2B	2-693
TS27M4		TLC27M4	2-727
TS27M4A		TLC27M4A	2-727
TS27M4B		TLC27M4B	2-727



**$\alpha_{IIO}$  Average Temperature Coefficient of Input Offset Current**

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{IIO} = \frac{\left( I_{IO} \text{ at } T_{A(1)} \right) - \left( I_{IO} \text{ at } T_{A(2)} \right)}{T_{A(1)} - T_{A(2)}}$$

where  $T_{A(1)}$  and  $T_{A(2)}$  are the specified temperature extremes.

**$\alpha_{VIO}$  Average Temperature Coefficient of Input Offset Voltage**

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range. The dc voltage that must be applied between the input terminals to force the quiescent dc output voltage to zero or other level, if specified.

$$\alpha_{VIO} = \frac{\left( V_{IO} \text{ at } T_{A(1)} \right) - \left( V_{IO} \text{ at } T_{A(2)} \right)}{T_{A(1)} - T_{A(2)}}$$

where  $T_{A(1)}$  and  $T_{A(2)}$  are the specified temperature extremes.

**$\Delta V_{CC}$**

See  $k_{SVS}$

**$\Delta V_{IO}$**

See  $k_{SVS}$

**$\phi_m$  Phase Margin**

The absolute value of the open-loop phase shift between the output and the inverting input at the frequency at which the modulus of the open-loop amplification is unity.

**$A_m$  Gain Margin**

The reciprocal of the open-loop voltage amplification at the lowest frequency at which the open-loop phase shift is such that the output is in phase with the inverting input.

**$A_V$  Large-Signal Voltage Amplification**

The ratio of the peak-to-peak output voltage swing to the change in input voltage required to drive the output

**$A_{VD}$  Differential Voltage Amplification**

The ratio of the change in output to the change in differential input voltage producing it with the common-mode input voltage held constant

**$B_1$  Unity-Gain Bandwidth**

The range of frequencies within which the maximum output voltage swing is above a specified value.

**$B_{OM}$  Maximum-Output-Swing Bandwidth**

The range of frequencies within which the maximum output voltage swing is above the specified value.

**$c_i$  Input Capacitance**

The capacitance between the input terminals with either input grounded

# OPERATIONAL AMPLIFIER GLOSSARY

---

## CMRR, $k_{CMR}$

### Common-Mode Rejection Ratio

The ratio of differential voltage amplification to common-mode voltage amplification.

NOTE: This is measured by determining the ratio of a change in input common-mode voltage to the resulting change in input offset voltage.

## $\bar{F}$ Average Noise Figure

The ratio of an ideal current source (having an internal impedance equal to infinity) in parallel with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a current source.

## $I_{CC+}$ , $I_{CC-}$ Supply Current

The current into the  $V_{CC+}$  or  $V_{CC-}$  terminal of an integrated circuit

## $I_{IB}$ Input Bias Current

The average of the currents into the two input terminals with the output at the specified level

## $I_{IO}$ Input Offset Current

The difference between the currents into the two input terminals with the output at the specified level

## $I_n$ Equivalent Input Noise Current

The current of an ideal current source (having internal impedance equal to infinity) in parallel with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a current source.

## $I_{OL}$ Low-Level Output Current

The current into an output with input conditions applied that according to the product specification will establish a low level at the output.

## $I_{OS}$ Short-Circuit Output Current

The maximum output current available from the amplifier with the output shorted to ground, to either supply, or to a specified point

## $k_{CMR}$

See CMRR

## $k_{SVS}$ , $\Delta V_{CC}$ , $\Delta V_{IO}$ Supply Voltage Sensitivity

The absolute value of the ratio of the change in supply voltages to the change in input offset voltage.

NOTES: 1. Unless otherwise noted, both supply voltages are varied symmetrically.

2. This is the reciprocal of supply voltage sensitivity.

## $k_{SVR}$ Supply Voltage Rejection Ratio

The absolute value of the ratio of the change in supply voltages to the change in input offset voltage.

NOTES: 1. Unless otherwise noted, both supply voltages are varied symmetrically.

2. This is the reciprocal of supply voltage sensitivity.

---

**P<sub>D</sub> Total Power Dissipation**

The total dc power supplied to the device less any power delivered from the device to a load.

NOTE: At no load:  $P_D = V_{CC+} \cdot I_{CC+} + V_{CC-} \cdot I_{CC-}$

**r<sub>i</sub> Input Resistance**

The resistance between the input terminals and either input grounded

**r<sub>id</sub> Differential Input Resistance**

The small-signal resistance between two ungrounded input terminals

**r<sub>o</sub> Output Resistance**

The resistance between an output terminal and ground

**SR Slew Rate**

The average time rate of change of the closed-loop amplifier output voltage for a step-signal input

**t<sub>r</sub> Rise Time**

The time required for an output voltage step to change from 10% to 90% of its final value

**t<sub>tot</sub> Total Response Time**

The time between a step-function change of the input signal and the instant at which the magnitude of the output signal reaches for the last time a specified level range ( $\pm\epsilon$ ) containing the final output signal level.

**V<sub>i</sub> Input Voltage Range**

The range of voltage that if exceeded at either input terminal may cause the operational amplifier to cease functioning properly.

**V<sub>IO</sub> Input Offset Voltage**

The dc voltage that must be applied between the input terminals to force the quiescent dc output voltage to zero or other level, if specified.

**V<sub>IC</sub> Common-Mode Input Voltage**

The average of the two input voltages

**V<sub>ICR</sub> Common-Mode Input Voltage Range**

The range of common-mode input voltage that if exceeded may cause the operational amplifier to cease functioning properly.

**V<sub>n</sub> Equivalent Input Noise Voltage**

The voltage of an ideal voltage source (having internal impedance equal to zero) in series with the input terminals of the device that represents the part of the internally generated noise that can properly be represented by a voltage source.

**V<sub>O1</sub>/V<sub>O2</sub> Crosstalk Attenuation**

The ratio of the change in output voltage of a driven channel to the resulting change in output voltage of another channel

## OPERATIONAL AMPLIFIER GLOSSARY

---

**$V_{OH}$  High-Level Output Voltage**

The voltage at an output with input conditions applied that according to the product specification will establish a high level at the output.

**$V_{OL}$  Low-Level Output Voltage**

The voltage at an output with input conditions applied that according to the product specification will establish a low level at the output.

**$V_{ID}$  Differential Input Voltage**

The voltage at the noninverting input with respect to the inverting input

**$V_{OM}$  Maximum Peak Output Voltage Swing**

The maximum positive or negative peak output voltage that can be obtained without waveform clipping when quiescent dc output voltage is zero.

**$V_{O(PP)}$  Maximum Peak-to-Peak Output Voltage Swing**

The maximum peak-to-peak output voltage that can be obtained without waveform clipping when quiescent dc output voltage is zero.

**$Z_{ic}$  Common-Mode Input Impedance**

The parallel sum of the small-signal impedance between each input terminal and ground

**$Z_o$  Output Impedance**

The small-signal impedance between the output terminal and ground

**Overshoot Factor**

The ratio of the largest deviation of the output signal value from its final steady-state value after a step-function change of the input signal to the absolute value of the difference between the steady-state output signal values before and after the step-function change of the input signal.

## COMPARATORS (Listed Alphanumerically)

DEVICE	V <sub>CC+</sub> V (nom)	V <sub>CC-</sub> V (nom)	V <sub>IO</sub> mV (max)	I <sub>IB</sub> μA (max)	I <sub>OL</sub> mA (min)	RESPONSE TIME ns (typ)	TEMP RANGE	DESCRIPTION	PAGE NO.
LM111	4-30	0	3	0.1	8	115	M	Single, low power, strobe	6-3
LM139	4-30	0	5	-0.1	6	300	M	Single, low power, bipolar	6-19
LM139A	4-30	0	2	-0.1	6	300	M	Single, precision input	6-19
LM193	4-30	0	5	0.1	6	300	M	Dual, low power, bipolar	6-27
LM211	4-30	0	3	0.1	8	115	I	Single, strobe	6-3
LM239	4-30	0	5	-0.25	6	300	I	Quad, low power, bipolar	6-19
LM239A	4-30	0	2	-0.25	6	300	I	Quad, low offset, low power, bipolar	6-19
LM293	4-30	0	5	0.25	6	300	I	Dual, precision input	6-27
LM293A	4-30	0	2	0.25	6	300	I	Dual, low offset, precision input, bipolar	6-27
LM306	12	-6	5	40	100	28	C	Single, strobe	6-33
LM311	4-30	0	7.5	0.25	8	115	C	Single, strobe	6-3
LM339	4-30	0	5	-0.25	6	300	C	Quad, low power, bipolar	6-19
LM339A	4-30	0	2	-0.25	6	300	C	Quad, precision input	6-19
LM339x2	4-30	0	2	-0.25	6	300	C	Octal, precision input	6-41
LM393	4-30	0	5	0.25	6	300	C	Dual, low power	6-27
LM393A	4-30	0	2	0.25	6	300	C	Dual, precision input	6-27
LM2901	4-30	0	7	-0.25	6	300	I	Quad, low power, bipolar	6-19
LM2901Q	4-30	0	7	-0.25	6	300	I	Quad, low power, bipolar	6-19
LM2903	4-30	0	7	0.25	6	300	I	Dual, low power	6-27
LM2903Q	4-30	0	7	0.25	6	300	I	Dual, low power	6-27
LM3902	4-26	0	20	0.5	6	300	I	Quad, low power, bipolar	6-45
LP111	4-30	0	7.5	0.1	1.6	1200	M	Single, low power, strobe	6-49
LP211	4-30	0	7.5	0.1	1.6	1200	I	Single, low power, strobe	6-49
LP239	4-30	0	±5	-0.025	20	8000	I	Quad, ultra-low power, bipolar	6-53
LP311	4-30	0	7.5	0.1	1.6	1200	C	Single, low power, strobe	6-49
LP339	4-30	0	±5	-0.025	6	8000	C	Quad, ultra-low power, bipolar	6-53
TL712	5	0	±1	—	max 16	25	C	Single, output enable	6-65
TL714C	5	0	—	—	max 16	7	C	Single, high speed	6-69
TLC139	4-16	0	5	typ 5 pA	—	2500	M	Quad, low power	6-73
TLC339	4-16	0	5	typ 5 pA	6	1100	C, I, M	Quad, ultra-low power, open-drain output	6-73

T C = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C

# COMPARATOR SELECTION GUIDE

## COMPARATORS (Listed Alphanumerically) (Continued)

DEVICE	VCC+ V (nom)	VCC- V (nom)	VIO mV (max)	I <sub>IB</sub> μA (max)	I <sub>OL</sub> mA (min)	RESPONSE TIME ns (typ)	TEMP <sup>t</sup> RANGE	DESCRIPTION	PAGE NO.
TLC352	1.4 - 18	0	10	typ 5 pA	6	200	C, I, M	Dual, ultra-low supply	6-89
TLC354	1.4 - 18	0	10	typ 5 pA	6	200	C, I, M	Quad, low supply	6-97
TLC371	3 - 16	0	3	typ 5 pA	6	200	C, I, M	Single, high speed	6-107
TLC372	3 - 16	0	5	typ 5 pA	6	200	C, I, M	Dual, high speed	6-117
TLC374	3 - 18	0	10	typ 5 pA	6	200	C, I, M	Quad, high speed, CMOS	6-128
TLC393	3 - 18	0	5	typ 5 pA	6	1100	C, I, M	Dual, ultra-low power, open-drain output	6-141
TLC3702	3 - 18	0	5	typ 5 pA	4	1300	C, I, M	Dual, ultra-low power, push-pull output	6-157
TLC3704	3 - 18	0	5	typ 5 pA	4	1300	C, I, M	Quad, ultra-low power, push-pull output	6-179
TLV1393	2 - 7	0	5	-0.25	0.5	700	I	Dual, low voltage	6-203
TLV2352	2 - 8	0	5	typ 5 pA	6	200	I	Dual, low voltage	6-217
TLV2354	2 - 8	0	5	typ 5 pA	6	200	I	Quad, low voltage	6-229
TLV2393	2 - 7	0	5	-0.25	4	450	I	Dual, low voltage	6-203

T<sub>C</sub> = 0°C to 70°C, I = -40°C to 85°C, M = -55°C to 125°C



## COMPARATOR CROSS-REFERENCE GUIDE

Replacements are based on similarity of electrical and mechanical characteristics shown in currently published data. Interchangeability in particular applications is not guaranteed. Before using a device as a substitute, compare the specifications of the substitute device with the specifications of the original.

Texas Instruments makes no warranty as to the information furnished and the buyer assumes all risk in the use thereof. No liability is assumed for damages resulting from the use of the information contained herein.

Manufacturers are arranged in alphabetical order.

<b>LINEAR TECHNOLOGY</b>			
<b>PART NO.</b>	<b>DIRECT TI REPLACEMENT</b>	<b>SUGGESTED TI REPLACEMENT</b>	<b>PAGE NO.</b>
LT1017		TLC352	6-89
		TLC3702	6-157
LT1018		TLC352	6-89
		TLC3702	6-157
<b>NATIONAL</b>			
LM311	LM311		6-3
LM339	LM339		6-19
		TLC339	6-73
LM393	LM393		6-27
		TLC393	6-141
LM2901	LM2901		6-19
		TLC339	6-73
LM3302	LM3302		6-45
LP339	LP339		6-53
		TLC339	6-73
<b>PMI</b>			
CMP04F		LM339	6-19
		LM2901	6-19
		LM3302	6-45
		TLC339	6-73



**$\alpha_{I_{IO}}$  Average Temperature Coefficient of Input Offset Current**

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{I_{IO}} = \frac{(I_{IO} \text{ at } T_{A(1)}) - (I_{IO} \text{ at } T_{A(2)})}{T_{A(1)} - T_{A(2)}}$$

where  $T_{A(1)}$  and  $T_{A(2)}$  are the specified temperature extremes.

**$\alpha_{V_{IO}}$  Average Temperature Coefficient of Input Offset Voltage**

The ratio of the change in input offset current to the change in free-air temperature. This is an average value for the specified temperature range.

$$\alpha_{V_{IO}} = \frac{(V_{IO} \text{ at } T_{A(1)}) - (V_{IO} \text{ at } T_{A(2)})}{T_{A(1)} - T_{A(2)}}$$

where  $T_{A(1)}$  and  $T_{A(2)}$  are the specified temperature extremes.

**$A_{VD}$  Differential Voltage Amplification**

The ratio of the change in output to the change in differential input voltage producing it with the common-mode input voltage held constant

**CMRR**

See  $k_{CMR}$

**$I_{CC+}, I_{CC-}$  Supply Current**

The current into the  $V_{CC+}$  or  $V_{CC-}$  terminal of an integrated circuit

**$I_{H(S)}$  High-Level Strobe Current**

The current flowing into or out of † the strobe at a high-level voltage

**$I_B$  Input Bias Current**

The average of the currents into the two input terminals with the output at the specified level

**$I_{L(S)}$  Low-Level Strobe Current**

The current flowing out of † the strobe at a low-level voltage

**$I_{IO}$  Input Offset Current**

The difference between the currents into the two input terminals with the output at the specified level

**$I_{OH}$  High-Level Output Current**

The current into an output with input conditions applied that according to the product specification will establish a high level at the output.

**$I_{OL}$  Low-Level Output Current**

The current into an output with input conditions applied that according to the product specification will establish a low level at the output.

**$k_{CMR}$  or CMRR Common-Mode Rejection Ratio**

The ratio of differential voltage amplification to common-mode voltage amplification.

NOTE: This is measured by determining the ratio of a change in input common-mode voltage to the resulting change in input offset voltage.

† Current out of a terminal is given as a negative value.

# COMPARATOR GLOSSARY

---

**P<sub>D</sub> Total Power Dissipation**

The total dc power supplied to the device less any power delivered from the device to a load.

NOTE: At no load:  $P_D = V_{CC+} \cdot I_{CC+} + V_{CC-} \cdot I_{CC-}$

**r<sub>o</sub> Output Resistance**

The resistance between an output terminal and ground

**V<sub>IC</sub> Common-Mode Input Voltage**

The average of the two input voltages

**V<sub>ICR</sub> Common-Mode Input Voltage Range**

The range of common-mode input voltage that if exceeded may cause the comparator to cease functioning properly.

**V<sub>ID</sub> Differential Input Voltage**

The voltage at the noninverting input with respect to the inverting input

**V<sub>ID</sub> Differential Input Voltage Range**

The range of voltage between the two input terminals that if exceeded may cause the comparator to cease functioning properly.

**V<sub>I</sub> Input Voltage Range**

The range of voltage that if exceeded at either input terminal may cause the comparator to cease functioning properly.

**V<sub>IH(S)</sub> High-Level Strobe Voltage**

For a device having an active-low strobe, a voltage within that range is guaranteed not to interfere with the operation of the comparator.

**V<sub>IL(S)</sub> Low-Level Strobe Voltage**

For a device having an active-low strobe, a voltage within the range that is guaranteed to force the output high or low, as specified, independently of the differential inputs.

**V<sub>IO</sub> Input Offset Voltage**

The dc voltage that must be applied between the input terminals to force the quiescent dc output voltage to the specified level.

**V<sub>OH</sub> High-Level Output Voltage**

The voltage at an output with input conditions applied that according to the product specification will establish a high level at the output.

**V<sub>OL</sub> Low-Level Output Voltage**

The voltage at an output with input conditions applied that according to the product specification will establish a low level at the output.

**Response Time**

The interval between the application of an input step function and the instant the output crosses the logic threshold voltage.

NOTE: The input step drives the comparator from some initial condition sufficient to saturate the output (or in the case of high-to-low-level response time, to turn the output off) to an input level just barely in excess of that required to bring the output back to the logic threshold voltage. This excess is referred to as the voltage overdrive.

**Strobe Release Time**

The time required for the output to rise to the logic threshold voltage after the strobe terminal has been driven from its active logic level to its inactive logic level.

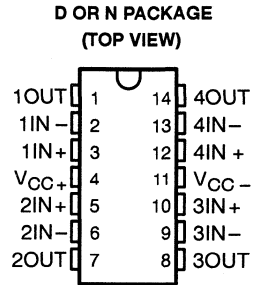


<b>General Information (Volume A)</b>	<b>1</b>
<b>Operational Amplifiers</b>	<b>2</b>
<b>Mechanical Data (Volume A)</b>	<b>3</b>
<b>General Information (Volume B)</b>	<b>4</b>
<b>Operational Amplifiers (continued)</b>	<b>5</b>
<b>Comparators</b>	<b>6</b>
<b>Mechanical Data (Volume B)</b>	<b>7</b>



# Operational Amplifiers

- Low Input Bias Current . . . 50 pA Typ
- Low Input Noise Current  
0.01 pA/√Hz Typ
- Low Total Harmonic Distortion
- Low Supply Current . . . 8 mA Typ
- Wide Gain Bandwidth . . . 3 MHz Typ
- High Slew Rate . . . 13 V/μs Typ
- Pin Compatible With the LM348



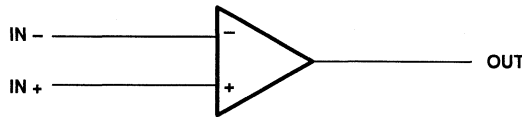
**description**

These devices are low-cost, high-speed, JFET-input operational amplifiers. They require low supply current yet maintain a large gain-bandwidth product and a fast slew rate. In addition, their matched high-voltage JFET inputs provide very low input bias and offset current.

The LF347 and LF347B can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF347 and LF347B are characterized for operation from 0°C to 70°C.

**symbol (each amplifier)**



**AVAILABLE OPTIONS**

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (N)
0°C to 70°C	10 mV	LF347D	LF347N
	5 mV	LF347BD	LF347BN

The D packages are available taped and reeled. Add R suffix to the device type (e.g., LF347DR).

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, V <sub>CC</sub> +	18 V
Supply voltage, V <sub>CC</sub> -	-18 V
Differential input voltage, V <sub>ID</sub>	±30 V
Input voltage, V <sub>I</sub> (see Note 1)	±15 V
Duration of output short circuit	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

# LF347, LF347B WIDE-BANDWIDTH JFET-INPUT QUAD OPERATIONAL AMPLIFIERS

SLOS013B – MARCH 1987 – REVISED AUGUST 1994

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^\circ\text{C}$ POWER RATING
D	608 mW	7.6 mW/°C	61°C	608 mW
N	680 mW	N/A	N/A	680 mW

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	3.5	18	V
Supply voltage, $V_{CC-}$	-3.5	-18	V

## electrical characteristics over operating free-air temperature range, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise specified)

PARAMETER	TEST CONDITIONS	$T_A$ †	LF347			LF347B			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 10\text{ k}\Omega$	25°C		5	10		3	5	mV
		Full range			13			7	
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage	$V_{IC} = 0$ , $R_S = 10\text{ k}\Omega$			18			18		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current‡	$V_{IC} = 0$	25°C		25	100		25	100	pA
		70°C			4			4	nA
$I_{IB}$ Input bias current‡	$V_{IC} = 0$	25°C		50	200		50	200	pA
		70°C			8			8	nA
$V_{ICR}$ Common-mode input voltage range			$\pm 11$	-12 to 15		$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$			$\pm 12$	$\pm 13.5$		$\pm 12$	$\pm 13.5$	V
$A_{VD}$ Large-signal differential voltage	$V_O = \pm 10\text{ V}$ , $R_L = 2\text{ k}\Omega$	25°C		25	100		50	100	V/mV
		Full range			15			25	
$r_i$ Input resistance	$T_A = 25^\circ\text{C}$			$10^{12}$			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$R_S \leq 2\text{ k}\Omega$			70	100		80	100	dB
$k_{SVR}$ Supply-voltage rejection ratio	See Note 2			70	100		80	100	dB
$I_{CC}$ Supply current				8	11		8	11	mA

† Full range is 0°C to 70°C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply-voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

## operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1\text{ kHz}$		120		dB
SR Slew rate		8	13		V/ $\mu\text{s}$
$B_1$ Unity-gain bandwidth			3		MHz
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\text{ }\Omega$		18		$\text{nV}/\sqrt{\text{Hz}}$
$I_n$ Equivalent input noise current	$f = 1\text{ kHz}$		0.01		$\text{pA}/\sqrt{\text{Hz}}$



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

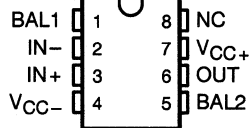


# LF351 WIDE-BANDWIDTH JFET-INPUT OPERATIONAL AMPLIFIER

SLOS014B – MARCH 1987 – REVISED AUGUST 1994

- Low Input Bias Current . . . 50 pA Typ
- Low Input Noise Voltage . . . 18 nV/ $\sqrt{\text{Hz}}$  Typ
- Low Input Noise Current  
0.01 pA/ $\sqrt{\text{Hz}}$  Typ
- Low Supply Current . . . 1.8 mA Typ
- High Input impedance . . .  $10^{12} \Omega$  Typ
- Low Total Harmonic Distortion
- Internally Trimmed Offset Voltage  
10 mV Typ
- High Slew Rate . . . 13 V/ $\mu\text{s}$  Typ
- Wide Gain Bandwidth . . . 3 MHz
- Pin Compatible With Standard 741

D OR P PACKAGE  
(TOP VIEW)



NC – No internal connection

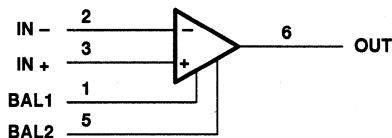
## description

This device is a low-cost, high-speed, JFET-input operational amplifier with an internally trimmed input offset voltage. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents. It uses the same offset voltage adjustment circuits as the 741.

The LF351 can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF351 is characterized for operation from 0°C to 70°C.

## symbol (each amplifier)



AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>I(O)max</sub> AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (P)
0°C to 70°C	10 mV	LF351D	LF351P

The D packages are available taped and reeled. Add the suffix R to the device type (ie., LF351DR).

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC+</sub>	18 V
Supply voltage, V <sub>CC-</sub>	-18 V
Differential input voltage, V <sub>ID</sub>	±30 V
Input voltage, V <sub>I</sub> (see Note 1)	±15 V
Duration of output short circuit	unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

**LF351**  
**WIDE-BANDWIDTH JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS014B – MARCH 1987 – REVISED AUGUST 1994

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	3.5	18	V
Supply voltage, $V_{CC-}$	-3.5	-18	V

**electrical characteristics over operating free-air temperature range,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	MIN	TYP	MAX	UNIT
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 10\text{ k}\Omega$	25°C		5	10	mV
		Full range			13	
$\alpha V_{IO}$ Average temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 10\text{ k}\Omega$			10		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current‡	$V_{IC} = 0$	25°C		25	100	pA
		70°C			4	nA
$I_{IB}$ Input bias current‡	$V_{IC} = 0$	25°C		50	200	pA
		70°C			8	nA
$V_{ICR}$ Common-mode input voltage range			$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$		$\pm 12$	$\pm 13.5$		V
$A_{VD}$ Large-signal differential voltage	$V_O = \pm 10\text{ V}, R_L = 2\text{ k}\Omega$	25°C		25	200	V/mV
		Full range		15	200	
$r_i$ Input resistance	$T_J = 25^\circ\text{C}$			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$R_S \leq 10\text{ k}\Omega$			70	100	dB
$k_{SVR}$ Supply-voltage rejection ratio	See Note 2			70	100	dB
$I_{CC}$ Supply current				1.8	3.4	mA

† Full range is 0°C to 70°C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply-voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

**operating characteristics,  $V_{CC\pm} = \pm 15$  V**

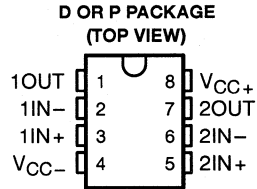
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate		8	13		V/ $\mu\text{s}$
$B_1$ Unity-gain bandwidth			3		MHz
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}, R_S = 20\ \Omega$		18		$\text{nV}/\sqrt{\text{Hz}}$
$I_n$ Equivalent input noise current	$f = 1\text{ kHz}$		0.01		$\text{pA}/\sqrt{\text{Hz}}$



# LF353 WIDE-BANDWIDTH JFET-INPUT DUAL OPERATIONAL AMPLIFIER

SLOS012B – MARCH 1987 – REVISED AUGUST 1994

- Low Input Bias Current . . . 50 pA Typ
- Low Input Noise Current  
0.01 pA/ $\sqrt{\text{Hz}}$  Typ
- Low Input Noise Voltage . . . 18 nV/ $\sqrt{\text{Hz}}$  Typ
- Low Supply Current . . . 3.6 mA Typ
- High Input Impedance . . .  $10^{12} \Omega$  Typ
- Internally Trimmed Offset Voltage
- Wide Gain Bandwidth . . . 3 MHz Typ
- High Slew Rate . . . 13 V/ $\mu\text{s}$  Typ



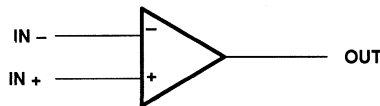
## description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF353 can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF353 is characterized for operation from 0°C to 70°C.

## symbol (each amplifier)



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (P)
0°C to 70°C	10 mV	LF353D	LF353P

The D packages are available taped and reeled. Add the suffix R to the device type (ie., LF353DR).

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC+</sub>	18 V
Supply voltage, V <sub>CC-</sub>	-18 V
Differential input voltage, V <sub>ID</sub>	±30 V
Input voltage, V <sub>I</sub> (see Note 1)	±15 V
Duration of output short circuit	unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

**LF353**  
**WIDE-BANDWIDTH JFET-INPUT**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS012B – MARCH 1987 – REVISED AUGUST 1994

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	3.5	18	V
Supply voltage, $V_{CC-}$	-3.5	-18	V

**electrical characteristics over operating free-air temperature range,  $V_{CC\pm} = \pm 15$  V (unless otherwise specified)**

PARAMETER	TEST CONDITIONS	$T_A$ †	MIN	TYP	MAX	UNIT
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$	25°C		5	10	mV
		Full range			13	
$\alpha V_{IO}$ Average temperature coefficient of input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$			10		$\mu V/^\circ C$
$I_{IO}$ Input offset current‡	$V_{IC} = 0$	25°C		25	100	pA
		70°C			4	nA
$I_{IB}$ Input bias current‡	$V_{IC} = 0$	25°C		50	200	pA
		70°C			8	nA
$V_{ICR}$ Common-mode input voltage range			$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10$ k $\Omega$		$\pm 12$	$\pm 13.5$		V
$A_{VD}$ Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k $\Omega$	25°C		25	100	V/mV
		Full range		15		
$r_i$ Input resistance	$T_J = 25^\circ C$			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$R_S \leq 10$ k $\Omega$		70	100		dB
kSVR Supply-voltage rejection ratio	See Note 2		70	100		dB
$I_{CC}$ Supply current				3.6	6.5	mA

† Full range is 0°C to 70°C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply-voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

**operating characteristics,  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ C$**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1$ kHz		120		dB
SR Slew rate		8	13		V/ $\mu s$
$B_1$ Unity-gain bandwidth			3		MHz
$V_n$ Equivalent input noise voltage	$f = 1$ kHz, $R_S = 20$ $\Omega$		18		nV/ $\sqrt{Hz}$
$I_n$ Equivalent input noise current	$f = 1$ kHz		0.01		pA/ $\sqrt{Hz}$

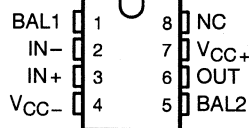


# LF411C JFET-INPUT OPERATIONAL AMPLIFIER

SLOS011B – MARCH 1987 – REVISED AUGUST 1994

- Low Input Bias Current . . . 50 pA Typ
- Low Input Noise Current  
0.01 pA/ $\sqrt{\text{Hz}}$  Typ
- Low Supply Current . . . 2 mA Typ
- High Input impedance . . .  $10^{12} \Omega$  Typ
- Low Total Harmonic Distortion
- Low 1/f Noise Corner . . . 50 Hz Typ

**D OR P PACKAGE  
(TOP VIEW)**



NC – No internal connection

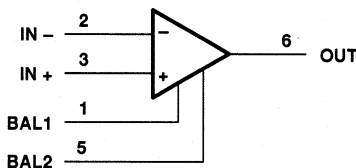
## description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage and a maximum input offset voltage drift. It requires low supply current yet maintains a large gain-bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF411C can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF411C is characterized for operation from 0°C to 70°C.

## symbol



**AVAILABLE OPTIONS**

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (P)
0°C to 70°C	2 mV	LF411CD	LF411CP

The D packages are available taped and reeled. Add the suffix R to the device type (ie., LF411CDR).

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC+</sub> .....	18 V
Supply voltage, V <sub>CC-</sub> .....	–18 V
Differential input voltage, V <sub>ID</sub> .....	±30 V
Input voltage, V <sub>I</sub> (see Note 1) .....	±15 V
Duration of output short circuit .....	unlimited
Continuous total power dissipation .....	500 mW
Operating temperature range .....	0°C to 70°C
Storage temperature range .....	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

# LF411C JFET-INPUT OPERATIONAL AMPLIFIER

SLOS011A – MARCH 1987 – REVISED AUGUST 1994

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	3.5	18	V
Supply voltage, $V_{CC-}$	-3.5	-18	V

## electrical characteristics over operating free-air temperature range, $V_{CC\pm} = \pm 15$ V (unless otherwise specified)

PARAMETER		TEST CONDITIONS	$T_A$ †	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$	25°C		0.8	2	mV
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$			10	20‡	$\mu$ V/°C
$I_{IO}$	Input offset current§	$V_{IC} = 0$	25°C		25	100	pA
			70°C			2	nA
$I_{IB}$	Input bias current§	$V_{IC} = 0$	25°C		50	200	pA
			70°C			4	nA
$V_{ICR}$	Common-mode input voltage range			$\pm 11$	-11.5 to 14.5		V
$V_{OM}$	Maximum peak output voltage swing	$R_L = 10$ k $\Omega$		$\pm 12$	$\pm 13.5$		V
$A_{VD}$	Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k $\Omega$	25°C		25	200	V/mV
			Full range		15	200	
$r_i$	Input resistance	$T_J = 25^\circ\text{C}$			$10^{12}$		$\Omega$
CMRR	Common-mode rejection ratio	$R_S \leq 10$ k $\Omega$		70	100		dB
$k_{SVR}$	Supply-voltage rejection ratio	See Note 2		70	100		dB
$I_{CC}$	Supply current				2	3.4	mA

† Full range is 0°C to 70°C.

‡ At least 90% of the devices meet this limit for  $\alpha_{VIO}$ .

§ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply-voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

## operating characteristics, $V_{CC\pm} = \pm 15$ V, $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR	Slew rate		8	13		V/ $\mu$ s
$B_1$	Unity-gain bandwidth		2.7	3		MHz
$V_n$	Equivalent input noise voltage	$f = 1$ kHz, $R_S = 20$ $\Omega$		18		nV/ $\sqrt{\text{Hz}}$
$I_n$	Equivalent input noise current	$f = 1$ kHz		0.01		pA/ $\sqrt{\text{Hz}}$

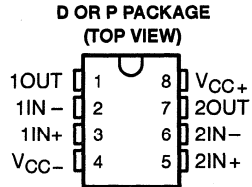


POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# LF412C DUAL JFET-INPUT OPERATIONAL AMPLIFIER

SLOS010B – MARCH 1987 – REVISED AUGUST 1994

- Low Input Bias Current . . . 50 pA Typ
- Low Input Noise Current  
0.01 pA/√Hz Typ
- Low Supply Current . . . 4.5 mA Typ
- High Input Impedance . . . 10<sup>12</sup> Ω Typ
- Internally Trimmed Offset Voltage
- Wide Gain Bandwidth . . . 3 MHz Typ
- High Slew Rate . . . 13 V/μs Typ



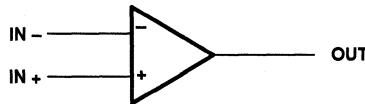
## description

This device is a low-cost, high-speed, JFET-input operational amplifier with very low input offset voltage and a specified maximum input offset voltage drift. It requires low supply current yet maintains a large gain bandwidth product and a fast slew rate. In addition, the matched high-voltage JFET input provides very low input bias and offset currents.

The LF412C can be used in applications such as high-speed integrators, digital-to-analog converters, sample-and-hold circuits, and many other circuits.

The LF412C is characterized for operation from 0°C to 70°C.

## symbol (each amplifier)



### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IOmax</sub> AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (P)
0°C to 70°C	3 mV	LF412CD	LF412CP

The D packages are available taped and reeled. Add the suffix R to the device type (ie., LF412CDR).

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC+</sub>	18 V
Supply voltage, V <sub>CC-</sub>	-18 V
Differential input voltage, V <sub>ID</sub>	±30 V
Input voltage, V <sub>I</sub> (see Note 1)	±15 V
Duration of output short circuit	unlimited
Continuous total power dissipation	500 mW
Operating temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: Unless otherwise specified, the absolute maximum negative input voltage is equal to the negative power supply voltage.

# LF412C

## DUAL JFET-INPUT OPERATIONAL AMPLIFIER

SLOS010B – MARCH 1987 – REVISED AUGUST 1994

### recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	3.5	18	V
Supply voltage, $V_{CC-}$	-3.5	-18	V

### electrical characteristics over operating free-air temperature range, $V_{CC\pm} = \pm 15$ V (unless otherwise specified)

PARAMETER	TEST CONDITIONS	$T_A^\dagger$	MIN	TYP	MAX	UNIT
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$	25°C		1	3	mV
$\alpha V_{IO}$ Average temperature coefficient of input offset voltage	$V_{IC} = 0$ , $R_S = 10$ k $\Omega$			10	20 $\ddagger$	$\mu$ V/°C
$I_{IO}$ Input offset current $\S$	$V_{IC} = 0$	25°C		25	100	pA
		70°C			4	nA
$I_{IB}$ Input bias current $\S$	$V_{IC} = 0$	25°C		50	200	pA
		70°C			8	nA
$V_{ICR}$ Common-mode input voltage range			$\pm 11$	-11.5 to 14.5		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10$ k $\Omega$		$\pm 12$	$\pm 13.5$		V
$A_{VD}$ Large-signal differential voltage	$V_O = \pm 10$ V, $R_L = 2$ k $\Omega$	25°C	25	200		V/mV
		Full range	15	200		
$r_i$ Input resistance	$T_A = 25^\circ\text{C}$			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$R_S \leq 10$ k $\Omega$		70	100		dB
kSVR Supply-voltage rejection ratio	See Note 2		70	100		dB
$I_{CC}$ Supply current				4.5	6.8	mA

$\dagger$  Full range is 0°C to 70°C.

$\ddagger$  At least 90% of the devices meet this limit for  $\alpha V_{IO}$ .

$\S$  Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as possible.

NOTE 2: Supply-voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously.

### operating characteristics, $V_{CC\pm} = \pm 15$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1$ kHz		120		dB
SR Slew rate		8	13		V/ $\mu$ s
$B_1$ Unity-gain bandwidth		2.7	3		MHz
$V_n$ Equivalent input noise voltage	$f = 1$ kHz, $R_S = 20$ $\Omega$		18		nV/ $\sqrt{\text{Hz}}$
$I_n$ Equivalent input noise current	$f = 1$ kHz		0.01		pA/ $\sqrt{\text{Hz}}$



# LM118, LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS063A – JUNE 1976 – REVISED APRIL 1994

- **Small-Signal Bandwidth . . . 15 MHz Typ**
- **Slew Rate . . . 50 V/μs Min**
- **Bias Current . . . 250 nA Max (LM118, LM218)**
- **Supply Voltage Range . . . ±5 V to ±20 V**
- **Internal Frequency Compensation**
- **Input and Output Overload Protection**
- **Same Pin Assignments as General-Purpose Operational Amplifiers**

## description

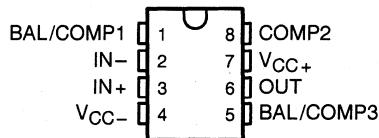
The LM118, LM218, and LM318 are precision, high-speed operational amplifiers designed for applications requiring wide bandwidth and high slew rate. They feature a factor-of-ten increase in speed over general-purpose devices without sacrificing dc performance.

These operational amplifiers have internal unity-gain frequency compensation. This considerably simplifies their application, since no external components are necessary for operation. However, unlike most internally compensated amplifiers, external frequency compensation may be added for optimum performance. For inverting applications, feed-forward compensation boosts the slew rate to over 150 V/μs and almost double the bandwidth. Overcompensation can be used with the amplifier for greater stability when maximum bandwidth is not needed. Further, a single capacitor may be added to reduce the settling time for 0.1% error band to under 1 μs.

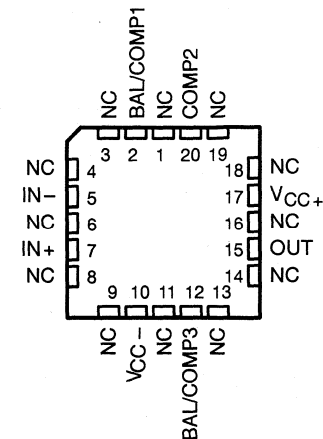
The high speed and fast settling time of these operational amplifiers make them useful in A/D converters, oscillators, active filters, sample-and-hold circuits, and general-purpose amplifiers.

The LM118 is characterized for operation from –55°C to 125°C. The LM218 is characterized for operation from –25°C to 85°C, and the LM318 is characterized for operation from 0°C to 70°C.

**D, JG, OR P PACKAGE  
(TOP VIEW)**

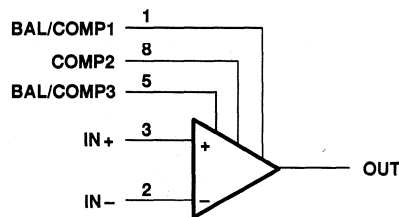


**FK PACKAGE  
(TOP VIEW)**



NC – No internal connection

## symbol



Pin numbers shown are for the D, JG, and P packages.

# LM118, LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

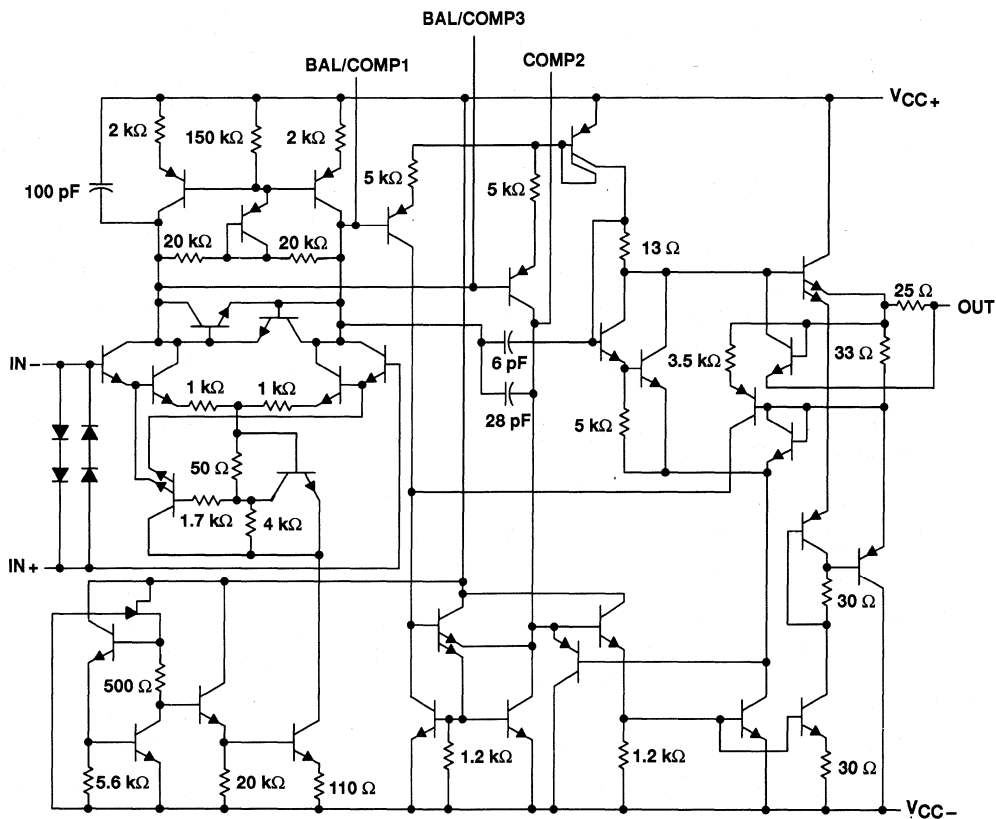
SLOS063A - JUNE 1976 - REVISED APRIL 1994

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	10 mV	LM318D	—	—	LM318P
-25°C to 85°C	4 mV	LM218D	—	—	LM218P
-55°C to 125°C	4 mV	LM118D	LM118FK	LM118JG	LM118P

The D package is available taped and reeled. Add the suffix R to the device type (e.g., LM318DR).

## schematic



Component values shown are nominal.

# LM118, LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS063A – JUNE 1976 – REVISED APRIL 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM118	LM218	LM318	UNIT	
Supply voltage, $V_{CC+}$ (see Note 1)	20	20	20	V	
Supply voltage, $V_{CC-}$ (see Note 1)	-20	-20	-20	V	
Input voltage, $V_I$ (either input, see Notes 1 and 2)	$\pm 15$	$\pm 15$	$\pm 15$	V	
Differential input current, $V_{ID}$ (see Note 3)	$\pm 10$	$\pm 10$	$\pm 10$	mA	
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited		
Continuous total power dissipation	See Dissipation Rating Table				
Operating free-air temperature range, $T_A$	-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$	
Storage temperature range	-65 to 150	-65 to 150	-65 to 50	$^{\circ}\text{C}$	
Case temperature for 60 seconds	FK package	260		$^{\circ}\text{C}$	
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	260	260	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG package	300		$^{\circ}\text{C}$	

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
3. The inputs are shunted with two opposite-facing base-emitter diodes for overvoltage protection. Therefore, excessive current flows if a different input voltage in excess of approximately 1 V is applied between the inputs unless some limiting resistance is used.
4. The output can be shorted to ground or either power supply. For the LM118 and LM218 only, the unlimited duration of the short circuit applies at (or below) 85 $^{\circ}\text{C}$  case temperature or 75 $^{\circ}\text{C}$  free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	500 mW	5.8 mW/ $^{\circ}\text{C}$	64 $^{\circ}\text{C}$	464 mW	377 mW	145 mW
FK	500 mW	11.0 mW/ $^{\circ}\text{C}$	105 $^{\circ}\text{C}$	500 mW	500 mW	275 mW
JG	500 mW	8.4 mW/ $^{\circ}\text{C}$	90 $^{\circ}\text{C}$	500 mW	500 mW	210 mW
P	500 mW	8.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	500 mW	500 mW	200 mW

# LM118, LM218, LM318 HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS063A – JUNE 1976 – REVISED APRIL 1994

## electrical characteristics at specified free-air temperature (see Note 5)

PARAMETER	TEST CONDITION†	T <sub>A</sub> ‡	LM118, LM218			LM318			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0	25°C		2	4		4	10	mV
		Full range			6			15	
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0	25°C		6	50		30	200	nA
		Full range			100			300	
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0	25°C		120	250		150	500	nA
		Full range			500			750	
V <sub>ICR</sub> Common-mode input voltage range	V <sub>CC±</sub> = ±15 V	Full range	±11.5			±11.5			V
V <sub>OM</sub> Maximum peak output voltage swing	V <sub>CC±</sub> = ±15 V, R <sub>L</sub> = 2 kΩ	Full range	±12	±13		±12	±13	V	
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>CC±</sub> = ±15 V, V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	25°C	50	200		25	200	V/mV	
		Full range	25			20			
B <sub>1</sub> Unity-gain bandwidth	V <sub>CC±</sub> = ±15 V	25°C		15		15		MHz	
r <sub>i</sub> Input resistance		25°C	1*	3		0.5	3	MΩ	
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	Full range	80	100		70	100	dB	
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC</sub> /ΔV <sub>IO</sub> )		Full range	70	80		65	80	dB	
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C		5	8		5	10	mA

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† All characteristics are measured under open-loop conditions with common-mode input voltage unless otherwise specified.

‡ Full range for LM118 is -55°C to 125°C, full range for LM218 is -25°C to 85°C, and full range for LM318 is 0°C to 70°C.

NOTE 5: Unless otherwise noted, V<sub>CC</sub> = ±5 V to ±20 V. All typical values are at V<sub>CC±</sub> = ±15 V and T<sub>A</sub> = 25°C.

## operating characteristics, V<sub>CC±</sub> = ±15 V, T<sub>A</sub> = 25°C

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slewing rate at unity gain	ΔV <sub>I</sub> = 10 V, C <sub>L</sub> = 100 pF, See Figure 1	50*	70		V/μs

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

## PARAMETER MEASUREMENT INFORMATION

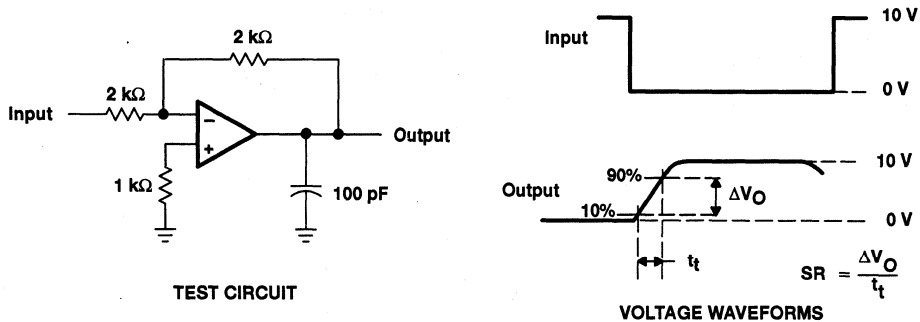


Figure 1. Slewing Rate

# LM124, LM124A, LM224, LM224A LM324, LM324A, LM324Y, LM2902, LM2902Q QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS066B – SEPTEMBER 1975 – REVISED MARCH 1994

- **Wide Range of Supply Voltages:**  
Single Supply . . . 3 V to 30 V  
(LM2902 and LM2902Q  
3 V to 26 V), or Dual Supplies
- **Low Supply Current Drain Independent of Supply Voltage** . . . 0.8 mA Typ
- **Common-Mode Input Voltage Range Includes Ground Allowing Direct Sensing Near Ground**
- **Low Input Bias and Offset Parameters:**  
Input Offset Voltage . . . 3 mV Typ  
A Versions . . . 2 mV Typ  
Input Offset Current . . . 2 nA Typ  
Input Bias Current . . . 20 nA Typ  
A Versions . . . 15 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage** . . . 32 V  
(26 V for LM2902 and LM2902Q)
- **Open-Loop Differential Voltage Amplification** . . . 100 V/mV Typ
- **Internal Frequency Compensation**

## description

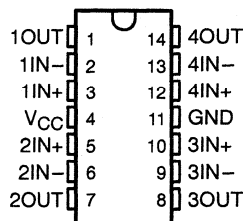
These devices consist of four independent high-gain frequency-compensated operational amplifiers that are designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible when the difference between the two supplies is 3 V to 30 V (for the LM2902 and LM2902Q, 3 V to 26 V) and  $V_{CC}$  is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, the LM124 can be operated directly from the standard 5-V supply that is used in digital systems and easily provides the required interface electronics without requiring additional  $\pm 15$ -V supplies.

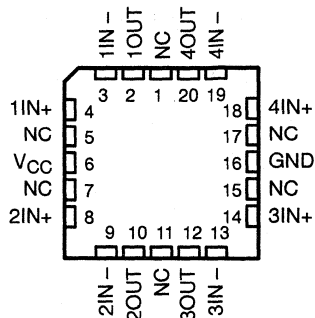
The LM2902Q is manufactured to demanding automotive requirements.

The LM124 and LM124A are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The LM224 and LM224A are characterized for operation from  $-25^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The LM324 and LM324A are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ . The LM2902 and LM2902Q are characterized for operation from  $-40^{\circ}\text{C}$  to  $105^{\circ}\text{C}$ .

LM124, LM224A . . . J OR W PACKAGE  
ALL OTHERS . . . D, DB, J, N OR PW PACKAGE  
(TOP VIEW)

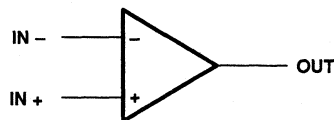


LM124, LM124A . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## symbol (each amplifier)



# LM124, LM124A, LM224, LM224A LM324, LM324A, LM324Y, LM2902, LM2902Q QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS066B - SEPTEMBER 1975 - REVISED MARCH 1994

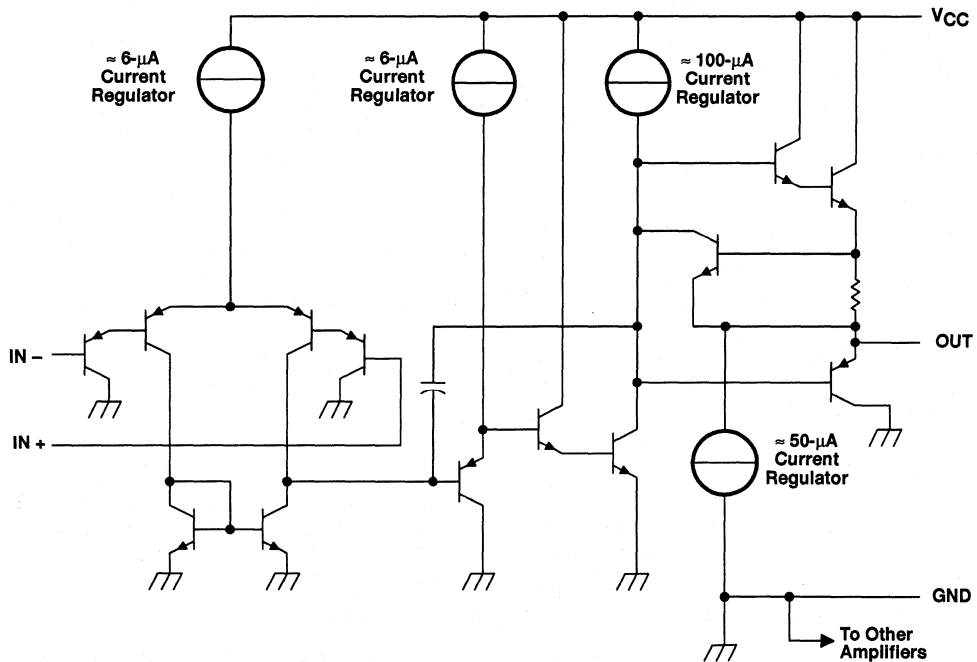
## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES							CHIP FORM (Y)
		SMALL OUTLINE (D)†	VERY SMALL OUTLINE (DB)‡	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)‡	FLAT PACK (W)	
0°C to 70°C	7 mV 3 mV	LM324D LM324AD	LM324DBLE —	—	—	LM324N LM324AN	LM324PWLE LM324APWLE	—	LM324Y
25°C to 85°C	5 mV 3 mV	LM224D LM224AD	—	—	—	LM224N LM224AN	—	—	—
-40°C to 105°C	7 mV	LM2902D LM2902QD	LM2902DBLE	—	—	LM2902N LM2902QN	LM2902PWLE	—	—
-55°C to 125°C	5 mV 2 mV	—	—	LM124FK LM124AFK	LM124J LM124AJ	—	—	LM124W	—

† The D package is available taped and reeled. Add the suffix R to the device type (e.g., LM324DR).

‡ The DB and PW packages are only available left-end taped and reeled.

## schematic (each amplifier)



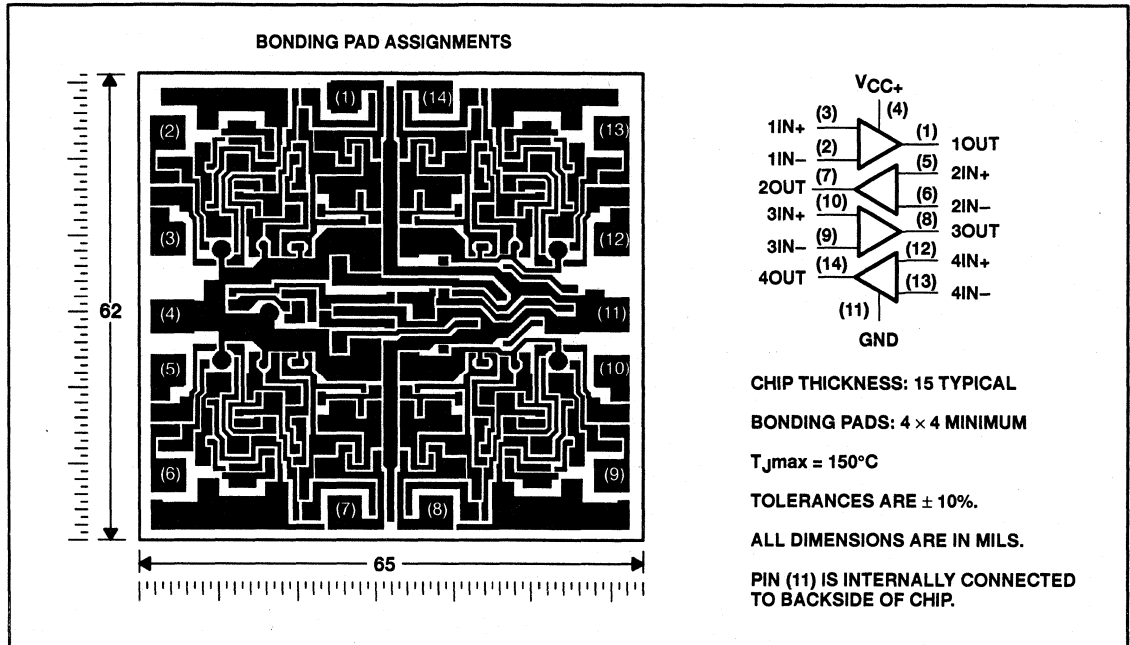
COMPONENT COUNT (total device)	
Epi-FET	1
Transistors	95
Diodes	4
Resistors	11
Capacitors	4

**LM124, LM124A, LM224, LM224A**  
**LM324, LM324A, LM324Y, LM2902, LM2902Q**  
**QUADRUPLE OPERATIONAL AMPLIFIERS**

SLOS066B – SEPTEMBER 1975 – REVISED MARCH 1994

**LM324Y chip information**

This chip, when properly assembled, displays characteristics similar to the LM324. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**LM124, LM124A, LM224, LM224A  
LM324, LM324A, LM324Y, LM2902, LM2902Q  
QUADRUPLE OPERATIONAL AMPLIFIERS**

SLOS066B – SEPTEMBER 1975 – REVISED MARCH 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

		LM124, LM124A LM224, LM224A LM324, LM324A	LM2902, LM2902Q	UNIT
Supply voltage, $V_{CC}$ (see Note 1)		32	26	V
Differential input voltage, $V_{ID}$ (see Note 2)		$\pm 32$	$\pm 26$	V
Input voltage, $V_I$ (either input)		-0.3 to 32	-0.3 to 26	V
Duration of output short circuit (one amplifier) to ground at (or below) $T_A = 25^\circ\text{C}$ , $V_{CC} \leq 15\text{ V}$ (see Note 3)		unlimited	unlimited	
Continuous total dissipation		See Dissipation Rating Table		
Operating free-air temperature range, $T_A$	LM124, LM124A	-55 to 125		°C
	LM224, LM224A	-25 to 85		
	LM324, LM324A	0 to 70		
	LM2902, LM2902Q		-40 to 105	
Storage temperature range		-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package	260		°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or W package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, DB, N, or PW package	260	260	°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values (except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{OS}$ ) are with respect to the network GND.  
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
3. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	900 mW	7.6 mW/°C	32°C	608 mW	494 mW	N/A
DB	775 mW	6.2 mW/°C	25°C	496 mW	403 mW	N/A
FK	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (LM124_)	900 mW	11.0 mW/°C	68°C	880 mW	715 mW	275 mW
J (all others)	900 mW	8.2 mW/°C	40°C	656 mW	533 mW	N/A
N	900 mW	9.2 mW/°C	52°C	736 mW	598 mW	N/A
PW	700 mW	5.6 mW/°C	25°C	448 mW	364 mW	N/A
W	900 mW	8.0 mW/°C	37°C	640 mW	520 mW	200 mW





**LM124, LM124A, LM224, LM224A  
LM324, LM324A, LM324Y, LM2902, LM2902Q  
QUADRUPLE OPERATIONAL AMPLIFIERS**

SLOS066B - SEPTEMBER 1975 - REVISED MARCH 1994

**electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONST	TA †	LM124, LM224			LM324			LM2902, LM2902Q			UNIT
			MIN	TYP§	MAX	MIN	TYP§	MAX	MIN	TYP§	MAX	
V <sub>IO</sub>	Input offset voltage	25°C	3	3	5	3	3	7	3	3	7	mV
		Full range			7			9			10	
I <sub>IO</sub>	Input offset current	25°C	2	30	30	2	50	50	2	50	50	nA
		Full range			100			150			200	
I <sub>IB</sub>	Input bias current	25°C	-20	-150	-300	-20	-250	-250	-20	-250	-250	nA
		Full range			-500			-500			-500	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$			0 to $V_{CC}-1.5$			V
		Full range	0 to $V_{CC}-2$			0 to $V_{CC}-2$			0 to $V_{CC}-2$			
V <sub>OH</sub>	High-level output voltage	25°C	$R_L = 2\text{ k}\Omega$			$R_L = 2\text{ k}\Omega$			$R_L = 2\text{ k}\Omega$			V
		Full range	$R_L = 10\text{ k}\Omega$			$R_L = 10\text{ k}\Omega$			$R_L = 10\text{ k}\Omega$			
V <sub>OL</sub>	Low-level output voltage	25°C	$V_{CC} = \text{MAX}$ , $R_L = 2\text{ k}\Omega$			$V_{CC} = \text{MAX}$ , $R_L = 2\text{ k}\Omega$			$V_{CC} = \text{MAX}$ , $R_L = 2\text{ k}\Omega$			V
		Full range	$V_{CC} = \text{MAX}$ , $R_L \geq 10\text{ k}\Omega$			$V_{CC} = \text{MAX}$ , $R_L \geq 10\text{ k}\Omega$			$V_{CC} = \text{MAX}$ , $R_L \geq 10\text{ k}\Omega$			
A <sub>V/D</sub>	Large-signal differential voltage amplification	25°C	$R_L \leq 10\text{ k}\Omega$	5	20	5	20	5	20	5	20	mV
		Full range	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V}$ to $11\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	50	100	25	100	100	100	100	100	
CMRR	Common-mode rejection ratio	25°C	$V_{IC} = V_{ICRmin}$	70	80	65	80	65	80	50	80	dB
		Full range		65	100	65	100	65	100	50	100	
K <sub>SVR</sub>	Supply-voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )	25°C		120		120		120	120		120	dB
		Full range		-20	-30	-60	-20	-30	-60	-20	-30	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	25°C	$f = 1\text{ kHz}$ to $20\text{ kHz}$									dB
		Full range	$V_{CC} = 15\text{ V}$ , $V_{ID} = 1\text{ V}$ , $V_O = 0$									
I <sub>O</sub>	Output current	25°C	$V_{CC} = 15\text{ V}$ , $V_O = 15\text{ V}$	10	20	10	20	10	20	10	20	mA
		Full range	$V_{ID} = -1\text{ V}$ , $V_O = 200\text{ mV}$	5	30	12	30	12	30	5	30	
I <sub>OS</sub>	Short-circuit output current	25°C	$V_{CC}$ at $5\text{ V}$ , $GND$ at $-5\text{ V}$	$\pm 40$	$\pm 60$	$\pm 40$	$\pm 60$	$\pm 40$	$\pm 60$	$\pm 40$	$\pm 60$	mA
		Full range	$V_O = 2.5\text{ V}$ , No load	0.7	1.2	0.7	1.2	0.7	1.2	0.7	1.2	
I <sub>CC</sub>	Supply current (four amplifiers)	25°C	$V_{CC} = \text{MAX}$ , $V_O = 0.5 V_{CC}$	1.4	3	1.4	3	1.4	3	1.4	3	mA
		Full range	$V_{CC} = \text{MAX}$ , $V_O = 0.5 V_{CC}$	1.4	3	1.4	3	1.4	3	1.4	3	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 26 V for LM2902 and LM2902Q, 30 V for the others.  
 ‡ Full range is -55°C to 125°C for LM124, -25°C to 85°C for LM224, 0°C to 70°C for LM324, and -40°C to 105°C for LM2902 and LM2902Q.  
 § All typical values are at  $T_A = 25^\circ\text{C}$ .



**LM124, LM124A, LM224, LM224A**  
**LM324, LM324A, LM324Y, LM2902, LM2902Q**  
**QUADRUPLE OPERATIONAL AMPLIFIERS**

SLOS068B – SEPTEMBER 1975 – REVISED MARCH 1994

**electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS†	T <sub>A</sub> ‡	LM124A			LM224A			LM324A			UNIT
			MIN	TYP§	MAX	MIN	TYP§	MAX	MIN	TYP§	MAX	
V <sub>IO</sub>	V <sub>CC</sub> = 5 V to 30 V, V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 1.4 V	25°C Full range	2 4	2 4	2 4	2 4	2 4	2 4	2 4	2 4	3 5	mV
I <sub>IO</sub>	V <sub>O</sub> = 1.4 V	25°C Full range	10 30	10 30	2 30	2 30	2 30	2 30	2 30	2 30	30 75	nA
I <sub>IB</sub>	V <sub>O</sub> = 1.4 V	25°C Full range	-50 -100	-50 -100	-15 -80	-15 -80	-15 -80	-15 -80	-15 -80	-15 -80	-100 -200	nA
V <sub>ICR</sub>	V <sub>CC</sub> = 30 V	25°C Full range	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	0 to V <sub>CC</sub> -1.5	V
V <sub>OH</sub>	R <sub>L</sub> = 2 kΩ V <sub>CC</sub> = 30 V, R <sub>L</sub> = 2 kΩ	25°C Full range	26	27	26	27	26	27	26	27	28	V
V <sub>OL</sub>	R <sub>L</sub> = 10 kΩ V <sub>CC</sub> = 30 V, R <sub>L</sub> = 10 kΩ	25°C Full range	20	20	5	20	5	20	5	20	20	mV
A <sub>VD</sub>	V <sub>CC</sub> = 15 V, V <sub>O</sub> = 1 V to 11 V, R <sub>L</sub> = ≥ 2 kΩ	25°C Full range	25	25	25	25	25	25	25	25	15	V/mV
CMRR	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	70	70	70	80	70	80	70	80	80	dB
K <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC</sub> /ΔV <sub>O</sub> )	25°C	65	65	65	100	65	100	65	100	100	dB
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	25°C	120	120	120	120	120	120	120	120	120	dB
I <sub>O</sub>	V <sub>CC</sub> = 15 V, V <sub>ID</sub> = 1 V, V <sub>O</sub> = 0	25°C Full range	-20 -10	-20 -10	-20 -10	-20 -10	-20 -10	-20 -10	-20 -10	-20 -10	-30 -60	mA
I <sub>OS</sub>	V <sub>CC</sub> = 15 V, V <sub>ID</sub> = -1 V, V <sub>O</sub> = 15 V	25°C Full range	10 5	10 5	10 5	10 5	10 5	10 5	10 5	10 5	20	mA
I <sub>CC</sub>	V <sub>ID</sub> = -1 V, V <sub>O</sub> = 200 mV V <sub>CC</sub> at 5 V, GND at -5 V, V <sub>O</sub> = 0	25°C	12	12	±40	±60	12	30	12	30	30	μA
	V <sub>O</sub> = 2.5 V, No load	Full range	0.7	1.2	0.7	1.2	0.7	1.2	0.7	1.2	1.2	mA
	V <sub>CC</sub> = 30 V, V <sub>O</sub> = 15 V, No load	Full range	1.4	3	1.4	3	1.4	3	1.4	3	3	mA

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

‡ Full range is -55°C to 125°C for LM124A, -25°C to 85°C for LM224A, and 0°C to 70°C for LM324A.

§ All typical values are at T<sub>A</sub> = 25°C.



**LM124, LM124A, LM224, LM224A  
LM324, LM324A, LM324Y, LM2902, LM2902Q  
QUADRUPLE OPERATIONAL AMPLIFIERS**

SLOS066B – SEPTEMBER 1975 – REVISED MARCH 1994

**electrical characteristics,  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITION†	LM324Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{CC} = 5\text{ V to MAX}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$		3	7	mV
$I_{IO}$ Input offset current			2	50	nA
$I_{IB}$ Input bias current			-20	-250	nA
$V_{ICR}$ Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	0 to $V_{CC}-1.5$			V
$V_{OH}$ High-level output voltage	$R_L = 10\text{ k}\Omega$	$V_{CC}-1.5$			V
$V_{OL}$ Low-level output voltage	$R_L \leq 10\text{ k}\Omega$		5	20	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V to }11\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	15	100		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	80		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )		65	100		dB
$I_O$ Output current	$V_{CC} = 15\text{ V}$ , $V_{ID} = 1\text{ V}$ , $V_O = 0$	-20	-30	-60	mA
	$V_{CC} = 15\text{ V}$ , $V_{ID} = -1\text{ V}$ , $V_O = 15\text{ V}$	10	20		
	$V_{ID} = 1\text{ V}$ , $V_O = 200\text{ mV}$	12	30		
$I_{OS}$ Short-circuit output current	$V_{CC}$ at 5 V, GND at -5 V, $V_O = 0$		$\pm 40$	$\pm 60$	mA
$I_{CC}$ Supply current (four amplifiers)	$V_O = 2.5 V_{CC}$ , No load		0.7	1.2	mA
	$V_{CC} = \text{MAX}$ , $V_O = 0.5 V_{CC}$ , No load		1.1	3	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 30 V.



# LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS058A – OCTOBER 1979 – REVISED DECEMBER 1992

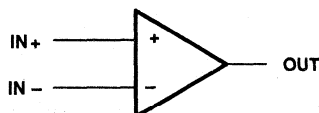
- $\mu$ A741 Operating Characteristics
- Low Supply Current Drain . . . 0.6 mA Typ (per amplifier)
- Low Input Offset Voltage
- Low Input Offset Current
- Class AB Output Stage
- Input/Output Overload Protection
- Designed to Be Interchangeable With National LM148, LM248, and LM348

## description

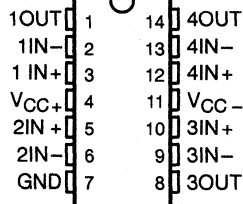
The LM148, LM248, and LM348 are quadruple, independent, high-gain, internally compensated operational amplifiers designed to have operating characteristics similar to the  $\mu$ A741. These amplifiers exhibit low supply current drain, and input bias and offset currents that are much less than those of the  $\mu$ A741.

The LM148 is characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ , the LM248 is characterized for operation from  $-25^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , and the LM348 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

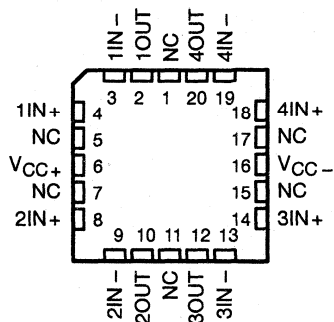
## symbol (each amplifier)



LM148 . . . J PACKAGE  
LM248, LM348 . . . D, N, OR PW PACKAGE  
(TOP VIEW)



LM148 . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## AVAILABLE OPTIONS

$T_A$	$V_{IOmax}$ AT $25^{\circ}\text{C}$	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)
$0^{\circ}\text{C}$ to $70^{\circ}\text{C}$	6 mV	LM348D	—	—	LM348N	LM348PW
$-25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	6 mV	LM248D	—	—	LM248N	—
$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$	5 mV	—	LM148FK	LM148J	—	—

The D package is available taped and reeled. Add the suffix R to the device type (e.g., LM348DR).

# LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS058A – OCTOBER 1979 – REVISED DECEMBER 1992

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM148	LM248	LM348	UNIT
Supply voltage, $V_{CC+}$ (see Note 1)	22	18	18	V
Supply voltage, $V_{CC-}$ (see Note 1)	-22	-18	-18	V
Differential input voltage, $V_{ID}$ (see Note 2)	44	36	36	V
Input voltage, $V_I$ (either input, see Notes 1 and 3)	$\pm 22$	$\pm 18$	$\pm 18$	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total power dissipation	See Dissipation Rating Table			
Operating free-air temperature range, $T_A$	-55 to 125	-25 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds	FK package	260		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J package	300		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N, or PW package		260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or the value specified in the table, whichever is less.  
 4. The output may be shorted to ground or either power supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	900 mW	7.6 mW/ $^{\circ}\text{C}$	32 $^{\circ}\text{C}$	608 mW	494 mW	N/A
FK	900 mW	11.0 mW/ $^{\circ}\text{C}$	68 $^{\circ}\text{C}$	880 mW	715 mW	275 mW
J	900 mW	11.0 mW/ $^{\circ}\text{C}$	68 $^{\circ}\text{C}$	880 mW	715 mW	275 mW
N	900 mW	9.2 mW/ $^{\circ}\text{C}$	52 $^{\circ}\text{C}$	736 mW	598 mW	N/A
PW	700 mW	5.6 mW/ $^{\circ}\text{C}$	N/A	448 mW	N/A	N/A

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	4	18	V
Supply voltage, $V_{CC-}$	-4	-18	V

# LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS058A – OCTOBER 1979 – REVISED DECEMBER 1992

electrical characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM148			LM248			LM348			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$	25°C	1	5	1	1	6	1	1	6	mV
		Full range		6		7.5				7.5	
$I_{IO}$ Input offset current	$V_O = 0$	25°C	4	25	4	4	50	4	4	50	nA
		Full range		75		125				100	
$I_{IB}$ Input bias current	$V_O = 0$	25°C	30	100	30	30	200	30	30	200	nA
		Full range		325		500				400	
$V_{ICR}$ Common-mode input voltage range		Full range	$\pm 12$		$\pm 12$			$\pm 12$			V
		25°C	$\pm 12$	$\pm 13$	$\pm 12$	$\pm 13$	$\pm 12$	$\pm 13$	$\pm 12$	$\pm 13$	
VOM Maximum peak output voltage swing		Full range	$\pm 10$		$\pm 12$			$\pm 10$			V
		25°C	$\pm 10$	$\pm 12$	$\pm 12$	$\pm 12$	$\pm 10$	$\pm 12$	$\pm 10$	$\pm 12$	
		25°C	$\pm 10$	$\pm 12$	$\pm 10$	$\pm 12$	$\pm 10$	$\pm 12$	$\pm 10$	$\pm 12$	
		Full range	$\pm 10$		$\pm 10$				$\pm 10$		
AVD Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	25°C	50	160	25	160	25	160	25	160	V/mV
		Full range	25		15				15		
$r_i$ Input resistance‡		25°C	0.8	2.5	0.8	2.5	0.8	2.5	0.8	2.5	M $\Omega$
$B_1$ Unity-gain bandwidth	$AVD = 1$	25°C	1		1			1			MHz
$\phi_m$ Phase margin	$AVD = 1$	25°C	60°		60°			60°			
		25°C	70	90	70	90	70	90	70	90	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$ , $V_O = 0$	25°C	70	90	70	90	70	90	70	90	dB
		Full range	70		70				70		
kSVR Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC\pm} = \pm 9\text{ V to } \pm 15\text{ V}$ , $V_O = 0$	25°C	77	96	77	96	77	96	77	96	dB
		Full range	77		77				77		
$I_{OS}$ Short-circuit output current		25°C	$\pm 25$		$\pm 25$			$\pm 25$			mA
$I_{CC}$ Supply current (four amplifiers)	No load	25°C	2.4	3.6	2.4	4.5	2.4	4.5	2.4	4.5	mA
		Full range									
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1\text{ Hz to } 20\text{ kHz}$	25°C	120		120			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for  $T_A$  is  $-55^\circ\text{C to } 125^\circ\text{C}$  for LM148,  $-25^\circ\text{C to } 85^\circ\text{C}$  for LM248, and  $0^\circ\text{C to } 70^\circ\text{C}$  for LM348.

‡ This parameter is not production tested.



# LM148, LM248, LM348 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS058A – OCTOBER 1979 – REVISED DECEMBER 1992

operating characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$ , See Figure 1		0.5		$\text{V}/\mu\text{s}$

## PARAMETER MEASUREMENT INFORMATION

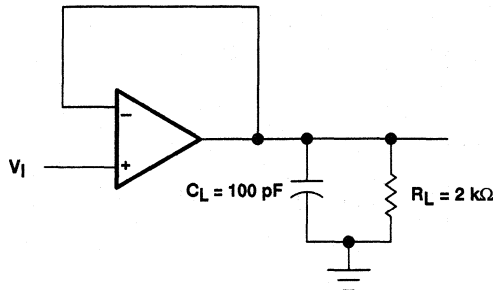


Figure 1. Unity-Gain Amplifier

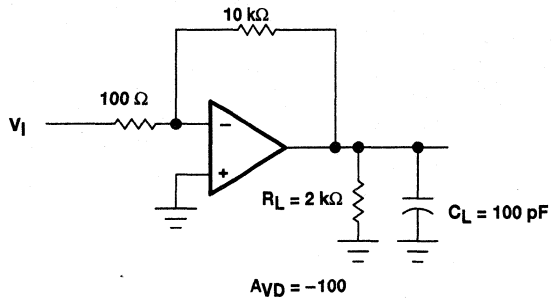


Figure 2. Inverting Amplifier

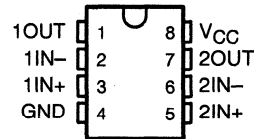


# LM158, LM258, LM358, LM158A LM258A, LM358A, LM358Y, LM2904, LM2904Q DUAL OPERATIONAL AMPLIFIERS

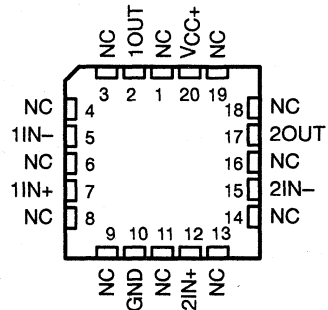
SLOS068 – JUNE 1976 – REVISED JULY 1991

- **Wide Range of Supply Voltages:**  
Single Supply . . . 3 V to 30 V  
(LM2904 and LM2904Q  
3 V to 26 V) or Dual Supplies
- **Low Supply Current Drain Independent of Supply Voltage . . . 0.7 mA Typ**
- **Common-Mode Input Voltage Range Includes Ground Allowing Direct Sensing Near Ground**
- **Low Input Bias and Offset Parameters:**  
Input Offset Voltage . . . 3 mV Typ  
A Versions . . . 2 mV Typ  
Input Offset Current . . . 2 nA Typ  
Input Bias Current . . . 20 nA Typ  
A Versions . . . 15 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . .  $\pm 32$  V**  
( $\pm 26$  V for LM2904 and LM2904Q)
- **Open-Loop Differential Voltage Amplification . . . 100 V/mV Typ**
- **Internal Frequency Compensation**

D, DB, JG, P, OR PW PACKAGE  
(TOP VIEW)



LM158, LM158A . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## description

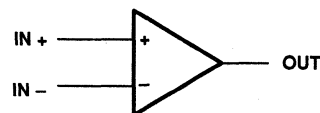
These devices consist of two independent, high-gain, frequency-compensated operational amplifiers that were designed specifically to operate from a single supply over a wide range of voltages. Operation from split supply is also possible so long as the difference between the two supplies is 3 V to 30 V (3 V to 26 V for the LM2904 and LM2904Q), and  $V_{CC}$  is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, these devices can be operated directly off of the standard 5-V supply that is used in digital systems and will easily provide the required interface electronics without requiring additional  $\pm 5$ -V supplies.

The LM2904Q is manufactured to demanding automotive requirements.

The LM158 and LM158A are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . The LM258 and LM258A are characterized for operation from  $-25^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , the LM358 and LM358A from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ , and the LM2904 and LM2904Q from  $-40^{\circ}\text{C}$  to  $105^{\circ}\text{C}$ .

## symbol (each amplifier)



**LM158, LM258, LM358, LM158A  
LM258A, LM358A, LM358Y, LM2904, LM2904Q  
DUAL OPERATIONAL AMPLIFIERS**

SLOS068 – JUNE 1976 – REVISED JULY 1991

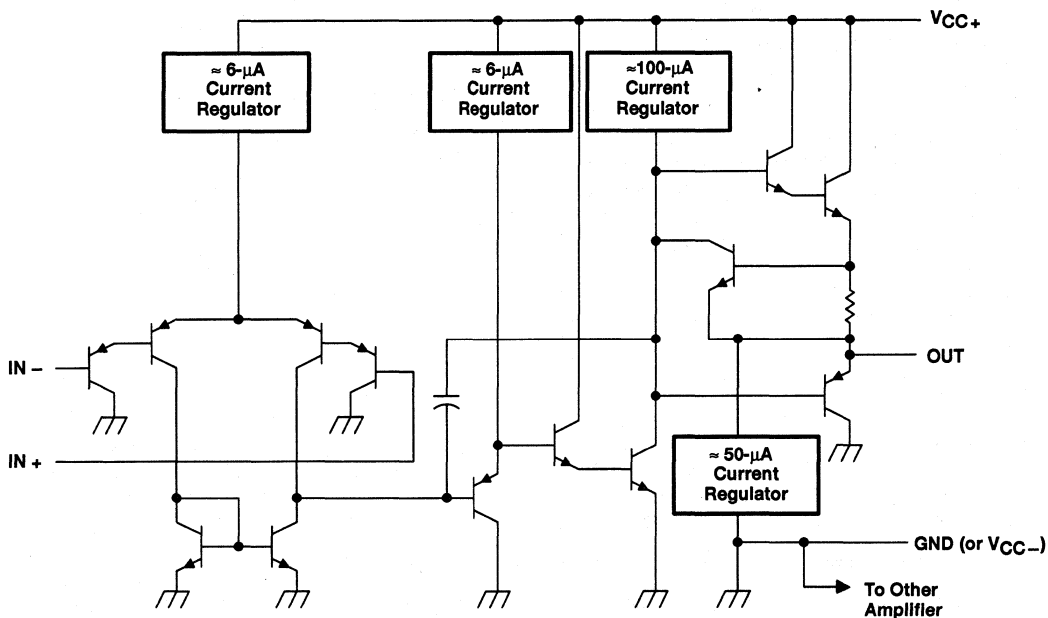
**AVAILABLE OPTIONS**

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES						CHIP FORM (Y)
		SMALL OUTLINE (D)†	SSOP (DB)‡	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	TSSOP (PW)‡	
0°C to 70°C	7 mV 3 mV	LM358D	LM358DB			LM358P LM358AP	LM358PW	LM358Y
-25°C to 85°C	5 mV 3 mV	LM258D				LM258P LM258AP		
-40°C to 105°C	7 mV	LM2904D LM2904QD	LM2904DB —			LM2904P LM2904QP	LM2904PW —	
-55°C to 125°C	5 mV 2 mV	LM158D		LM158FK LM158AFK	LM158JG LM158AJG	LM158P		

† The D package is available taped and reeled. Add the suffix R to the device type (e.g., LM358DR).

‡ The DB and PW packages are only available left-end taped and reeled. Add the suffix LE to the device type (e.g., LM358DBLE).

**schematic (each amplifier)**



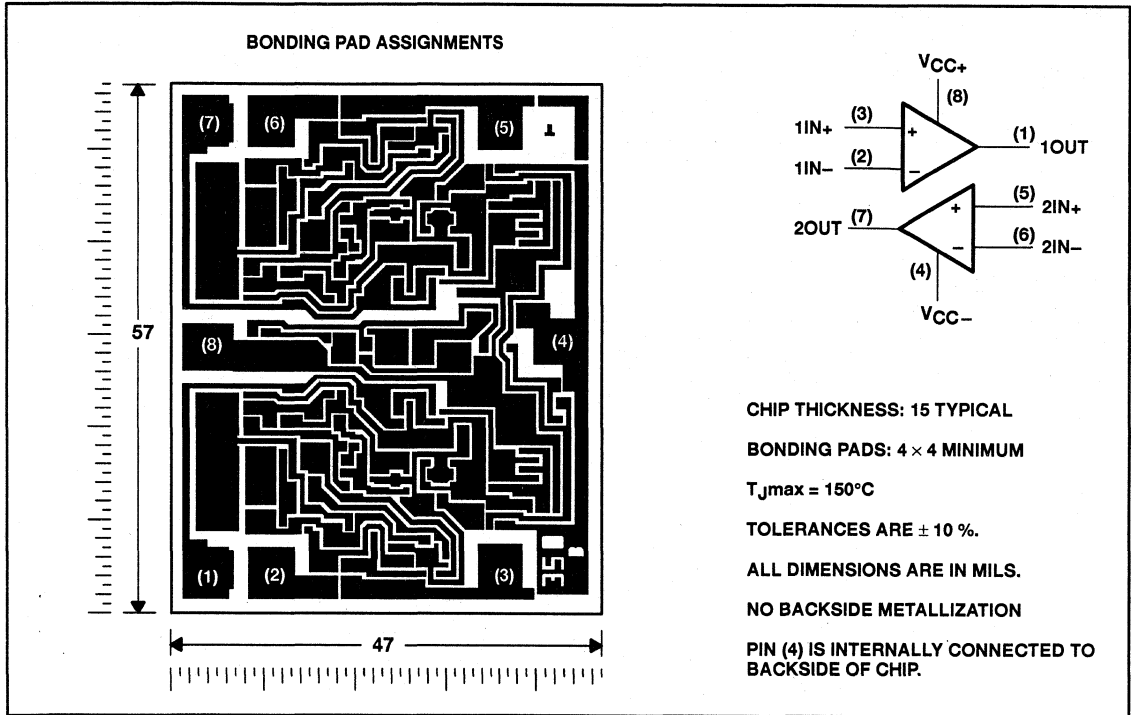
COMPONENT COUNT	
Epi-FET	1
Diodes	2
Resistors	7
Transistors	51
Capacitors	2

LM158, LM258, LM358, LM158A  
LM258A, LM358A, LM358Y, LM2904, LM2904Q  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS068 - JUNE 1976 - REVISED JULY 1991

**LM358Y chip information**

These chips, when properly assembled, display characteristics similar to the LM358. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**LM158, LM258, LM358, LM158A  
LM258A, LM358A, LM358Y, LM2904, LM2904Q  
DUAL OPERATIONAL AMPLIFIERS**

SLOS068 – JUNE 1976 – REVISED JULY 1991

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

		LM158, LM158A LM258, LM258A LM358, LM358A	LM2904, LM2904Q	UNIT	
Supply voltage $V_{CC}$ (see Note 1)		32	26	V	
Differential input voltage (see Note 2)		$\pm 32$	$\pm 26$	V	
Input voltage (either input)		0.3 to 32	0.3 to 26	V	
Duration of output short circuit (one amplifier) to ground at (or below) 25°C free-air temperature ( $V_{CC} \leq 15$ V) (see Note 3)		unlimited	unlimited		
Continuous total dissipation		See Dissipation Rating Table			
Operating free-air temperature range	LM158, LM158A	-55 to 125		°C	
	LM258, LM258A	-25 to 85			
	LM358, LM358A	0 to 70			
	LM2904, LM2904Q		-40 to 105		
Storage temperature range		-65 to 150	-65 to 150	°C	
Case temperature for 60 seconds		FK package	260	°C	
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds		JG package	300	300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds		D, DB, P, or PW package	260	260	°C

- NOTES: 1. All voltage values, except differential voltages and  $V_{CC}$  specified for measurement of  $I_{OS}$ , are with respect to the network ground terminal.  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
DB	525 mW	4.2 mW/°C	336 mW	273 mW	–
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW
PW	525 mW	4.2 mW/°C	336 mW	273 mW	–



**LM158, LM258, LM358, LM2904, LM2904Q**  
**LM258A, LM358A, LM358Y, LM2904, LM2904Q**  
**DUAL OPERATIONAL AMPLIFIERS**  
SLOS068 - JUNE 1976 - REVISED JULY 1991

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITION†	LM158, LM258		LM358		LM2904, LM2904Q		UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
$V_{IO}$	$V_{CC} = 5\text{ V to MAX}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$	3	5	3	7	3	7	mV
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			7		7		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	$V_O = 1.4\text{ V}$	2	30	2	50	2	50	nA
$\alpha_{IIO}$	Average temperature coefficient of input offset current			100		150		nA
$I_{IB}$	$V_O = 1.4\text{ V}$			10		10		$\text{pA}/^\circ\text{C}$
$V_{ICR}$	Common-mode input voltage range			-20	-150	-20	-250	nA
$V_{OH}$	High-level output voltage			-300		-500		nA
$V_{OL}$	Low-level output voltage			0 to $V_{CC}-1$		0 to $V_{CC}-1$		V
$A_{VD}$	Large-signal differential voltage amplification			0 to $V_{CC}-2$		0 to $V_{CC}-2$		V
$CMRR$	Common-mode rejection ratio			$V_{CC}-1.5$		$V_{CC}-1.5$		dB
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )			26		26		dB
$V_{O1}/V_{O2}$	Crosstalk attenuation			27		23		dB
				28		24		dB
				5		5		dB
				20		20		dB
				50		100		dB
				25		15		dB
				70		80		dB
				65		100		dB
				120		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 26 V for LM2904 and 30 V for others. Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$  for LM158,  $-25^\circ\text{C}$  to  $85^\circ\text{C}$  for LM258,  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for LM358, and  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  for LM2904 and LM2904Q.  
‡ All typical values are at  $T_A = 25^\circ\text{C}$ .



**LM158, LM258, LM358, LM2904, LM2904Q**  
**LM258A, LM358A, LM358Y, LM2904, LM2904Q**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS068 - JUNE 1976 - REVISED JULY 1991

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted) (continued)

PARAMETER	TEST CONDITIONS†	LM158, LM258		LM358		LM2904, LM2904Q		UNIT	
		MIN	TYP‡	MAX	MIN	TYP‡	MAX		MIN
IO Output current	VCC = 15 V, VID = 1 V, VO = 0 Full range	25°C	-20	-30	-20	-30	-20	-30	mA
		25°C	-10	-10	-10	-10	-10	-10	
		25°C	10	20	10	20	10	20	
LOS Short-circuit output current	VID = -1 V, VO = 200 mV VCC at 5 V, GND at -5 V, VO = 0	25°C	5	5	5	5	5	5	mA
		25°C	12	30	12	30	12	30	
		25°C	±40	±60	±40	±60	±40	±60	
ICC Supply current (two amplifiers)	VO = 2.5 V, No load VCC = MAX, VO = 0.5 V, No load	25°C	0.7	1.2	0.7	1.2	0.7	1.2	mA
		25°C	1	2	1	2	1	2	
		25°C	0.7	1.2	0.7	1.2	0.7	1.2	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX VCC for testing purposes is 26 V for LM 2904 and 30 V for others. Full range is -55°C to 125°C for LM158, -25°C to 85°C for LM258, 0°C to 70°C for LM358, and -40°C to 85°C for LM2904 and LM2904Q.  
‡ All typical values are at TA = 25°C.



LM158, LM258, LM358, LM2904, LM2904Q  
 LM258A, LM358A, LM358Y, LM2904, LM2904Q  
 DUAL OPERATIONAL AMPLIFIERS

SLOS068 - JUNE 1976 - REVISED JULY 1991

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	LM158A		LM258A		LM358A		UNIT	
		MIN	TYP‡	MAX	MIN	TYP‡	MAX		
$V_{IO}$	$V_{CC} = 5\text{ V}$ to $30\text{ V}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$			2	2	3	2	3	mV
	Full range			4		4		5	
$\alpha_{VIO}$	Full range			7	7	15	7	20	$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	$V_O = 1.4\text{ V}$			2	2	10	2	30	nA
	Full range			30		30		75	
$\alpha_{IIO}$	Full range			10	10	200	10	300	$\text{pA}/^\circ\text{C}$
$I_{IB}$	$V_O = 1.4\text{ V}$			-15	-15	-80	-15	-100	nA
	Full range			-100		-100		-200	
$V_{ICR}$	$V_{CC} = 30\text{ V}$	0 to $V_{CC-1}$		0 to $V_{CC-1.5}$	0 to $V_{CC-1.5}$		0 to $V_{CC-1.5}$		V
	Full range	0 to $V_{CC-2}$		0 to $V_{CC-2}$	0 to $V_{CC-2}$		0 to $V_{CC-2}$		
$V_{OH}$	$R_L \geq 2\text{ k}\Omega$	$V_{CC-1.5}$		$V_{CC-1.5}$	$V_{CC-1.5}$		$V_{CC-1.5}$		V
	Full range	26		26	26		26		
$V_{OL}$	$V_{CC} = 30\text{ V}$ , $R_L = 2\text{ k}\Omega$	27		28	27	28	27	28	
	Full range	5		20	5	20	5	100	mV
$A_{VD}$	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V}$ to $11\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	50		100	50	100	25	100	V/mV
	Full range	25		25	25		15		
CMRR	Common-mode rejection ratio	70		80	70	80	65	80	dB
KSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	65		100	65	100	65	100	dB
$V_{O1}/V_{O2}$	Crosstalk attenuation $f = 1\text{ kHz}$ to $20\text{ kHz}$	120		120	120		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$  for LM158A,  $-25^\circ\text{C}$  to  $85^\circ\text{C}$  for LM258A, and  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for LM358A.

‡ All typical values are at  $T_A = 25^\circ\text{C}$ .

**LM158, LM258, LM358, LM2904, LM2904Q  
LM258A, LM358A, LM358Y, LM2904, LM2904Q  
DUAL OPERATIONAL AMPLIFIERS**

SLOS068 - JUNE 1976 - REVISED JULY 1991

electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted) (continued)

PARAMETER	TEST CONDITIONS†	LM158A		LM258A		LM358A		UNIT
		MIN	TYP‡	MAX	MIN	TYP‡	MAX	
$I_O$ Output current	$V_{CC} = 15\text{ V}$ , $V_{ID} = 1\text{ V}$ , $V_O = 0$	-20	-30	-60	-20	-30	-60	mA
	Full range	-10			-10			
	$V_{CC} = 15\text{ V}$ , $V_{ID} = -1\text{ V}$ , $V_O = 15\text{ V}$	10	20		10	20		
$I_{OS}$ Short-circuit output current	$V_{ID} = -1\text{ V}$ , $V_O = 200\text{ mV}$	12	30		12	30		$\mu\text{A}$
	$V_{CC}$ at $5\text{ V}$ , $V_O = 0$		$\pm 40$	$\pm 60$		$\pm 40$	$\pm 60$	
$I_{CC}$ Supply current (two amplifiers)	$V_{CC} = 2.5\text{ V}$ , No load		0.7	1.2		0.7	1.2	mA
	$V_{CC} = 30\text{ V}$ , $V_O = 15\text{ V}$ , No load		1	2		1	2	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$  for LM158A,  $-25^\circ\text{C}$  to  $85^\circ\text{C}$  for LM258A, and  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for LM358A.

‡ All typical values are at  $T_A = 25^\circ\text{C}$ .





**LM158, LM258, LM358, LM158A  
LM258A, LM358A, LM358Y, LM2904, LM2904Q  
DUAL OPERATIONAL AMPLIFIERS**

SLOS088 – JUNE 1976 – REVISED JULY 1991

**electrical characteristics  $V_{CC} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONST	LM358Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{CC} = 5\text{ V to MAX}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$		3	7	mV
$I_{IO}$ Input offset current			2	50	nA
$I_{IB}$ Input bias current			-20	-250	nA
$V_{ICR}$ Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	0 to $V_{CC}-1.5$			V
$V_{OH+}$ High-level output voltage	$R_L \geq 10\text{ k}\Omega$	$V_{CC}-1.5$			V
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V to }11\text{ V}$ , $R_L = \geq 2\text{ k}\Omega$	15	100		V/mV
$CMRR$ Common-mode rejection ratio	$V_{IC} = V_{ICR min}$	65	80		dB
$K_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )		65	100		dB
$I_O$ Output current	$V_{CC} = 15\text{ V}$ , $V_{ID} = 1\text{ V}$ , $V_O = 0$	-20	-30	-60	mA
	$V_{CC} = 15\text{ V}$ , $V_{ID} = -1\text{ V}$ , $V_O = 15\text{ V}$	10	20		
	$V_{ID} = 1\text{ V}$ , $V_O = 200\text{ mV}$	12	30		
$I_{OS}$ Short-circuit output current	$V_{CC}$ at 5 V, GND at -5 V, $V_O = 0$		$\pm 40$	$\pm 60$	mA
$I_{CC}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load		0.7	1.2	mA
	$V_{CC} = MAX$ , $V_O = 0.5\text{ V}$ , No load		1	2	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 30 V.

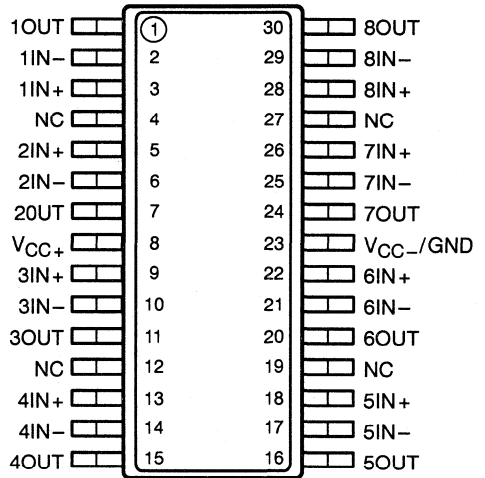


# LM324x2 OCTAL OPERATIONAL AMPLIFIER

SLOS133 – APRIL 1994

- **Wide Range of Supply Voltages:**  
Single Supply . . . 3 V to 30 V  
or Dual Supplies
- **Low Supply-Current Drain Independent of Supply Voltage . . . 1.4 mA Typ**
- **Common-Mode Input Voltage Range**  
Includes Ground Allowing Direct Sensing  
Near Ground
- **Low Input Bias and Offset Parameters:**  
Input Offset Voltage . . . 3 mV Typ  
Input Offset Current . . . 2 nA Typ  
Input Bias Current . . . -20 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage . . . 32 V**
- **Open-Loop Differential Voltage Amplification . . . 100 V/mV Typ**
- **Internal Frequency Compensation**

**DB PACKAGE  
(TOP VIEW)**



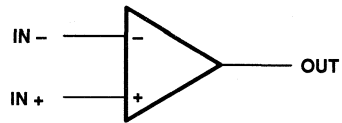
NC – No internal connection

## description

The LM324x2 device consists of eight independent, high-gain frequency-compensated operational amplifiers that are designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible when the difference between the two supplies is 3 V to 30 V and  $V_{CC}$  is at least 1.5 V more positive than the input common-mode voltage. The low supply-current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational-amplifier circuits that now can be more easily implemented in single-supply-voltage systems.

## symbol (each amplifier)



## AVAILABLE OPTION

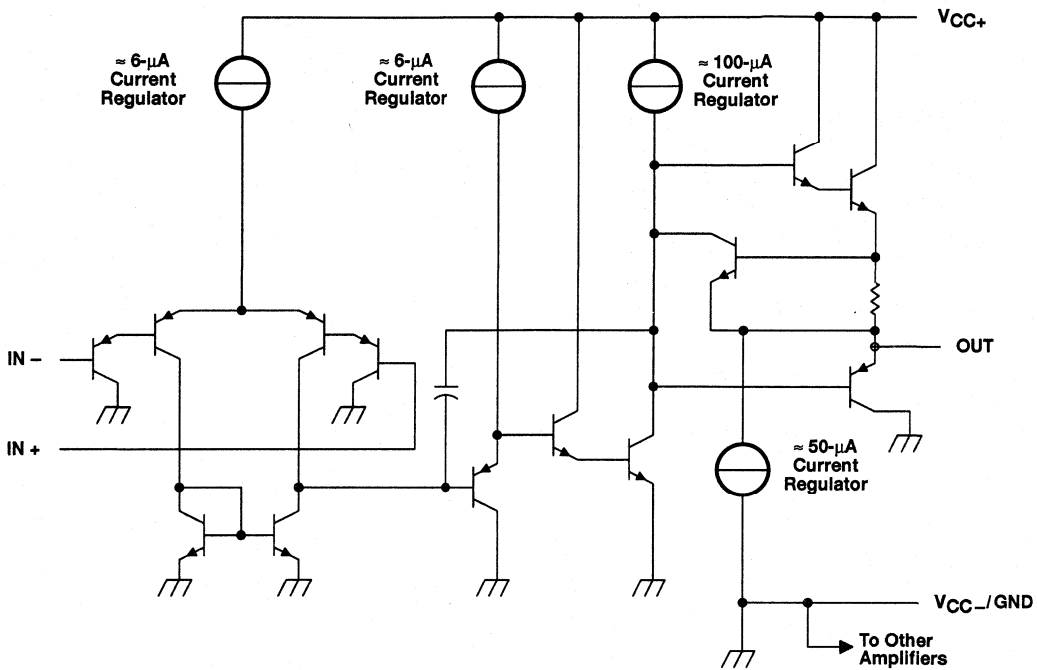
$T_A$	$V_{IOmax}$ AT 25°C	PACKAGE
		SMALL OUTLINE (DB)†
0°C to 70°C	7 mV	LM324x2DBLE

† The DB package is only available left-end taped and reeled.

# LM324x2 OCTAL OPERATIONAL AMPLIFIER

SLOS133 – APRIL 1994

## schematic (each amplifier)



COMPONENT COUNT (total device)	
Epi-FET	2
Transistors	190
Diodes	8
Resistors	22
Capacitors	8

# LM324x2 OCTAL OPERATIONAL AMPLIFIER

SLOS133 – APRIL 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{CC}$ (see Note 1)	32 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 32$ V
Input voltage range, $V_I$ (any input)	-0.3 V to 32 V
Duration of output short circuit to ground (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these conditions beyond those indicated is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{OS}$ , are with respect to GND.  
 2. Differential voltages are at IN + with respect to IN -.  
 3. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DB	1024 mW	8.2 mW/°C	655 mW



# LM324x2 OCTAL OPERATIONAL AMPLIFIER

SLOS133 – APRIL 1994

## electrical characteristics at specified free-air temperature, $V_{CC} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	$T_A$ ‡	MIN	TYP§	MAX	UNIT	
$V_{IO}$ Input offset voltage	$V_{CC} = 5\text{ V to MAX}$ , $V_{IC} = V_{ICRmin}$ , $V_O = 1.4\text{ V}$	25°C		3	7	mV	
		Full range			9		
$I_{IO}$ Input offset current	$V_O = 1.4\text{ V}$	25°C		2	50	nA	
		Full range			150		
$I_{IB}$ Input bias current	$V_O = 1.4\text{ V}$	25°C		-20	-250	nA	
		Full range			-500		
$V_{ICR}$ Common-mode input voltage range	$V_{CC} = 5\text{ V to MAX}$	25°C		0 to $V_{CC}-1.5$		V	
		Full range		0 to $V_{CC}-2$			
$V_{OH}$ High-level output voltage	$R_L = 2\text{ k}\Omega$	25°C		$V_{CC}-1.5$		V	
	$V_{CC} = \text{MAX}$ , $R_L = 2\text{ k}\Omega$	Full range		26			
	$V_{CC} = \text{MAX}$ , $R_L \geq 10\text{ k}\Omega$	Full range		27	28		
$V_{OL}$ Low-level output voltage	$R_L \leq 10\text{ k}\Omega$	Full range		5	20	mV	
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC} = 15\text{ V}$ , $V_O = 1\text{ V to }11\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	25°C		25	100	V/mV	
		Full range		15			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C		65	80	dB	
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )		25°C		65	100	dB	
$V_{O1}/V_{O2}$ Crosstalk attenuation	$f = 1\text{ kHz to }20\text{ kHz}$	25°C			120	dB	
$I_O$ Output current	$V_{CC} = 15\text{ V}$ , $V_{ID} = 1\text{ V}$ , $V_O = 0$	25°C		-20	-30	-60	mA
		Full range		-10			
	$V_{CC} = 15\text{ V}$ , $V_{ID} = -1\text{ V}$ , $V_O = 15\text{ V}$	25°C		10	20		
		Full range		5			
$I_{OS}$ Short-circuit output current	$V_O = 0$ , GND = -5 V	25°C		$\pm 40$	$\pm 60$	mA	
		Full range		1.4	2.4		
$I_{CC}$ Supply current (eight amplifiers)	$V_O = 2.5\text{ V}$ , No load	Full range				mA	
		$V_{CC} = \text{MAX}$ , No load	Full range		2.2		6

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. MAX  $V_{CC}$  for testing purposes is 30 V.

‡ Full range is 0°C to 70°C.

§ All typical values are at  $T_A = 25^\circ\text{C}$ .



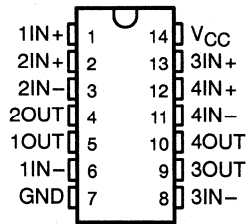
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS059 – JULY 1979 – REVISED SEPTEMBER 1990

- Wide Range of Supply Voltages, Single or Dual Supplies
- Wide Bandwidth
- Large Output Voltage Swing
- Output Short-Circuit Protection
- Internal Frequency Compensation
- Low Input Bias Current
- Designed to Be Interchangeable With National Semiconductor LM2900 and LM3900, Respectively

N PACKAGE  
(TOP VIEW)

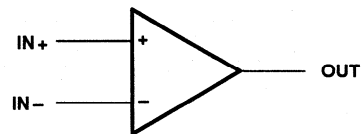


## description

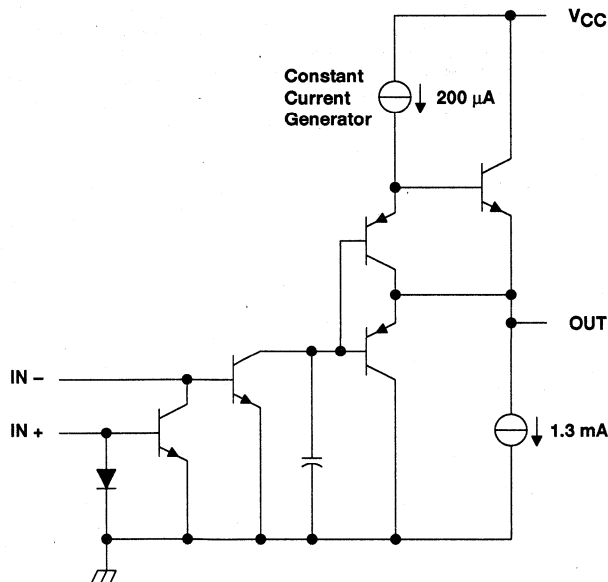
These devices consist of four independent, high-gain frequency-compensated Norton operational amplifiers that were designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible. The low supply current drain is essentially independent of the magnitude of the supply voltage. These devices provide wide bandwidth and large output voltage swing.

The LM2900 is characterized for operation from -40°C to 85°C, and the LM3900 is characterized for operation from 0°C to 70°C.

## symbol (each amplifier)



## schematic (each amplifier)



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1990, Texas Instruments Incorporated

# LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS059 – JULY 1979 – REVISED SEPTEMBER 1990

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	LM2900	LM3900	UNIT
Supply voltage, $V_{CC}$ (see Note 1)	36	36	V
Input current	20	20	mA
Duration of output short circuit (one amplifier) to ground at (or below) 25°C free-air temperature (see Note 2)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-40 to 85	0 to 70	°C
Storage temperature range	-65 to 150	-65 to 150	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260	260	°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the network ground terminal.  
2. Short circuits from outputs to  $V_{CC}$  can cause excessive heating and eventual destruction.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
N	1150 mW	9.2 mW/°C	736 mW	598 mW

## recommended operating conditions

	LM2900		LM3900		UNIT
	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC}$ (single supply)	4.5	32	4.5	32	V
Supply voltage, $V_{CC+}$ (dual supply)	2.2	16	2.2	16	V
Supply voltage, $V_{CC-}$ (dual supply)	-2.2	-16	-2.2	-16	V
Input current (see Note 3)		-1		-1	mA
Operating free-air temperature, $T_A$	-40	85	0	70	°C

NOTE 3: Clamp transistors are included that prevent the input voltages from swinging below ground more than approximately -0.3 V. The negative input currents that may result from large signal overdrive with capacitive input coupling must be limited externally to values of approximately -1 mA. Negative input currents in excess of -4 mA causes the output voltage to drop to a low voltage. These values apply for any one of the input terminals. If more than one of the input terminals are simultaneously driven negative, maximum currents are reduced. Common-mode current biasing can be used to prevent negative input voltages.



# LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS059 – JULY 1979 – REVISED SEPTEMBER 1990

## electrical characteristics, $V_{CC} = 15\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONST		LM2900			LM3900			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$I_{IB}$ Input bias current (inverting input)	$I_{I+} = 0$	$T_A = 25^\circ\text{C}$	30	200		30	200	nA	
		$T_A = \text{Full range}$	300		300				
Mirror gain	$I_{I+} = 20\ \mu\text{A}$ to $200\ \mu\text{A}$ $T_A = \text{Full range}$ , See Note 4		0.9		1.1	0.9		1.1	$\mu\text{A}/\mu\text{A}$
Change in mirror gain	See Note 4		2%		5%	2%		5%	
Mirror current	$V_{I+} = V_{I-}$ , $T_A = \text{Full range}$ , See Note 4		10		500	10		500	$\mu\text{A}$
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 10\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $f = 100\text{ Hz}$		1.2		2.8	1.2		2.8	V/mV
$r_i$ Input resistance (inverting input)			1			1			$\text{M}\Omega$
$r_o$ Output resistance			8			8			k $\Omega$
$B_1$ Unity-gain bandwidth (inverting input)			2.5			2.5			MHz
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )			70			70			dB
$V_{OH}$ High-level output voltage	$I_{I+} = 0$ , $I_{I-} = 0$	$R_L = 2\text{ k}\Omega$	13.5			13.5			V
		$V_{CC} = 30\text{ V}$ , No load	29.5			29.5			
$V_{OL}$ Low-level output voltage	$I_{I+} = 0$ , $R_L = 2\text{ k}\Omega$	$I_{I-} = 10\ \mu\text{A}$ ,	0.09		0.2	0.09		0.2	V
$I_{OS}$ Short-circuit output current (output internally high)	$I_{I+} = 0$ , $V_O = 0$	$I_{I-} = 0$ ,	-6		-18	-6		-10	mA
			0.5		1.3	0.5		1.3	
$I_{OL}$ Low-level output current‡	$I_{I-} = 5\ \mu\text{A}$	$V_{OL} = 1\text{ V}$	5			5			mA
$I_{CC}$ Supply current (four amplifiers)	No load		6.2		10	6.2		10	mA

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for  $T_A$  is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  for LM2900 and  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for LM3900.

‡ The output current-sink capability can be increased for large-signal conditions by overdriving the inverting input.

NOTE 4: These parameters are measured with the output balanced midway between  $V_{CC}$  and GND.

## operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
SR	Low-to-high output	$V_O = 10\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 2\text{ k}\Omega$				0.5		V/ $\mu\text{s}$
	High-to-low output					20		

# LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS059 – JULY 1979 – REVISED SEPTEMBER 1990

## TYPICAL CHARACTERISTICS†

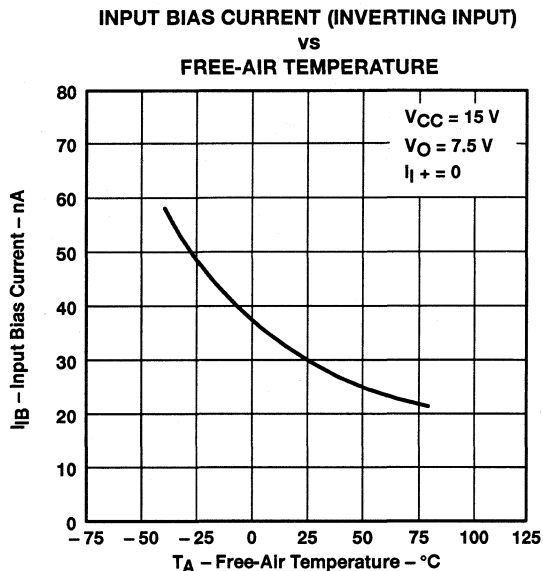


Figure 1

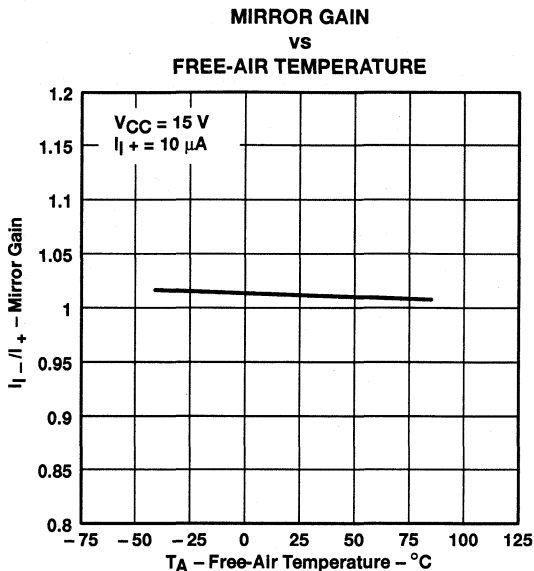


Figure 2

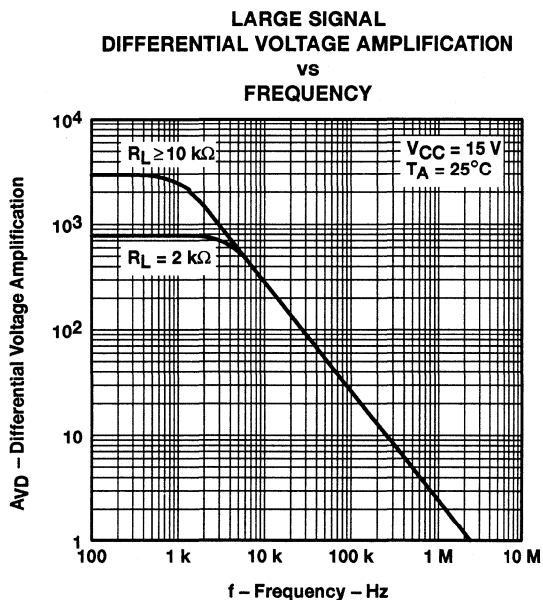


Figure 3

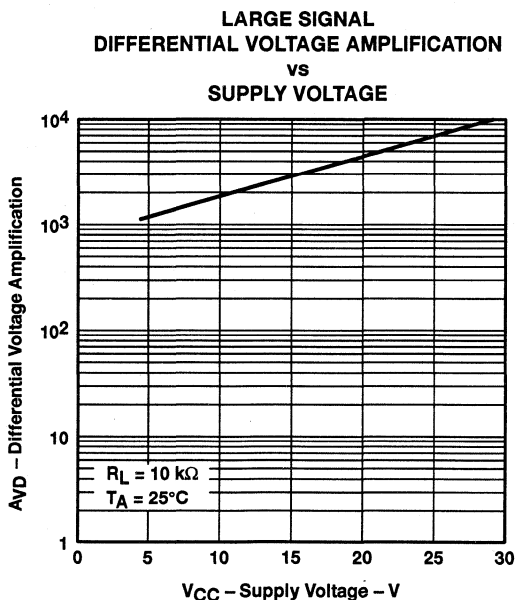
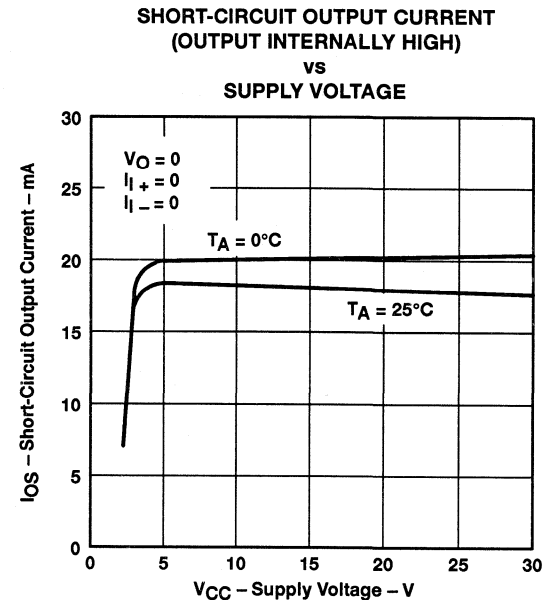
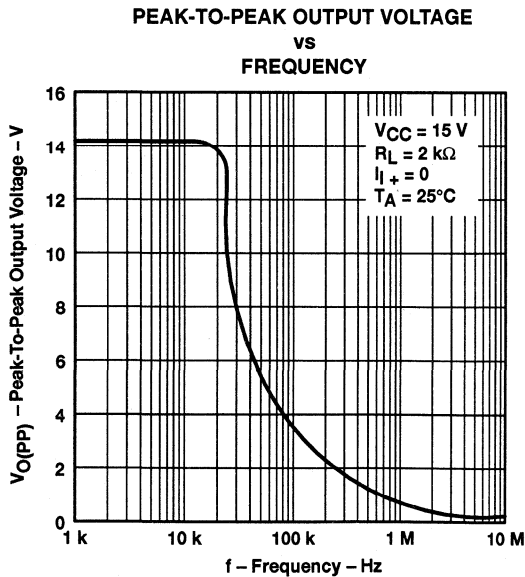
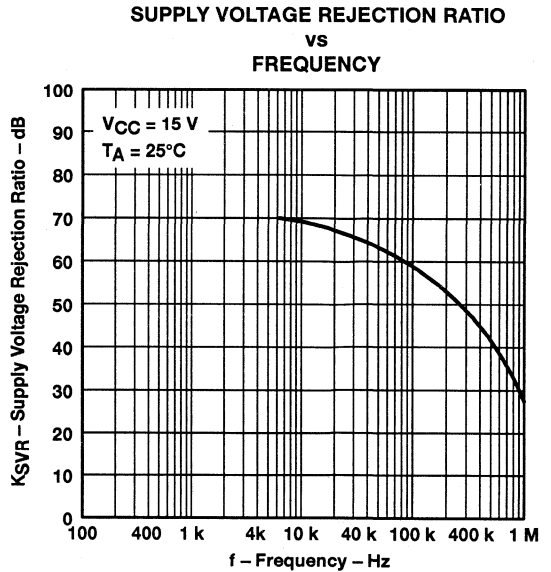
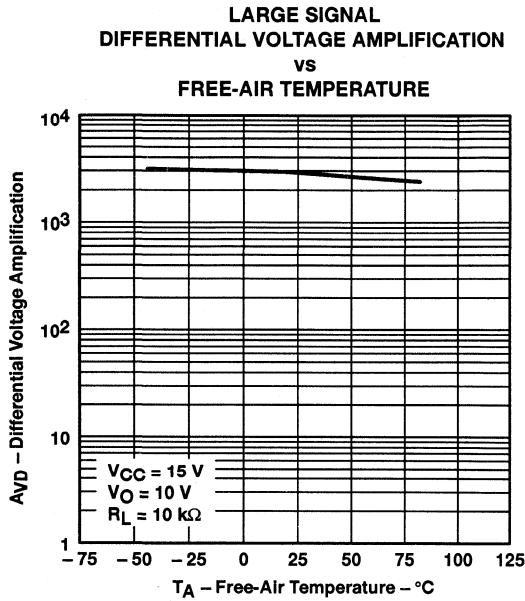


Figure 4

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# LM2900, LM3900 QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS059 – JULY 1979 – REVISED SEPTEMBER 1990

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE

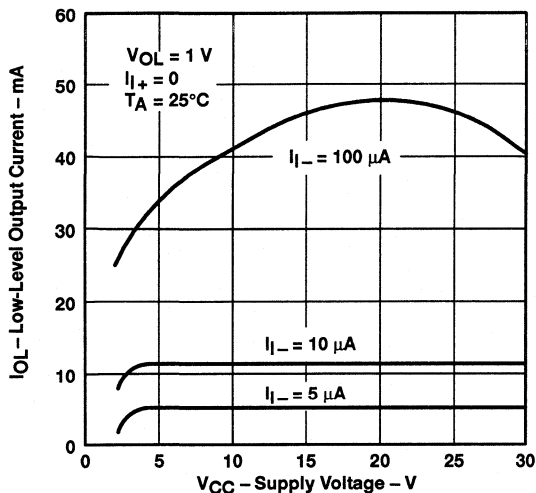


Figure 9

PULLDOWN CURRENT  
vs  
SUPPLY VOLTAGE

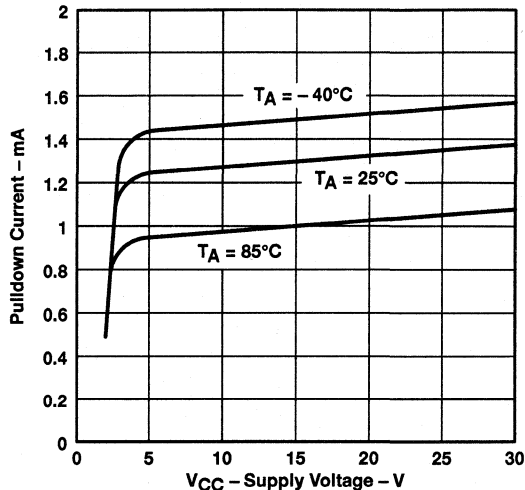


Figure 10

PULLDOWN CURRENT  
vs  
FREE-AIR TEMPERATURE

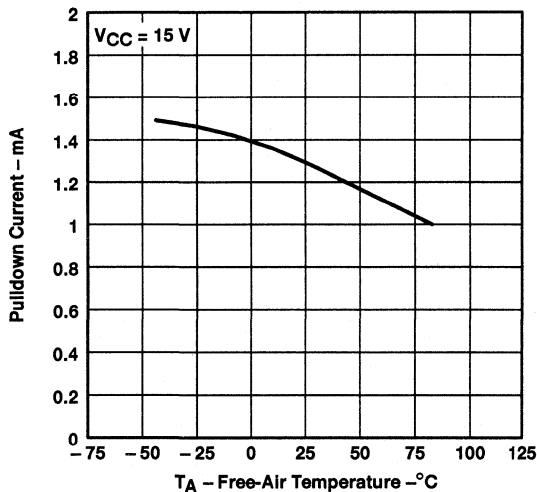


Figure 11

TOTAL SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

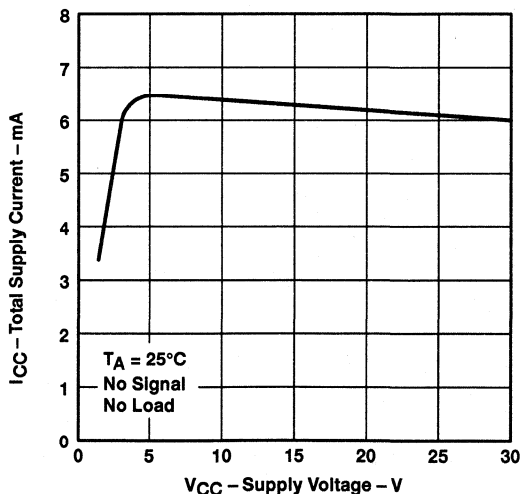


Figure 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

APPLICATION INFORMATION

Norton (or current-differencing) amplifiers can be used in most standard general-purpose operational amplifier applications. Performance as a dc amplifier in a single-power-supply mode is not as precise as a standard integrated-circuit operational amplifier operating from dual supplies. Operation of the amplifier can best be understood by noting that input currents are differenced at the inverting input terminal and this current then flows through the external feedback resistor to produce the output voltage. Common-mode current biasing is generally useful to allow operating with signal levels near (or even below) ground.

Internal transistors clamp negative input voltages at approximately  $-0.3\text{ V}$  but the magnitude of current flow has to be limited by the external input network. For operation at high temperature, this limit should be approximately  $-100\text{ }\mu\text{A}$ .

Noise immunity of a Norton amplifier is less than that of standard bipolar amplifiers. Circuit layout is more critical since coupling from the output to the noninverting input can cause oscillations. Care must also be exercised when driving either input from a low-impedance source. A limiting resistor should be placed in series with the input lead to limit the peak input current. Current up to  $20\text{ mA}$  will not damage the device, but the current mirror on the noninverting input will saturate and cause a loss of mirror gain at higher current levels, especially at high operating temperatures.

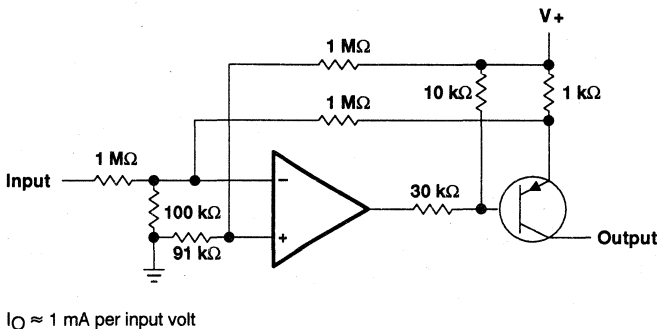


Figure 13. Voltage-Controlled Current Source

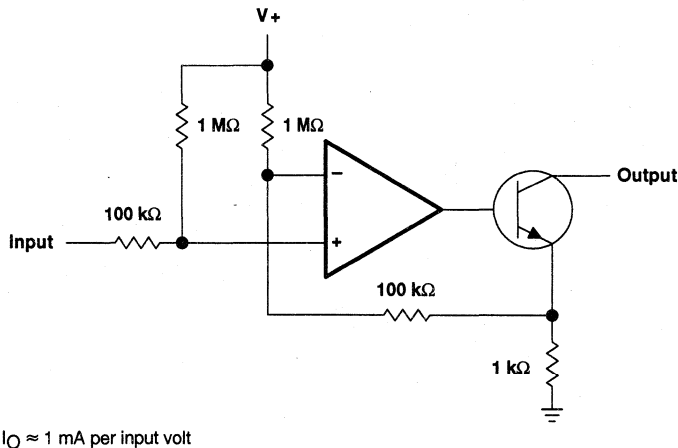


Figure 14. Voltage-Controlled Current Sink



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

- **Single-Supply Operation:**  
Input Voltage Range Extends to Ground  
Output Swings to Ground While Sinking Current
- **Input Offset Voltage**  
150  $\mu\text{V}$  Max at 25°C for LT1013A
- **Offset Voltage Temperature Coefficient**  
2.5  $\mu\text{V}/^\circ\text{C}$  Max for LT1013A
- **Input Offset Current**  
0.8 nA Max at 25°C for LT1013A
- **High Gain . . .** 1.5 V/ $\mu\text{V}$  Min ( $R_L = 2 \text{ k}\Omega$ ),  
0.8 V/ $\mu\text{V}$  Min ( $R_L = 600 \text{ k}\Omega$ ) for LT1013A
- **Low Supply Current . . .** 0.5 mA Max at  
 $T_A = 25^\circ\text{C}$  for LT1013A
- **Low Peak-to-Peak Noise Voltage**  
0.55  $\mu\text{V}$  Typ
- **Low Current Noise . . .** 0.07  $\text{pA}/\sqrt{\text{Hz}}$  Typ

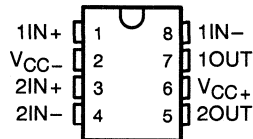
## description

The LT1013 is a dual precision operational amplifier featuring low offset voltage temperature coefficient, high gain, low supply current, and low noise.

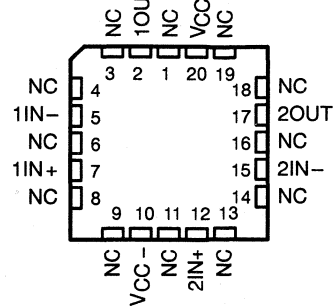
The LT1013 can be operated from a single 5-V power supply; the common-mode input voltage range includes ground, and the output can also swing to within a few millivolts of ground. Crossover distortion is eliminated. The LT1013 can be operated with both dual  $\pm 15\text{-V}$  and single 5-V supplies.

The LT1013C and LT1013AC, and LT1013D are characterized for operation from 0°C to 70°C. The LT1013I and LT1013AI, and LT1013DI are characterized for operation from -40°C to 105°C. The LT1013M and LT1013AM, and LT1013DM are characterized for operation over the full military temperature range of -55°C to 125°C.

**D PACKAGE  
(TOP VIEW)**

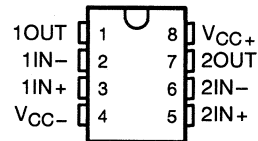


**FK PACKAGE  
(TOP VIEW)**



NC – No internal connection

**JG OR P PACKAGE  
(TOP VIEW)**



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES				CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	
0°C to 70°C	150 $\mu\text{V}$	—	—	—	LT1013ACP	LT1013Y
	300 $\mu\text{V}$	—	—	—	LT1013CP	
	800 $\mu\text{V}$	LT1013DD	—	—	LT1013DP	
-40°C to 105°C	150 $\mu\text{V}$	—	—	—	LT1013AIP	—
	300 $\mu\text{V}$	—	—	—	LT1013IP	
	800 $\mu\text{V}$	LT1013DID	—	—	LT1013DIP	
-55°C to 125°C	150 $\mu\text{V}$	—	LT11013AMFK	—	LT1013AMP	—
	300 $\mu\text{V}$	—	LT11013MFK	LT11013MJG	LT1013MP	
	800 $\mu\text{V}$	LT1013DMD	—	LT11013DMJG	LT1013DMP	

The D package is available taped and reeled. Add the suffix R to the device type (e.g., LT1013DDR).

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

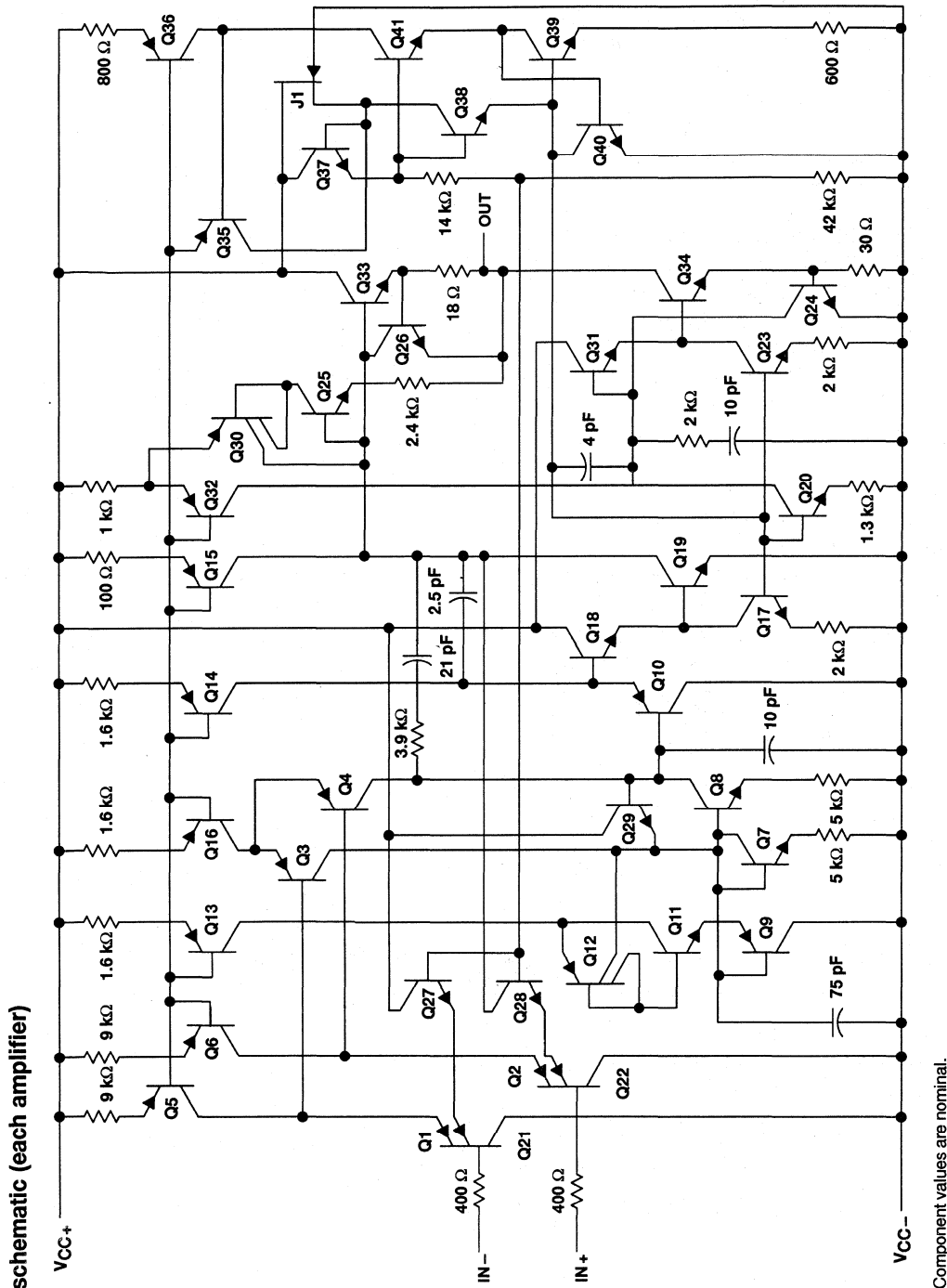
**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1991, Texas Instruments Incorporated  
On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991



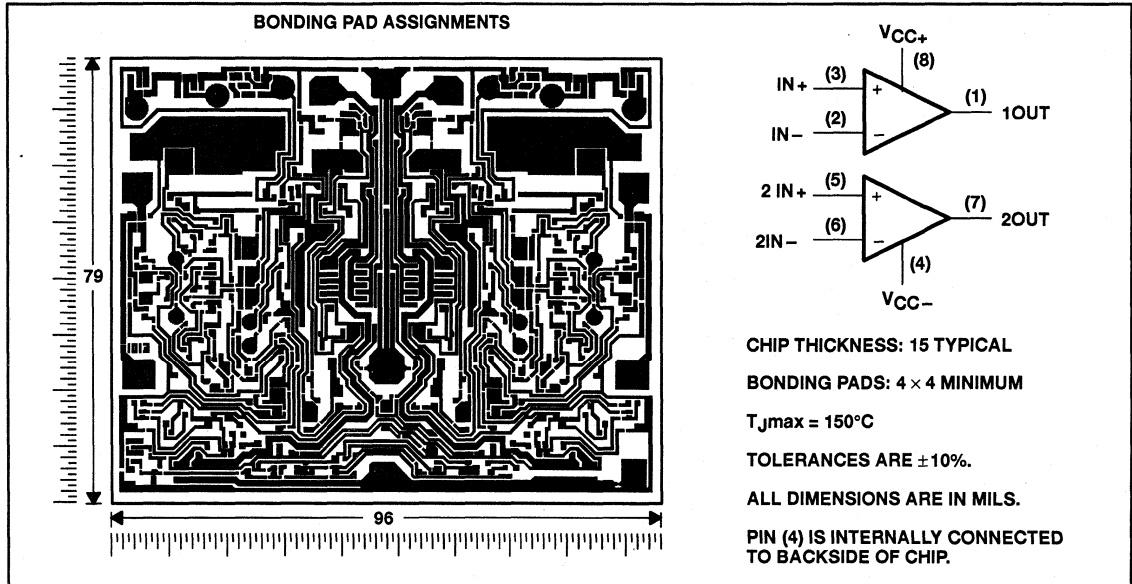


# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## LT1013Y chip information

This chip, when properly assembled, displays characteristics similar to the LT1013. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1) .....	22 V
Supply voltage, $V_{CC-}$ (see Note 1) .....	-22 V
Differential input voltage (see Note 2) .....	$\pm 30$ V
Input voltage range, $V_I$ (any input, see Note 1) .....	$V_{CC-} - 5$ V to $V_{CC+}$
Duration of short-circuit current at (or below) $25^{\circ}\text{C}$ (see Note 3) .....	unlimited
Operating free-air temperature range, $T_A$ : LT1013C, LT1013AC, LT1013D .....	$-0^{\circ}\text{C}$ to $70^{\circ}\text{C}$
LT1013I, LT1013AI, LT1013DI .....	$-40^{\circ}\text{C}$ to $105^{\circ}\text{C}$
LT1013M, LT1013AM, LT1013DM .....	$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$
Storage temperature range .....	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds: D or P package .....	$260^{\circ}\text{C}$
Case temperature for 60 seconds: FK package .....	$260^{\circ}\text{C}$
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds: JG package .....	$300^{\circ}\text{C}$

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at IN+ with respect to IN-.  
 3. The output may be shorted to either supply.

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

electrical characteristics at specified free-air temperature,  $V_{CC} \pm \pm 15\text{ V}$ ,  $V_{IC} = 0$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	LT1013C		LT1013AC		LT1013DC		UNIT
			MIN	TYP‡	MAX	MIN	TYP‡	MAX	
V <sub>IO</sub>	RS = 50 Ω	25°C	60	300	40	150	200	800	μV
		Full range		400		240		1000	
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	Full range	0.4	2.5	0.3	2	0.7	5	μV/°C
	Long-term drift of input offset voltage	25°C	0.5		0.4		0.5		μV/mo
I <sub>IO</sub>	Input offset current	25°C	0.2	1.5	0.15	0.8	0.2	1.5	nA
		Full range		2.8		1.5		2.8	
I <sub>IB</sub>	Input bias current	25°C	-15	-30	-12	-20	-15	-30	nA
		Full range		-38		-25		-38	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-15	-15.3	-15	-15.3	-15	-15.3	V
			to	to	to	to	to	to	
			13.5	13.8	13.5	13.8	13.5	13.8	
		Full range	-15		-15		-15		V
			to	to	to	to	to	to	
			13		13		13		
V <sub>OM</sub>	Maximum peak output voltage swing	25°C	±12.5	±14	±13	±14	±12.5	±14	V
		Full range							
		25°C	0.5	0.2	0.8	2.5	0.5	2	V/μV
		25°C	1.2	7	1.5	8	1.2	7	
		Full range	0.7		1		0.7		
CMRR	Common-mode rejection ratio	25°C	97	114	100	117	97	114	dB
		Full range							
		25°C	94		98		94		dB
		25°C	100	117	103	120	100	117	
		Full range	97		101		97		
KSVR	Supply-voltage rejection ratio (ΔV <sub>CC</sub> /AV <sub>IO</sub> )	25°C	120	137	123	140	120	137	dB
		25°C	70	300	100	400	70	300	
r <sub>id</sub>	Channel separation	25°C	4		5		4		MΩ
		25°C							
r <sub>ic</sub>	Differential input resistance	25°C							GΩ
		25°C							
I <sub>CC</sub>	Supply current per amplifier	25°C	0.35	0.55	0.35	0.5	0.35	0.55	mA
		Full range		0.7		0.55		0.6	

† Full range is 0°C to 70°C.

‡ All typical values are at T<sub>A</sub> = 25°C.



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

electrical characteristics at specified free-air temperature,  $V_{CC+} = 5V$ ,  $V_{CC-} = 0$ ,  $V_O = 1.4V$ ,  $V_{IC} = 0$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1013C		LT1013AC		LT1013DC		UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	25°C	90	450	60	250	250	950	$\mu V$
	Full range		570		350		1200	
$I_{IO}$ Input offset current	25°C	0.3	2	0.2	1.3	0.3	2	nA
	Full range		6		3.5		6	
$I_{IB}$ Input bias current	25°C	-18	-50	-15	-35	-18	-50	nA
	Full range		-90		-55		-90	
$V_{ICR}$ Common-mode input voltage range	25°C	0	-0.3	0	-0.3	0	0.3	
		to	to	to	to	to	to	
		3.5	3.8	3.5	3.8	3.5	3.8	
$V_{OM}$ Maximum-peak output voltage swing	Full range	0	to	0	to	0	to	V
		to	3	to	3	to	3	
$V_{OM}$ Maximum-peak output voltage swing	Output low, No load	15	25	15	25	15	25	mV
	Output low, $R_L = 600\ \Omega$ to GND	5	10	5	10	5	10	
	Output low, $I_{s\text{ink}} = 1\ \text{mA}$		13		13		13	
	Output high, No load	220	350	220	350	220	350	
	Output high, $R_L = 600\ \Omega$ to GND	4	4.4	4	4.4	4	4.4	
$A_{VD}$ Large-signal differential voltage amplification	25°C	3.4	4	3.4	4	3.4	4	V
	Full range	3.2		3.3		3.2		
	$V_O = 5\ \text{mV}$ to $4\ \text{V}$ , $R_L = 500\ \Omega$	1		1		1		
$I_{CC}$ Supply current per amplifier	25°C	0.32	0.5	0.31	0.45	0.32	0.5	mA
	Full range		0.55		0.5		0.55	

† Full range is  $-0^\circ\text{C}$  to  $70^\circ\text{C}$ .

operating characteristics,  $V_{CC\pm} = \pm 15\ \text{V}$ ,  $V_{IC} = 0$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate		0.2	0.4		$V/\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		24		$nV/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$		22		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{Hz}$ to $10\ \text{Hz}$		0.55		$\mu V$
$I_n$ Equivalent input noise current	$f = 10\ \text{Hz}$		0.07		$pA/\sqrt{\text{Hz}}$

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

electrical characteristics at specified free-air temperature,  $V_{CC} = \pm 15V$ ,  $V_{IC} = 0$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TA †	LT1013		LT1013A ‡		LT1013DI		UNIT
			MIN	TYP ‡	MAX	MIN	TYP ‡	MAX	
V <sub>IO</sub>	R <sub>S</sub> = 50 Ω	25°C	60	300	40	150	200	800	μV
		Full range		550	300		1000		
α <sub>V</sub> IO	Temperature coefficient of input offset voltage	25°C	0.4	2.5	0.3	2	0.7	5	μV/°C
		Full range							
I <sub>IO</sub>	Long-term drift of input offset voltage	25°C	0.5		0.4		0.5		μV/mo
		Full range							
I <sub>IB</sub>	Input offset current	25°C	0.2	1.5	0.15	0.8	0.2	1.5	nA
		Full range		2.8	1.5		2.8		
I <sub>IB</sub>	Input bias current	25°C	-15	-30	-12	-20	-15	-30	nA
		Full range		-38		-25		-38	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-15	-15.3	-15	-15.3	-15	-15.3	V
		Full range		to 13		to 13		to 13	
V <sub>OM</sub>	Maximum peak output voltage swing	25°C	±12.5	±14	±13	±14	±12.5	±14	V
		Full range		±12		±12		±12	
A <sub>VD</sub>	Large-signal differential voltage amplification	25°C	0.5	0.2	0.8	2.5	0.5	2	V/μV
		Full range		1.2		1.5		1.2	
CMRR	Common-mode rejection ratio	25°C	97	114	100	117	97	114	dB
		Full range		94		97		94	
K <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC</sub> /ΔV <sub>IO</sub> )	25°C	100	117	103	120	100	117	dB
		Full range		97		101		97	
f <sub>id</sub>	Channel separation	25°C	120	137	123	140	120	137	dB
		Full range		70		100		70	
f <sub>ic</sub>	Common-mode input resistance	25°C	4		5		4		GΩ
		Full range		0.35		0.35		0.35	
I <sub>CC</sub>	Supply current per amplifier	25°C	0.35	0.55	0.35	0.5	0.35	0.55	mA
		Full range		0.7		0.55		0.6	

† Full range is -40°C to 105°C.  
‡ All typical values are at T<sub>A</sub> = 25°C.



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

electrical characteristics at specified free-air temperature,  $V_{CC+} = 5V$ ,  $V_{CC-} = 0V$ ,  $V_O = 1.4V$ ,  $V_{IC} = 0$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1013†			LT1013AI			LT1013DI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$R_S = 50 \Omega$	25°C	90	450	60	250	250	950	1200	$\mu V$	
		Full range		570	350						
$I_{IO}$ Input offset current		25°C	0.3	2	0.2	1.3	0.3	2	6	nA	
		Full range		6	3.5						
$I_{IB}$ Input bias current		25°C	-18	-50	-15	-35	-18	-50	-90	nA	
		Full range		-90	-55						
$V_{ICR}$ Common-mode input voltage range		25°C	0	-0.3	0	-0.3	0	0.3	0.3	V	
		Full range	to 3.5	to 3.8	to 3.5	to 3.8	to 3.5	to 3.8	to 3		
$V_{OM}$ Maximum-peak output voltage swing	Output low, No load	25°C	15	25	15	25	15	25	25	mV	
		Full range		13	13						
		25°C	220	350	220	350	220	350	350		
		Full range		3							
$A_{VD}$ Large-signal differential voltage amplification	Output high, $R_L = 600 \Omega$ to GND	25°C	4	4.4	4	4.4	4	4.4	4	V	
		Full range		3.2	3.3						
		25°C	3.4	4	3.4	4	3.4	4	4		
		Full range		3.2	3.3						
$I_{CC}$ Supply current per amplifier	$V_O = 5 mV$ to $4V$ , $R_L = 500 \Omega$	25°C	0.32	0.5	0.31	0.45	0.32	0.5	0.55	mA	
		Full range		0.55		0.5					

† Full range is  $-40^\circ C$  to  $105^\circ C$ .

operating characteristics,  $V_{CC\pm} = \pm 15V$ ,  $V_{IC} = 0$ ,  $T_A = 25^\circ C$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate		0.2	0.4		V/ $\mu s$
$V_n$ Equivalent input noise voltage	$f = 10 Hz$		24		nV/ $\sqrt{Hz}$
$V_n(PP)$ Peak-to-peak equivalent input noise voltage	$f = 1 kHz$		22		$\mu V$
$I_n$ Equivalent input noise current	$f = 0.1 Hz$ to $10 Hz$		0.55		pA/ $\sqrt{Hz}$
	$f = 10 Hz$		0.07		



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

electrical characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $V_{IC} = 0$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TA †	LT1013M		LT1013AM		LT1013DM		UNIT
			MIN	TYP ‡	MAX	MIN	TYP ‡	MAX	
V <sub>IO</sub>	PS = 50 Ω	25°C	60	300	40	150	200	800	μV
		Full range		550	300		1000		
α <sub>VIO</sub>	Temperature coefficient of input offset voltage	Full range	0.5	2.5*	0.4	2*	0.5	2.5*	μV/°C
		25°C	0.5		0.4		0.5		
I <sub>IO</sub>	Long-term drift of input offset voltage	25°C	0.2	1.5	0.15	0.8	0.2	1.5	nA
		Full range		5	2.5		5		
I <sub>IB</sub>	Input bias current	25°C	-15	-30	-12	-20	-15	-30	nA
		Full range		-45		-30		-45	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-15 to 13.8	-15.3 to 13.8	-15 to 13.8	-15.3 to 13.8	-15 to 13.8	-15.3 to 13.8	V
		Full range		-14.9 to 13		-14.9 to 13		-14.9 to 13	
V <sub>OM</sub>	Maximum peak output voltage swing	25°C	±12.5	±14	±13	±14	±12.5	±14	V
		Full range		±11.5		±11.5		±11.5	
A <sub>VD</sub>	Large-signal differential voltage amplification	25°C	0.5	2	0.8	2.5	0.5	2	V/μV
		Full range		0.25		0.5		0.25	
CMRR	Common-mode rejection ratio	25°C	97	117	100	117	97	114	dB
		Full range		94		97		94	
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC</sub> /ΔV <sub>IO</sub> )	25°C	100	117	103	120	100	117	dB
		Full range		97		100		97	
r <sub>id</sub>	Channel separation	25°C	120	137	123	140	120	137	dB
		Full range		70		100		70	
r <sub>ic</sub>	Common-mode input resistance	25°C	4		5		4		GΩ
		Full range		0.35		0.35		0.35	
I <sub>CC</sub>	Supply current per amplifier	25°C	0.35	0.55	0.35	0.5	0.35	0.55	mA
		Full range		0.7		0.6		0.7	

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is -55°C to 125°C.

‡ All typical values are at T<sub>A</sub> = 25°C.



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

**electrical characteristics at specified free-air temperature,  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = 0$ ,  $V_O = 1.4\text{ V}$ ,  $V_{IC} = 0$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TA†	LT1013M			LT1013AM			LT1013DM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	R <sub>S</sub> = 50 Ω	25°C	90	450	250	60	250	250	950			
		Full range	400	1500	900	250	900	800	2000			μV
I <sub>IO</sub> Input offset current	V <sub>IC</sub> = 0.1 V	25°C	200	750	450	120	450	560	1200			
		Full range	0.3	2	1.3	0.2	1.3	0.3	2			nA
I <sub>IB</sub> Input bias current		25°C	-18	-50	-35	-15	-35	-18	-50			
		Full range		-120	-80		-80		-120			nA
V <sub>ICR</sub> Common-mode input voltage range		25°C	0	-0.3	0	0	-0.3	0	-0.3			
		Full range	0	to 3.5	to 3.8	0	to 3.5	to 3.8	0	to 3.5	to 3.8	
V <sub>OM</sub> Maximum-peak output voltage swing	Output low, No load	25°C	15	25	25	15	25	15	25			
		Full range	5	10	10	5	10	5	10			mV
	Output high, No load	25°C	220	350	350	220	350	220	350			
		Full range	4	4.4	4.4	4	4.4	4	4.4			V/μV
	Output high, R <sub>L</sub> = 600 Ω to GND	25°C	3.4	4	4	3.4	4	3.4	4			
		Full range	3.1	3.1	3.2	3.2	3.1	3.1	3.1			
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>O</sub> = 5 mV to 4 V, R <sub>L</sub> = 500 Ω	25°C	1	1	1	1	1	1				
I <sub>CC</sub> Supply current per amplifier		25°C	0.32	0.5	0.45	0.31	0.45	0.32	0.5			mA
Full range		0.65	0.65	0.55	0.55	0.65	0.65	0.65				

† Full range is -55°C to 125°C.

**operating characteristics,  $V_{CC±} = ±15\text{ V}$ ,  $V_{IC} = 0$ , T<sub>A</sub> = 25°C**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate		0.2	0.4		V/μs
V <sub>n</sub> Equivalent input noise voltage	f = 10 Hz		24		nV/√Hz
V <sub>n(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 1 kHz		22		μV
I <sub>n</sub> Equivalent input noise current	f = 0.1 Hz to 10 Hz		0.55		pA/√Hz
	f = 10 Hz		0.07		



# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

electrical characteristics at  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = 0$ ,  $V_O = 1.4\text{ V}$ ,  $V_{IC} = 0$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1013Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$R_S = 50\ \Omega$		250	950	$\mu\text{V}$
$I_{IO}$ Input offset current			0.3	2	nA
$I_{IB}$ Input bias current			-18	-50	nA
$V_{ICR}$ Common-mode input voltage range		0 to 3.5	0.3 to 3.8		V
$V_{OM}$ Maximum peak output voltage swing	Output low, No load		15	25	mV
	Output low, $R_L = 600\ \Omega$ to GND		5	10	
	Output low, $I_{\text{sink}} = 1\text{ mA}$		220	350	
	Output high, No load		4	4.4	V
	Output high, $R_L = 600\ \Omega$ to GND		3.4	4	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 5\text{ mV}$ to $4\text{ V}$ , $R_L = 500\ \Omega$		1		$\text{V}/\mu\text{V}$
$I_{CC}$ Supply current per amplifier			0.32	0.5	mA

electrical characteristics at  $V_{CC+} = \pm 15\text{ V}$ ,  $V_{IC} = 0$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	LT1013Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$R_S = 50\ \Omega$		200	800	$\mu\text{V}$
	Long-term drift of input offset voltage		0.5		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current			0.2	1.5	nA
$I_{IB}$ Input bias current			-15	-30	nA
$V_{ICR}$ Common-mode input voltage range		-15 to 13.5	-15.3 to 13.8		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 2\text{ k}\Omega$		$\pm 12.5$	$\pm 14$	V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$ , $R_L = 600\ \Omega$		0.5	2	$\text{V}/\mu\text{V}$
	$V_O = \pm 10\text{ V}$ , $R_L = 2\ \Omega$		1.2	7	
CMRR Common-mode rejection ratio	$V_{IC} = -15\text{ V}$ to $13.5\text{ V}$		97	114	dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC} / \Delta V_{IO}$ )	$V_{CC\pm} = \pm 2\text{ V}$ to $\pm 18\text{ V}$		100	117	
	Channel separation	$V_O = \pm 10\text{ V}$ , $R_L = 2\ \Omega$	120	137	dB
$r_{id}$ Differential input resistance			70	300	$\text{M}\Omega$
$r_{ic}$ Common-mode input resistance			4		$\text{G}\Omega$
$I_{CC}$ Supply current per amplifier			0.35	0.55	mA

operating characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $V_{IC} = 0$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	LT1013Y			UNIT
		MIN	TYP	MAX	
SR Slew rate		0.2	0.4		$\text{V}/\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$		24		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$		22		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$		0.55		$\mu\text{V}$
$I_n$ Equivalent input noise current	$f = 10\text{ Hz}$		0.07		$\text{pA}/\sqrt{\text{Hz}}$





# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
$V_{IO}$	Input offset voltage	vs Source resistance	1
		vs Temperature	2
$\Delta V_{IO}$	Change in input offset voltage	vs Time	3
$I_{IO}$	Input offset current	vs Temperature	4
$I_{IB}$	Input bias current	vs Temperature	5
$V_{IC}$	Common-mode input voltage	vs Input bias current	6
$A_{VD}$	Differential voltage amplification	vs Load resistance	7, 8
		vs Frequency	9, 10
	Channel separation	vs Frequency	11
	Output saturation voltage	vs Temperature	12
CMRR	Common-mode rejection ratio	vs Frequency	13
kSVR	Supply voltage rejection ratio	vs Frequency	14
$I_{CC}$	Supply current	vs Temperature	15
$I_{OS}$	Short-circuit output current	vs Time	16
$V_n$	Equivalent input noise voltage	vs Frequency	17
$I_n$	Equivalent input noise current	vs Frequency	17
$V_{n(PP)}$	Peak-to-peak input noise voltage	vs Time	18
	Pulse response	Small signal	19, 21
		Large signal	20, 22, 23
	Phase shift	vs Frequency	9

# LT1013, LT1013A, LT1013D, LT1013V DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

## TYPICAL CHARACTERISTICS†

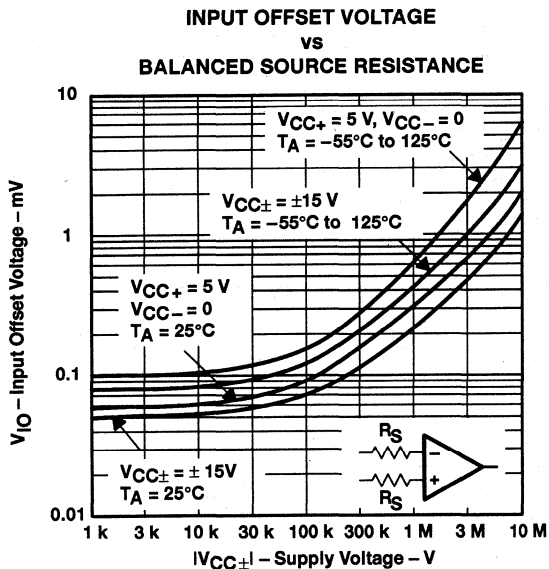


Figure 1

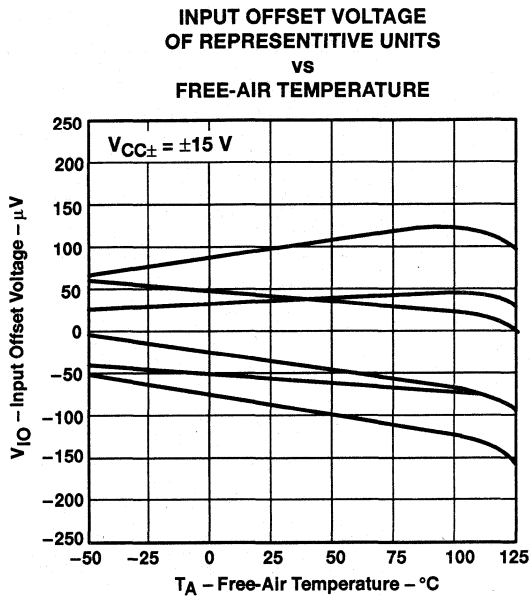


Figure 2

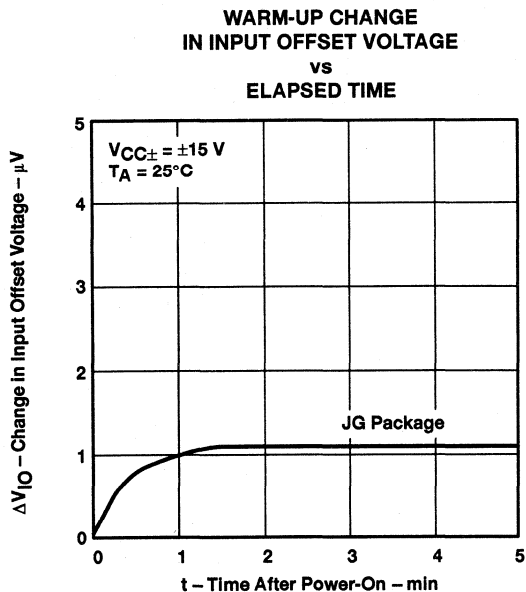


Figure 3

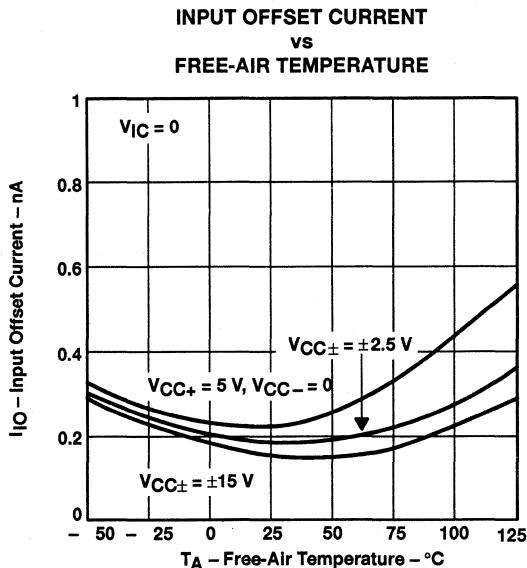


Figure 4

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

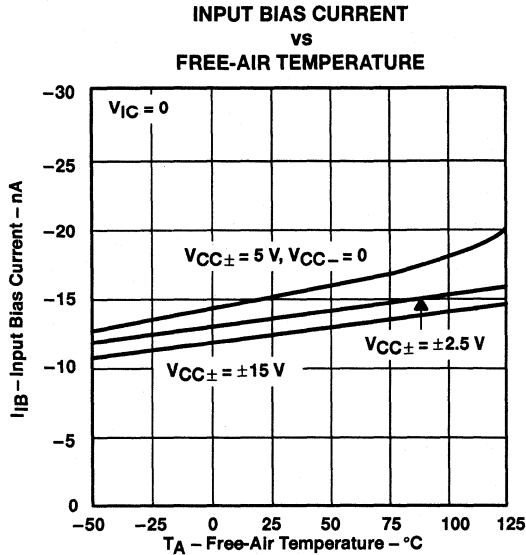


Figure 5

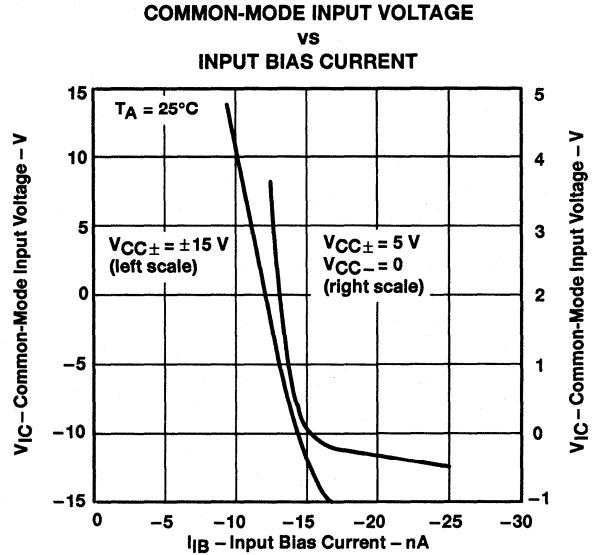


Figure 6

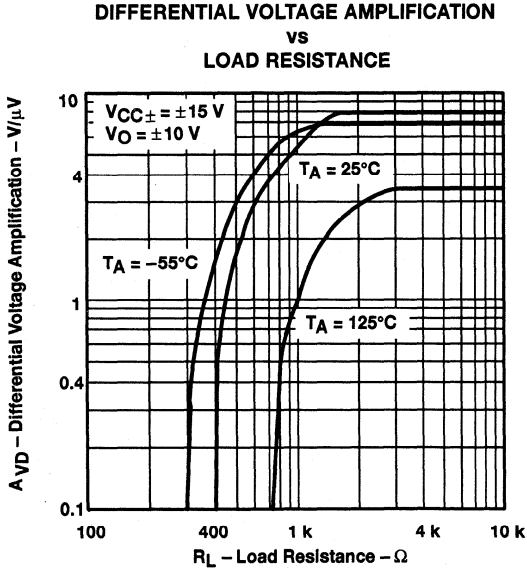


Figure 7

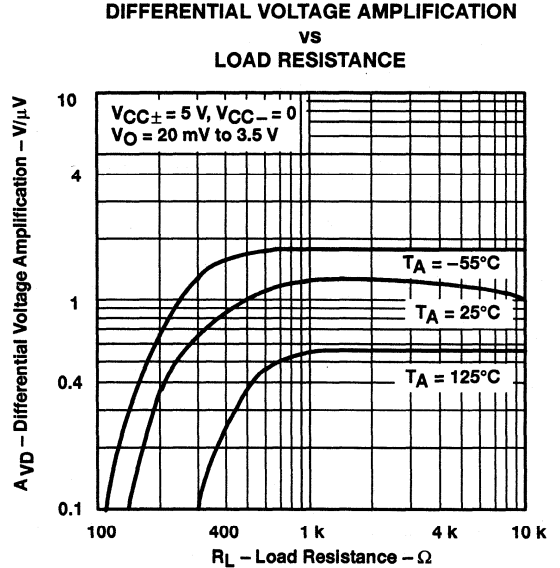


Figure 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

## TYPICAL CHARACTERISTICS†

**DIFFERENTIAL VOLTAGE AMPLIFICATION  
AND PHASE SHIFT  
VS  
FREQUENCY**

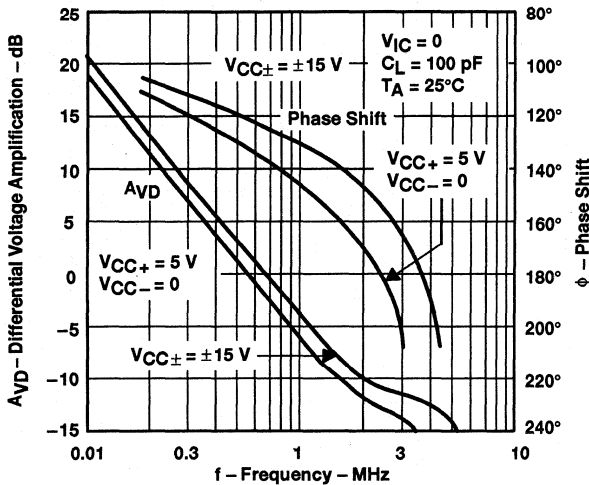


Figure 9

**DIFFERENTIAL VOLTAGE AMPLIFICATION  
VS  
FREQUENCY**

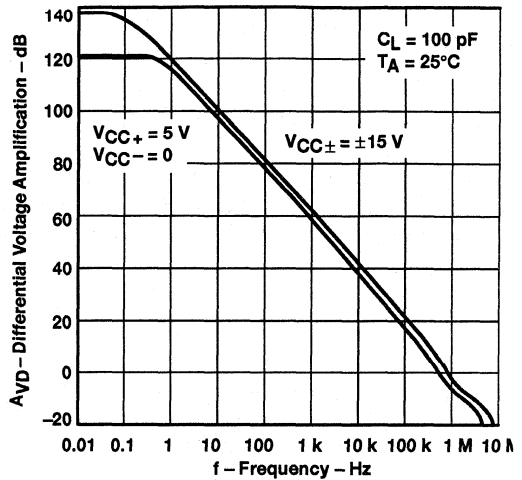


Figure 10

**CHANNEL SEPARATION  
VS  
FREQUENCY**

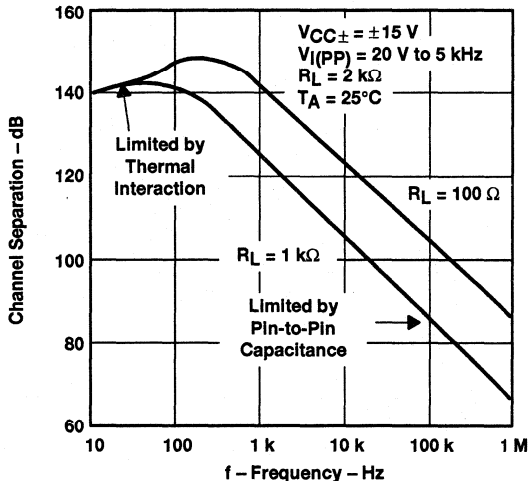


Figure 11

**OUTPUT SATURATION VOLTAGE  
VS  
FREE-AIR TEMPERATURE**

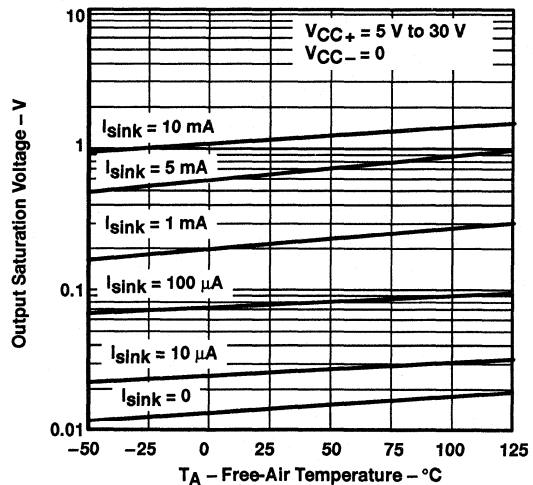


Figure 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

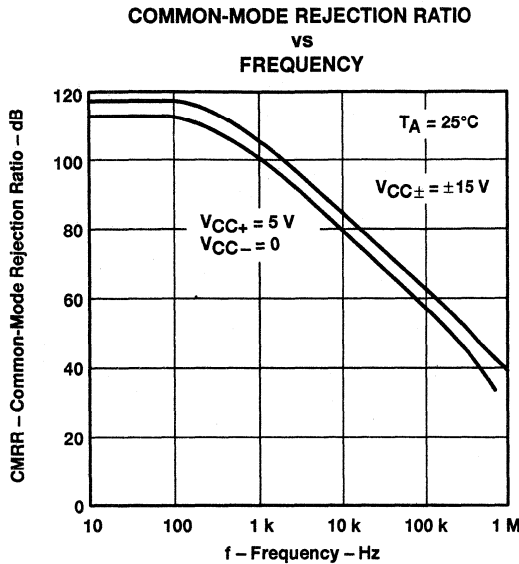


Figure 13

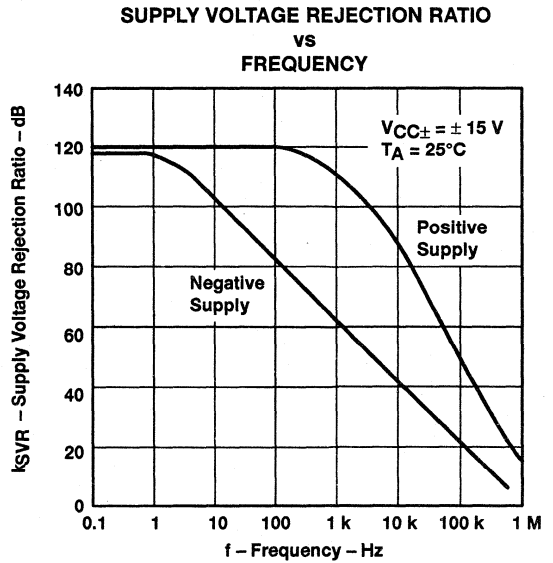


Figure 14

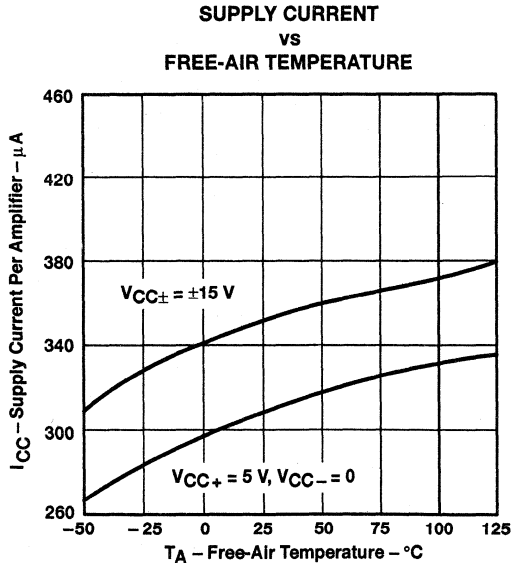


Figure 15

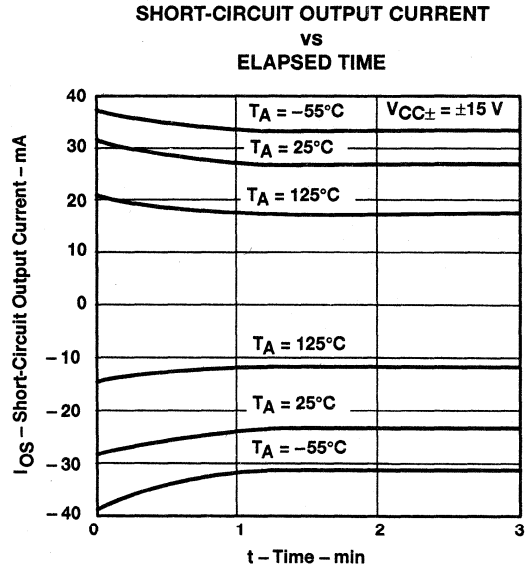


Figure 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## TYPICAL CHARACTERISTICS

**EQUIVALENT INPUT NOISE VOLTAGE  
AND EQUIVALENT INPUT NOISE CURRENT  
vs  
FREQUENCY**

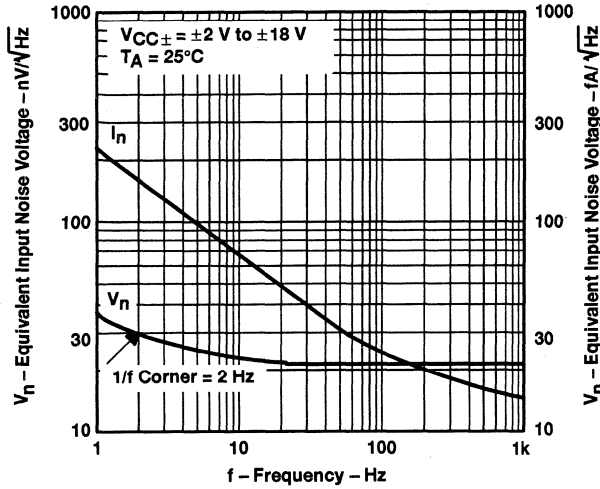


Figure 17

**PEAK-TO-PEAK INPUT NOISE VOLTAGE  
OVER A  
10-SECOND PERIOD**

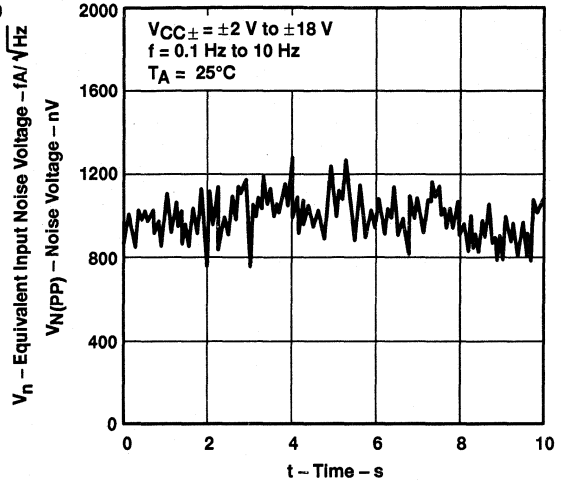


Figure 18

**VOLTAGE-FOLLOWER  
SMALL-SIGNAL  
PULSE RESPONSE**

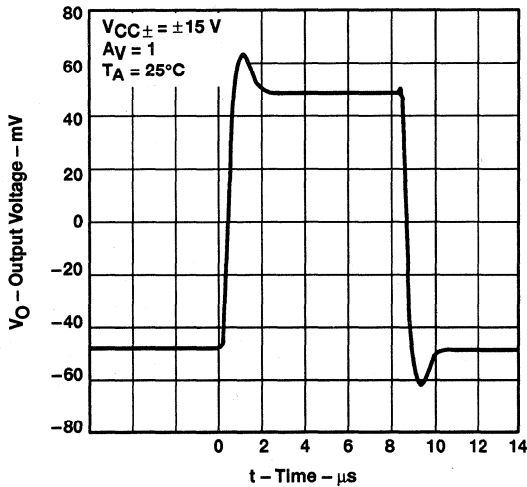


Figure 19

**VOLTAGE-FOLLOWER  
LARGE-SIGNAL  
PULSE-RESPONSE**

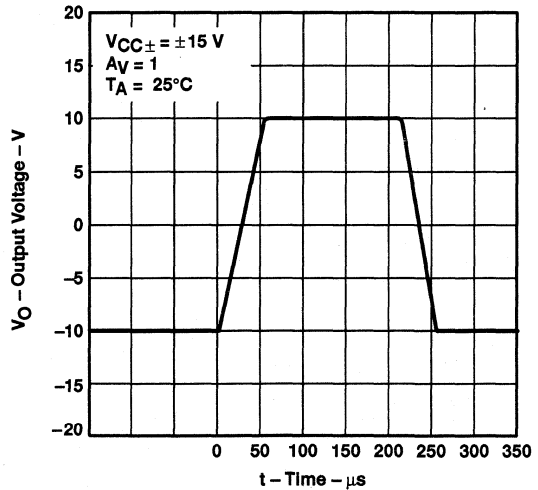


Figure 20

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE

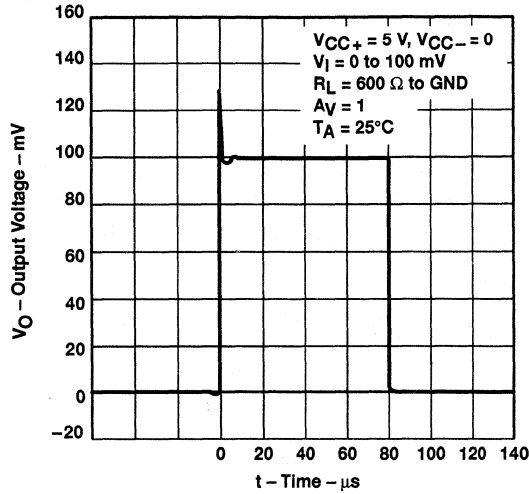


Figure 21

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE

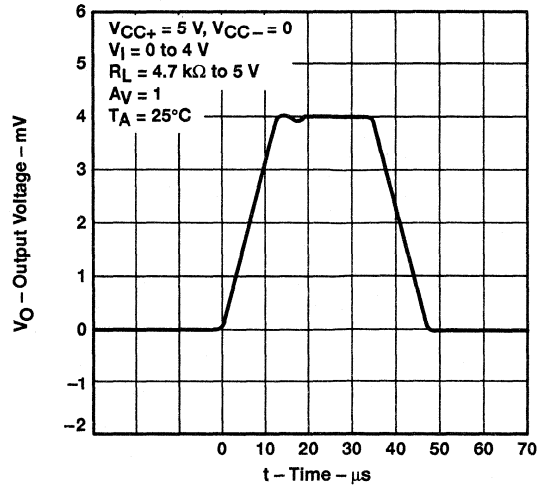


Figure 22

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE

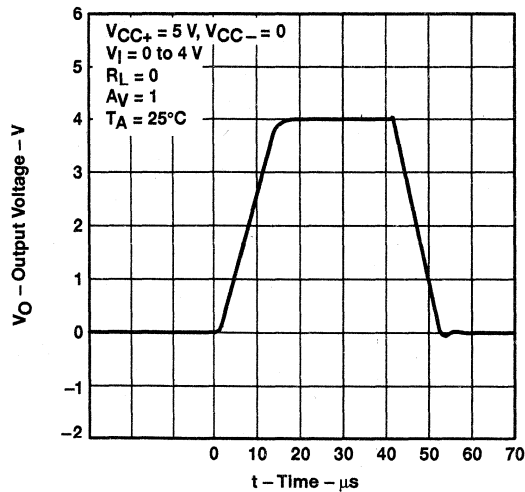


Figure 23

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## APPLICATION INFORMATION

### single-supply operation

The LT1013 is fully specified for single-supply operation ( $V_{CC-} = 0$ ). The common-mode input voltage range includes ground, and the output swings within a few millivolts of ground.

Furthermore, the LT1013 has specific circuitry that addresses the difficulties of single-supply operation, both at the input and at the output. At the input, the driving signal can fall below 0 V, either inadvertently or on a transient basis. If the input is more than a few hundred millivolts below ground, the LT1013 is designed to deal with the following two problems that can occur:

1. On many other operational amplifiers, when the input is more than a diode drop below ground, unlimited current will flow from the substrate ( $V_{CC-}$  terminal) to the input, which can destroy the unit. On the LT1013, the 400- $\Omega$  resistors in series with the input (see schematic) protect the device even when the input is 5 V below ground.
2. When the input is more than 400 mV below ground (at  $T_A = 25^\circ\text{C}$ ), the input stage of similar type operational amplifiers saturates and phase reversal occurs at the output. This can cause lock up in servo systems. Because of a unique phase-reversal protection circuitry (Q21, Q22, Q27, and Q28), the LT1013 outputs do not reverse, even when the inputs are at -1.5 V (see Figure 24).

This phase-reversal protection circuitry does not function when the other operational amplifier on the LT1013 is driven hard into negative saturation at the output. Phase-reversal protection does not work on amplifier 1 when 2's output is in negative saturation or on amplifier 2 when 1s output is in negative saturation.

At the output, other single-supply designs either cannot swing to within 600 mV of ground or cannot sink more than a few microamperes while swinging to ground. The all-NPN output stage of the LT1013 maintains its low output resistance and high gain characteristics until the output is saturated. In dual-supply operations, the output stage is free of crossover distortion.

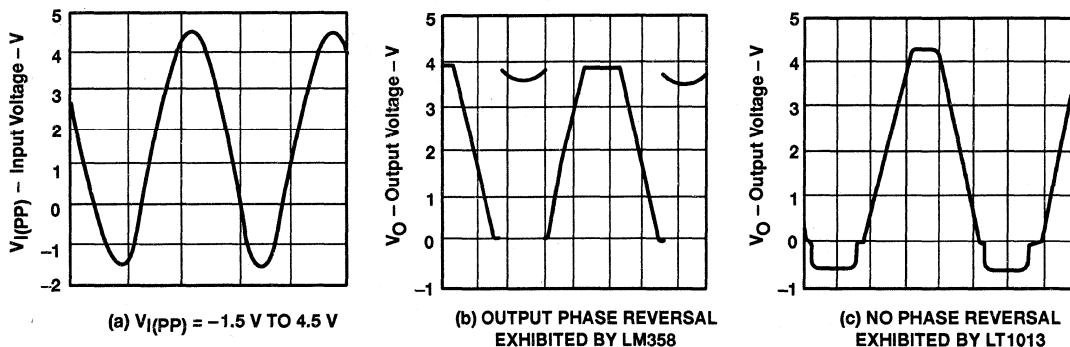


Figure 24. Voltage-Follower Response With Input Exceeding the Negative Common-Mode Input Voltage Range



APPLICATION INFORMATION

comparator applications

The single-supply operation of the LT1013 lends itself for use as a precision comparator with TTL-compatible output. In systems using both operational amplifiers and comparators, the LT1013 can perform multiple duties. Refer to Figures 25 and 26.

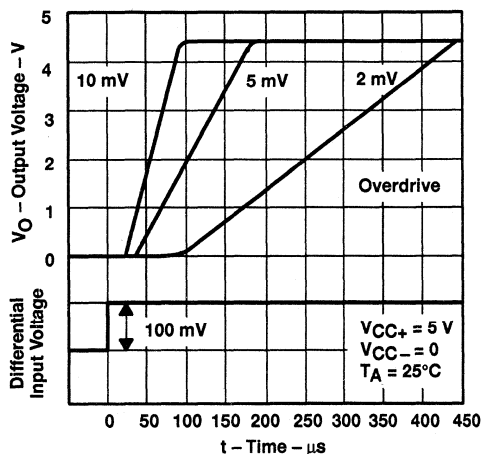


Figure 25. Low-to-High-Level Output Response for Various Input Overdrives

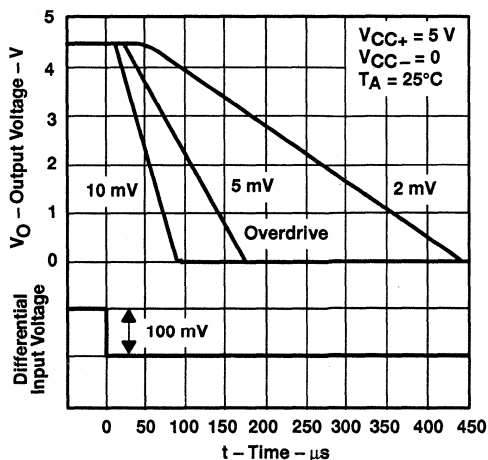


Figure 26. High-to-Low-Level Output Response for Various Input Overdrives

low-supply operation

The minimum supply voltage for proper operation of the LT1013 is 3.4 V (three Ni-Cad batteries). Typical supply current at this voltage is 290  $\mu$ A; therefore, power dissipation is only 1mW per amplifier.

offset voltage and noise testing

The test circuit for measuring input offset voltage and its temperature coefficient is shown in Figure 30. This circuit with supply voltages increased to  $\pm 20$  V is also used as the burn-in configuration.

The peak-to-peak equivalent input noise voltage of the LT1013 is measured using the test circuit shown in Figure 27. The frequency response of the noise tester indicates that the 0.1-Hz corner is defined by only one zero. The test time to measure 0.1-Hz to 10-Hz noise should not exceed 10 seconds, as this time limit acts as an additional zero to eliminate noise contribution from the frequency band below 0.1 Hz.

An input noise voltage test is recommended when measuring the noise of a large number of units. A 10-Hz input noise voltage measurement correlates well with a 0.1-Hz peak-to-peak noise reading because both results are determined by the white noise and the location of the  $1/f$  corner frequency.

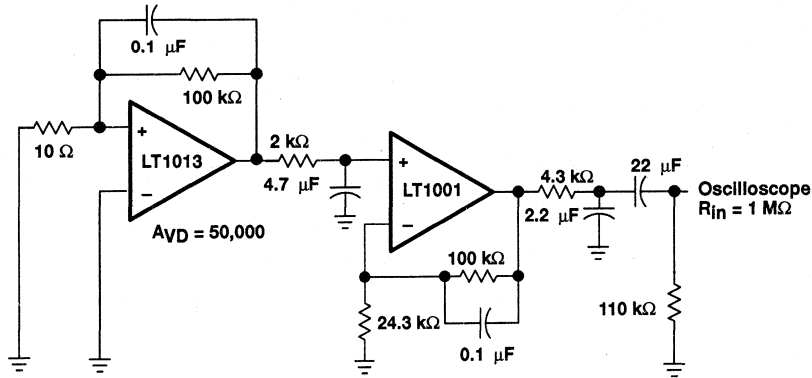
Current noise is measured by the circuit and formula shown in Figure 28. The noise of the source resistors is subtracted.

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A - MAY 1988 - REVISED AUGUST 1991

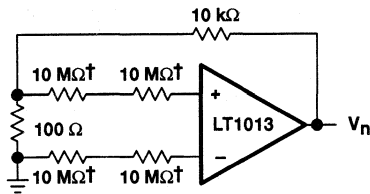
## APPLICATION INFORMATION

### offset voltage and noise testing (continued)



NOTE: All capacitor values are for nonpolarized capacitors only.

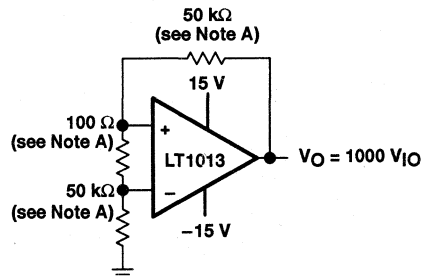
Figure 27. 0.1-Hz to 10-Hz Peak-to-Peak Noise Test Circuit



$$I_n = \frac{[V_{no}^2 - (820 \text{ nV})^2]^{1/2}}{40 \text{ M}\Omega / 100}$$

† Metal-film resistor

Figure 28. Noise-Current Test Circuit and Formula

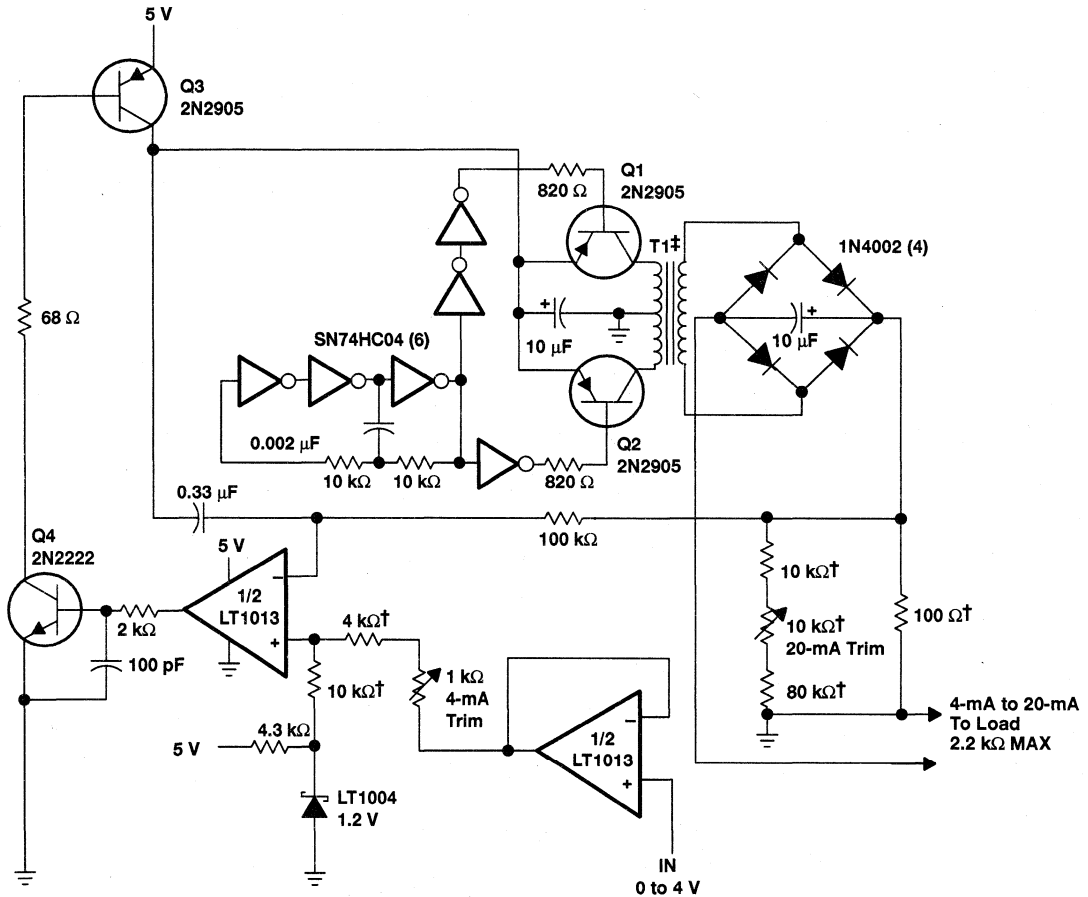


NOTE A: Resistors must have low thermoelectric potential.

Figure 29. Test Circuit for  $V_{IO}$  and  $\alpha V_{IO}$

APPLICATION INFORMATION

typical applications



† 1% film resistor. Match 10-kΩ resistors 0.05%.

‡ T1 = PICO-31080

Figure 30. 5-V Powered 4-mA – 20-mA Current Loop Transmitter With 12-Bit Accuracy

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## APPLICATION INFORMATION

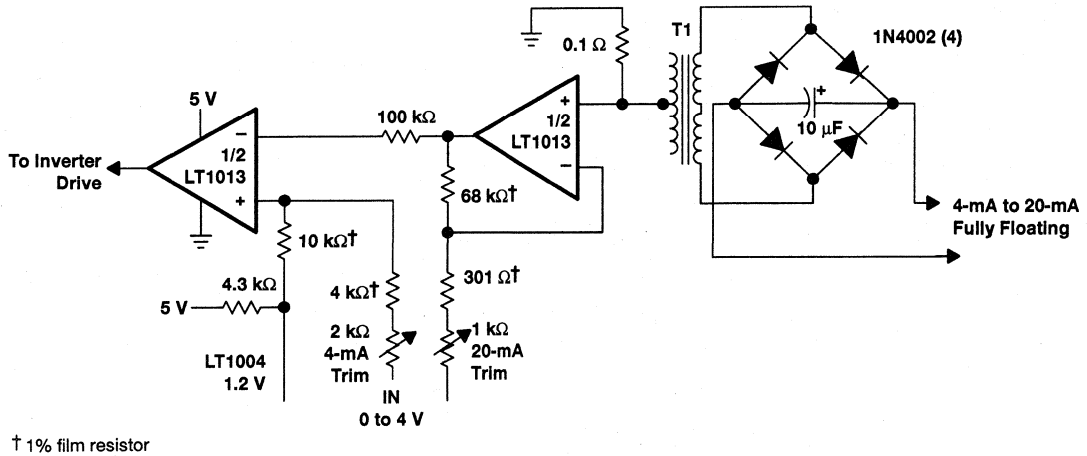
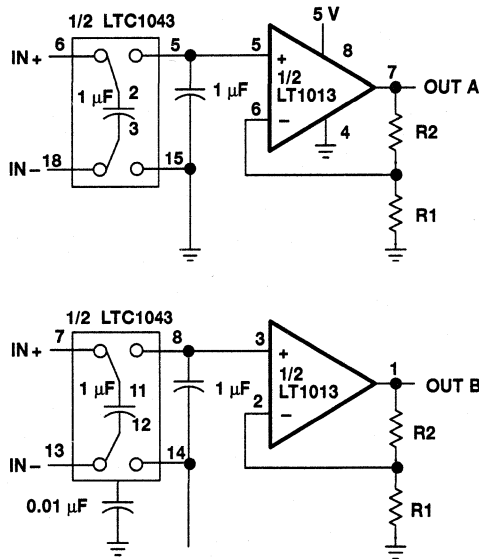


Figure 31. Fully Floating Modification to 4-mA – 20-mA Current Loop Transmitter With 8-Bit Accuracy



NOTE:  $V_{IO} = 150 \mu\text{V}$ ,  $A_{VD} = (R1/R2) + 1$ ,  $CMRR = 120 \text{ dB}$ ,  $V_{ICR} = 0 \text{ to } 5 \text{ V}$

Figure 32. 5-V Single-Supply Dual Instrumentation Amplifier

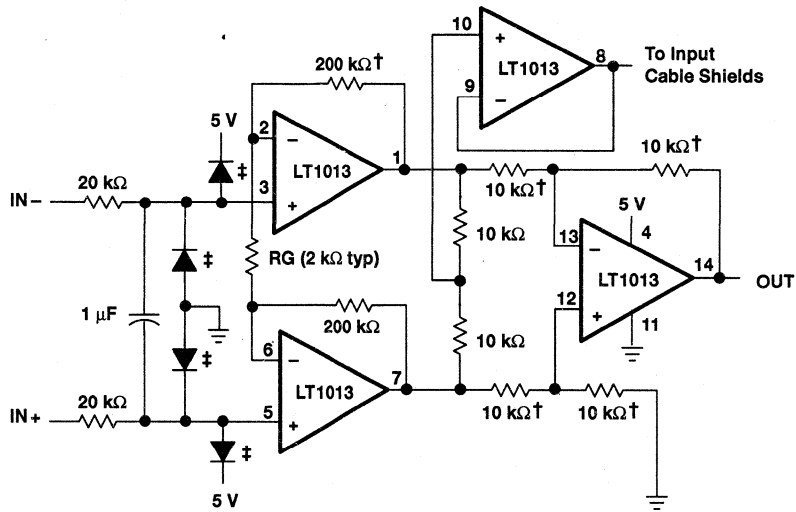
TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# LT1013, LT1013A, LT1013D, LT1013Y DUAL PRECISION OPERATIONAL AMPLIFIERS

SLOS018A – MAY 1988 – REVISED AUGUST 1991

## APPLICATION INFORMATION



† 1% film resistor. Match 10-kΩ resistors 0.05%.  
‡ For high source impedances, use 2N2222 as diodes.  
NOTE:  $A_{VD} = (400,000/RG) + 1$

Figure 33. 5-V Powered Precision Instrumentation Amplifier



# MC1458, MC1558 DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

SLOS069 – FEBRUARY 1971 – REVISED OCTOBER 1990

- Short-Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-Up
- Designed to Be Interchangeable With Motorola MC1558/MC1458 and Signetics S5558/N5558

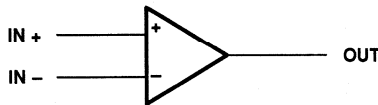
## description

The MC1458 and MC1558 are dual general-purpose operational amplifiers with each half electrically similar to the  $\mu$ A741 except that offset null capability is not provided.

The high-common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

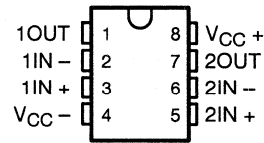
The MC1458 is characterized for operation from 0°C to 70°C. The MC1558 is characterized for operation over the full military temperature range of -55°C to 125°C.

## symbol (each amplifier)

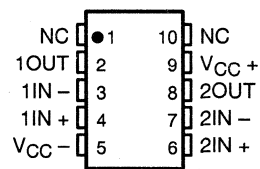


MC1458 ... D OR P PACKAGE  
MC1558 ... JG PACKAGE

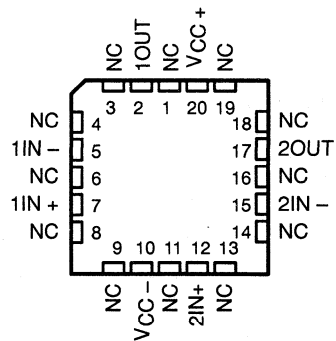
(TOP VIEW)



MC1558 ... U PACKAGE  
(TOP VIEW)



MC1558 ... FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	CERAMIC FLAT PACK (U)
0°C to 70°C	6 mV	MC1458CD	—	—	MC1458CP	—
-55°C to 125°C	5 mV	—	MC1558MFK	MC1558MSG	—	MC1558MU

The D packages are available taped and reeled. Add the suffix R to the device type (i.e., MC1458DR)

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

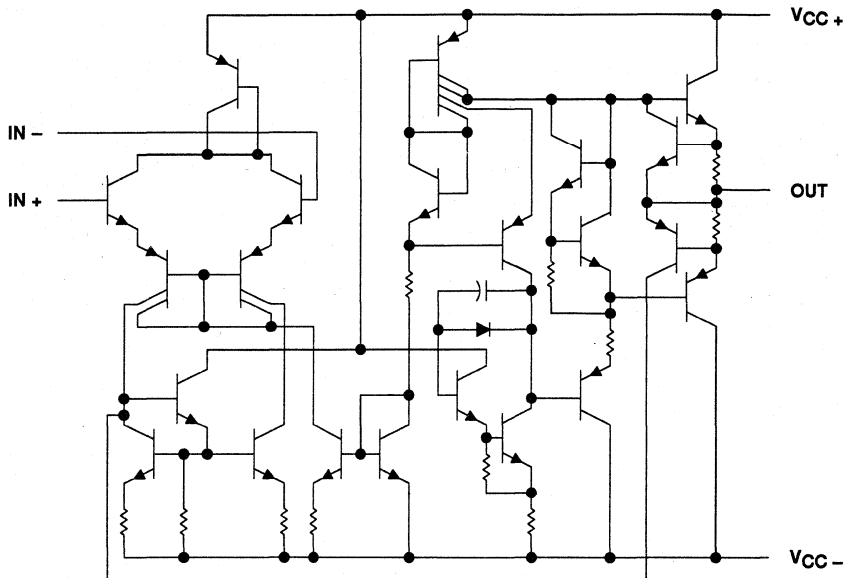
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1990, Texas Instruments Incorporated  
On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

# MC1458, MC1558 DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

SLOS069 – FEBRUARY 1971 – REVISED OCTOBER 1990

## schematic (each amplifier)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	MC1458	MC1558	UNIT
Supply voltage $V_{CC+}$ (see Note 1)	18	22	V
Supply voltage $V_{CC-}$ (see Note 1)	-18	-22	V
Differential input voltage (see Note 2)	$\pm 30$	$\pm 30$	V
Input voltage at either input (see Notes 1 and 3)	$\pm 15$	$\pm 15$	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	0 to 70	-55 to 125	$^{\circ}\text{C}$
Storage temperature range	65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds: FK package		260	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG or U package	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $\text{IN}+$  with respect to  $\text{IN}-$ .  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
 4. The output can be shorted to ground or either power supply. For the MC1558 only, the unlimited duration of the short circuit applies at (or below)  $125^{\circ}\text{C}$  case temperature or  $70^{\circ}\text{C}$  free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	680 mW	5.8 mW/ $^{\circ}\text{C}$	33 $^{\circ}\text{C}$	464 mW	—
FK	680 mW	11.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	880 mW	275 mW
JG	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	210 mW
P	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	—
U	675 mW	5.4 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	432 mW	135 mW

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# MC1458, MC1558

## DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

SLOS069 – FEBRUARY 1971 – REVISED OCTOBER 1990

### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC\pm}$	$\pm 5$		$\pm 15$	V

### electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15$ V

PARAMETER	TEST CONDITION <sup>†</sup>	MC1458			MC1558			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$	25°C	1	6	1	5	mV	
		Full range		7.5		6		
$I_{IO}$ Input offset current	$V_O = 0$	25°C	20	200	20	200	nA	
		Full range		300		500		
$I_{IB}$ Input bias current	$V_O = 0$	25°C	80	500	80	500	nA	
		Full range		800		1500		
$V_{ICR}$ Common-mode input voltage range		25°C	$\pm 12$	$\pm 13$	$\pm 12$	$\pm 13$	V	
		Full range	$\pm 12$		$\pm 12$			
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10$ k $\Omega$	25°C	$\pm 12$	$\pm 14$	$\pm 12$	$\pm 14$	V	
	$R_L \geq 10$ k $\Omega$	Full range	$\pm 12$		$\pm 12$			
	$R_L = 2$ k $\Omega$	25°C	$\pm 10$	$\pm 13$	$\pm 10$	$\pm 13$		
	$R_L \geq 2$ k $\Omega$	Full range	$\pm 10$		$\pm 10$			
$A_{VD}$ Large-signal differential voltage amplification	$R_L \geq 2$ k $\Omega$ , $V_O = \pm 10$ V	25°C	20	200	50	200	V/mV	
		Full range	15		25			
$B_{OM}$ Maximum-output-swing bandwidth (closed loop)	$R_L = 2$ k $\Omega$ , $V_O \geq \pm 10$ V, $A_{VD} = 1$ , THD $\geq 5\%$	25°C	14		14		kHz	
$B_1$ Unity-gain bandwidth		25°C	1		1		MHz	
$\phi_m$ Phase margin	$A_{VD} = 1$	25°C	65		65		°C	
		Gain margin	25°C	11		11		dB
$r_i$ Input resistance		25°C	0.3*	2	0.3*	2	M $\Omega$	
$r_o$ Output resistance	$V_O = 0$ , See Note 5	25°C	75		75		$\Omega$	
$C_i$ Input capacitance		25°C	1.4		1.4		pF	
$z_{ic}$ Common-mode input impedance	$f = 20$ Hz	25°C	200		200		M $\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR}$ min, $V_O = 0$	25°C	70	90	70	90	dB	
		Full range	70		70			
$k_{SVS}$ Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC} = \pm 9$ V to $\pm 15$ V, $V_O = 0$	25°C	30	150	30	150	$\mu$ V/V	
		Full range		150		150		
$V_n$ Equivalent input noise voltage (closed loop)	$A_{VD} = 100$ , $R_S = 0$ , $f = 1$ kHz, $BW = 1$ Hz	25°C	45		45		nV/ $\sqrt{Hz}$	

\*This parameter is not production tested.

<sup>†</sup> All characteristics are specified under open-loop operating conditions with zero common-mode input voltage unless otherwise specified. Full range for MC1458 is 0°C to 70°C and for MC1558 is -55°C to 125°C.

NOTE 5: This typical value applies only at frequencies above a few hundred hertz because of the effect of drift and thermal feedback.

# MC1458, MC1558 DUAL GENERAL-PURPOSE OPERATIONAL AMPLIFIERS

SLOS069 – FEBRUARY 1971 – REVISED OCTOBER 1990

## electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15\text{ V}$ (continued)

PARAMETER	TEST CONDITION†	MC1458			MC1558			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
$I_{OS}$	Short-circuit output current		25°C	±25	±40		±25	±40	mA
$I_{CC}$	Supply current (both amplifiers)	$V_O = 0$ , No load	25°C	3.4	5.6		3.4	5	mA
			Full range			6.6		6.6	
$P_D$	Total power dissipation (both amplifiers)	$V_O = 0$ , No load	25°C	100	170		100	150	mW
			Full range		200			200	
$V_{O1}/V_{O2}$	Crosstalk attenuation		25°C		120		120	dB	

† All characteristics are specified under open-loop operating conditions with zero common-mode input voltage unless otherwise specified. Full range for MC1458 is 0°C to 70°C and for MC1558 is -55°C to 125°C.

## operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MC1458			MC1558			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$t_r$	Rise time	$V_I = 20\text{ mV}$ , $C_L = 100\text{ pF}$	$R_L = 2\text{ k}\Omega$ , See Figure 1		0.3		0.3	$\mu\text{s}$
	Overshoot factor				5%		5%	
SR	Slew rate at unity gain	$V_I = 10\text{ V}$ , $C_L = 100\text{ pF}$	$R_L = 2\text{ k}\Omega$ , See Figure 1		0.5		0.5	$\text{V}/\mu\text{s}$

## PARAMETER MEASUREMENT INFORMATION

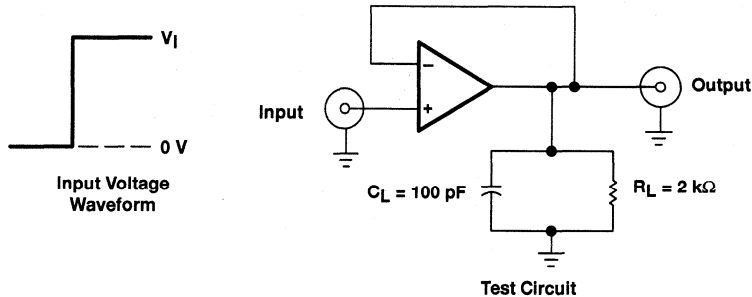


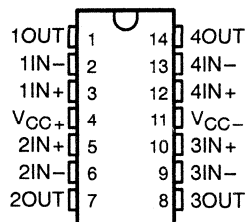
Figure 1. Rise Time, Overshoot, and Slew Rate Waveform and Test Circuit

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – D2517, FEBRUARY 1979 – REVISED SEPTEMBER 1990

- Wide Range of Supply Voltages Single Supply . . . 3 V to 36 V or Dual Supplies
- Class AB Output Stage
- True Differential Input Stage
- Low Input Bias Current
- Internal Frequency Compensation
- Short-Circuit Protection
- Designed to Be Interchangeable With Motorola MC3303, MC3403

D OR N PACKAGE  
(TOP VIEW)

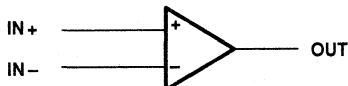


## description

The MC3303 and the MC3403 are quadruple operational amplifiers similar in performance to the  $\mu$ A741 but with several distinct advantages. They are designed to operate from a single supply over a range of voltages from 3 V to 36 V. Operation from split supplies is also possible provided the difference between the two supplies is 3 V to 36 V. The common-mode input range includes the negative supply. Output range is from the negative supply to  $V_{CC} - 1.5$  V. Quiescent supply currents are less than one-half those of the  $\mu$ A741.

The MC3303 is characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , and the MC3403 is characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ .

## symbol (each amplifier)



## AVAILABLE OPTIONS

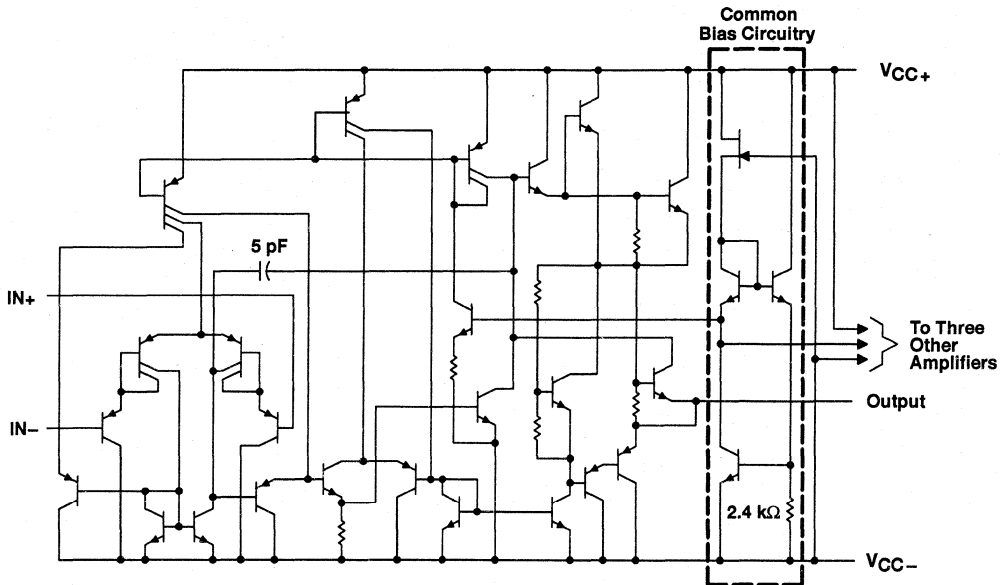
T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE	
		SMALL OUTLINE (D)	PLASTIC DIP (N)
0°C to 70°C	10 mV	MC3403D	MC3403N
-40°C to 85°C	8 mV	MC3303D	MC3303N

The D packages are available taped and reeled. Add R suffix to the device type (e.g., MC3403DR).

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – FEBRUARY 1979 – REVISED SEPTEMBER 1990

## schematic (each amplifier)



Component values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	MC3303	MC3403	UNIT
Supply voltage $V_{CC+}$ (see Note 1)	18	18	V
Supply voltage $V_{CC-}$ (see Note 1)	-18	-18	V
Supply voltage $V_{CC+}$ with respect to $V_{CC-}$	36	36	V
Differential input voltage (see Note 2)	$\pm 36$	$\pm 36$	V
Input voltage (see Notes 1 and 3)	$\pm 18$	$\pm 18$	V
Continuous total power dissipation	See Dissipation Rating Table		
Operating free-air temperature range	-40 to 85	0 to 70	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260	260	$^{\circ}\text{C}$

- NOTES: 1. These voltage values are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $\text{IN}+$  with respect to  $\text{IN}-$ .  
 3. Neither input must ever be more positive than  $V_{CC+}$  or more negative than  $V_{CC-}$ .

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR		$T_A = 85^{\circ}\text{C}$
	POWER RATING	ABOVE $T_A = 25^{\circ}\text{C}$		POWER RATING
D	950 mW	7.6 mW/ $^{\circ}\text{C}$	608 mW	494 mW
N	1150 mW	9.2 mW/ $^{\circ}\text{C}$	736 mW	598 mW

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – FEBRUARY 1979 – REVISED SEPTEMBER 1990

## recommended operating conditions

	MIN	MAX	UNIT
Single-supply voltage, $V_{CC}$	5	30	V
Dual-supply voltage, $V_{CC+}$	2.5	15	V
Dual-supply voltage, $V_{CC-}$	-2.5	-15	V

**electrical characteristics at specified free-air temperature,  $V_{CC+} = 14$  V,  $V_{CC-} = 0$  V for MC3303,  $V_{CC\pm} = \pm 15$  V for MC3403 (unless otherwise noted)**

PARAMETER	TEST CONDITIONS†	MC3303			MC3403			UNIT	
		MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage	See Note 4	25°C	2	8	2	10	mV		
		Full range	10			12			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	See Note 4	Full range	10		10		$\mu\text{V}/^\circ\text{C}$		
$I_{IO}$ Input offset current	See Note 4	25°C	30	75	30	50	nA		
		Full range	250			200			
$\alpha_{IIO}$ Temperature coefficient of input offset current	See Note 4	Full range	50		50		pA/C		
$I_{IB}$ Input bias current	See Note 4	25°C	-0.2	-0.5	-0.2	-0.5	$\mu\text{A}$		
		Full range	-1			-0.8			
$V_{ICR}$ Common-mode input voltage range‡		25°C	$V_{CC-}$ to 12	$V_{CC-}$ to 12.5	$V_{CC-}$ to 13	$V_{CC-}$ to 13.5	V		
$V_{OM}$ Peak output voltage swing	$R_L = 10$ k $\Omega$	25°C	12	12.5	$\pm 12$	$\pm 13.5$	V		
	$R_L = 2$ k $\Omega$	25°C	10	12	$\pm 10$	$\pm 13$			
	$R_L = 2$ k $\Omega$	Full range	10		$\pm 10$				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10$ V, $R_L = 2$ k $\Omega$	25°C	20	200	20	200	V/mV		
		Full range	15		15				
$B_{OM}$ Maximum-output-swing bandwidth	$V_{OPP} = 20$ V, $A_{VD} = 1$ , THD $\leq 5\%$ , $R_L = 2$ k $\Omega$	25°C	9		9		kHz		
$B_1$ Unity-gain bandwidth	$V_O = 50$ mV, $R_L = 10$ k $\Omega$	25°C	1		1		MHz		
$\phi_m$ Phase margin	$C_L = 200$ pF, $R_L = 2$ k $\Omega$	25°C	60°		60°				
$r_i$ Input resistance	$f = 20$ Hz	25°C	0.3	1	0.3	1	M $\Omega$		
$r_o$ Output resistance	$f = 20$ Hz	25°C	75		75		$\Omega$		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	70	90	70	90	dB		
$k_{SVS}$ Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC\pm} = \pm 2.5$ to $\pm 15$ V	25°C	30	150	30	150	$\mu\text{V}/\text{V}$		
$I_{OS}$ Short-circuit output current§		25°C	$\pm 10$	$\pm 30$	$\pm 45$	$\pm 10$	$\pm 30$	$\pm 45$	mA
$I_{CC}$ Total supply current	No load, See Note 4	25°C	2.8	7	2.8	7	mA		

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for  $T_A$  is -40°C to 85°C for MC3303, and 0°C to 70°C for MC3403.

‡ The  $V_{ICR}$  limits are directly linked volt-for-volt to supply voltage; the positive limit is 2 V less than  $V_{CC+}$ .

§ Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

NOTE 4:  $V_{IO}$ ,  $I_{IO}$ ,  $I_{IB}$ , and  $I_{CC}$  are defined at  $V_O = 0$  for MC3403 and  $V_O = 7$  V for MC3303.

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – FEBRUARY 1979 – REVISED SEPTEMBER 1990

electrical characteristics,  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = 0\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITION†	MC3303			MC3403			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 2.5\text{ V}$			10			mV
$I_{IO}$	Input offset current	$V_O = 2.5\text{ V}$			75			nA
$I_{IB}$	Input bias current	$V_O = 2.5\text{ V}$			-0.5			pA
$V_{OM}$	Peak output voltage swing‡	$R_L = 10\text{ k}\Omega$			3.3 3.5			V
		$R_L = 10\text{ k}\Omega$ , $V_{CC+} = 5\text{ V to } 30\text{ V}$			$V_{CC+} - 1.7$			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1.7\text{ V to } 3.3\text{ V}$ , $R_L = 2\text{ k}\Omega$			20 200			V/mV
$k_{SVS}$	Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC\pm}$ )	$V_{CC\pm} = \pm 2.5\text{ V to } \pm 15\text{ V}$			150			$\mu\text{V/V}$
$I_{CC}$	Supply current	$V_O = 2.5\text{ V}$ , No load			2.5 7			mA
$V_{O1}/V_{O2}$	Crosstalk attenuation	$f = 1\text{ kHz to } 20\text{ kHz}$			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

‡ Output will swing essentially to ground.

operating characteristics,  $V_{CC+} = 14\text{ V}$ ,  $V_{CC-} = 0\text{ V}$  for MC3303,  $V_{CC\pm} = \pm 15\text{ V}$  for MC3403,  $T_A = 25^\circ\text{C}$ ,  $A_{VD} = 1$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
SR	Slew rate at unity gain	$V_I = \pm 10\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 2\text{ k}\Omega$ , See Figure 1			0.6	$\text{V}/\mu\text{s}$
$t_r$	Rise time	$\Delta V_O = 50\text{ mV}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$ , See Figure 1			0.35	$\mu\text{s}$
$t_f$	Fall time	$\Delta V_O = 50\text{ mV}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$ , See Figure 1			0.35	$\mu\text{s}$
	Overshoot factor				20%	
	Crossover distortion	$V_{I(PP)} = 30\text{ mV}$ , $V_{OPP} = 2\text{ V}$ , $f = 10\text{ kHz}$			1%	

## PARAMETER MEASUREMENT INFORMATION

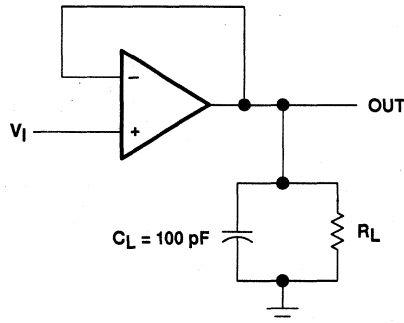


Figure 1. Unity-Gain Amplifier

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – FEBRUARY 1979 – REVISED SEPTEMBER 1990

## TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
$I_{IB}$	Input bias current	vs Free-air temperature
		vs Supply voltage
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Supply voltage
		vs Frequency
$A_{VD}$	Large-signal differential voltage amplification	vs Frequency
		vs Time

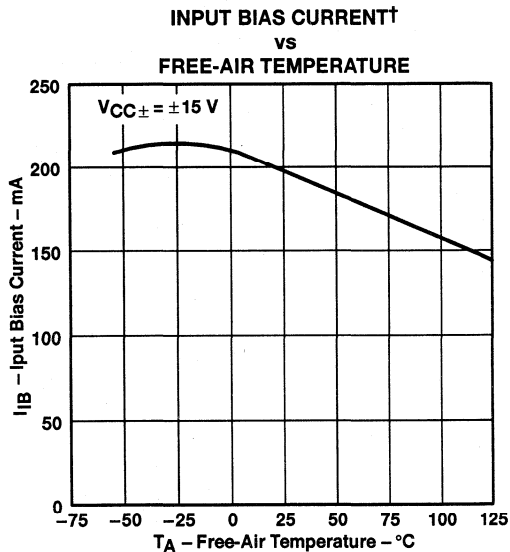


Figure 2

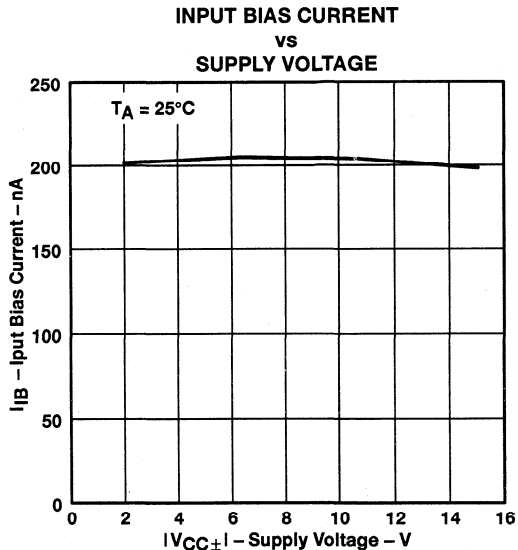


Figure 3

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# MC3303, MC3403 QUADRUPLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS101 – FEBRUARY 1979 – REVISED SEPTEMBER 1990

## TYPICAL CHARACTERISTICS†

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE

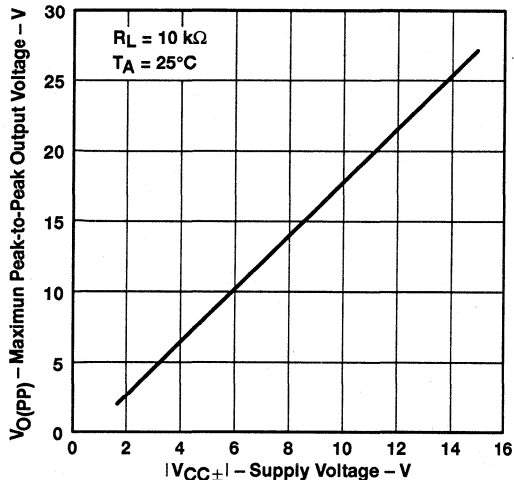


Figure 4

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
vs  
FREQUENCY

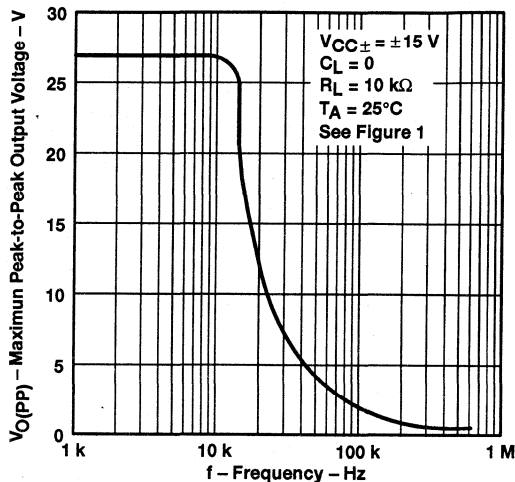


Figure 5

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
FREQUENCY

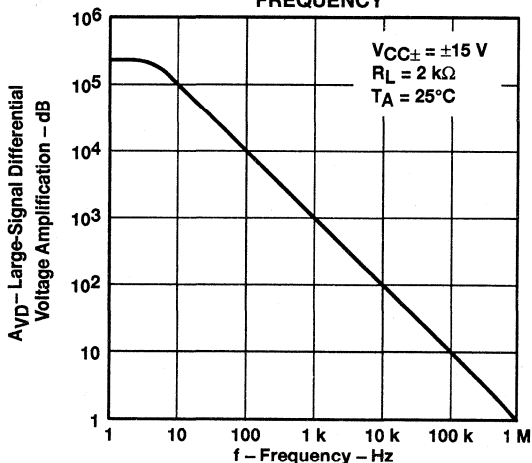


Figure 6

VOLTAGE-FOLLOWER  
LARGE-SIGNAL PULSE RESPONSE

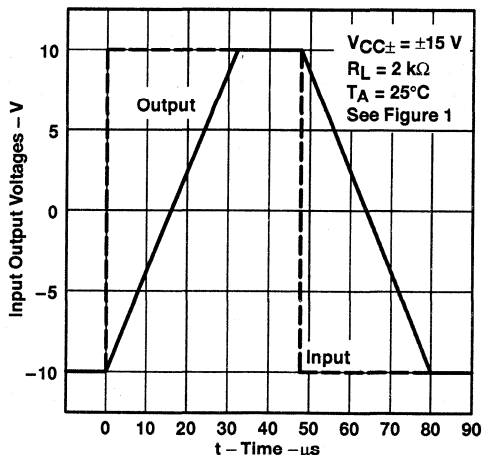


Figure 7

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

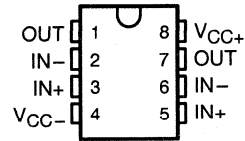


# NE5532, NE5532A, NE5532I, NE5532AI DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS075 – NOVEMBER 1979 – REVISED SEPTEMBER 1990

- Equivalent Input Noise Voltage  
5  $\text{nv}/\sqrt{\text{Hz}}$  Typ at 1 kHz
- Unity-Gain Bandwidth . . . 10 MHz Typ
- Common-Mode Rejection Ratio  
100 dB Typ
- High DC Voltage Gain . . . 100 V/mV Typ
- Peak-to-Peak Output Voltage Swing  
32 V Typ With  $V_{CC\pm} = \pm 18 \text{ V}$  and  
 $R_L = 600 \Omega$
- High Slew Rate . . . 9  $\text{V}/\mu\text{s}$  Typ
- Wide Supply Voltage Range . . .  $\pm 3 \text{ V}$   
to  $\pm 20 \text{ V}$
- Designed to Be Interchangeable With  
Signetics NE5532 and NE5532A

P PACKAGE  
(TOP VIEW)

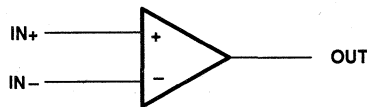


## description

The NE5532 and NE5532A are monolithic high-performance operational amplifiers combining excellent dc and ac characteristics. They feature very low noise, high output drive capability, high unity-gain and maximum-output-swing bandwidths, low distortion, high slew rate, input-protection diodes, and output short-circuit protection. These operational amplifiers are internally compensated for unity-gain operation. The NE5532A has specified maximum limits for equivalent input noise voltage.

The NE5532 and NE5532A are characterized for operation from 0°C to 70°C. The NE5532I and NE5532AI are characterized for operation from -40°C to 85°C.

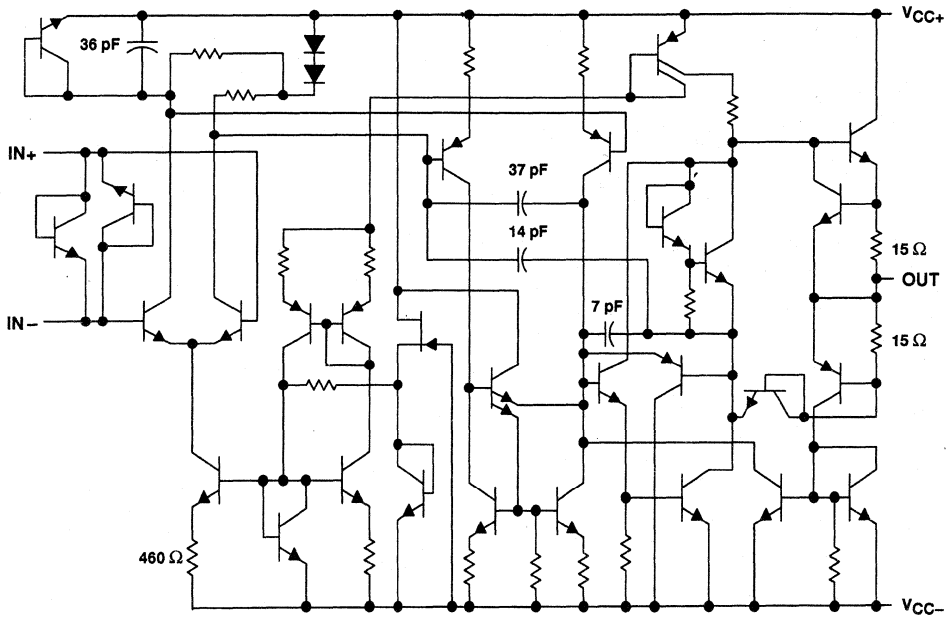
## symbol (each amplifier)



# NE5532, NE5532A, NE5532I, NE5532AI DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS075 – NOVEMBER 1979 – REVISED SEPTEMBER 1990

## schematic (each amplifier)



Component values shown are nominal.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	22 V
Supply voltage, $V_{CC-}$ (see Note 1)	-22 V
Input voltage, either input (see Notes 1 and 2)	$V_{CC\pm}$
Input current (see Note 3)	$\pm 10$ mA
Duration of output short circuit (see Note 4)	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: NE5532, NE5532A	0°C to 70°C
NE5532I, NE5532AI	-40°C to 85°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. The magnitude of the input voltage must never exceed the magnitude of the supply voltage.
3. Excessive input current will flow if a differential input voltage in excess of approximately 0.6 V is applied between the inputs unless some limiting resistance is used.
4. The output may be shorted to ground or either power supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
P	1000 mW	8 mW/°C	640 mW	520 mW

 **TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# NE5532, NE5532A, NE5532I, NE5532AI DUAL LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS075 – NOVEMBER 1979 – REVISED SEPTEMBER 1990

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	5		15	V
Supply voltage, $V_{CC-}$	-5		-15	V

## electrical characteristics, $V_{CC\pm} = \pm 15$ V, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 0$	$T_A = 25^\circ\text{C}$	0.5		4	mV
			$T_A = \text{Full range}$			5	
$I_{IO}$	Input offset current	$T_A = 25^\circ\text{C}$		10		150	nA
		$T_A = \text{Full range}$				200	
$I_{IB}$	Input bias current	$T_A = 25^\circ\text{C}$		200		800	nA
		$T_A = \text{Full range}$				1000	
$V_{ICR}$	Common-mode input voltage range			$\pm 12$	$\pm 13$		V
$V_{OPP}$	Maximum peak-to-peak output voltage swing	$R_L \geq 600 \Omega$	$V_{CC\pm} = \pm 15$ V	24		26	V
			$V_{CC\pm} = \pm 18$ V	30		32	
$A_{VD}$	Large-signal differential voltage amplification	$R_L \geq 600 \Omega$ , $V_O = \pm 10$ V	$T_A = 25^\circ\text{C}$	15		50	V/mV
			$T_A = \text{Full range}$	10			
		$R_L \geq 2 \text{ k}\Omega$ , $V_O = \pm 10$ V	$T_A = 25^\circ\text{C}$	25		100	
			$T_A = \text{Full range}$	15			
$A_{vd}$	Small-signal differential voltage amplification	$f = 10 \text{ kHz}$			2.2		V/mV
$B_{OM}$	Maximum-output-swing bandwidth	$R_L = 600 \Omega$	$V_O = \pm 10$ V			140	kHz
			$V_{CC\pm} = \pm 18$ V, $V_O = \pm 14$ V			100	
$B_1$	Unity-gain bandwidth	$R_L = 600 \Omega$ ,	$C_L = 100 \text{ pF}$			10	MHz
$r_i$	Input resistance			30	300		$\text{k}\Omega$
$z_o$	Output impedance	$A_{VD} = 30 \text{ dB}$ , $R_L = 600 \Omega$ , $f = 10 \text{ kHz}$			0.3		$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR} \text{ min}$		70	100		dB
kSVR	Supply voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC\pm} = \pm 9$ V to $\pm 15$ V, $V_O = 0$		80	100		dB
$I_{OS}$	Output short-circuit current				38		mA
$I_{CC}$	Total supply current	$V_O = 0$ , No load			8	16	mA
	Crosstalk attenuation ( $V_{O1}/V_{O2}$ )	$V_{O1} = 10$ V peak, $f = 1 \text{ kHz}$			110		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for  $T_A$  is  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for NE5532/NE5532A and  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  for NE5532I/NE5532AI.

## operating characteristics, $V_{CC\pm} = \pm 15$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	NE5532/NE5532I			NE5532A/NE5532AI			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	9			9			$\text{V}/\mu\text{s}$
	Overshoot factor	10%			10%			
$V_n$	Equivalent input noise voltage	$V_I = 100 \text{ mV}$ , $A_{VD} = 1$ , $R_L = 600 \Omega$ , $C_L = 100 \text{ pF}$		10%		10%		
		$f = 30 \text{ Hz}$	8		8	10	$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1 \text{ kHz}$	5		5	6			
$I_n$	Equivalent input noise current	$f = 30 \text{ Hz}$	2.7		2.7		$\text{pA}/\sqrt{\text{Hz}}$	
		$f = 1 \text{ kHz}$	0.7		0.7			





# NE5534, NE5534A, SE5534, SE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS070 – JULY 1979 – REVISED SEPTEMBER 1990

- **Equivalent Input Noise Voltage**  
3.5 nV/√Hz
- **Unity-Gain Bandwidth** . . . 10 MHz Typ
- **Common-Mode Rejection Ratio**  
100 dB Typ
- **High DC Voltage Gain** . . . 100 V/mV Typ
- **Peak-to-Peak Output Voltage Swing**  
32 V Typ With  $V_{CC\pm} = \pm 18$  V and  $R_L = 600 \Omega$
- **High Slew Rate** . . . 13 V/μs Typ
- **Wide Supply Voltage Range**  $\pm 3$  V to  $\pm 20$  V
- **Low Harmonic Distortion**
- **Designed to Be Interchangeable With**  
Signetics NE5534, NE5534A, SE5534,  
and SE5534A

## description

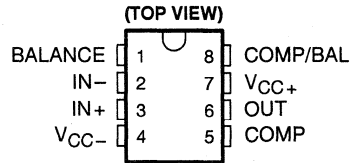
The NE5534, NE5534A, SE5534, and SE5534A are monolithic high-performance operational amplifiers combining excellent dc and ac characteristics. Some of the features include very low noise, high output drive capability, high unity-gain and maximum-output-swing bandwidths, low distortion, and high slew rate.

These operational amplifiers are internally compensated for a gain equal to or greater than three. Optimization of the frequency response for various applications can be obtained by use of an external compensation capacitor between COMP and COMP/BAL. The devices feature input-protection diodes, output short-circuit protection, and offset-voltage nulling capability.

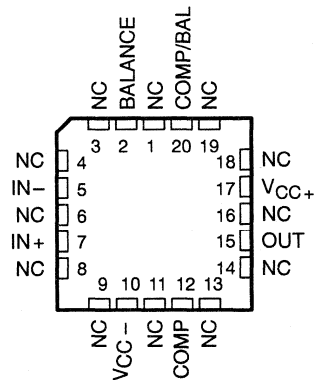
For the NE5534A, a maximum limit is specified for equivalent input noise voltage.

The NE5534 and NE5534A are characterized for operation from 0°C to 70°C. The SE5534 and SE5534A are characterized for operation over the full military temperature range of –55°C to 125°C.

NE5534, NE5534A . . . D OR P PACKAGE  
SE5534, SE5534A . . . JG PACKAGE

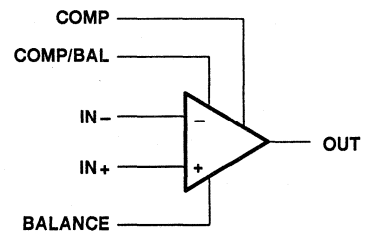


SE5534, SE5534A . . . FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## symbol



**SE5534A FROM TI NOT RECOMMENDED  
FOR NEW DESIGNS**

## AVAILABLE OPTIONS

$T_A$	$V_{IO}$ max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	4 mV	NE5534D NE5534AD	— —	— —	NE5534P NE5534AP
–55°C to 125°C	2 mV	— —	SE5534FK SE5534AFK	SE5534JG SE5534AJG	— —

The D package is available taped and reeled. Add the suffix R to the device type (e.g., NE5534DR).

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

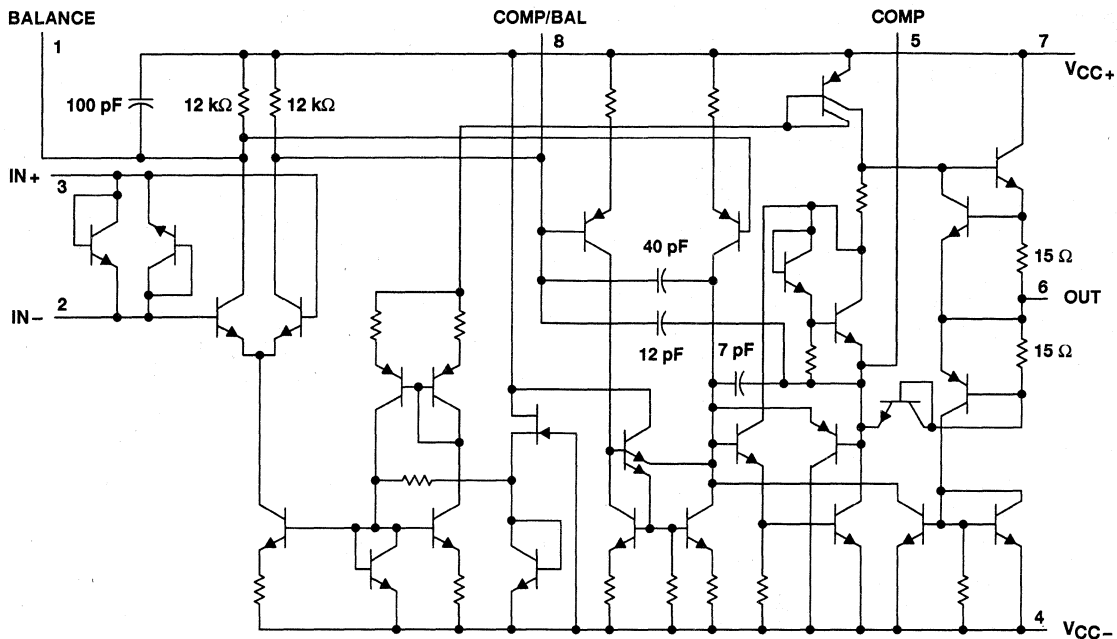
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1990, Texas Instruments Incorporated

# NE5534, NE5534A, SE5534, SE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS070 – JULY 1979 – REVISED SEPTEMBER 1990

## schematic



All component values shown are nominal.

Pin numbers shown are for D, JG, and P packages.

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	22 V
Supply voltage, $V_{CC-}$ (see Note 1)	-22 V
Input voltage either input (see Notes 1 and 2)	$V_{CC+}$
Input current (see Note 3)	$\pm 10$ mA
Duration of output short circuit (see Note 4)	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range: NE5534, NE5534A	0°C to 70°C
SE5534, SE5534A	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature range 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C
Lead temperature range 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C

- NOTES:
- All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  - The magnitude of the input voltage must never exceed the magnitude of the supply voltage.
  - Excessive current will flow if a differential input voltage in excess of approximately 0.6 V is applied between the inputs unless some limiting resistance is used.
  - The output may be shorted to ground or to either power supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

TEXAS  
INSTRUMENTS

# NE5534, NE5534A, SE5534, SE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS070 – JULY 1979 – REVISED SEPTEMBER 1990

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	N/A
FK (see Note 5)	1375 mW	11.0 mW/°C	880 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	N/A

NOTE 5: For the FK package, power rating and derating factor will vary with actual mounting technique used. The values stated here are believed to be conservative.

## recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC+}$	5		15	V
Supply voltage, $V_{CC-}$	-5		-15	V

## electrical characteristics, $V_{CC} \pm = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITION†	NE5534, NE5534A			SE5534, SE5534A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$		0.5	4	$T_A = 25^\circ\text{C}$		mV
		$T_A = \text{Full range}$		5		3		
$I_{IO}$ Input offset current	$V_O = 0$	$T_A = 25^\circ\text{C}$		20	300	$T_A = 25^\circ\text{C}$		nA
		$T_A = \text{Full range}$		400		500		
$I_{IB}$ Input bias current	$V_O = 0$	$T_A = 25^\circ\text{C}$		500	1500	$T_A = 25^\circ\text{C}$		nA
		$T_A = \text{Full range}$		2000		1500		
$V_{ICR}$ Common-mode input voltage range		$\pm 12$	$\pm 13$	$\pm 12$	$\pm 13$		V	
$V_{O(PP)}$ Maximum peak-to-peak output voltage swing	$R_L \geq 600\ \Omega$	$V_{CC\pm} = \pm 15\text{ V}$		24	26	$V_{CC\pm} = \pm 15\text{ V}$		V
		$V_{CC\pm} = \pm 18\text{ V}$		30	32	$V_{CC\pm} = \pm 18\text{ V}$		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$ , $R_L \geq 600\ \Omega$	$T_A = 25^\circ\text{C}$		25	100	$T_A = 25^\circ\text{C}$		V/mV
		$T_A = \text{Full range}$		15		25		
$A_{vd}$ Small-signal differential voltage amplification	$f = 10\text{ kHz}$	$C_C = 0$		6		$C_C = 0$		V/mV
		$C_C = 22\text{ pF}$		2.2		$C_C = 22\text{ pF}$		
$B_{OM}$ Maximum-output-swing bandwidth	$V_O = \pm 10\text{ V}$ , $C_C = 0$			200		200		kHz
	$V_O = \pm 10\text{ V}$ , $C_C = 22\text{ pF}$			95		95		
	$V_{CC\pm} = \pm 18\text{ V}$ , $R_L \geq 600\ \Omega$ , $V_O = \pm 14\text{ V}$ , $C_C = 22\text{ pF}$			70		70		
$B_1$ Unity-gain bandwidth	$C_C = 22\text{ pF}$ , $C_L = 100\text{ pF}$			10		10		MHz
$r_i$ Input resistance		30	100	50	100		k $\Omega$	
$z_o$ Output impedance	$A_{VD} = 30\text{ dB}$ , $C_C = 22\text{ pF}$ , $R_L \geq 600\ \Omega$ , $f = 10\text{ kHz}$			0.3		0.3		$\Omega$
CMRR Common-mode rejection ratio	$V_O = 0$ , $R_S = 50\ \Omega$	$V_{IC} = V_{ICRmin}$		70	100	80	100	dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )	$V_{CC+} = +9\text{ V to } +15\text{ V}$ , $V_O = 0$ , $R_S = 50\ \Omega$			80	100	86	100	dB
$I_{OS}$ Output short-circuit current				38		38		mA
$I_{CC}$ Supply current	$V_O = 0$ , No load	$T_A = 25^\circ\text{C}$		4	8	$T_A = 25^\circ\text{C}$		mA
		$T_A = \text{Full range}$				9		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is  $T_A = 0^\circ\text{C to } 70^\circ\text{C}$  for NE5534 and NE5534A and  $-55^\circ\text{C to } 125^\circ\text{C}$  for SE5534 and SE5534A.



# NE5534, NE5534A, SE5534, SE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS070 – JULY 1979 – REVISED SEPTEMBER 1990

operating characteristics,  $V_{CC} \pm = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	SE5534, NE5534			SE5534A, NE5534A			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$C_C = 0$		13			13		$\text{V}/\mu\text{s}$
	$C_C = 22\text{ pF}$		6			6		
$t_r$ Rise time	$V_I = 50\text{ mV}$ , $A_{VD} = 1$ , $R_L = 600\ \Omega$ , $C_C = 22\text{ pF}$		20			20		ns
Overshoot factor	$C_L = 100\text{ pF}$		20%			20%		
$t_r$ Rise time	$V_I = 50\text{ mV}$ , $A_{VD} = 1$ , $R_L = 600\ \Omega$ , $C_C = 47\text{ pF}$		50			50		ns
Overshoot factor	$C_L = 500\text{ pF}$		35%			35%		
$V_n$ Equivalent input noise voltage	$f = 30\text{ Hz}$		7		5.5	7		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$		4		3.5	4.5		
$I_n$ Equivalent input noise current	$f = 30\text{ Hz}$		2.5		1.5		$\text{pA}/\sqrt{\text{Hz}}$	
	$f = 1\text{ kHz}$		0.6		0.4			
$\bar{F}$ Average noise figure	$R_S = 5\text{ k}\Omega$ , $f = 10\text{ Hz to } 20\text{ kHz}$				0.9			dB

## TYPICAL CHARACTERISTICS†

NORMALIZED INPUT BIAS CURRENT  
AND INPUT OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE

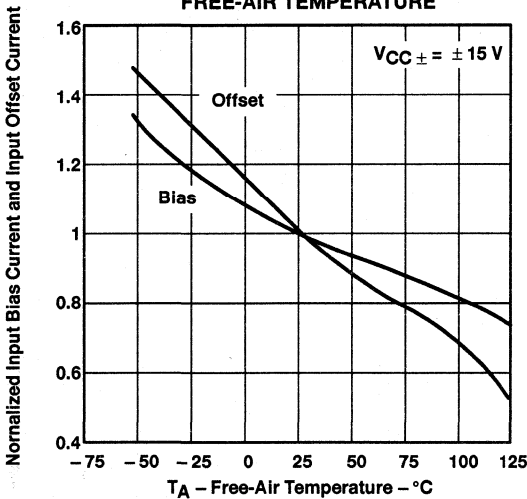


Figure 1

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
vs  
FREQUENCY

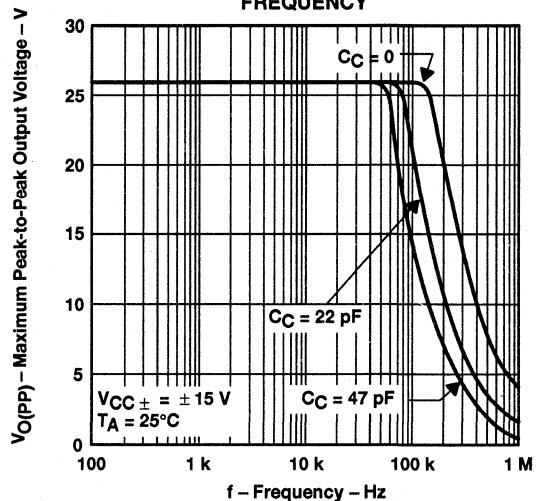


Figure 2

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

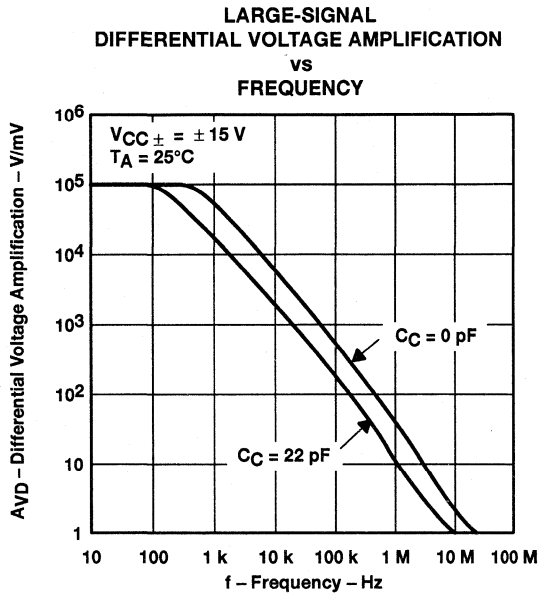


Figure 3

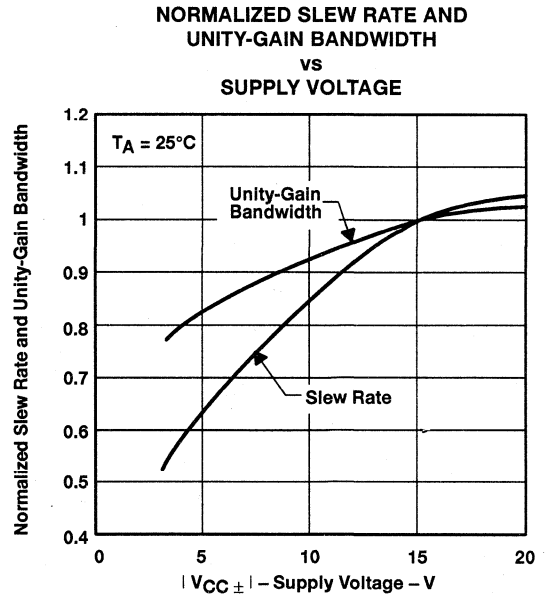


Figure 4

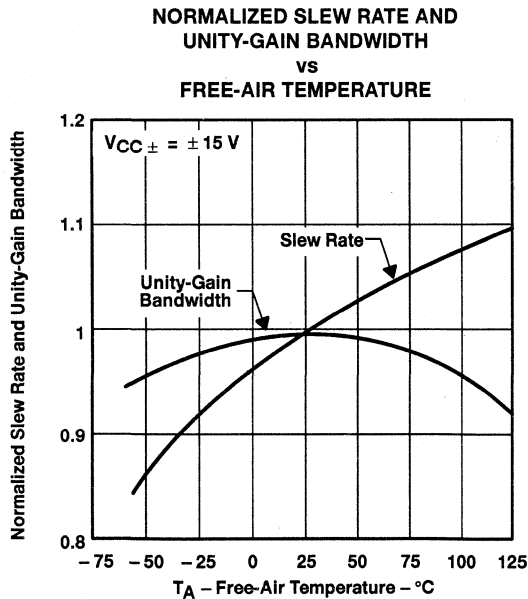


Figure 5

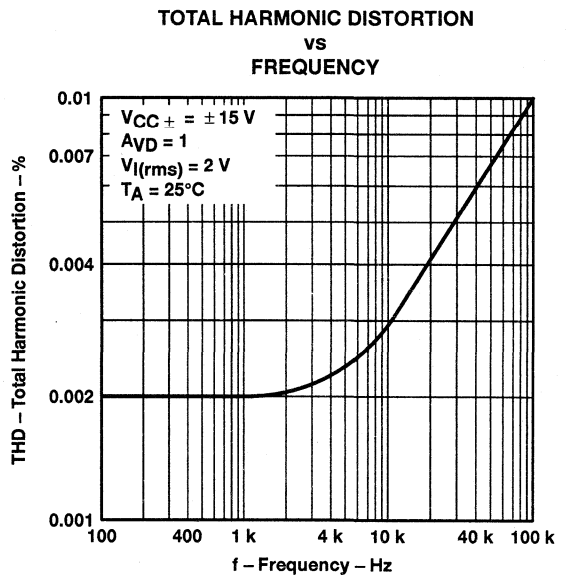


Figure 6

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# NE5534, NE5534A, SE5534, SE5534A LOW-NOISE OPERATIONAL AMPLIFIERS

SLOS070 – JULY 1979 – REVISED SEPTEMBER 1990

## TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE  
vs  
FREQUENCY

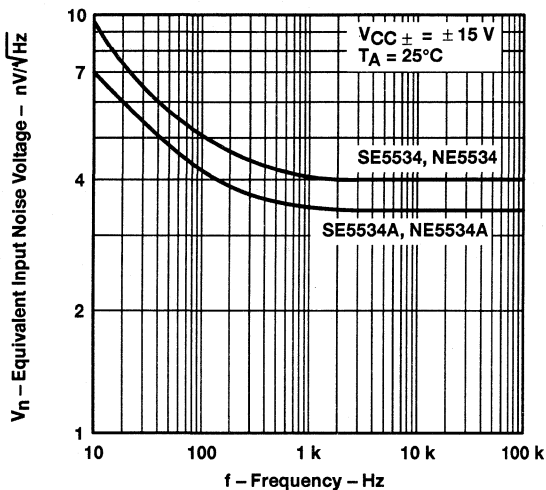


Figure 7

EQUIVALENT INPUT NOISE CURRENT  
vs  
FREQUENCY

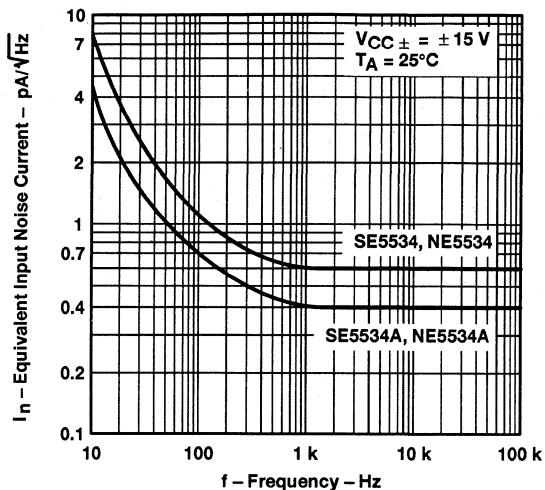


Figure 8

TOTAL EQUIVALENT INPUT NOISE VOLTAGE  
vs  
SOURCE RESISTANCE

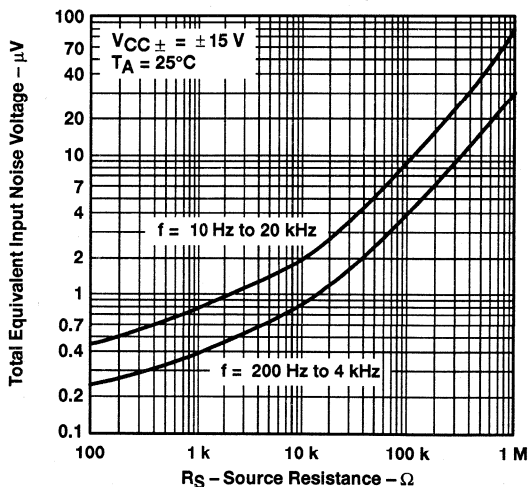


Figure 9

# OP07C, OP07D, OP07Y LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

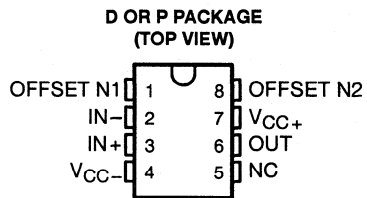
SLOS099A – OCTOBER 1983 – REVISED SEPTEMBER 1991

- Low Noise
- No External Components Required
- Replaces Chopper Amplifiers at a Lower Cost
- Single-Chip Monolithic Fabrication
- Wide Input Voltage Range  
0 to  $\pm 14$  V Typ
- Wide Supply Voltage Range  
 $\pm 3$  V to  $\pm 18$  V
- Essentially Equivalent to Fairchild  $\mu A714$  Operational Amplifiers
- Direct Replacement for PMI OP07C and OP07D

## description

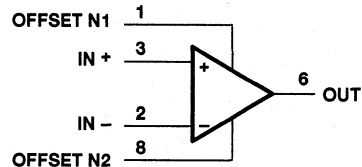
These devices represent a breakthrough in operational amplifier performance. Low offset and long-term stability are achieved by means of a low-noise, chopperless, bipolar-input-transistor amplifier circuit. For most applications, external components are not required for offset nulling and frequency compensation. The true differential input, with a wide input voltage range and outstanding common-mode rejection, provides maximum flexibility and performance in high-noise environments and in noninverting applications. Low bias currents and extremely high input impedances are maintained over the entire temperature range. The OP07 is unsurpassed for low-noise, high-accuracy amplification of very low-level signals.

These devices are characterized for operation from 0°C to 70°C.



NC—No internal connection

## symbol



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
0°C to 70°C	150 $\mu$ V	OP07CD QP07DD	OP07CP OP07DP	OP07Y

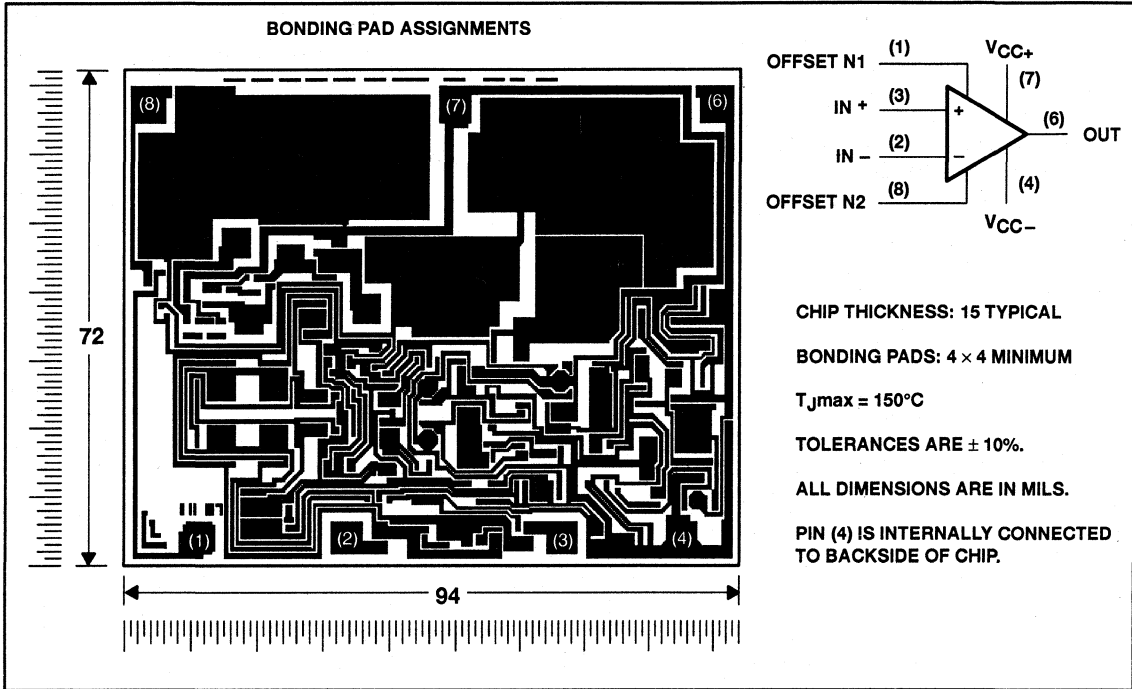
The D package is available taped and reeled. Add the suffix R to the device type (e.g., OP07CDR). The chip form is tested at T<sub>A</sub> = 25°C.

# OP07C, OP07D, OP07Y LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

SLOS099A – OCTOBER 1983 – REVISED SEPTEMBER 1991

## OP07Y chip information

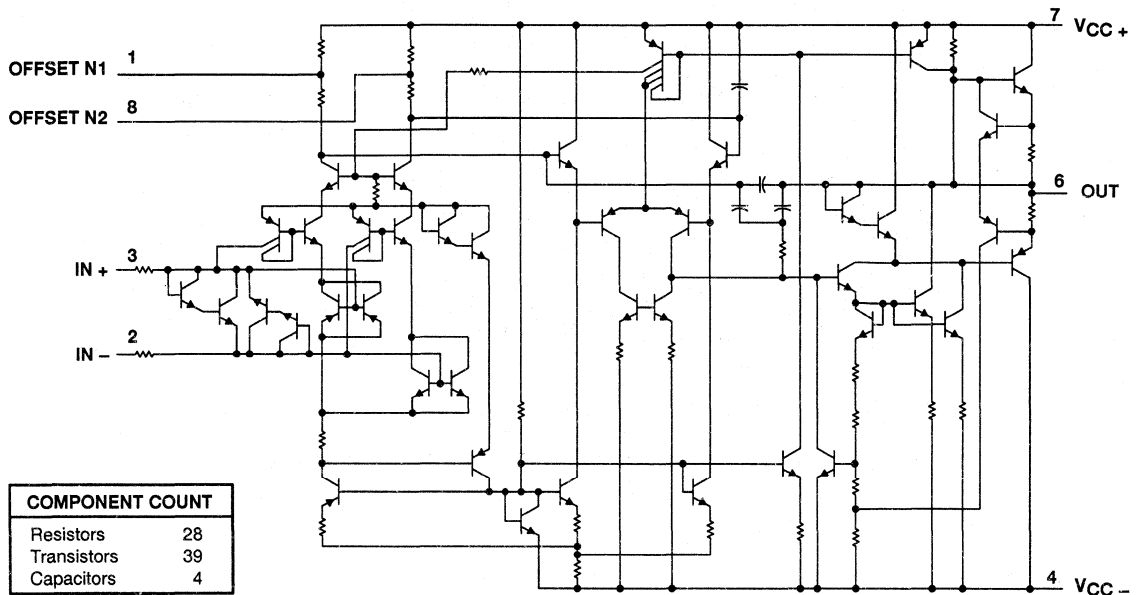
These chips, properly assembled, display characteristics similar to the OP07. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# OP07C, OP07D, OP07E LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

SLOS099A – OCTOBER 1983 – REVISED SEPTEMBER 1991

## schematic



COMPONENT COUNT	
Resistors	28
Transistors	39
Capacitors	4

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, $V_{CC+}$ (see Note 1)	22 V
Supply voltage, $V_{CC-}$	-22 V
Differential input voltage (see Note 2)	$\pm 30$ V
Input voltage, $V_I$ (either input, see Note 3)	$\pm 22$ V
Duration of output short circuit (see Note 4)	unlimited
Continuous total dissipation at (or below) 25°C free-air temperature (see Note 5)	500 mW
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to ground or either power supply.
5. For operation above 64°C free-air temperature, derate the D package to 464 mW at 70°C at the rate of 5.8 mW/°C.

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC\pm}$	$\pm 3$	$\pm 18$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} \pm 15$ V		V
Operating free-air temperature, $T_A$	0	70	°C

# OP07C, OP07D, OP07Y LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

SLOS099A – OCTOBER 1993 – REVISED SEPTEMBER 1991

electrical characteristics at specified free-air temperature,  $V_{CC} \pm = \pm 15 \text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	T <sub>A</sub>	OP07C			OP07D			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	60	150	60	150	60	150	μV
α <sub>VIO</sub> Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	0°C to 70°C	85	250	85	250	85	250	μV/°C
		0°C to 70°C	0.5	1.8	0.5	1.8	0.7	2.5	μV/°C
Long-term drift of input offset voltage	See Note 6		0.4		0.4		0.5		μV/mo
Offset adjustment range	R <sub>S</sub> = 20 kΩ, See Figure 1	25°C	±4		±4		±4		mV
I <sub>IO</sub> Input offset current		25°C	0.8	6	0.8	6	0.8	6	nA
α <sub>IIO</sub> Temperature coefficient of input offset current		0°C to 70°C	1.6	8	1.6	8	1.6	8	nA
		0°C to 70°C	12	50	12	50	12	50	pA/°C
I <sub>IB</sub> Input bias current		25°C	±1.8	±7	±2	±12	±2	±12	nA
α <sub>IIB</sub> Temperature coefficient of input bias current		0°C to 70°C	±2.2	±9	±3	±14	±3	±14	nA
		0°C to 70°C	18	50	18	50	18	50	pA/°C
V <sub>ICR</sub> Common-mode input voltage range		25°C	±13	±14	±13	±14	±13	±14	V
V <sub>OM</sub> Peak output voltage	R <sub>L</sub> ≥ 10 kΩ	0°C to 70°C	±13	±13.5	±13	±13.5	±13	±13.5	V
		25°C	±12	±13	±12	±13	±12	±13	V
		0°C to 70°C	±11.5	±12.8	±11.5	±12.8	±11.5	±12.8	V
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>O</sub> = ±10 V, R <sub>L</sub> = 2 kΩ	0°C to 70°C	±12		±12		±12		V/mV
		25°C	±11	±12.6	±11	±12.6	±11	±12.6	V/mV
		0°C to 70°C	100	400	100	400	100	400	V/mV
B <sub>1</sub> Unity-gain bandwidth		25°C	120	400	120	400	120	400	MHz
r <sub>i</sub> Input resistance		25°C	0.4	0.6	0.4	0.6	0.4	0.6	MΩ
CMRR Common-mode rejection ratio	V <sub>IC</sub> = ±13 V, R <sub>S</sub> = 50 Ω	25°C	8	33	7	31	7	31	dB
		0°C to 70°C	100	120	94	110	94	106	dB
KSVS Supply voltage sensitivity (ΔV <sub>IO</sub> /ΔV <sub>CC</sub> )	V <sub>O</sub> = ±10 V, R <sub>L</sub> = 2 kΩ	25°C	7	32	7	32	7	32	μV/V
		0°C to 70°C	10	51	10	51	10	51	μV/V
P <sub>D</sub> Power dissipation	V <sub>O</sub> = 0, No load	25°C	80	150	80	150	80	150	mW
		0°C to 70°C	4	8	4	8	4	8	mW

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise noted.  
NOTE 6: Since long-term drift cannot be measured on the individual devices prior to shipment, this specification is not intended to be a warranty. It is an engineering estimate of the averaged trend line of drift versus time over extended periods after the first thirty days of operation.



# OP07C, OP07D, OP07Y LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

SLOS099A – OCTOBER 1983 – REVISED SEPTEMBER 1991

## operating characteristics, $V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS†	OP07C			OP07D			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$	10.5			10.5			$nV/\sqrt{\text{Hz}}$
	$f = 100\text{ Hz}$	10.2			10.3			
	$f = 1\text{ kHz}$	9.8			9.8			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }10\text{ Hz}$	0.38			0.38			$\mu\text{V}$
$I_n$ Equivalent input noise current	$f = 10\text{ Hz}$	0.35			0.35			$pA/\sqrt{\text{Hz}}$
	$f = 100\text{ Hz}$	0.15			0.15			
	$f = 1\text{ kHz}$	0.13			0.13			
$I_{N(PP)}$ Peak-to-peak equivalent input noise current	$f = 0.1\text{ Hz to }10\text{ Hz}$	15			15			$pA$
SR Slew rate	$R_L \geq 2\text{ k}\Omega$	0.3			0.3			$V/\mu\text{s}$

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise noted.

## electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	OP07Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$R_S = 50\ \Omega$	60 150			$\mu\text{V}$
Long-term drift of input offset voltage	See Note 6	0.5			$\mu\text{V}/\text{mo}$
Offset adjustment range	$R_S = 20\text{ k}\Omega$ , See Figure 1	$\pm 4$			$\text{mV}$
$I_{IO}$ Input offset current		0.8 6			$\text{nA}$
$I_{IB}$ Input bias current		$\pm 2$ $\pm 12$			$\text{nA}$
$V_{ICR}$ Common-mode input voltage range		$\pm 13$ $\pm 14$			$\text{V}$
$V_{OM}$ Peak output voltage	$R_L \leq 10\text{ k}\Omega$	$\pm 12$ $\pm 13$			$\text{V}$
	$R_L \leq 2\text{ k}\Omega$	$\pm 11.5$ $\pm 12.8$			
	$R_L \leq 1\text{ k}\Omega$	$\pm 12$			
$A_{VD}$ Large-signal differential voltage amplification	$V_{CC\pm} = \pm 3\text{ V}$ , $V_O = \pm 0.5\text{ V}$ , $R_L \leq 500\text{ k}\Omega$	400			
	$V_O = \pm 10\text{ V}$ , $R_L = 2\text{ k}\Omega$	120 400			
$B_1$ Unity-gain bandwidth		0.4 0.6			$\text{MHz}$
$r_i$ Input resistance		7 31			$\text{M}\Omega$
CMRR Common-mode input resistance	$V_{IC} = \pm 13\text{ V}$ , $R_S = 50\ \Omega$	94 110			$\text{dB}$
$k_{SVS}$ Supply-voltage rejection ratio ( $\Delta V_{CC}/\Delta V_{IO}$ )	$V_{CC\pm} = \pm 3\text{ V to } \pm 18\text{ V}$ , $R_S = 50\ \Omega$	7 32			$\mu\text{V}/\text{V}$
$P_D$ Power dissipation	$V_O = 0$ , No load	80 150			$\text{M}\Omega$
	$V_{CC\pm} = \pm 3\text{ V}$ , $V_O = 0$ , No load	4 8			

NOTE 6: Since long-term drift cannot be measured on the individual devices prior to shipment, this specification is not intended to be a warranty. It is an engineering estimate of the averaged trend line of drift versus time over extended periods after the first thirty days of operation.

# OP07C, OP07D, OP07Y LOW-OFFSET VOLTAGE OPERATIONAL AMPLIFIERS

SLOS099A – OCTOBER 1983 – REVISED SEPTEMBER 1991

operating characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONST	OP07Y			UNIT
		MIN	TYP	MAX	
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$		10.5		nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$		10.3		
	$f = 0.1\text{ Hz to }10\text{ Hz}$		9.8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }10\text{ Hz}$		0.38		$\mu\text{V}$
$I_n$ Equivalent input noise current	$f = 10\text{ Hz}$		0.35		pA/ $\sqrt{\text{Hz}}$
	$f = 100\text{ Hz}$		0.15		
	$f = 1\text{ kHz}$		0.13		
$I_{N(PP)}$ Peak-to-peak equivalent input noise current	$f = 0.1\text{ Hz to }10\text{ Hz}$		15		pA
SR Slew rate	$R_L = 2\text{ k}\Omega$		0.3		V/ $\mu\text{s}$

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise noted.

## APPLICATION INFORMATION

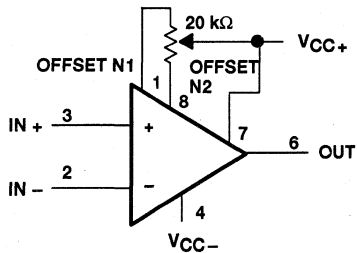


Figure 1. Input Offset Voltage Null Circuit



# RC4136, RM4136, RV4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS072 – MARCH 1978 – REVISED SEPTEMBER 1990

- **Continuous-Short-Circuit Protection**
- **Wide Common-Mode and Differential Voltage Ranges**
- **No Frequency Compensation Required**
- **Low Power Consumption**
- **No Latch-Up**
- **Unity Gain Bandwidth . . . 3 MHz Typ**
- **Gain and Phase Match Between Amplifiers**
- **Designed To Be Interchangeable With Raytheon RC4136, RM4136, and RV4136**
- **Low Noise . . . 8 nV $\sqrt{\text{Hz}}$  Typ at 1 kHz**

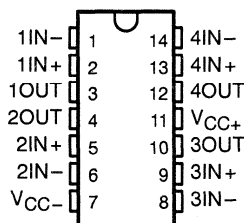
## description

The RC4136, RM4136, and RV4136 are quad high-performance operational amplifiers with each amplifier electrically similar to the  $\mu\text{A}741$  except that offset null capability is not provided.

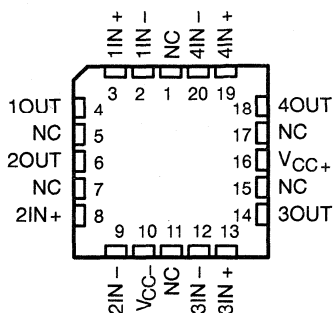
The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short circuit protected and the internal frequency compensation ensures stability without external components.

The RC4136 is characterized for operation from 0°C to 70°C, the RM4136 is characterized for operation over the full military temperature range of -55°C to 125°C, and the RV4136 is characterized for operation from -40°C to 85°C.

**RM4136 . . . J OR W PACKAGE  
ALL OTHERS . . . D OR N PACKAGE  
(TOP VIEW)**

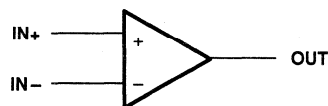


**RM4136 . . . FK PACKAGE  
(TOP VIEW)**



NC – No internal connection

## symbol (each amplifier)



## AVAILABLE OPTIONS

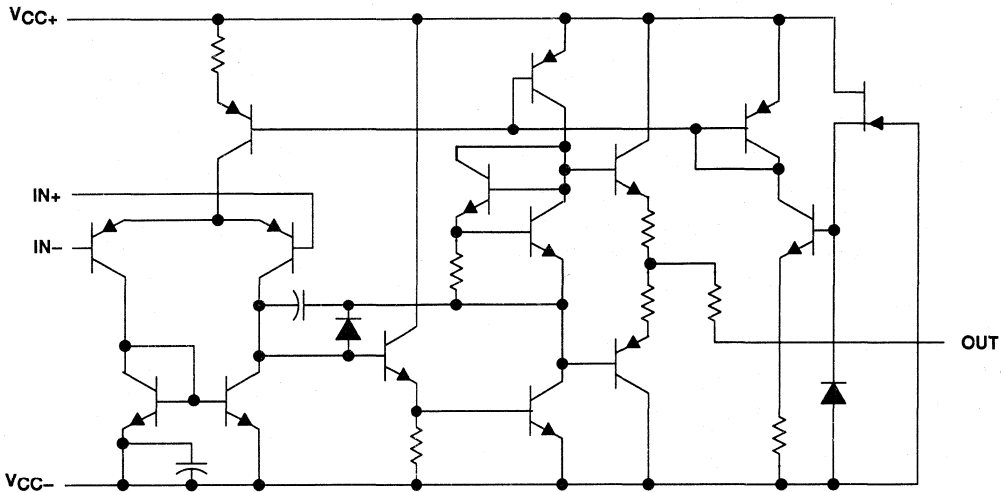
T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	FLAT (W)
0°C to 70°C	6 mV	RC4136D	—	—	RC4136N	—
-40°C to 85°C	6 mV	RV4136D	—	—	RV4136N	—
-55°C to 125°C	4 mV	—	RM4136FK	RM4136J	—	RM4136W

The D packages are available taped and reeled. Add the suffix R to the device type (e.g., RC4136DR).

# RC4136, RM4136, RV4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS072 – MARCH 1978 – REVISED SEPTEMBER 1990

## schematic (each amplifier)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	RC4136	RM4136	RV4136	UNIT
Supply voltage $V_{CC+}$ (see Note 1)	18	22	18	V
Supply voltage $V_{CC-}$ (see Note 1)	-18	-22	-18	V
Differential input voltage (see Note 2)	$\pm 30$	$\pm 30$	$\pm 30$	V
Input voltage (any input, see Notes 1 and 3)	$\pm 15$	$\pm 15$	$\pm 15$	V
Duration of output short circuit to ground, one amplifier at a time (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	0 to 70	-55 to 125	-40 to 85	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds		FK package	—	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds		J or W package	—	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds		D or N package	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at IN+ with respect to IN-.  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
 4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	800 mW	7.6 mW/ $^{\circ}\text{C}$	45 $^{\circ}\text{C}$	608 mW	494 mW	—
FK	800 mW	11.0 mW/ $^{\circ}\text{C}$	77 $^{\circ}\text{C}$	800 mW	715 mW	275 mW
J	800 mW	11.0 mW/ $^{\circ}\text{C}$	77 $^{\circ}\text{C}$	800 mW	715 mW	275 mW
N	800 mW	9.2 mW/ $^{\circ}\text{C}$	63 $^{\circ}\text{C}$	736 mW	598 mW	—
W	800 mW	8.0 mW/ $^{\circ}\text{C}$	50 $^{\circ}\text{C}$	640 mW	520 mW	200 mW



# RC4136, RM4136, RV4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS072 – MARCH 1978 – REVISED SEPTEMBER 1990

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	5	15	V
Supply voltage, $V_{CC-}$	-5	-15	V

## electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$ , $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	RC4136			RM4136			RV4136			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IL}$ Input offset voltage	$V_O = 0$	25°C	0.5	6	0.5	4	0.5	6	mV		
		Full range	7.5		6		7.5				
$I_{IO}$ Input offset current	$V_O = 0$	25°C	5	200	5	1.50	5	200	nA		
		Full range	300		500		500				
$I_{IB}$ Input bias current	$V_O = 0$	25°C	140	500	140	400	140	500	nA		
		Full range	800		1500		1500				
$V_i$ Input voltage range		25°C	±12	±14	±12	±14	±12	±14	V		
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14	±12	±14	±12	±14	V		
	$R_L = 2\text{ k}\Omega$	25°C	±10	±13	±10	±13	±10	±13			
	$R_L \geq 2\text{ k}\Omega$	Full range	±10		±10		±10				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\text{ V}$ , $R_L \geq 2\text{ k}\Omega$	25°C	20	300	50	350	20	300	V/mV		
		Full range	15		25		15				
$B_1$ Unity-gain bandwidth		25°C	3		3.5		3		MHz		
$r_i$ Input resistance		25°C	0.3*	5	0.3*	5	0.3*	5	M $\Omega$		
CMRR Common-mode rejection ratio	$V_O = 0$ , $R_S = 50\text{ }\Omega$	25°C	70	90	70	90	70	90	dB		
$k_{SVS}$ Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC} = \pm 9\text{ V}$ to $\pm 15\text{ V}$ , $V_O = 0$	25°C	30	150	30	150	30	150	$\mu\text{V/V}$		
$V_n$ Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$ , $BW = 1\text{ Hz}$ , $f = 1\text{ kHz}$ , $R_S = 100\text{ }\Omega$	25°C	8		8		8		nV/ $\sqrt{\text{Hz}}$		
$I_{CC}$ Supply current (all four amplifiers)	$V_O = 0$ , No load	25°C	5	11.3	5	11.3	5	11.3	mA		
		MIN $T_A$	6	13.7	6	13.3	6	13.7			
		MAX $T_A$	4.5	10	4.5	10	4.5	10			
$P_D$ Total power dissipation (all four amplifiers)	$V_O = 0$ , No load	25°C	150	340	150	340	150	340	mW		
		MIN $T_A$	180	400	180	400	180	400			
		MAX $T_A$	135	300	135	300	135	300			
Crosstalk attenuation ( $V_{O1}/V_{O2}$ )	$A_{VD} = 100$ , $f = 10\text{ kHz}$ , $R_S = 1\text{ k}\Omega$	25°C	105		105		105		dB		

\* This parameter is not production tested.

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is 0°C to 70°C for RC4136, -55°C to 125°C for RM4136, and -40°C to 85°C for RV4136. Minimum  $T_A$  is 0°C for RC4136, -55°C for RM4136, and -40°C for RV4136. Maximum  $T_A$  is 70°C for RC4136, 125°C for RM4136, and 85°C for RV4136.



# RC4136, RM4136, RV4136 QUAD HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS072 – MARCH 1978 – REVISED SEPTEMBER 1990

operating characteristics,  $V_{CC+} = 15\text{ V}$ ,  $V_{CC-} = -15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	RC4136, RV4136			RM4136			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$t_r$ Rise time	$V_I = 20\text{ mV}$ , $R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$	0.13			0.13			$\mu\text{s}$
Overshoot factor		5%			5%			
SR Slew rate at unity gain	$V_I = 10\text{ V}$ , $R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$	1.7			1.7			$\text{V}/\mu\text{s}$

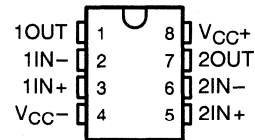


# RC4558, RC4558Y, RM4558, RV4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS073 – MARCH 1976 – REVISED AUGUST 1991

- Continuous-Short-Circuit Protection
- Wide Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- Low Power Consumption
- No Latch-Up
- Unity Gain Bandwidth . . . 3 MHz Typ
- Gain and Phase Match Between Amplifiers
- Low Noise . . . 8 nV/√Hz Typ at 1 kHz
- Designed To Be Interchangeable With Raytheon RC4558, RM4558, and RV4558

D, DB, JG, P, OR PW PACKAGE  
(TOP VIEW)



## description

The RC4558, RM4558, and RV4558 are dual high-performance operational amplifiers with each half electrically similar to the  $\mu$ A741 except that offset null capability is not provided.

The high common-mode input voltage range and the absence of latch-up make these amplifiers ideal for voltage-follower applications. The devices are short-circuit protected and the internal frequency compensation ensures stability without external components.

The RC4558 is characterized for operation from 0°C to 70°C, the RM4558 is characterized for operation over the full military temperature range of -55°C to 125°C, and the RV4558 is characterized for operation from -40°C to 85°C.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	SSOP (DBLE)	CERAMIC DIP (JG)	PLASTIC DIP (P)	SSOP (PWLE)	
0°C to 70°C	6 mV	RC4558D	RC4558DBLE	—	RC4558P	RC4558PWLE	RC4558Y
-40°C to 85°C	6 mV	RV4558D	—	—	RV4558P	—	—
-55°C to 125°C	6 mV	—	—	RM4558JG	—	—	—

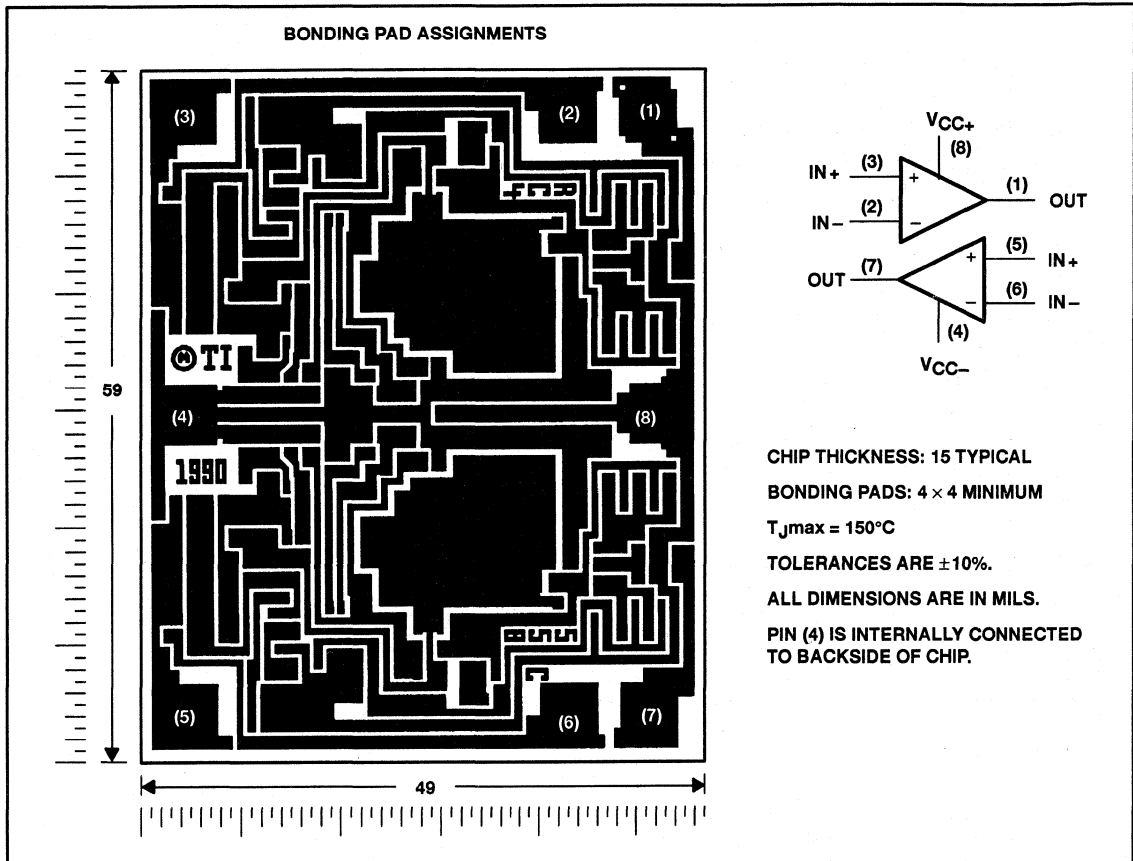
The D package is available taped and reeled. Add the suffix R to the device type (e.g., RC4558DR). The DB and PW packages are available only left-end taped and reeled. RC4558Y is tested at 25°C.

# RC4558, RC4558Y, RM4558, RV4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS073 – MARCH 1976 – REVISED AUGUST 1991

## RC4558Y chip information

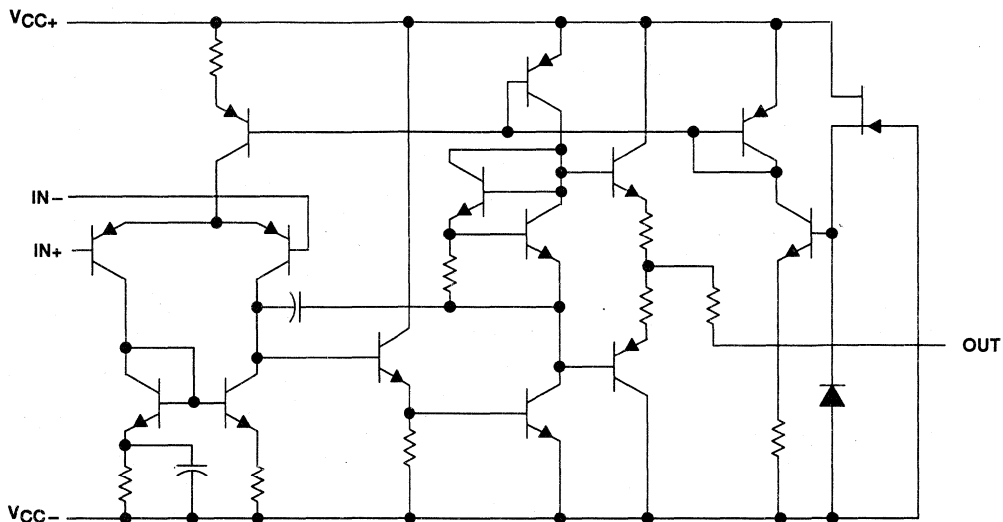
These chips, properly assembled, display characteristics similar to the RC4558. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# RC4558, RC4558Y, RM4558, RV4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS073 – MARCH 1976 – REVISED AUGUST 1991

## schematic (each amplifier)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	RC4558	RM4558	RV4558	UNIT
Supply voltage $V_{CC+}$ (see Note 1)	18	22	18	V
Supply voltage $V_{CC-}$ (see Note 1)	-18	-22	-18	V
Differential input voltage (see Note 2)	$\pm 30$	$\pm 30$	$\pm 30$	V
Input voltage (any input, see Notes 1 and 3)	$\pm 15$	$\pm 15$	$\pm 15$	V
Duration of output short circuit to ground, one amplifier at a time (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	0 to 70	-55 to 125	-40 to 85	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package		300		$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, DB, P, or PW package	260		260	$^{\circ}\text{C}$

NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .

2. Differential voltages are at  $\text{IN+}$  with respect to  $\text{IN-}$ .

3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.

4. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$	DERATING FACTOR ABOVE $T_A = 25^{\circ}\text{C}$	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$	$T_A = 85^{\circ}\text{C}$	$T_A = 125^{\circ}\text{C}$
	POWER RATING			POWER RATING	POWER RATING	POWER RATING
D	680 mW	5.8 mW/ $^{\circ}\text{C}$	33 $^{\circ}\text{C}$	464 mW	377 mW	N/A
DB or PW	525 mW	4.2 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	336 mW	N/A	N/A
JG	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	546 mW	210 mW
P	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	520 mW	N/A

# RC4558, RC4558Y, RM4558, RV4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS073 – MARCH 1976 – REVISED AUGUST 1991

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	5	15	V
Supply voltage, $V_{CC-}$	-5	-15	V

## electrical characteristics at specified free-air temperature, $V_{CC+} = 15\text{ V}$ , $V_{CC-} = -15\text{ V}$

PARAMETER	TEST CONDITIONS†	RC4558			RM4558			RV4558			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$	25°C	0.5	6	0.5	5	0.5	6	mV		
		Full range	7.5			6			7.5		
$I_{IO}$ Input offset current	$V_O = 0$	25°C	5	200	5	200	5	200	nA		
		Full range	300			500			500		
$I_{IB}$ Input bias current	$V_O = 0$	25°C	150	500	140	500	140	500	nA		
		Full range	800			1500			1500		
$V_{ICR}$ Common-mode input voltage range		25°C	±12	±14	±12	±14	±12	±14	V		
$V_{OM}$ Maximum output voltage swing	$R_L = 10\text{ k}\Omega$	25°C	±12	±14	±12	±14	±12	±14	V		
		$R_L = 2\text{ k}\Omega$	25°C	±10	±13	±10	±13	±10		±13	
		$R_L \geq 2\text{ k}\Omega$	Full range	±10			±10				
$A_{VD}$ Large-signal differential voltage amplification	$R_L \geq 2\text{ k}\Omega$ , $V_O = \pm 10\text{ V}$	25°C	20	300	50	350	20	300	V/mV		
		Full range	15			25					
$B_1$ Unity-gain bandwidth		25°C	3		2	3.5	3		MHz		
$r_i$ Input resistance		25°C	0.3	5	0.3	5	0.3	5	M $\Omega$		
CMRR Common-mode rejection ratio		25°C	70	90	70	90	70	90	dB		
$k_{svs}$ Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC} = \pm 15\text{ V}$ to $\pm 9\text{ V}$	25°C	30	150	30	150	30	150	$\mu\text{V/V}$		
$V_n$ Equivalent input noise voltage (closed loop)	$A_{VD} = 100$ , $R_S = 100\ \Omega$ , $f = 1\text{ kHz}$ , $BW = 1\text{ Hz}$	25°C	8		8		8		nV/ $\sqrt{\text{Hz}}$		
$I_{CC}$ Supply current (both amplifiers)	$V_O = 0$ , No load	25°C	2.5	5.6	2.5	5.6	2.5	5.6	mA		
		MIN $T_A$	3		3		3				
		MAX $T_A$	2.3		2		2.3				
$P_D$ Total power dissipation (both amplifiers)	$V_O = 0$ , No load	25°C	75	170	75	170	75	170	mW		
		MIN $T_A$	90		90		90				
		MAX $T_A$	70		60		70				
$V_{O1}/V_{O2}$ Crosstalk attenuation	Open loop	$R_S = 1\text{ k}\Omega$ , $f = 10\text{ kHz}$	25°C		85		85		dB		
	$A_{VD} = 100$		105		105		105				

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range is 0°C to 70°C for RC4558, -55°C to 125°C for RM4558, and -40°C to 85°C for RV4558. Minimum  $T_A$  is 0°C for RC4558, -55°C for RM4558, and -40°C for RV4558. Maximum  $T_A$  is 70°C for RC4558, 125°C for RM4558, and 85°C for RV4558.

## operating characteristics, $V_{CC+} = 15\text{ V}$ , $V_{CC-} = -15\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS			MIN	TYP	MAX	UNIT
$t_r$ Rise time	$V_I = 20\text{ mV}$ , $R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$			0.13		ns	
Overshoot				5%			
SR Slew rate at unity gain	$V_I = 10\text{ V}$ , $R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$			1.1	1.7	V/ $\mu\text{s}$	





# RC4558, RC4558Y, RM4558, RV4558 DUAL HIGH-PERFORMANCE OPERATIONAL AMPLIFIERS

SLOS073 – MARCH 1976 – REVISED AUGUST 1991

**electrical characteristics,  $V_{CC+} = 15\text{ V}$ ,  $V_{CC-} = -15\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONST	RC4558Y			UNIT
			MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 0$		0.5	6	mV
$I_{IO}$	Input offset current	$V_O = 0$		5	200	nA
$I_{IB}$	Input bias current	$V_O = 0$		150	500	nA
$V_{ICR}$	Common-mode input voltage range		$\pm 12$	$\pm 14$		V
$V_{OM}$	Maximum output voltage swing	$R_L = 10\text{ k}\Omega$	$\pm 12$	$\pm 14$		V
		$R_L = 2\text{ k}\Omega$	$\pm 12$	$\pm 13$		
$A_{VD}$	Large-signal differential voltage amplification	$R_L = 2\text{ k}\Omega$ , $V_O = \pm 10\text{ V}$	20	300		V/mV
$B_1$	Unity-gain bandwidth			3		MHz
$r_i$	Input resistance		0.3	5		M $\Omega$
CMRR	Common-mode rejection ratio		70	90		dB
$k_{SVS}$	Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC} = \pm 15\text{ V}$ to $\pm 9\text{ V}$		30	150	$\mu\text{V/V}$
$V_n$	Equivalent input noise voltage (closed-loop)	$A_{VD} = 100$ , $R_S = 100\ \Omega$ , $f = 1\text{ kHz}$ , $BW = 1\text{ Hz}$		8		$\text{nV}/\sqrt{\text{Hz}}$
$I_{CC}$	Supply current (both amplifiers)	$V_O = 0$ , No load		2.5	5.6	mA
$P_D$	Total power dissipation (both amplifiers)	$V_O = 0$ , No load		75	170	mW
$V_{O1}/V_{O2}$	Crosstalk attenuation	Open loop		85		dB
		$A_{VD} = 100$	$R_S = 1\text{ k}\Omega$ , $f = 10\text{ kHz}$	105		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

**operating characteristics,  $V_{CC+} = 15\text{ V}$ ,  $V_{CC-} = -15\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
$t_r$	Rise time	$V_I = 20\text{ mV}$ ,	$R_L = 2\text{ k}\Omega$ ,	$C_L = 100\text{ pF}$		0.13		ns
	Overshoot					5%		
SR	Slew rate at unity gain	$V_I = 10\text{ V}$ ,	$R_L = 2\text{ k}\Omega$ ,	$C_L = 100\text{ pF}$	1.1	1.7		V/ $\mu\text{s}$





# TL022C, TL022M DUAL LOW-POWER OPERATIONAL AMPLIFIERS

SLOS076 – SEPTEMBER 1973 – REVISED SEPTEMBER 1990

- Very Low Power Consumption
- Power Dissipation With  $\pm 2$ -V Supplies  
170  $\mu$ W Typ
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Input Offset Voltage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- Popular Dual Operational Amplifier Pinout

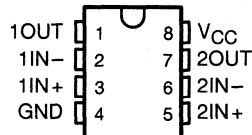
**TL022M IS NOT RECOMMENDED FOR  
NEW DESIGNS**

## description

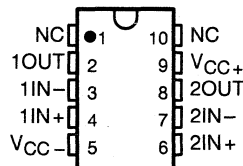
The TL022 is a dual low-power operational amplifier designed to replace higher power devices in many applications without sacrificing system performance. High input impedance, low supply currents, and low equivalent input noise voltage over a wide range of operating supply voltages result in an extremely versatile operational amplifier for use in a variety of analog applications including battery-operated circuits. Internal frequency compensation, absence of latch-up, high slew rate, and output short-circuit protection assure ease of use.

The TL022C is characterized for operation from 0°C to 70°C. The TL022M is characterized for operation over the full military temperature range of -55°C to 125°C.

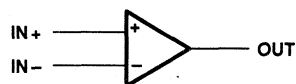
TL022M ... JG PACKAGE  
TL022C ... D OR P PACKAGE  
(TOP VIEW)



TL022M ... U PACKAGE  
(TOP VIEW)



## symbol (each amplifier)



## AVAILABLE OPTIONS

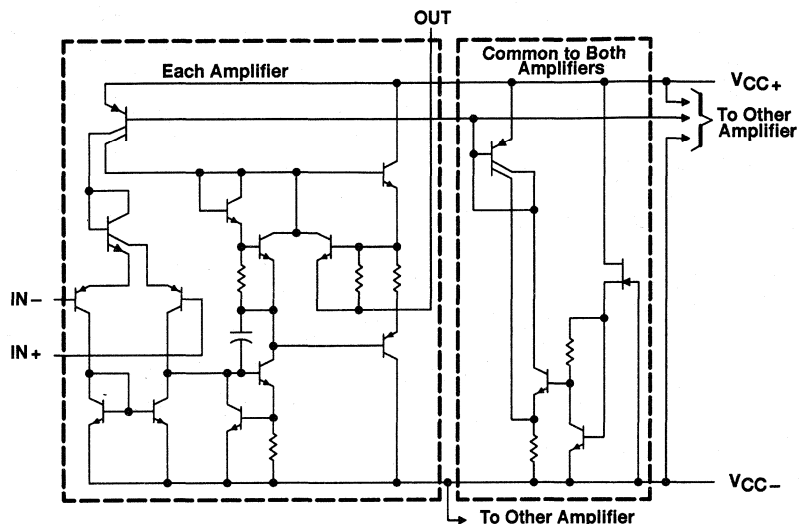
T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)	CERAMIC FLAT PACK (U)
0°C to 70°C	5 mV	TL022CD	—	TL022CP	—
-55°C to 125°C	5 mV	—	TL022MJG	—	TL022MU

The D package is available taped and reeled. Add the suffix R to the device type (i.e. TL022CDR).

# TL022C, TL022M DUAL LOW-POWER OPERATIONAL AMPLIFIERS

SLOS076 – SEPTEMBER 1973 – REVISED SEPTEMBER 1990

## schematic



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

	TL022C	TL022M	UNIT
Supply voltage, $V_{CC+}$ (see Note 1)	18	22	V
Supply voltage, $V_{CC-}$ (see Note 1)	-18	-22	V
Differential input voltage (see Note 2)	$\pm 30$	$\pm 30$	V
Input voltage (any input, see Notes 1 and 3)	$\pm 15$	$\pm 15$	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table		
Operating free-air temperature range	0 to 70	-55 to 125	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	JG or U package	300	$^{\circ}\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D or P package	260	$^{\circ}\text{C}$

- NOTES: 1. All voltage values, unless otherwise noted, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $\text{IN}+$  with respect to  $\text{IN}-$ .  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
 4. The output may be shorted to ground or either power supply. For the TL022M only, the unlimited duration of the short circuit applies at (or below) 125 $^{\circ}\text{C}$  case temperature or 75 $^{\circ}\text{C}$  free-air temperature.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D	680 mW	5.8 mW/ $^{\circ}\text{C}$	33 $^{\circ}\text{C}$	464 mW	—
JG	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	210 mW
P	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	—
U	675 mW	5.4 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	432 mW	135 mW

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TL022C, TL022M DUAL LOW-POWER OPERATIONAL AMPLIFIERS

SLOS076 – SEPTEMBER 1973 – REVISED SEPTEMBER 1990

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC+}$	5	15	V
Supply voltage, $V_{CC-}$	-5	-15	V

## electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm 15$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONST	TL022C			TL022M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0,$ $R_S = 50 \Omega$	25°C	1	5	1	5	mV	
		Full range		7.5		6		
$I_{IO}$ Input offset current	$V_O = 0$	25°C	15	80	5	40	nA	
		Full range		200		100		
$I_{IB}$ Input bias current	$V_O = 0$	25°C	100	250	50	100	nA	
		Full range		400		250		
$V_{ICR}$ Common-mode input voltage range		25°C	±12	±13	±12	±13	V	
		Full range	±12		±12			
$V_{O(PP)}$ Maximum peak-to-peak output voltage swing	$R_L = 10 \text{ k}\Omega$	25°C	20	26	20	26	V	
	$R_L \geq 10 \text{ k}\Omega$	Full range	20		20			
$A_{VD}$ Large-signal differential voltage amplification	$R_L \geq 10 \text{ k}\Omega,$ $V_O = \pm 10 \text{ V}$	25°C	60	80	72	86	dB	
		Full range	60		66			
$B_1$ Unity-gain bandwidth		25°C	0.5		0.5		MHz	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $R_S = 50 \Omega$	25°C	60	72	60	72	dB	
		Full range	60		60			
$k_{SVS}$ Supply voltage sensitivity ( $\Delta V_{IO}/\Delta V_{CC}$ )	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V},$ $R_S = 50 \Omega$	25°C	30	200	30	150	$\mu\text{V/V}$	
		Full range		200		150		
$V_n$ Equivalent input noise voltage	$A_{VD} = 20 \text{ dB},$ $B = 1 \text{ Hz}, f = 1 \text{ kHz}$	25°C	50		50		nV/Hz	
$I_{OS}$ Short-circuit output current		25°C	±6		±6		mA	
$I_{CC}$ Supply current (both amplifiers)	$V_O = 0,$ No load	25°C	130	250	130	250	$\mu\text{A}$	
		Full range		250		250		
$P_D$ Total dissipation (both amplifiers)	$V_O = 0,$ No load	25°C	3.9	7.5	3.9	6	mW	
		Full range		7.5		6		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for TL022C is 0°C to 70°C and for TL022M is -55°C to 125°C.

## operating characteristics, $V_{CC\pm} = \pm 15$ V, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_r$ Rise time	$V_I = 20 \text{ mV}, R_L = 10 \text{ k}\Omega, C_L = 100 \text{ pF},$ See Figure 1		0.3		$\mu\text{s}$
Overshoot factor			5%		
SR Slew rate at unity gain	$V_I = 10 \text{ V}, R_L = 10 \text{ k}\Omega, C_L = 100 \text{ pF},$ See Figure 1		0.5		$\text{V}/\mu\text{s}$

# TL022C, TL022M DUAL LOW-POWER OPERATIONAL AMPLIFIERS

SLOS076 – SEPTEMBER 1973 – REVISED SEPTEMBER 1990

## PARAMETER MEASUREMENT INFORMATION

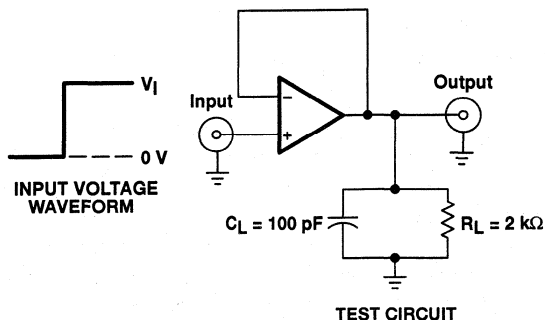


Figure 1. Rise Time, Overshoot Factor, and Slew Rate

## TYPICAL CHARACTERISTICS

TOTAL POWER DISSIPATION  
vs  
SUPPLY RATE

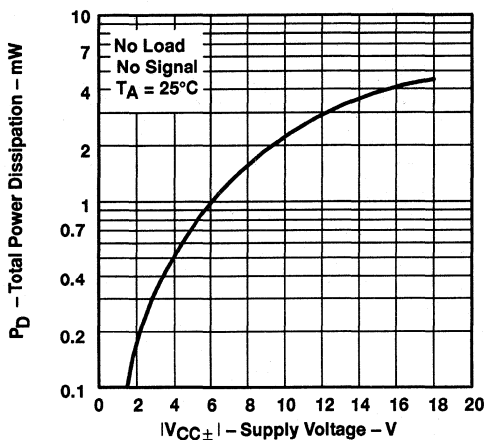


Figure 2

# TL031, TL031A ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C – JULY 1988 – REVISED AUGUST 1994

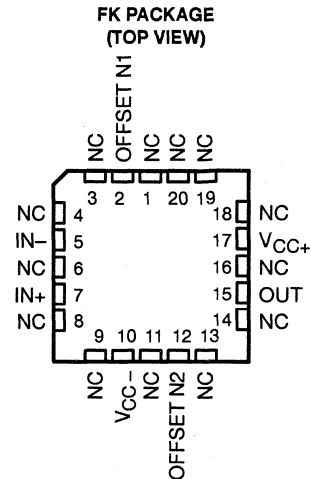
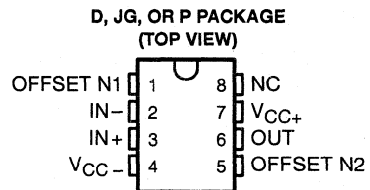
- **Maximum Offset Voltage . . . 800  $\mu$ V**
- **High Slew Rate . . . 2.9 V/ $\mu$ s Typ**
- **Low Input Bias Current . . . 2 pA Typ**
- **Very Low Power Consumption  
6.5 mW Typ**
- **Output Short-Circuit Protection**

## description

The TL031 and TL031A operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

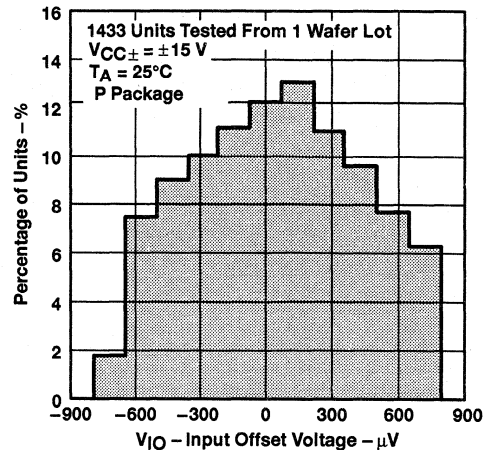
This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages, coupled with low power consumption, make the TL031 well suited for new state-of-the-art designs as well as existing design upgrades. The TL031 has been designed to be functionally compatible and pin compatible with the TL061. Two offset voltage grades are available: TL031 (1.5 mV max) and TL031A (800  $\mu$ V max).

A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.



NC – No internal connection

**DISTRIBUTION OF TL031A  
INPUT OFFSET VOLTAGE**



**AVAILABLE OPTIONS**

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D) <sup>†</sup>	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	0.8 mV 1.5 mV	TL031ACD TL031CD	—	—	TL031ACP TL031CP
-40°C to 85°C	0.8 mV 1.5 mV	TL031AID TL031ID	—	—	TL031AIP TL031IP
-55°C to 125°C	0.8 mV 1.5 mV	TL031AMD TL031MD	TL031AMFK TL031MFK	TL031AMJG TL031MJG	TL031AMP TL031MP

<sup>†</sup> The D packages are available taped and reeled. Add R suffix to device type (e.g., TL031CDR).

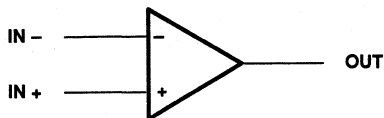
# TL031, TL031A ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C - JULY 1988 - REVISED AUGUST 1994

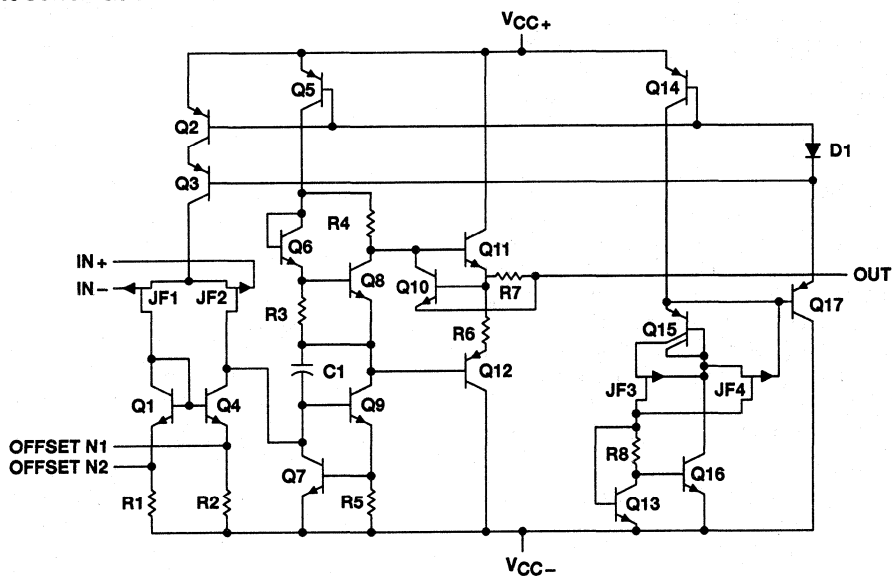
## description (continued)

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

## symbol (each amplifier)



## equivalent schematic





**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C - JULY 1988 - REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$ V
Input voltage range, $V_I$ (any input) (see Notes 1 and 3)	$\pm 15$ V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$	$\pm 40$ mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
 4. The output can be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC\pm}$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5$ V		-1.5	4	-1.5	4	V
	$V_{CC\pm} = \pm 15$ V		-11.5	14	-11.5	14	
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



# TL031, TL031A ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C – JULY 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL031C, TL031AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031C	25°C	0.54	3.5	0.5	1.5	mV	
			Full range	4.5			2.5		
		TL031AC	25°C	0.41	2.8	0.34	0.8		
			Full range	3.8			1.8		
α <sub>VIO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031C	25°C to 70°C	7.1		5.9		μV/°C	
TL031AC		25°C to 70°C	7.1		5.9	25			
Input offset voltage long-term drift (see Note 6)			25°C	0.04		0.04		μV/mo	
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5		25°C	1	100	1	100	pA	
			70°C	9	200	12	200		
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5		25°C	2	200	2	200	pA	
			70°C	50	400	80	400		
V <sub>ICR</sub> Common-mode input voltage range			25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V	
			Full range	-1.5 to .4		-11.5 to 14			
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3	4.3	13	14	V	
			0°C	3	4.2	13	14		
			70°C	3	4.3	13	14		
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-3	-4.2	-12.5	-13.9	V	
			0°C	-3	-4.1	-12.5	-13.9		
			70°C	-3	-4.2	-12.5	-14		
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7		25°C	4	12	5	14.3	V/mV	
			0°C	3	11.1	4	13.5		
			70°C	4	13.3	5	15.2		
r <sub>i</sub> Input resistance			25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω	
c <sub>i</sub> Input capacitance			25°C	5		4		pF	
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	70	87	75	94	dB	
			0°C	70	87	75	94		
			70°C	70	87	75	94		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	75	96	75	96	dB	
			0°C	75	96	75	96		
			70°C	75	96	75	96		

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis for the TL031A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031C, TL031AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation	V <sub>O</sub> = 0, No load	25°C		1.9	2.5		6.5	8.4	mW
		0°C		1.8	2.5		6.3	8.4	
		70°C		1.9	2.5		6.3	8.4	
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C		192	250		217	280	μA
		0°C		184	250		211	280	
		70°C		189	250		210	280	

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031C, TL031AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR <sub>+</sub> Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C		2		1.5	2.9	V/μs	
		0°C		1.8		1	2.6		
		70°C		2.2		1.5	3.2		
SR <sub>-</sub> Negative slew rate at unity gain		R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		3.9		1.5	5.1	V/μs
			0°C		3.7		1.5	5	
			70°C		4		1.5	5	
t <sub>r</sub> Rise time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		138		132	ns	
			0°C		134		127		
			70°C		150		142		
t <sub>f</sub> Fall time		V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns	
			0°C		134		127		
			70°C		150		142		
Overshoot factor	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		11%		5%		
			0°C		10%		4%		
			70°C		12%		6%		
V <sub>n</sub> Equivalent input noise voltage (see Note 9)		R <sub>S</sub> = 20 Ω, See Figure 3	25°C	TL031C	f = 10 Hz		61	nV/√Hz	
				TL031AC	f = 1 kHz		41		
			25°C	TL031C	f = 10 Hz		61		
	TL031AC			f = 1 kHz		41	60		
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C		0.003		0.003	pA/√Hz		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		1		1.1	MHz		
		0°C		1		1.1			
		70°C		1		1			
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		61°		65°			
		0°C		61°		65°			
		70°C		60°		64°			

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



# TL031, TL031A

## ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C – JULY 1988 – REVISED AUGUST 1994

### electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL031I, TL031AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031I	25°C	0.54	3.5		0.5	1.5	mV
			Full range		5.3		3.3		
		TL031AI	25°C	0.41	2.8		0.34	0.8	
			Full range		4.6		2.6		
α <sub>VIO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031I	25°C to 85°C	6.5			6.2		μV/°C
		TL031AI	25°C to 85°C	6.5			6.2	25	
Input offset voltage long-term drift (see Note 6)			25°C	0.04			0.04		μV/mo
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	1	100		1	100	pA
			85°C	0.02	0.45		0.02	0.45	nA
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	2	200		2	200	pA
			85°C	0.2	0.9		0.2	0.9	nA
V <sub>ICR</sub> Common-mode input voltage range			25°C	-1.5 to 4	-3.4 to 5.4		-11.5 to 14	-13.4 to 15.4	V
			Full range	-1.5 to 4			-11.5 to 14		
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3	4.3		13	14	V
			-40°C	3	4.1		13	14	
			85°C	3	4.4		13	14	
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-3	-4.2		-12.5	-13.9	V
			-40°C	-3	-4.1		-12.5	-13.8	
			85°C	-3	-4.2		-12.5	-14	
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7		25°C	4	12		5	14.3	V/mV
			-40°C	3	8.4		4	11.6	
			85°C	4	13.5		5	15.3	
r <sub>i</sub> Input resistance			25°C	10 <sup>12</sup>			10 <sup>12</sup>	Ω	
c <sub>i</sub> Input capacitance			25°C	5			4	pF	
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	70	87		75	94	dB
			-40°C	70	87		75	94	
			85°C	70	87		75	94	
K <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	75	96		75	96	dB
			-40°C	75	96		75	96	
			85°C	75	96		75	96	

† Full range is -40°C to 85°C.

NOTES: 5. This parameter is tested on a sample basis for the TL031A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031I, TL031AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation	V <sub>O</sub> = 0, No load	25°C		1.9	2.5		6.5	8.4	mW
		-40°C		1.4	2.5		5.4	8.4	
		85°C		1.9	2.5		6.2	8.4	
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C		192	250		217	280	μA
		-40°C		144	250		181	280	
		85°C		189	250		207	280	

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031I, TL031AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR <sup>+</sup> Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C		2		1.5	2.9	V/μs	
		-40°C		1.6		1	2.1		
		85°C		2.3		1.5	3.3		
SR <sup>-</sup> Negative slew rate at unity gain		R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C		3.9		1.5	5.1	V/μs
			-40°C		3.3		1.5	4.8	
			85°C		4.1		1.5	4.9	
t <sub>r</sub> Rise time	V <sub>I(pp)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		138		132	ns	
			-40°C		132		123		
			85°C		154		146		
t <sub>f</sub> Fall time		V <sub>I(pp)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns	
			-40°C		132		123		
			85°C		154		146		
Overshoot factor	V <sub>I(pp)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		11%		5%		
			-40°C		12%		5%		
			85°C		13%		7%		
V <sub>n</sub> Equivalent input noise voltage (see Note 9)		TL031I TL031AI	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C		61		nV/√Hz
				f = 1 kHz	25°C		41		
			f = 10 Hz	25°C		61		61	
	f = 1 kHz			25°C		41		41	
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C		0.003		0.003	pA/√Hz		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C		1		1.1	MHz		
		-40°C		1		1.1			
		85°C		0.9		1			
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C		61°		65°			
		-40°C		60°		65°			
		85°C		60°		64°			

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I(pp)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(pp)</sub> = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



# TL031, TL031A ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C – JULY 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL031M, TL031AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031M	25°C	0.54 3.5		0.5 1.5		mV	
			Full range	6.5		4.5			
		TL031AM	25°C	0.41 2.8		0.34 0.8			
			Full range	5.8		3.8			
αV <sub>IO</sub> Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL031M	25°C to 125°C	5.1		4.3		μV/°C	
		TL031AM	25°C to 125°C	5.1		4.3			
Input offset voltage long-term drift (see Note 6)		25°C	0.04		0.04		μV/mo		
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	1 100		1 100		pA		
		125°C	0.2 10		0.2 10		nA		
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	2 200		2 200		pA		
		125°C	7 20		8 20		nA		
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3 4.3		13 14		V		
		-55°C	3 4.1		13 14				
		125°C	3 4.4		13 14				
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-3 -4.2		-12.5 -13.9		V		
		-55°C	-3 -4		-12.5 -13.8				
		125°C	-3 -4.3		-12.5 -14				
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7	25°C	4 12		5 14.3		V/mV		
		-55°C	3 7.1		4 10.4				
		125°C	3 12.9		4 15				
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub> Input capacitance		25°C	5		4		pF		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70 87		75 94		dB		
		-55°C	70 87		70 94				
		125°C	70 87		70 94				
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75 96		75 96		dB		
		-55°C	75 96		75 95				
		125°C	75 96		75 96				
P <sub>D</sub> Total power dissipation	V <sub>O</sub> = 0, No load	25°C	1.9 2.5		6.5 8.4		mW		
		-55°C	1.1 2.5		4.7 8.4				
		125°C	1.8 2.5		5.8 8.4				

† Full range is -55°C to 125°C.

NOTES: 6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031M, TL031AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C		192	250		217	280	μA
		-55°C		114	250		156	280	
		125°C		178	250		197	280	

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL031M, TL031AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR <sub>+</sub> Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C		2		1.5	2.9	V/μs	
		-55°C		1.4		1	1.9		
		125°C		2.4		1	3.5		
SR <sub>-</sub> Negative slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C		3.9		1.5	5.1	V/μs	
		-55°C		3.2		1	4.6		
		125°C		4.1		1	4.7		
t <sub>r</sub> Rise time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns		
		-55°C		142		123			
		125°C		166		158			
t <sub>f</sub> Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns		
		-55°C		142		123			
		125°C		166		158			
Overshoot factor	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		11%		5%			
		-55°C		16%		6%			
		125°C		14%		8%			
V <sub>n</sub> Equivalent input noise voltage	R <sub>S</sub> = 20 Ω, See Figure 3	25°C	TL031 M	f = 10 Hz		61	nV/√Hz		
			TL031 M	f = 1 kHz		41			
		TL031 AM	f = 10 Hz	25°C		61			
			f = 1 kHz	25°C		41			
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C		0.003		0.003	pA/√Hz		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		1		1.1	MHz		
		-55°C		1		1.1			
		125°C		0.9		0.9			
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		61°		65°			
		-55°C		57°		64°			
		125°C		59°		62°			

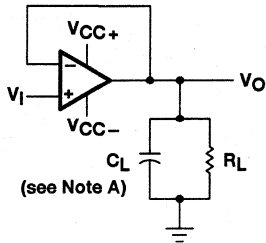
NOTE 8.: For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.



**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

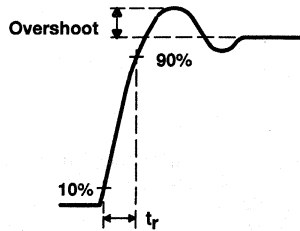
SLOS032C – JULY 1988 – REVISED AUGUST 1994

**PARAMETER MEASUREMENT INFORMATION**

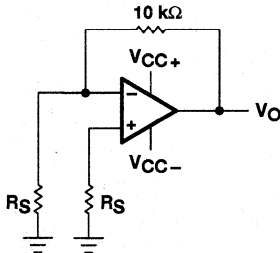


NOTE A:  $C_L$  includes fixture capacitance.

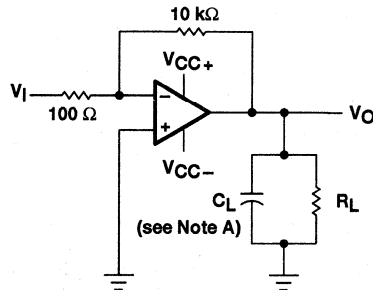
**Figure 1. Slew Rate, Rise/Fall Time, and Overshoot Test Circuit**



**Figure 2. Rise Time and Overshoot Waveform**

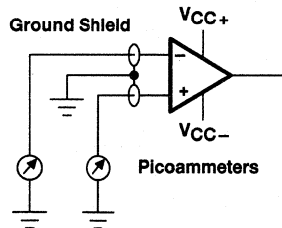


**Figure 3. Noise-Voltage Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit**



**Figure 5. Input-Bias and Offset-Current Test Circuit**

**typical values**

Typical values presented in this data sheet represent the median (50% point) of device parametric performance.



## PARAMETER MEASUREMENT INFORMATION

### input bias and offset current

At the picoampere bias-current level typical of the TL031 and TL031A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

### noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Input bias current	vs Free-air temperature	8
		vs Common-mode input voltage	9
$V_{IC}$	Common-mode input voltage	vs Supply voltage	10
		vs Free-air temperature	11
$V_O$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage	vs Supply voltage	14
		vs Output current	16, 17
		vs Free-air temperature	18, 19
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	15
$A_{VD}$	Large-signal differential voltage amplification	vs Load resistance	20
		vs Frequency	21
		vs Free-air temperature	22
$z_o$	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency	24, 25
		vs Free-air temperature	26
kSVR	Supply-voltage rejection ratio	vs Free-air temperature	27
$I_{OS}$	Short-circuit output current	vs Supply voltage	28
		vs Time	29
		vs Free-air temperature	30
$V_n$	Equivalent input noise voltage	vs Frequency	31
$I_{CC}$	Supply current	vs Supply voltage	32
		vs Free-air temperature	33
SR	Slew rate	vs Load resistance	34, 35
		vs Free-air temperature	36, 37
	Overshoot factor	vs Load capacitance	38
THD	Total harmonic distortion	vs Frequency	39
$B_1$	Unity-gain bandwidth	vs Supply voltage	40
		vs Free-air temperature	41
$\phi_m$	Phase margin	vs Supply voltage	42
		vs Load capacitance	43
		vs Free-air temperature	44
	Phase shift	vs Frequency	21
	Pulse response	Small signal	45
		Large signal	46, 47



**TYPICAL CHARACTERISTICS†**

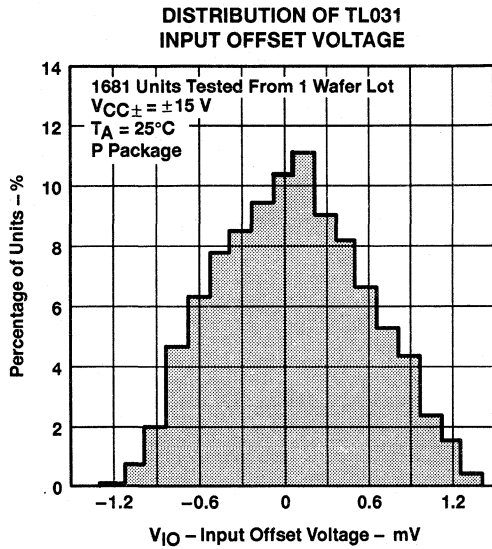


Figure 6

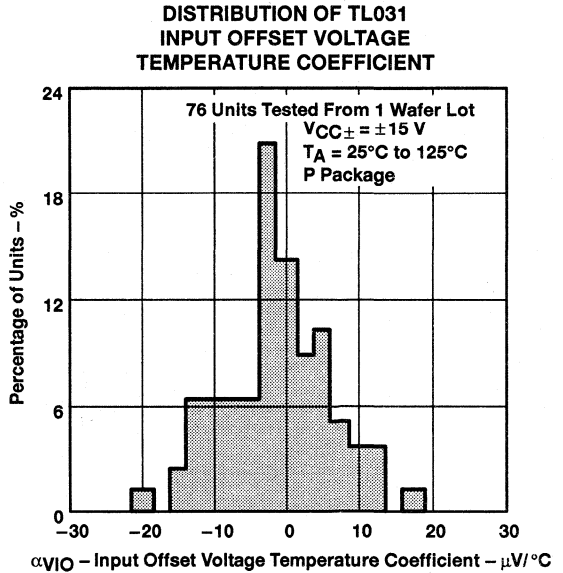


Figure 7

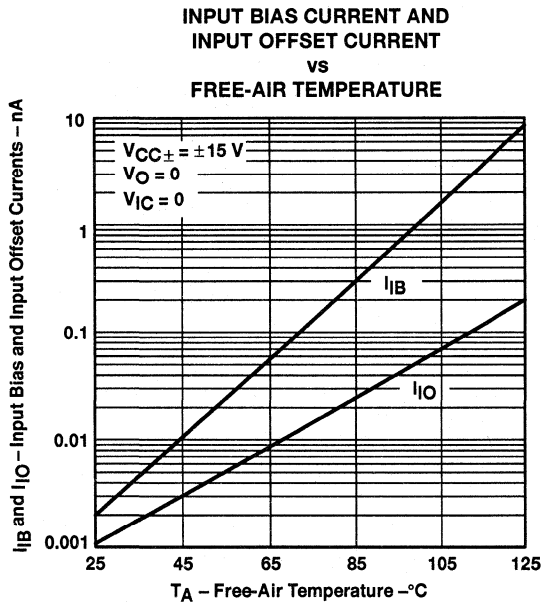


Figure 8

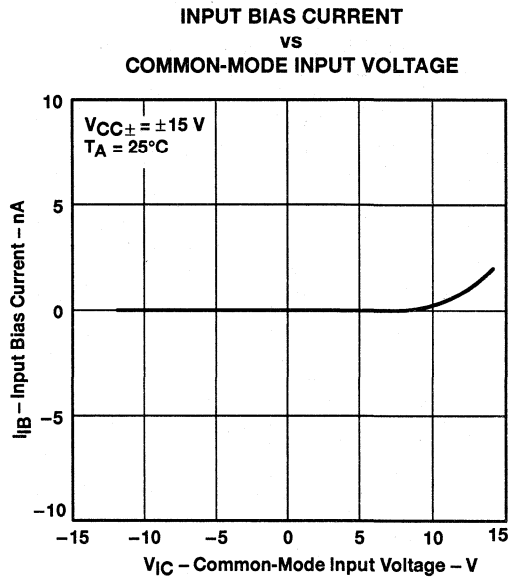


Figure 9

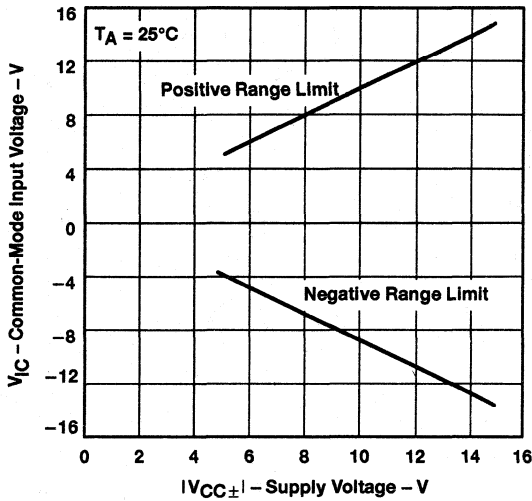
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

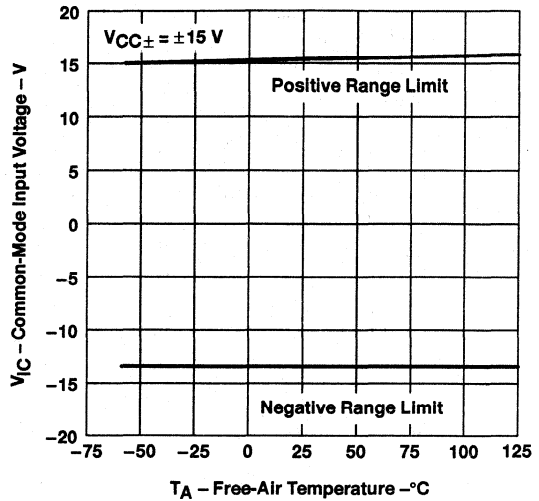
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE INPUT VOLTAGE**  
**VS**  
**SUPPLY VOLTAGE**



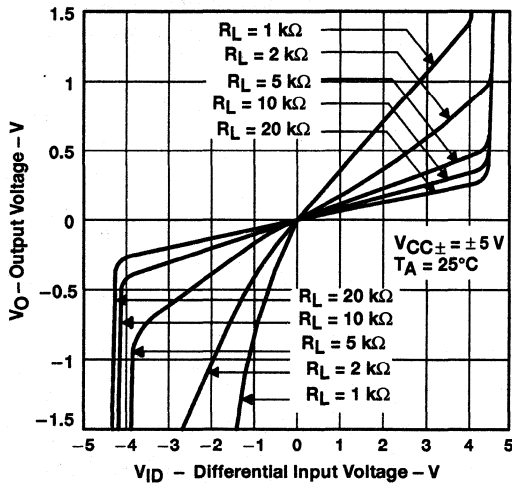
**Figure 10**

**COMMON-MODE INPUT VOLTAGE**  
**VS**  
**FREE-AIR TEMPERATURE**



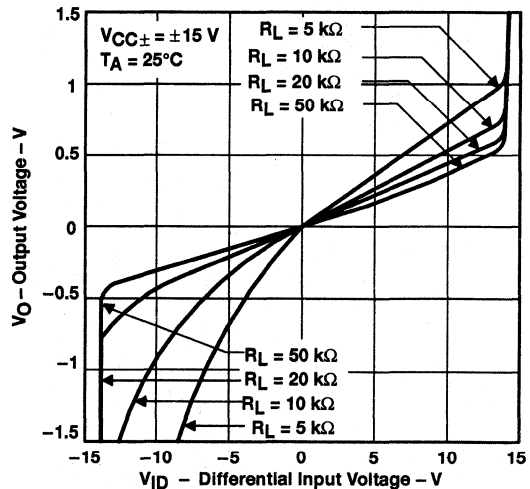
**Figure 11**

**OUTPUT VOLTAGE**  
**VS**  
**DIFFERENTIAL INPUT VOLTAGE**



**Figure 12**

**OUTPUT VOLTAGE**  
**VS**  
**DIFFERENTIAL INPUT VOLTAGE**

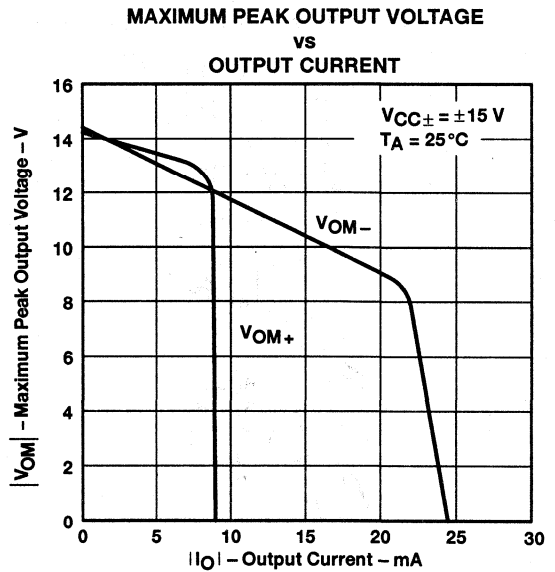
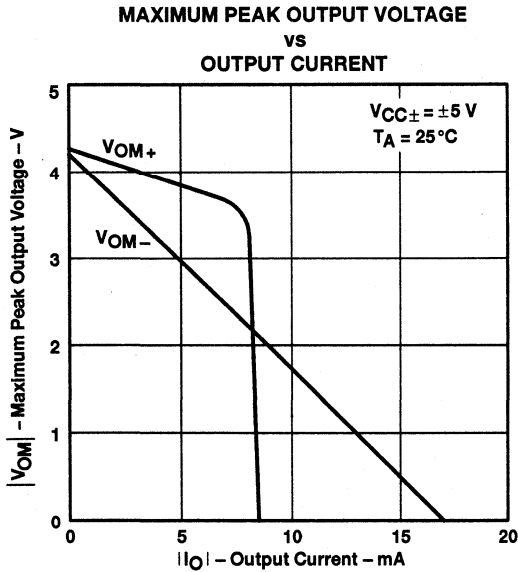
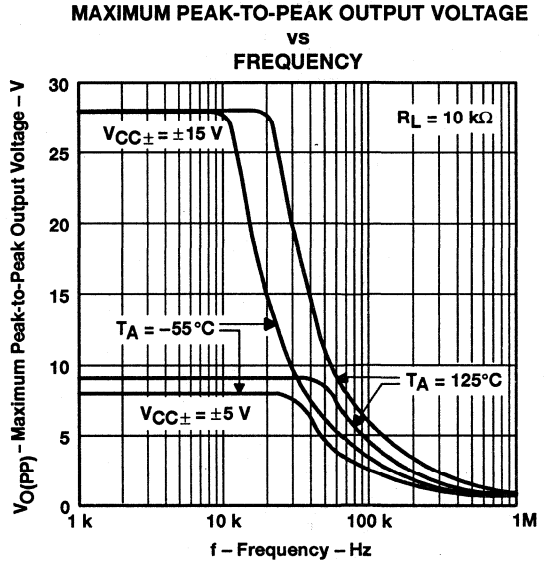
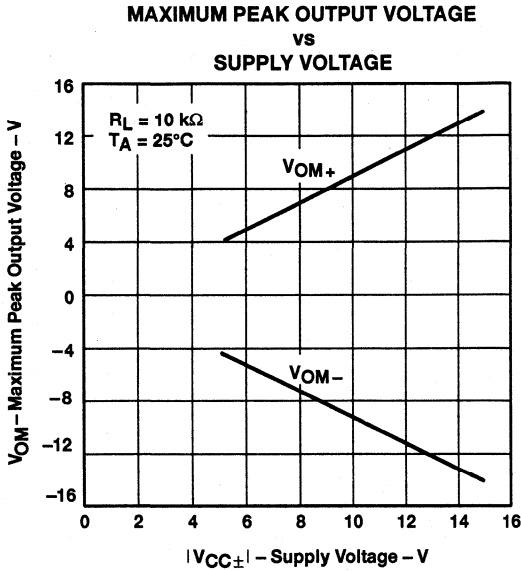


**Figure 13**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS†**



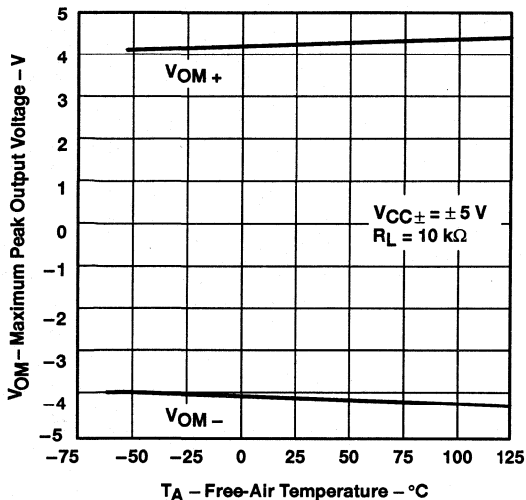
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C - JULY 1988 - REVISED AUGUST 1994

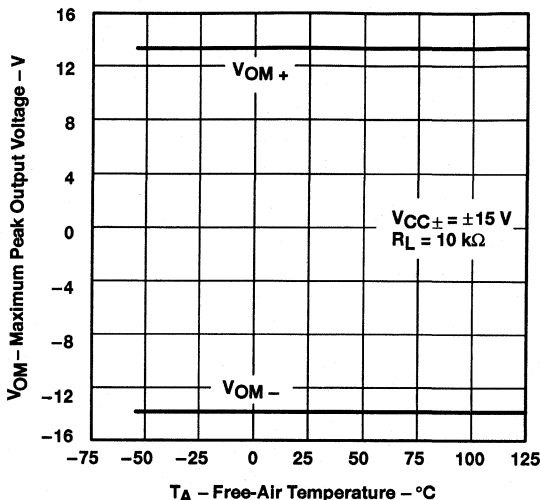
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



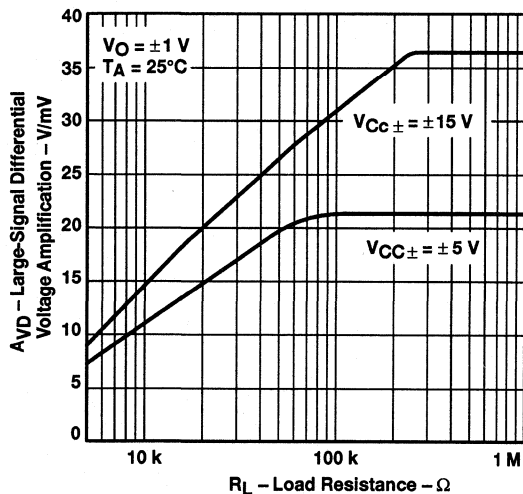
**Figure 18**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



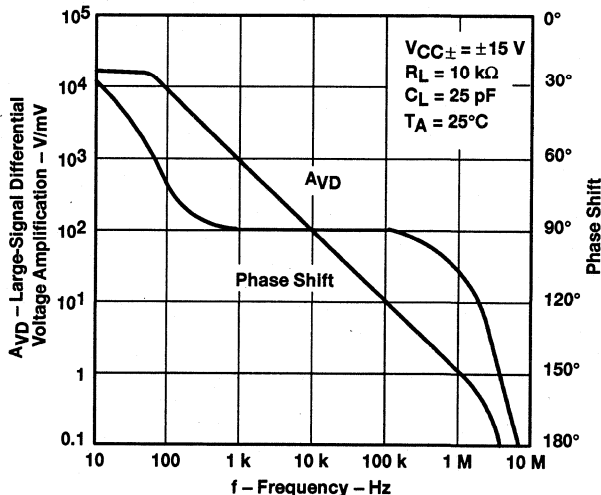
**Figure 19**

**LARGE-SIGNAL DIFFERENTIAL**  
**VOLTAGE AMPLIFICATION**  
**vs**  
**LOAD RESISTANCE**



**Figure 20**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION AND PHASE SHIFT**  
**vs**  
**FREQUENCY**



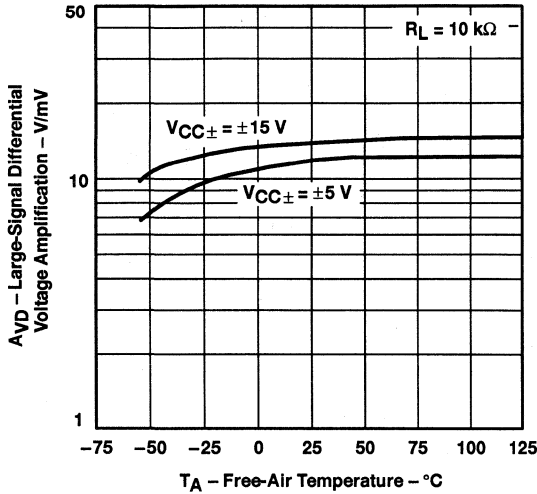
**Figure 21**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

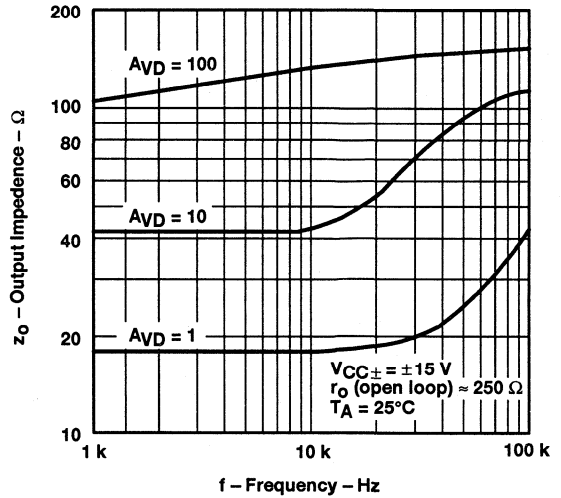


**TYPICAL CHARACTERISTICS†**

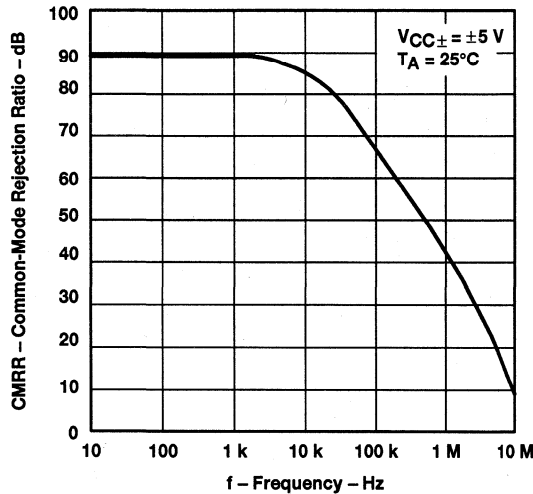
**LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE**



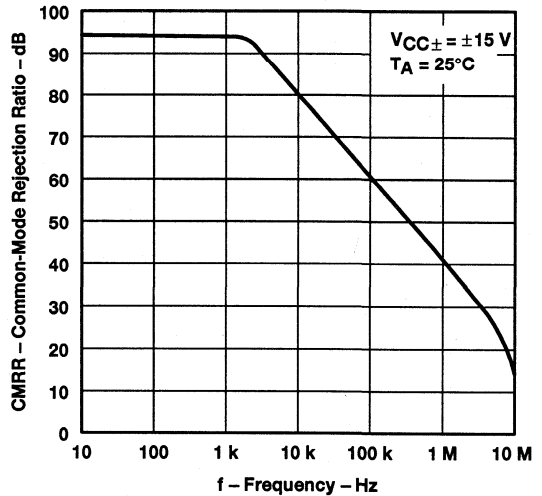
**OUTPUT IMPEDANCE  
 VS  
 FREQUENCY**



**COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY**



**COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY**



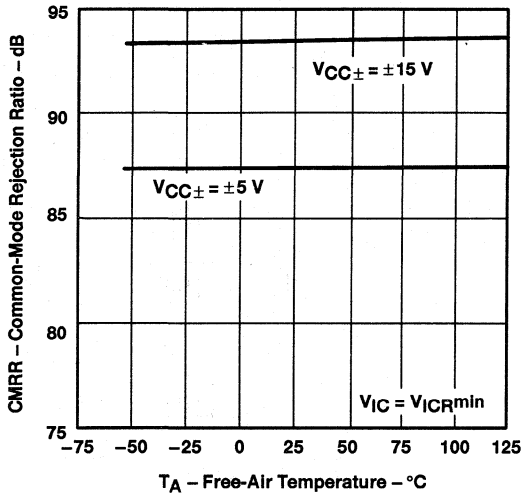
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

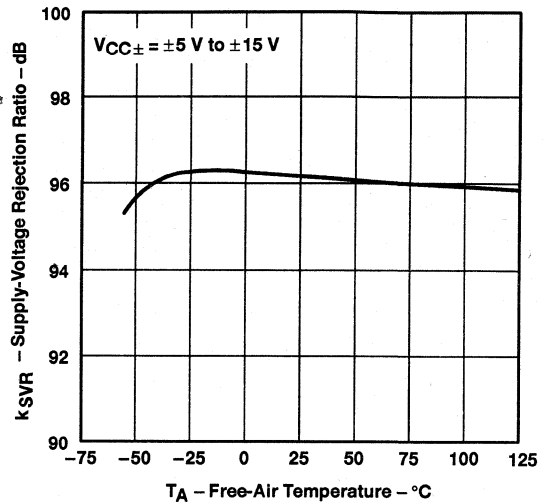
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



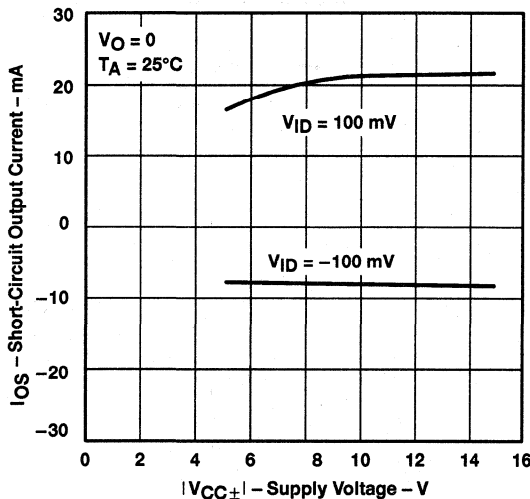
**Figure 26**

**SUPPLY-VOLTAGE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



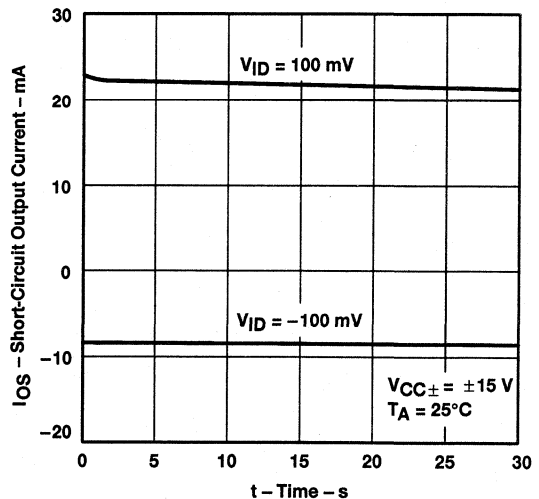
**Figure 27**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 28**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**TIME**



**Figure 29**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





**TYPICAL CHARACTERISTICS†**

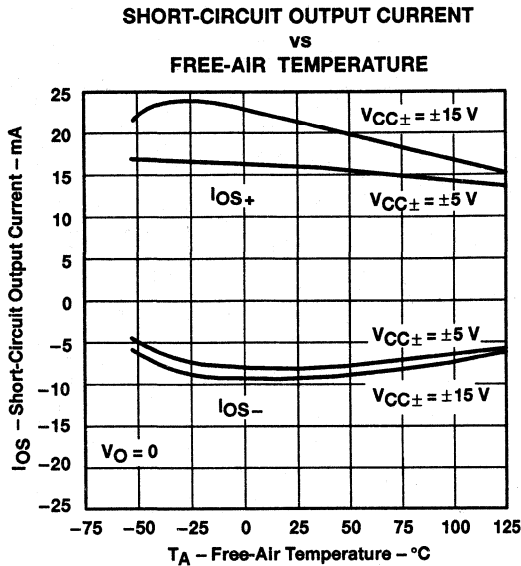


Figure 30

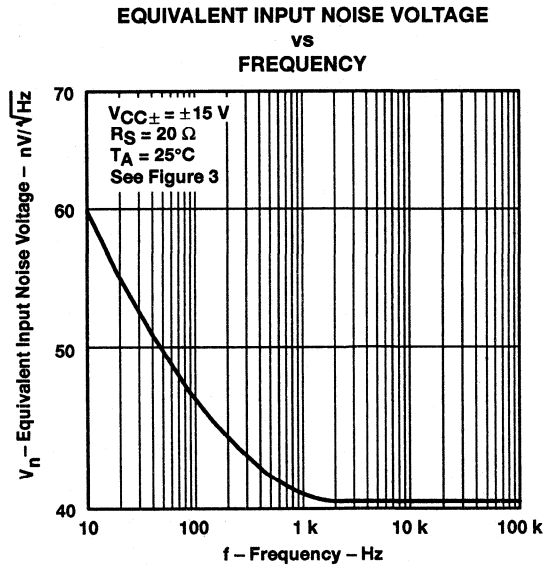


Figure 31

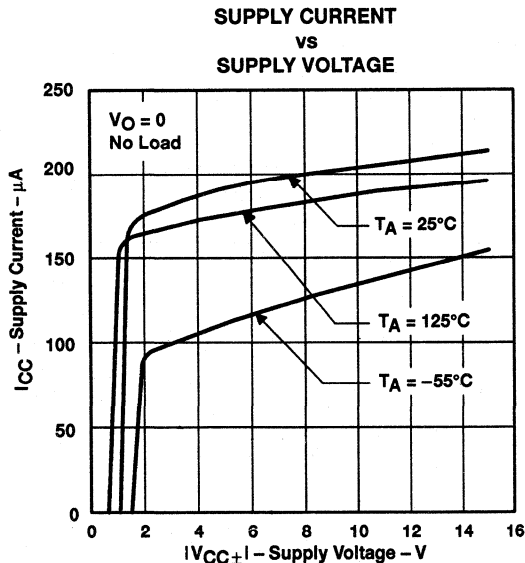


Figure 32

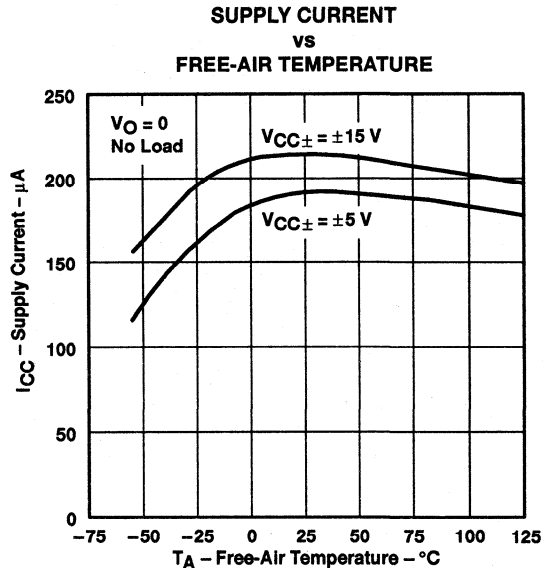


Figure 33

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

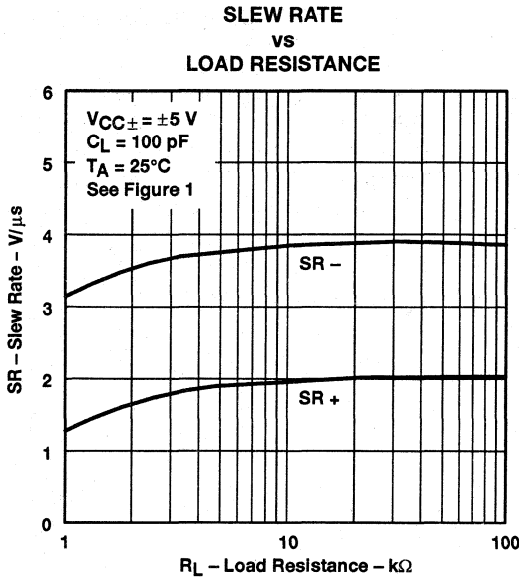


Figure 34

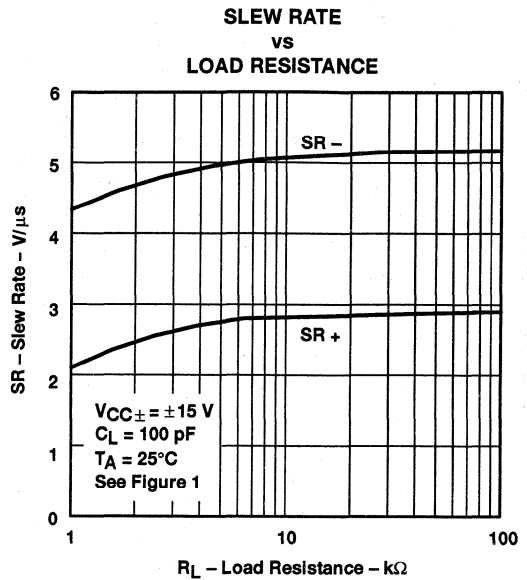


Figure 35

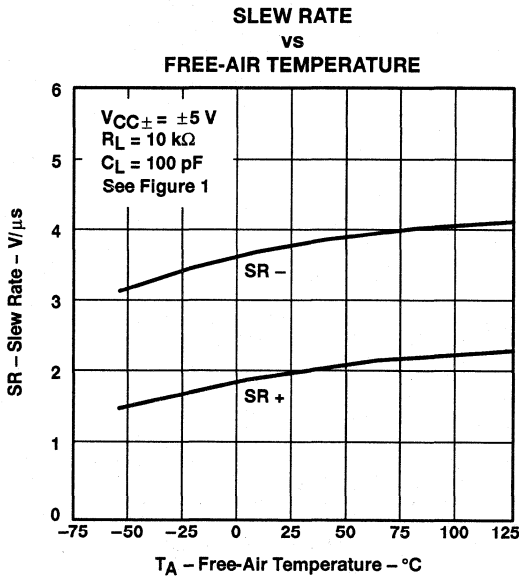


Figure 36

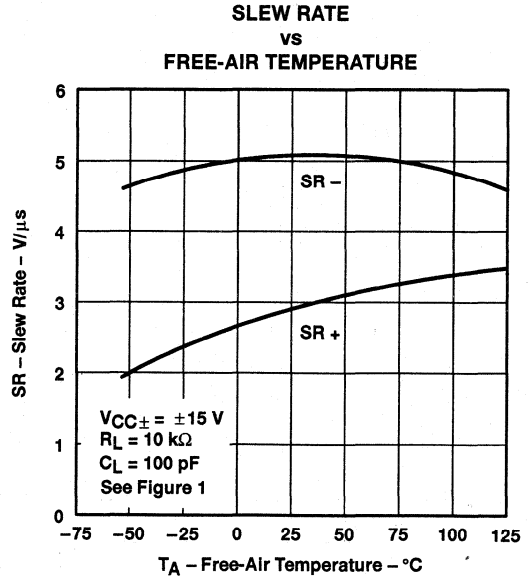
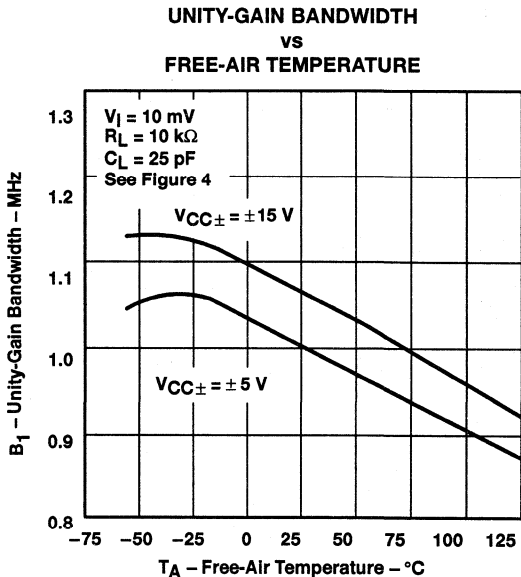
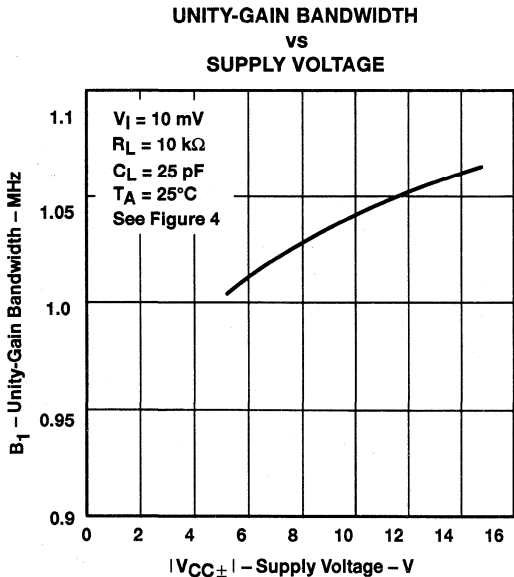
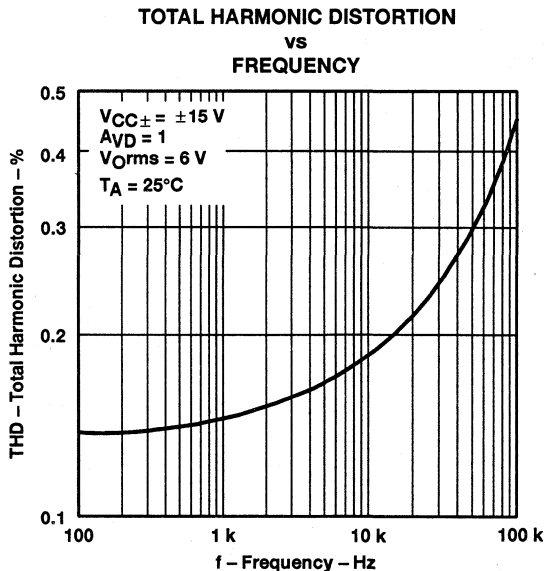
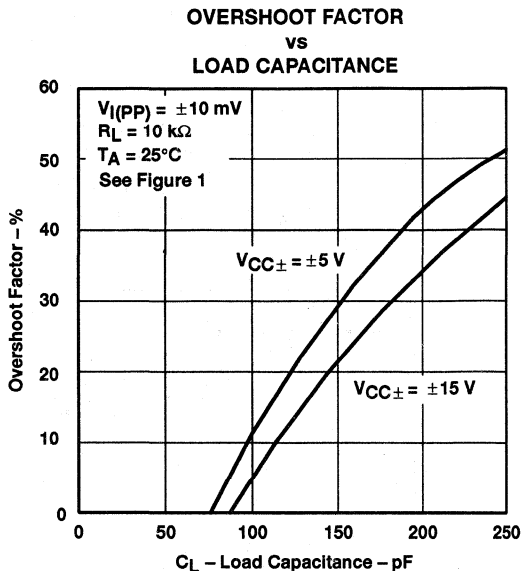


Figure 37

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

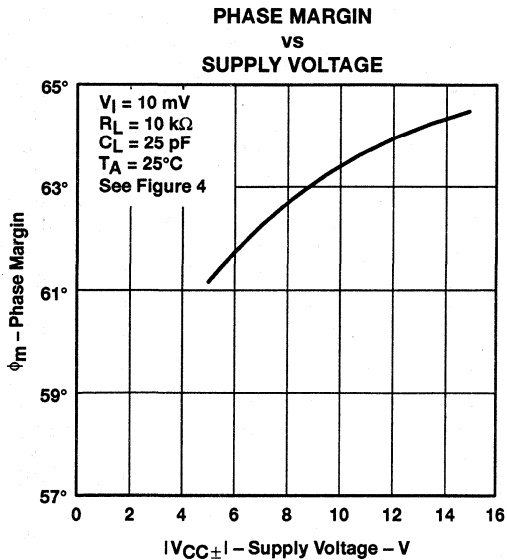


† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

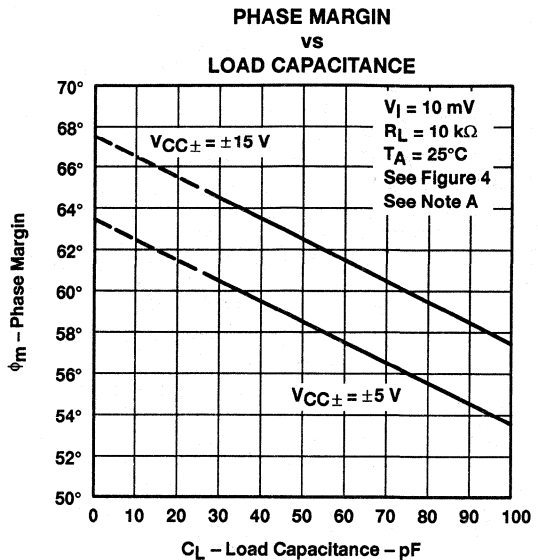
**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C - JULY 1988 - REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

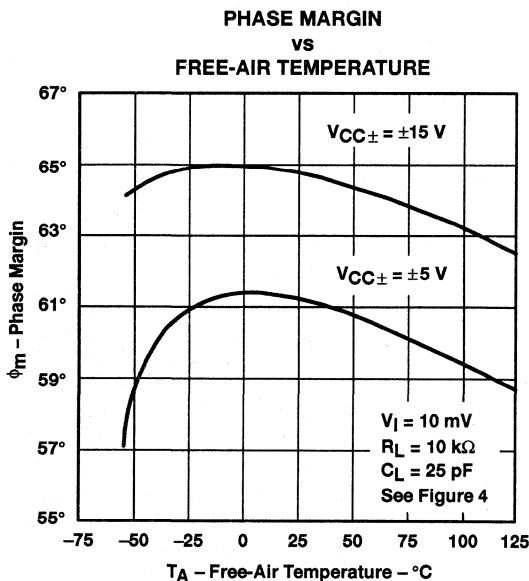


**Figure 42**

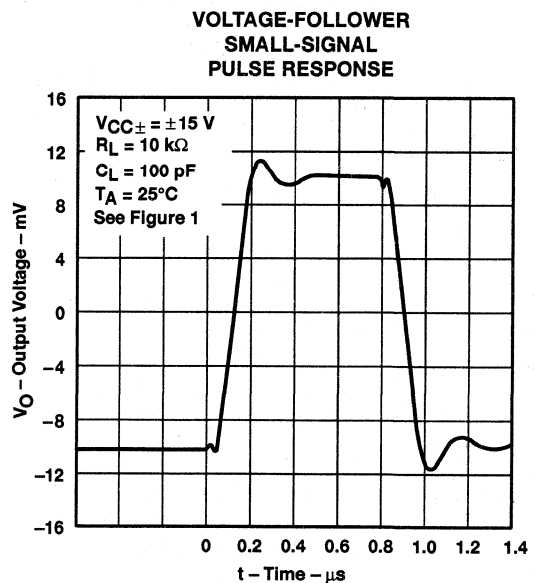


NOTE A: Values of phase margin below a load capacitance of 25 pF were estimated.

**Figure 43**



**Figure 44**



**Figure 45**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**

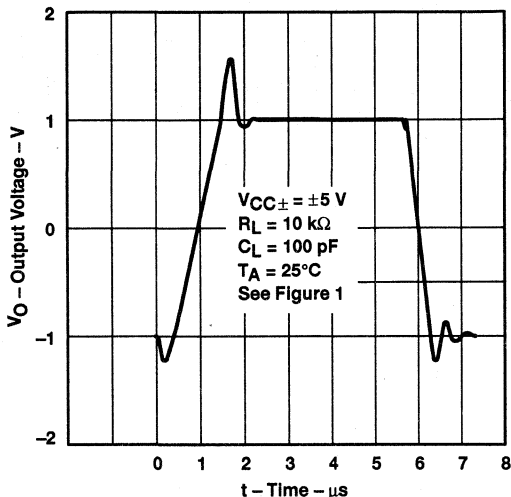


Figure 46

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**

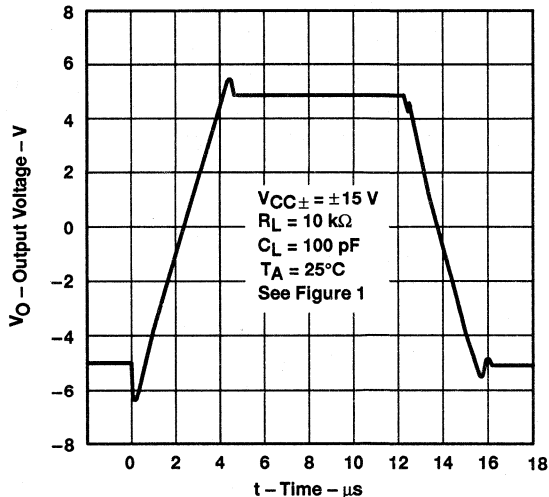


Figure 47

**APPLICATION INFORMATION**

**input characteristics**

The TL031 and TL031A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Because of the extremely high input impedance and resulting low bias current requirements, the TL031 and TL031A are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

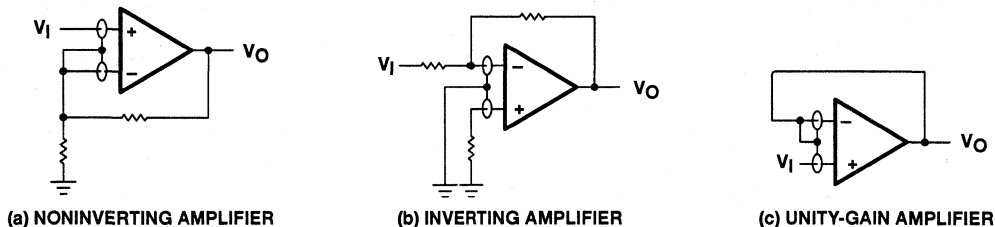


Figure 48. Use of Guard Rings

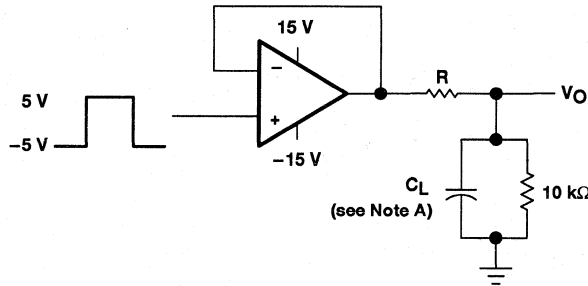
**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

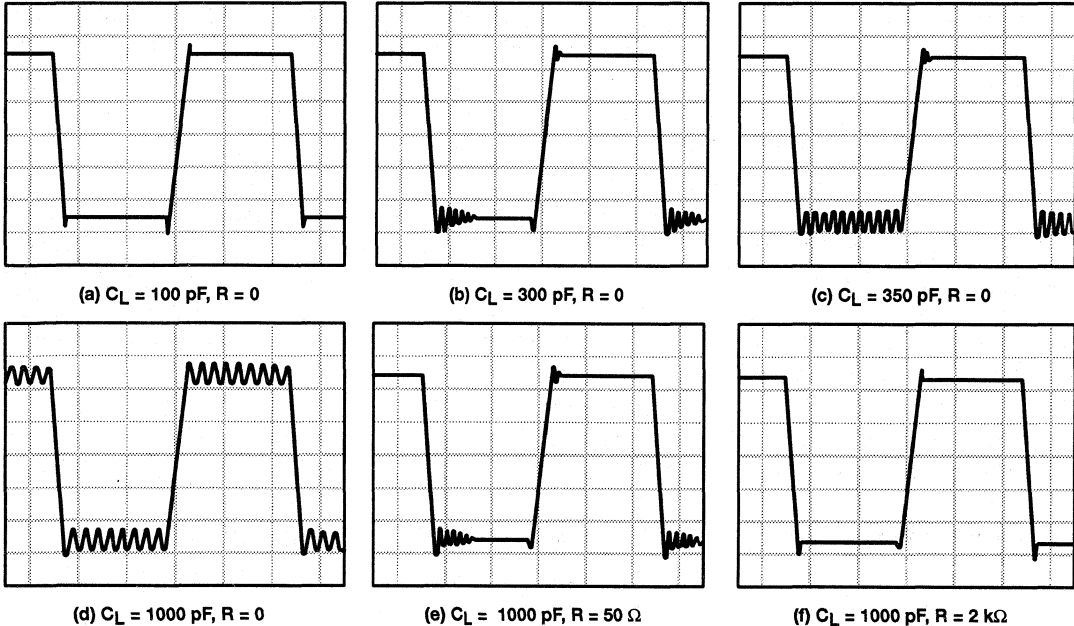
**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL031 and TL031A drives higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem (see Figure 49). Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 49. Test Circuit for Output Characteristics**



**Figure 50. Effect of Capacitive Loads**

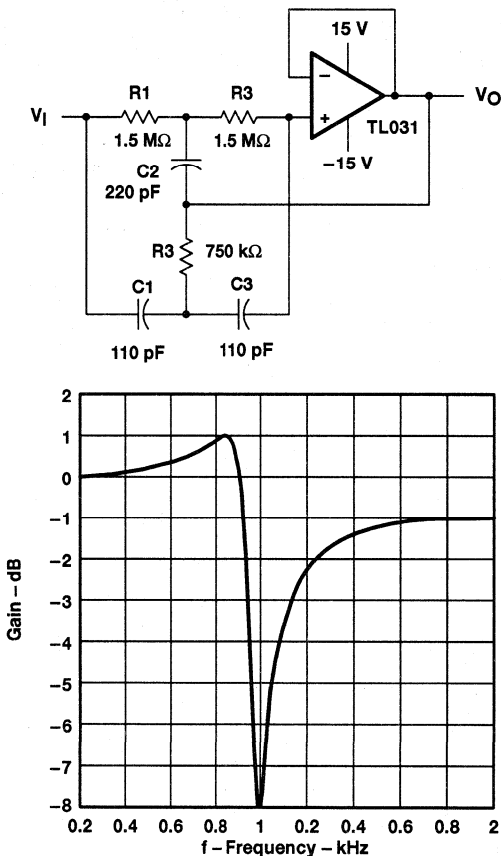
**APPLICATION INFORMATION**

**high-Q notch filter**

In general, Texas Instruments enhanced JFET operational amplifiers serve as excellent filters. The circuit in Figure 51 provides a narrow notch at a specific frequency. Notch filters are designed to eliminate frequencies that are interfering with the operation of an application. For this filter, the center frequency can be calculated as:

$$f_o = \frac{1}{2\pi R_1 C_1}$$

With the resistors and capacitors shown in Figure 51, the center frequency is 1 kHz.  $C_1 = C_3 = C_2 + 2$  and  $R_1 = R_3 = 2 \times R_2$ . The center frequency can be modified by varying these values. When adjusting the center frequency, ensure that the operational amplifier has sufficient gain at the frequency required.



**Figure 51. High-Q Notch Filter**

# TL031, TL031A ENHANCED-JFET LOW-POWER LOW-OFFSET OPERATIONAL AMPLIFIERS

SLOS032C – JULY 1988 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### transimpedance amplifier

The low-power precision TL031 allows accurate measurement of low currents. The high input impedance and low offset voltage of the TL031A greatly simplify the design of a transimpedance amplifier. At room temperature, this design achieves ten-bit accuracy with an error of less than 1/2 LSB.

Assuming that  $R_2$  is much less than  $R_1$  and ignoring error terms, the output voltage can be expressed as:

$$V_O = -I_{IN} \times R_F \left( \frac{R_1 + R_2}{R_2} \right)$$

Using the resistor values shown in the schematic for a 1-nA input current, the output voltage equals  $-0.1$  V. If the  $V_O$  limit for the TL031A is measured at  $\pm 12$  V, the maximum input current for these resistor values is  $\pm 120$  nA. Similarly, one LSB on a ten-bit scale corresponds to 12 mV of output voltage, or 120 pA of input current.

The following equation shows the effect of input offset voltage and input bias current on the output voltage:

$$V_O = - \left[ V_{IO} + R_F (I_{IO} + I_{IB}) \right] \left( \frac{R_1 + R_2}{R_2} \right)$$

If the application requires input protection for the transimpedance amplifier, do not use standard PN diodes. Instead, use low-leakage Siliconix SN4117 JFETs (or equivalent) connected as diodes across the TL031A inputs as shown in Figure 52.

As with all precision applications, special care must be taken to eliminate external sources of leakage and interference. Other precautions include using high-quality insulation, cleaning insulating surfaces to remove fluxes and other residue, and enclosing the application within a protective box.

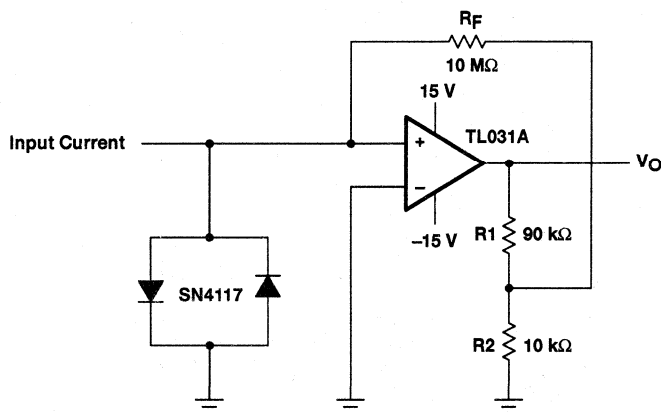


Figure 52. Transimpedance Amplifier



## APPLICATION INFORMATION

### 4-mA to 20-mA current loops

Often, information from an analog sensor must be sent over a distance to the receiving circuitry. For many applications, the most feasible method involves converting voltage information to a current before transmission. The following circuits give two variations of low-power current loops. The circuit in Figure 53 requires three wires from the transmitting to receiving circuitry while the second variation in Figure 54 requires only two wires but includes an extra integrated circuit. Both circuits benefit from the high input impedance of the TL031A since many inexpensive sensors do not have low output impedance.

Assuming that the voltage at the noninverting input of the TL031A is zero, the following equation determines the output current:

$$I_O = V_I \left( \frac{R_3}{R_1 \times R_S} \right) + 5 V \left( \frac{R_3}{R_2 \times R_S} \right) = 0.16 \times V_I + 4 \text{ mA}$$

The circuits presently provide 4-mA to 20-mA output for an input voltage of 0 to 100 mV. By modifying R1, R2, and R3, the input voltage range or the output current range can be adjusted.

Including the offset voltage of the operational amplifier in the above equation clearly illustrates why the low offset TL031A was chosen:

$$I_O = V_I \left( \frac{R_3}{R_1 \times R_S} \right) + 5 V \left( \frac{R_3}{R_2 \times R_S} \right) - V_I \left( \frac{R_3}{R_1 \times R_S} + \frac{R_3}{R_2 \times R_S} + \frac{R_1}{R_S} \right)$$

$$= 0.16 \times V_I + 4 \text{ mA} - 0.17 \times V_I$$

For example, an offset voltage of 1 mV decreases the output current by 0.17 mA.

Due to the low power consumption of the TL031A, both circuits have at least 2 mA available to drive the actual sensor from the 5-V reference node.

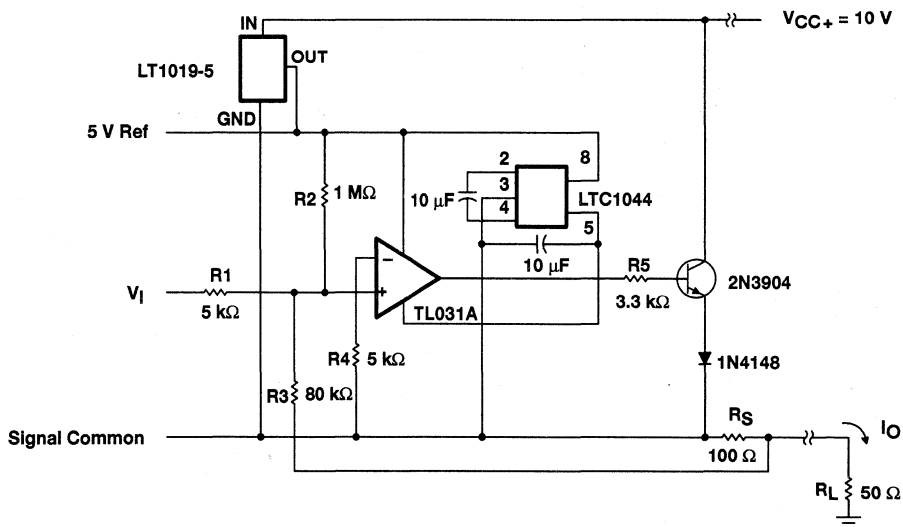


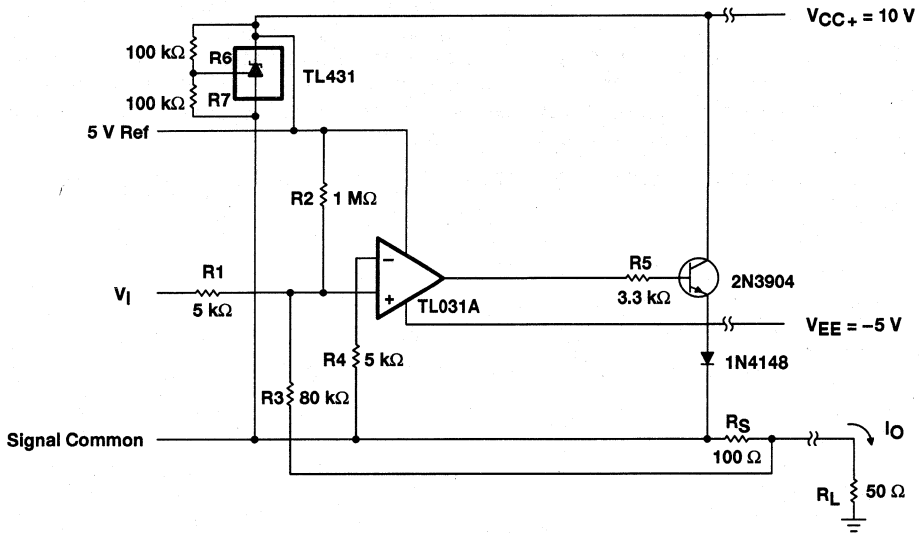
Figure 53. 2-Wire 4-mA to 20-mA Current Loop

**TL031, TL031A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**OPERATIONAL AMPLIFIERS**

SLOS032C – JULY 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

**4-mA to 20-mA current loops (continued)**



**Figure 54. 3-Wire 4-mA to 20-mA Current Loop**

# TL032, TL032A

## ENHANCED-JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

SLOS033D – JULY 1988 – REVISED AUGUST 1994

- Maximum Offset Voltage . . . 800  $\mu\text{V}$
- High Slew Rate . . . 2.9 V/ $\mu\text{s}$  Typ
- Low Input Bias Current . . . 2 pA Typ
- Very Low Power Consumption  
13 mW Typ
- Output Short-Circuit Protection

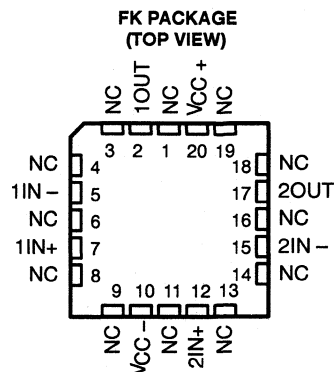
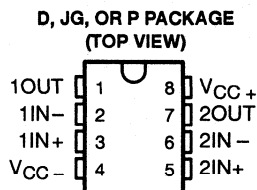
### description

The TL032 and TL032A dual operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

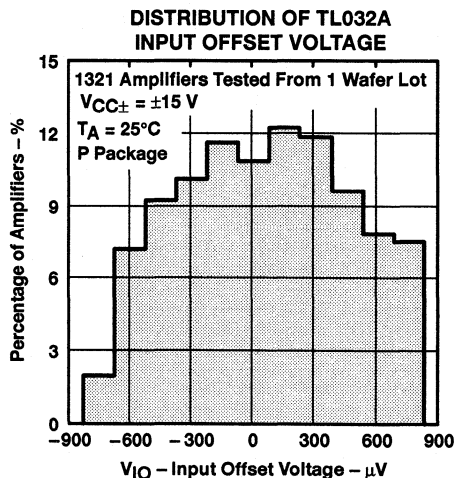
This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low power consumption make the TL032 well suited for new state-of-the-art designs as well as existing design upgrades.

The TL032 has been designed to be functionally compatible and pin compatible with the TL062. Two offset voltage grades are available: TL032 (1.5 mV max) and TL032A (800  $\mu\text{V}$  max).

A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.



NC – No internal connection



### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	0.8 mV 1.5 mV	TL032ACD TL032CD	—	—	TL032ACP TL032CP
-40°C to 85°C	0.8 mV 1.5 mV	TL032AID TL032ID	—	—	TL032AIP TL032IP
-55°C to 125°C	0.8 mV 1.5 mV	TL032AMD TL032MD	TL032AMFK TL032MFK	TL032AMJG TL032MJG	TL032AMP TL032MP

The D packages are available taped and reeled. Add R suffix to device type (e.g., TL032CDR).

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2-143

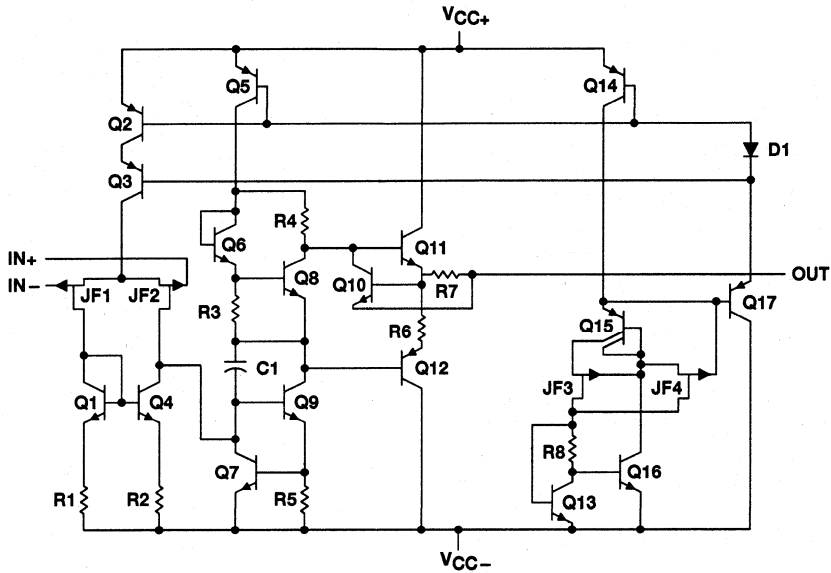
**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

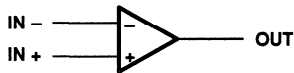
**description (continued)**

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

**equivalent schematic (each amplifier)**



**symbol (each amplifier)**



**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage (see Note 2)	$\pm 30$ V
Input voltage, $V_I$ (any input) (see Notes 1 and 3)	$\pm 15$ V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$ (each output)	$\pm 40$ mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1 /16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1 /16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC}$		$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5$ V	-1.5	4	-1.5	4	-1.5	4	V
	$V_{CC\pm} = \pm 15$ V	-11.5	14	-11.5	14	-11.5	14	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TL032, TL032A ENHANCED-JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

SLOS033D – JULY 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL032C, TL032AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL032C	25°C	0.69	3.5	0.57	1.5	mV	
			Full range	4.5			2.5		
		TL032AC	25°C	0.53	2.8	0.39	0.8		
			Full range	3.8			1.8		
αV <sub>IO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL032C	25°C to 70°C	11.5		10.8		μV/°C	
TL032AC		25°C to 70°C	11.5		10.8		25		
Input offset voltage long-term drift (see Note 6)		25°C	0.04		0.04		μV/mo		
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	1	100	1	100	pA		
		70°C	9	200	12	200			
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	2	200	2	200	pA		
		70°C	50	400	80	400			
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V		
		Full range	-1.5 to 4		-11.5 to 14				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3	4.3	13	14	V		
		0°C	3	4.2	13	14			
		70°C	3	4.3	13	14			
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V		
		0°C	-3	-4.1	-12.5	-13.9			
		70°C	-3	-4.2	-12.5	-14			
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7	25°C	4	12	5	14.3	V/mV		
		0°C	3	11.1	4	13.5			
		70°C	4	13.3	5	15.2			
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub> Input capacitance		25°C	5		14		pF		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70	87	75	94	dB		
		0°C	70	87	75	94			
		70°C	70	87	75	94			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	96	75	96	dB		
		0°C	75	96	75	96			
		70°C	75	96	75	96			

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis for the TL032A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = 2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032C, TL032AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (two amplifiers)	V <sub>O</sub> = 0, No load	25°C	3.8 5			13 17			mW
		0°C	3.7 5			12.7 17			
		70°C	3.8 5			12.6 17			
I <sub>CC</sub> Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	0°C	368 500			422 560			mA
		70°C	378 500			420 560			
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>V</sub> D = 100 dB	25°C	120			120			dB

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032C, TL032AC						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR <sub>+</sub> Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	12			1.5 2.9			V/μs	
		0°C	1.8			1 2.6				
		70°C	2.2			1.5 3.2				
SR <sub>-</sub> Negative slew rate at unity gain		25°C	3.9			1.5 5.1			V/μs	
		0°C	3.7			1.5 5				
		70°C	4			1.5 5				
t <sub>r</sub> Rise time	V <sub>I</sub> (PP) = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	138			132			ns	
		0°C	134			127				
		70°C	150			142				
t <sub>f</sub> Fall time		25°C	138			132			ns	
		0°C	134			127				
		70°C	150			142				
Overshoot factor		25°C	11%			5%				
		0°C	10%			4%				
		70°C	12%			6%				
V <sub>n</sub> Equivalent input noise voltage (see Note 9)	TL032C	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			49 49			nV/√Hz
			f = 1 kHz	41 41						
	TL032AC		f = 10 Hz	25°C			49 49			
			f = 1 kHz	41 41			60			
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C	0.003			0.003			pA/√Hz	
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C	1			1.1			MHz	
		0°C	1			1.1				
		70°C	1			1				
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C	61°			65°				
		0°C	61°			65°				
		70°C	60°			64°				

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



# TL032, TL032A

## ENHANCED-JFET LOW-POWER LOW-OFFSET

### DUAL OPERATIONAL AMPLIFIERS

SLOS033D – JULY 1988 – REVISED AUGUST 1994

#### electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL032I, TL032AI						UNIT			
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V						
			MIN	TYP	MAX	MIN	TYP	MAX				
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL032I	25°C	0.69		3.5		0.57		1.5		mV
			Full range			5.3				3.3		
		TL032AI	25°C	0.53		2.8		0.39		0.8		
			Full range			4.6				2.6		
α <sub>VIO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL032I	25°C to 85°C	11.4				10.8				μV/°C
		TL032AI	25°C to 85°C	11.4				10.8		25		
Input offset voltage long-term drift (see Note 6)		25°C	0.04				0.04				μV/mo	
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	1		100		1		100		pA	
		85°C	0.02		0.45		0.02		0.45		nA	
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	2		200		2		200		pA	
		85°C	0.2		0.9		0.3		0.9		nA	
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4			-11.5 to 14	-13.4 to 15.4			V	
		Full range	-1.5 to 4					-11.5 to 14				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3		4.3		13		14		V	
		-40°C	3		4.2		13		14			
		85°C	3		4.4		13		14			
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-3		-4.2		-12.5		-13.9		V	
		-40°C	-3		-4.1		-12.5		-13.8			
		85°C	-3		-4.2		-12.5		-14			
A <sub>VD</sub> voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7	-40°C	3		8.4		4		11.6		V/mV	
		85°C	4		13.5		5		15.3			
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>				10 <sup>12</sup>				Ω	
c <sub>i</sub> Input capacitance		25°C	5				4				pF	
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70		87		75		94		dB	
		-40°C	70		87		75		94			
		85°C	70		87		75		94			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75		96		75		96		dB	
		-40°C	75		96		75		96			
		85°C	75		96		75		96			

† Full range is -40°C to 85°C.

NOTES: 5. This parameter is tested on a sample basis for the TL032A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = 2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.





**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032I, TL032AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (two amplifiers)	V <sub>O</sub> = 0, No load	25°C		3.8	5		13	17	mW
		-40°C		2.9	5		10.9	17	
		85°C		3.7	5		12.4	17	
I <sub>CC</sub> Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C		384	500		434	560	μA
		-40°C		288	500		362	560	
		85°C		372	500		414	560	
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>VD</sub> = 100 dB	25°C		120		120		dB	

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032I, TL032AI						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+ Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C		2		1.5	2.9	V/μs		
		-40°C		1.6		1	2.1			
		85°C		2.3		1.5	3.3			
SR- Negative slew rate at unity gain		R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C		3.9		1.5	5.1	V/μs	
			-40°C		3.3		1.5	4.8		
			85°C		4.1		1.5	4.9		
t <sub>r</sub> Rise time	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		138		132	ns		
			-40°C		132		123			
			85°C		154		146			
t <sub>f</sub> Fall time		V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns		
			-40°C		132		123			
			85°C		154		146			
Overshoot factor			V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		11%		5%		
				-40°C		12%		5%		
				85°C		13%		7%		
V <sub>n</sub> Equivalent input noise voltage (see Note 9)	R <sub>S</sub> = 20 Ω, See Figure 3			25°C	TL032I	f = 10 Hz		49		nV/√Hz
						f = 1 kHz		41		
				25°C	TL032AI	f = 10 Hz		49		
		f = 1 kHz					41		41	
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C			0.003		0.003	pA/√Hz		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C			1		1.1	MHz		
		-40°C		1		1.1				
		85°C		0.9		1				
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C		61°		65°				
		-40°C		61°		65°				
		85°C		60°		64°				

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



# TL032, TL032A

## ENHANCED-JFET LOW-POWER LOW-OFFSET DUAL OPERATIONAL AMPLIFIERS

SLOS033D – JULY 1988 – REVISED AUGUST 1994

### electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL032M, TL032AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	25°C	TL032M	0.69	3.5	0.57	1.5	mV	
			Full range	6.5			4.5		
			TL032AM	0.53	2.8	0.39	0.8		
			Full range	5.8			3.8		
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	25°C to 125°C	TL032M	9.7			9.7	μV/°C	
			TL032AM	9.7			9.7		
	Input offset voltage long-term drift (see Note 6)	25°C	0.04			0.04	μV/mo		
I <sub>IO</sub>	Input offset current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	1	100	1	100	pA	
			125°C	0.2	10	0.2	10	nA	
I <sub>IB</sub>	Input bias current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	2	200	2	200	pA	
			125°C	7	20	8	20	nA	
V <sub>ICR</sub>	Common-mode input voltage range	25°C		-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V	
			Full range	-1.5 to 4		-11.5 to 14			
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3	4.3	13	14	V	
			-55°C	3	4.1	13	14		
			125°C	3	4.4	13	14		
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9	V	
			-55°C	-3	-4	-12.5	-13.8		
			125°C	-3	-4.3	-12.5	-14		
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 7	25°C	4	12	5	14.3	V/mV	
			-55°C	3	7.1	4	10.4		
			125°C	3	12.9	4	15		
r <sub>i</sub>	Input resistance	25°C	10 <sup>12</sup>			10 <sup>12</sup>	Ω		
c <sub>i</sub>	Input capacitance	25°C	5			4	pF		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70	87	75	94	dB	
			-55°C	70	87	70	94		
			125°C	70	87	70	94		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	96	75	96	dB	
			-55°C	75	95	75	95		
			125°C	75	96	75	96		

† Full range is -55°C to 125°C.

NOTES: 6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = 2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032M, TL032AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (two amplifiers)	V <sub>O</sub> = 0, No load	25°C	3.8 5			13 17			mW
		-55°C	2.3 5			9.4 17			
		125°C	3.6 5			11.8 17			
I <sub>CC</sub> Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C	384 500			434 560			μA
		-55°C	228 500			312 560			
		125°C	356 500			394 560			
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>VD</sub> = 100 dB	25°C	120			120			dB

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL032M, TL032AM						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR <sub>+</sub> Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C	2			1.5 2.9			V/μs	
		-55°C	1.4			1 1.9				
		125°C	2.4			1 3.5				
SR <sub>-</sub> Negative slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Note 8 and Figure 1	25°C	3.9			1.5 5.1			V/μs	
		-55°C	3.2			1 4.6				
		125°C	4.1			1 4.7				
t <sub>r</sub> Rise time	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	138			132			ns	
		-55°C	142			123				
		125°C	166			58				
t <sub>f</sub> Fall time	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	138			132			ns	
		-55°C	142			123				
		125°C	166			158				
Overshoot factor	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	11%			5%				
		-55°C	16%			6%				
		125°C	14%			8%				
V <sub>n</sub> Equivalent input noise voltage	TL032M	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	49			49			nV/√Hz
			f = 1 kHz	41			41			
	TL032AM	f = 10 Hz	49			49				
		f = 1 kHz	41			41				
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C	0.003			0.003			pA/√Hz	
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C	1			1.1			MHz	
		-55°C	1			1.1				
		125°C	0.9			0.9				
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C	61°			65°				
		-55°C	57°			64°				
		125°C	59°			62°				

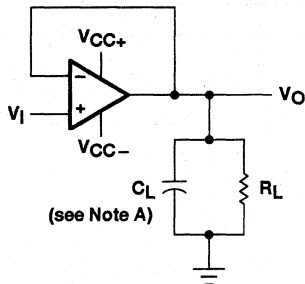
NOTE: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.



**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

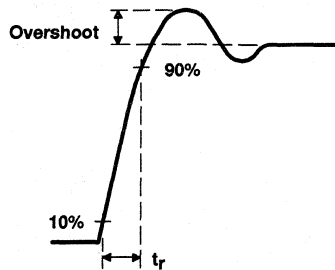
SLOS033D - JULY 1988 - REVISED AUGUST 1994

**PARAMETER MEASUREMENT INFORMATION**

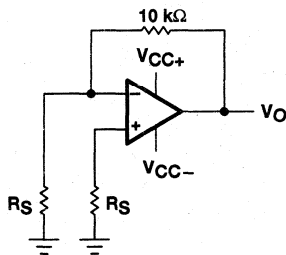


NOTE A:  $C_L$  includes fixture capacitance.

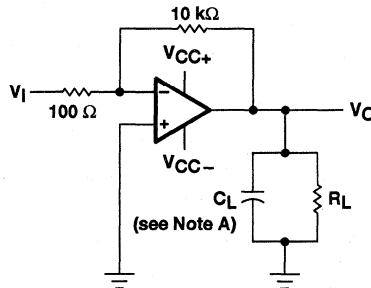
**Figure 1. Slew Rate, Rise/Fall Time, and Overshoot Test Circuit**



**Figure 2. Rise Time and Overshoot Waveform**

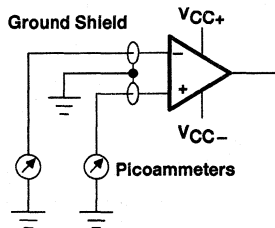


**Figure 3. Noise-Voltage Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit**



**Figure 5. Input-Bias and Offset-Current Test Circuit**

### typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

### input bias and offset current

At the picoamp bias-current level typical of the TL032 and TL032A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

### noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D – JULY 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Input bias current	vs Free-air temperature vs Common-mode input voltage	8 9
$V_{IC}$	Common-mode input voltage range	vs Supply voltage vs Free-air temperature	10 11
$V_O$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage	vs Supply voltage vs Output current vs Free-air temperature	14 16, 17 18, 19
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	15
$A_{VD}$	Large-signal differential voltage amplification	vs Load resistance vs Frequency vs Free-air temperature	20 21 22
$z_o$	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency vs Free-air temperature	24, 25 26
kSVR	Supply-voltage rejection ratio	vs Free-air temperature	27
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Time vs Free-air temperature	28 29 30
$V_n$	Equivalent input noise voltage	vs Frequency	31
$I_{CC}$	Supply current	vs Supply voltage vs Free-air temperature	32 33
SR	Slew rate	vs Load resistance vs Free-air temperature	34, 35 36, 37
	Overshoot factor	vs Load capacitance	38
THD	Total harmonic distortion	vs Frequency	39
$B_1$	Unity-gain bandwidth	vs Supply voltage vs Free-air temperature	40 41
$\phi_m$	Phase margin	vs Supply voltage vs Load capacitance vs Free-air temperature	42 43 44
	Pulse response	Small signal Large signal	45 46, 47
	Phase shift	vs Frequency	21



**TYPICAL CHARACTERISTICS†**

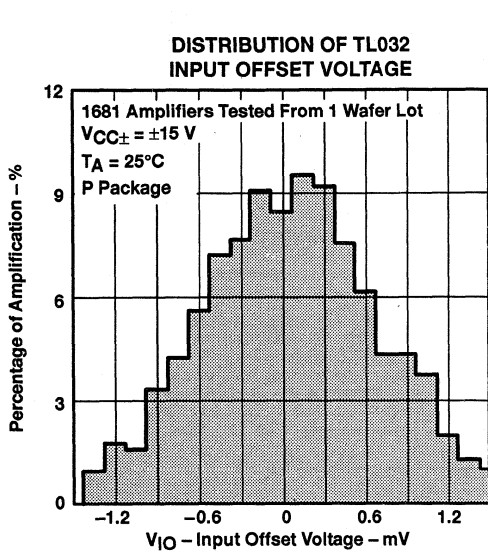


Figure 6

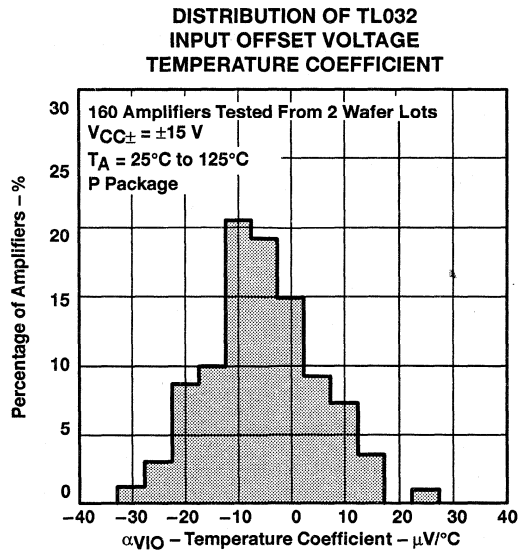


Figure 7

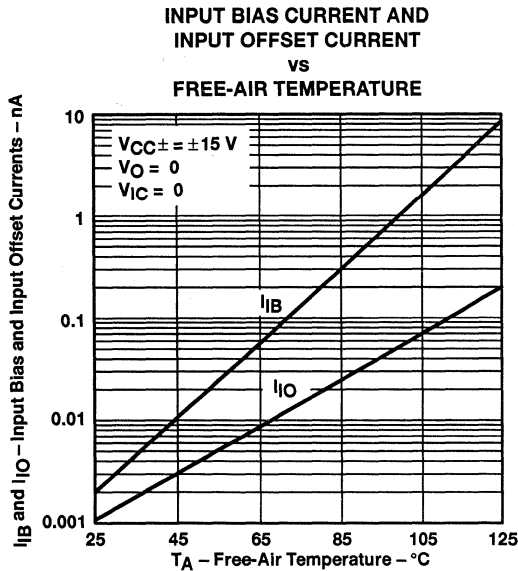


Figure 8

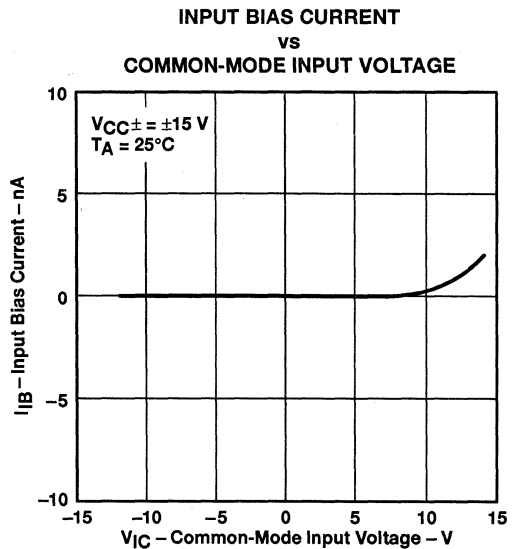


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

**COMMON-MODE  
 INPUT VOLTAGE RANGE LIMITS  
 vs  
 SUPPLY VOLTAGE**

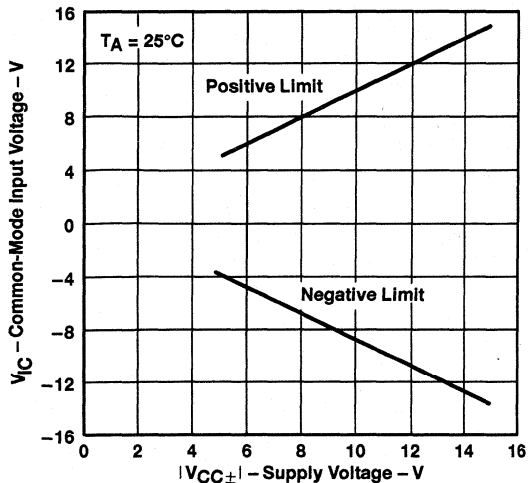


Figure 10

**COMMON-MODE  
 INPUT VOLTAGE RANGE LIMITS  
 vs  
 FREE-AIR TEMPERATURE**

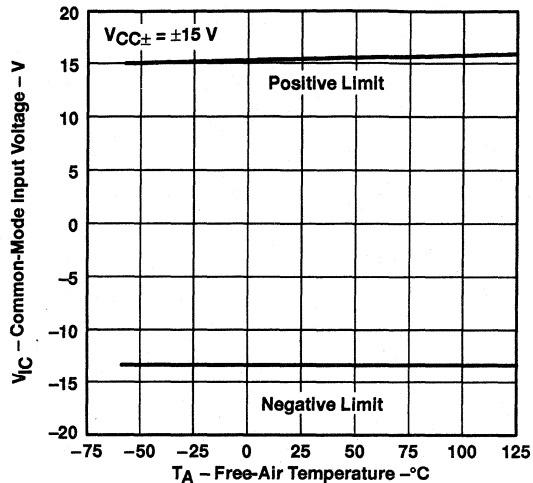


Figure 11

**OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE**

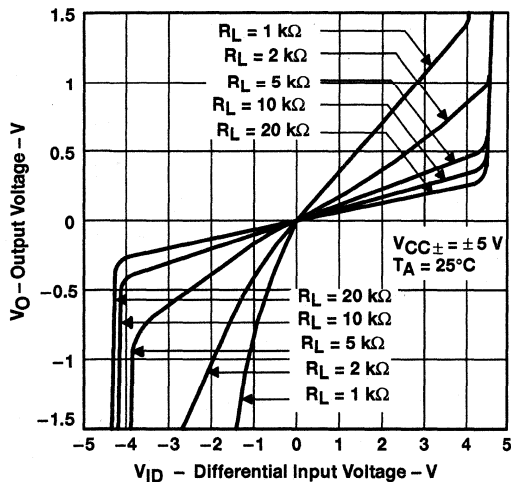


Figure 12

**OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE**

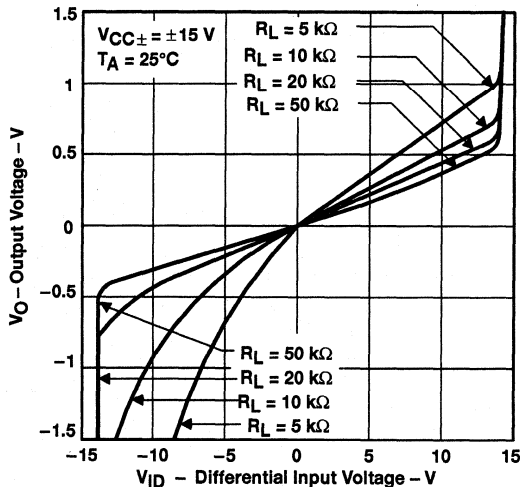


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE

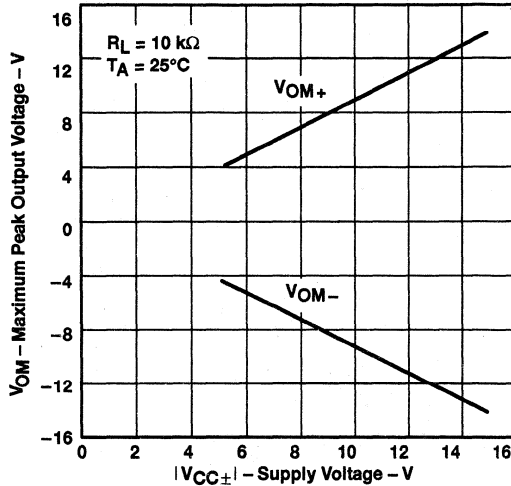


Figure 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY

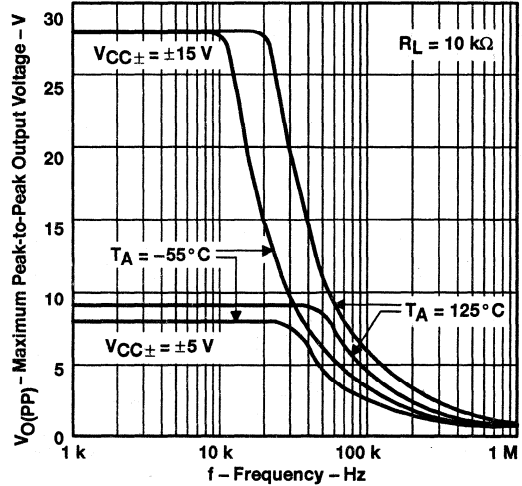


Figure 15

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

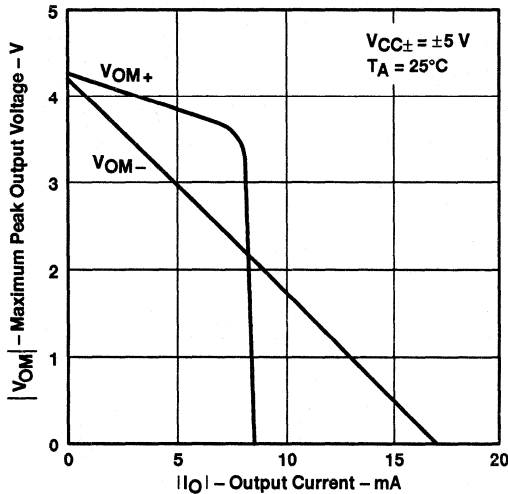


Figure 16

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

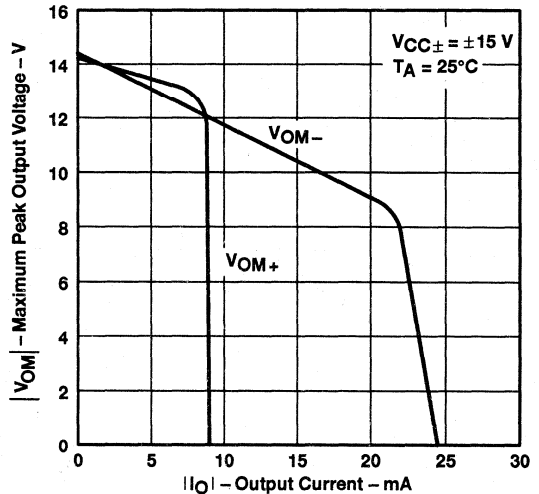


Figure 17

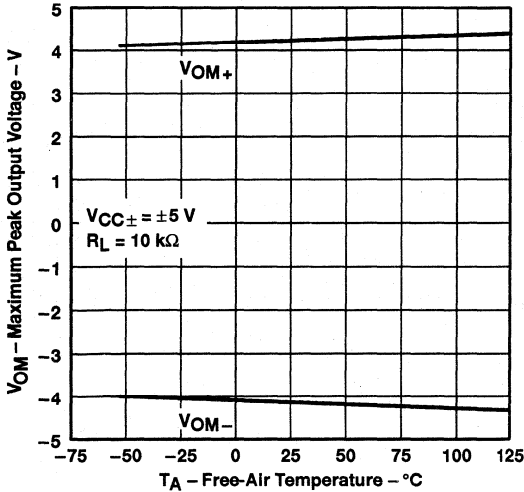
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

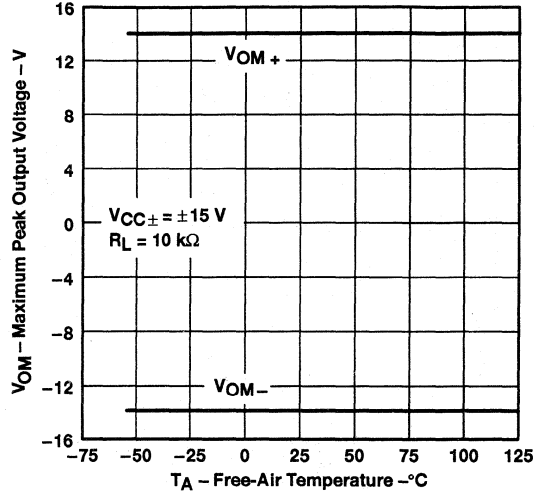
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**VS**  
**FREE-AIR TEMPERATURE**



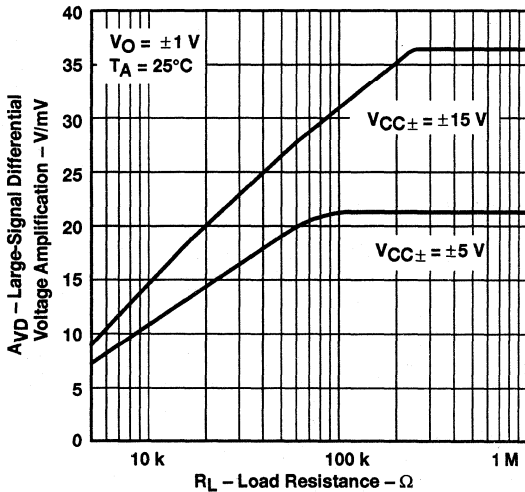
**Figure 18**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**VS**  
**FREE-AIR TEMPERATURE**



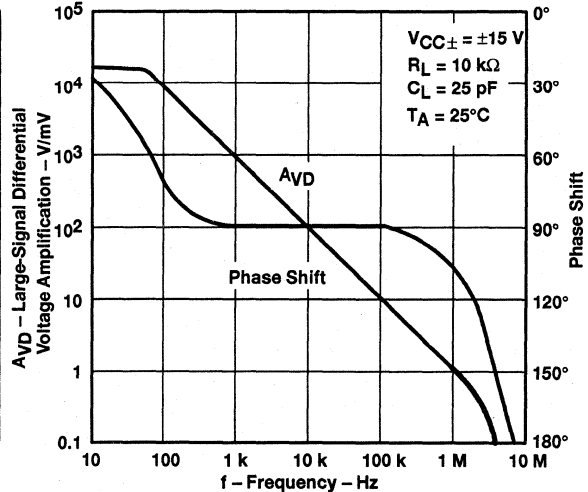
**Figure 19**

**LARGE-SIGNAL DIFFERENTIAL**  
**VOLTAGE AMPLIFICATION**  
**VS**  
**LOAD RESISTANCE**



**Figure 20**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION AND PHASE SHIFT**  
**VS**  
**FREQUENCY**



**Figure 21**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

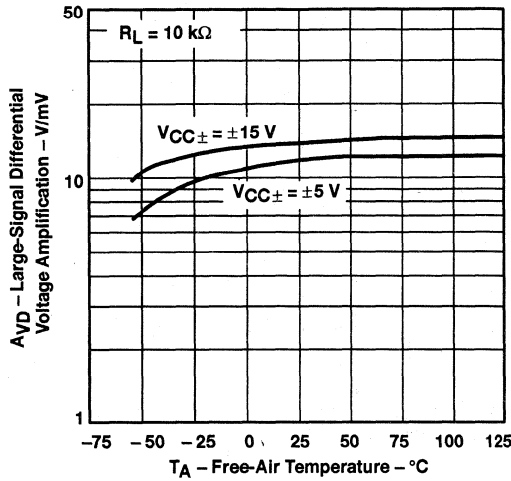


Figure 22

OUTPUT IMPEDANCE  
 VS  
 FREQUENCY

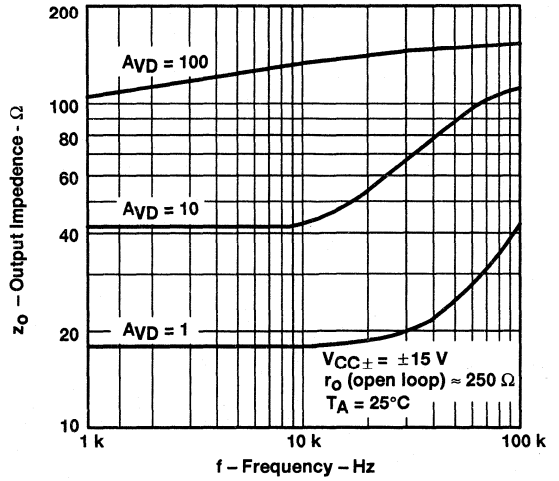


Figure 23

COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY

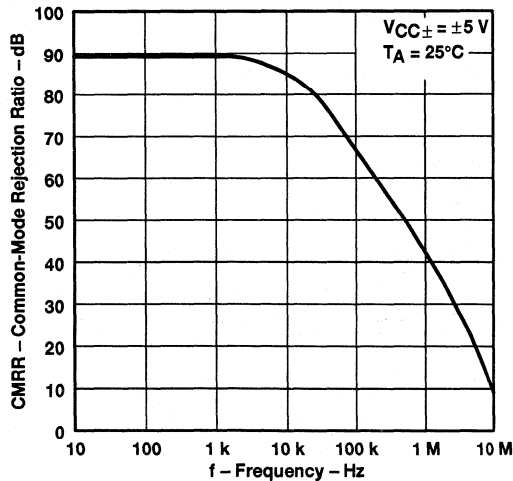


Figure 24

COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY

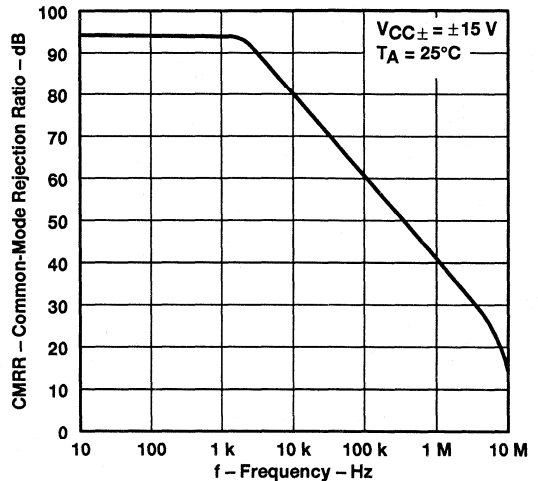


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

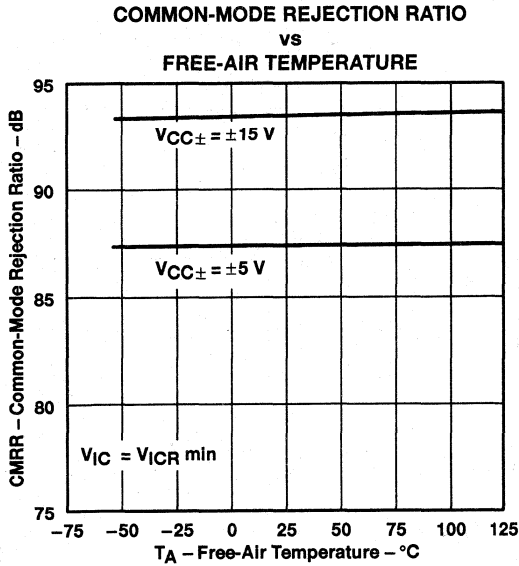


Figure 26

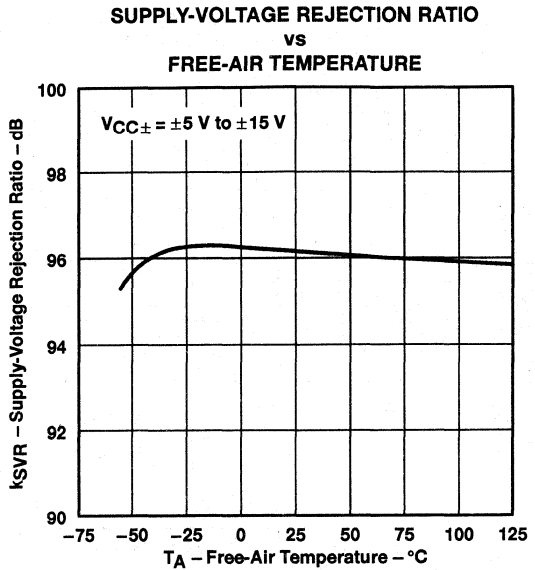


Figure 27

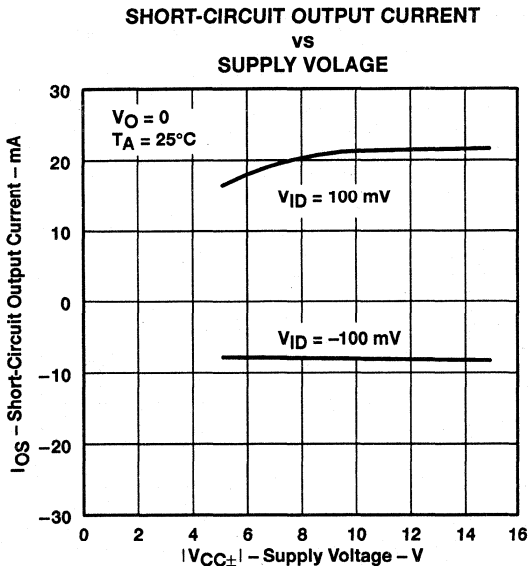


Figure 28

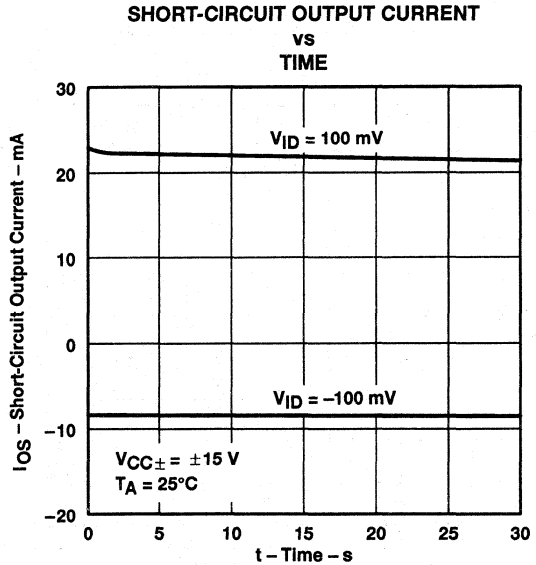


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

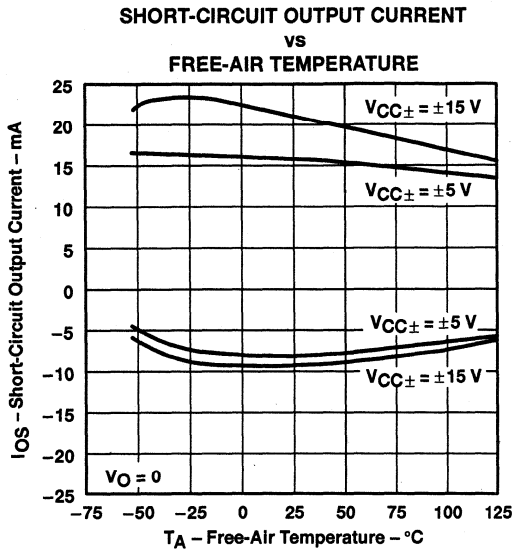


Figure 30

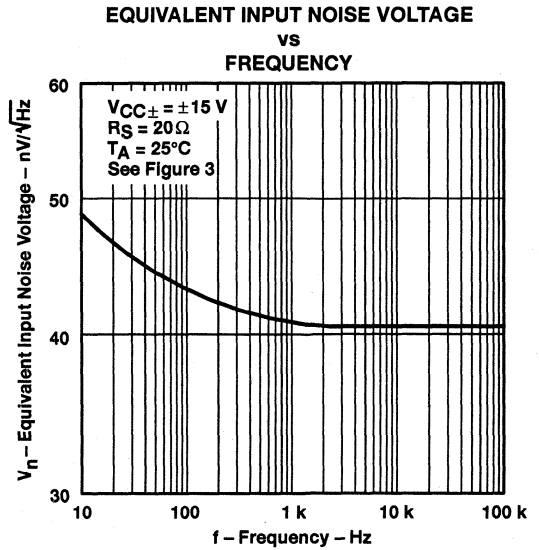


Figure 31

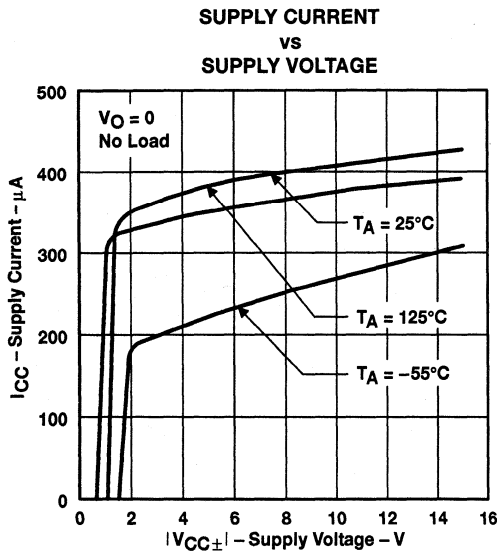


Figure 32

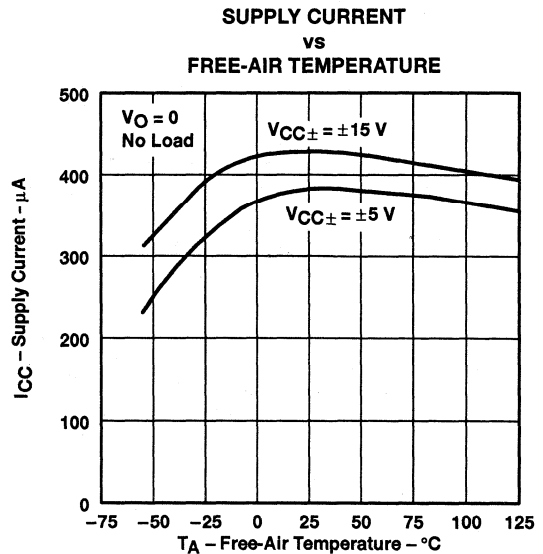


Figure 33

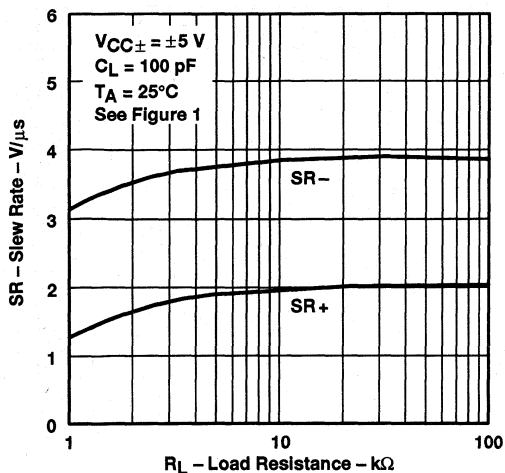
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

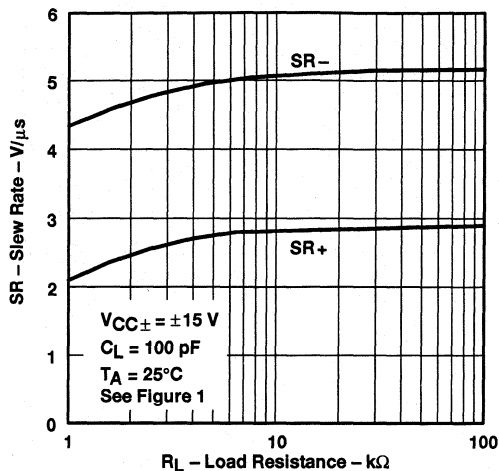
**TYPICAL CHARACTERISTICS†**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**



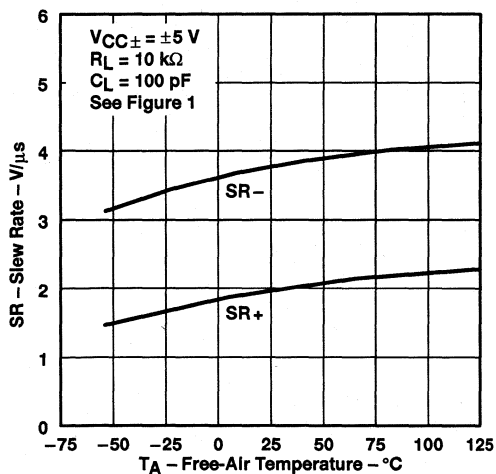
**Figure 34**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**



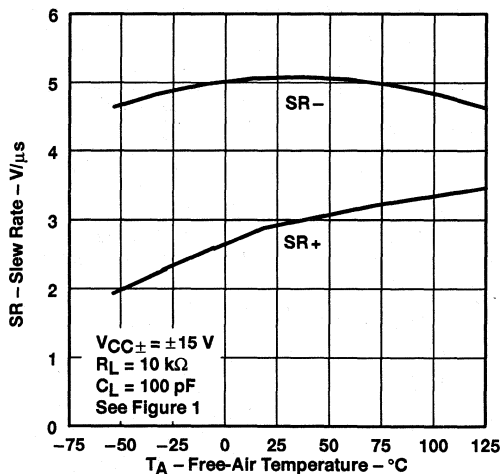
**Figure 35**

**SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 36**

**SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**

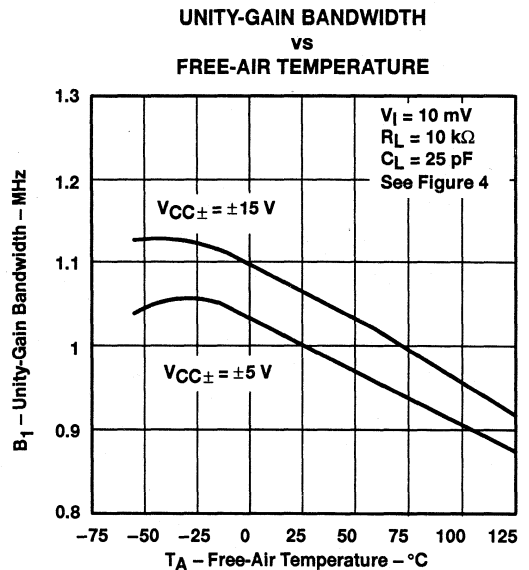
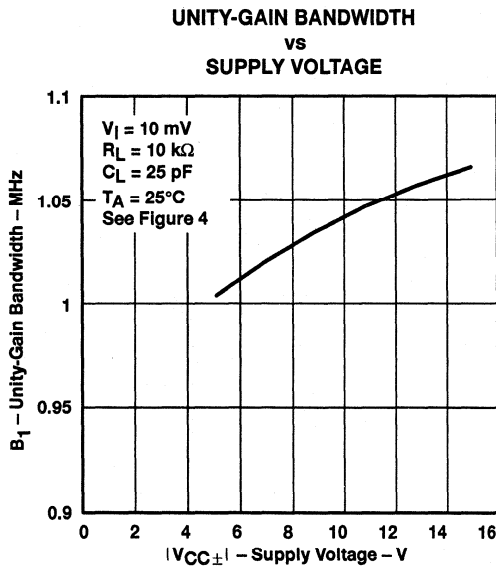
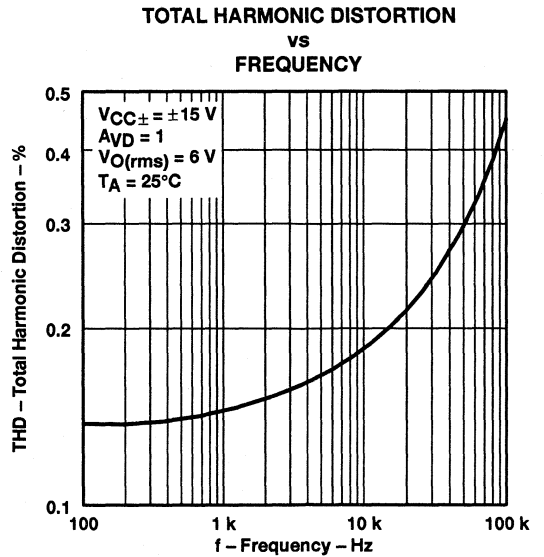
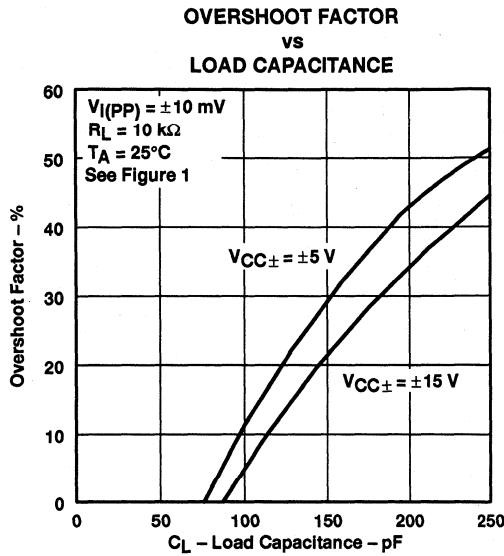


**Figure 37**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†



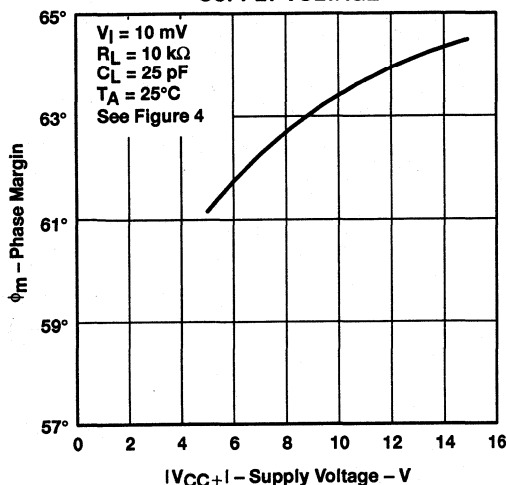
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D - JULY 1988 - REVISED AUGUST 1994

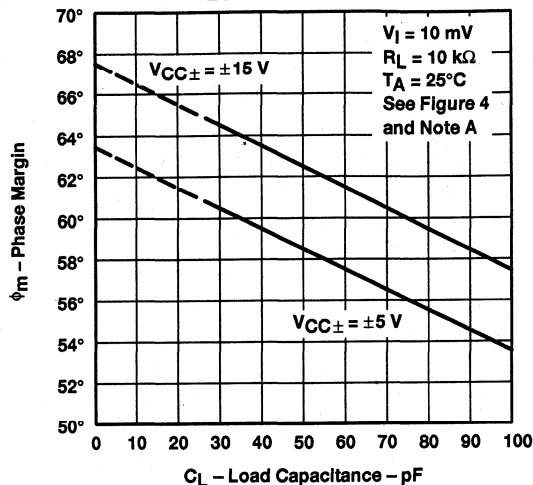
**TYPICAL CHARACTERISTICS†**

**PHASE MARGIN**  
**VS**  
**SUPPLY VOLTAGE**



**Figure 42**

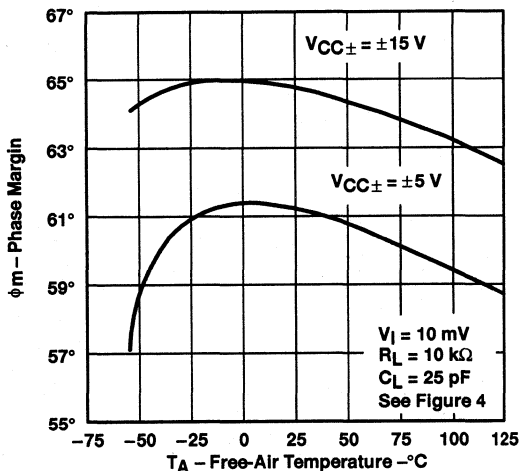
**PHASE MARGIN**  
**VS**  
**LOAD CAPACITANCE**



NOTE A: Values of phase margin below a load capacitance of 25 pF were estimated.

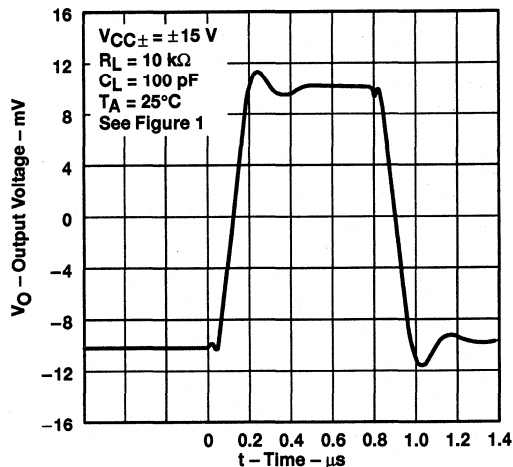
**Figure 43**

**PHASE MARGIN**  
**VS**  
**FREE-AIR TEMPERATURE**



**Figure 44**

**VOLTAGE-FOLLOWER**  
**SMALL-SIGNAL**  
**PULSE RESPONSE**



**Figure 45**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**

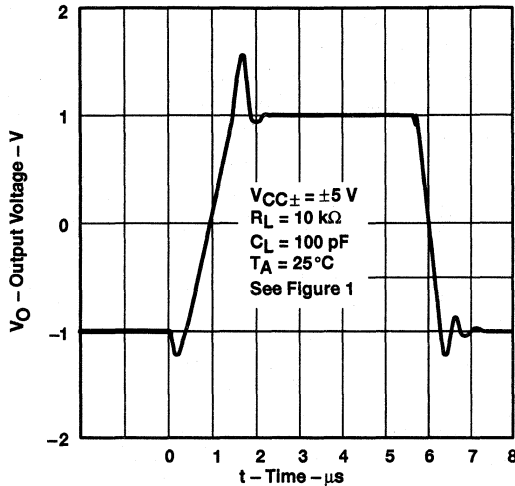


Figure 46

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**

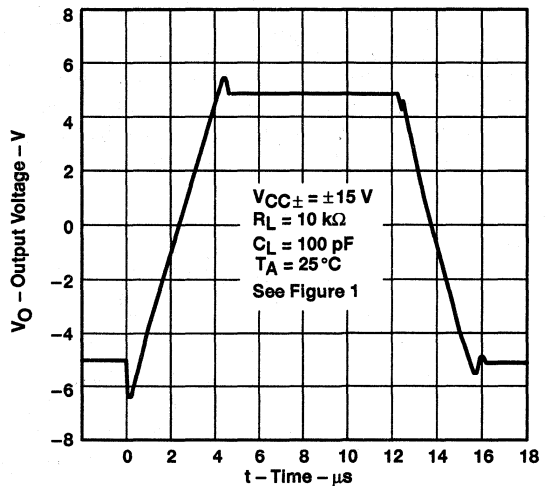


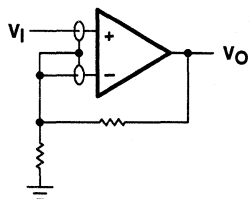
Figure 47

**APPLICATION INFORMATION**

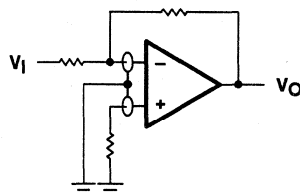
**input characteristics**

The TL032 and TL032A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Because of the extremely high input impedance and resulting low bias current requirements, the TL032 and TL032A are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

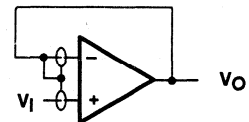
Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



(c) UNITY-GAIN AMPLIFIER

Figure 48. Use of Guard Rings

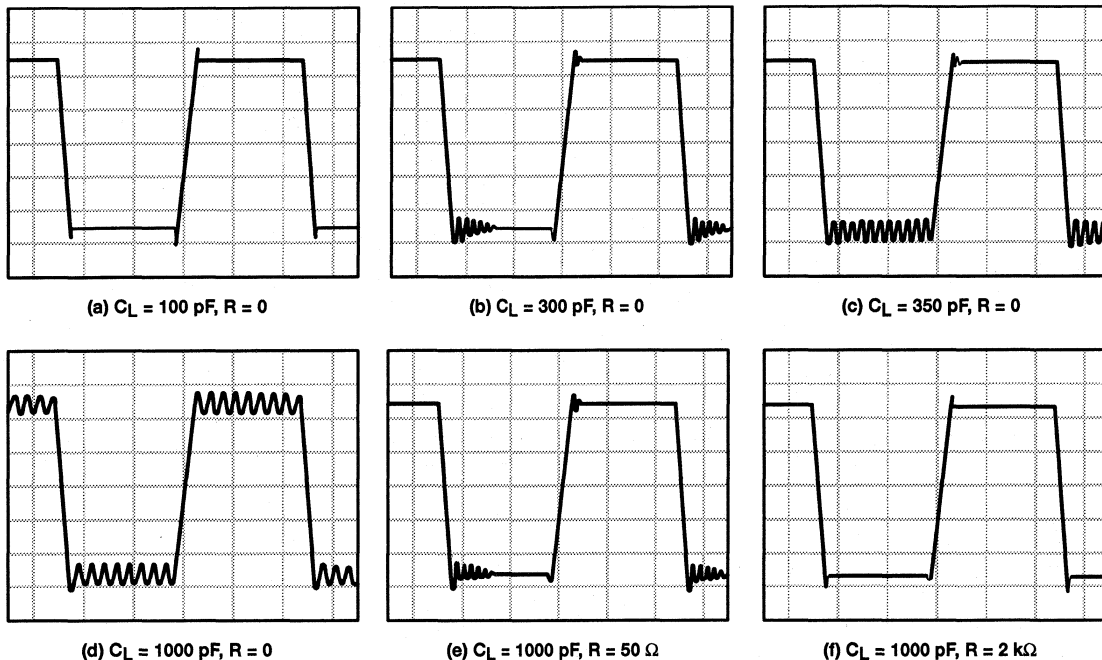
**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D – JULY 1988 – REVISED AUGUST 1994

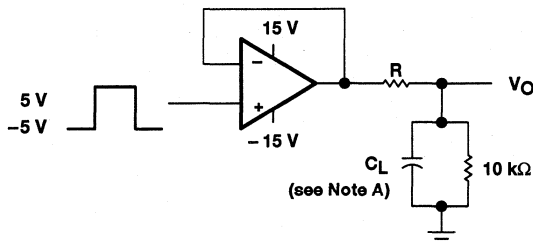
**APPLICATION INFORMATION**

**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL032 and TL032A will drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem (see Figure 50). Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 49).



**Figure 49. Effect of Capacitive Loads**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 50. Test Circuit for Output Characteristics**

## APPLICATION INFORMATION

### high-Q notch filter

In general, Texas Instruments enhanced JFET operational amplifiers serve as excellent filters. This circuit provides a narrow notch at a specific frequency. Notch filters are designed to eliminate frequencies that are interfering with the operation of an application. For this filter, the center frequency can be calculated as:

$$f_O = \frac{1}{2\pi R_1 C_1}$$

With the resistors and capacitors shown in Figure 51, the center frequency is 1 kHz.  $C_1 = C_3 = C_2 + 2$  and  $R_1 = R_3 = 2 \times R_2$ . The center frequency can be modified by varying these values. When adjusting the center frequency, be sure that the operational amplifier still has sufficient gain at the frequency required.

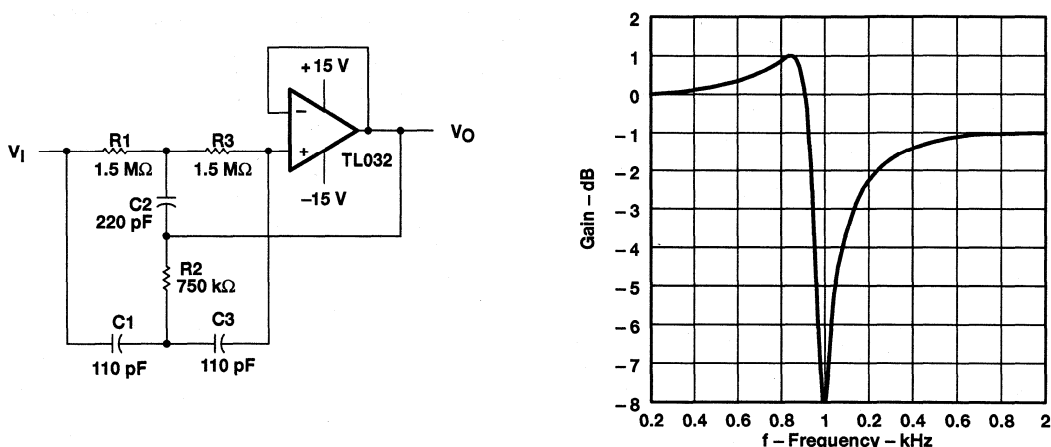


Figure 51. High-Q Notch Filter

### 2-wire 4-mA to 20-mA current loop

Often information from an analog sensor must be sent over a distance to the receiving circuitry. For many applications, the most feasible method involves converting voltage information to a current before transmission. The following circuit benefits from the high input impedance of the TL032A since many inexpensive sensors do not have low output impedance.

Assuming that the voltage at the TL032A's noninverting input is zero, the following equation determines the output current:

$$I_O = V_I \left( \frac{R_3}{R_1 \times R_S} \right) + 5 \text{ V} \left( \frac{R_3}{R_2 \times R_S} \right) = 0.16 \times V_I + 4 \text{ mA}$$

The current presently provides 4-mA to 20-mA output for an input voltage of 0 to 100 mV. By modifying  $R_1$ ,  $R_2$ , and  $R_3$ , the input voltage range or the output current range can be adjusted.

**TL032, TL032A**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS033D – JULY 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

**2-wire 4-mA to 20-mA current loop (continued)**

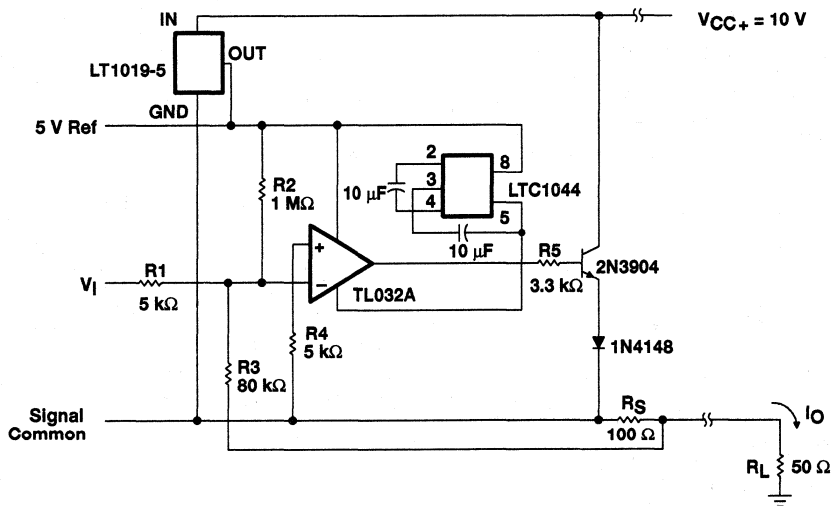
Including the offset voltage of the operational amplifier in the following equation clearly illustrates why the low offset TL032A was chosen:

$$I_O = V_I \left( \frac{R_3}{R_1 \times R_S} \right) + 5 \text{ V} \left( \frac{R_3}{R_2 \times R_S} \right) - V_{IO} \left( \frac{R_3}{R_1 \times R_S} + \frac{R_3}{R_2 \times R_S} + \frac{R_1}{R_S} \right)$$

$$= 0.16 \times V_I + 4 \text{ mA} - 0.17 \times V_{IO}$$

For example, an offset voltage of 1 mV decreases the output current by 0.17 mA.

Because of the low-power consumption of the TL032A, this circuit has at least 2 mA available to drive the actual sensor from the 5-V reference node.



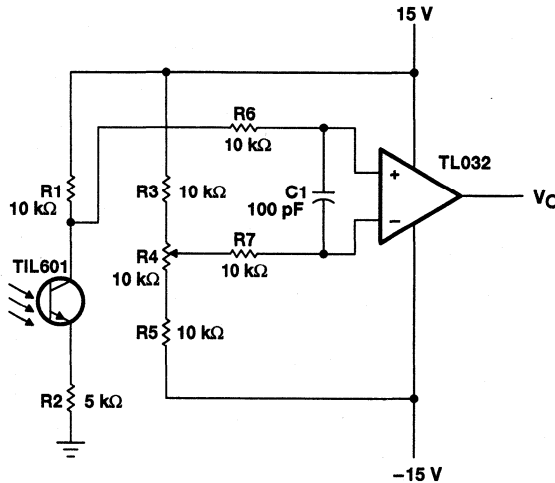
**Figure 52. 2-Wire 4-mA to 20-mA Current Loop**

**APPLICATION INFORMATION**

**low-level light detector preamplifier**

Applications that need to detect small currents require high input-impedance operational amplifiers; otherwise, the bias currents of the operational amplifier camouflage the current being monitored. Phototransistors provide a current that is proportional to the light reaching the transistor. The TL032 allows even the small currents resulting from low-level light to be detected.

In Figure 53, if there is no light, the phototransistor is off and the output is high. As light is detected, the operational amplifier output begins pulling low. Adjusting R4 both compensates for offset voltage of the amplifier and adjusts the point of light detection by the amplifier.



**Figure 53. Low-Level Light Detector Preamplifier**



# TL034, TL034A, TL034Y ENHANCED-JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

SLOS034D – JULY 1988 – REVISED AUGUST 1994

- **Maximum Offset Voltage . . . 1.5 mV**
- **High Slew Rate . . . 2.9 V/ $\mu$ s Typ**
- **Low Input Bias Current . . . 2 pA Typ**
- **Very Low Power Consumption  
26 mW Typ**
- **Output Short-Circuit Protection**
- **Monolithic Construction**

## description

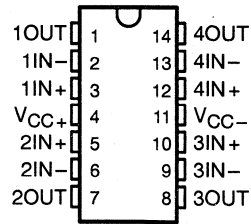
The TL034 and TL034A quadruple operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of the TI enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability, but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages, coupled with low power consumption, make the TL034 well suited for new state-of-the-art designs as well as existing design upgrades. The TL034 has been designed to be functionally compatible and pin compatible with the TL064. Two offset voltage grades are available: TL034 (4 mV max) and TL034A (1.5 mV max).

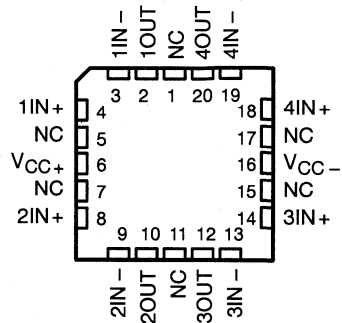
A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

**D, J, N, OR PW PACKAGE  
(TOP VIEW)**

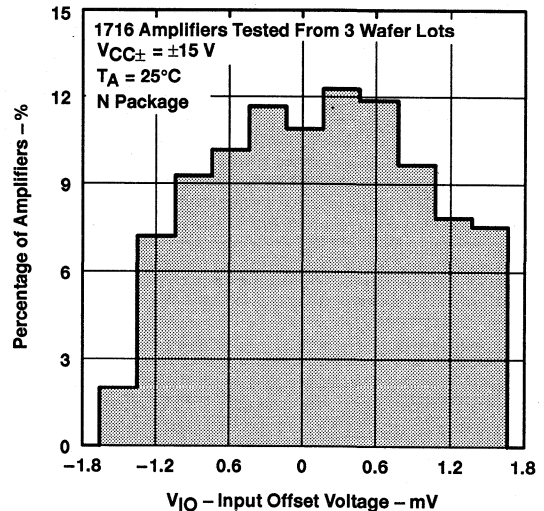


**FK PACKAGE  
(TOP VIEW)**



NC – No internal connection

**DISTRIBUTION OF TL034A  
INPUT OFFSET VOLTAGE**



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



# TL034, TL034A, TL034Y ENHANCED-JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

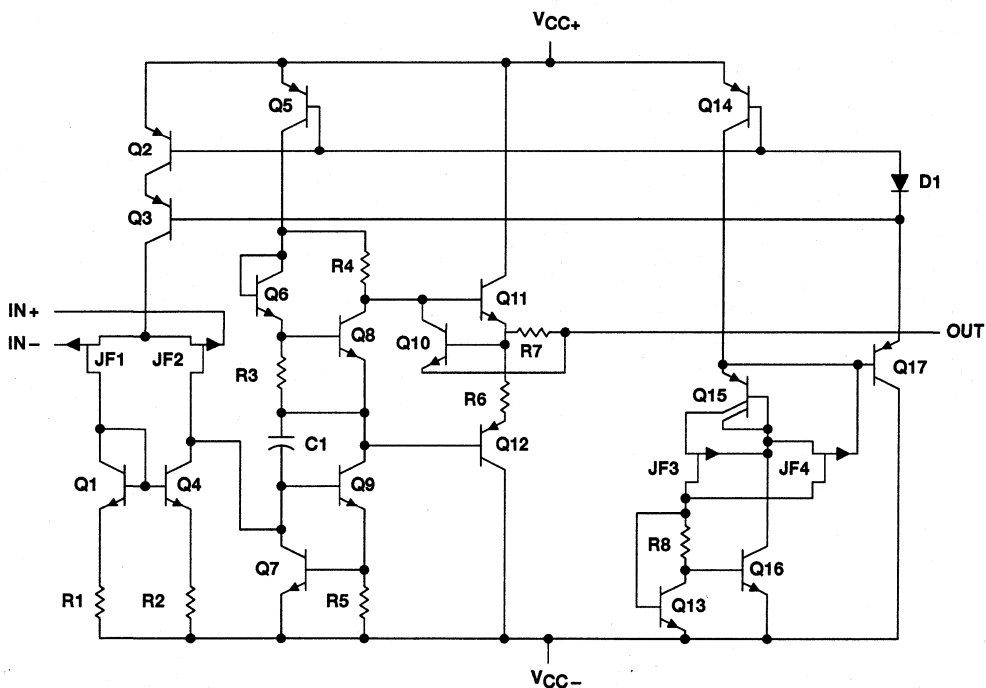
SLOS034D - JULY 1988 - REVISED AUGUST 1994

## AVAILABLE OPTIONS

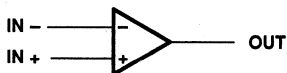
T <sub>A</sub>	V <sub>I</sub> Omax AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	1.5 mV	TL034ACD	—	—	TL034ACN	—	TL034Y
	4 mV	TL034CD	—	—	TL04CN	TL034CPW	
-40°C to 85°C	1.5 mV	TL034AID	—	—	TL034AIN	—	—
	4 mV	TL034ID	—	—	TL034IN	—	
-55°C to 125°C	1.5 mV	TL034AMD	TL034AMFK	TL034AMJ	TL034AMN	—	—
	4 mV	TL034MD	TL034MFK	TL034MJ	TL034MN	—	

The D packages are available taped and reeled. Add R suffix to device type (e.g., TL034CDR).

## equivalent schematic (each amplifier)



## symbol (each amplifier)



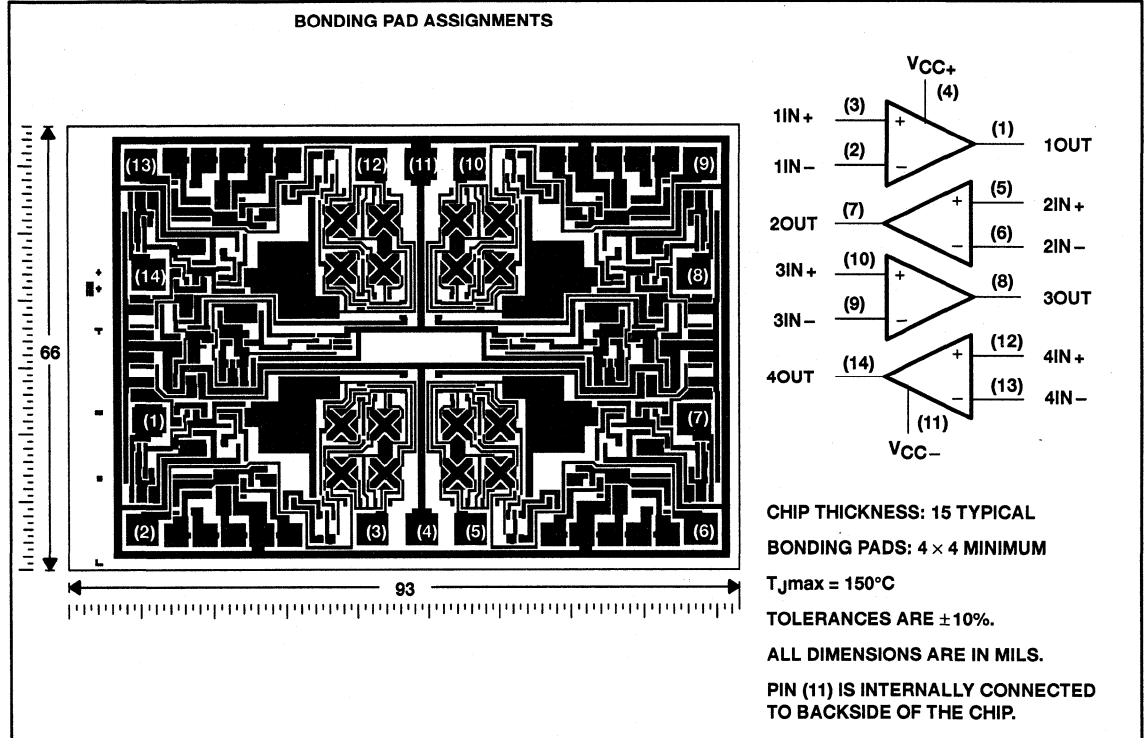


**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D - JULY 1988 - REVISED AUGUST 1994

**TL034 chip information**

This chip, when properly assembled, displays characteristics similar to the TLC34C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$ V
Input voltage, $V_I$ (any input) (see Notes 1 and 3)	$\pm 15$ V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$ (each output)	$\pm 40$ mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	230 mW
PW	700 mW	5.6 mW/°C	448 mW	N/A	N/A

**recommended operating conditions**

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC\pm}$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5$ V		-1.5	4	-1.5	4	V
	$V_{CC\pm} = \pm 15$ V		-11.5	14	-11.5	14	
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL034C, TL034AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL034C	25°C	0.91	6	0.79	4	mV	
			Full range	8.2			6.2		
		TL034AC	25°C	0.7	3.5	0.58	1.5		
			Full range	5.7			3.7		
α <sub>VIO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL034C	25°C to 70°C	11.6		12		μV/°C	
		TL034AC	25°C to 70°C	11.6		12	25		
Input offset voltage long-term drift (see Note 6)			25°C	0.04		0.04		μV/mo	
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5		25°C	1	100	1	100	pA	
			70°C	9	200	12	200		
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5		25°C	2	200	2	200	pA	
			70°C	50	400	80	400		
V <sub>ICR</sub> Common-mode input voltage range			25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4	V	
			Full range	-1.5 to 4		-11.5 to 14			
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3	4.3	13	14	V	
			0°C	3	4.2	13	14		
			70°C	3	4.3	13	14		
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-3	-4.2	-12.5	-13.9	V	
			0°C	-3	-4.1	-12.5	-13.9		
			70°C	-3	-4.2	-12.5	-14		
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 6		25°C	4	12	5	14.3	V/mV	
			0°C	3	11.1	4	13.5		
			70°C	4	13.3	5	15.2		
r <sub>i</sub> Input resistance			25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω	
c <sub>i</sub> Input capacitance			25°C	5		14		pF	
CMRR      Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	70	87	75	94	dB	
			0°C	70	87	75	94		
			70°C	70	87	75	94		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0,      R <sub>S</sub> = 50 Ω		25°C	75	96	75	96	dB	
			0°C	75	96	75	96		
			70°C	75	96	75	96		

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL034, TL034A, TL034Y ENHANCED-JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

SLOS034D – JULY 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034C, TL034AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (two amplifiers)	V <sub>O</sub> = 0, No load	25°C		7.7	10		26	34	mW
		0°C		7.4	10		25.3	34	
		70°C		7.6	10		25.2	34	
I <sub>CC</sub> Supply current (four amplifiers)	V <sub>O</sub> = 0, No load	25°C		0.77	1		0.87	1.12	mA
		0°C		0.74	1		0.85	1.12	
		70°C		0.76	1		0.84	1.12	
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>V</sub> D = 100	25°C		120		120		dB	

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034C, TL034AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C		2		1.5	2.9	V/μs	
		0°C		1.8		1	2.6		
		70°C		2.2		1.5	3.2		
SR- Negative slew rate at unity gain		R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C		3.9		1.5	5.1	V/μs
			0°C		3.7		1.5	5	
			70°C		4		1.5	5	
t <sub>r</sub> Rise time	V <sub>I</sub> (PP) = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		138		132	ns	
			0°C		134		127		
			70°C		150		142		
t <sub>f</sub> Fall time		V <sub>I</sub> (PP) = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns	
			0°C		134		127		
			70°C		150		142		
Overshoot factor	V <sub>I</sub> (PP) = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		11%		5%		
			0°C		10%		4%		
			70°C		12%		6%		
V <sub>n</sub> Equivalent input noise voltage (see Note 9)		TL034C	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C	83	83	nV/√Hz	
				f = 1 kHz	43	43			
		TL034AC	f = 10 Hz	25°C	83	83			
	f = 1 kHz		43	43	60				
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C	0.003	0.003	pA/√Hz				
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		1		1.1	MHz		
		0°C		1		1.1			
		70°C		1		1			
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 10 kΩ, See Figure 4	25°C		61°		65°			
		0°C		61°		65°			
		70°C		60°		64°			

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL034I, TL034AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	25°C	TL034I			TL034AI			mV
			25°C			25°C			
			Full range			Full range			
			25°C			25°C			
α <sub>VIO</sub>	Temperature coefficient of input offset voltage (see Note 5)	25°C	TL034I			TL034AI			μV/°C
			25°C to 85°C			25°C to 85°C			
			Full range			Full range			
			25°C to 85°C			25°C to 85°C			
	Input offset voltage long-term drift (see Note 6)	25°C	0.04			0.04			μV/mo
I <sub>IO</sub>	Input offset current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5			V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5			pA
			85°C	0.02 0.45			0.02 0.45		
I <sub>IB</sub>	Input bias current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5			V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5			pA
			85°C	0.2 0.9			0.3 0.9		
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-1.5 to 4			-3.4 to 5.4			V
			Full range			Full range			
V <sub>OM+</sub>	Maximum positive peak output voltage swing	RL = 10 kΩ	25°C			25°C			V
			-40°C			-40°C			
			85°C			85°C			
V <sub>OM-</sub>	Maximum negative peak output voltage swing	RL = 10 kΩ	25°C			25°C			V
			-40°C			-40°C			
			85°C			85°C			
A <sub>VD</sub>	Large-signal differential voltage amplification	RL = 10 kΩ, See Note 7	-40°C			-40°C			V/mV
			85°C			85°C			
r <sub>i</sub>	Input resistance	25°C	10 <sup>12</sup>			10 <sup>12</sup>			Ω
c <sub>i</sub>	Input capacitance	25°C	5			4			pF
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			25°C			dB
			-40°C			-40°C			
			85°C			85°C			
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			25°C			dB
			-40°C			-40°C			
			85°C			85°C			

† Full range is -40°C to 85°C.

- NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.
6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.
7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL034, TL034A, TL034Y

## ENHANCED-JFET LOW-POWER LOW-OFFSET

### QUAD OPERATIONAL AMPLIFIERS

SLOS034D – JULY 1988 – REVISED AUGUST 1994

#### electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034I, TL034AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (four amplifiers)	V <sub>O</sub> = 0, No load	25°C		7.7	10		26	34	mW
		-40°C		5.8	10		21.7	34	
		85°C		7.4	10		24.8	34	
I <sub>CC</sub> Supply current (four amplifiers)	V <sub>O</sub> = 0, No load	25°C		0.77	1		0.87	1.12	μA
		-40°C		0.58	1		0.72	1.12	
		85°C		0.74	1		0.83	1.12	
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>VD</sub> = 100	25°C		120		120		dB	

#### operating characteristics

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034I, TL034AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C		2		1.5	2.9	V/μs	
		-40°C		1.6		1	2.1		
		85°C		2.3		1.5	3.3		
SR- Negative slew rate at unity gain		R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		3.9		1.5	5.1	V/μs
			-40°C		3.3		1.5	4.8	
			85°C		4.1		1.5	4.9	
t <sub>r</sub> Rise time	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		138		132	ns	
			-40°C		132		123		
			85°C		154		146		
t <sub>f</sub> Fall time		V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C		138		132	ns	
			-40°C		132		123		
			85°C		154		146		
Overshoot factor	V <sub>I(PP)</sub> = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2		25°C		11%		5%		
			-40°C		12%		5%		
			85°C		13%		7%		
V <sub>n</sub> Equivalent input noise voltage (see Note 9)		TL034I	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C		83		nV/√Hz
				f = 1 kHz	25°C		43		
		TL034AI		f = 10 Hz	25°C		83		
	f = 1 kHz			25°C		43			
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C		0.003		0.003	pA/√Hz		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C		1		1.1	MHz		
		-40°C		1		1.1			
		85°C		0.9		1			
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C		61°		65°			
		-40°C		61°		65°			
		85°C		60°		64°			

NOTES: 8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D - JULY 1988 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL034M, TL034AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	VO = 0, VIC = 0, RS = 50 Ω	25°C	TL034M		0.91	3.6	0.78		4
			Full range				11		9
		Full range	TL034AM		0.7	3.5	0.58		1.5
			Full range				8.5		6.5
αV <sub>IO</sub> Temperature coefficient of input offset voltage	VO = 0, VIC = 0, RS = 50 Ω	25°C to 125°C	TL034M		10.6		10.9		μV/°C
		25°C to 125°C	TL034AM		10.6		10.9		
Input offset voltage long-term drift (see Note 5)		25°C			0.04		0.04		μV/mo
I <sub>IO</sub> Input offset current	VO = 0, VIC = 0, See Figure 5	25°C			1	100	1	100	pA
		125°C			0.2	10	0.2	10	nA
I <sub>IB</sub> Input bias current	VO = 0, VIC = 0, See Figure 5	25°C			2	200	2	200	pA
		125°C			7	20	8	20	nA
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1.5 to 4	-3.4 to 5.4	-11.5 to 14	-13.4 to 15.4			V
		Full range	-1.5 to 4		-11.5 to 14				
V <sub>OM+</sub> Maximum positive peak output voltage swing	RL = 10 kΩ	25°C	3	4.3	13	14			V
		-55°C	3	4.1	13	14			
		125°C	3	4.4	13	14			
V <sub>OM-</sub> Maximum negative peak output voltage swing	RL = 10 kΩ	25°C	-3	-4.2	-12.5	-13.9			V
		-55°C	-3	-4	-12.5	-13.8			
		125°C	-3	-4.3	-12.5	-14			
A <sub>VD</sub> Large-signal differential voltage amplification	RL = 10 kΩ, See Note 7	25°C	4	12	5	14.3			V/mV
		-55°C	3	7.1	4	10.4			
		125°C	3	12.9	4	15			
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>				Ω
c <sub>i</sub> Input capacitance		25°C	5		4				pF
CMRR Common-mode rejection ratio	VIC = V <sub>ICRmin</sub> , VO = 0, RS = 50 Ω	25°C	70	87	75	94			dB
		-55°C	70	87	70	94			
		125°C	70	87	70	94			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	VO = 0, RS = 50 Ω	25°C	75	96	75	96			dB
		-55°C	75	95	75	95			
		125°C	75	96	75	96			

† Full range is -55°C to 125°C.

NOTES: 6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.

# TL034, TL034A, TL034Y ENHANCED-JFET LOW-POWER LOW-OFFSET QUAD OPERATIONAL AMPLIFIERS

SLOS034D – JULY 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034M, TL034AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
P <sub>D</sub> Total power dissipation (two amplifiers)	V <sub>O</sub> = 0, No load	25°C	7.7 10			26 34			mW
		-55°C	4.6 12			18.7 45			
		125°C	7.1 12			23.6 45			
I <sub>CC</sub> Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C	0.77 1			0.87 1.12			μA
		-55°C	0.46 1.2			0.62 1.5			
		125°C	0.71 1.2			0.79 1.5			
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>V</sub> D = 100	25°C	120			120			dB

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL034M, TL034AM						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+ Positive slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	2			1.5 2.9			V/μs	
		-55°C	1.4			1 1.9				
		125°C	2.4			1 3.5				
SR- Negative slew rate at unity gain		25°C	3.9			1.5 5.1			V/μs	
		-55°C	3.2			1 4.6				
		125°C	4.1			1 4.7				
t <sub>r</sub> Rise time	V <sub>I</sub> (PP) = ±10 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	138			132			ns	
		-55°C	142			123				
		125°C	166			58				
t <sub>f</sub> Fall time		25°C	138			132			ns	
		-55°C	142			123				
		125°C	166			158				
Overshoot factor	25°C	11%			5%					
	-55°C	16%			6%					
	125°C	14%			8%					
V <sub>n</sub> Equivalent input noise voltage	TL034M	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	83			83			nV/√Hz
			f = 1 kHz	43			43			
	TL034AM	f = 10 Hz	83			83				
		f = 1 kHz	43			43				
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C	0.003			0.003			pA/√Hz	
B1 Unity-gain bandwidth	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C	1			1.1			MHz	
		-55°C	1			1.1				
		125°C	0.9			0.9				
φ <sub>m</sub> Phase margin at unity gain	V <sub>I</sub> = 10 mV, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 25 pF, See Figure 4	25°C	61°			65°				
		-55°C	57°			64°				
		125°C	59°			62°				

NOTE 8: For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

**electrical characteristics,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TL034Y						UNIT		
		$V_{CC\pm} = \pm 5\text{ V}$			$V_{CC\pm} = \pm 15\text{ V}$					
		MIN	TYP	MAX	MIN	TYP	MAX			
$V_{IO}$ Input offset voltage	$V_O = 0, V_{IC} = 0,$ $R_S = 50\ \Omega$	0.91			6			0.79	4	mV
$\alpha V_{IO}$ Temperature coefficient of input offset voltage (see Note 5)		11.6			12					$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current	$V_O = 0, V_{IC} = 0,$ See Figure 5	1			100			1	100	pA
		2			200			2	200	pA
$I_{IB}$ Input bias current	$V_O = 0, V_{IC} = 0,$ See Figure 5	2			200			2	200	pA
		7			20			8	20	nA
$V_{ICR}$ Common-mode input voltage range		-1.5 to 4	-3.4 to 5.4		-11.5 to 14	-13.4 to 15.4				V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	3	4.3		13	14				V
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$	-3	-4.2		-12.5	-13.9				V
$A_{VD}$ Large-signal differential voltage amplification	$R_L = 10\ \text{k}\Omega$ , See Note 7	4	12		5	14.3				V/mV
$r_i$ Input resistance		10 <sup>12</sup>			10 <sup>12</sup>					$\Omega$
$c_i$ Input capacitance		5			4					pF
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0, R_S = 50\ \Omega$	70	87		70	94				dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC\pm} / \Delta V_{IO}$ )	$V_O = 0, R_S = 50\ \Omega$	75	96		75	96				dB
$P_D$ Total power dissipation (four amplifiers)	$V_O = 0$ , No load	7.7			10			26	34	mW
$I_{CC}$ Supply current (four amplifiers)	$V_O = 0$ , No load	0.77			1			0.87	1.12	$\mu\text{A}$
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$	120			120					dB

NOTE 7: At  $V_{CC\pm} = \pm 5\text{ V}$ ,  $V_O = \pm 2.3\text{ V}$ ; at  $V_{CC\pm} = \pm 15\text{ V}$ ,  $V_O = \pm 10\text{ V}$ .

**operating characteristics,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TL034Y						UNIT	
		$V_{CC\pm} = \pm 5\text{ V}$			$V_{CC\pm} = \pm 15\text{ V}$				
		MIN	TYP	MAX	MIN	TYP	MAX		
$SR+$ Positive slew rate at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF},$ See Figure 1	2			1.5			2.9	$\text{V}/\mu\text{s}$
$SR-$ Negative slew rate at unity gain		3.9			1.5			5.1	$\text{V}/\mu\text{s}$
$t_r$ Rise time	$V_{I(PP)} = \pm 10\text{ V},$ $R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF},$ See Figures 1 and 2	138			132				ns
$t_f$ Fall time		138			132				ns
Overshoot factor		11%			5%				
$V_n$ Equivalent input noise voltage	$R_S = 20\ \Omega,$ See Figure 3	$f = 10\ \text{kHz}$			83				nV/ $\sqrt{\text{Hz}}$
		$f = \text{kHz}$			43				
$I_n$ Equivalent input noise current	$f = 1\ \text{kHz}$	0.003			0.003				pA/ $\sqrt{\text{Hz}}$
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}, R_L = 10\ \text{k}\Omega,$ $C_L = 25\ \text{pF},$ See Figure 4	1			1.1				MHz
$\phi_m$ Phase margin at unity gain	$V_I = 10\ \text{mV}, R_L = 10\ \text{k}\Omega,$ $C_L = 25\ \text{pF},$ See Figure 4	61°			65°				

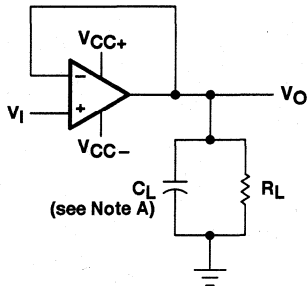
NOTE 5: This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

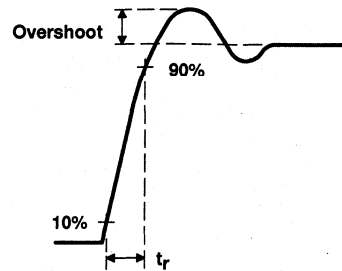
SLOS034D - JULY 1988 - REVISED AUGUST 1994

**PARAMETER MEASUREMENT INFORMATION**

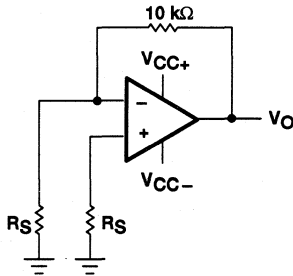


NOTE A:  $C_L$  includes fixture capacitance.

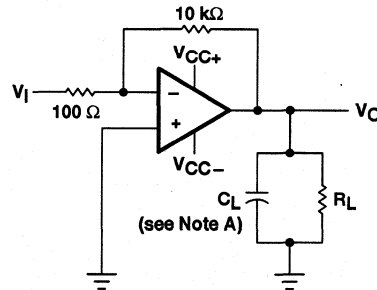
**Figure 1. Slew-Rate and Overshoot Test Circuit**



**Figure 2. Rise Time and Overshoot Waveform**

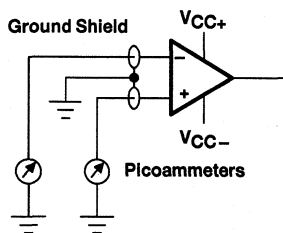


**Figure 3. Noise-Voltage Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit**



**Figure 5. Input-Bias and Offset-Current Test Circuit**

## PARAMETER MEASUREMENT INFORMATION

### typical values

Typical values presented in this data sheet represent the median (50% point) of device parametric performance.

### input bias and offset current

At the picoampere bias current level typical of the TL034 and TL034A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted into the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

### noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha V_{IO}$	Temperature coefficient input offset voltage	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Input bias current	vs Common-mode input voltage vs Free-air temperature	9 8
$V_{IC}$	Common-mode input voltage range	vs Supply voltage vs Free-air temperature	10 11
$V_{ID}$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage	vs Supply voltage vs Output current vs Free-air temperature	14 16, 17 18, 19
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	15
$A_{VD}$	Large-signal differential voltage amplification	vs Load resistance vs Frequency vs Free-air temperature	20 21 22
$z_o$	Output impedance	vs Frequency	23
CMRR	Common-mode rejection ratio	vs Frequency vs Free-air temperature	24, 25 26
kSVR	Supply voltage rejection ratio	vs Free-air temperature	27
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Time vs Free-air temperature	28 29 30
$V_n$	Equivalent input noise voltage	vs Frequency	31
$I_{CC}$	Supply current	vs Supply voltage vs Free-air temperature	32 33
SR	Slew rate	vs Load resistance vs Free-air temperature	34, 35 36, 37
	Overshoot factor	vs Load capacitance	38
THD	Total harmonic distortion	vs Frequency	39
$B_1$	Unity-gain bandwidth	vs Supply voltage vs Free-air temperature	40 41
$\phi_m$	Phase margin	vs Supply voltage vs Load capacitance vs Free-air temperature	42 43 44
	Pulse response	Small signal Large signal	45 46, 47
	Phase shift	vs Frequency	21

TYPICAL CHARACTERISTICS†

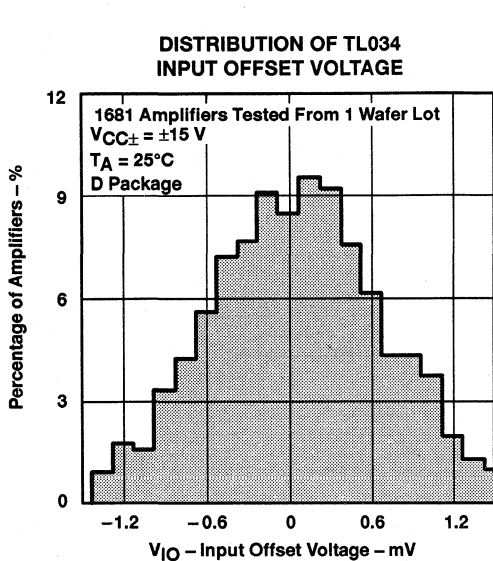


Figure 6

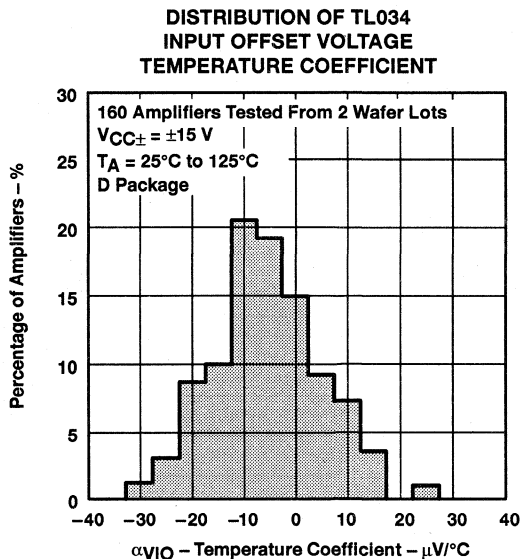


Figure 7

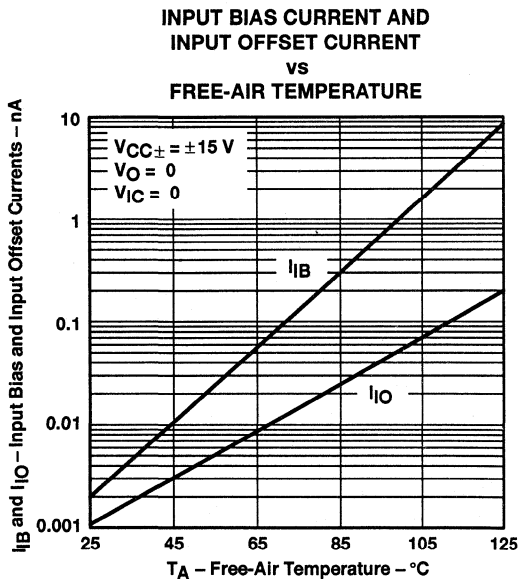


Figure 8

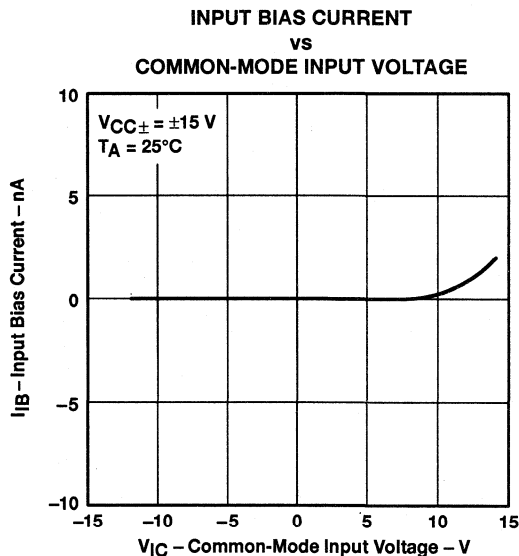


Figure 9

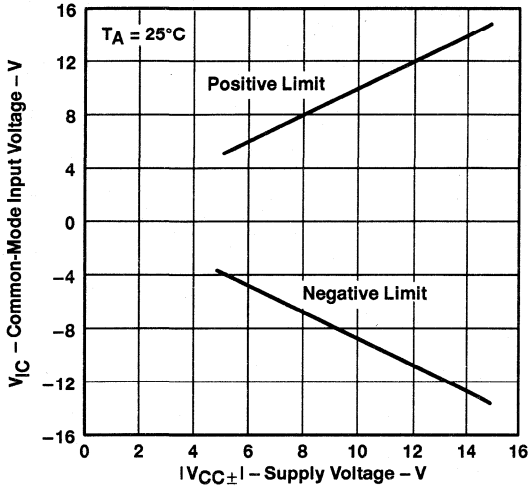
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

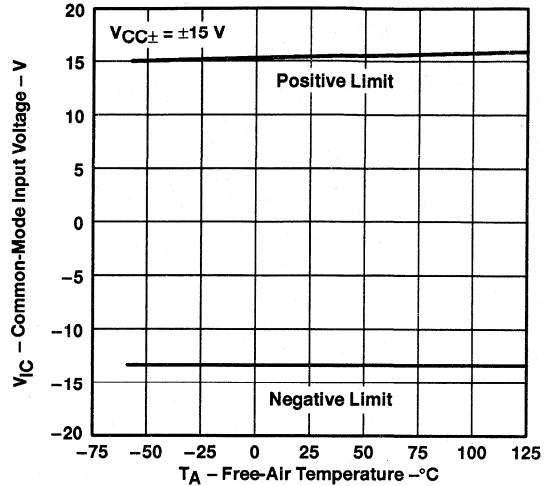
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE INPUT VOLTAGE**  
**VS**  
**SUPPLY VOLTAGE**



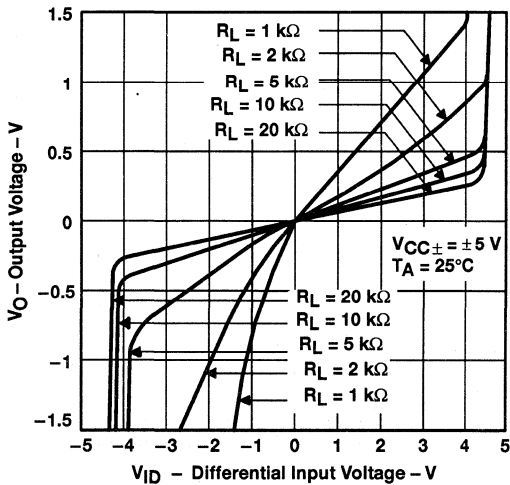
**Figure 10**

**COMMON-MODE INPUT VOLTAGE**  
**VS**  
**FREE-AIR TEMPERATURE**



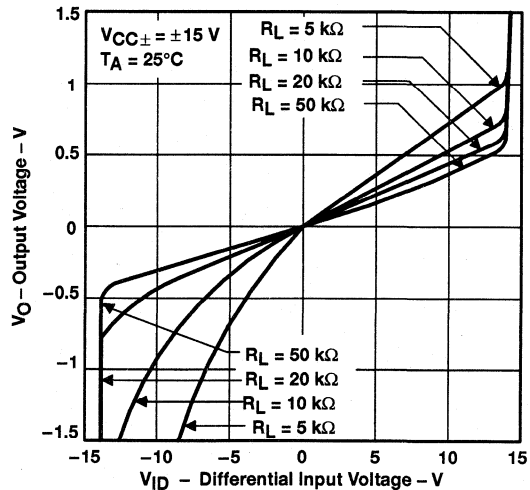
**Figure 11**

**OUTPUT VOLTAGE**  
**VS**  
**DIFFERENTIAL INPUT VOLTAGE**



**Figure 12**

**OUTPUT VOLTAGE**  
**VS**  
**DIFFERENTIAL INPUT VOLTAGE**



**Figure 13**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

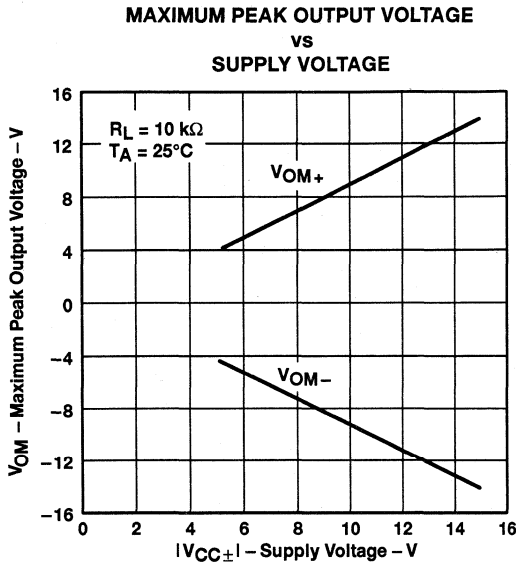


Figure 14

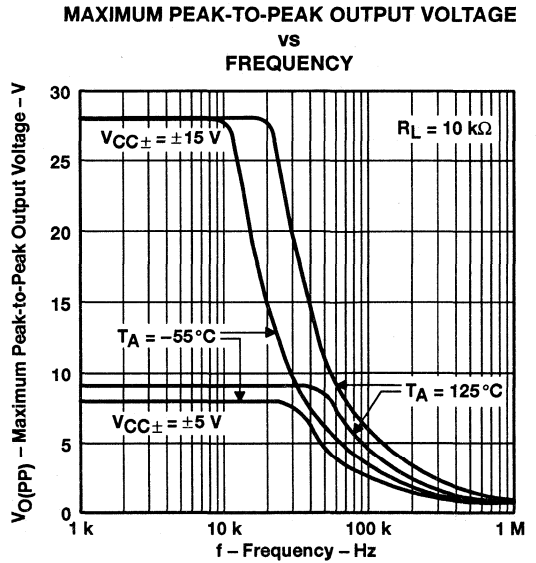


Figure 15

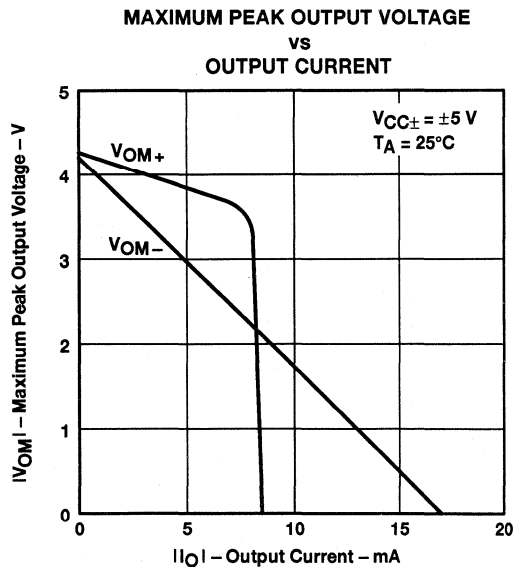


Figure 16

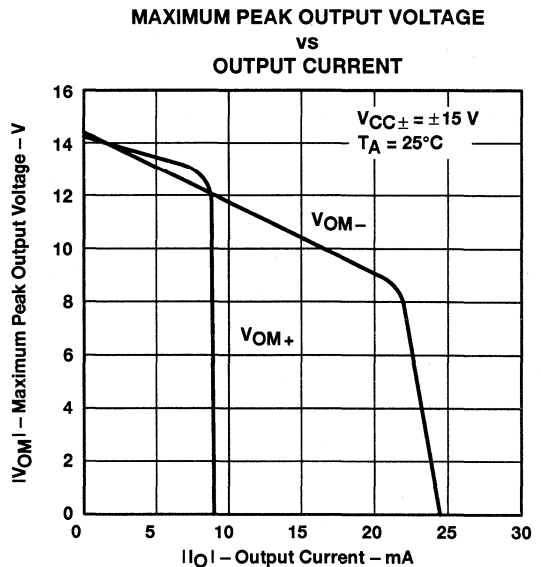


Figure 17

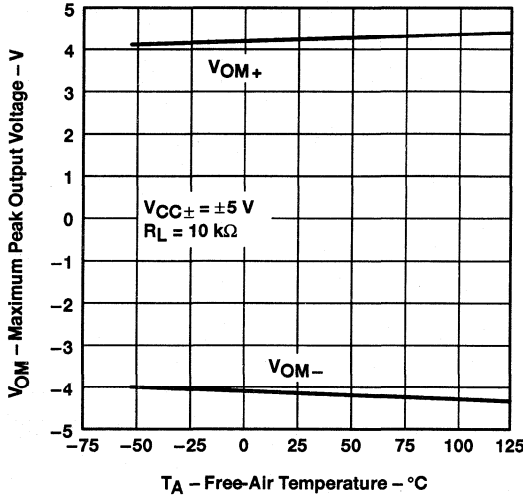
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

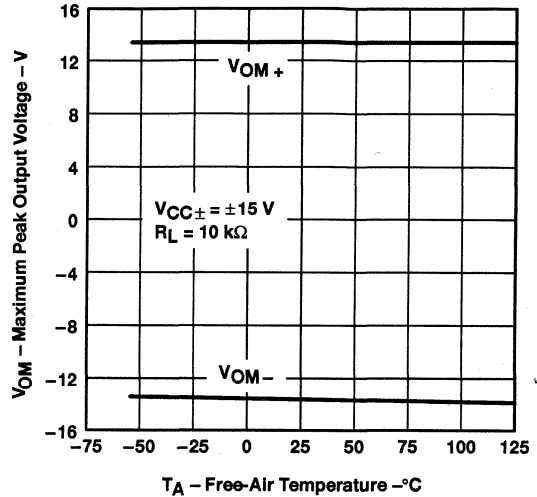
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



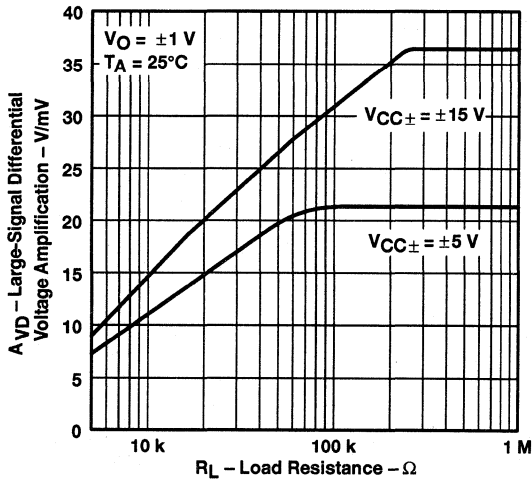
**Figure 18**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



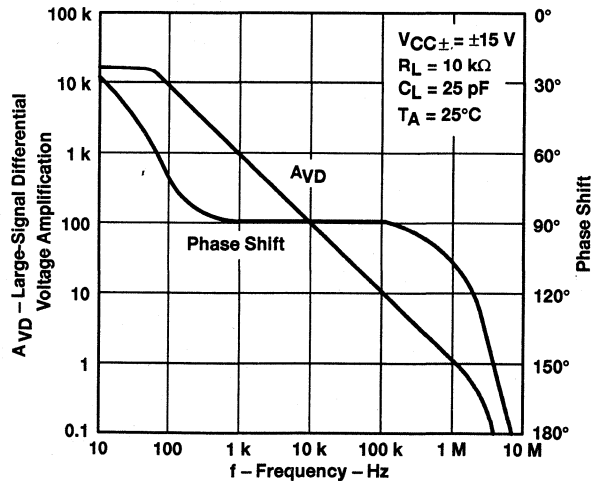
**Figure 19**

**LARGE-SIGNAL DIFFERENTIAL**  
**VOLTAGE AMPLIFICATION**  
**vs**  
**LOAD RESISTANCE**



**Figure 20**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION AND PHASE SHIFT**  
**vs**  
**FREQUENCY**



**Figure 21**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

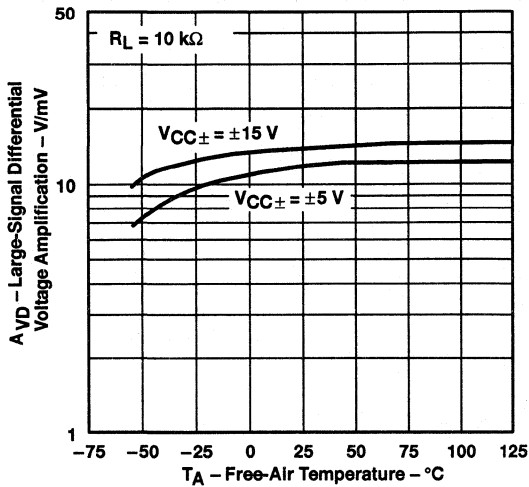


Figure 22

OUTPUT IMPEDANCE  
 VS  
 FREQUENCY

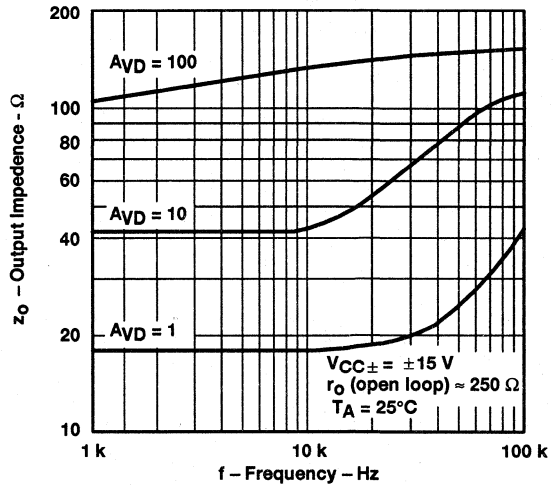


Figure 23

COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY

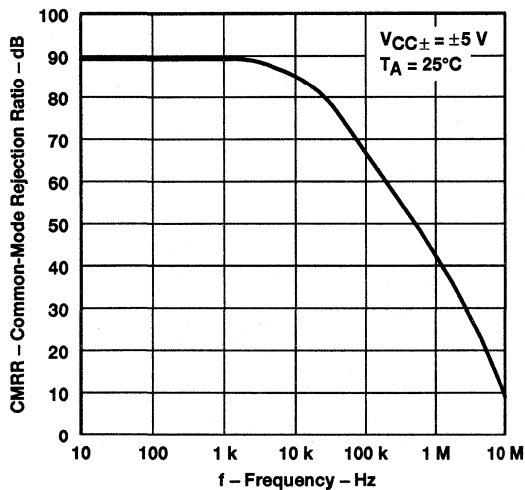


Figure 24

COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY

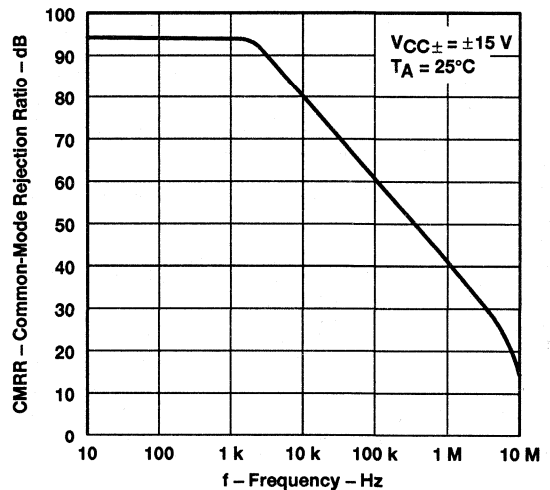


Figure 25

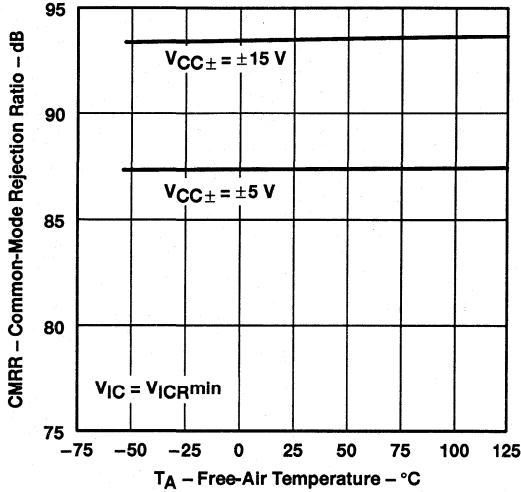
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

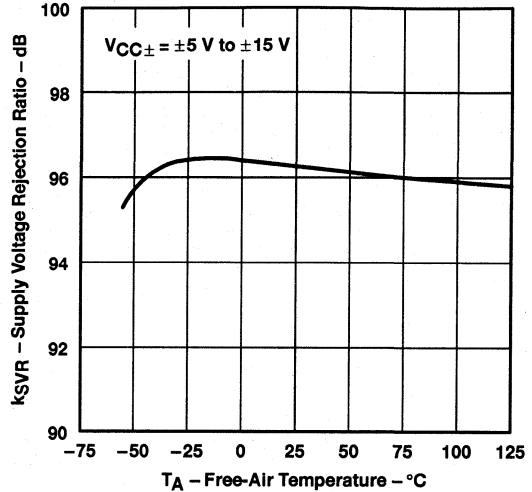
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



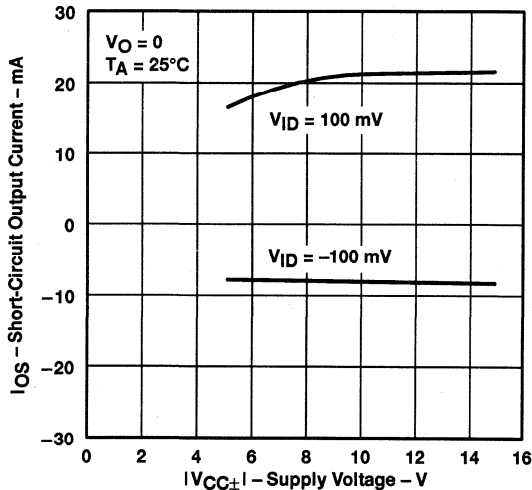
**Figure 26**

**SUPPLY VOLTAGE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



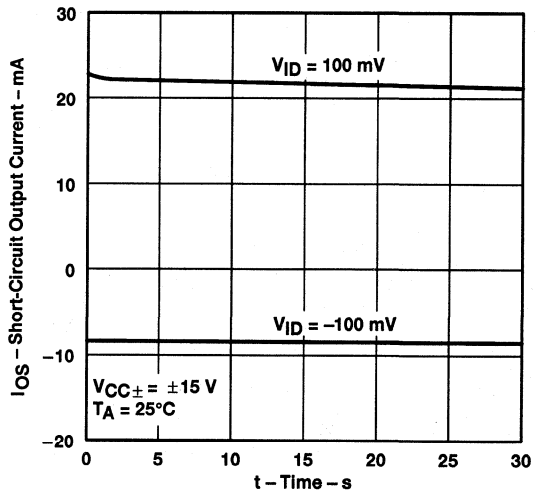
**Figure 27**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 28**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**TIME**



**Figure 29**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

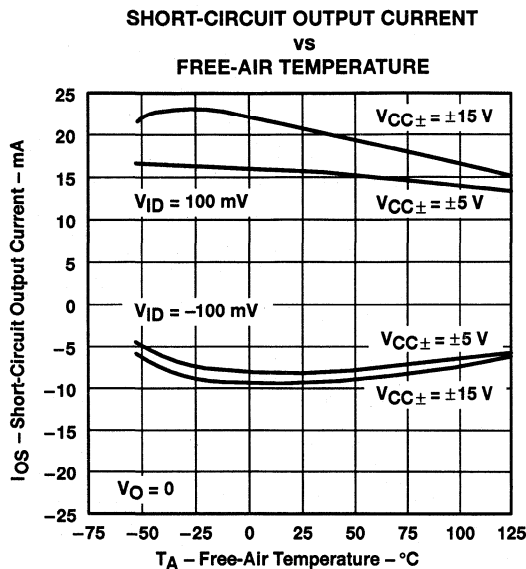


Figure 30

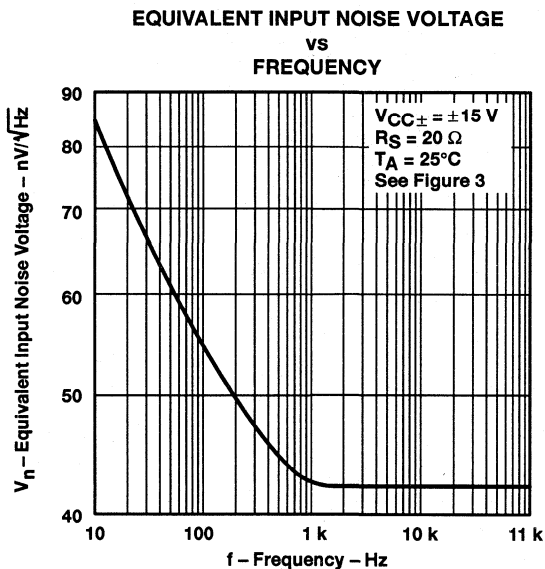


Figure 31

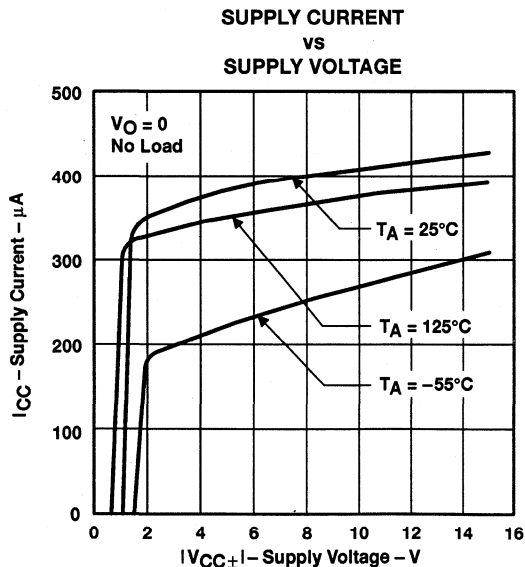


Figure 32

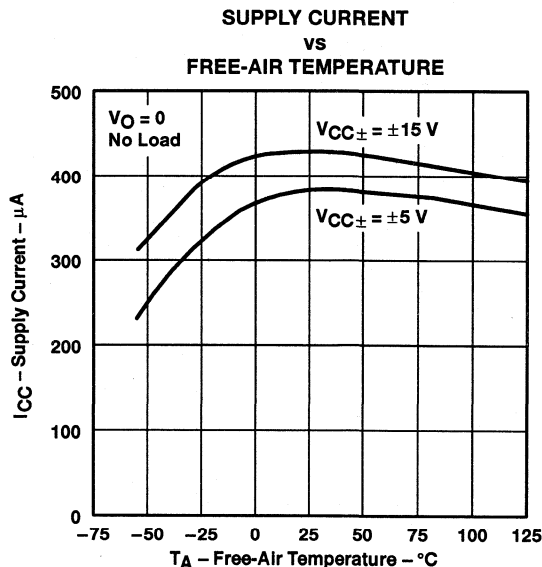


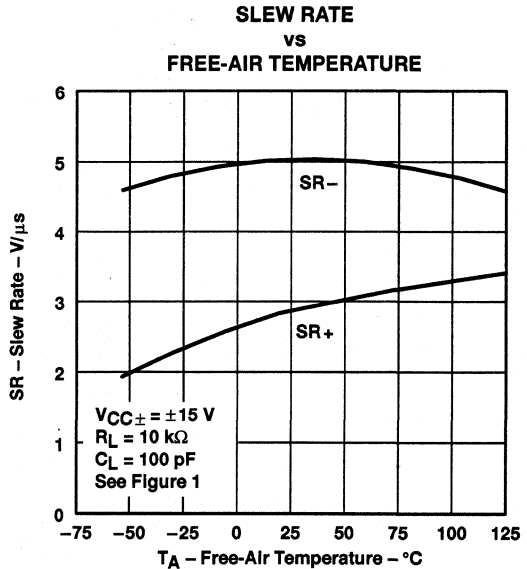
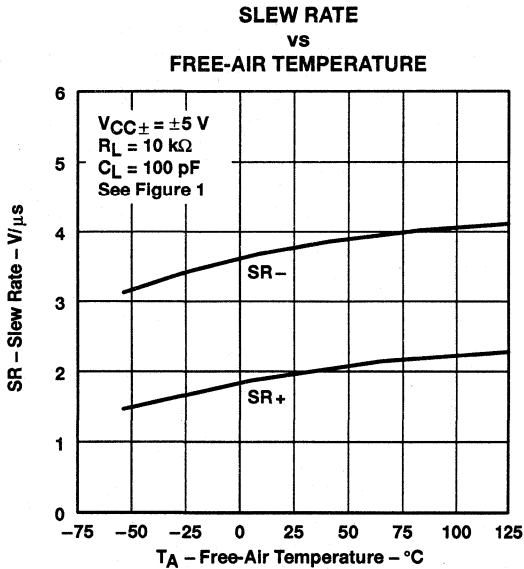
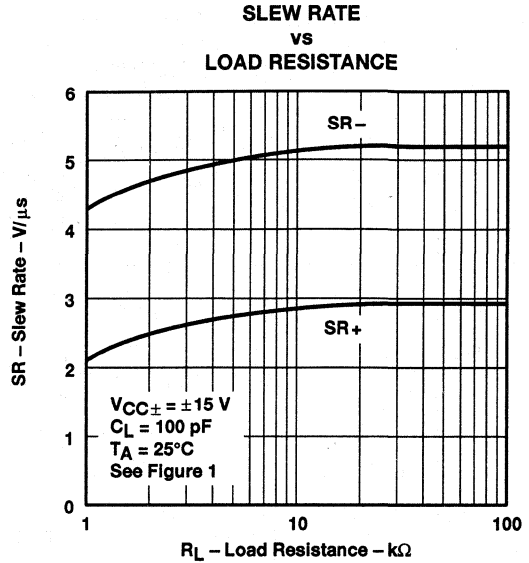
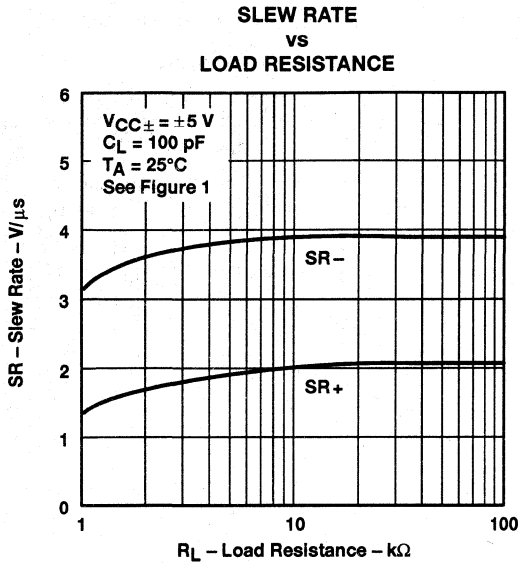
Figure 33

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D - JULY 1988 - REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

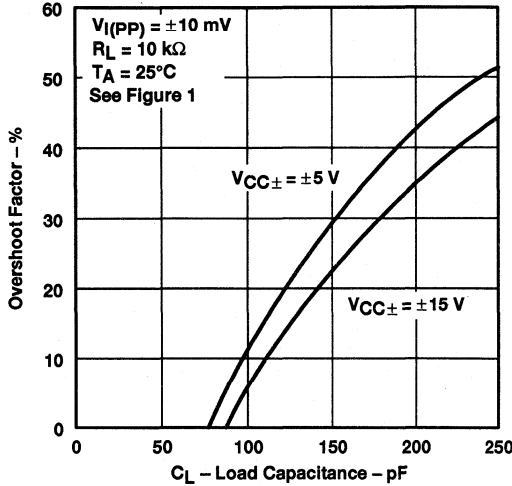


**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

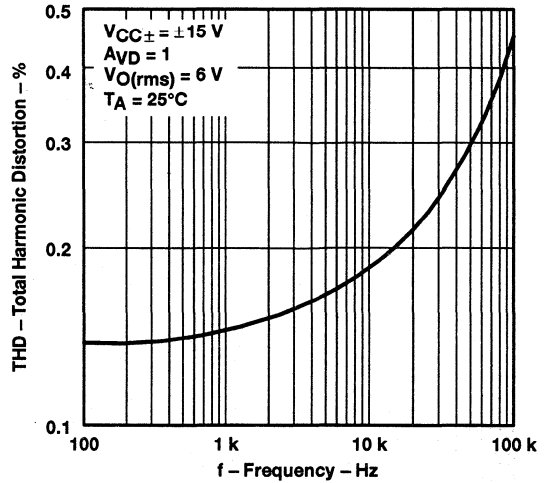
**TYPICAL CHARACTERISTICS†**

**OVERSHOOT FACTOR  
vs  
LOAD CAPACITANCE**



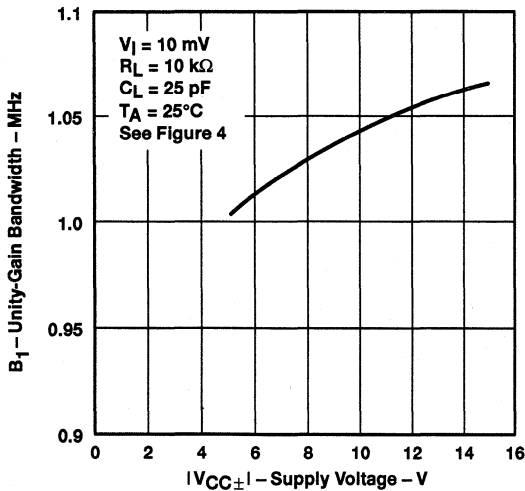
**Figure 38**

**TOTAL HARMONIC DISTORTION  
vs  
FREQUENCY**



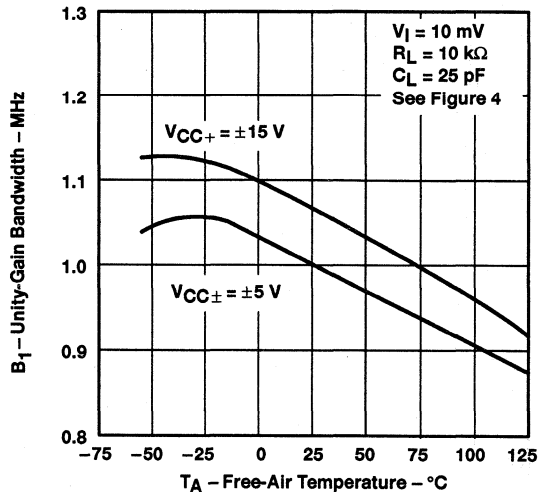
**Figure 39**

**UNITY-GAIN BANDWIDTH  
vs  
SUPPLY VOLTAGE**



**Figure 40**

**UNITY-GAIN BANDWIDTH  
vs  
FREE-AIR TEMPERATURE**



**Figure 41**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TL034, TL034A, TL034Y

## ENHANCED-JFET LOW-POWER LOW-OFFSET

### QUAD OPERATIONAL AMPLIFIERS

SLOS034D - JULY 1988 - REVISED AUGUST 1994

#### TYPICAL CHARACTERISTICS†

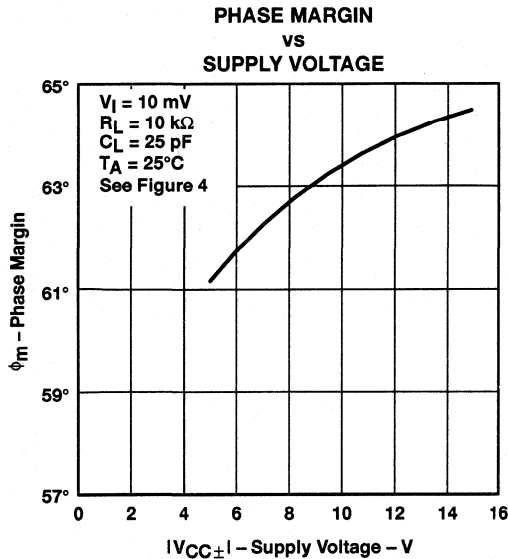
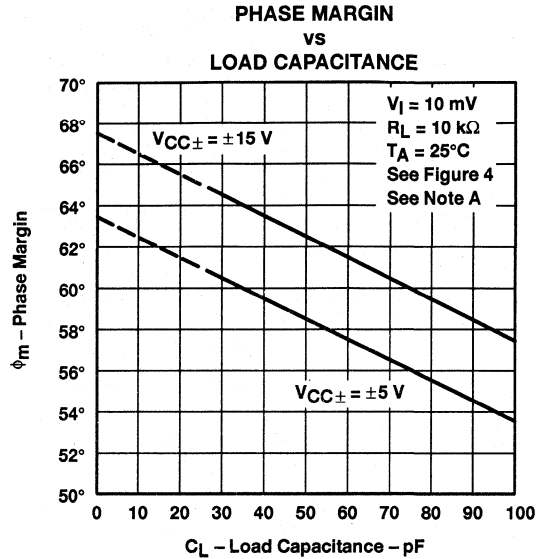


Figure 42



NOTE A: Values of phase margin below a load capacitance of 25 pF were estimated.

Figure 43

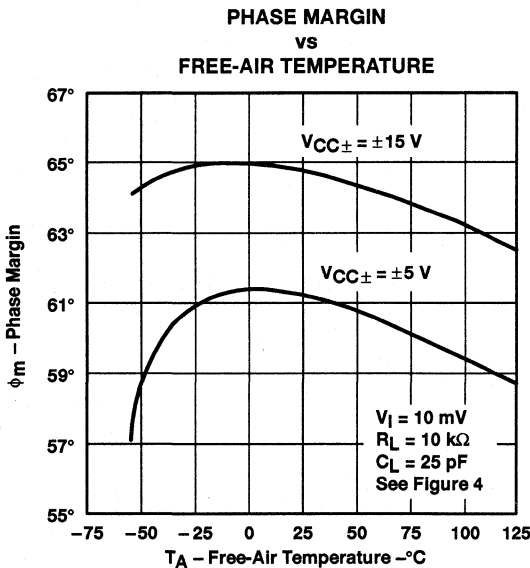


Figure 44

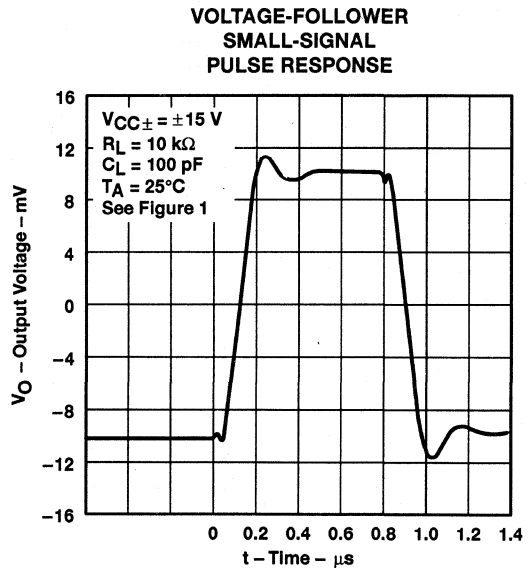


Figure 45

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS**

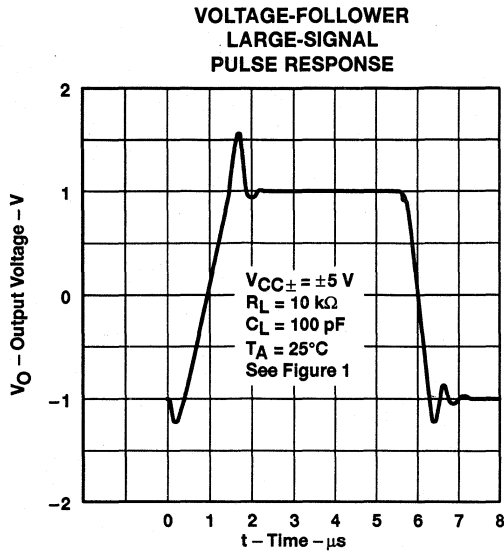


Figure 46

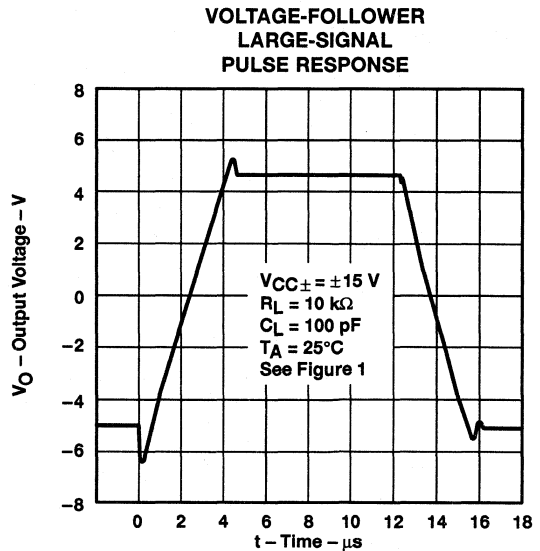


Figure 47

**APPLICATION INFORMATION**

**input characteristics**

The TL034 and TL034A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL034 and TL034A are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is a good practice to include guard rings around inputs (see Figure 48). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

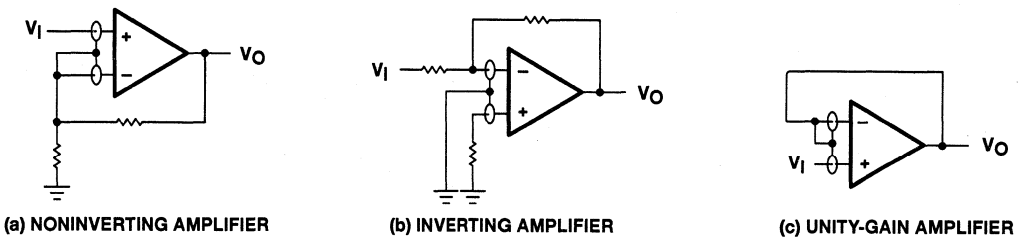


Figure 48. Use of Guard Rings

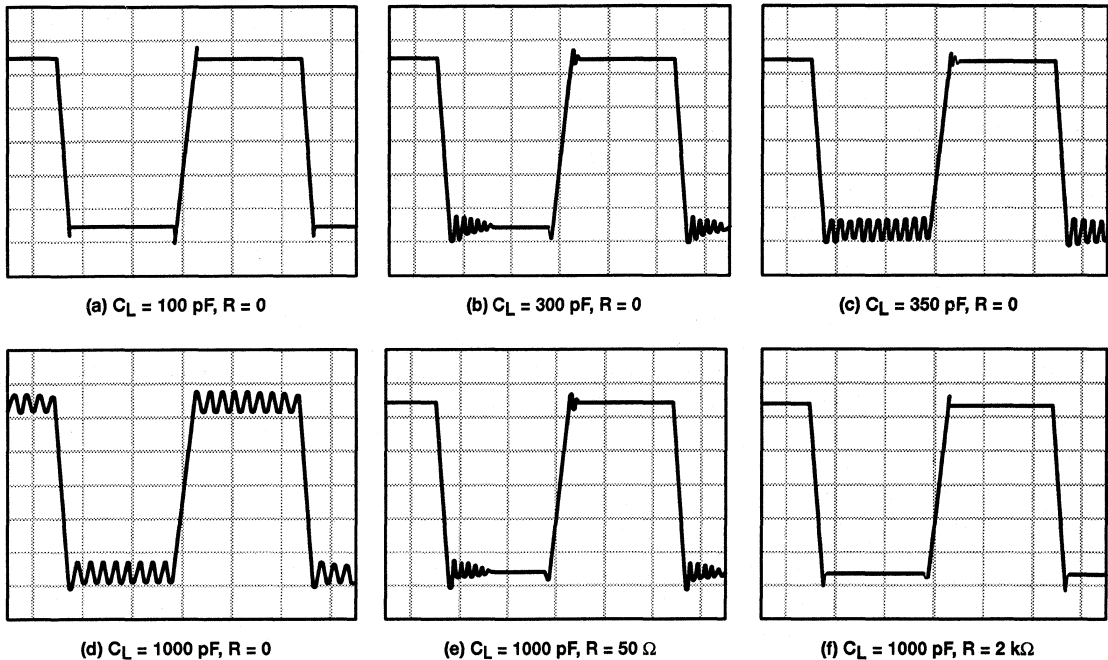
**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

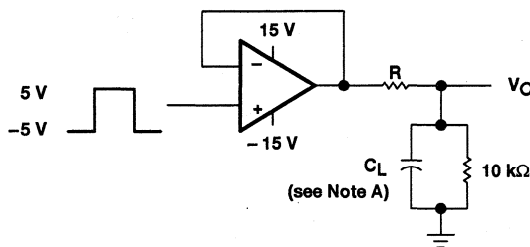
**APPLICATION INFORMATION**

**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL034 and TL034A drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem (see Figure 50). Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 49).



**Figure 49. Effect of Capacitive Loads**



NOTE A:  $C_L$  includes fixture capacitance.

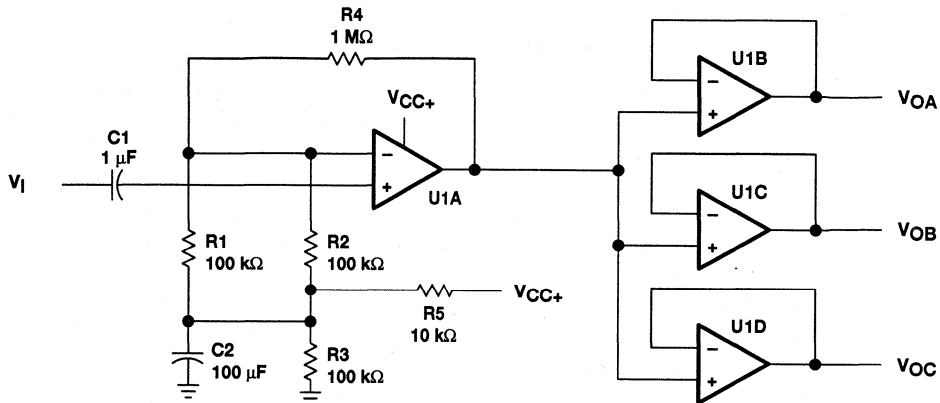
**Figure 50. Test Circuit for Output Characteristics**



APPLICATION INFORMATION

audio-distribution amplifier

This audio-distribution amplifier (see Figure 51) feeds the input signal to three separate output channels. U1A amplifies the input signal with a gain of 10, while U1B, U1C, and U1D serve as buffers to the output channels. The gain response of this circuit is very flat from 20 Hz to 20 kHz. The TL034 allows quick response to the input signal while maintaining low power consumption.



NOTE: U1A through U1D = TL034;  $V_{CC+} = 5\text{ V}$ .

Figure 51. Audio-Distribution Amplifier Circuit

**TL034, TL034A, TL034Y**  
**ENHANCED-JFET LOW-POWER LOW-OFFSET**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS034D – JULY 1988 – REVISED AUGUST 1994

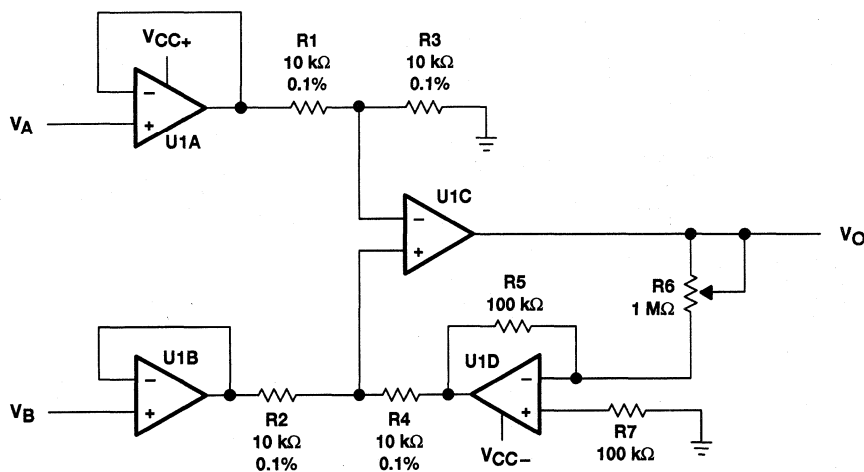
**APPLICATION INFORMATION**

**instrumentation amplifier with linear gain adjust**

The low offset voltage and low power consumption of the TL034 provide an accurate but inexpensive instrumentation amplifier (see Figure 52). This particular configuration offers the advantage that the gain can be linearly set by one resistor:

$$V_O = \frac{R6}{R5} \times (V_B - V_A)$$

Adjusting R6 varies the gain. The value of R6 should always be greater or equal to the value of R5 in order to ensure stability. The disadvantage of this instrumentation amplifier topology is the high degree of CMRR degradation resulting from mismatches between R1, R2, R3, and R4. For this reason, these four resistors should be 0.1% tolerance resistors.



NOTE: U1A through U1D = TL034;  $V_{CC\pm} = \pm 15$  V.

**Figure 52. Instrumentation Amplifier With Linear Gain-Adjust Circuit**

# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D - JUNE 1988 - REVISED AUGUST 1994

- **Maximum Offset Voltage . . . 800  $\mu$ V (TL051A)**
- **High Slew Rate . . . 19.8 V/ $\mu$ s Typ at 25°C**
- **Low Total Harmonic Distortion**  
**0.003% Typ at  $R_L = 2\text{ k}\Omega$**
- **Low Noise Voltage . . . 18 nV/ $\sqrt{\text{Hz}}$**   
**Typ at  $f = 1\text{ kHz}$**
- **Low Input Bias Currents . . . 30 pA Typ**

## description

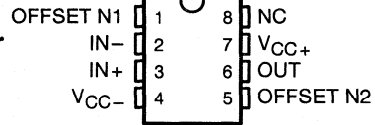
The TL051 and TL051A operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL051 well suited for new state-of-the-art designs as well as existing design upgrades. The TL051 has been designed to be functionally compatible, as well as pin compatible, with the TL071 and TL081. Two offset voltage grades are available: TL051 (1.5 mV max) and TL051A (800  $\mu$ V max).

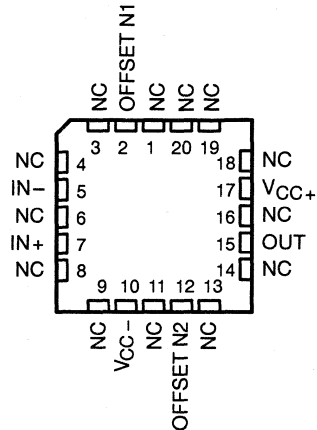
A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

D, JG, OR P PACKAGE  
(TOP VIEW)

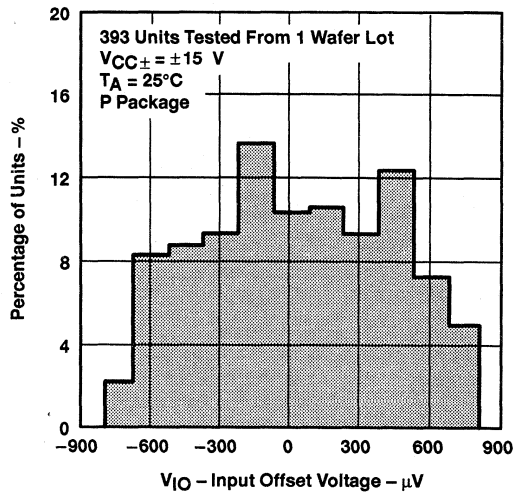


FK PACKAGE  
(TOP VIEW)



NC - No internal connection

DISTRIBUTION OF TL051A  
INPUT OFFSET VOLTAGE



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

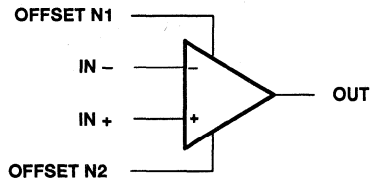
SLOS035D - JUNE 1988 - REVISED AUGUST 1994

## AVAILABLE OPTIONS

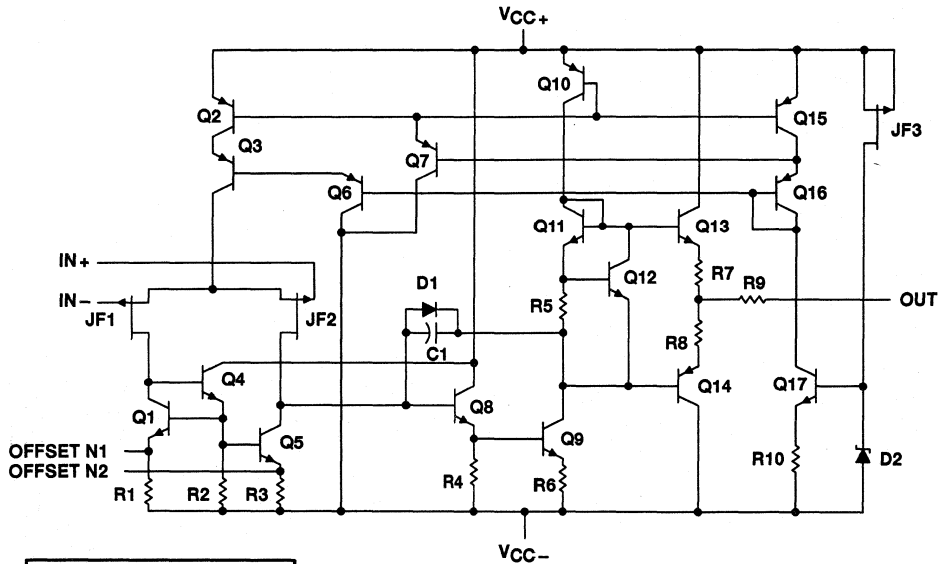
T <sub>A</sub>	V <sub>I</sub> Omax AT 25°C	PACKAGED DEVICES				CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	
0°C to 70°C	800 μV 1500 μV	TL051ACD TL051CD	—	—	TL051ACP TL051CP	TL051Y
-40°C to 85°C	800 μV 1500 μV	TL051AID TL051ID	—	—	TL051AIP TL051IP	—
-55°C to 125°C	800 μV 1500 μV	TL051AMD TL051MD	TL051AMFK TL051MFK	TL051AMJG TL051MJG	TL051AMP TL051MP	—

The D packages are available taped and reeled. Add R suffix to device type (e.g., TL051CDR).

## symbol



## equivalent schematic



COMPONENT COUNT	
Transistors	20
Resistors	10
Capacitors	1
Diodes	2

 **TEXAS  
INSTRUMENTS**

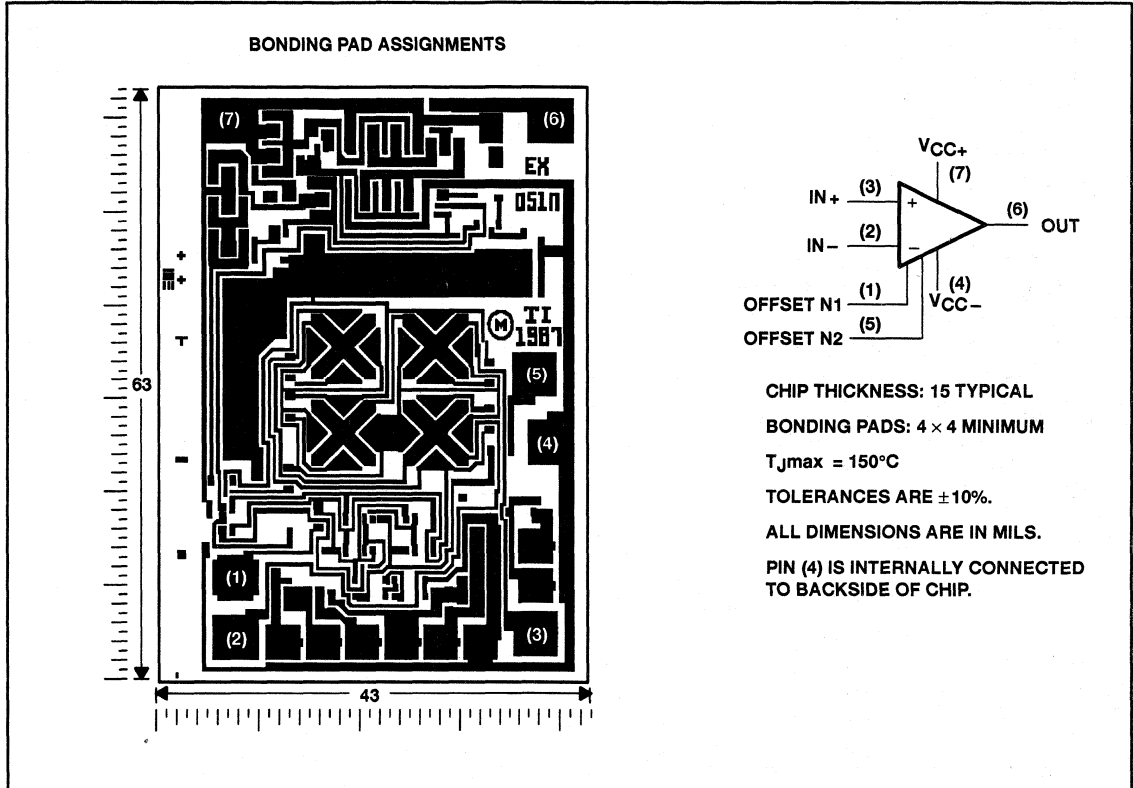
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL051, TL051A, TL051Y  
ENHANCED-JFET PRECISION  
OPERATIONAL AMPLIFIERS**

SLOS035D - JUNE 1988 - REVISED AUGUST 1994

**TL051Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL051. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$ V
Input voltage range, $V_I$ (any input) (see Notes 1 and 3)	$\pm 15$ V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$ (each output)	$\pm 80$ mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC\pm}$		$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5$ V	-1	4	-1	4	-1	4	V
	$V_{CC\pm} = \pm 15$ V	-11	11	-11	11	-11	11	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL051C, TL051AC						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL051C	25°C	0.75	3.5	0.59	1.5	mV	
			Full range		4.5		2.5		
		TL051AC	25°C	0.55	2.8	0.35	0.8		
			Full range		3.8		1.8		
αV <sub>IO</sub> Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL051C	25°C to 70°C	8		8		μV/°C	
		TL051AC	25°C to 70°C	8		8	25		
Input offset voltage long-term drift (see Note 6)		25°C	0.04		0.04		μV/mo		
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	4	100	5	100	pA		
		70°C	0.02	1	0.025	1	nA		
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	20	200	30	200	pA		
		70°C	0.15	4	0.2	4	nA		
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3		13				
	R <sub>L</sub> = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5		11.5				
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5		-12				
	R <sub>L</sub> = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3		-11				
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 7	25°C	25	59	50	105	V/mV		
		0°C	30	65	60	129			
		70°C	20	46	30	85			
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub> Input capacitance		25°C	10		12		pF		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	65	85	75	93	dB		
		0°C	65	84	75	92			
		70°C	65	84	75	91			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	99	75	99	dB		
		0°C	75	98	75	98			
		70°C	75	97	75	97			
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C	2.6	3.2	2.7	3.2	mA		
		0°C	2.7	3.2	2.8	3.2			
		70°C	2.6	3.2	2.7	3.2			

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V, or for V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL051C, TL051AC						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	Positive slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	16			13 20			V/μs
			Full range	16.4			11 22.6			
SR-	Negative slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	15			13 18			
			Full range	16			11 19.3			
t <sub>r</sub>	Rise time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56			ns
			0°C	54			55			
			70°C	63			63			
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57			
			0°C	54			56			
			70°C	62			64			
	Overshoot factor	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	24%			19%			
			0°C	24%			19%			
			70°C	24%			19%			
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			75			nV/√Hz
			f = 1 kHz	25°C			18 18 30			
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage	f = 10 Hz to 10 kHz	25°C	4			4			μV
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01			pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 9	25°C	0.003%			0.003%			
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3.1			MHz
			0°C	3.2			3.3			
			70°C	2.7			2.8			
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	59°			62°			
			0°C	58°			62°			
			70°C	59°			62°			

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.





**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS		T <sub>A</sub> †	TL051I, TL051AI						UNIT	
				V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
				MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub> Input offset voltage		TL051I	25°C	0.75		3.5	0.59		1.5	mV	
			Full range			5.3			3.3		
			TL051AI	25°C	0.55		2.8	0.35			0.8
				Full range			4.6				2.6
α <sub>VIO</sub> Temperature coefficient of input offset voltage (see Note 5)		TL051I	25°C to 85°C	7		8				μV/°C	
			TL051AI	25°C to 85°C	8		8		25		
Input offset voltage long-term drift (see Note 6)			25°C	0.04		0.04				μV/mo	
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	4		100	5		100	pA	
			85°C	0.06		10	0.07		10	nA	
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	20		200	30		200	pA	
			85°C	0.6		20	0.7		20	nA	
V <sub>ICR</sub> Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6		-11 to 11	-12.3 to 15.6	V		
			Full range	-1 to 4			-11 to 11				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3		4.2	13		13.9	V	
			Full range	3			13				
	R <sub>L</sub> = 2 kΩ	25°C	2.5		3.8	11.5		12.7			
		Full range	2.5			11.5					
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-2.5		-3.5	-12		-13.2	V	
			Full range	-2.5			-12				
	R <sub>L</sub> = 2 kΩ	25°C	-2.3		-3.2	-11		-12			
		Full range	-2.3			-11					
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 7		25°C	25		59	50		105	V/mV	
			-40°C	30		74	60		145		
			85°C	20		43	30		76		
r <sub>i</sub> Input resistance			25°C	10 <sup>12</sup>			10 <sup>12</sup>		Ω		
c <sub>i</sub> Input capacitance			25°C	10			12		pF		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	65		85	75		93	dB	
			-40°C	65		83	75		90		
			85°C	65		84	75		93		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	75		99	75		99	dB	
			-40°C	75		98	75		98		
			85°C	75		99	75		99		
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load		25°C	2.6		3.2	2.7		3.2	mA	
			-40°C	2.4		3.2	2.6		3.2		
			85°C	2.5		3.2	2.6		3.2		

† Full range is -40°C to 85°C

- NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.
6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.
7. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V, or for V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D - JUNE 1988 - REVISED AUGUST 1994

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL051I, TL051AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+	Positive slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	16			13	20	V/μs
			Full range				11		
SR-	Negative slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	15			13	18	
			Full range				11		
t <sub>r</sub>	Rise time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56		ns
			-40°C	52			53		
			85°C	64			65		
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57		
			-40°C	51			53		
			85°C	64			65		
Overshoot factor			25°C	24%			19%		
			-40°C	24%			19%		
			85°C	24%			19%		
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C	75			nV/√Hz	
			f = 1 kHz	25°C	18				18
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C	4			μV	
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz		25°C	0.01			0.01	pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 9		25°C	0.003%			0.003%	
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3.1		MHz
			-40°C	3.5			3.6		
			85°C	2.6			2.7		
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	59°			62°		
			-40°C	58°			61°		
			85°C	59°			62°		

† Full range is -40°C to 85°C.

NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.



**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> <sup>†</sup>	TL051M, TL051AM						UNIT
			V <sub>CC±</sub> = ± 5 V			V <sub>CC±</sub> = ± 15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL051M	25°C	0.75	3.5	0.59	1.5	mV	
			Full range		6.5		4.5		
		TL051AM	25°C	0.55	2.8	0.35	0.8		
			Full range		5.8		3.8		
αV <sub>IO</sub> Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL051M	25°C to 125°C	8		8		μV/°C	
		TL051AM	25°C to 125°C	8		8			
Input offset voltage long-term drift (see Note 6)		25°C	0.04		0.04		μV/mo		
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	4	100	5	100	pA		
		125°C	1	20	2	20	nA		
I <sub>IB</sub> Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	20	200	30	200	pA		
		125°C	10	50	20	50	nA		
V <sub>ICR</sub> Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
		Full range	-1 to 4		-11 to 11				
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3	4.2	13	13.9	V		
		Full range	3		13				
	R <sub>L</sub> = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
		Full range	2.5		11.5				
V <sub>OM-</sub> Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
		Full range	-2.5		-12				
	R <sub>L</sub> = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
		Full range	-2.3		-11				
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 7	25°C	25	59	50	105	V/mV		
		-55°C	30	76	60	149			
		125°C	10	32	15	49			
r <sub>i</sub> Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub> Input capacitance		25°C	10		12		pF		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	65	85	75	93	dB		
		-55°C	65	83	75	92			
		125°C	65	84	75	94			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	99	75	99	dB		
		-55°C	75	98	75	98			
		125°C	75	100	75	100			
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C	2.6	3.2	2.7	3.2	mA		
		-55°C	2.3	3.2	2.4	3.2			
		125°C	2.4	3.2	2.5	3.2			

<sup>†</sup> Full range is -55°C to 125°C.

NOTES: 6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. For V<sub>CC±</sub> = ± 5 V, V<sub>O</sub> = ± 2.3 V, or for V<sub>CC±</sub> = ± 15 V, V<sub>O</sub> = ± 10 V.



# TL051, TL051A, TL051Y ENHANCED-JFET PRECISION OPERATIONAL AMPLIFIERS

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL051M, TL051AM						UNIT		
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V					
			MIN	TYP	MAX	MIN	TYP	MAX			
SR+	Positive slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	16			13 20			V/μs	
SR-	Negative slew rate at unity gain		25°C	15			13				
t <sub>r</sub>	Rise time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56			ns	
t <sub>f</sub>	Fall time		-55°C	51			52				
			125°C	68			68				
			25°C	55			57				
Overshoot factor			-55°C	51			52				
			125°C	68			69				
			25°C	24%			19%				
			-55°C	25%			19%				
			125°C	25%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C	75			75			nV/√Hz
			f = 1 kHz	25°C	18			19			
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C	4			4			μV
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz		25°C	0.01			0.01			pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz,	R <sub>L</sub> = 2kΩ, See Note 9	25°C	0.003%			0.003%			
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3.1			MHz
				-55°C	3.6			3.7			
				125°C	2.3			2.4			
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	59°			62°			
				-55°C	57°			61°			
				125°C	59°			62°			

NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.



**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TL051Y						UNIT
		$V_{CC} \pm = \pm 5 \text{ V}$			$V_{CC} \pm = \pm 15 \text{ V}$			
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0, V_{IC} = 0, R_S = 50 \Omega$		0.75	3.5		0.59	1.5	mV
$I_{IO}$ Input offset current	$V_O = 0, V_{IC} = 0, \text{See Figure 5}$		4	100		5	100	pA
$I_{IB}$ Input bias current	$V_O = 0, V_{IC} = 0, \text{See Figure 5}$		20	200		30	200	pA
$V_{ICR}$ Common-mode input voltage range			-1 to 4	-2.3 to 5.6		-11 to 11	-12.3 to 15.6	V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 \text{ k}\Omega$		3	4.2		13	13.9	V
	$R_L = 2 \text{ k}\Omega$		2.5	3.8		11.5	12.7	
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10 \text{ k}\Omega$		-2.5	-3.5		-12	-13.2	V
	$R_L = 2 \text{ k}\Omega$		-2.3	-3.2		-11	-12	
$A_{VD}$ Large-signal differential voltage amplification	$R_L = 2 \text{ k}\Omega, \text{See Note 7}$		25	59		50	105	V/mV
$r_i$ Input resistance			$10^{12}$			$10^{12}$		$\Omega$
$c_i$ Input capacitance			10			12		pF
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}, V_O = 0, R_S = 50 \Omega$		65	85		75	93	dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC} / \Delta V_{IO}$ )	$V_O = 0, R_S = 50 \Omega$		75	99		75	99	dB
$I_{CC}$ Supply current	$V_O = 0, \text{No load}$		2.6	3.2		2.7	3.2	mA

- NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.  
7. For  $V_{CC} \pm = \pm 5 \text{ V}$ ,  $V_O = \pm 2.3 \text{ V}$ , or for  $V_{CC} \pm = \pm 15 \text{ V}$ ,  $V_O = \pm 10 \text{ V}$ .

**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1984

**operating characteristics,  $T_A = 25^\circ\text{C}$**

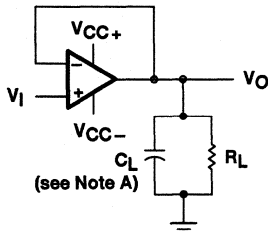
PARAMETER	TEST CONDITIONS	TL051Y						UNIT
		$V_{CC\pm} = \pm 5\text{ V}$			$V_{CC\pm} = \pm 15\text{ V}$			
		MIN	TYP	MAX	MIN	TYP	MAX	
SR+	Positive slew rate at unity gain	$R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$ , See Figure 1 and Note 8		16		13 20		$\text{V}/\mu\text{s}$
SR-	Negative slew rate at unity gain			15		13 18		
$t_r$	Rise time	$V_I(\text{PP}) = \pm 10\text{ mV}$ , $R_L = 2\text{ k}\Omega$ , $C_L = 100\text{ pF}$ , See Figures 1 and 2		55		56		ns
$t_f$	Fall time			55		57		
	Overshoot factor			24%		19%		
$V_n$	Equivalent input noise voltage (see Note 5)	$R_S = 20\ \Omega$ , See Figure 3	$f = 10\text{ Hz}$	75		75		$\text{nV}/\sqrt{\text{Hz}}$
			$f = 1\text{ kHz}$	18		18 30		
$V_N(\text{PP})$	Peak-to-peak equivalent input noise voltage		$f = 10\text{ Hz to }10\text{ kHz}$	4		4		$\mu\text{V}$
$I_n$	Equivalent input noise current	$f = 1\text{ kHz}$		0.01		0.01		$\text{pA}/\sqrt{\text{Hz}}$
THD	Total harmonic distortion	$R_S = 1\text{ k}\Omega$ , $f = 1\text{ kHz}$ ,	$R_L = 2\text{ k}\Omega$ , See Note 9	0.003%		0.003%		
$B_1$	Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 25\text{ pF}$ ,	$R_L = 2\text{ k}\Omega$ , See Figure 4	3		3.1		MHz
$\phi_m$	Phase margin at unity gain	$V_I = 10\text{ mV}$ , $C_L = 25\text{ pF}$ ,	$R_L = 2\text{ k}\Omega$ , See Figure 4	59°		62°		

NOTES: 5. This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For  $V_{CC\pm} = \pm 5\text{ V}$ ,  $V_I(\text{PP}) = \pm 1\text{ V}$ ; for  $V_{CC\pm} = \pm 15\text{ V}$ ,  $V_I(\text{PP}) = \pm 5\text{ V}$ .  
 9. For  $V_{CC\pm} = \pm 5\text{ V}$ ,  $V_{O\text{rms}} = 1\text{ V}$ ; for  $V_{CC\pm} = \pm 15\text{ V}$ ,  $V_{O\text{rms}} = 6\text{ V}$ .



PARAMETER MEASUREMENT INFORMATION



NOTE A:  $C_L$  includes fixture capacitance.

Figure 1. Test Circuit

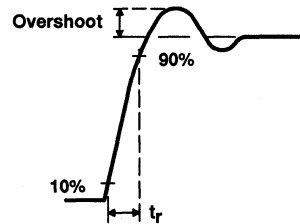


Figure 2. Rise Time and Overshoot Waveform

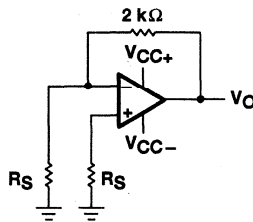


Figure 3. Noise-Voltage Test Circuit

typical values

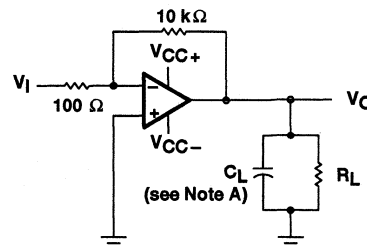
Typical values presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL051 and TL051A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.



NOTE A:  $C_L$  includes fixture capacitance.

Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit

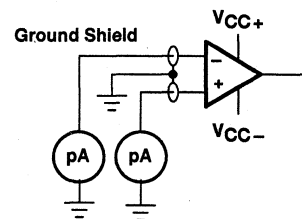


Figure 5. Input-Bias and Offset-Current Test Circuit

**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Common-mode input bias current	vs Common-mode input voltage vs Free-air temperature	9 8
$V_I$	Input voltage range	vs Supply voltage vs Free-air temperature	10 11
$V_O$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage swing	vs Supply voltage vs Output current vs Frequency vs Free-air temperature	14 18, 19 15, 16, 17 20, 21
$A_{VD}$	Differential voltage amplification	vs Load resistance vs Frequency vs Free-air temperature	22 23 24, 25
$CMRR$	Common-mode rejection ratio	vs Frequency vs Free-air temperature	26, 27 28
$Z_o$	Output impedance	vs Frequency	29
$KS_{VR}$	Supply-voltage rejection ratio	vs Free-air temperature	30
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Time vs Free-air temperature	31 32 33
$I_{CC}$	Supply current	vs Supply voltage vs Free-air temperature	34 35
$SR$	Slew rate	vs Load resistance vs Free-air temperature	36, 37 38, 39
	Overshoot factor	vs Load capacitance	40
$V_n$	Equivalent input noise voltage	vs Frequency	41
$THD$	Total harmonic distortion	vs Frequency	42
$B_1$	Unity-gain bandwidth	vs Supply voltage vs Free-air temperature	43 44
$\phi_m$	Phase margin	vs Supply voltage vs Load capacitance vs Free-air temperature	45 46 47
	Phase shift	vs Frequency	23
	Pulse response	Small signal	48 49



TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TL051  
 INPUT OFFSET VOLTAGE

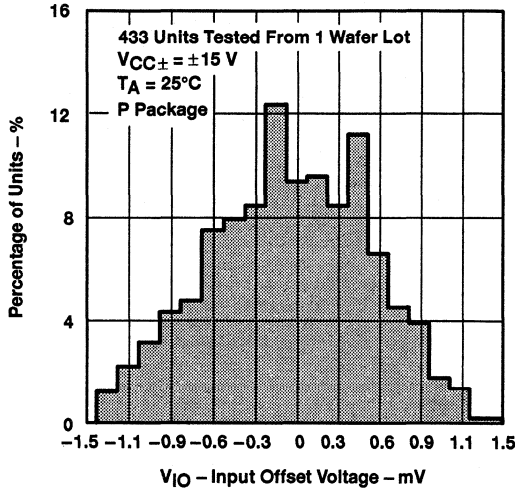


Figure 6

DISTRIBUTION OF TL051  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

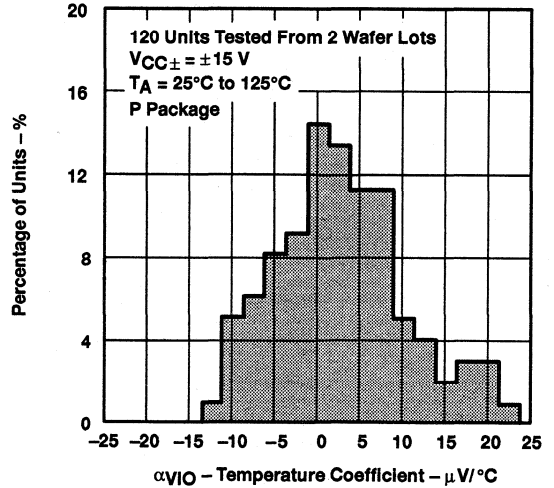


Figure 7

INPUT BIAS CURRENT AND  
 INPUT OFFSET CURRENT  
 vs  
 FREE-AIR TEMPERATURE

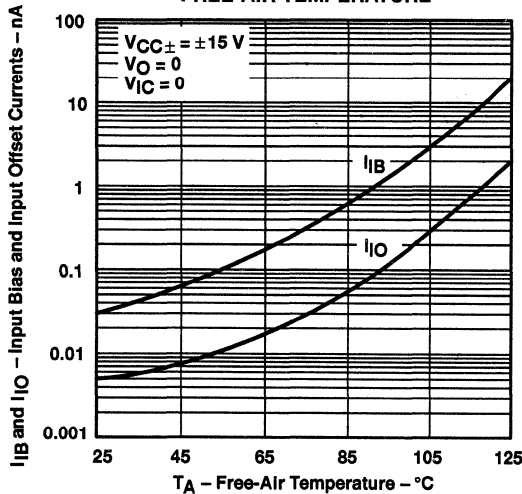


Figure 8

INPUT BIAS CURRENT  
 vs  
 COMMON-MODE INPUT VOLTAGE

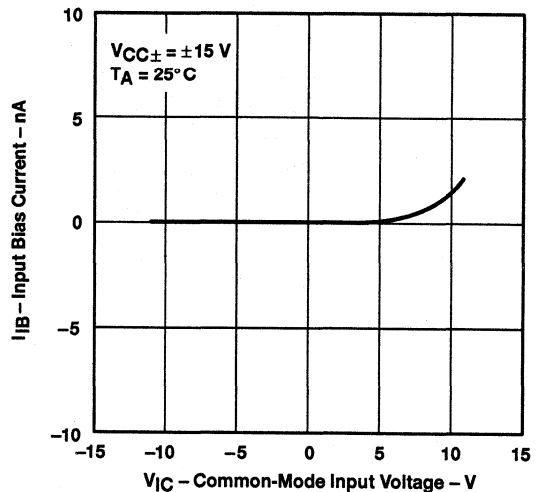
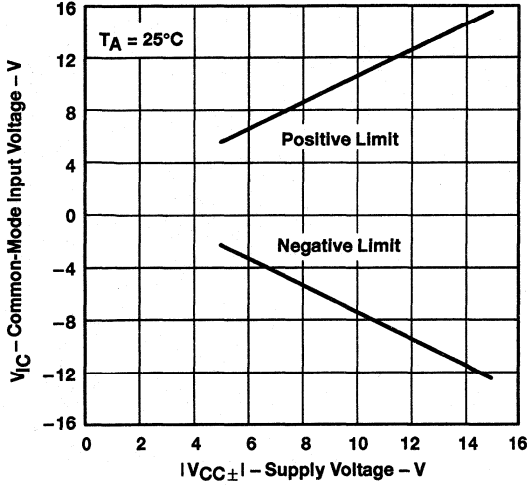


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

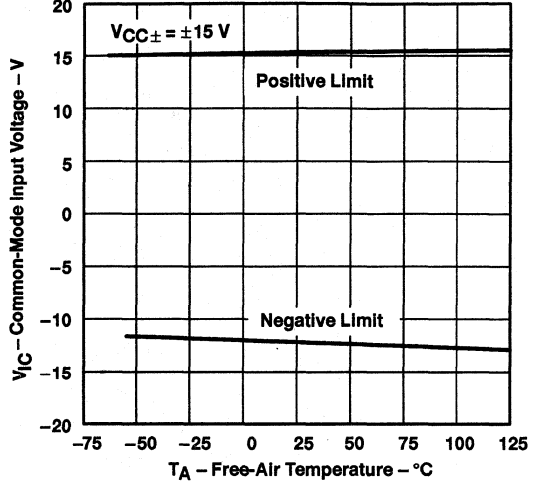
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS vs SUPPLY VOLTAGE**



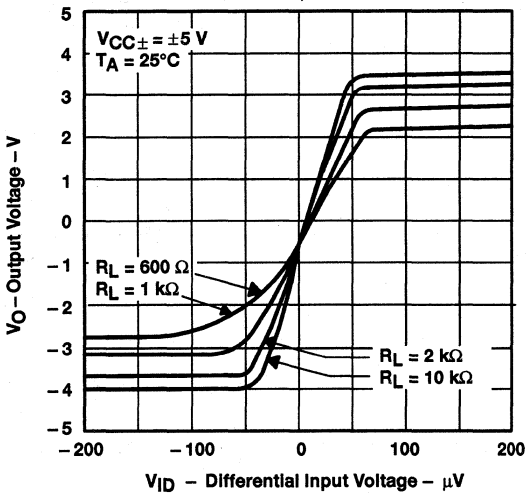
**Figure 10**

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS vs FREE-AIR TEMPERATURE**



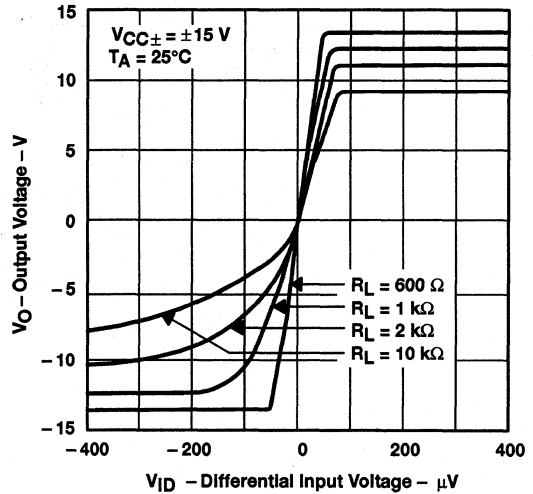
**Figure 11**

**OUTPUT VOLTAGE vs DIFFERENTIAL INPUT VOLTAGE**



**Figure 12**

**OUTPUT VOLTAGE vs DIFFERENTIAL INPUT VOLTAGE**



**Figure 13**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

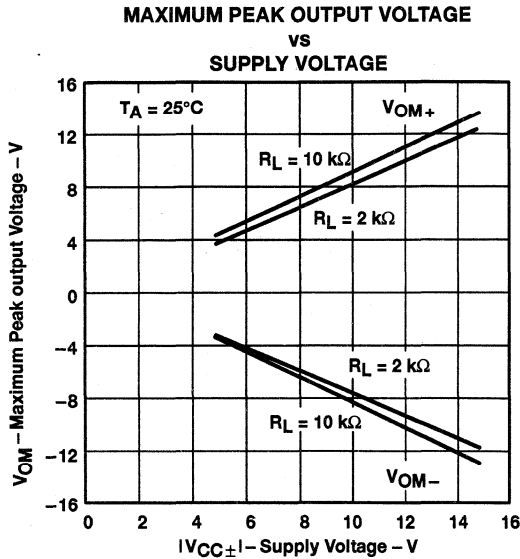


Figure 14

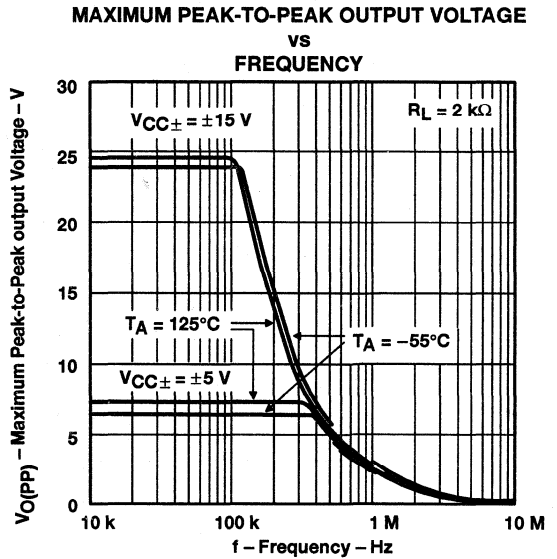


Figure 15

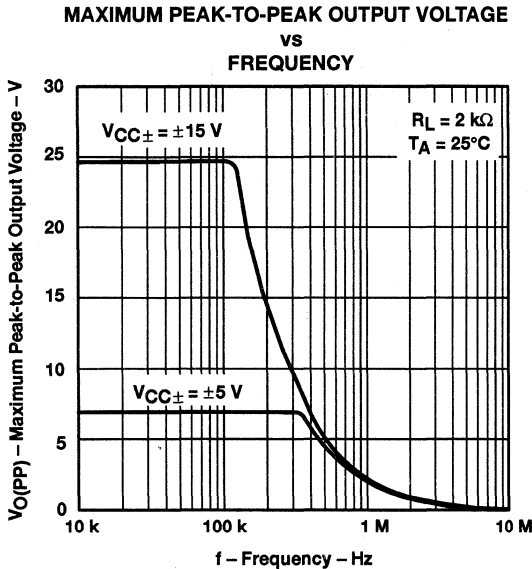


Figure 16

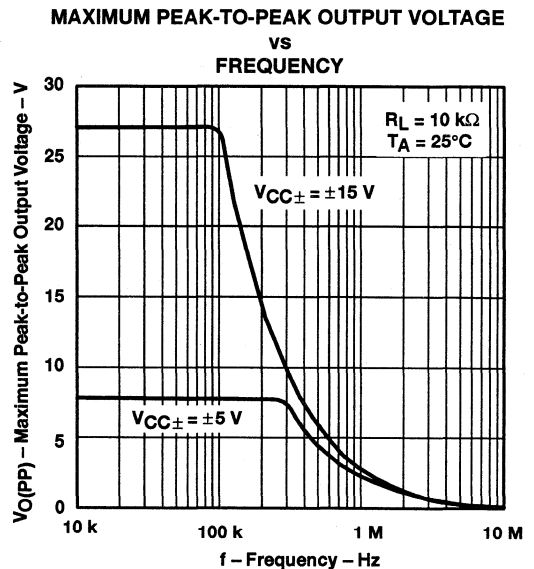
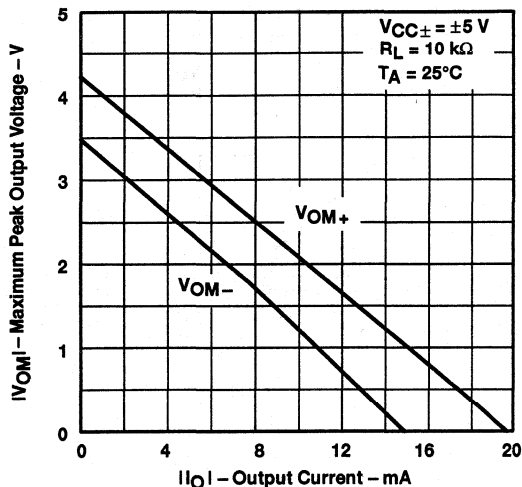


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

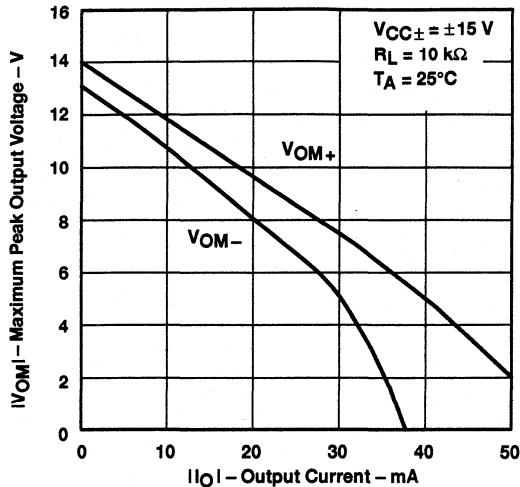
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
 vs  
**OUTPUT CURRENT**



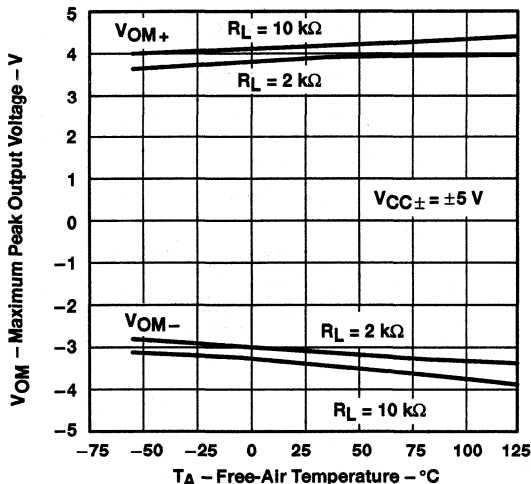
**Figure 18**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
 vs  
**OUTPUT CURRENT**



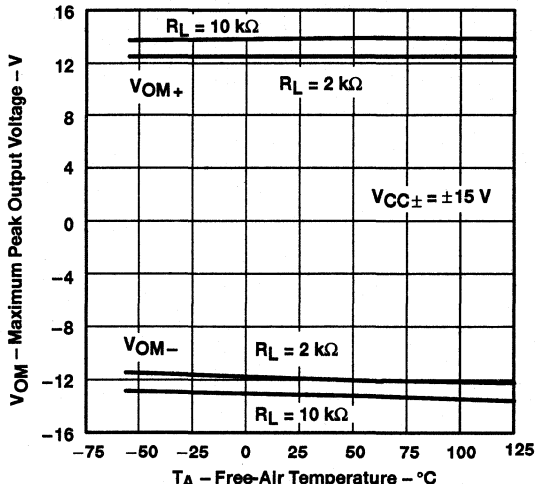
**Figure 19**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
 vs  
**FREE-AIR TEMPERATURE**



**Figure 20**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
 vs  
**FREE-AIR TEMPERATURE**



**Figure 21**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

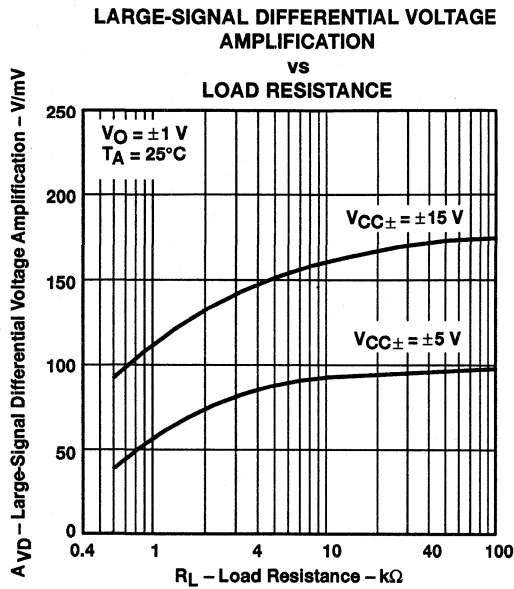


Figure 22

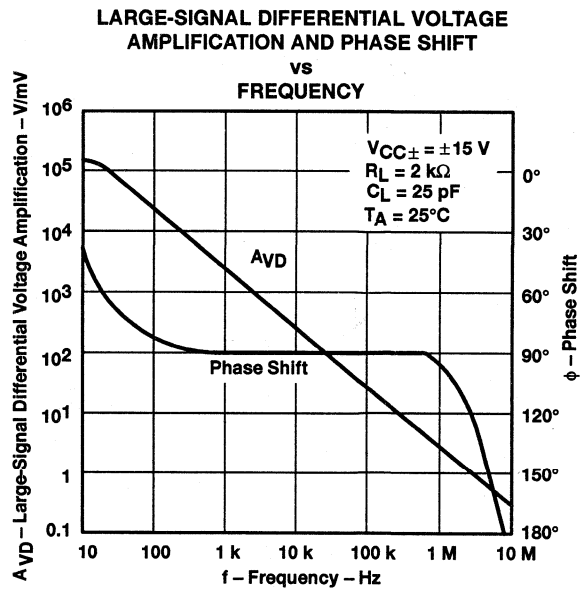


Figure 23

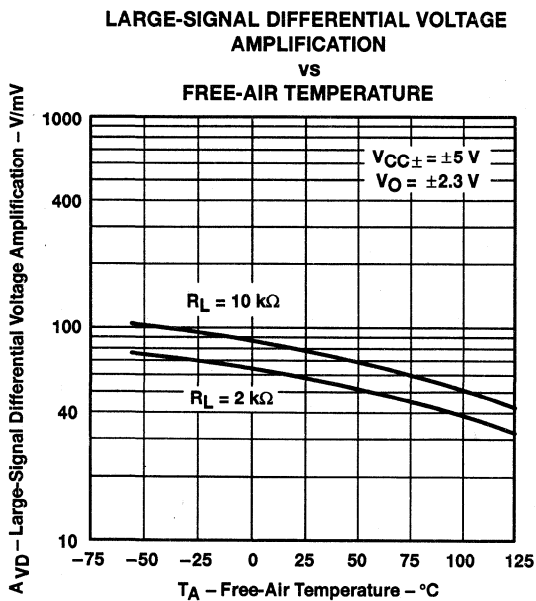


Figure 24

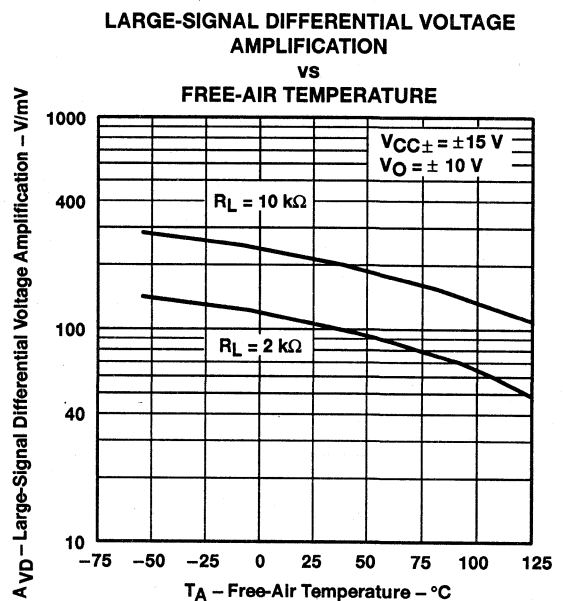


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

COMMON-MODE REJECTION RATIO  
 vs  
 FREQUENCY

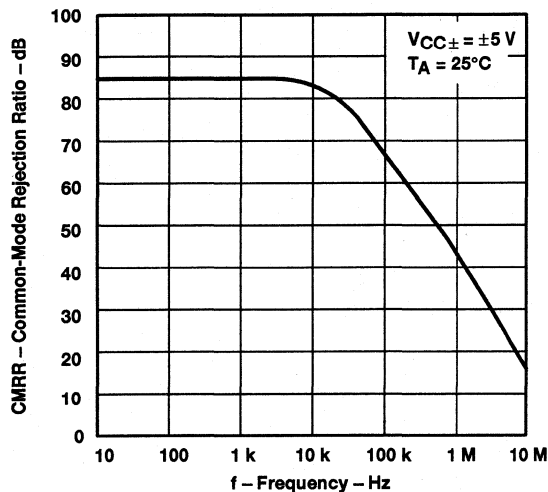


Figure 26

COMMON-MODE REJECTION RATIO  
 vs  
 FREQUENCY

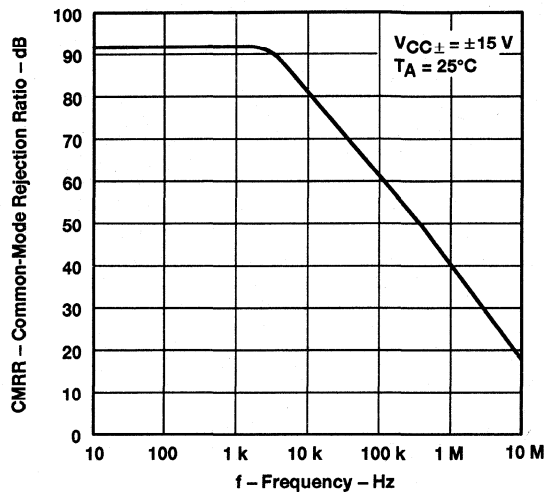


Figure 27

COMMON-MODE REJECTION RATIO  
 vs  
 FREE-AIR TEMPERATURE

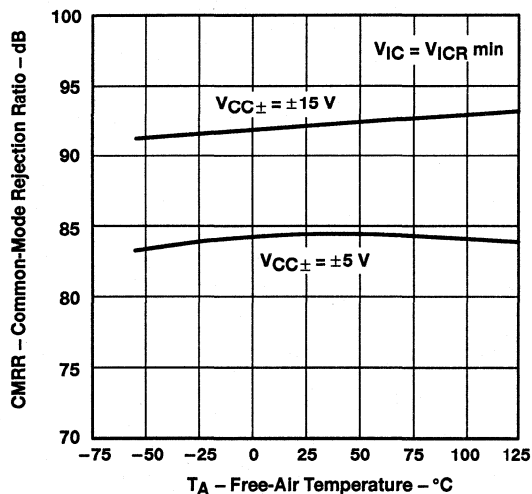


Figure 28

OUTPUT IMPEDANCE  
 vs  
 FREQUENCY

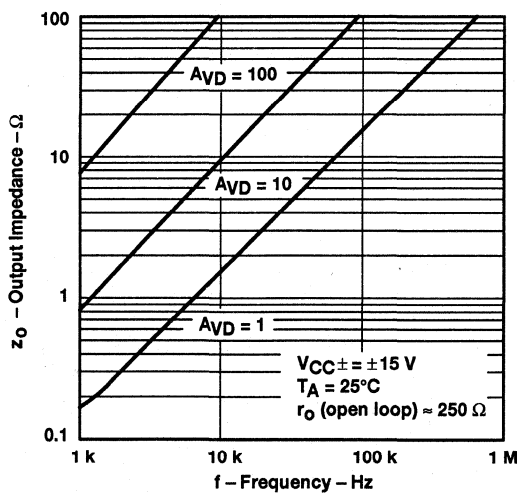


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

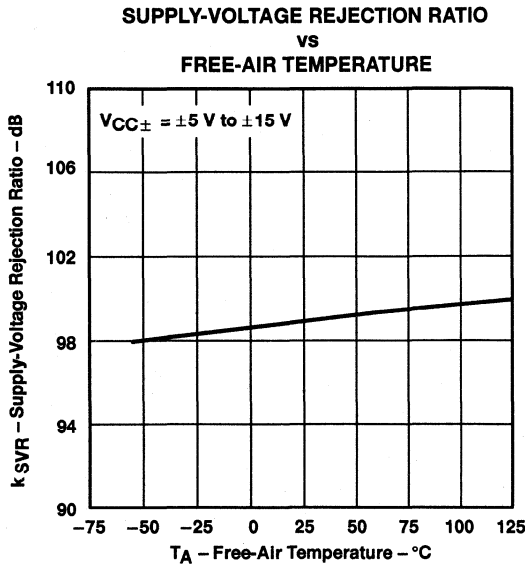


Figure 30

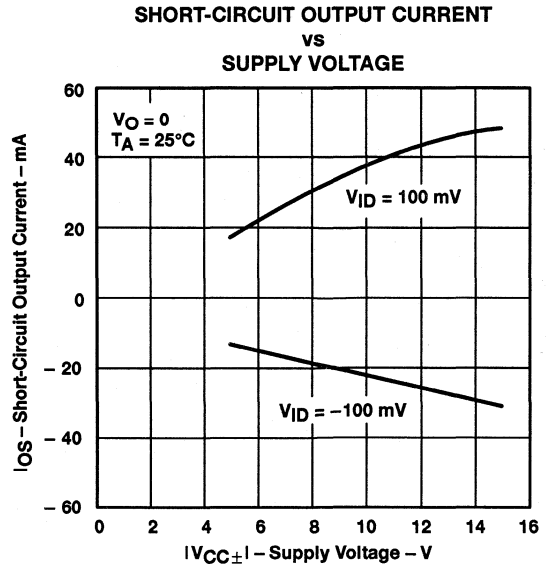


Figure 31

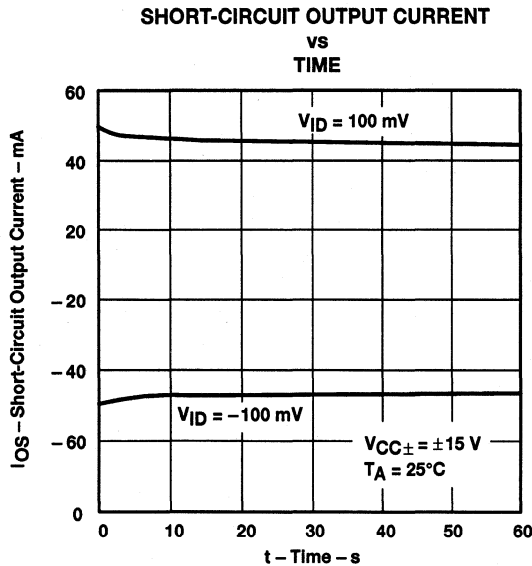


Figure 32

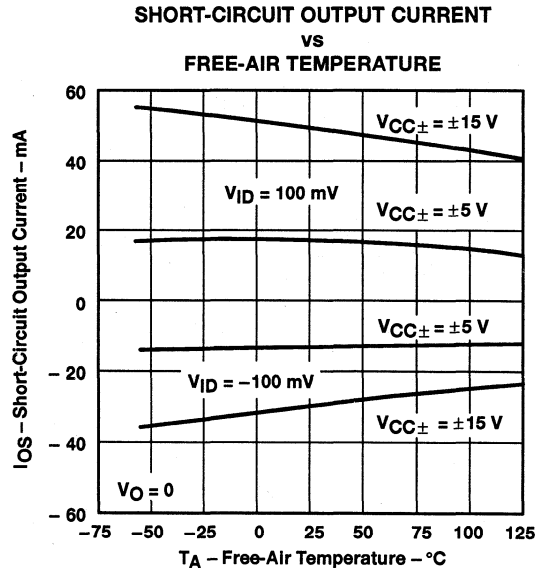


Figure 33

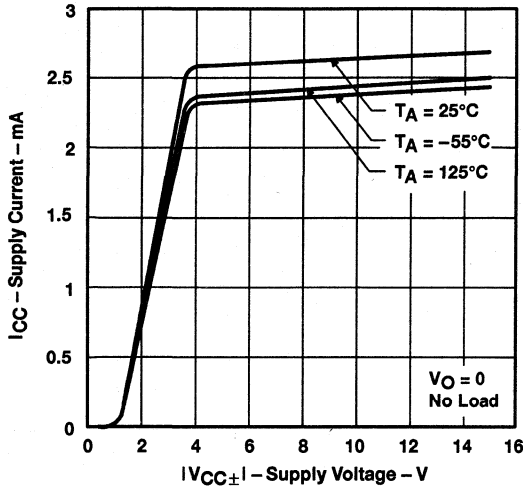
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

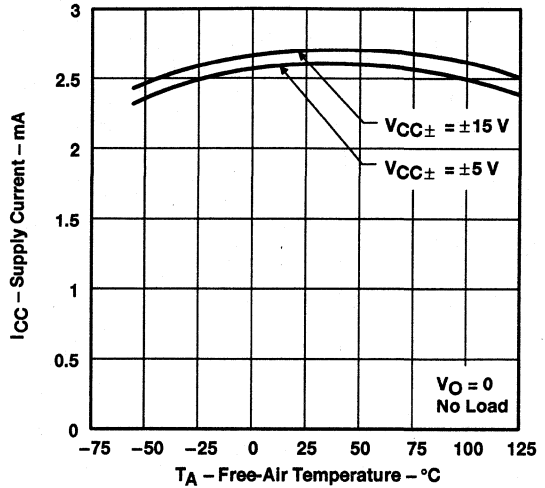
**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



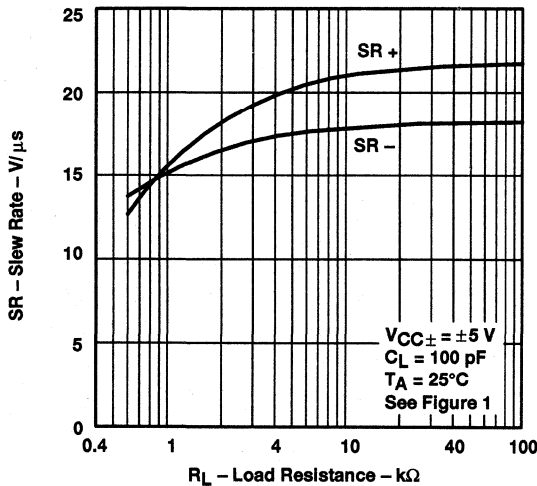
**Figure 34**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



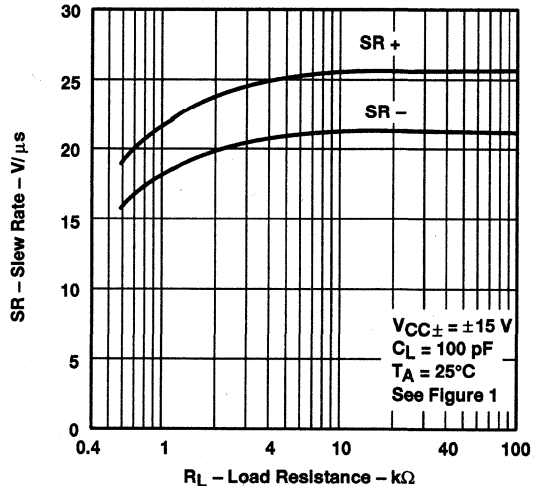
**Figure 35**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**



**Figure 36**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**



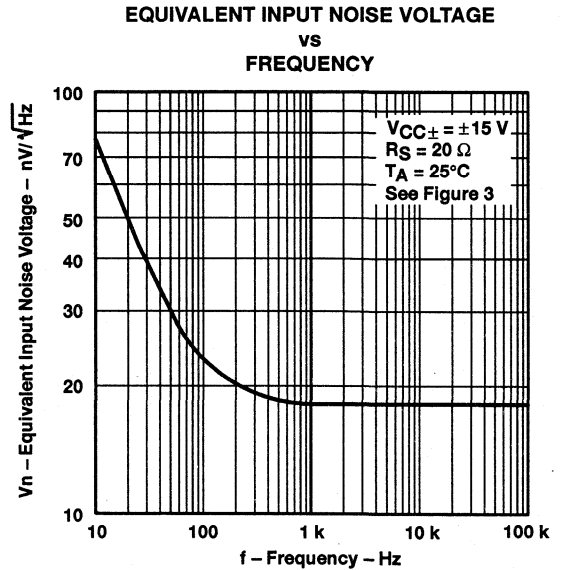
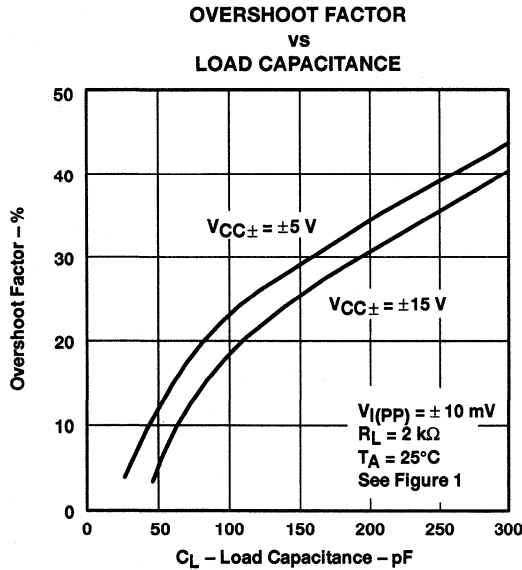
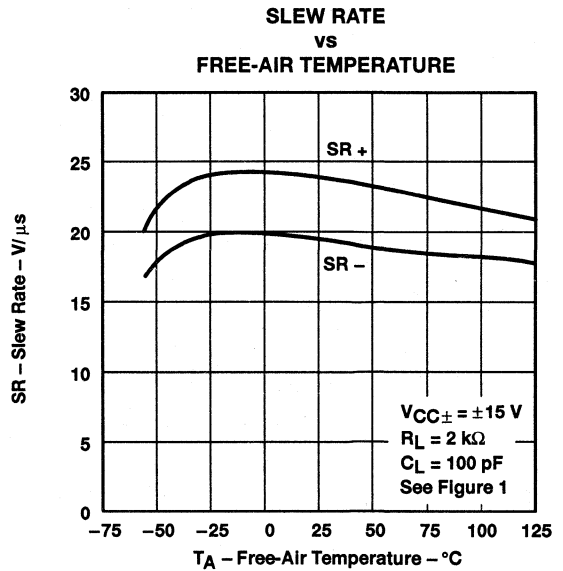
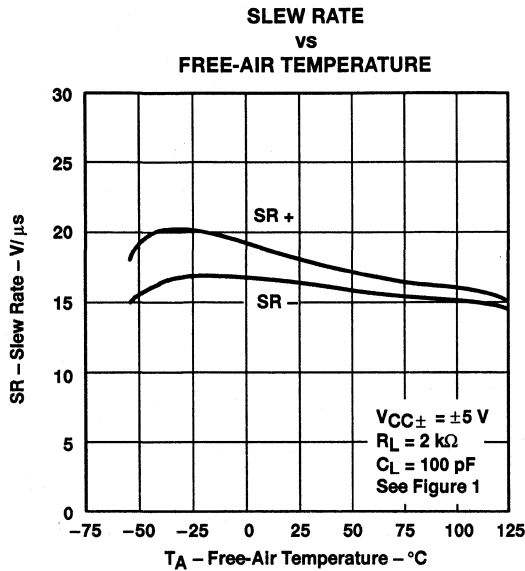
**Figure 37**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TYPICAL CHARACTERISTICS†



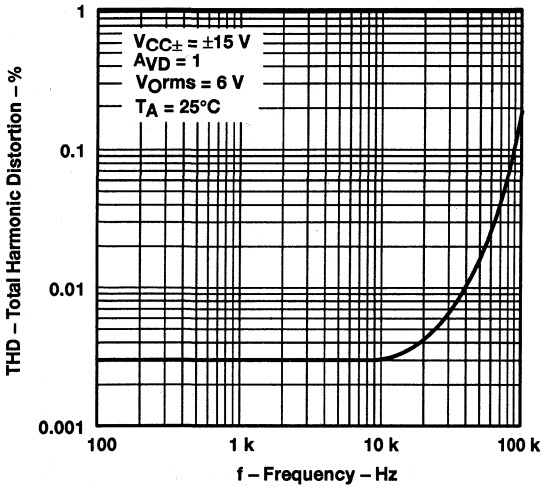
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D - JUNE 1988 - REVISED AUGUST 1994

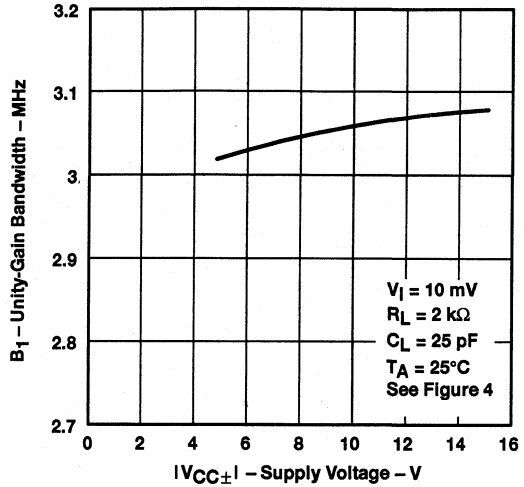
**TYPICAL CHARACTERISTICS†**

**TOTAL HARMONIC DISTORTION**  
**vs**  
**FREQUENCY**



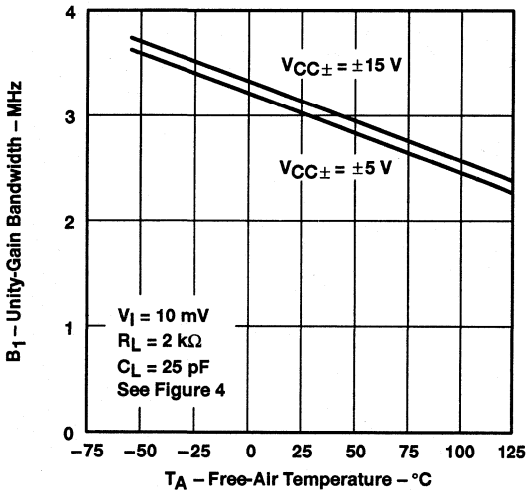
**Figure 42**

**UNITY-GAIN BANDWIDTH**  
**vs**  
**SUPPLY VOLTAGE**



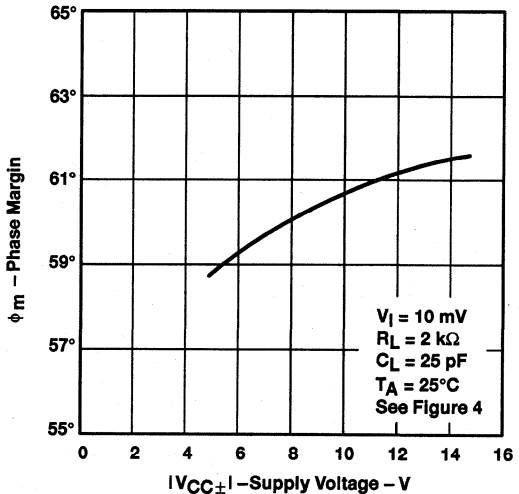
**Figure 43**

**UNITY-GAIN BANDWIDTH**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 44**

**PHASE MARGIN**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 45**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS†**

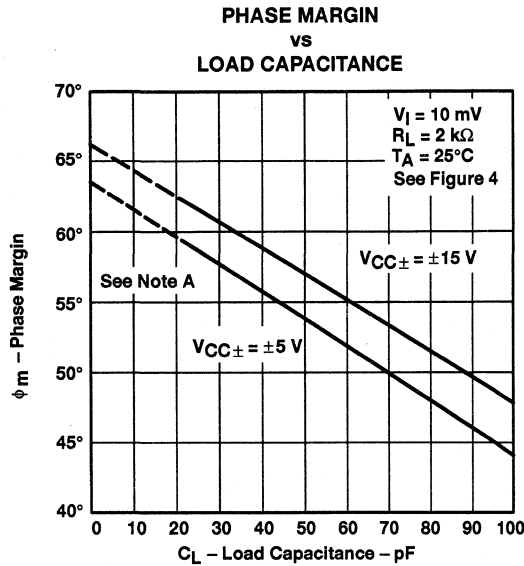


Figure 46

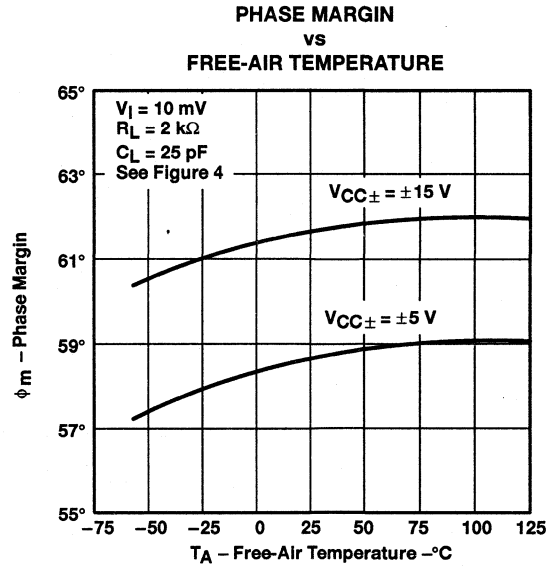


Figure 47

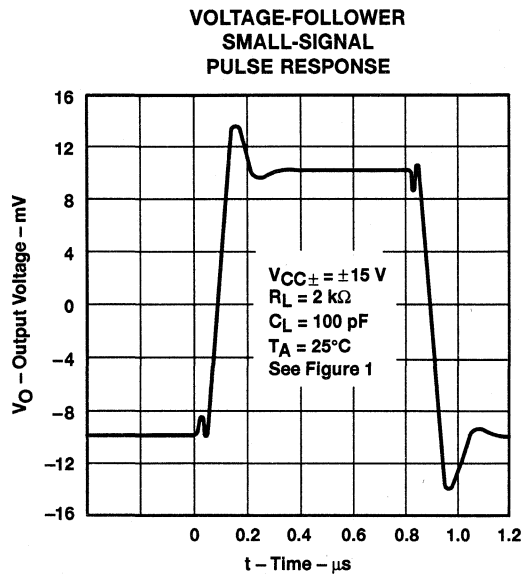


Figure 48

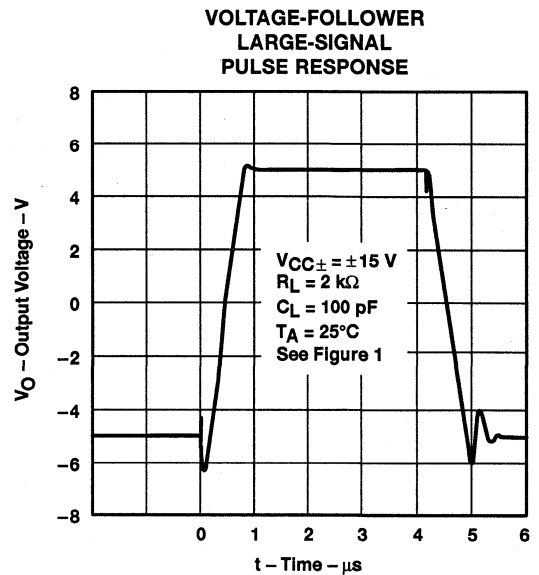


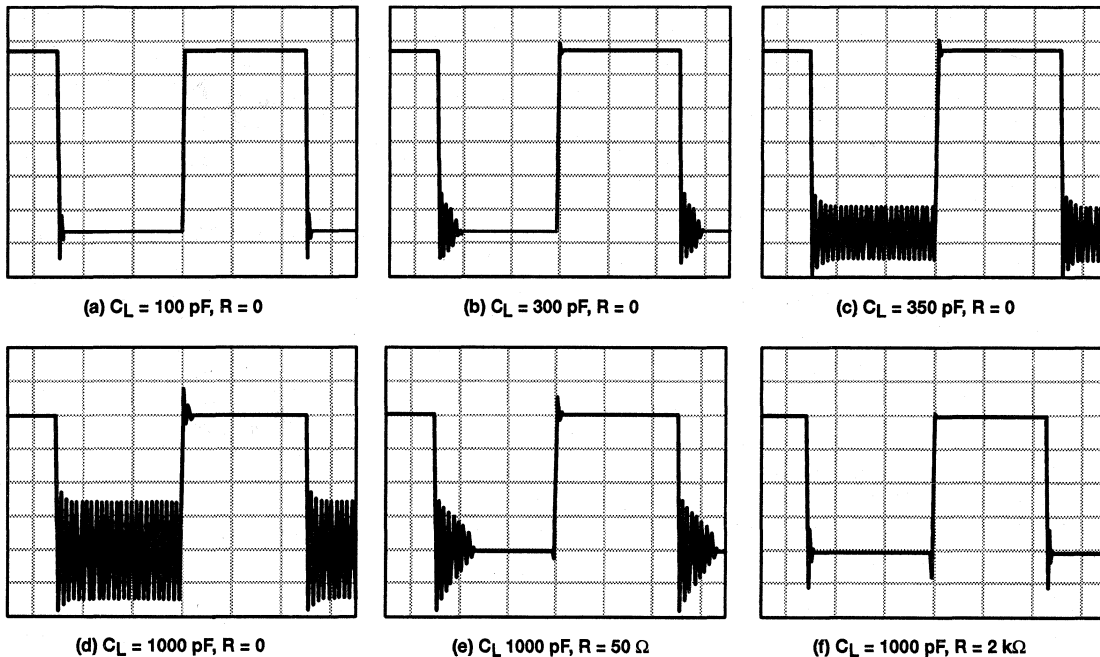
Figure 49

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

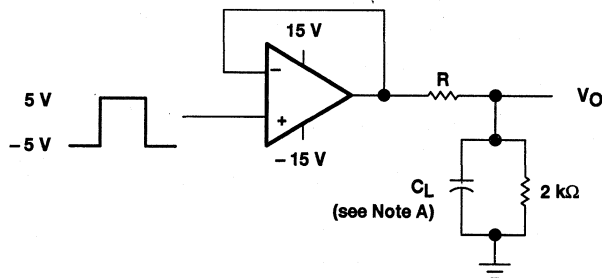
**APPLICATION INFORMATION**

**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with a 100-pF load capacitance (see Figure 51). The TL051 and TL051A drives higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).



**Figure 50. Effect of Capacitive Loads**

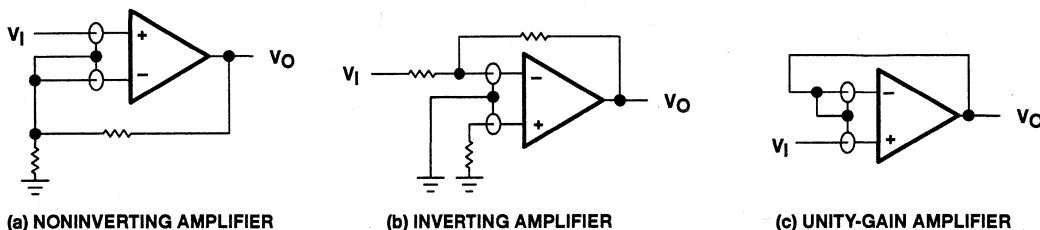


**Figure 51. Test Circuit for Output Characteristics**

**APPLICATION INFORMATION**

**input characteristics**

The TL051 and TL051A are specified with a minimum and a maximum input voltage that if exceeded at either input could cause the device to malfunction. Because of the extremely high input impedance and resulting low bias current requirements, the TL051 and TL051A are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.



**Figure 52. Use of Guard Rings**

**noise performance**

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL051 and TL051A result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 kΩ.

**TL051, TL051A, TL051Y**  
**ENHANCED-JFET PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS035D – JUNE 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

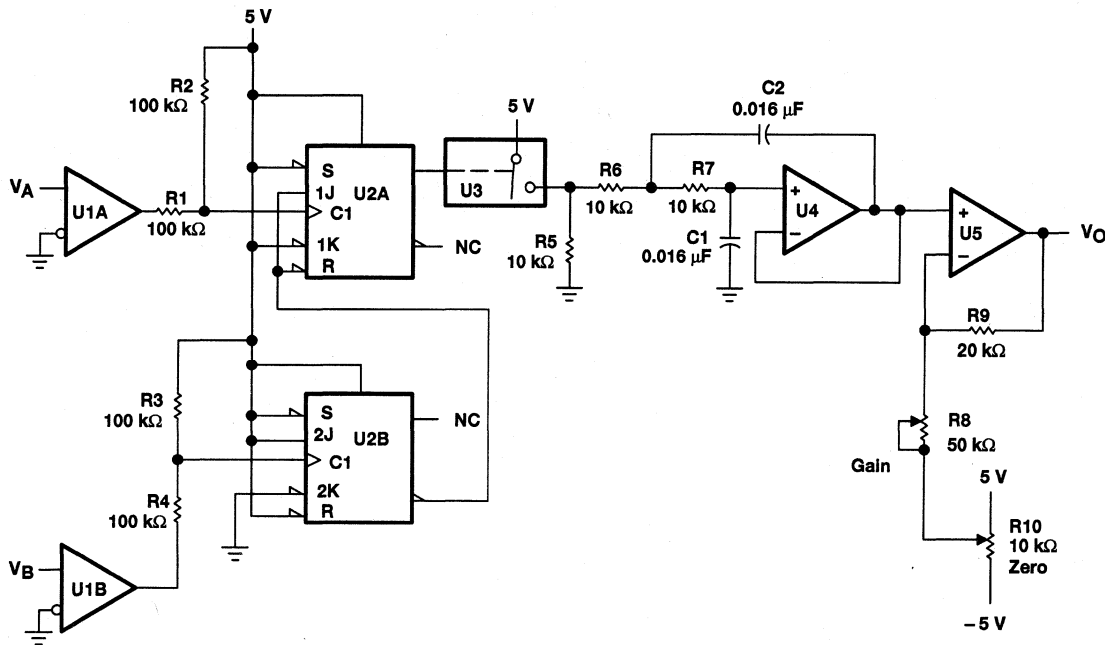
**phase meter**

The phase meter in Figure 53 produces an output voltage of 10 mV per degree of phase delay between the two input signals  $V_A$  and  $V_B$ . The reference signal  $V_A$  must be the same frequency as  $V_B$ . The TLC3702 comparators (U1) convert these two input sine waves into  $\pm 5$ -V square waves. Then R1 and R4 provide level shifting prior to the SN74HC109 dual J-K flip flop.

Flip-flop U2B is connected as a toggle flip-flop and generates a square wave at half the frequency of  $V_B$ . Flip-flop U2A also produces a square wave at half the input frequency. The pulse duration of U2A varies from zero to half the period where zero corresponds to zero phase delay between  $V_A$  and  $V_B$ , and half the period corresponds to  $V_B$  lagging  $V_A$  by 360 degrees.

The output pulse from U2A causes the TLC4066 (U3) switch to charge the TL051 (U4) integrator capacitors C1 and C2. As the phase delay approaches 360 degrees, the output of U2A approximates a square wave, and U4 has an output of almost 2.5 V. U5 acts as a noninverting amplifier with a gain of 1.44 in order to scale the 0-V to 2.5-V integrator output to a 0-V to 3.6-V output range.

R8 and R10 provide output gain and zero-level calibration. This circuit operates over a 100-Hz to 10-kHz frequency range.



NOTES: U1 = TLC3702;  $V_{CC\pm} = \pm 5$  V  
 U2 = SN74HC109  
 U3 = TLC4066  
 U4, U5 = TL051;  $V_{CC\pm} = \pm 5$  V

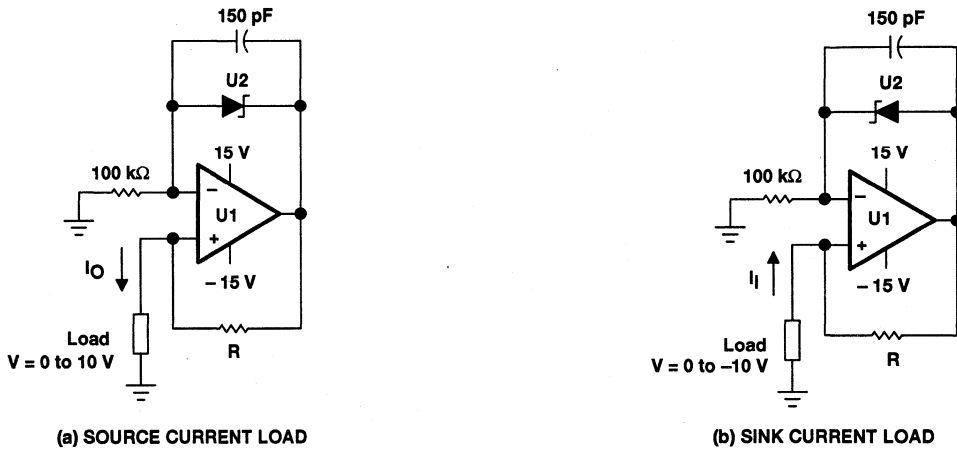
**Figure 53. Phase Meter**

**APPLICATION INFORMATION**

**precision constant-current source over temperature**

A precision current source benefits from the high input impedance and stability of Texas Instruments enhanced-JFET process. A low-current shunt regulator maintains 2.5 V between the inverting input and the output of the TL051. The negative feedback then forces 2.5 V across the current setting resistor R; therefore, the current to the load is simply 2.5 V divided by R.

Possible choices for the shunt regulator include the LT1004, LT1009, and LM385. If the regulator cathode connects to the operational amplifier output, this circuit sources load current. Similarly, if the cathode connects to the inverting input, the circuit sinks current from the load. To minimize output current change with temperature, R should be a metal-film resistor with a low temperature coefficient. Also, this circuit must be operated with split-voltage supplies.



NOTES: U1 = TL051  
 U2 = LM385, LT1004, or LT1009 voltage reference  
 $I = \frac{2.5 \text{ V}}{R}$ , R = low-temperature coefficient metal-film resistor

**Figure 54. Precision Constant-Current Source**





# TL052, TL052A, TL052Y ENHANCED-JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

- **Maximum Offset Voltage**  
800  $\mu\text{V}$  (TL052A)
- **High Slew Rate** . . . 17.8  $\text{V}/\mu\text{s}$  Typ at 25°C
- **Low Total Harmonic Distortion**  
0.003% Typ at  $R_L = 2 \text{ k}\Omega$
- **Low Noise Voltage** . . . 19  $\text{nV}/\sqrt{\text{Hz}}$
- **Low Input Bias Currents** . . . 30 pA Typ

## description

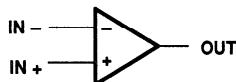
The TL052 and TL052A dual operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset voltage drift.

This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL052 well suited for new state-of-the-art designs as well as existing design upgrades. The TL052 has been designed to be functionally compatible, as well as pin compatible, with the TL072 and TL082. Two offset voltage grades are available: TL052 (1.5 mV max) and TL052A (800  $\mu\text{V}$  max).

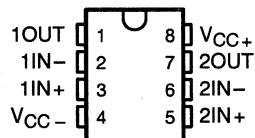
A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

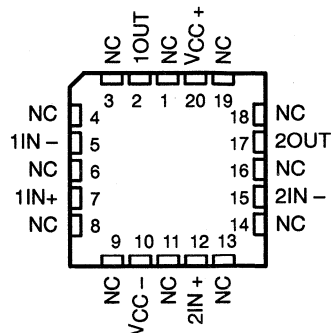
## symbol (each amplifier)



D, JG, OR P PACKAGE  
(TOP VIEW)

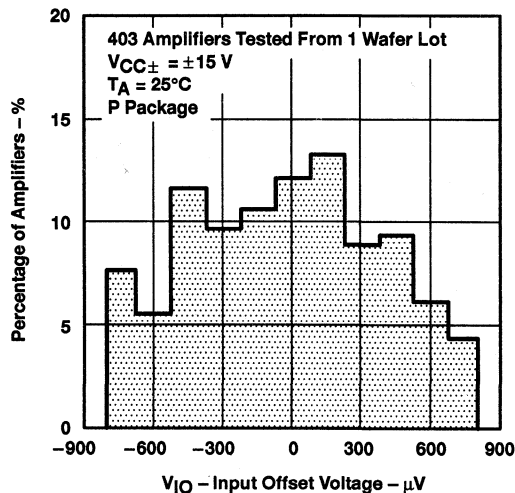


FK PACKAGE  
(TOP VIEW)



NC – No internal connection

DISTRIBUTION OF TL052A  
INPUT OFFSET VOLTAGE



PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TL052, TL052A, TL052Y ENHANCED-JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

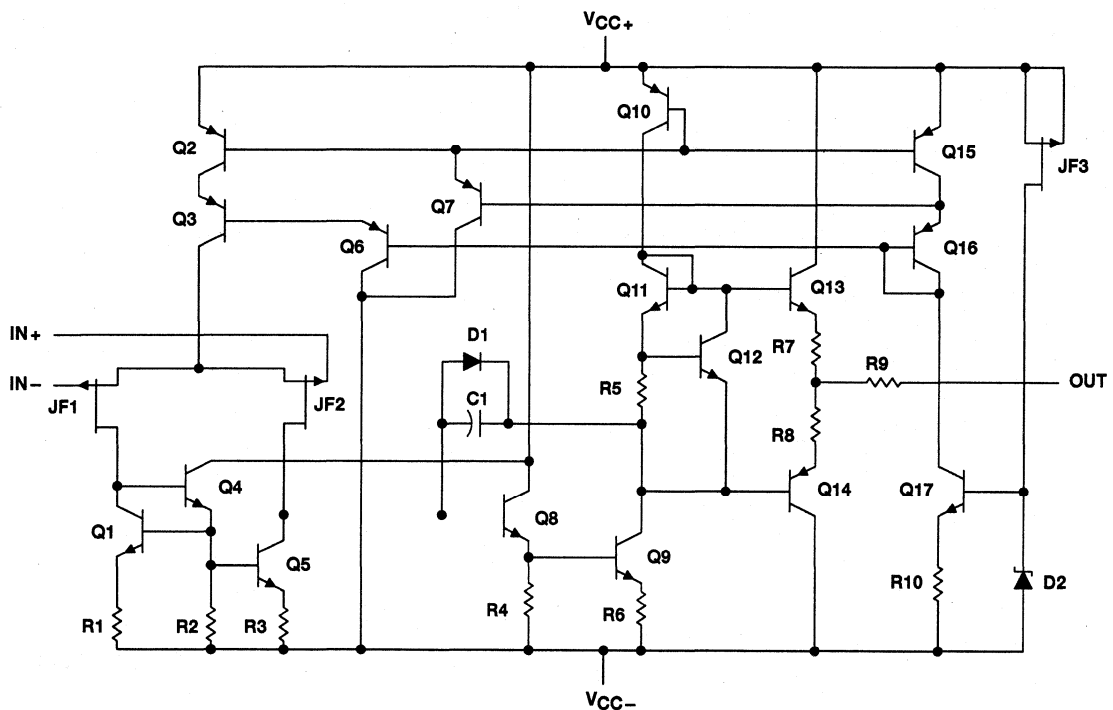
SLOS036C - JUNE 1988 - REVISED AUGUST 1994

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES				CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	
0°C to 70°C	800 μV 1500 μV	TL052ACD TL052CD	—	—	TL052ACP TL052CP	TL052Y
-40°C to 85°C	800 μV 1500 μV	TL052AID TL052ID	—	—	TL052AIP TL052IP	—
-55°C to 125°C	800 μV 1500 μV	TL052AMD TL052MD	TL052AMFK TL052MFK	TL052AMJG TL052MJG	TL052AMP TL052MP	—

The D packages are available taped and reeled. Add R suffix to device type (e.g., TL052CDR).

## equivalent schematic (each amplifier)

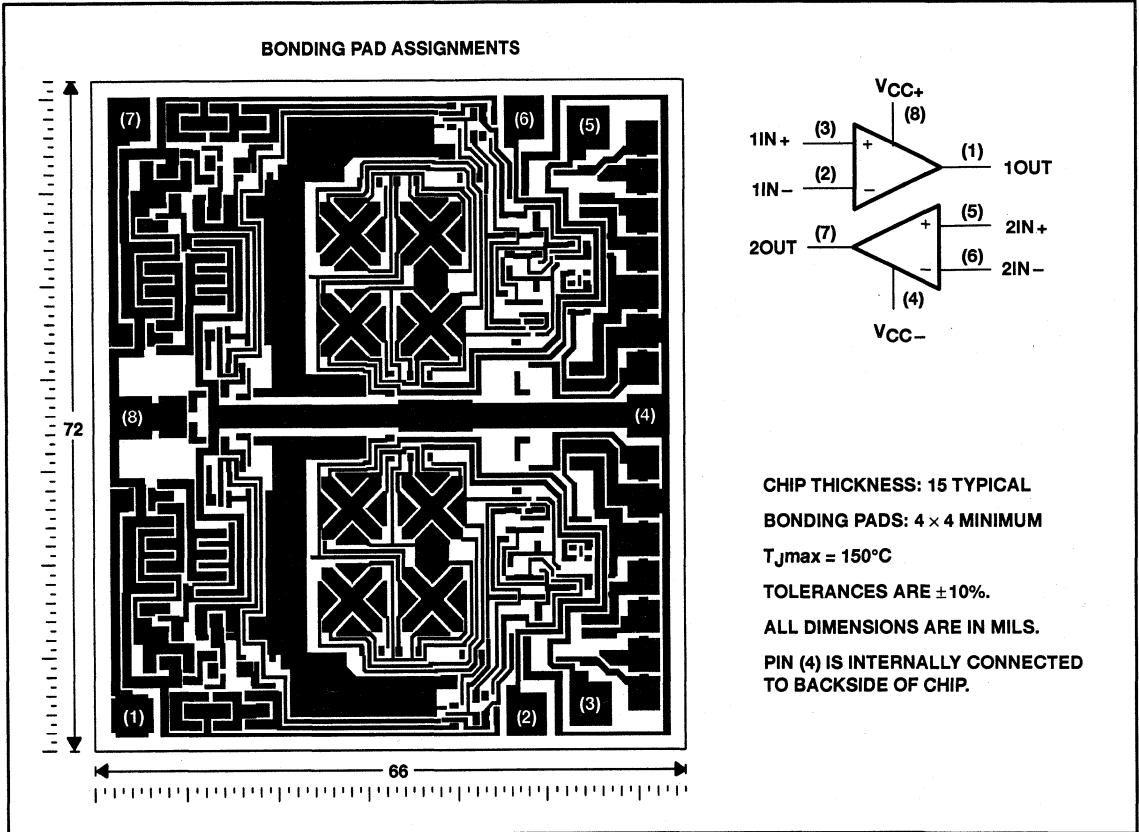


COMPONENT COUNT†	
Transistors	34
Resistors	19
Diodes	3
Capacitors	2

† Includes both amplifiers and all bias and trim circuitry

**TL052Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL052. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage (see Note 2)	±30 V
Input voltage range, $V_I$ (any input, see Notes 1 and 3)	±15 V
Input current, $I_I$ (each input)	±1 mA
Output current, $I_O$ (each output)	±80 mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC\pm}$		±5	±15	±5	±15	±5	±15	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5\text{ V}$	-1	4	-1	4	-1	4	V
	$V_{CC\pm} = \pm 15\text{ V}$	-11	11	-11	11	-11	11	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS		T <sub>A</sub> †	TL052C, TL052AC						UNIT
				V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL052C	25°C	0.73	3.5	0.65	1.5	mV	
				Full range		4.5	2.5			
			TL052AC	25°C	0.51	2.8	0.4	0.8		
				Full range		3.8	1.8			
α <sub>VIO</sub>	Temperature coefficient of input offset voltage (see Note 5)	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL052C	25°C to 70°C	8		8		μV/°C	
			TL052AC	25°C to 70°C	8		6	25		
	Input offset voltage long-term drift (see Note 6)	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	V <sub>IC</sub> = 0,	25°C	0.04		0.04		μV/mo	
I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	4	100	5	100	pA	
				70°C	0.02	1	0.025	1	nA	
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	20	200	30	200	pA	
				70°C	0.15	4	0.2	4	nA	
V <sub>ICR</sub>	Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V	
				Full range	-1 to 4		-11 to 11			
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3	4.2	13	13.9	V	
				Full range	3		13			
		R <sub>L</sub> = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
			Full range	2.5		11.5				
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-2.5	-3.5	-12	-13.2	V	
				Full range	-2.5		-12			
		R <sub>L</sub> = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
			Full range	-2.3		-11				
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ,	See Note 7	25°C	25	59	50	105	V/mV	
				0°C	30	65	60	129		
				70°C	20	46	30	85		
r <sub>i</sub>	Input resistance			25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω	
c <sub>i</sub>	Input capacitance			25°C	10		12		pF	
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	65	85	75	93	dB	
				0°C	65	84	75	92		
				70°C	65	84	75	91		

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature (continued)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052C, TL052AC						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	99		75	99	dB	
			0°C	75	98		75	98		
			70°C	75	97		75	97		
I <sub>CC</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C		4.6	5.6		4.8	5.6	mA
			0°C		4.7	6.4		4.8	6.4	
			70°C		4.4	6.4		4.6	6.4	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C	120			120			dB

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052C, TL052AC						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR +	Slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	17.8			9 20.7			V/μs
SR -	Negative slew rate at unity gain		Full range				8			
			25°C	15.4			9 17.8			
t <sub>r</sub>	Rise time	V <sub>I</sub> (PP) = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56			ns
			0°C	54			55			
t <sub>f</sub>	Fall time		70°C	63			63			
			25°C	55			57			
Overshoot factor			0°C	54			56			
			70°C	62			64			
		25°C	24%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	71			71			nV/√Hz
			f = 1 kHz	19			19 30			
			25°C	4			4			
V <sub>N</sub> (PP)	Peak-to-peak equivalent input noise current	f = 10 Hz to 10 kHz	25°C	4			4			μV
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01			pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 9	25°C	0.003%			0.003%			
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3			MHz
			0°C	3.2			3.2			
			70°C	2.6			2.7			
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	60°			63°			
			0°C	59°			63°			
			70°C	60°			63°			

† Full range is 0°C to 70°C.

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.  
 9. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub>(RMS) = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O</sub>(RMS) = 6 V.



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS		T <sub>A</sub> †	TL052I, TL052AI						UNIT
				V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL052I	25°C	0.73 3.5		0.65 1.5		mV	
				Full range	5.3		3.3			
			TL052AI	25°C	0.51 2.8		0.4 0.8			
				Full range	4.6		2.6			
α <sub>VIO</sub>	Temperature coefficient (see Note 5)		TL052I	25°C to 85°C	7		6		μV/°C	
			TL052AI	25°C to 85°C	6		6 25			
	Input offset voltage long-term drift (see Note 6)	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	V <sub>IC</sub> = 0,	25°C	0.04		0.04		μV/mo	
I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	4 100		5 100		pA	
				85°C	0.06 10		0.07 10		nA	
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	20 200		30 200		pA	
				85°C	0.6 20		0.7 20		nA	
V <sub>ICR</sub>	Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V	
				Full range	-1 to 4		-11 to 11			
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3 4.2		13 13.9		V	
				Full range	3		13			
				25°C	2.5 3.8		11.5 12.7			
				Full range	2.5		11.5			
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-2.5 -3.5		-12 -13.2		V	
				Full range	-2.5		-12			
				25°C	-2.3 -3.2		-11 -12			
				Full range	-2.3		-11			
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 7		25°C	25 59		50 105		V/mV	
				-40°C	30 74		60 145			
				85°C	20 43		30 76			
r <sub>i</sub>	Input resistance			25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω	
c <sub>i</sub>	Input capacitance			25°C	10		12		pF	
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	65 85		75 93		dB	
				-40°C	65 83		75 90			
				85°C	65 84		75 93			

† Full range is -40°C to 85°C.

- NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters
6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.
7. At V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL052, TL052A, TL052Y ENHANCED-JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052I, TL052AI						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	99		75	99	dB	
			-40°C	75	98		75	98		
			85°C	75	99		75	99		
I <sub>CC</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C		4.6	5.6		4.8	5.6	mA
			-40°C		4.5	6.4		4.7	6.4	
			85°C		4.4	6.4		4.6	6.4	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C		120			120	dB	

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052I, TL052AI						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR +	Slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	17.8			9	20.7		V/μs
SR -	Negative slew rate at unity gain		Full range				8			
			25°C	15.4			9	17.8		
t <sub>r</sub>	Rise time	V <sub>I</sub> (PP) = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56		ns	
			-40°C	52			53			
			85°C	64			65			
t <sub>f</sub>	Fall time		25°C	55			57			
			-40°C	51			53			
			85°C	64			65			
Overshoot factor		25°C	24%			19%				
		-40°C	24%			19%				
		85°C	24%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			71			
			f = 1 kHz	25°C			19			30
V <sub>N</sub> (PP)	Peak-to-peak equivalent input noise current	f = 10 Hz to 10 kHz	25°C	4			4		μV	
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01		pA/√Hz	
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 9	25°C	0.003%			0.003%			
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3		MHz	
			-40°C	3.5			3.6			
			85°C	2.5			2.6			
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	60°			63°			
			-40°C	58°			61°			
			85°C	60°			63°			

† Full range is -40°C to 85°C.

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.

9. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub>(RMS) = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O</sub>(RMS) = 6 V.





**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052M, TL052AM				UNIT		
			V <sub>CC±</sub> = ±5 V		V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN		TYP	MAX
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL052M	25°C	0.73	3.5	0.65	1.5	mV
				Full range	6.5		4.5		
			TL052AM	25°C	0.51	2.8	0.4	0.8	
				Full range	5.8		3.8		
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	TL052M	25°C to 125°C	10		9		μV/°C	
			TL052AM	25°C to 125°C	9		8		
	Input offset voltage long-term drift (see Note 6)	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	V <sub>IC</sub> = 0,	25°C	0.04		0.04		μV/mo
I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	4	100	5	100	pA
				125°C	1	20	2	20	nA
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	20	200	30	200	pA
				125°C	10	50	20	50	nA
V <sub>ICR</sub>	Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V
				Full range	-1 to 4		-11 to 11		
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	3	4.2	13	13.9	V
				Full range	3		13		
		R <sub>L</sub> = 2 kΩ	25°C	2.5	3.8	11.5	12.7		
			Full range	2.5		11.5			
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ		25°C	-2.5	-3.5	-12	-13.2	V
				Full range	-2.5		-12		
		R <sub>L</sub> = 2 kΩ	25°C	-2.3	-3.2	-11	-12		
			Full range	-2.3		-11			
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ,	See Note 7	25°C	25	59	50	105	V/mV
				-55°C	30	76	60	149	
				125°C	10	32	15	49	
r <sub>i</sub>	Input resistance			25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω
c <sub>i</sub>	Input capacitance			25°C	10		12		pF
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω		25°C	65	85	75	93	dB
				-55°C	65	83	75	92	
				125°C	65	84	75	94	
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0,	R <sub>S</sub> = 50 Ω	25°C	75	99	75	99	dB
				-55°C	75	98	75	98	
				125°C	75	100	75	100	
I <sub>CC</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 0,	No load	25°C	4.6	5.6	4.8	5.6	mA
				-55°C	4.4	6.4	4.5	6.4	
				125°C	4.2	6.4	4.4	6.4	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100		25°C	120		120		dB

† Full range is -55°C to 125°C.

NOTES: 6. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

7. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



# TL052, TL052A, TL052Y ENHANCED-JFET PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL052M, TL052AM						UNIT		
			V <sub>CC±</sub> = ± 5 V			V <sub>CC±</sub> = ± 15 V					
			MIN	TYP	MAX	MIN	TYP	MAX			
SR +	Positive slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	17.8			9 20.7			V/μs	
			Full range				8				
SR –	Negative slew rate at unity gain		25°C	15.4			9 17.8				
			Full range				8				
t <sub>r</sub>	Rise time		25°C	55			56			ns	
			–55°C	51			52				
			125°C	68			68				
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ± 10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57				
			–55°C	51			52				
			125°C	68			69				
Overshoot factor			25°C	24%			19%				
			–55°C	25%			19%				
			125°C	25%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			71			nV/√Hz	
			f = 1 kHz	25°C			19				
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise current		f = 10 Hz to 10 kHz	25°C			4			μV	
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01			pA/√Hz	
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz,	R <sub>L</sub> = 2 kΩ, See Note 9	25°C			0.003%			0.003%	
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3			MHz
				–55°C	3.6			3.7			
				125°C	2.3			2.4			
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	60°			63°			
				–55°C	57°			61°			
				125°C	60°			63°			

† Full range is –55°C to 125°C.

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

9. For V<sub>CC±</sub> = ±5 V, V<sub>O(RMS)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(RMS)</sub> = 6 V.



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL052Y						UNIT				
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V							
			MIN	TYP	MAX	MIN	TYP	MAX					
V <sub>IO</sub>	Input offset voltage	25°C	0.73			3.5			0.65	1.5	mV		
α <sub>VIO</sub>	Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	25°C to 70°C			8			8		μV/°C		
	Input offset voltage long-term drift	25°C	0.04			0.04					μV/mo		
I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	4			100			5		100	pA	
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	20			200			30		200	pA	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-1 to 4	-2.3 to 5.6		-11 to 11		-12.3 to 15.6				V	
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ R <sub>L</sub> = 2 kΩ	25°C			3 to 2.5		4.2 to 3.8		13 to 11.5		13.9 to 12.7	V
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ R <sub>L</sub> = 2 kΩ	25°C			-2.5 to -2.3		-3.5 to -3.2		-12 to -11		-13.2 to -12	V
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 7	25°C			25		59		50		105	V/mV
r <sub>i</sub>	Input resistance		25°C			10 <sup>12</sup>			10 <sup>12</sup>			Ω	
c <sub>i</sub>	Input capacitance		25°C			10			12			pF	
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			65		85		75		93	dB
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			75		99		75		99	dB
I <sub>CC</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 0, No load	25°C			4.6		5.6		4.8		5.6	mA
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C			120			120			dB	

NOTE 7. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V; at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature**

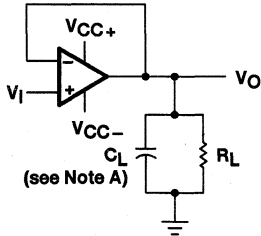
PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TL052Y						UNIT		
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V					
			MIN	TYP	MAX	MIN	TYP	MAX			
SR +	Positive slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 8	25°C	17.8			9 20.7			V/μs	
SR -	Negative slew rate at unity gain		25°C	15.4			9 17.8				
t <sub>r</sub>	Rise time	V <sub>I</sub> (PP) = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			56			ns	
t <sub>f</sub>	Fall time			55			57				
	Overshoot factor			24%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 5)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C	71			71			nV/√Hz
			f = 1 kHz	25°C	19			19 30			
V <sub>N</sub> (PP)	Peak-to-peak equivalent input noise current		f = 10 Hz to 10 kHz	25°C	4			4			μV
I <sub>n</sub>	Equivalent input noise current		f = 1 kHz	25°C	0.01			0.01			pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz,	R <sub>L</sub> = 2 kΩ, See Note 9	25°C	0.003%			0.003%			
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	3			3			MHz
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	60°			63°			

NOTES: 5. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

8. For V<sub>CC±</sub> = ±5 V, V<sub>I</sub>(PP) = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I</sub>(PP) = ±5 V.  
 9. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub>(RMS) = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O</sub>(RMS) = 6 V.



PARAMETER MEASUREMENT INFORMATION



NOTE A:  $C_L$  includes fixture capacitance.

Figure 1. Slew Rate, Rise/Fall Time, and Overshoot Test Circuit

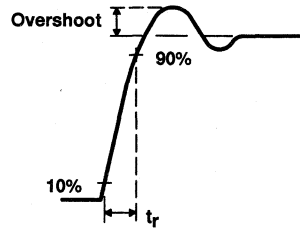


Figure 2. Rise Time and Overshoot Waveform

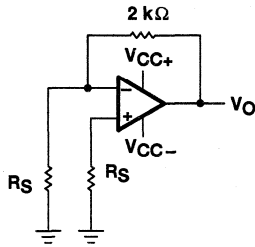
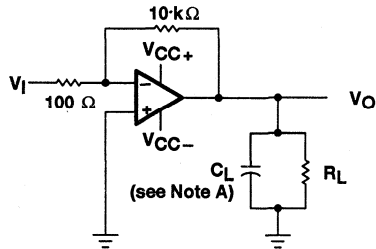


Figure 3. Noise-Voltage Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit

typical values

Typical values as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL052 and TL052A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket, and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.

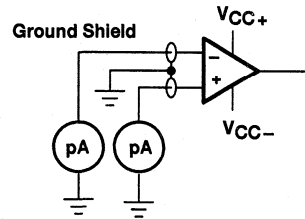


Figure 5. Input-Bias and Offset-Current Test Circuit

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Input bias current	vs Common-mode input voltage	9
		vs Free-air temperature	8
$V_{IC}$	Common-mode input voltage	vs Supply voltage	10
		vs Free-air temperature	11
$V_O$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage swing	vs Supply voltage	14
		vs Frequency	15, 16, 17
		vs Output current	18, 19
		vs Free-air temperature	20, 21
$A_{VD}$	Differential voltage amplification	vs Load resistance	22
		vs Frequency	23
		vs Free-air temperature	24, 25
$Z_o$	Output impedance	vs Frequency	29
$CMRR$	Common-mode rejection ratio	vs Frequency	26, 27
		vs Free-air temperature	28
$KSVR$	Supply-voltage rejection ratio	vs Free-air temperature	30
$I_{OS}$	Short-circuit output current	vs Supply voltage	31
		vs Time	32
		vs Free-air temperature	33
$I_{CC}$	Supply current	vs Supply voltage	34
		vs Free-air temperature	35
$SR$	Slew rate	vs Load resistance	36, 37
		vs Free-air temperature	38, 39
	Overshoot factor	vs Load capacitance	40
$V_n$	Equivalent input noise voltage	vs Frequency	41
$THD$	Total harmonic distortion	vs Frequency	42
$B_1$	Unity-gain bandwidth	vs Supply voltage	43
		vs Free-air temperature	44
$\phi_m^*$	Phase margin	vs Supply voltage	45
		vs Load capacitance	46
		vs Free-air temperature	47
	Phase shift	vs Frequency	23
	Pulse response	Small-signal	48
		Large-signal	49



TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TL052  
 INPUT OFFSET VOLTAGE

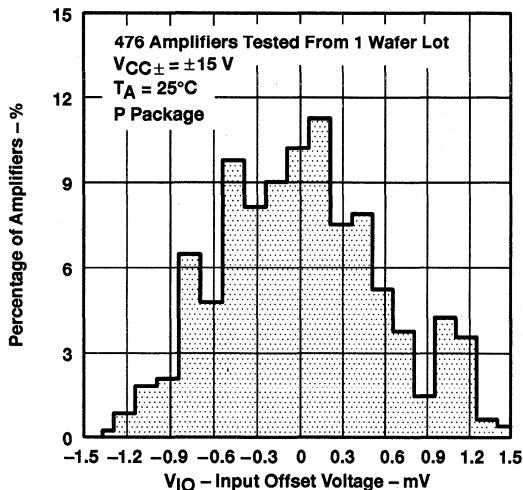


Figure 6

DISTRIBUTION OF TL052  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

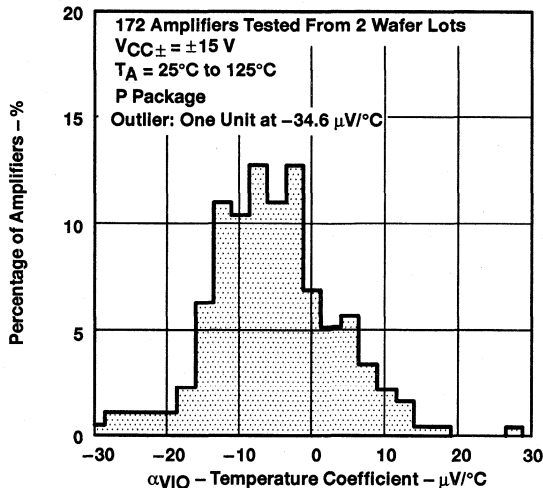


Figure 7

INPUT BIAS CURRENT AND  
 INPUT OFFSET CURRENT  
 vs  
 FREE-AIR TEMPERATURE

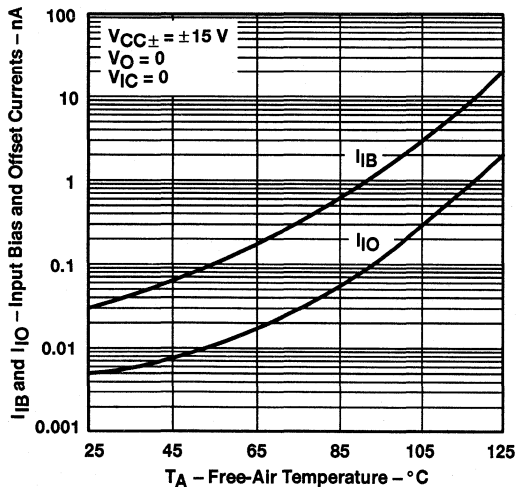


Figure 8

INPUT BIAS CURRENT  
 vs  
 COMMON-MODE INPUT VOLTAGE

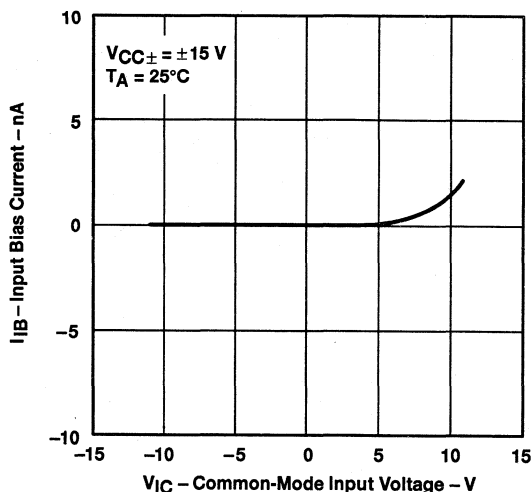


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS**  
**vs**  
**SUPPLY VOLTAGE**

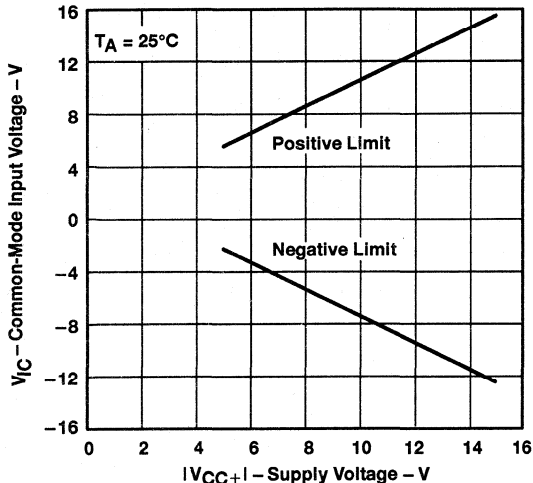


Figure 10

**COMMON-MODE INPUT VOLTAGE RANGE LIMITS**  
**vs**  
**FREE-AIR TEMPERATURE**

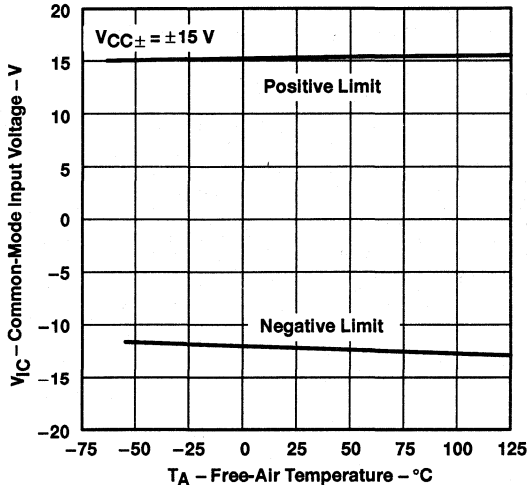


Figure 11

**OUTPUT VOLTAGE**  
**vs**  
**DIFFERENTIAL INPUT VOLTAGE**

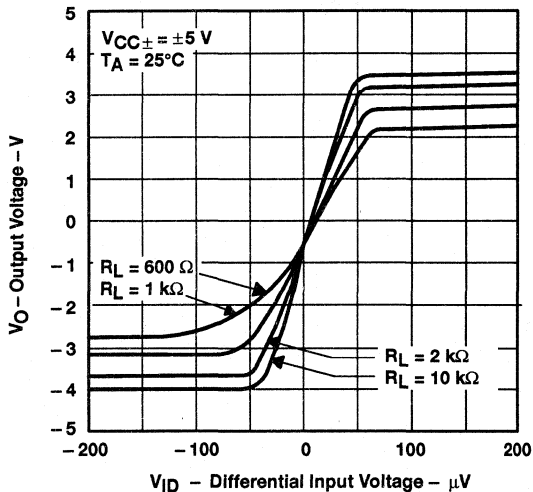


Figure 12

**OUTPUT VOLTAGE**  
**vs**  
**DIFFERENTIAL INPUT VOLTAGE**

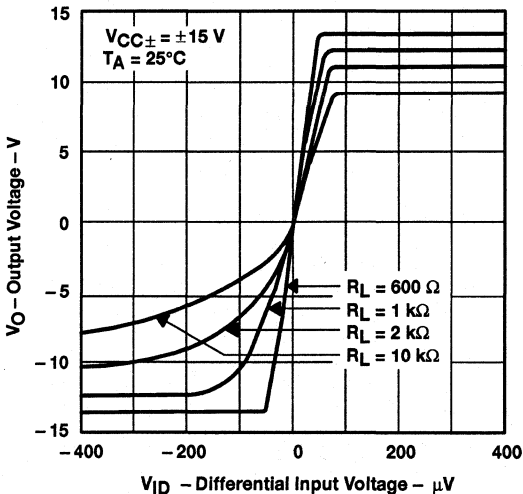


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TYPICAL CHARACTERISTICS†

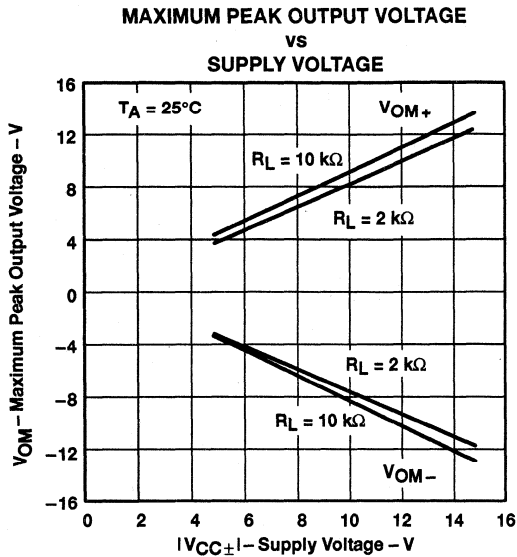


Figure 14

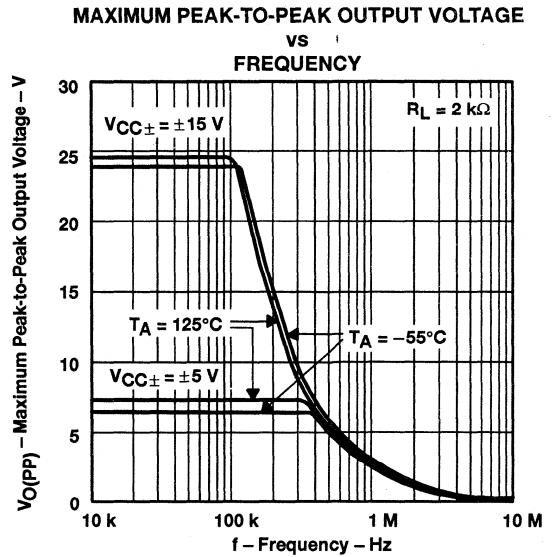


Figure 15

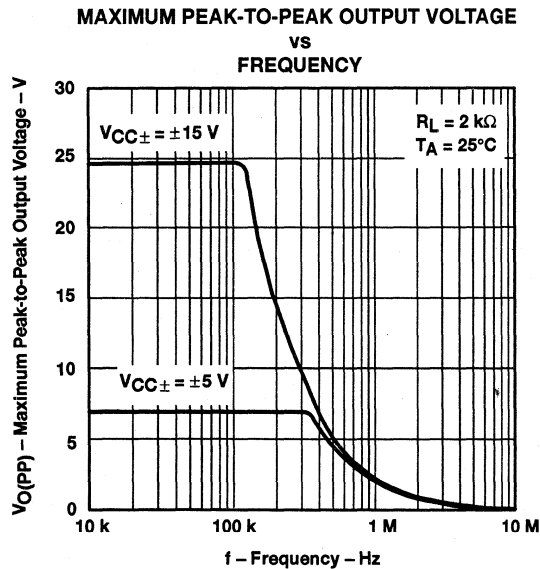


Figure 16

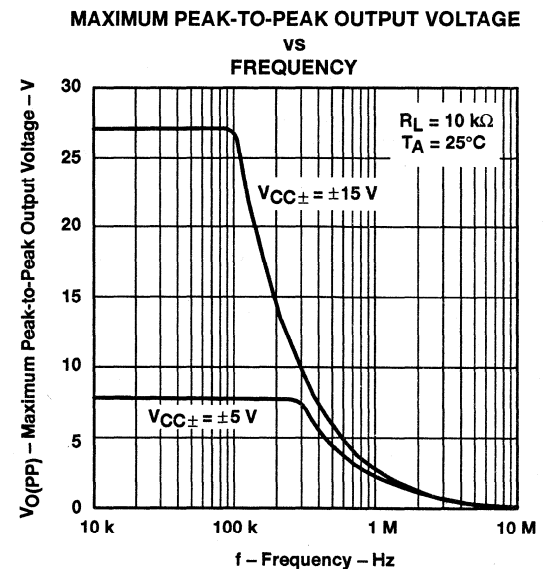


Figure 17

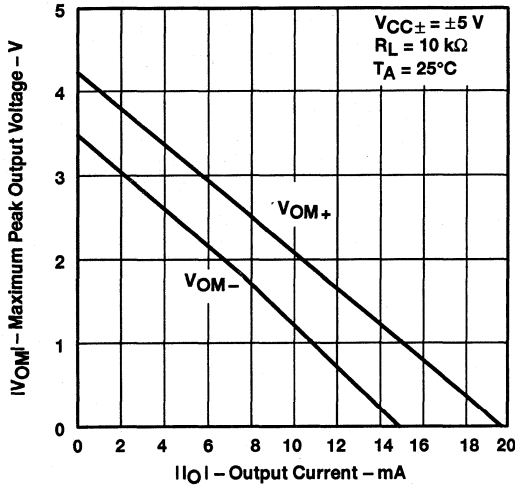
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

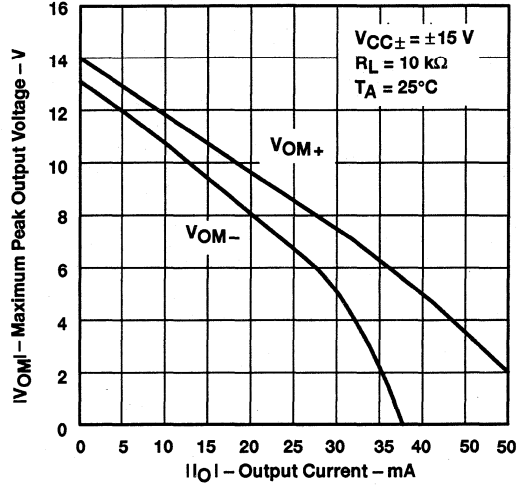
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**OUTPUT CURRENT**



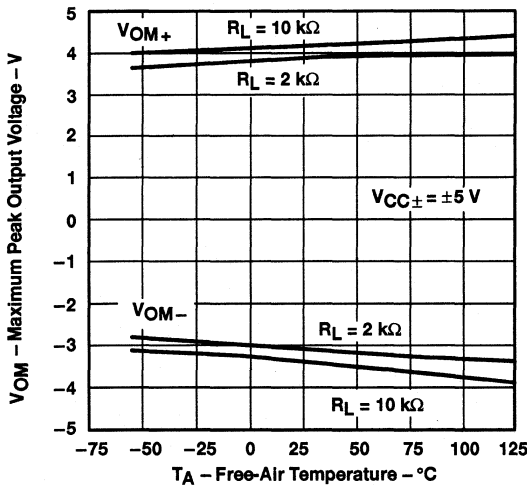
**Figure 18**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**OUTPUT CURRENT**



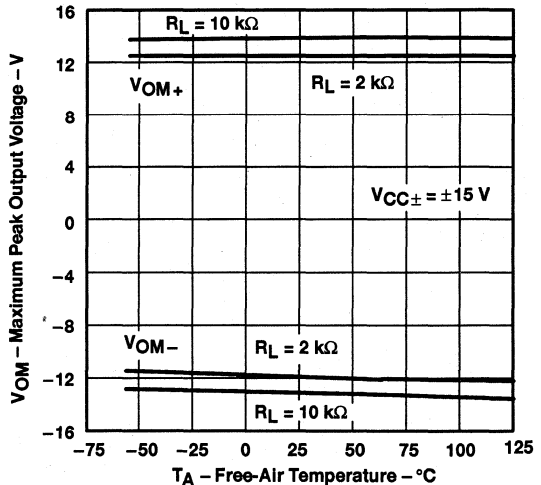
**Figure 19**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 20**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 21**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION  
 vs  
 LOAD RESISTANCE

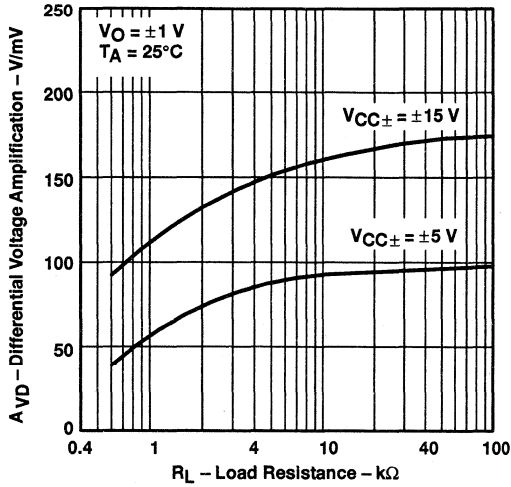


Figure 22

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

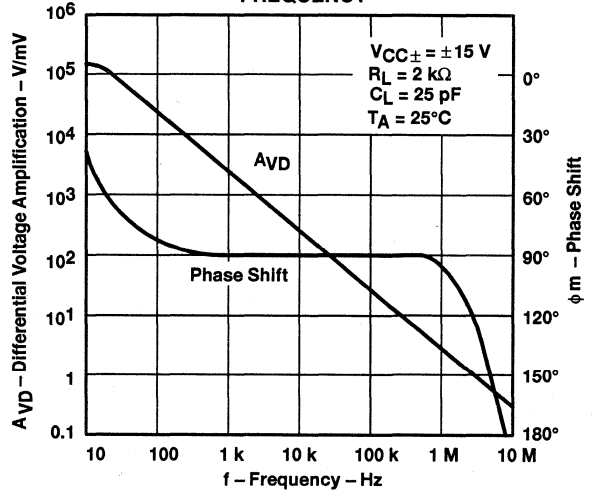


Figure 23

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

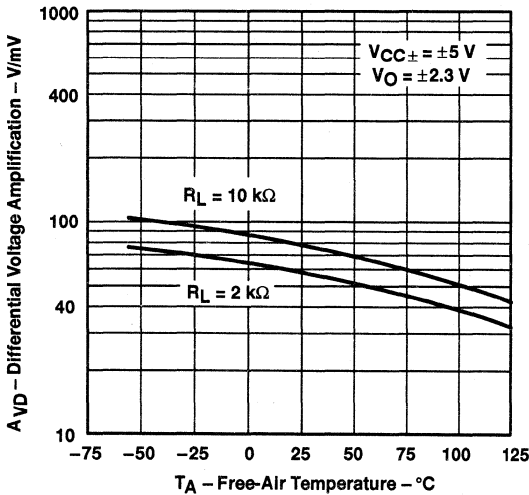


Figure 24

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

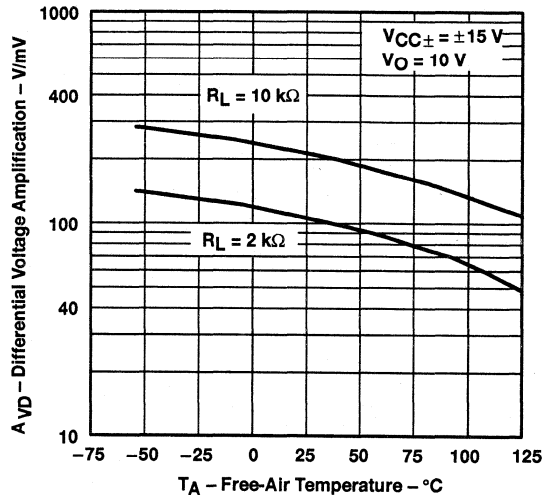


Figure 25

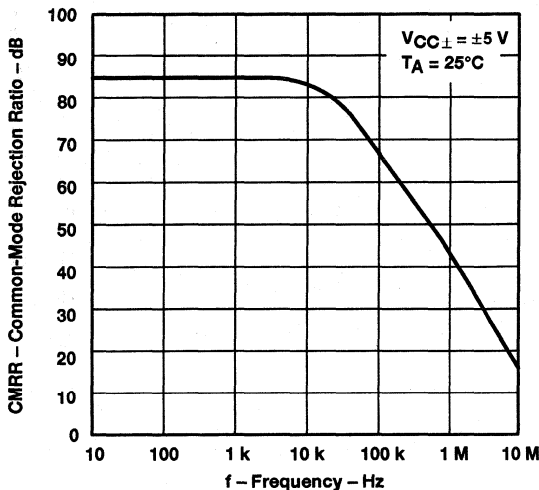
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

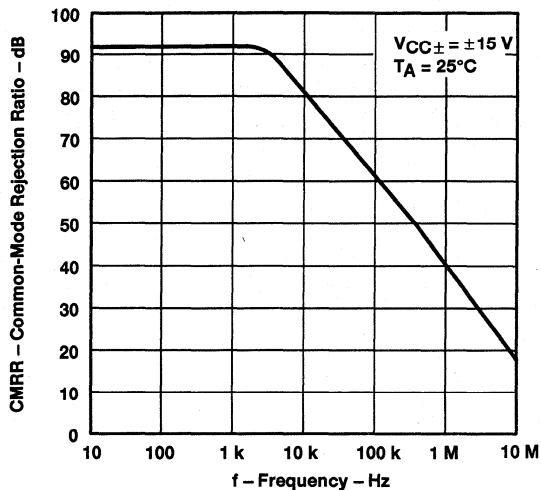
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO**  
**VS**  
**FREQUENCY**



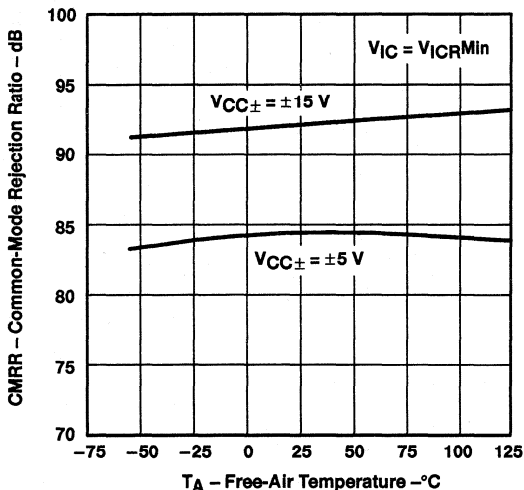
**Figure 26**

**COMMON-MODE REJECTION RATIO**  
**VS**  
**FREQUENCY**



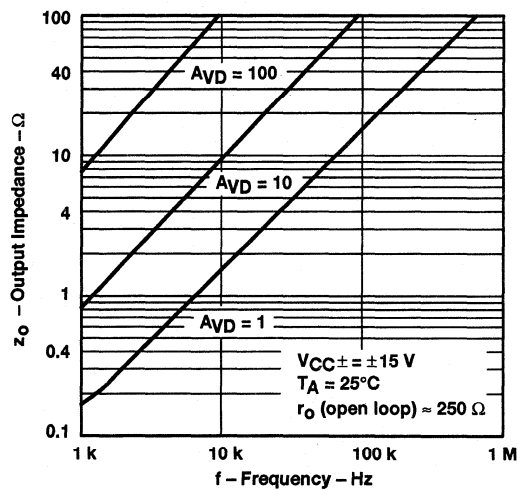
**Figure 27**

**COMMON-MODE REJECTION RATIO**  
**VS**  
**FREE-AIR TEMPERATURE**



**Figure 28**

**OUTPUT IMPEDANCE**  
**VS**  
**FREQUENCY**



**Figure 29**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

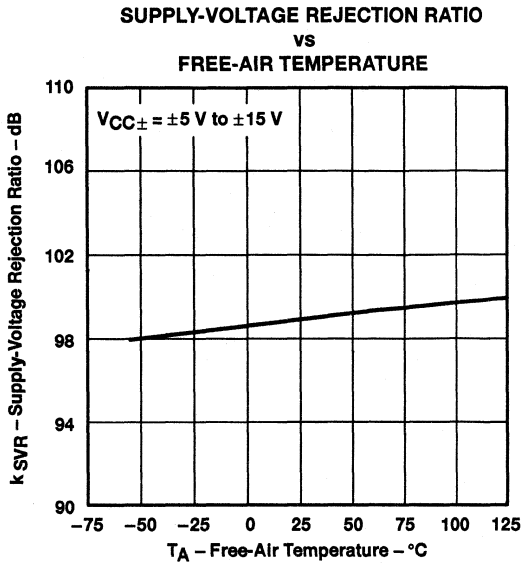


Figure 30

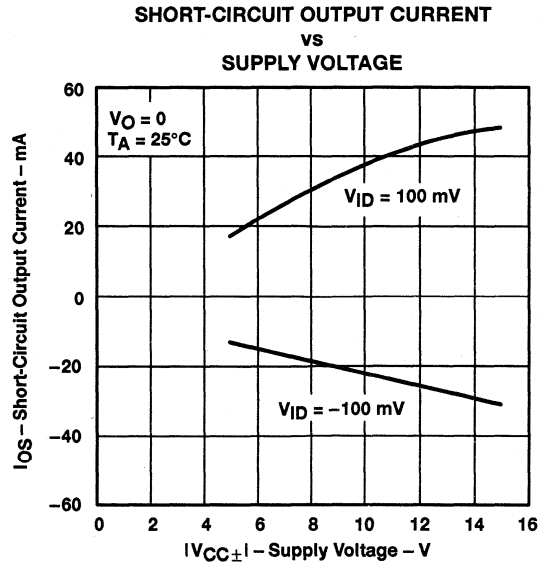


Figure 31

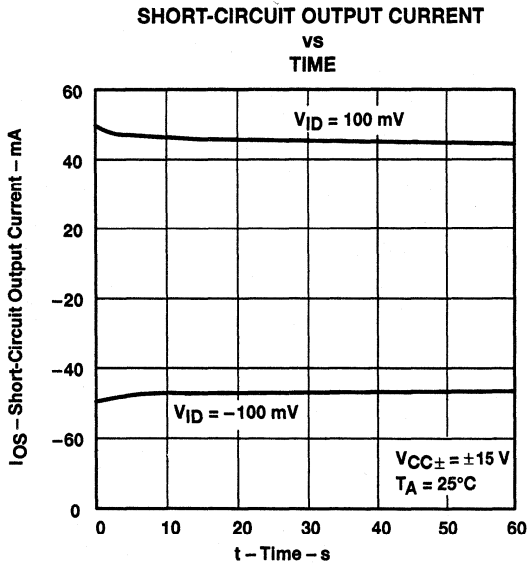


Figure 32

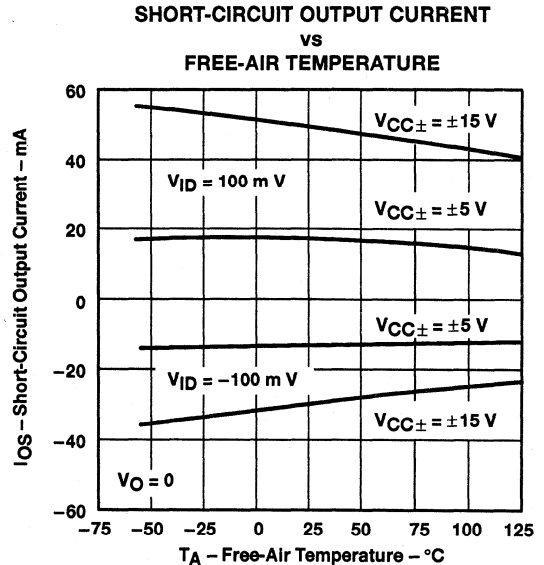


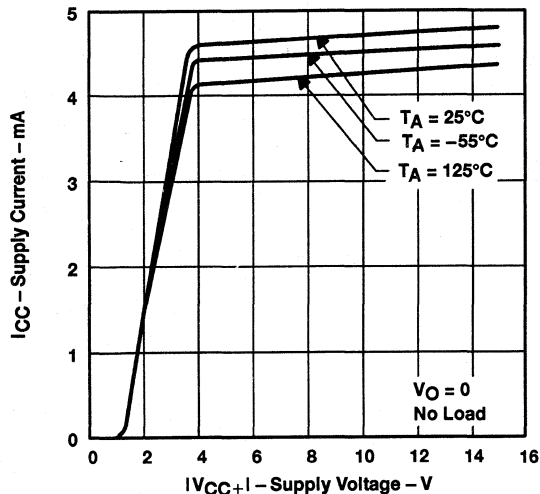
Figure 33

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS036C - JUNE 1988 - REVISED AUGUST 1994

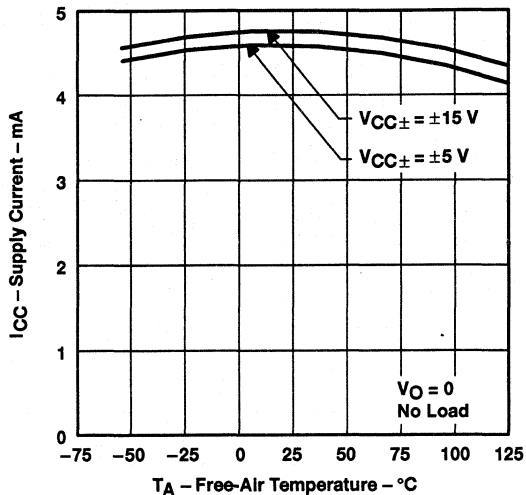
**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



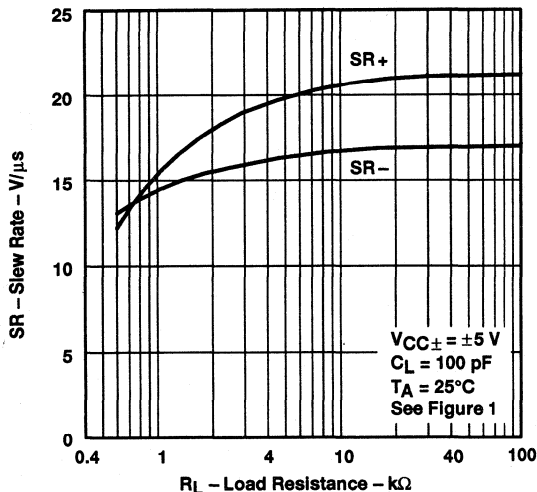
**Figure 34**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



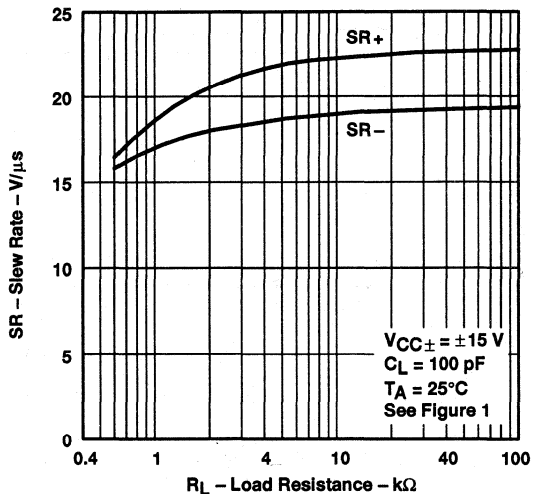
**Figure 35**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**



**Figure 36**

**SLEW RATE**  
**vs**  
**LOAD RESISTANCE**

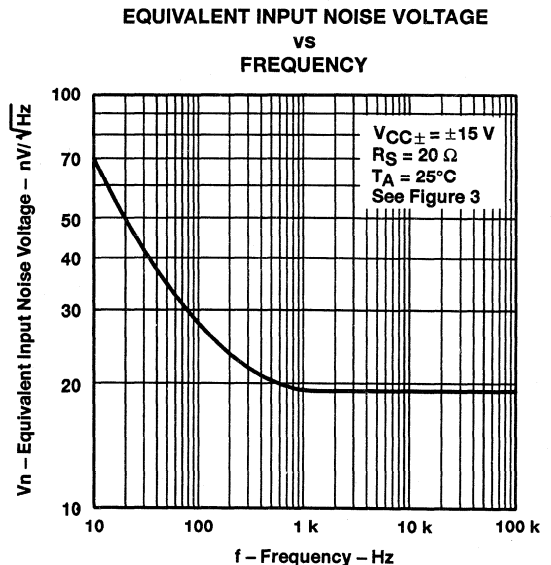
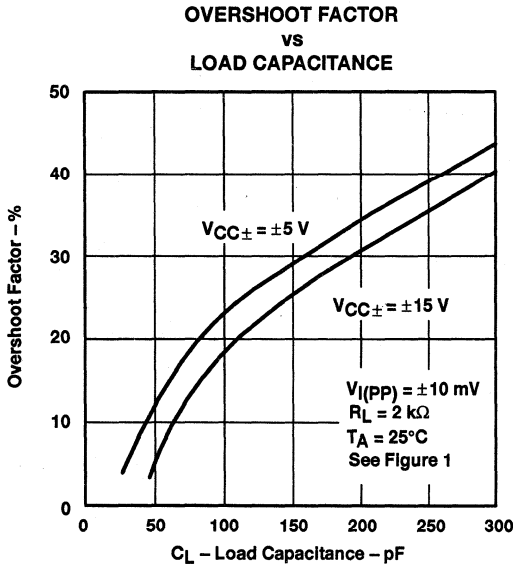
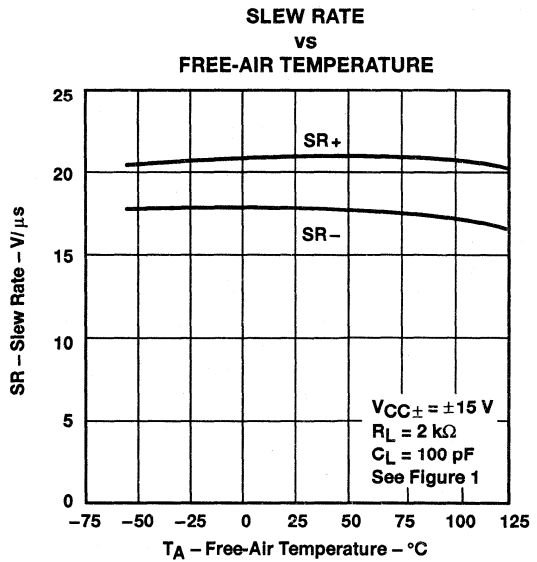
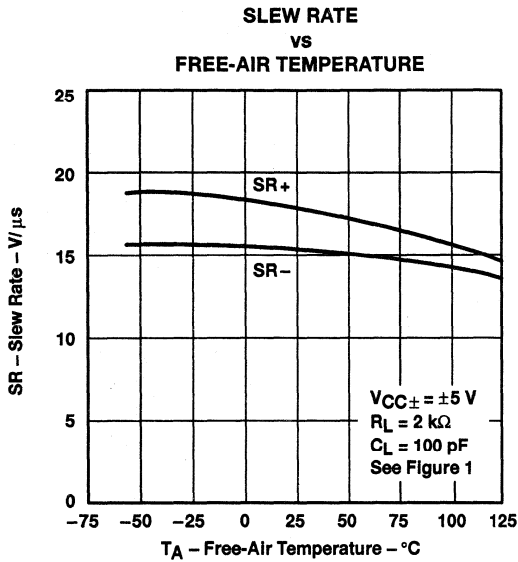


**Figure 37**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†



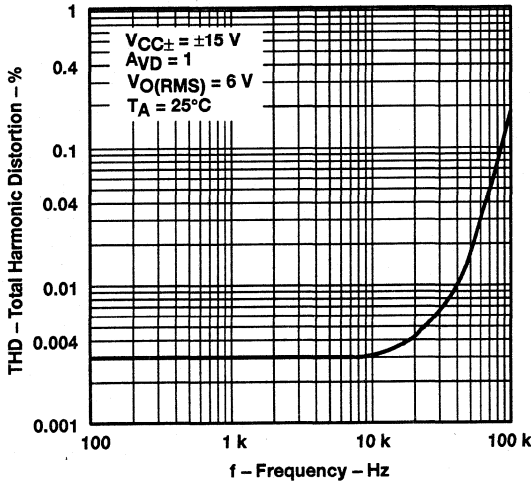
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

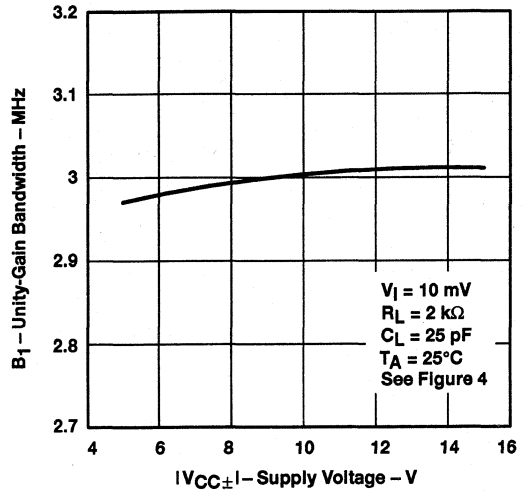
**TYPICAL CHARACTERISTICS†**

**TOTAL HARMONIC DISTORTION  
 vs  
 FREQUENCY**



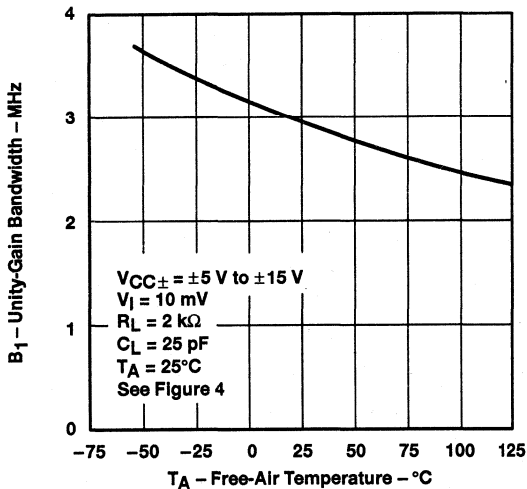
**Figure 42**

**UNITY-GAIN BANDWIDTH  
 vs  
 SUPPLY VOLTAGE**



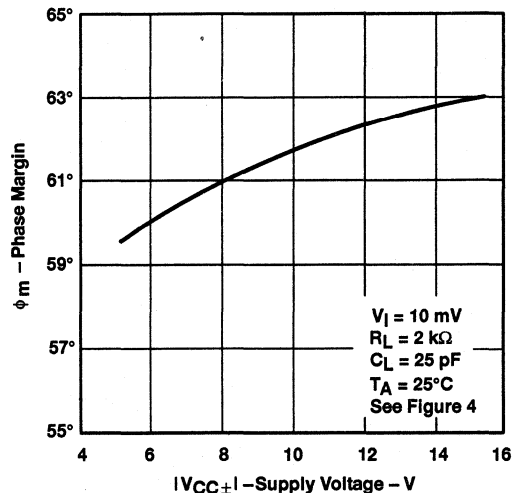
**Figure 43**

**UNITY-GAIN BANDWIDTH  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 44**

**PHASE MARGIN  
 vs  
 SUPPLY VOLTAGE**



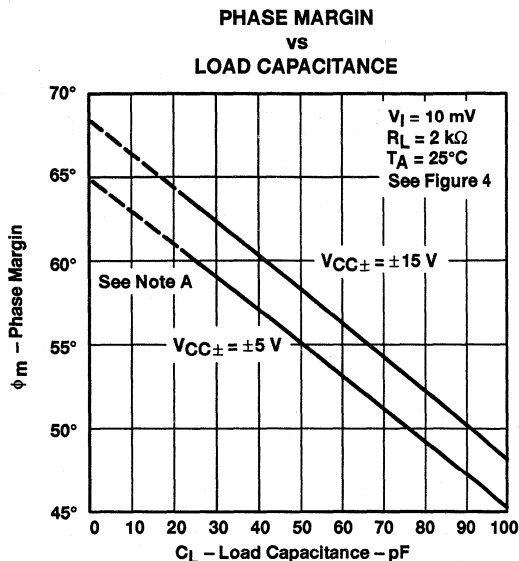
**Figure 45**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TYPICAL CHARACTERISTICS†



NOTE A: Values of phase margin below a load capacitance of 25 pF were estimated.

Figure 46

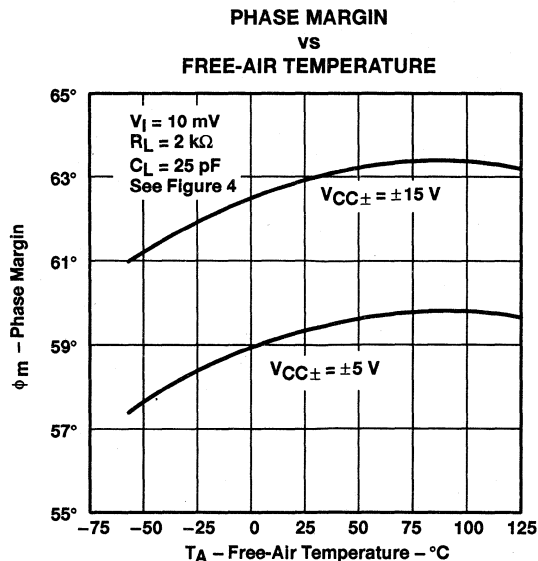


Figure 47

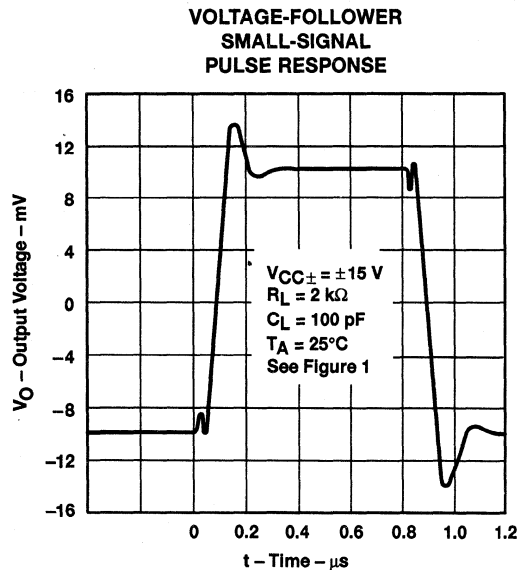


Figure 48

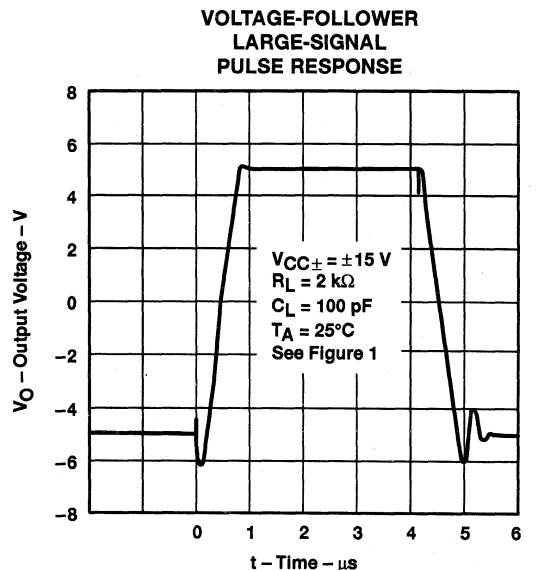


Figure 49

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

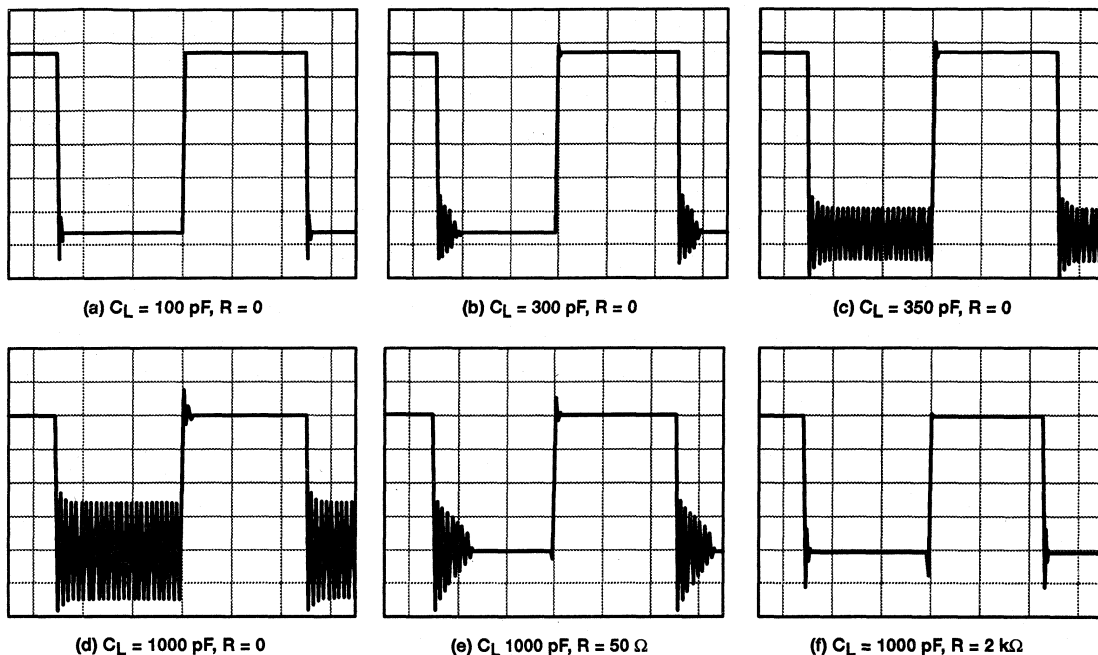
**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C – JUNE 1988 – REVISED AUGUST 1994

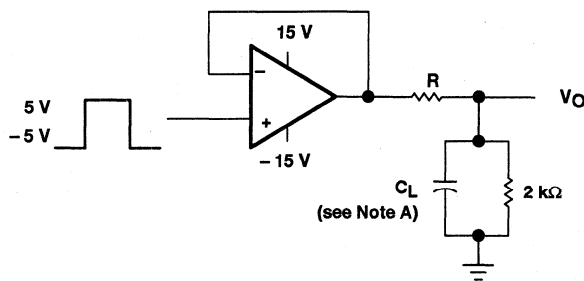
**APPLICATION INFORMATION**

**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL052 and TL052A drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).



**Figure 50. Effect of Capacitive Loads**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 51. Test Circuit for Output Characteristics**

## APPLICATION INFORMATION

### input characteristics

The TL052 and TL052A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL052 and TL052A are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

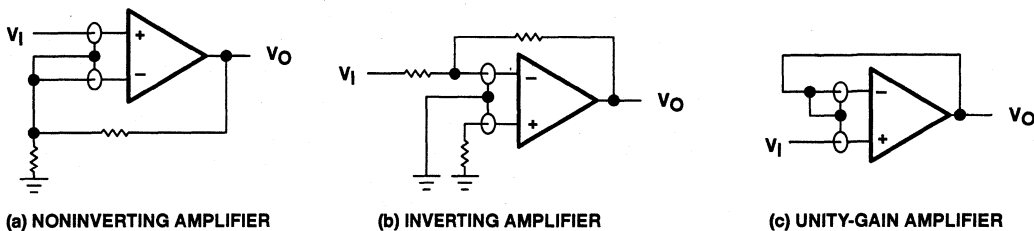


Figure 52. Use of Guard Rings

### noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL052 and TL052A result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k $\Omega$ .

**TL052, TL052A, TL052Y**  
**ENHANCED-JFET PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS036C - JUNE 1988 - REVISED AUGUST 1994

**APPLICATION INFORMATION**

**instrumentation amplifier with adjustable gain/null**

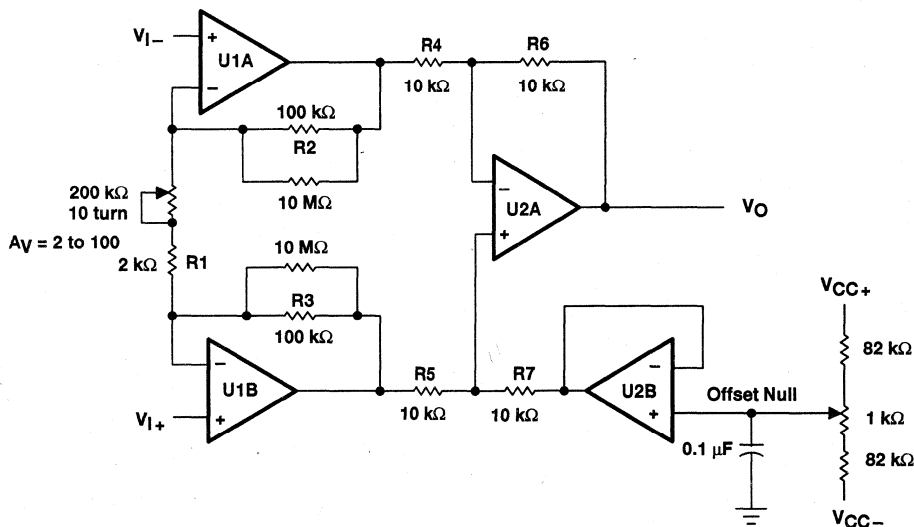
The instrumentation amplifier in Figure 53 benefits greatly from the high input impedance and stable input offset voltage of the TL052A. Amplifiers U1A, U1B, and U2A form the actual instrumentation amplifier, while U2B provides offset null. Potentiometer R1 provides gain adjust. With  $R1 = 2\text{ k}\Omega$ , the circuit gain equals 100, while with  $R1 = 200\text{ k}\Omega$ , the circuit gain equals two. The following equation shows the instrumentation amplifier gain as a function of  $R1$ :

$$A_v = 1 + \left( \frac{R2 + R3}{R1} \right)$$

Readjusting the offset null is necessary whenever the circuit gain is changed. If U2B is needed for another application, R7 can be terminated at ground. The low input offset voltage of the TL052A minimizes the dc error of the circuit. For best matching, all resistors should be one percent tolerance. The matching between R4, R5, R6, and R7 controls the CMRR of this application.

The following equation shows the output voltages when the input voltage equals zero. This dc error can be nulled by adjusting the offset null potentiometer; however, any change in offset voltage over time or temperature also creates an error. To calculate the error from changes in offset, consider the three offset components in the equation as delta offsets rather than initial offsets. The improved stability of Texas Instruments enhanced JFETs minimizes the error resulting from change in input offset voltage with time. Assuming  $V_1$  equals zero,  $V_O$  can be shown as a function of the offset voltage:

$$V_O = V_{IO2} \left[ \left( 1 + \frac{R3}{R1} \right) \left( \frac{R7}{R5 + R7} \right) \left( 1 + \frac{R6}{R4} \right) + \frac{R2}{R1} \left( \frac{R6}{R4} \right) \right] - V_{IO1} \left[ \frac{R3}{R1} \left( \frac{R7}{R5 + R7} \right) \left( 1 + \frac{R6}{R4} \right) + \frac{R6}{R4} \left( 1 + \frac{R2}{R1} \right) \right] + V_{IO3} \left( 1 + \frac{R6}{R4} \right)$$



**Figure 53. Instrumentation Amplifier**



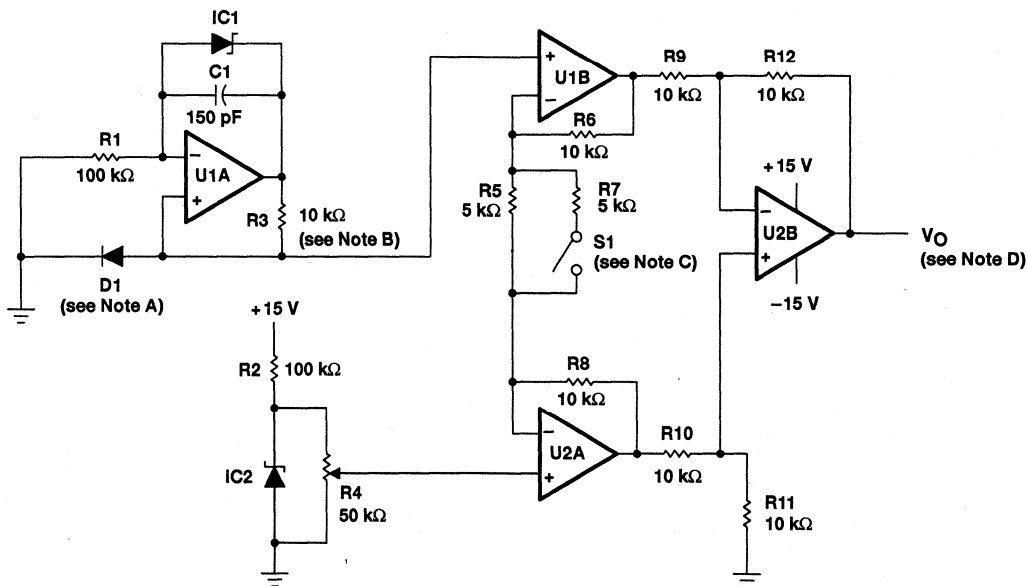
**APPLICATION INFORMATION**

**analog thermometer**

By combining a current source that does not vary over temperature with an instrumentation amplifier, a precise analog thermometer can be built (see Figure 54). Amplifier U1A and IC1 establish a constant current through the temperature-sensing diode D1. For this section of the circuit to operate correctly, the TL052 must use split supplies and R3 must be a metal-film resistor with a low temperature coefficient.

The temperature-sensitive voltage from the diode is compared to a temperature-stable voltage reference set by IC2. R4 should be adjusted to provide the correct output voltage when the diode is at a known temperature. Although this potentiometer resistance varies with temperature, the divider ratio of the potentiometer remains constant.

Amplifiers U1B, U2A, and U2B form the instrumentation amplifier that converts the difference between the diode and reference voltage to a voltage proportional to the temperature. With switch S1 closed, the amplifier gain equals 5 and the output voltage is proportional to temperature in degrees Celsius. With S1 open, the amplifier gain is 9 and the output is proportional to temperature in degrees Fahrenheit. Every time that S1 is changed, R4 must be recalibrated. By setting S1 correctly, the output voltage equals 10 mV per degree (C or F).



- NOTES: A. Temperature-sensing diode  $\approx (-2 \text{ mV}/^\circ\text{C})$   
 B. Metal-film resistor (low temperature coefficient)  
 C. Switch open for  $^\circ\text{F}$  and closed for  $^\circ\text{C}$   
 D.  $V_O \propto$  temperature;  $10 \text{ mV}/^\circ\text{C}$  or  $10 \text{ mV}/^\circ\text{F}$   
 E. U1, U2 = TL052. IC1, IC2 = LM385, LT1004, or LT1009 voltage reference

**Figure 54. Analog Thermometer**

**APPLICATION INFORMATION**

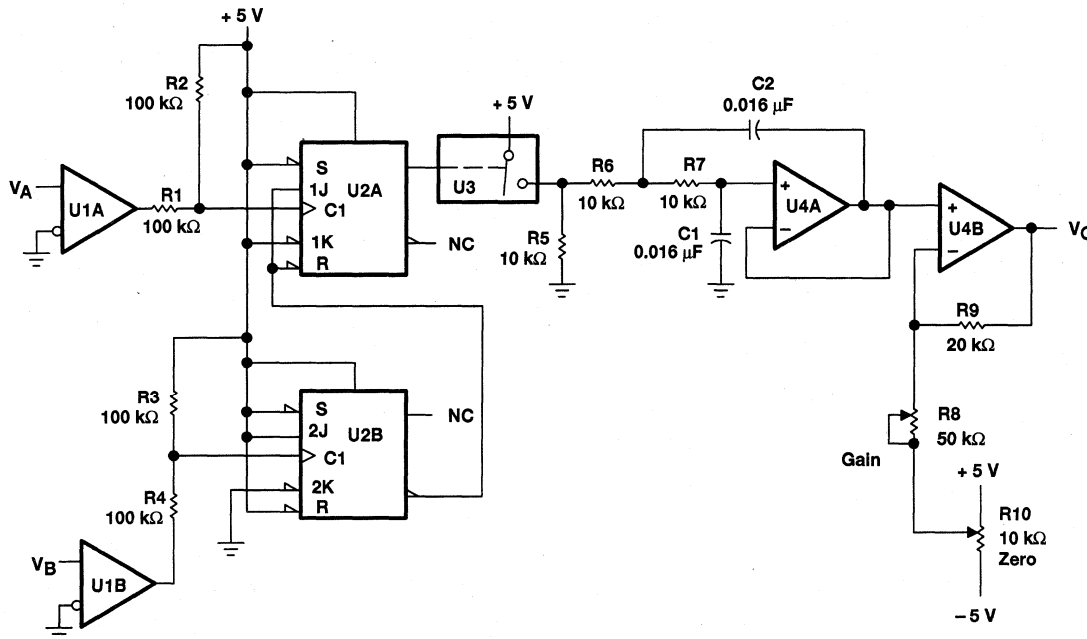
**phase meter**

The phase meter in Figure 55 produces an output voltage of 10 mV per degree of phase delay between the two input signals  $V_A$  and  $V_B$ . The reference signal  $V_A$  must be the same frequency as  $V_B$ . The TLC3702 comparators (U1) convert these two input sine waves into  $\pm 5$ -V square waves. Then R1 and R4 provide level shifting prior to the SN74HC109 dual J-K flip flops.

Flip-flop U2B is connected as a toggle flip-flop and generates a square wave at half the frequency of  $V_B$ . Flip-flop U2A also produces a square wave at half the input frequency. The pulse duration of U2A varies from zero to half the period, where zero corresponds to zero phase delay between  $V_A$  and  $V_B$  and half the period corresponds to  $V_B$  lagging  $V_A$  by 360 degrees.

The output pulse from U2A causes the TLC4066 (U3) switch to charge the TL052 (U4) integrator capacitors C1 and C2. As the phase delay approaches 360 degrees, the output of U4A approximates a square wave and U2A has an output of almost 2.5 V. U4B acts as a noninverting amplifier with a gain of 1.44 in order to scale the 0- to 2.5-V integrator output to a 0- to 3.6-V output range.

R8 and R10 provide output gain and zero-level calibration. This circuit operates over a 100-Hz to 10-kHz frequency range.



- NOTES: U1 = TLC3702;  $V_{CC} \pm = \pm 5$  V  
 U2 = SN74HC109  
 U3 = TLC4066  
 U4, U5 = TL051;  $V_{CC} \pm = \pm 5$  V

**Figure 55. Phase Meter**

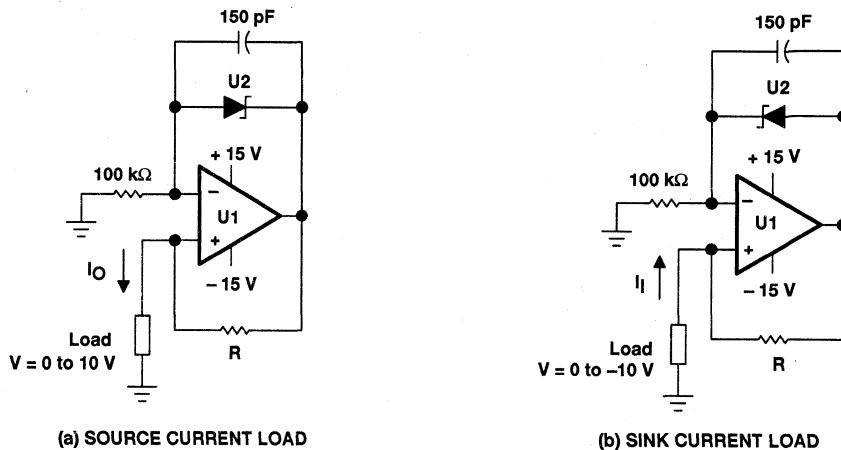


**APPLICATION INFORMATION**

**precision constant-current source over temperature**

A precision current source benefits from the high input impedance and stability of Texas Instruments enhanced-JFET process. A low-current shunt regulator maintains 2.5 V between the inverting input and the output of the TL052. The negative feedback then forces 2.5 V across the current setting resistor R; therefore, the current to the load is simply 2.5 V divided by R.

Possible choices for the shunt regulator include the LT1004, LT1009, and LM385. If the regulator's cathode connects to the operational amplifier output, this circuit sources load current. Similarly, if the cathode connects to the inverting input, the circuit sinks current from the load. To minimize output current change with temperature, R should be a metal film resistor with a low temperature coefficient. Also, this circuit must be operated with split-voltage supplies.



NOTES: U1 = 1/2 TL052  
 U2 = LM385, LT1004, or LT1009 voltage reference  
 $I = \frac{2.5 \text{ V}}{R}$ , R = Low temperature coefficient metal film resistor

**Figure 56. Precision Constant-Current Source**





# TL054, TL054A ENHANCED-JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS038C – JUNE 1988 – REVISED AUGUST 1994

- **Maximum Offset Voltage . . . 1.5 mV (TL054A)**
- **High Slew Rate . . . 15.9 V/μs Typ at 25°C**
- **Low Total Harmonic Distortion 0.003% Typ at  $R_L = 2\text{ k}\Omega$**
- **Low Noise Voltage . . . 21 nV/√Hz Typ at  $f = 1\text{ kHz}$**
- **Low Input Bias Currents . . . 30 pA Typ**
- **Monolithic Construction**

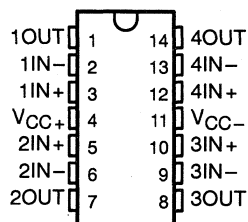
## description

The TL054 and TL054A quad operational amplifiers incorporate well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. These devices offer the significant advantages of Texas Instruments new enhanced-JFET process. This process affords not only low initial offset voltage due to the on-chip zener trim capability but also stable offset voltage over time and temperature. In comparison, traditional JFET processes are plagued by significant offset-voltage drift.

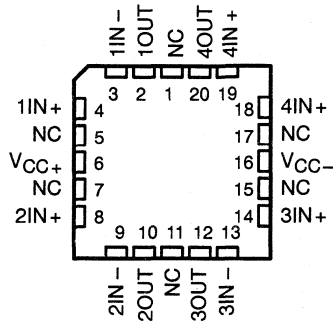
This new enhanced process still maintains the traditional JFET advantages of fast slew rates and low input bias and offset currents. These advantages coupled with low noise and low harmonic distortion make the TL054 well suited for new state-of-the-art designs as well as existing design upgrades. The TL054 has been designed to be functionally compatible, as well as pin compatible, with the TL074 and TL084.

Two offset-voltage grades are available: TL054 (4 mV MAX) and TL054A (1.5 mV MAX).

**D, J, OR N PACKAGE  
(TOP VIEW)**



**FK PACKAGE  
(TOP VIEW)**



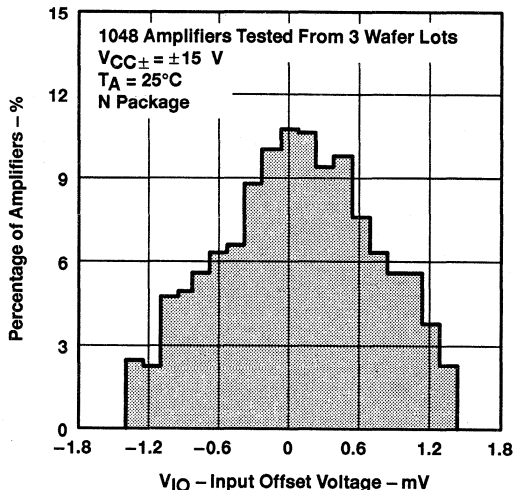
NC – No internal connection

**AVAILABLE OPTIONS**

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)
0°C to 70°C	1.5 mV 4 mV	TL054ACD TL054CD	—	—	TL054ACN TL054CN
-40°C to 85°C	1.5 mV 4 mV	TL054AID TL054ID	—	—	TL054AIN TL054IN
-55°C to 125°C	1.5 mV 4 mV	TL054AMD TL054MD	TL054AMFK TL054MFK	TL054AMJ TL054MJ	TL054AMN TL054MN

The D packages are available taped and reeled. Add R suffix to device type (e.g., TL054CDR).

**DISTRIBUTION OF TL054A  
INPUT OFFSET VOLTAGE**



# TL054, TL054A ENHANCED-JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

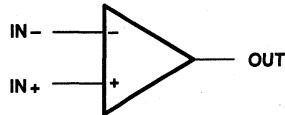
SLOS038C - JUNE 1988 - REVISED AUGUST 1994

## description (continued)

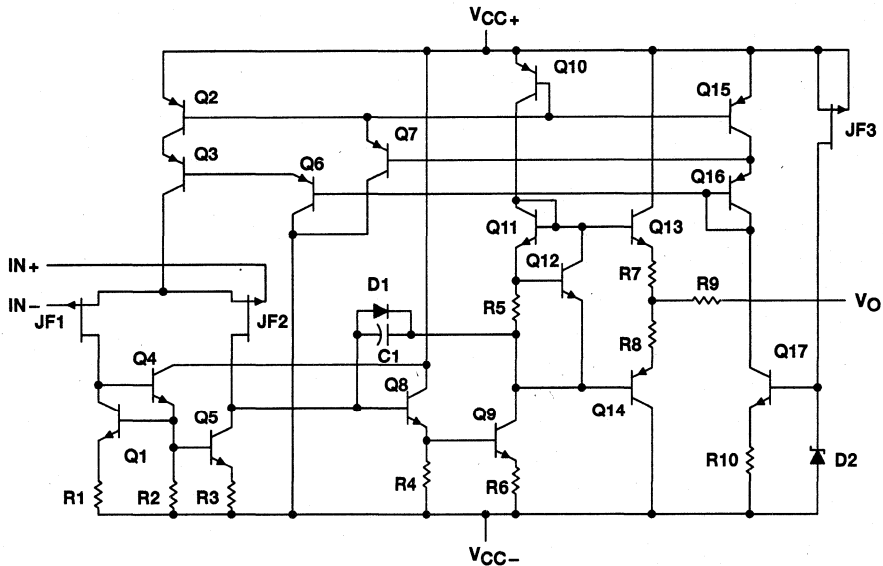
A variety of available packaging options includes small-outline and chip-carrier versions for high-density system applications.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

## symbol (each amplifier)



## equivalent schematic (each amplifier)



**TL054, TL054A**  
**ENHANCED-JFET PRECISION**  
**QUAD OPERATIONAL AMPLIFIERS**  
SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage (see Note 2)	$\pm 30$ V
Input voltage range, $V_I$ (any input, see Notes 1 and 3)	$\pm 15$ V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$ (each output)	$\pm 80$ mA
Total current into $V_{CC+}$	160 mA
Total current out of $V_{CC-}$	160 mA
Duration of short-circuit current at (or below) 25°C (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
4. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	230 mW

**recommended operating conditions**

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{CC\pm}$		$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	$\pm 5$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 5$ V	-1	4	-1	4	-1	4	V
	$V_{CC\pm} = \pm 15$ V	-11	11	-11	11	-11	11	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C

# TL054, TL054A ENHANCED-JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS038C - JUNE 1988 - REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL054C, TL054AC						UNIT	
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub>	Input offset voltage	TL054C	25°C	0.64	5.5	0.56	4	mV		
			Full range	7.7		6.2				
			TL054AC	25°C	0.57	3.5	0.5		1.5	
				Full range	5.7		3.7			
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL054C	25°C to 70°C	25		23		μV/°C	
				TL054AC	25°C to 70°C	24		23		
			Input offset voltage long-term drift (see Note 5)		25°C	0.04		0.04		μV/mo
			I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	4	100	5	100
70°C	0.02	1				0.025	1	nA		
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	25°C	20	200	30	200	pA		
			70°C	0.15	4	0.2	4	nA		
V <sub>ICR</sub>	Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V		
			Full range	-1 to 4		-11 to 11				
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3	4.2	13	13.9	V		
			Full range	3		13				
		R <sub>L</sub> = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
			Full range	2.5		11.5				
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	V		
			Full range	-2.5		-12				
		R <sub>L</sub> = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
			Full range	-2.3		-11				
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 6	25°C	25	72	50	133	V/mV		
			0°C	30	88	60	173			
			70°C	20	57	30	85			
r <sub>i</sub>	Input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub>	Input capacitance		25°C	10		12		pF		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	65	84	75	92	dB		
			0°C	65	84	75	92			
			70°C	65	84	75	93			
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75	99	75	99	dB		
			0°C	75	99	75	99			
			70°C	75	99	75	99			
I <sub>CC</sub>	Supply current (four amplifiers)	V <sub>O</sub> = 0, No load	25°C	8.1	11.2	8.4	11.2	mA		
			0°C	8.2	12.8	8.5	12.8			
			70°C	7.9	11.2	8.2	11.2			
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C	120		120		dB		

† Full range is 0°C to 70°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V, at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.A



# TL054, TL054A ENHANCED-JFET PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS038C – JUNE 1988 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL054C, TL054A						UNIT		
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V					
			MIN	TYP	MAX	MIN	TYP	MAX			
SR <sub>+</sub>	Positive slew rate at unity gain		25°C	15.4			10 17.8			V/μs	
			0°C	15.7			8 17.9				
SR <sub>-</sub>	Negative slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 7	70°C	14.4			8 17.5				
			25°C	13.9			10 15.9				
			0°C	14.3			8 16.1				
			70°C	13.3			8 15.5				
t <sub>r</sub>	Rise time		25°C	55			56				ns
			0°C	54			55				
			70°C	63			63				
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57				
			0°C	54			56				
			70°C	62			64				
			25°C	24%			19%				
Overshoot factor			0°C	24%			19%				
			70°C	24%			19%				
			70°C	24%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 9)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			75			nV/√Hz	
			f = 1 kHz	25°C			21 45				
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C			4 4			μV	
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01			pA/√Hz	
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 8	25°C	0.003%			0.003%				
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	2.7			2.7			MHz	
			0°C	3			3				
			70°C	2.4			2.4				
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	61°			64°				
			0°C	60°			64°				
			70°C	61°			63°				

† Full range is 0°C to 70°C.

NOTES: 7. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

8. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.

9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

**TL054, TL054A**  
**ENHANCED-JFET PRECISION**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL054I, TL054AI						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	25°C	TL054I	0.64	5.5	0.56	4	mV	
			Full range		8.8		7.3		
			TL054AI	0.57	3.5	0.5	1.5		
			Full range		6.8		4.8		
α <sub>VIO</sub>	Temperature coefficient of input offset voltage	25°C to 85°C	TL054I	25		24		μV/°C	
			TL054AI	25		23			
	Input offset voltage long-term drift (see Note 5)	25°C	0.04		0.04		μV/mo		
I <sub>IO</sub>	Input offset current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	4	100	5	100	pA	
			85°C	0.06	10	0.07	10	nA	
I <sub>B</sub>	Input bias current	25°C	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, See Figure 5	20	200	30	200	pA	
			85°C	0.6	20	0.7	20	nA	
V <sub>ICR</sub>	Common-mode input voltage range	25°C		-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	V	
			Full range	-1 to 4		-11 to 11			
V <sub>OM+</sub>	Maximum positive peak output voltage swing	25°C	R <sub>L</sub> = 10 kΩ	3	4.2	13	13.9	V	
			Full range	3		13			
		Full range	R <sub>L</sub> = 2 kΩ	2.5	3.8	11.5	12.7		
				2.5		11.5			
V <sub>OM-</sub>	Maximum negative peak output voltage swing	25°C	R <sub>L</sub> = 10 kΩ	-2.5	-3.5	-12	-13.2	V	
			Full range	-2.5		-12			
		Full range	R <sub>L</sub> = 2 kΩ	-2.3	-3.2	-11	-12		
				-2.3		-11			
A <sub>VD</sub>	Large-signal differential voltage amplification	25°C	R <sub>L</sub> = 2 kΩ, See Note 6	25	72	50	133	V/mV	
			-40°C	30	101	60	212		
			85°C	20	50	30	70		
r <sub>i</sub>	Input resistance	25°C	10 <sup>12</sup>		10 <sup>12</sup>		Ω		
c <sub>i</sub>	Input capacitance	25°C	10		12		pF		
CMRR	Common-mode rejection ratio	25°C	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	65	84	75	92	dB	
			-40°C	65	83	75	92		
			85°C	65	84	75	93		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	25°C	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	75	99	75	99	dB	
			-40°C	75	98	75	99		
			85°C	75	99	75	99		
I <sub>CC</sub>	Supply current (four amplifiers)	25°C	V <sub>O</sub> = 0, No load	8.1	11.2	8.4	11.2	mA	
			-40°C	7.9	12.8	8.2	12.8		
			85°C	7.6	11.2	7.9	11.2		
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	25°C	120		120		dB		

† Full range is -40°C to 85°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V, at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.



**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL054I, TL054AI						UNIT		
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V					
			MIN	TYP	MAX	MIN	TYP	MAX			
SR+	Positive slew rate at unity gain		25°C	15.4			10 17.8			V/μs	
			-40°C	16.4			8 18				
SR-	Negative slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 7	85°C	14			8 17.3				
			25°C	13.9			10 15.9				
			-40°C	14.7			8 16.1				
			85°C	13			8 15.3				
t <sub>r</sub>	Rise time		25°C	55			56				ns
			-40°C	52			53				
			85°C	64			65				
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57				
			-40°C	51			53				
			85°C	64			65				
Overshoot factor			25°C	24%			19%				
			-40°C	24%			19%				
			85°C	24%			19%				
V <sub>n</sub>	Equivalent input noise voltage (see Note 9)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C 75			75			nV/√Hz	
			f = 1 kHz	25°C 21			21 45				
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C 4			4			μV	
I <sub>n</sub>	Equivalent input noise current		f = 1 kHz	25°C 0.01			0.01			pA/√Hz	
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz,	R <sub>L</sub> = 2 kΩ, See Note 8	25°C 0.003%			0.003%				
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C 2.7			2.7			MHz	
				-40°C 3.3			3.3				
				85°C 2.3			2.4				
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF,	R <sub>L</sub> = 2 kΩ, See Figure 4	25°C 61°			64°				
				-40°C 59°			62°				
				85°C 61°			64°				

† Full range is -40°C to 85°C.

- NOTES: 7. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.  
 8. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.  
 9. This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

**TL054, TL054A**  
**ENHANCED-JFET PRECISION**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS		T <sub>A</sub> †	TL054M, TL054AM						UNIT		
				V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V					
				MIN	TYP	MAX	MIN	TYP	MAX			
V <sub>IO</sub>	Input offset voltage	TL054M	25°C	0.64		5.5		0.56		4		
			Full range			10.5				9		
			TL054AM	25°C	0.57		3.5		0.5		1.5	
				Full range			8.5				6.5	
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	TL054M	25°C to 85°C		21		20		μV/°C		
			TL054AM	25°C to 85°C		21		20				
	Input offset voltage long-term drift (see Note 5)		25°C	0.04				0.04		μV/mo		
I <sub>IO</sub>	Input offset current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	4		100		5		100	
				125°C	1		20		2		20	
I <sub>IB</sub>	Input bias current	V <sub>O</sub> = 0, See Figure 5	V <sub>IC</sub> = 0,	25°C	20		200		30		200	
				125°C	10		50		20		50	
V <sub>ICR</sub>	Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6			V		
			Full range	-1 to 4		-11 to 11						
V <sub>OM+</sub>	Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	3		4.2		13		13.9		
			Full range	3				13				
		R <sub>L</sub> = 2 kΩ	25°C	2.5		3.8		11.5		12.7		
			Full range	2.5				11.5				
V <sub>OM-</sub>	Maximum negative peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	-2.5		-3.5		-12		-13.2		
			Full range	-2.5				-12				
		R <sub>L</sub> = 2 kΩ	25°C	-2.3		-3.2		-11		-12		
			Full range	-2.3				-11				
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 2 kΩ, See Note 6	25°C	25		72		50		133		
			-55°C	30		99		60		209		
			125°C	10		35		15		35		
r <sub>i</sub>	Input resistance		25°C	10 <sup>12</sup>				10 <sup>12</sup>		Ω		
c <sub>i</sub>	Input capacitance		25°C	10				12		pF		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	65		84		75		92		
			-55°C	65		83		75		92		
			125°C	65		84		75		93		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC±</sub> = ±5 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	75		99		75		99		
			-40°C	75		98		75		98		
			85°C	75		100		75		100		
I <sub>CC</sub>	Supply current (four amplifiers)	V <sub>O</sub> = 0, No load	25°C	8.1		11.2		8.4		11.2		
			-55°C	7.8		12.8		8.1		12.8		
			125°C	7.1		11.2		7.5		11.2		
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C	120				120		dB		

† Full range is -55°C to 125°C.

NOTES: 5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. For V<sub>CC±</sub> = ±5 V, V<sub>O</sub> = ±2.3 V, at V<sub>CC±</sub> = ±15 V, V<sub>O</sub> = ±10 V.





**TL054, TL054A**  
**ENHANCED-JFET PRECISION**  
**QUAD OPERATIONAL AMPLIFIERS**  
SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL054M, TL054AM						UNIT
			V <sub>CC±</sub> = ±5 V			V <sub>CC±</sub> = ±15 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+	Positive slew rate at unity gain		25°C	15.4			10 17.8		V/μs
			-55°C	16.7			18.3		
			125°C	12.9			16.7		
SR-	Negative slew rate at unity gain	R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1 and Note 7	25°C	13.9			10 15.9		
			-55°C	14.7			16.3		
			125°C	12.2			14.5		
t <sub>r</sub>	Rise time		25°C	55			56		ns
			-55°C	51			52		
			125°C	68			68		
t <sub>f</sub>	Fall time	V <sub>I(PP)</sub> = ±10 mV, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figures 1 and 2	25°C	55			57		
			-55°C	51			52		
			125°C	68			69		
	Overshoot factor		25°C	24%			19%		
			-55°C	25%			19%		
			125°C	25%			19%		
V <sub>n</sub>	Equivalent input noise voltage (see Note 9)	R <sub>S</sub> = 20 Ω, See Figure 3	f = 10 Hz	25°C			75		nV/√Hz
			f = 1 kHz	25°C			21 45		
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage		f = 10 Hz to 10 kHz	25°C			4 4		μV
I <sub>n</sub>	Equivalent input noise current	f = 1 kHz	25°C	0.01			0.01		pA/√Hz
THD	Total harmonic distortion	R <sub>S</sub> = 1 kΩ, f = 1 kHz, R <sub>L</sub> = 2 kΩ, See Note 8	25°C	0.003%			0.003%		
B <sub>1</sub>	Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	2.7			2.7		MHz
			-55°C	3.4			3.4		
			125°C	2.1			2.1		
φ <sub>m</sub>	Phase margin at unity gain	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 25 pF, R <sub>L</sub> = 2 kΩ, See Figure 4	25°C	61°			64°		
			-55°C	58°			62°		
			125°C	60°			64°		

† Full range is -55°C to 125°C.

NOTES: 7. For V<sub>CC±</sub> = ±5 V, V<sub>I(PP)</sub> = ±1 V; for V<sub>CC±</sub> = ±15 V, V<sub>I(PP)</sub> = ±5 V.

8. For V<sub>CC±</sub> = ±5 V, V<sub>O(rms)</sub> = 1 V; for V<sub>CC±</sub> = ±15 V, V<sub>O(rms)</sub> = 6 V.

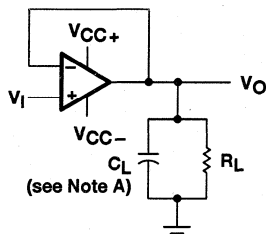
# TL054, TL054A

## ENHANCED-JFET PRECISION

### QUAD OPERATIONAL AMPLIFIERS

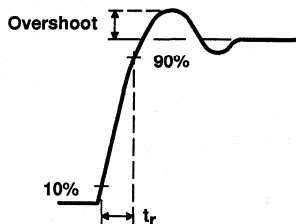
SLOS038C – JUNE 1988 – REVISED AUGUST 1994

#### PARAMETER MEASUREMENT INFORMATION

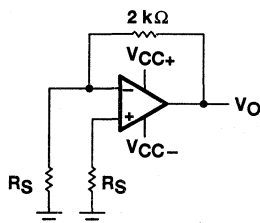


NOTE A:  $C_L$  includes fixture capacitance.

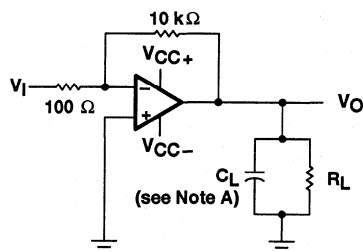
**Figure 1. Slew Rate, Rise/Fall Time, and Overshoot Test Circuit**



**Figure 2 Rise Time and Overshoot Waveform**



**Figure 3. Noise-Voltage Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit**

#### typical values

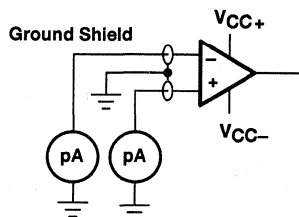
Typical values are presented in this data sheet represent the median (50% point) of device parametric performance.

#### input bias and offset current

At the picoamp-bias-current level typical of the TL054 and TL054A, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

#### noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample-tested at  $f = 1$  kHz. Texas Instruments also has additional noise testing capability to meet specific application requirements. Please contact the factory for details.



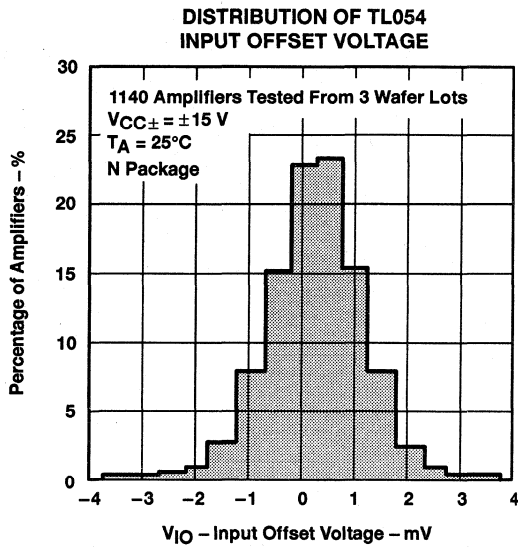
**Figure 5. Input-Bias and Offset-Current Test Circuit**

**TYPICAL CHARACTERISTICS**

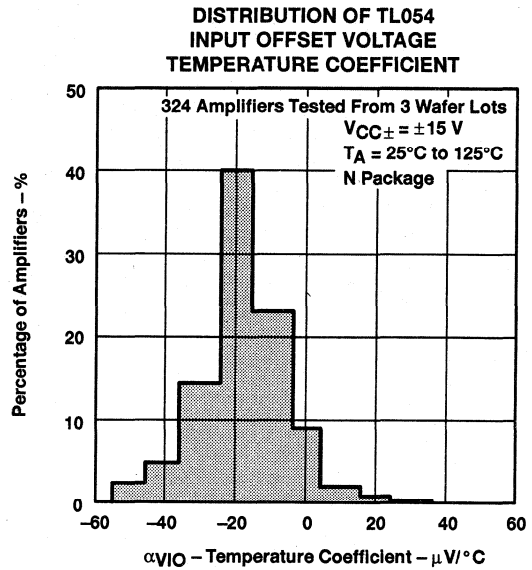
**Table of Graphs**

			FIGURE
$V_{IO}$	Input offset voltage	Distribution	6
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	Distribution	7
$I_{IO}$	Input offset current	vs Free-air temperature	8
$I_{IB}$	Input bias current	vs Temperature	8
		vs Common-mode input voltage	9
$V_I$	Input voltage range	vs Supply voltage	10
		vs Free-air temperature	11
$V_O$	Output voltage	vs Differential input voltage	12, 13
$V_{OM}$	Maximum peak output voltage swing	vs Supply voltage	14
		vs Frequency	15, 16, 17
		vs Output current	18, 19
		vs Free-air temperature	20, 21
$A_{VD}$	Differential voltage amplification	vs Resistance load	22
		vs Frequency	23
		vs Free-air temperature	24, 25
$z_o$	Output impedance	vs Frequency	29
CMRR	Common-mode rejection ratio	vs Frequency	26, 27
		vs Free-air temperature	28
$k_{SVR}$	Supply-voltage rejection ratio	vs free air temperature	30
$I_{OS}$	Short-circuit output current	vs Supply voltage	31
		vs Time	32
		vs Free-air temperature	33
$I_{CC}$	Supply current	vs Supply voltage	34
		vs free-air temperature	35
SR	Slew rate	Resistance load	36, 37
		vs Free-air temperature	38, 39
	Overshoot factor	vs Capacitance load	40
$V_n$	Equivalent input noise voltage	vs Frequency	41
THD	Total harmonic distortion	vs Frequency	42
$B_1$	Unity-gain bandwidth	vs Supply voltage	43
		vs Free-air temperature	44
$\phi_m$	Phase margin	vs Supply voltage	45
		vs Capacitance load	46
		vs Free-air temperature	47
	Phase shift	vs Frequency	23
	Pulse response	Small signal	48
		Large signal	49

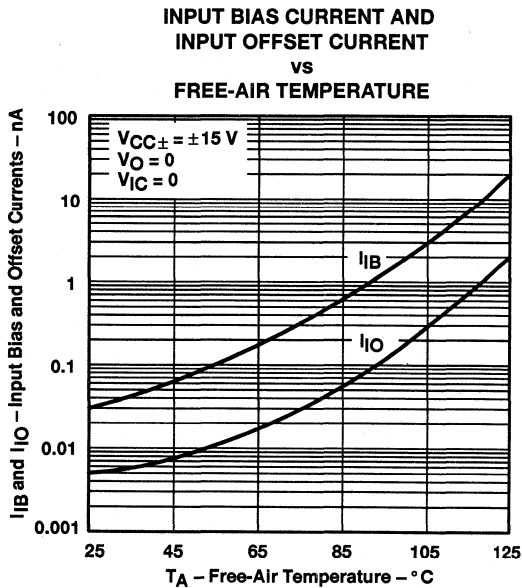
**TYPICAL CHARACTERISTICS†**



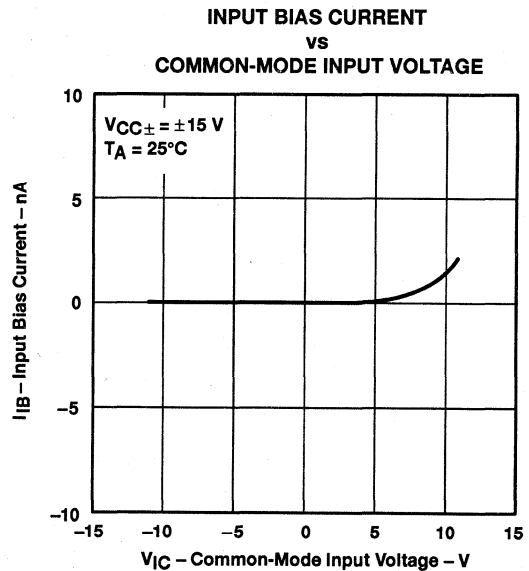
**Figure 6**



**Figure 7**



**Figure 8**



**Figure 9**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

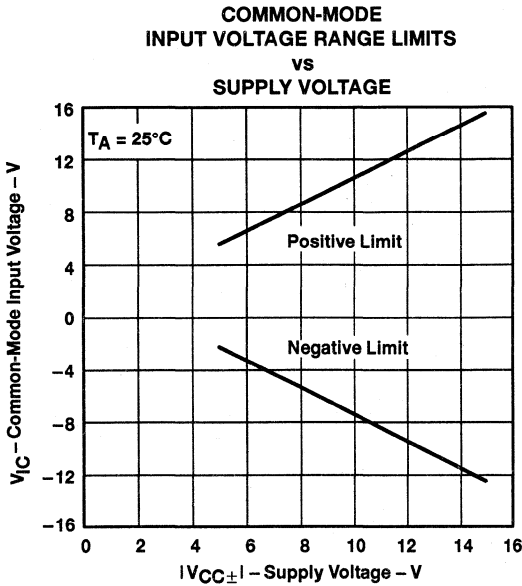


Figure 10

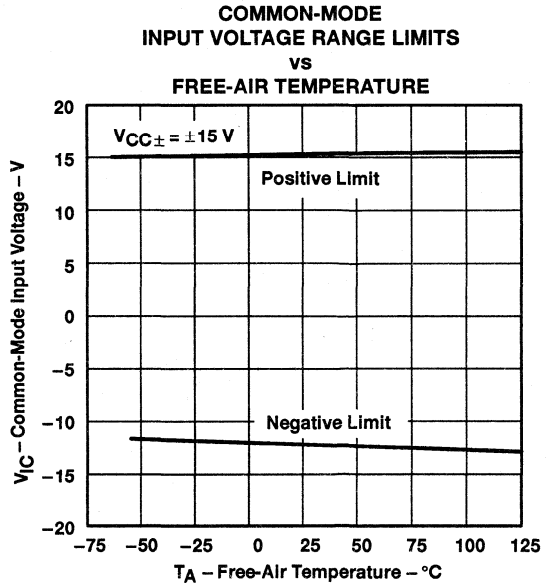


Figure 11

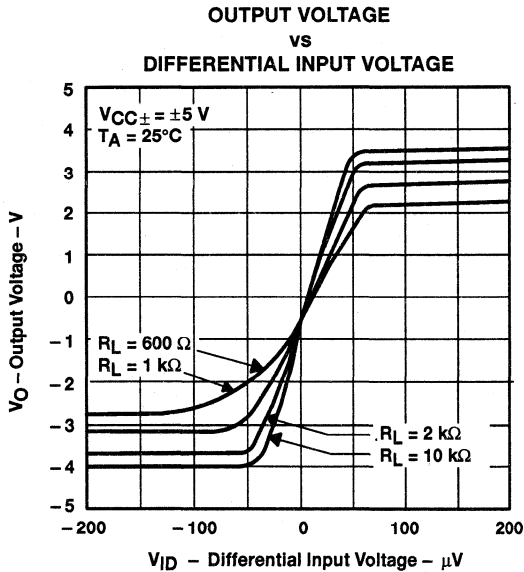


Figure 12

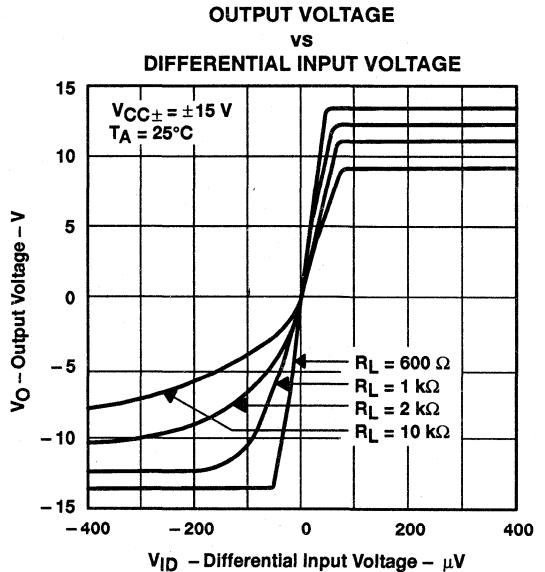
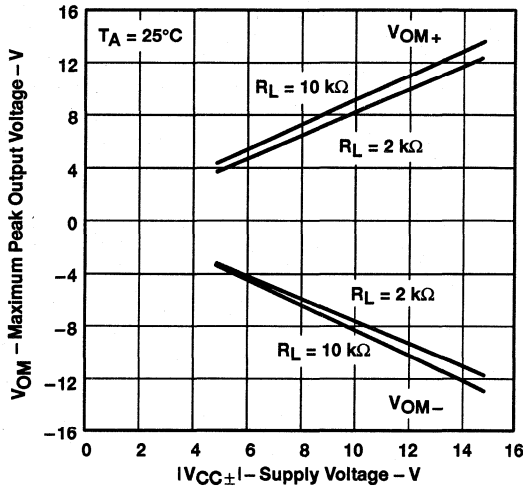


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

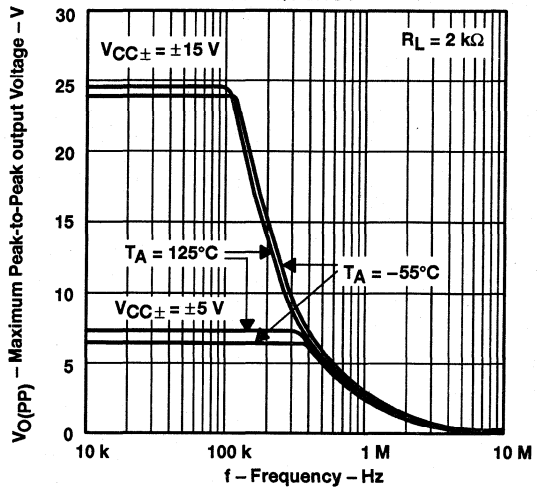
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**SUPPLY VOLTAGE**



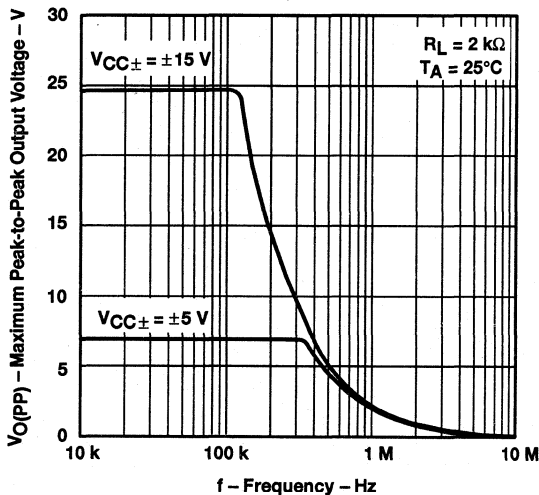
**Figure 14**

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



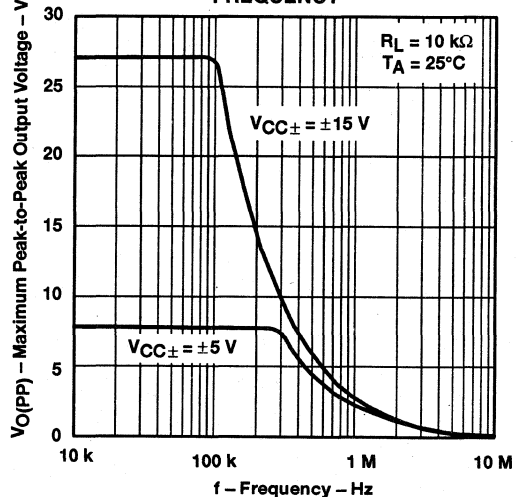
**Figure 15**

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 16**

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 17**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

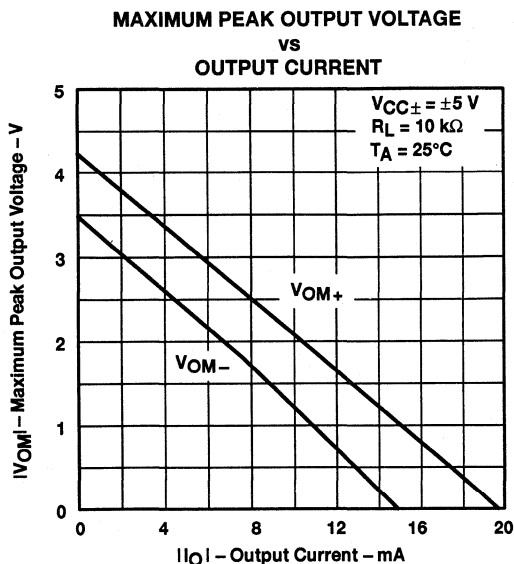


Figure 18

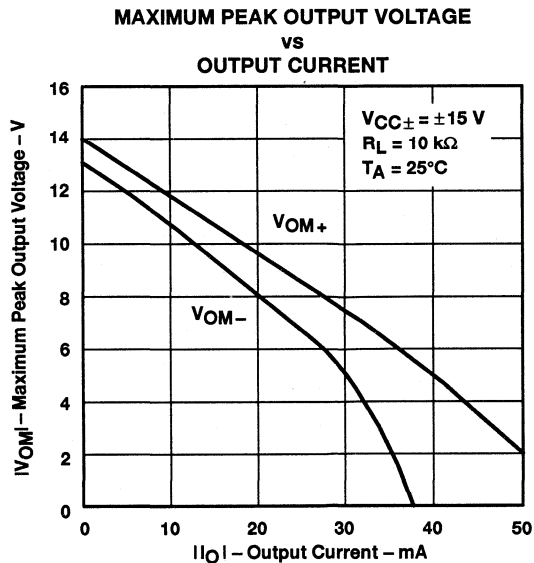


Figure 19

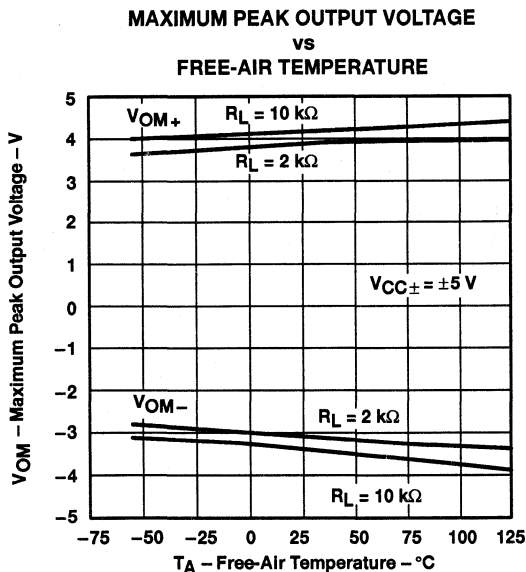


Figure 20

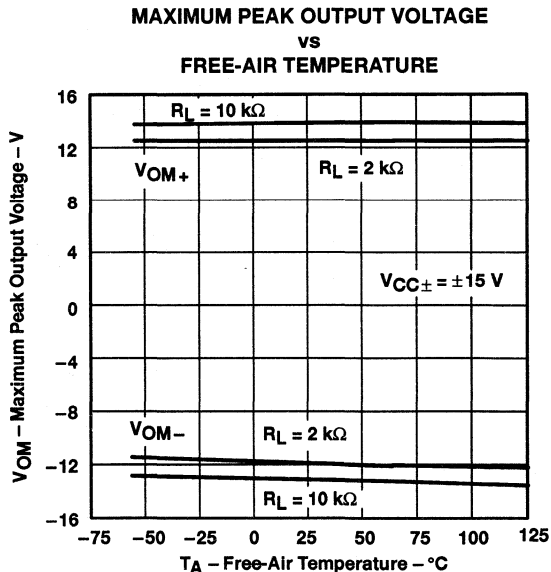
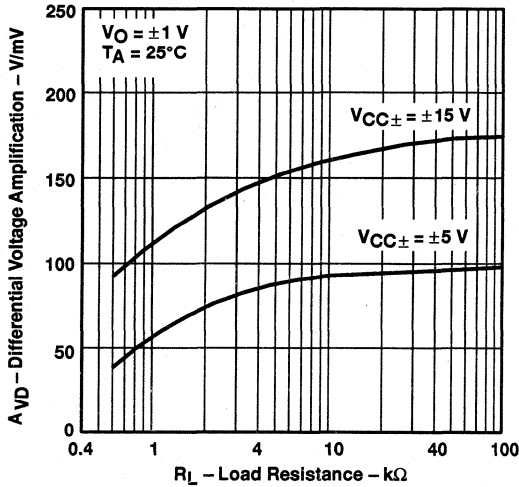


Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

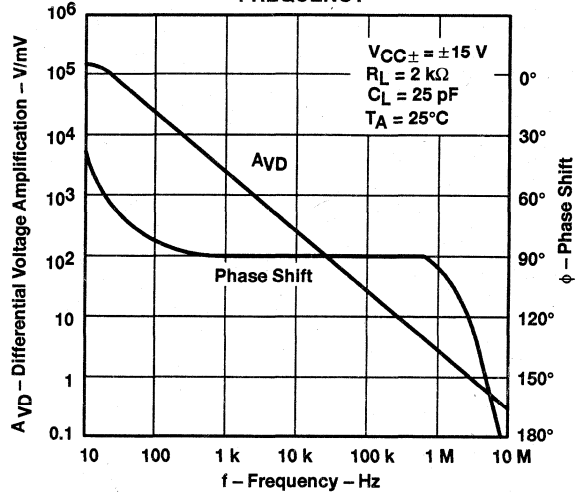
**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL VOLTAGE AMPLIFICATION**  
**vs**  
**LOAD RESISTANCE**



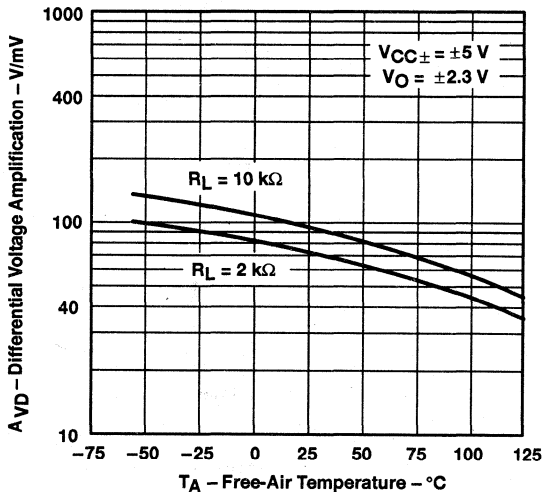
**Figure 22**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION AND PHASE SHIFT**  
**vs**  
**FREQUENCY**



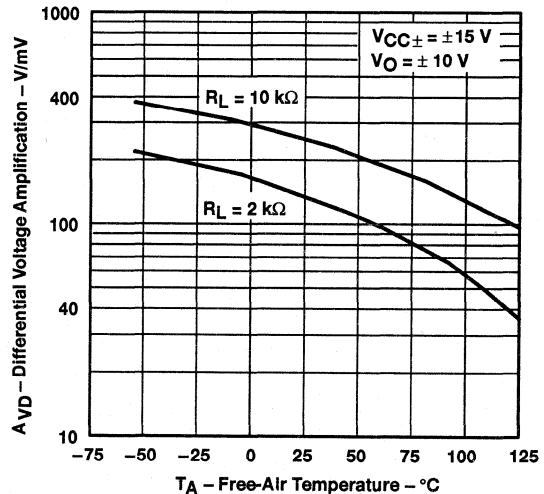
**Figure 23**

**LARGE-SIGNAL VOLTAGE AMPLIFICATION**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 24**

**LARGE-SIGNAL VOLTAGE AMPLIFICATION**  
**vs**  
**FREE-AIR TEMPERATURE**



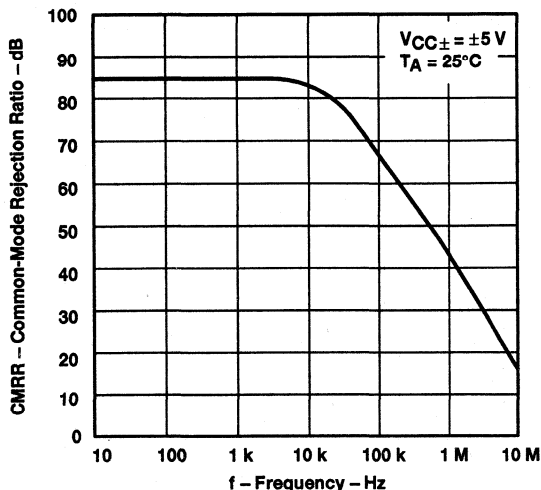
**Figure 25**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



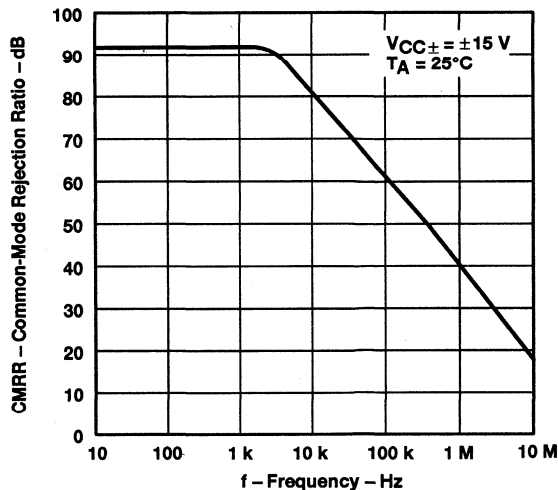
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY**



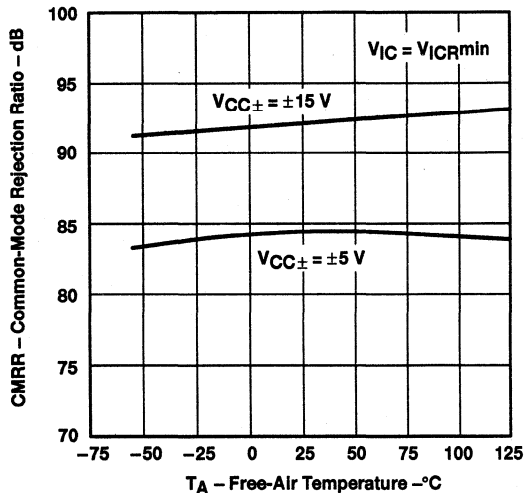
**Figure 26**

**COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY**



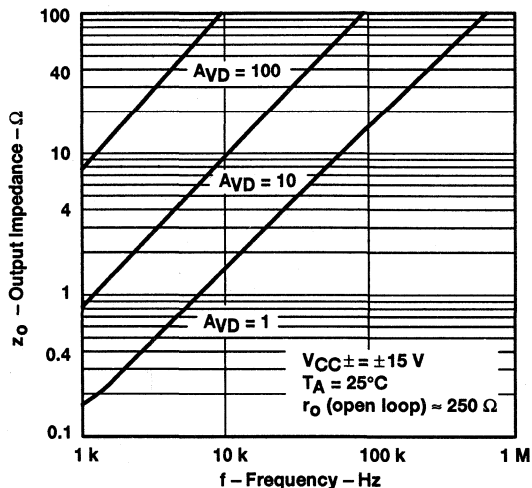
**Figure 27**

**COMMON-MODE REJECTION RATIO  
 VS  
 FREE-AIR TEMPERATURE**



**Figure 28**

**OUTPUT IMPEDANCE  
 VS  
 FREQUENCY**



**Figure 29**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

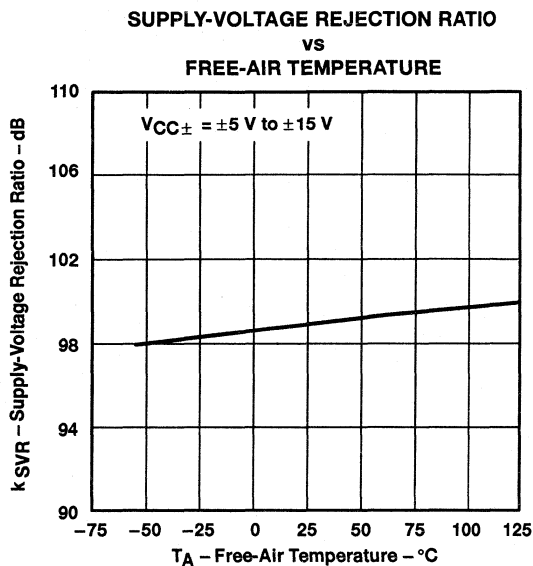


Figure 30

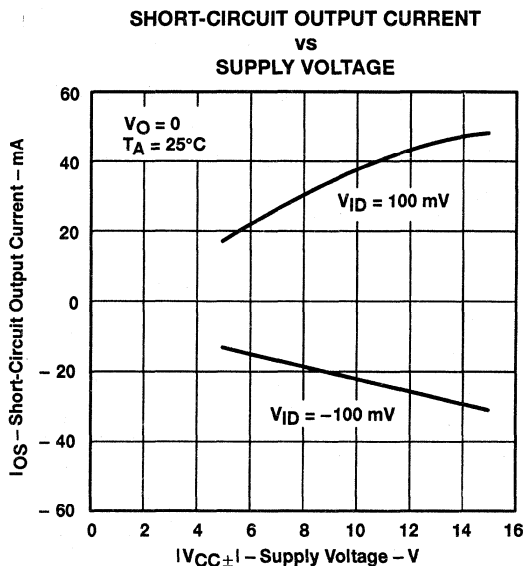


Figure 31

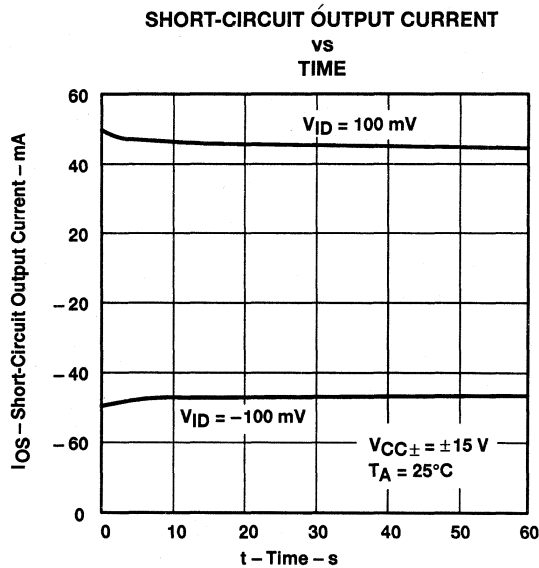


Figure 32

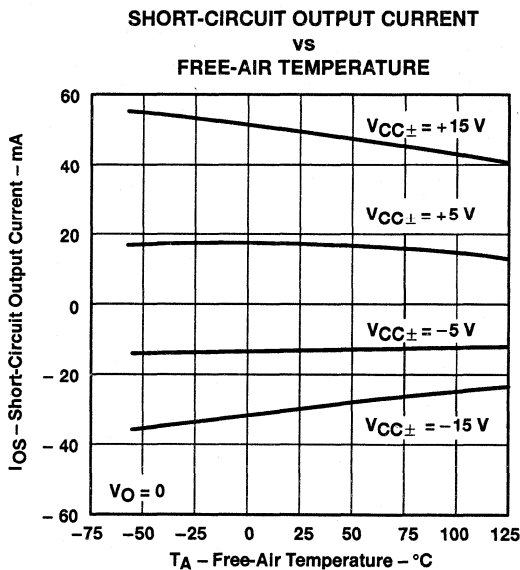


Figure 33

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

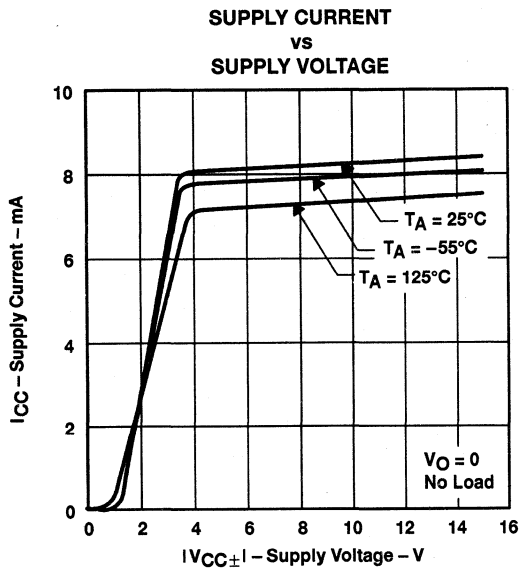


Figure 34

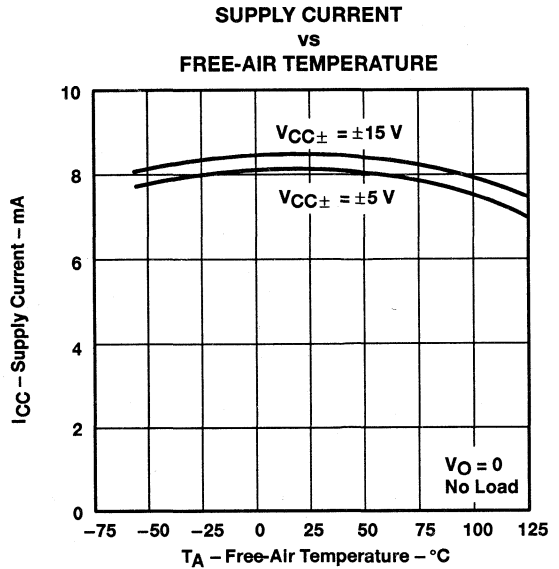


Figure 35

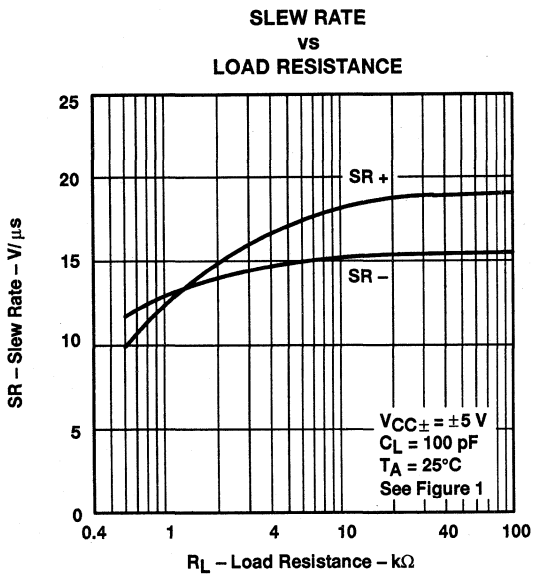


Figure 36

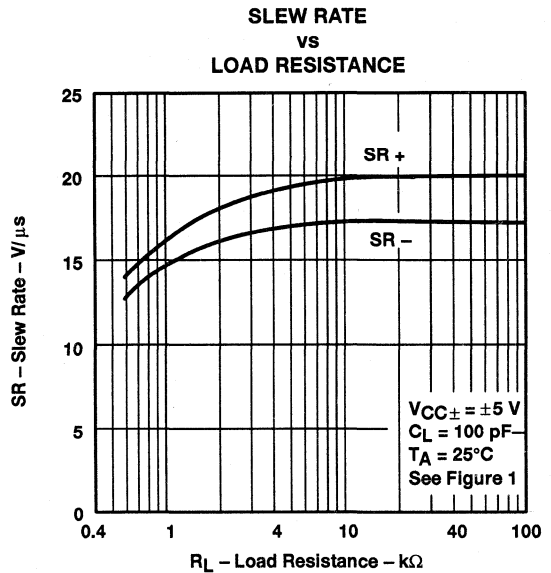


Figure 37

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

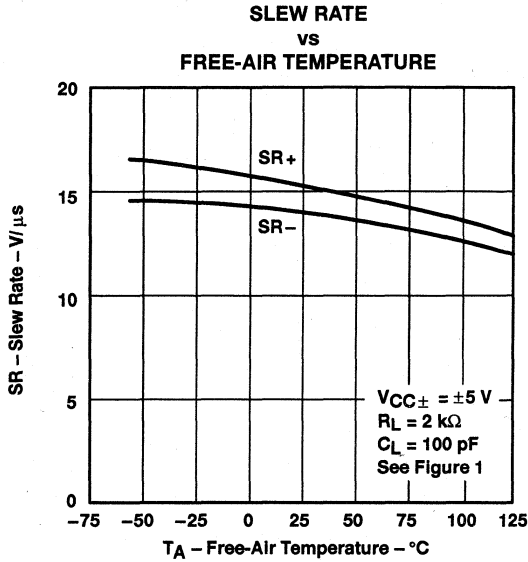


Figure 38

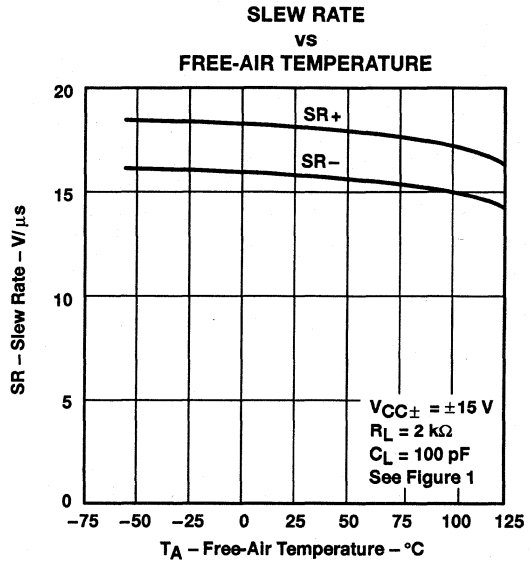


Figure 39

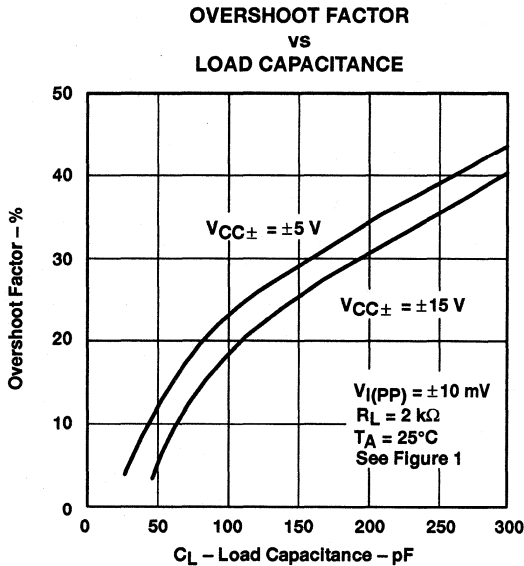


Figure 40

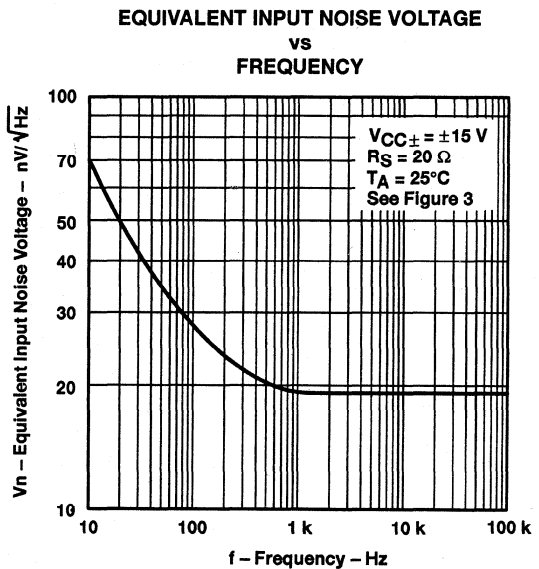


Figure 41

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

TOTAL HARMONIC DISTORTION  
 vs  
 FREQUENCY

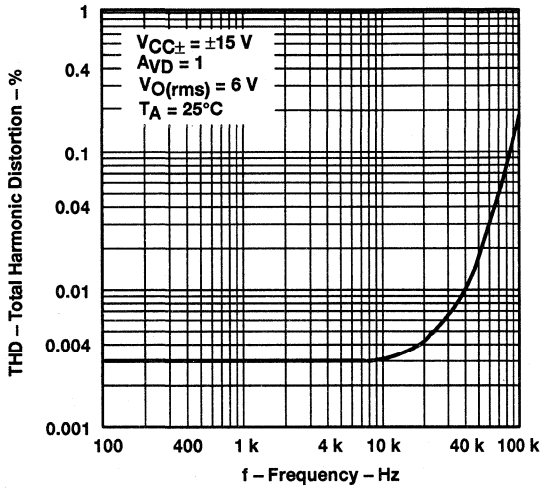


Figure 42

UNITY-GAIN BANDWIDTH  
 vs  
 SUPPLY VOLTAGE

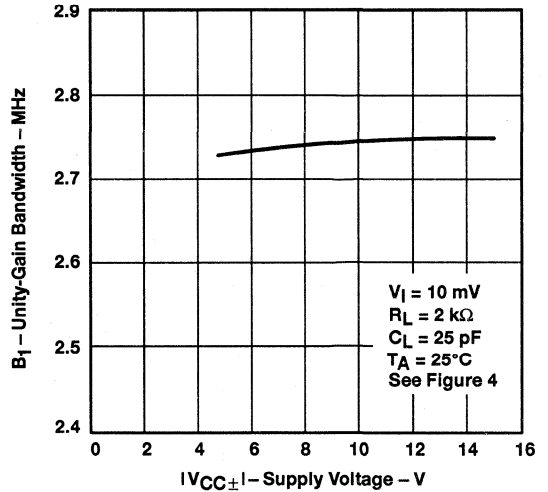


Figure 43

UNITY-GAIN BANDWIDTH  
 vs  
 FREE-AIR TEMPERATURE

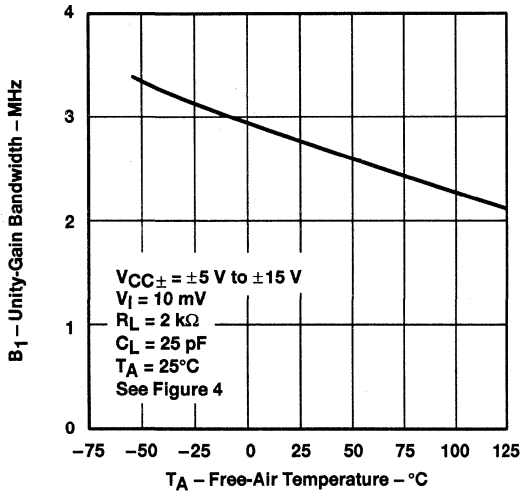


Figure 44

PHASE MARGIN  
 vs  
 SUPPLY VOLTAGE

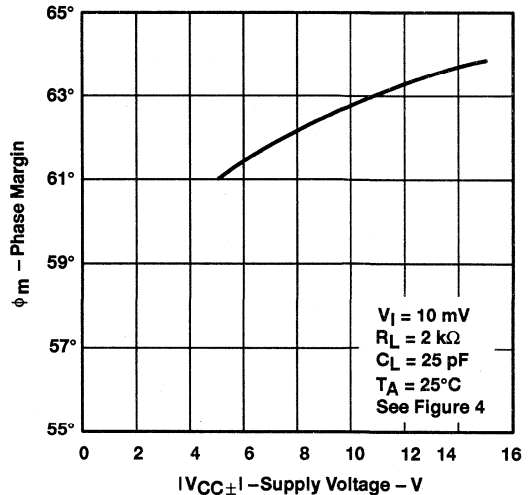
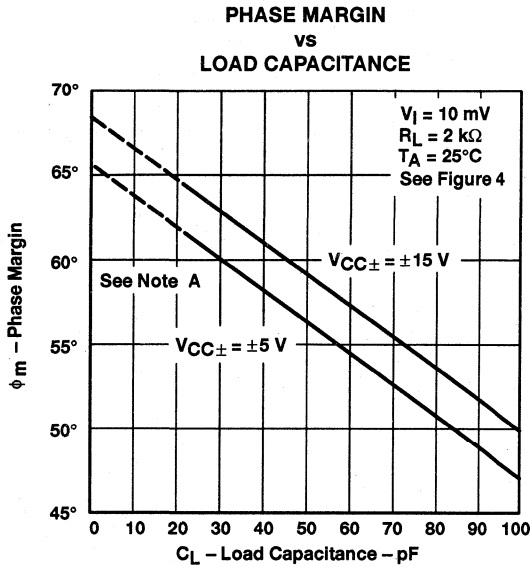


Figure 45

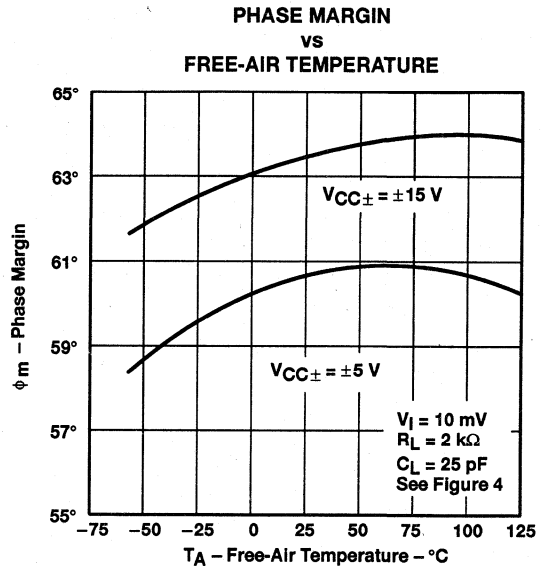
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

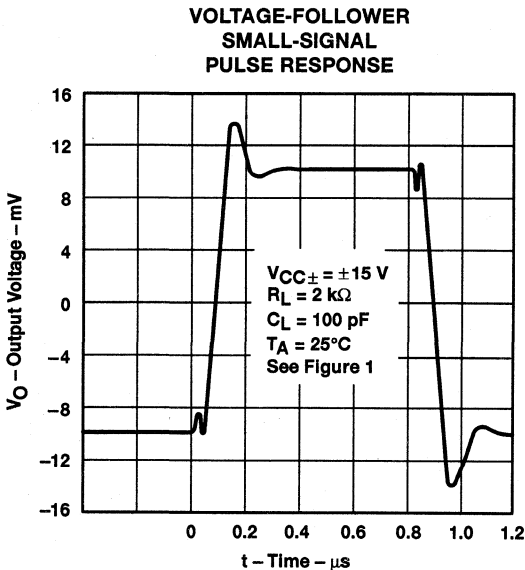


NOTE A: Values of phase margin below a load capacitance of 25 pF were estimated.

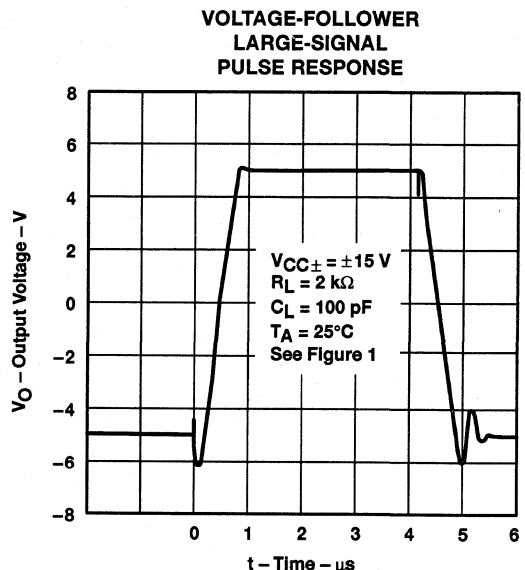
**Figure 46**



**Figure 47**



**Figure 48**



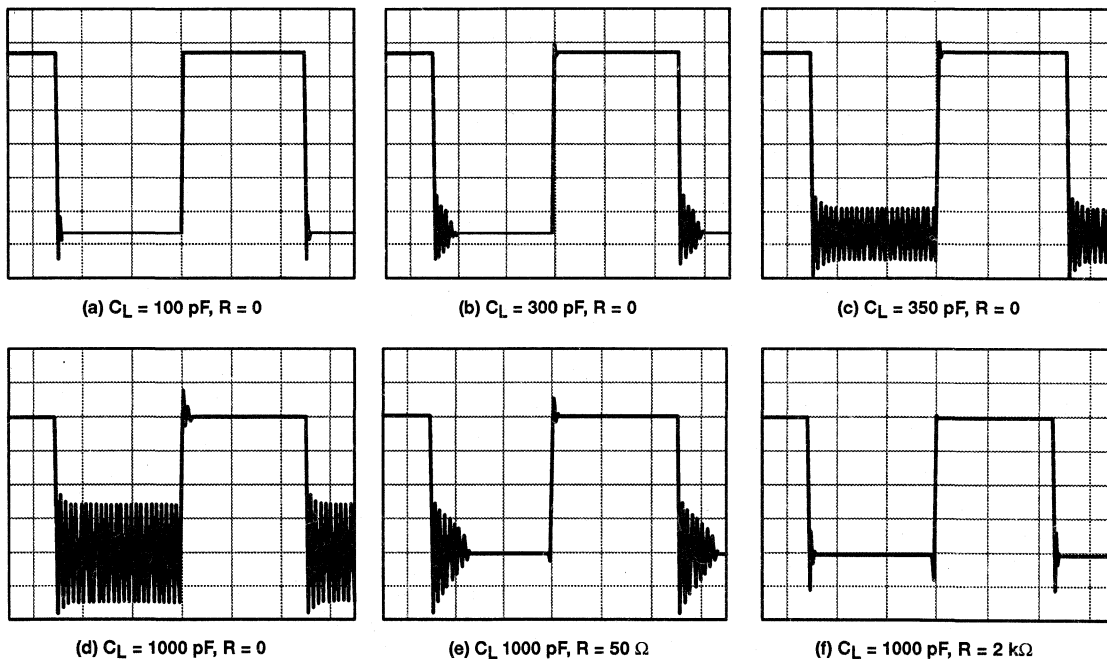
**Figure 49**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

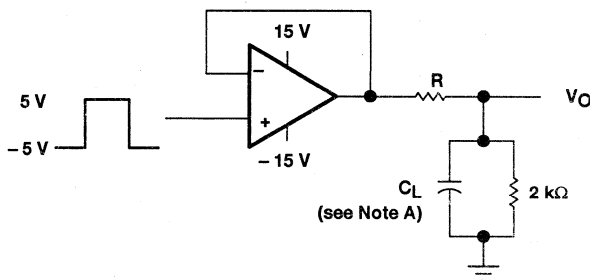
**APPLICATION INFORMATION**

**output characteristics**

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL054 and TL054A drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF and larger may be driven if enough resistance is added in series with the output (see Figure 50).



**Figure 50. Effect of Capacitive Loads**



**Figure 51. Test Circuit for Output Characteristics**

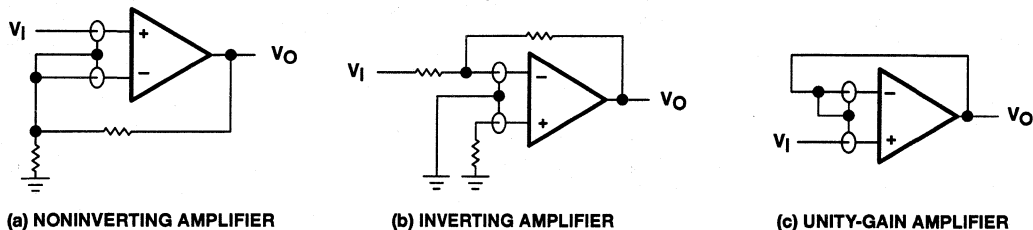
**APPLICATION INFORMATION**

**Input characteristics**

The TL054 and TL054A are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low bias current requirements, the TL054 and TL054A are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 52). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.



**Figure 52. Use of Guard Rings**

**noise performance**

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TL054 and TL054A result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k $\Omega$ .



## APPLICATION INFORMATION

### instrumentation amplifier with adjustable gain/null

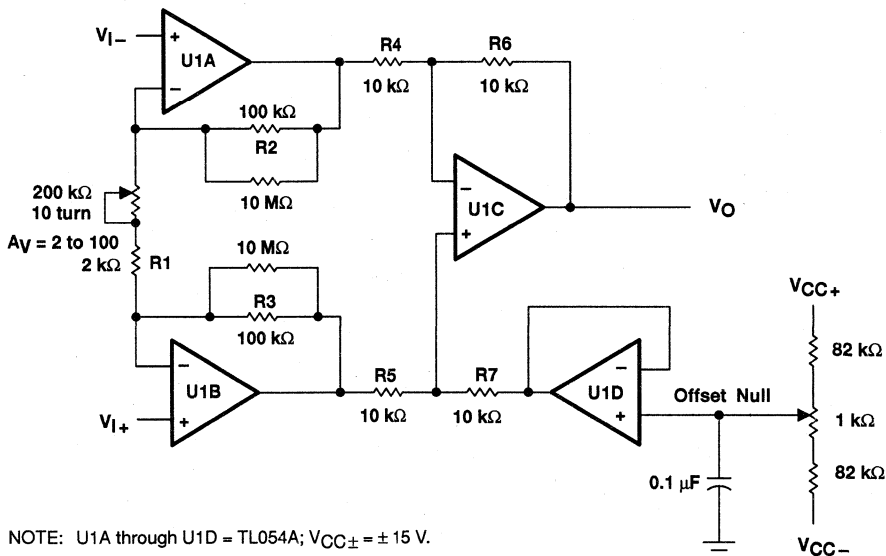
The instrumentation amplifier in Figure 53 benefits greatly from the high input impedance and stable input offset voltage of the TL054A. Amplifiers U1A, U1B, and U1C form the actual instrumentation amplifier, while U1D provides offset null. Potentiometer R1 provides gain adjust. With  $R1 = 2\text{ k}\Omega$ , the circuit gain equals 100, while with  $R1 = 200\text{ k}\Omega$ , the circuit gain equals two. The following equation shows the instrumentation amplifier gain as a function of  $R1$ :

$$A_V = 1 + \left( \frac{R2 + R3}{R1} \right)$$

Readjusting the offset null is necessary whenever the circuit gain is changed. If U1D is needed for another application, R7 can be terminated at ground. The low input offset voltage of the TL054A minimizes the dc error of the circuit. For best matching, all resistors should be one percent tolerance. The matching between R4, R5, R6, and R7 controls the CMRR of this application.

The following equation shows the output voltages when the input voltage equals zero. This dc error can be nulled by adjusting the offset null potentiometer; however, any change in offset voltage over time or temperature also creates an error. To calculate the error from changes in offset, consider the three offset components in the equation as delta offsets rather than initial offsets. The improved stability of Texas Instruments enhanced JFETs minimizes the error resulting from change in input offset voltage with time. Assuming  $V_i$  equals zero,  $V_o$  can be shown as a function of the offset voltage:

$$V_o = V_{IO2} \left[ \left( 1 + \frac{R3}{R1} \right) \left( \frac{R7}{R5 + R7} \right) \left( 1 + \frac{R6}{R4} \right) + \frac{R2}{R1} \left( \frac{R6}{R4} \right) \right] - V_{IO1} \left[ \frac{R3}{R1} \left( \frac{R7}{R5 + R7} \right) \left( 1 + \frac{R6}{R4} \right) + \frac{R6}{R4} \left( 1 + \frac{R2}{R1} \right) \right] + V_{IO3} \left( 1 + \frac{R6}{R4} \right)$$



**Figure 53. Instrumentation Amplifier**

**TL054, TL054A**  
**ENHANCED-JFET PRECISION**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS038C – JUNE 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

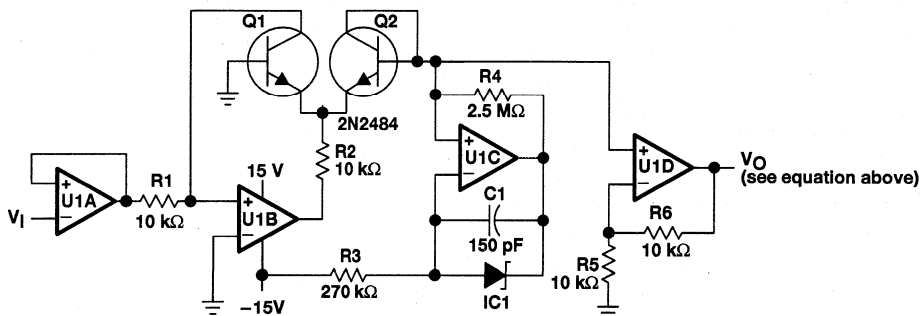
**high input impedance log amplifier**

The low input offset voltage and high input impedance of the TL054a create a precision log amplifier (see Figure 54). IC1 is a 2.5-V, low-current precision, shunt regulator. Transistors Q1 and Q2 must be a closely matched NPN pair. For best performance over temperature, R4 should be a metal film resistor with a low temperature coefficient.

In this circuit, U1A serves as a high-impedance unity-gain buffer. Amplifier U1B converts the input voltage to a current through R1 and Q1. Amplifier U1C, IC1, and R4 form a 1 μA temperature-stable current source that sets the base-emitter voltage of Q2. Amplifies the difference between the base-emitter voltage of Q1 and Q2. The output voltage is given by the following equation:

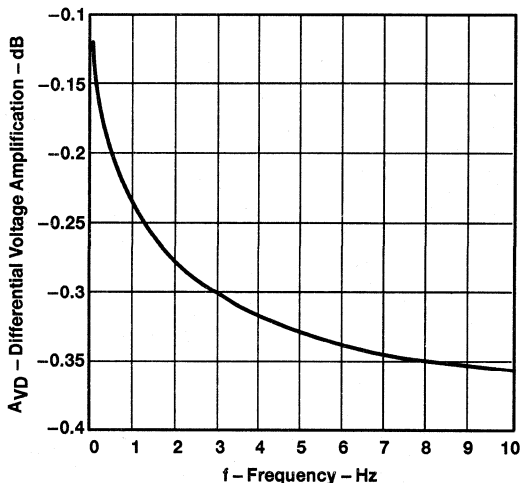
$$V_O = - \left[ 1 + \frac{R_6}{R_5} \right] \frac{kT}{q} \ln \left[ \frac{V_I}{(R_1 \times 1 \times 10^{-6})} \right]$$

where  $k = 1.38 \times 10^{-23}$ ,  $q = 1.602 \times 10^{-19}$ , and T is in degrees kelvin.



NOTE: U1A through U1D = TL054A. IC1 = LM385, LT1004, or LT1009 voltage reference.

**Figure 54. Log Amplifier**



**Figure 55. Output Voltage vs Input Voltage for Log Amplifier**



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

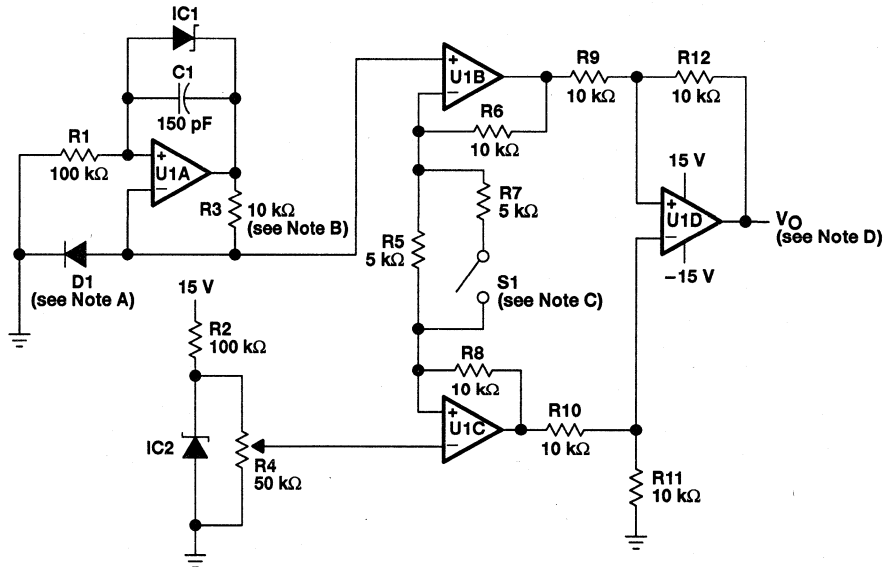
**APPLICATION INFORMATION**

**analog thermometer**

By combining a current source that does not vary over temperature with an instrumentation amplifier, a precise analog thermometer can be built (see Figure 56). Amplifier U1A and IC1 establish a constant current through the temperature sensing diode D1. For this section of the circuit to operate correctly, the TL054 must use split supplies and R3 must be a metal-film resistor with a low temperature coefficient.

The temperature-sensitive voltage from the diode is compared to a temperature-stable voltage reference set by IC2. R4 should be adjusted to provide the correct output voltage when the diode is at a known temperature. Although this potentiometer resistance varies with temperature, the divider ratio of the potentiometer remains constant.

Amplifiers U1B, U1C, and U1D form the instrumentation amplifier that converts the difference between the diode and reference voltage to a voltage proportional to the temperature. With switch S1 closed, the amplifier gain equals 5, and the output voltage is proportional to temperature in degrees Celsius. With S1 open, the amplifier gain is 9, and the output is proportional to temperature in degrees Fahrenheit. Every time that S1 is changed, R4 must be recalibrated. By setting S1 correctly, the output voltage equals 10 mV per degree (C or F).



- NOTES: A. Temperature-sensing diode  $\approx (-2 \text{ mV}/^\circ\text{C})$ .  
 B. Metal-film resistor (low temperature coefficient).  
 C. Switch open for  $^\circ\text{F}$  and closed for  $^\circ\text{C}$ .  
 D.  $V_O \propto \text{temperature}$ ;  $10 \text{ mV}/^\circ\text{C}$  or  $10 \text{ mV}/^\circ\text{F}$ .  
 E. U1A thru U1D = TL054. IC1, IC2 = LM385, LT1004, or LT1009 voltage reference.

**Figure 56. Analog Thermometer**

**APPLICATION INFORMATION**

**voltage-ratio-to-dB converter**

The application in Figure 57 measures the amplitude ratio of two signals and then converts the ratio to decibels. The output voltage provides a resolution of 100 mV/dB. The two inputs can be either dc or sinusoidal ac signals. When using ac signals, both signals should be the same frequency or output glitches will occur. For measuring two input signals of different frequencies, extra filtering should be added after the rectifiers.

The circuit contains three low-offset TL054A devices. Two of these devices provide the rectification and logarithmic conversion of the inputs. The third TL054A forms an instrumentation amplifier. The stage performing the logarithmic conversion also requires two well-matched NPN transistors.

The input signal first passes through a high impedance unity-gain buffer U1A (U2A). Then U1B (U2B) rectifies the input signal at a gain of 0.5, and U1C (U2C) provides a noninverting gain of 2 so that the system gain is still one. U1D (U2D), R6 (R13), and Q1 (Q2) perform the logarithmic conversion of the rectified input signal. The instrumentation amplifier formed by U3A, U3B, U3D scales the difference of the two logarithmic voltages by a gain of 33.6. As a result, the output voltage equals 100 mV/dB. The 1-kΩ potentiometer on the input of U3C calibrates the zero dB reference level. The following equations are used to derive the relationship between the input voltage ratio expressed in decibels and the output voltage.

$$X \text{ dB} = 20 \log \left[ \frac{V_A}{V_B} \right] = 20 \left[ \frac{\ln(V_A) - \ln(V_B)}{\ln(10)} \right]$$

$$X \text{ dB} = 8.686 \left[ \ln(V_A) - \ln(V_B) \right]$$

$$V_{BE(Q1)} = \frac{kT}{q} \ln \left[ \frac{V_A}{R \times I_S} \right] \quad V_{BE(Q2)} = \frac{kT}{q} \ln \left[ \frac{V_B}{R \times I_S} \right]$$

$$\Delta V_{BE} = V_{BE(Q1)} - V_{BE(Q2)} = \frac{kT}{q} \left[ \ln(V_A) - \ln(V_B) \right]$$

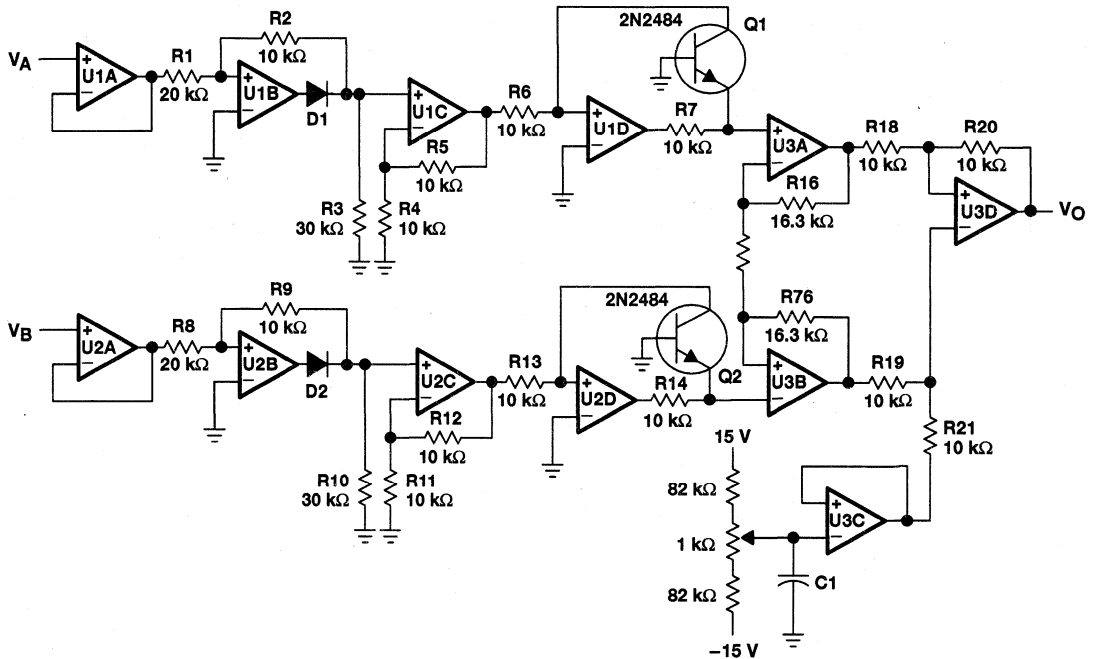
$$X \text{ dB} = \frac{8.686}{kT/q} \left[ V_{BE(Q1)} - V_{BE(Q2)} \right] = 336 \left[ V_{BE(Q1)} - V_{BE(Q2)} \right] \text{ at } 25^\circ\text{C}$$

where

$$k = 1.38 \times 10^{-23}, \quad q = 1.602 \times 10^{-19}, \quad \text{and } T \text{ is in kelvins.}$$

This would give a resolution of 1 V/dB. Therefore, the gain of the instrumentation amplifier is set at 33.6 to obtain 100 mV/dB.

APPLICATION INFORMATION



NOTE: U1A through U3D = TL054A,  $V_{CC} \pm = \pm 15$  V. D1 and D2 = 1N914.

Figure 57. Voltage-Ratio-to-dB Converter

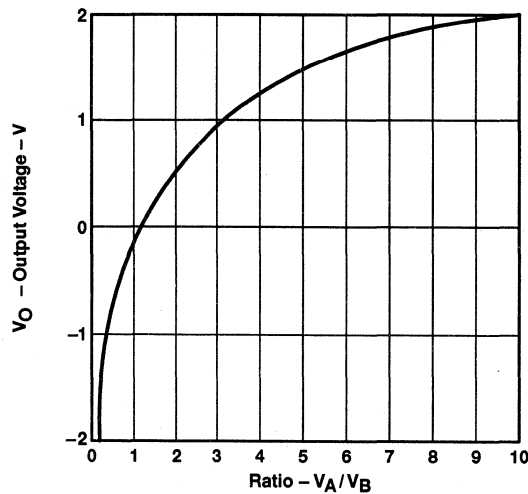


Figure 58. Output Voltage Versus the Ratio of the Input Voltages for Voltage-to-dB Converter



# TL061, TL061A, TL061B, TL061Y, TL062, TL062A TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

## 15 DEVICES COVER MILITARY, INDUSTRIAL, AND COMMERCIAL TEMPERATURE RANGES

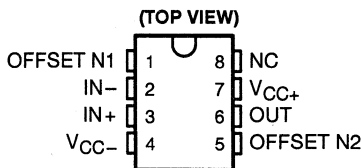
- Very Low Power Consumption
- Typical Supply Current . . . 200  $\mu$ A (per Amplifier)
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Common-Mode Input Voltage Range Includes  $V_{CC+}$
- Output Short-Circuit Protection
- High Input impedance . . . JFET-Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . . 3.5 V/ $\mu$ s Typ

### description

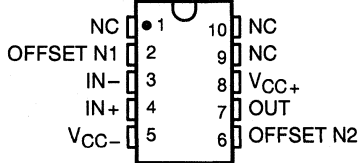
The JFET-input operational amplifiers of the TL06\_ series are designed as low-power versions of the TL08\_ series amplifiers. They feature high input impedance, wide bandwidth, high slew rate, and low input offset and bias currents. The TL06\_ series features the same terminal assignments as the TL07\_ and TL08\_ series. Each of these JFET-input operational amplifiers incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C, and the M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

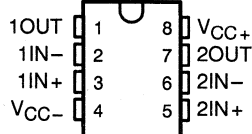
**TL061, TL061A, TL061B  
D, JG, P, OR PW PACKAGE**



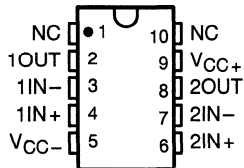
**TL061 . . . U PACKAGE  
(TOP VIEW)**



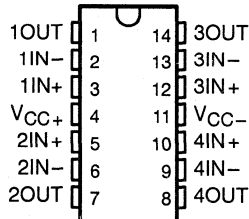
**TL062, TL062A, TL062B  
D, JG, P, OR PW PACKAGE  
(TOP VIEW)**



**TL062 . . . U PACKAGE  
(TOP VIEW)**



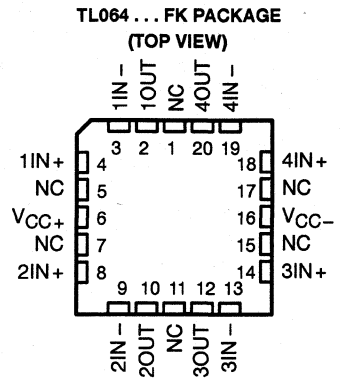
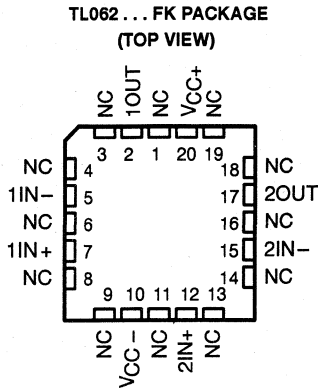
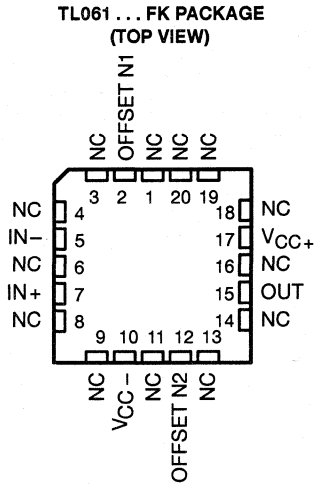
**TL064 . . . D, J, N, PW, OR W PACKAGE  
TL064A, TL064B . . . D OR N PACKAGE  
(TOP VIEW)**



NC – No internal connection

# TL061, TL061A, TL061B, TL061Y, TL062, TL062A TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994



NC – No internal connection

## AVAILABLE OPTIONS

TA	V <sub>IOMax</sub> AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D008)†	SMALL OUTLINE (D014)†	PLASTIC DIP (N)	PLASTIC DIP (P)	TSSOP (PW)	
0°C to 70°C	15 mV 6 mV 3 mV	TL061CD TL061ACD TL061BCD			TL061CP TL061ACP TL061BCP	TL061CPW	TL061Y
	15 mV 6 mV 3 mV	TL062CD TL062ACD TL062BCD			TL062CP TL062ACP TL062BCP	TL062CPW	TL062Y
	15 mV 6 mV 3 mV		TL064CD TL064ACD TL064BCD	TL064CN TL064ACN TL064BCN		TL064CPW	TL064Y

TA	V <sub>IOMax</sub> AT 25°C	PACKAGES								
		SMALL OUTLINE (D008)†	SMALL OUTLINE (D014)†	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	PLASTIC DIP (N)	PLASTIC DIP (P)	FLAT PACK (U)	FLAT PACK (W)
-40°C to 85°C	6 mV	TL0611D TL0621D	TL0641D				TL0641N	TL0611P TL0621P		
-55°C to 125°C	6 mV 6 mV 9 mV			TL061MFK TL062MFK TL064MFK	TL064MJ	TL061MJG TL062MJG			TL061MU TL062MU	TL064MW

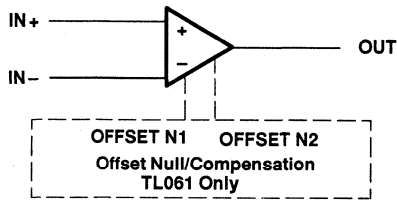
† The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL061CDR).



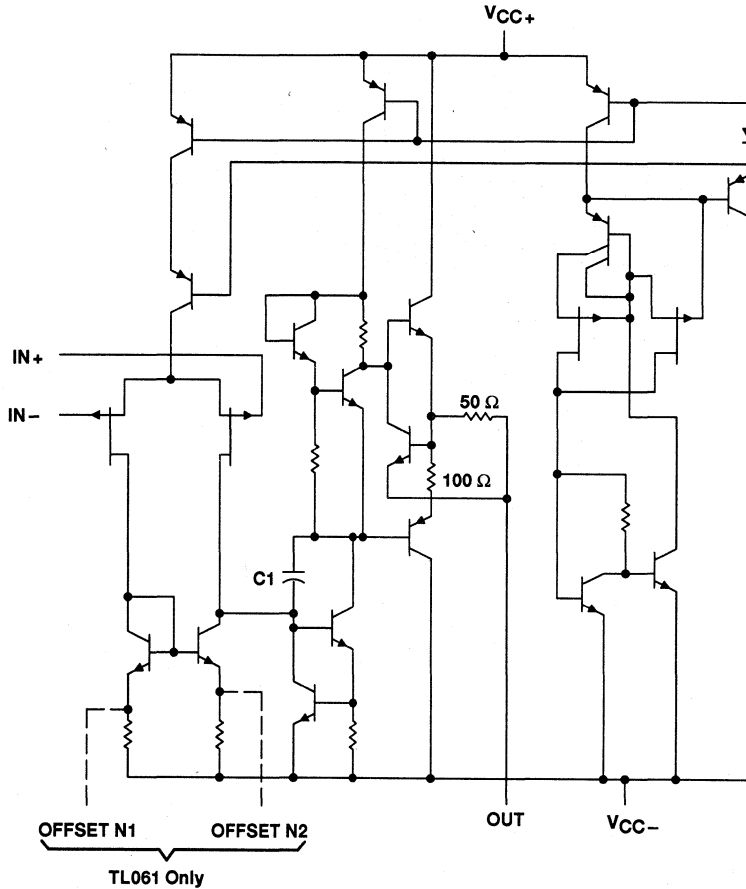
TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B - NOVEMBER 1978 - REVISED AUGUST 1994

symbol (each amplifier)



schematic (each amplifier)

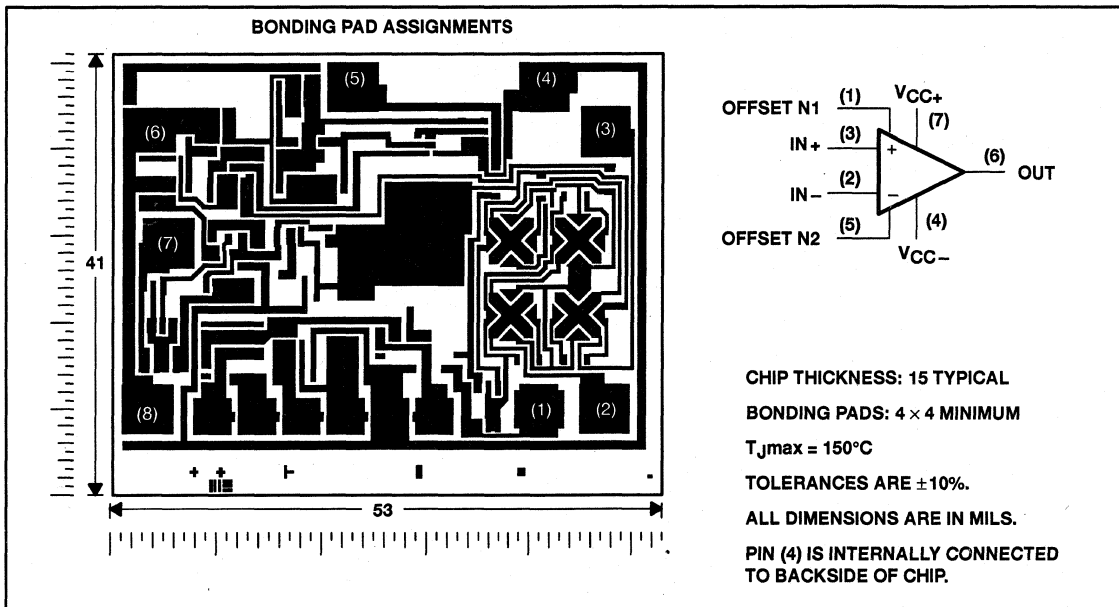


C1 = 10 pF on TL061, TL062, and TL064  
 Component values shown are nominal.

**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**  
 SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**TL061Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL061. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.

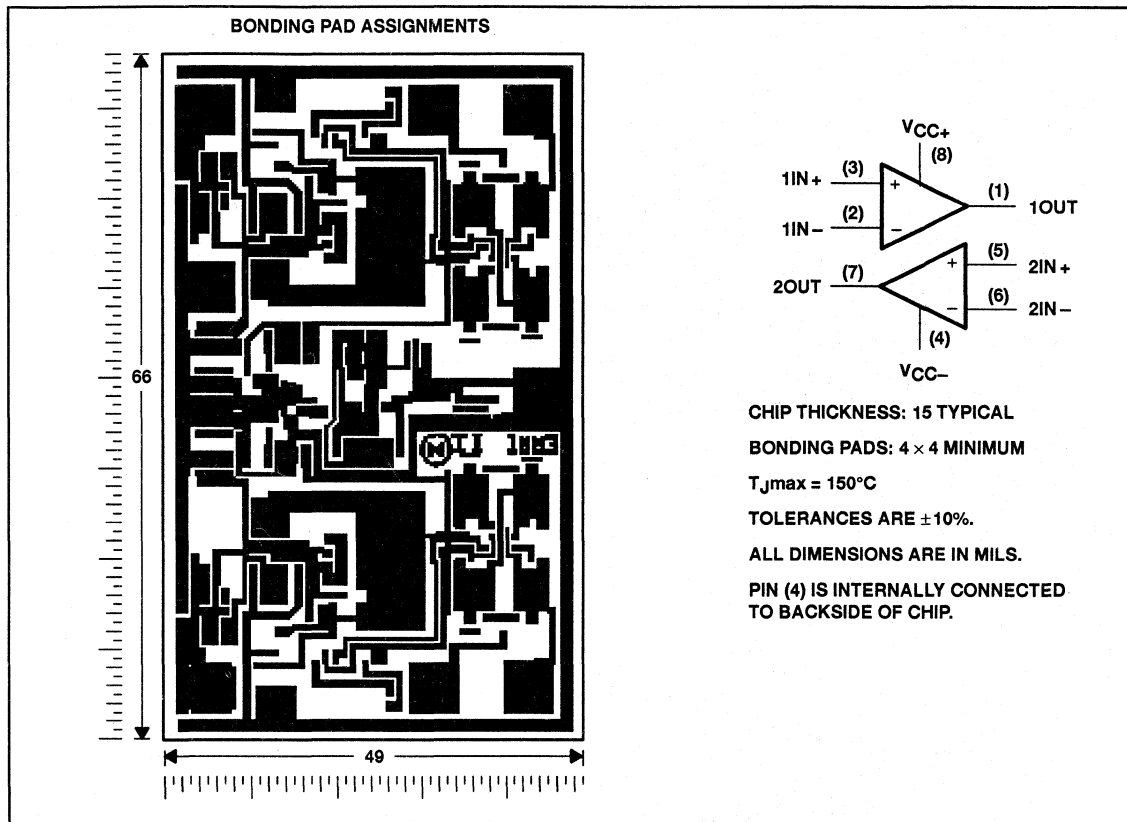


TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B - NOVEMBER 1978 - REVISED AUGUST 1994

**TL062Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL062. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.

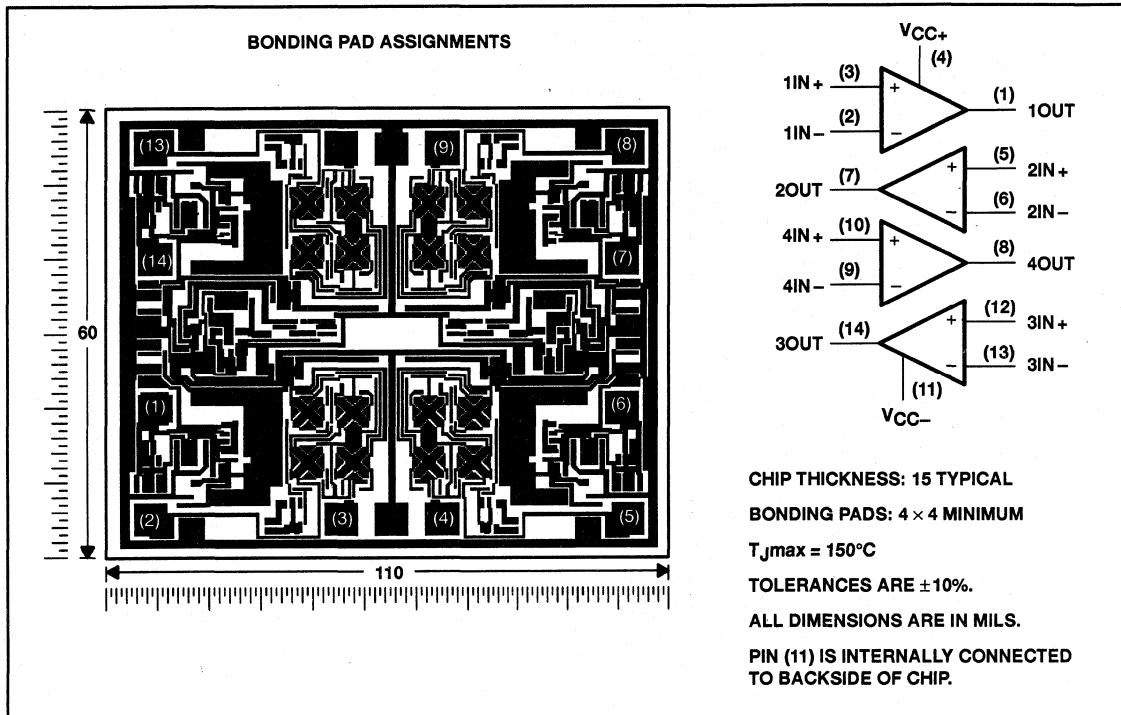


**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B - NOVEMBER 1978 - REVISED AUGUST 1994

**TL064Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL064. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

	TL06_C TL06_AC TL06_BC	TL06_I	TL06_M	UNIT
Supply voltage, $V_{CC+}$ (see Note 1)	18	18	18	V
Supply voltage, $V_{CC-}$ (see Note 1)	-18	-18	-18	V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$	$\pm 30$	$\pm 30$	V
Input voltage, $V_I$ (see Notes 1 and 3)	$\pm 15$	$\pm 15$	$\pm 15$	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range	0 to 70	-40 to 85	-55 to 125	°C
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	°C
Case temperature for 60 seconds	FK package		260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J, JG, U, or W package		300	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N, P, or PW package	260	260	°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values except differential voltages are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.  
 4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D (8 pin)	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	N/A
D (14 pin)	680 mW	7.6 mW/°C	60°C	608 mW	494 mW	N/A
FK	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
JG	680 mW	8.4 mW/°C	69°C	672 mW	546 mW	210 mW
N	680 mW	9.2 mW/°C	76°C	680 mW	598 mW	N/A
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	N/A
PW (8 pin)	525 mW	4.2 mW/°C	25°C	336 mW	N/A	N/A
PW (14 pin)	700 mW	5.6 mW/°C	25°C	448 mW	N/A	N/A
U	675 mW	5.4 mW/°C	25°C	432 mW	351 mW	135 mW
W	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	200 mW

# TL061, TL061A, TL061B, TL061Y, TL062, TL062A TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

electrical characteristics,  $V_{CC\pm} = \pm 15\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†		TL061C			TL061AC			TL061BC			TL0611			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$	$T_A = 25^\circ\text{C}$	3	15	3	6	3	6	2	3	3	6	6	mV	
		$T_A = \text{Full range}$		20		7.5		5		5		9			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_O = 0$ , $T_A = \text{Full range}$	$R_S = 50\ \Omega$	10		10		10		10		10		10	$\mu\text{V}/^\circ\text{C}$	
		$T_A = 25^\circ\text{C}$	5	200	5	100	5	100	5	100	5	100	5		100
$I_{IO}$ Input offset current	$V_O = 0$	$T_A = 25^\circ\text{C}$		5		3		3		3		3		pA	
		$T_A = \text{Full range}$		5		3		3		3		3			
$I_{IB}$ Input bias current‡	$V_O = 0$	$T_A = 25^\circ\text{C}$	30	400	30	200	30	200	30	200	30	200	30	pA	
		$T_A = \text{Full range}$		10		7		7		7		20			
$V_{ICR}$ Common-mode input voltage range	$T_A = 25^\circ\text{C}$		-12	to	-12	to	-12	to	-12	to	-12	to	-12	V	
			$\pm 11$	to	$\pm 11$	to	$\pm 11$	to	$\pm 11$	to	$\pm 11$	to	$\pm 11$		
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , $R_L \geq 10\ \text{k}\Omega$	$T_A = 25^\circ\text{C}$	$\pm 10$	$\pm 13.5$	$\pm 10$	$\pm 13.5$	$\pm 10$	$\pm 13.5$	$\pm 10$	$\pm 13.5$	$\pm 10$	$\pm 13.5$	$\pm 10$	V	
		$T_A = \text{Full range}$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$	$\pm 10$		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V}$ , $R_L \geq 10\ \text{k}\Omega$	$T_A = 25^\circ\text{C}$	3	6	4	6	4	6	4	6	4	6	4	V/mV	
		$T_A = \text{Full range}$	3	3	4	4	4	4	4	4	4	4	4		
$B_1$ Unity-gain bandwidth	$R_L = 10\ \text{k}\Omega$ , $T_A = 25^\circ\text{C}$		1		1		1		1		1		1	MHz	
			10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		
$f_i$ Input resistance	$V_{IC} = V_{ICRmin}$ , $V_O = 0$ , $R_S = 50\ \Omega$ , $T_A = 25^\circ\text{C}$		70	86	80	86	80	86	80	86	80	86	80	$\Omega$	
			70	86	80	86	80	86	80	86	80	86	80		
CMRR Common-mode rejection ratio	$V_{CC\pm} = \pm 9\ \text{V}$ to $\pm 15\ \text{V}$ , $V_O = 0$ , $R_S = 50\ \Omega$ , $T_A = 25^\circ\text{C}$		70	95	80	95	80	95	80	95	80	95	80	dB	
			70	95	80	95	80	95	80	95	80	95	80		
kSVR Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_O = 0$ , No load	$T_A = 25^\circ\text{C}$	6	7.5	6	7.5	6	7.5	6	7.5	6	7.5	6	dB	
		$T_A = \text{Full range}$		6		6		6		6		6			
PD Total power dissipation (each amplifier)	$V_O = 0$ , No load	$T_A = 25^\circ\text{C}$	200	250	200	250	200	250	200	250	200	250	200	mW	
		$T_A = \text{Full range}$		200		200		200		200		200			
I <sub>CC</sub> Supply current (each amplifier)	$V_O = 0$ , No load	$T_A = 25^\circ\text{C}$	120		120		120		120		120		120	$\mu\text{A}$	
		$T_A = \text{Full range}$		120		120		120		120		120			
$V_{O1}/V_{O2}$ Crosstalk attenuation	$AVD = 100$ , $T_A = 25^\circ\text{C}$		120		120		120		120		120		120	dB	

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Full range for  $T_A$  is  $0^\circ\text{C}$  to  $70^\circ\text{C}$  for TL06\_C, TL06\_AC, and TL06\_BC and  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  for TL06\_I.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 15. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.



**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITION†	TL061M TL062M			TL064M			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0,$ $R_S = 50 \Omega$	$T_A = 25^\circ\text{C}$		3	6	$T_A = 25^\circ\text{C}$		mV
		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		9		15		
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_O = 0,$ $T_A = -55^\circ\text{C to } 125^\circ\text{C}$	$R_S = 50 \Omega,$		10		10		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current	$V_O = 0$	$T_A = 25^\circ\text{C}$		5	100	$T_A = 25^\circ\text{C}$		pA
		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		20		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		nA
$I_{IB}$ Input bias current‡	$V_O = 0$	$T_A = 25^\circ\text{C}$		30	200	$T_A = 25^\circ\text{C}$		pA
		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		50		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		nA
$V_{ICR}$ Common-mode input voltage range	$T_A = 25^\circ\text{C}$	$\pm 11.5$	-12 to 15			$\pm 11.5$	-12 to 15	V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10 \text{ k}\Omega,$ $R_L \geq 10 \text{ k}\Omega,$	$T_A = 25^\circ\text{C}$		$\pm 10$	$\pm 13.5$	$T_A = 25^\circ\text{C}$		V
		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		$\pm 10$		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10 \text{ V},$ $R_L \geq 10 \text{ k}\Omega$	$T_A = 25^\circ\text{C}$		4	6	$T_A = 25^\circ\text{C}$		V/mV
		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		4		$T_A = -55^\circ\text{C to } 125^\circ\text{C}$		
$B_1$ Unity-gain bandwidth	$R_L = 10 \text{ k}\Omega,$ $T_A = 25^\circ\text{C}$							MHz
$r_i$ Input resistance	$T_A = 25^\circ\text{C}$	$10^{12}$			$10^{12}$			$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0,$ $R_S = 50 \Omega,$ $T_A = 25^\circ\text{C}$	80	86			80	86	dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V},$ $R_S = 50 \Omega,$ $V_O = 0,$ $T_A = 25^\circ\text{C}$	80	95			80	95	dB
$P_D$ Total power dissipation (each amplifier)	$V_O = 0,$ No load	$T_A = 25^\circ\text{C},$		6	7.5	$T_A = 25^\circ\text{C},$		mW
$I_{CC}$ Supply current (each amplifier)	$V_O = 0,$ No load	$T_A = 25^\circ\text{C},$		200	250	$T_A = 25^\circ\text{C},$		$\mu\text{A}$
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100,$ $T_A = 25^\circ\text{C}$	120			120			dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 15. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**operating characteristics,  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain (see Note 5)	$V_I = 10 \text{ V},$ $C_L = 100 \text{ pF},$ $R_L = 10 \text{ k}\Omega,$ See Figure 1	1.5	3.5		V/ $\mu\text{s}$
$t_r$ Rise time	$V_I = 20 \text{ V},$ $C_L = 100 \text{ pF},$ $R_L = 10 \text{ k}\Omega,$ See Figure 1	0.2			$\mu\text{s}$
Overshoot factor		10%			
$V_n$ Equivalent input noise voltage	$R_S = 20 \Omega,$ $f = 1 \text{ kHz}$	42			nV/ $\sqrt{\text{Hz}}$

NOTE 5: Slew rate at  $-55^\circ\text{C to } 125^\circ\text{C}$  is 0.7 V/ $\mu\text{s}$  min.



**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONST	TL061Y TL062Y TL064Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$		3	15	mV
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$		10		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current	$V_O = 0$		5	200	pA
$I_{IB}$ Input bias current‡	$V_O = 0$		30	400	pA
$V_{ICR}$ Common-mode input voltage range		$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega$	$\pm 10$	$\pm 13.5$		V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V}$ , $R_L \geq 2\ \text{k}\Omega$	3	6		V/mV
$B_1$ Unity-gain bandwidth	$R_L = 10\ \text{k}\Omega$		1		MHz
$r_i$ Input resistance			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$ , $V_O = 0$ , $R_S = 50\ \Omega$	70	86		dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC\pm} / \Delta V_{IO}$ )	$V_{CC} = \pm 9\ \text{V}$ to $\pm 15\ \text{V}$ , $V_O = 0$ , $R_S = 50\ \Omega$	70	95		dB
$P_D$ Total power dissipation (each amplifier)	$V_O = 0$ , No load		6	7.5	mW
$I_{CC}$ Supply current (per amplifier)	$V_O = 0$ , No load		200	250	$\mu\text{A}$
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 15. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**operating characteristics,  $V_{CC\pm} = \pm 15\ \text{V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TL061Y TL062Y TL064Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$V_I = 10\ \text{mV}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$ , See Figure 1	1.5	3.5		V/ $\mu\text{s}$
$t_r$ Rise time	$V_I = 20\ \text{V}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$ , See Figure 1		0.2		$\mu\text{s}$
Overshoot factor	$C_L = 100\ \text{pF}$ , See Figure 1		10%		
$V_n$ Equivalent input noise voltage	$R_S = 20\ \Omega$ , $f = 1\ \text{kHz}$		42		nV/ $\sqrt{\text{Hz}}$



PARAMETER MEASUREMENT INFORMATION

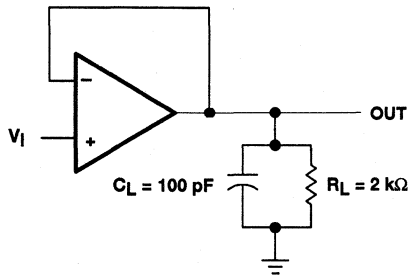


Figure 1. Unity-Gain Amplifier

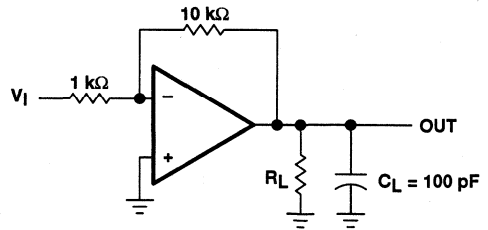


Figure 2. Gain-of-10 Inverting Amplifier

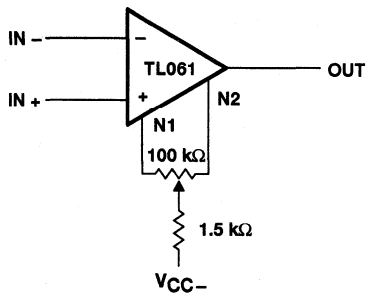


Figure 3. Input Offset Voltage Null Circuit

**TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			FIGURE
V <sub>OM</sub>	Maximum output voltage	vs Supply voltage	4
		vs Free-air temperature	5
		vs Load resistance	6
		vs Frequency	7
A <sub>VD</sub>	Differential voltage amplification	vs Free-air temperature	8
A <sub>VD</sub>	Large-signal differential voltage amplification	vs Frequency	9
	Phase shift	vs Frequency	9
I <sub>CC</sub>	Supply current	vs Supply voltage	10
		vs Free-air temperature	11
P <sub>D</sub>	Total power dissipation	vs Free-air temperature	12
CMRR	Common-mode rejection ratio	vs Free-air temperature	13
		Normalized unity-gain bandwidth	14
		Normalized slew rate	14
		Normalized phase shift	14
I <sub>IB</sub>	Input bias current	vs Free-air temperature	15
		Large-signal pulse response	16
V <sub>O</sub>	Output voltage	vs Elapsed time	17
V <sub>n</sub>	Equivalent input noise voltage	vs Frequency	18

TYPICAL CHARACTERISTICS†

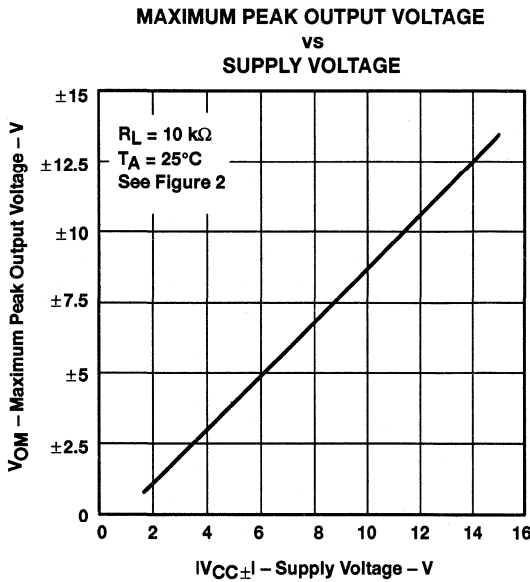


Figure 4

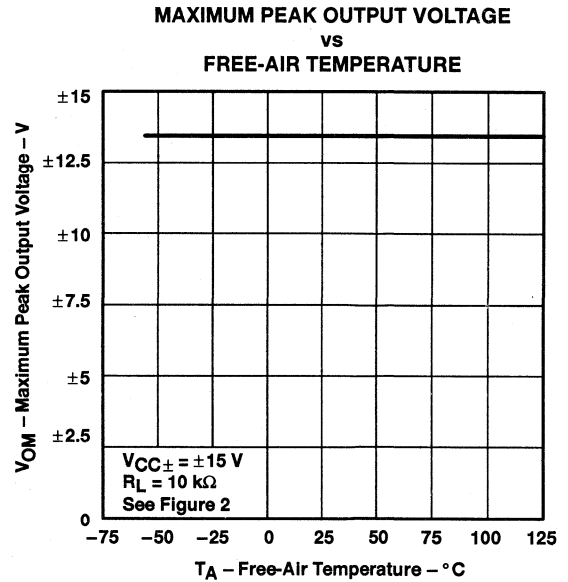


Figure 5

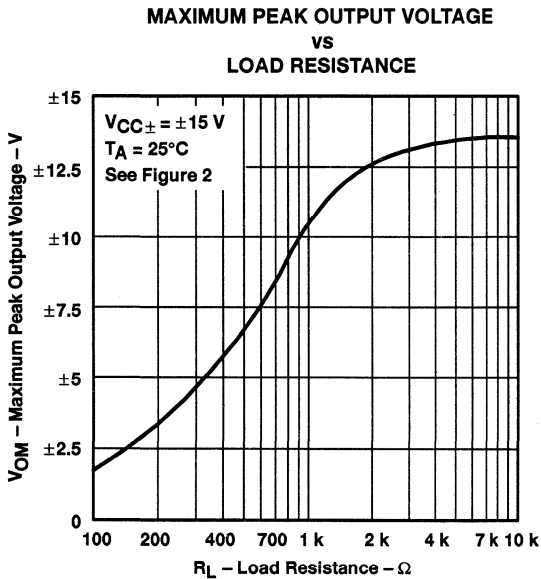


Figure 6

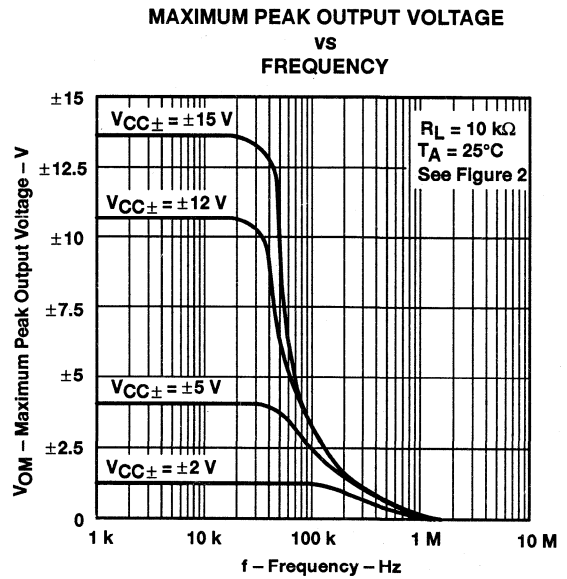


Figure 7

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

DIFFERENTIAL VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

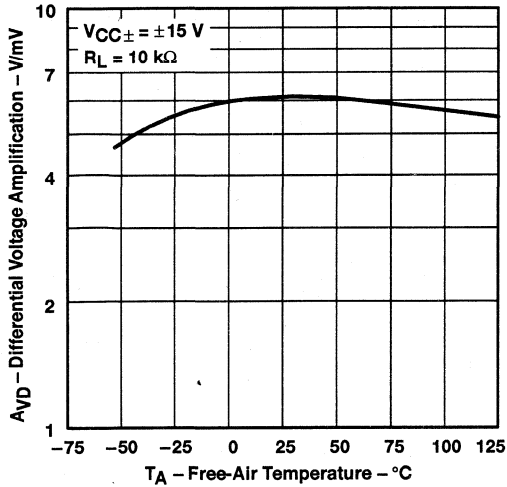


Figure 8

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY

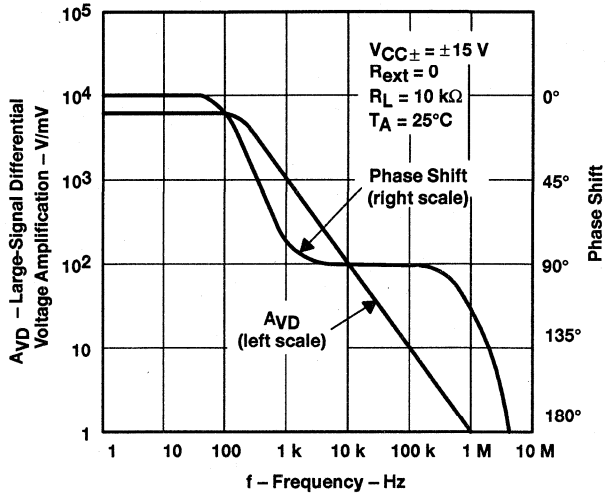


Figure 9

SUPPLY CURRENT  
 VS  
 SUPPLY VOLTAGE

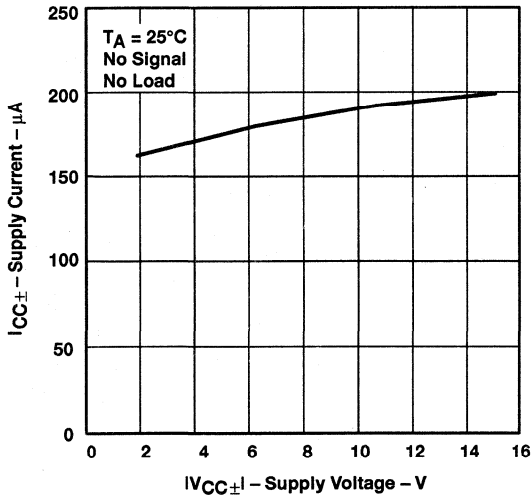


Figure 10

SUPPLY CURRENT  
 VS  
 FREE-AIR TEMPERATURE

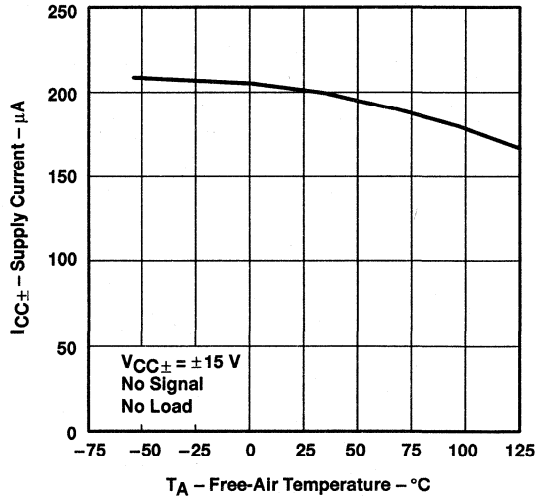


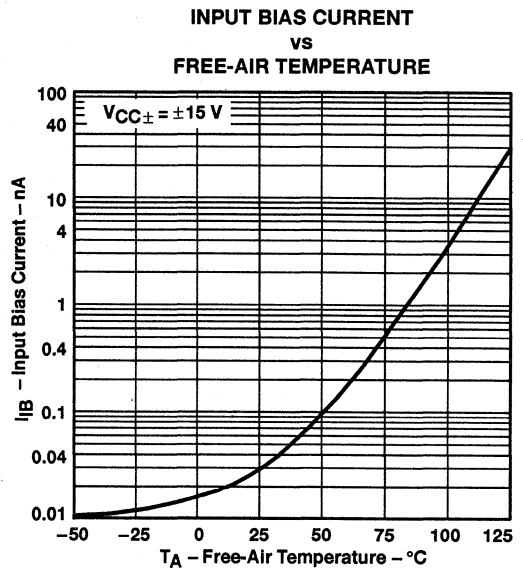
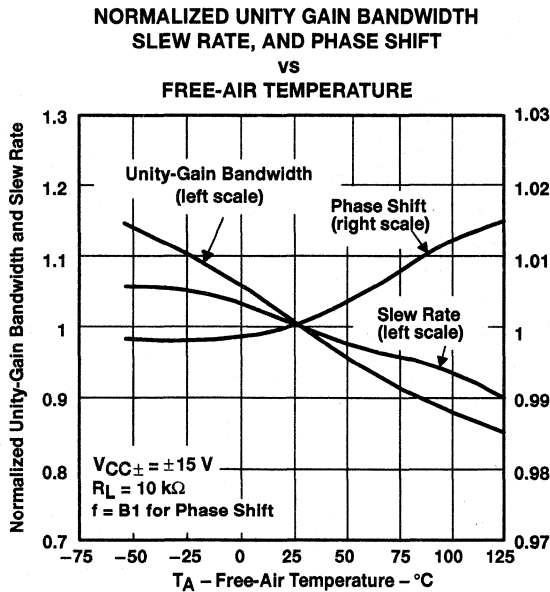
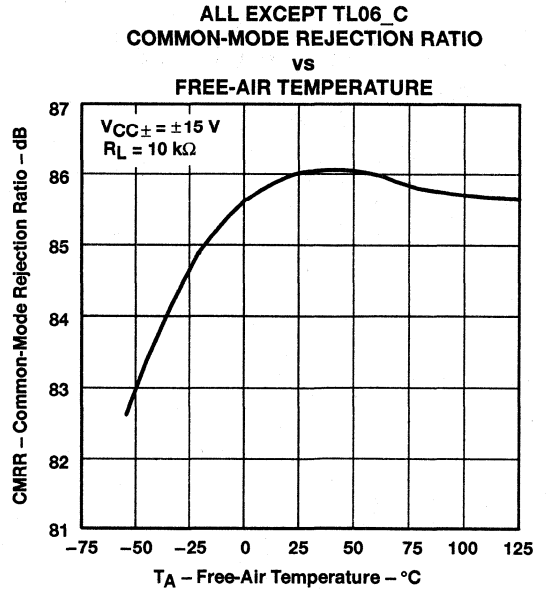
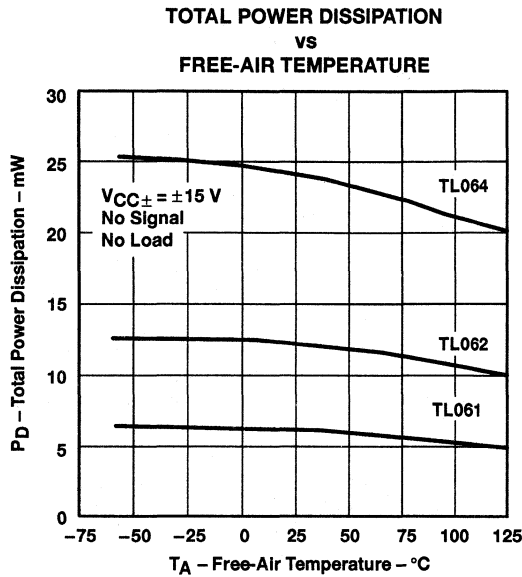
Figure 11

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
 LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS078B - NOVEMBER 1978 - REVISED AUGUST 1994

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

VOLTAGE FOLLOWER  
 LARGE SIGNAL PULSE RESPONSE

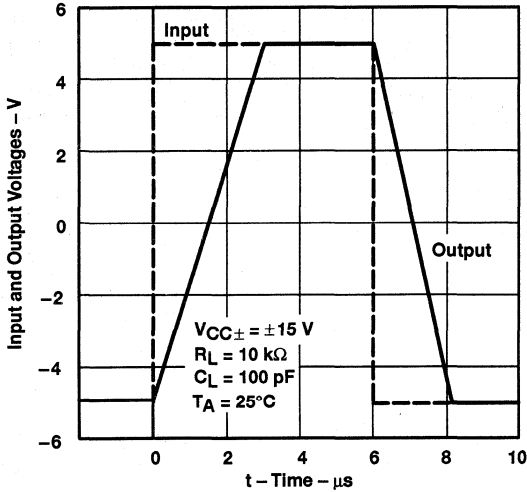


Figure 16

OUTPUT VOLTAGE  
 VS  
 ELAPSED TIME

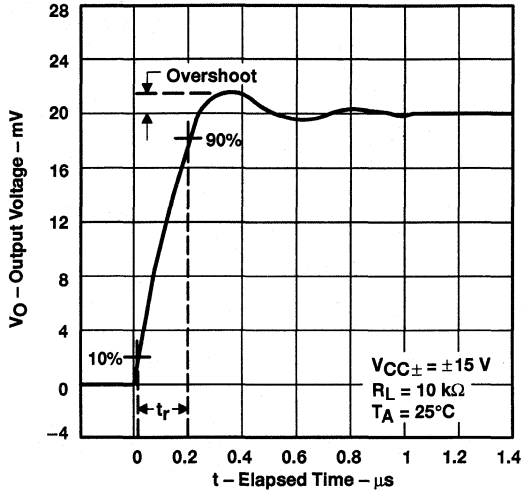


Figure 17

EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY

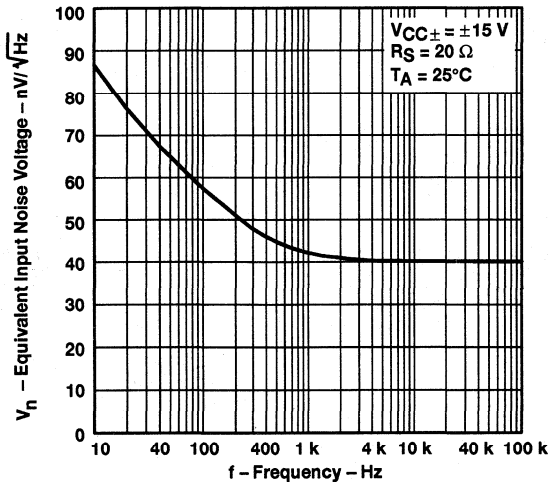


Figure 18

APPLICATION INFORMATION

Table of Application Diagrams

APPLICATION DIAGRAM	PART NUMBER	FIGURE
Instrumentation filter	TL064	19
0.5-Hz square-wave oscillator	TL061	20
High-Q notch filter	TL061	21
Audio-distribution amplifier	TL064	22
Low-level light detector preamplifier	TL061	23
AC amplifier	TL061	24
Microphone preamplifier with tone control	TL061	25
Instrumentation amplifier	TL062	26
IC preamplifier	TL062	27

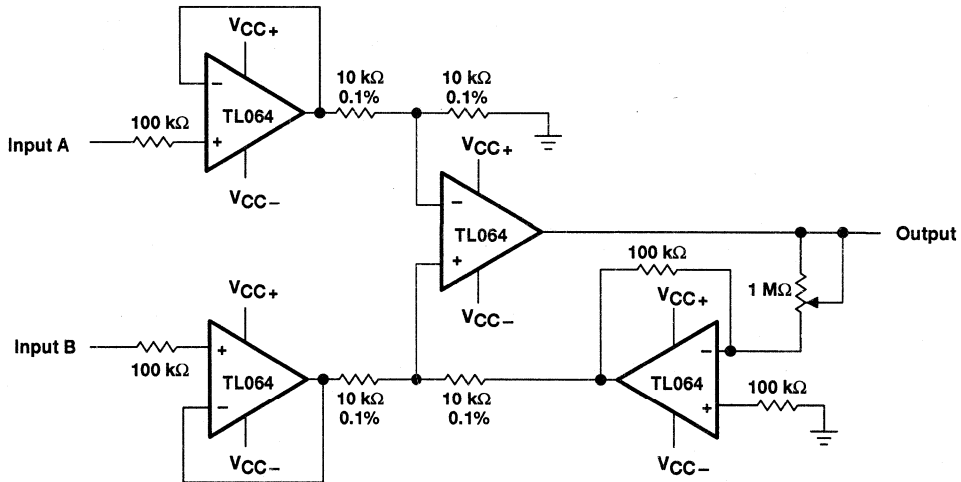


Figure 19. Instrumentation Amplifier

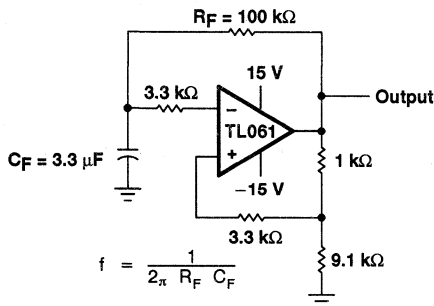


Figure 20. 0.5-Hz Square-Wave Oscillator

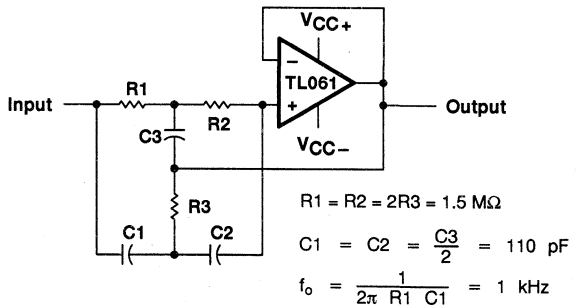
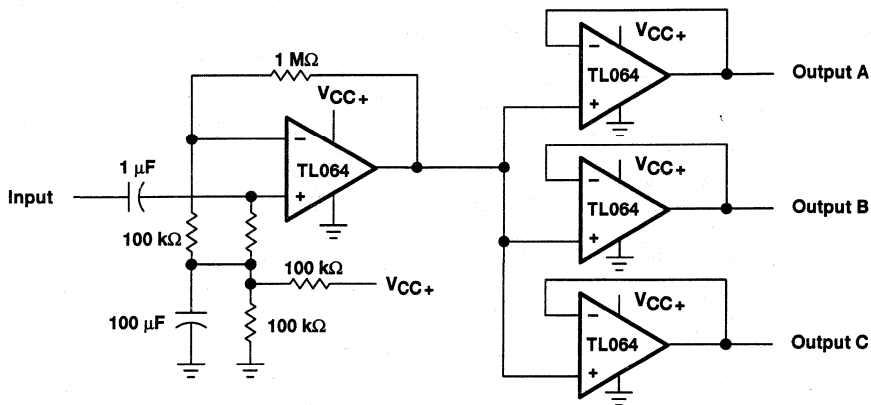


Figure 21. High-Q Notch Filter

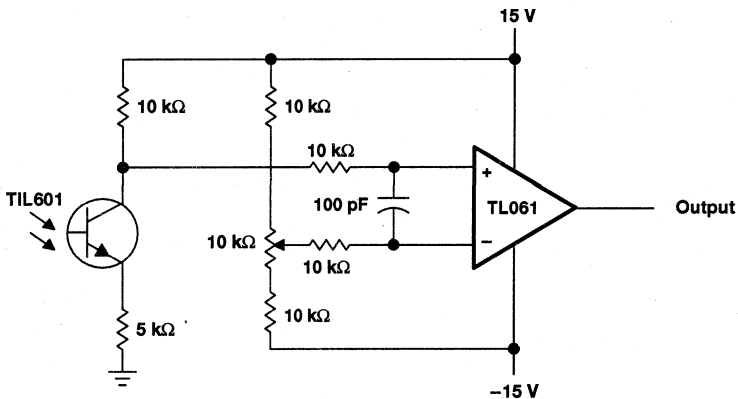
TL061, TL061A, TL061B, TL061Y, TL062, TL062A  
 TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y  
**LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B – NOVEMBER 1978 – REVISED AUGUST 1994

**APPLICATION INFORMATION**



**Figure 22. Audio-Distribution Amplifier**



**Figure 23. Low-Level Light-Detector Preampifier**



APPLICATION INFORMATION

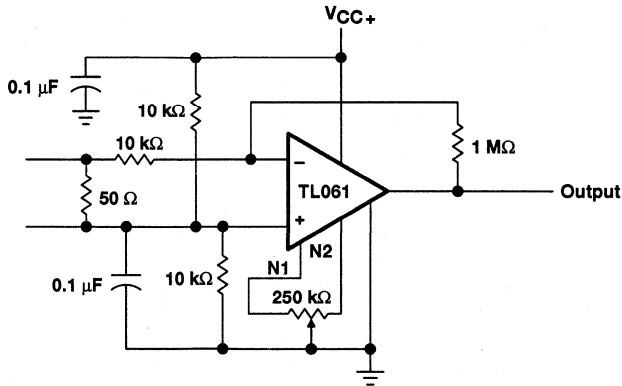


Figure 24. AC Amplifier

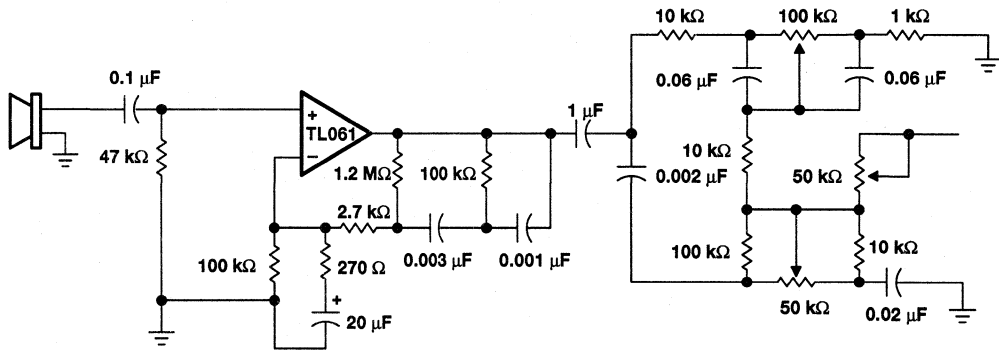


Figure 25. Microphone Preamp With Tone Control

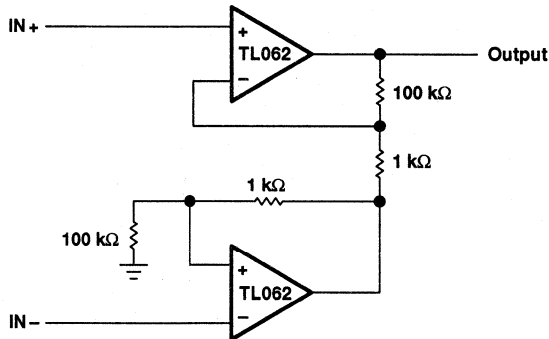


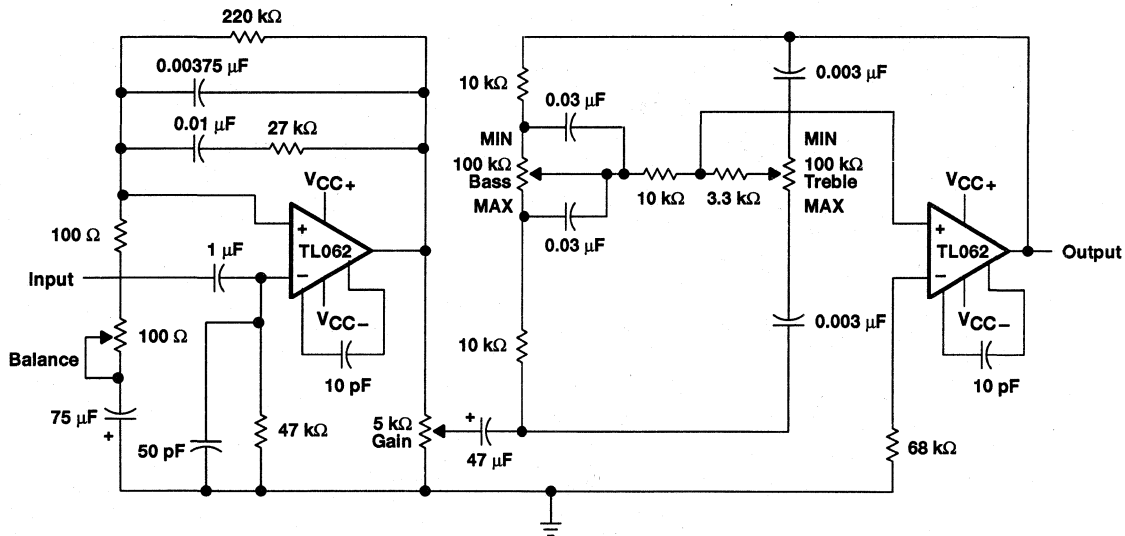
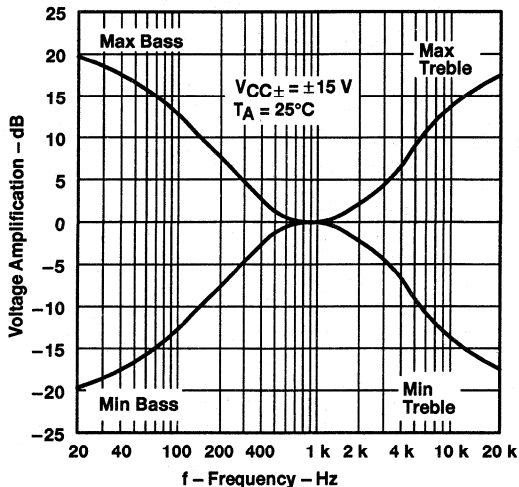
Figure 26. Instrumentation Amplifier

**TL061, TL061A, TL061B, TL061Y, TL062, TL062A,  
TL062B, TL062Y, TL064, TL064A, TL064B, TL064Y**  
**LOW-POWER JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS078B - NOVEMBER 1978 - REVISED AUGUST 1994

**APPLICATION INFORMATION**

**IC PREAMPLIFIER RESPONSE CHARACTERISTICS**



**Figure 27. IC Preamplifier**

# TL064x2 LOW-POWER JFET-INPUT OCTAL OPERATIONAL AMPLIFIER

SLOS134 – APRIL 1994

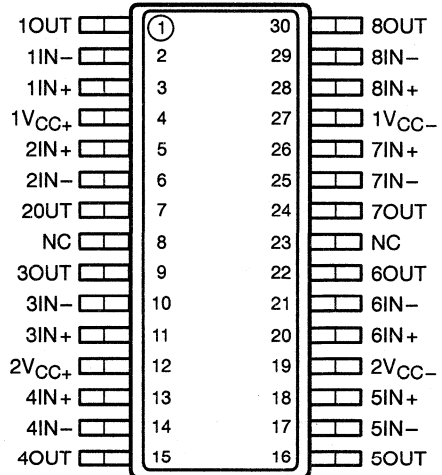
- Very Low Power Consumption
- Typical Supply Current . . . 200  $\mu$ A (Per Amplifier)
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Common-Mode Input Voltage Range Includes  $V_{CC+}$
- Output Short-Circuit Protection
- High Input Impedance . . . JFET-Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . . 3.5 V/ $\mu$ s Typ

### description

The TL064x2 JFET-input operational amplifier is designed as a low-power version of the TL084x2 amplifier. It features high input impedance, wide bandwidth, high slew rate, and low input offset and bias currents. The TL064x2 features the same terminal assignments as the TL074x2 and TL084x2. Each of these JFET-input operational amplifiers incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit.

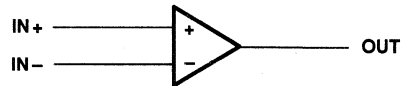
The TL064x2 is characterized for operation from 0°C to 70°C.

DB PACKAGE  
(TOP VIEW)



NC – No internal connection

### symbol (each amplifier)



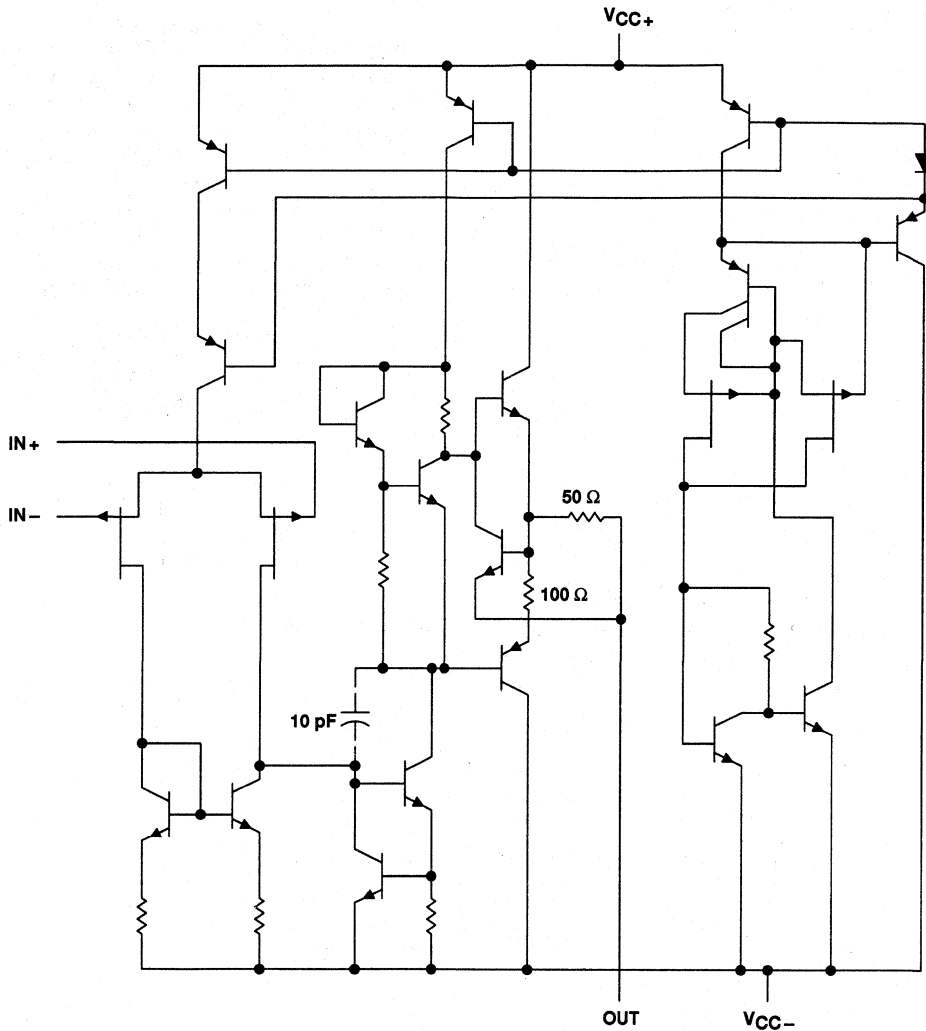
### AVAILABLE OPTION

$T_A$	$V_{IOmax}$ AT 25°C	PACKAGE
		SMALL OUTLINE (DB) <sup>†</sup>
0°C to 70°C	7 mV	TL064x2DBLE

<sup>†</sup> The DB package is only available left-end taped and reeled.

**TL064x2**  
**LOW-POWER JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**  
 SLOS134 – APRIL 1994

schematic (each amplifier)



All component values shown are nominal.

ACTUAL DEVICE COMPONENT COUNT	
Transistors	116
Resistors	60
JFET	24
Capacitors	8
Diodes	4

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1) .....	18 V
Supply voltage, $V_{CC-}$ (see Note 1) .....	-18 V
Differential input voltage, $V_{ID}$ (see Note 2) .....	$\pm 30$ V
Input voltage, $V_I$ (any input) (see Notes 1 and 3) .....	$\pm 15$ V
Duration of output short circuit to ground (see Note 4) .....	unlimited
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{OS}$ , are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  2. Differential voltages are at IN + with respect to IN -.
  3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
  4. The output can be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DB	1024 mW	8.2 mW/°C	655 mW

**TL064x2**  
**LOW-POWER JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**

SLOS134 – APRIL 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		$T_A$ ‡	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 0,$	$R_S = 50 \Omega$	25°C		3	15	mV
				Full range			20	
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	$V_O = 0,$	$R_S = 50 \Omega$	Full range		10		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current	$V_O = 0$		25°C		5	200	pA
				Full range				5
$I_{IB}$	Input bias current§	$V_O = 0$		25°C		30	400	pA
				Full range				10
$V_{ICR}$	Common-mode input voltage range			25°C	$\pm 11$	-12 to 15		V
$V_{OM}$	Maximum peak output voltage swing	$R_L = 10 \text{ k}\Omega$		25°C	$\pm 10$	$\pm 13.5$		V
				Full range	$\pm 10$			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 10 \text{ V},$	$R_L \geq 10 \text{ k}\Omega$	25°C	3	6		V/mV
				Full range	3			
$B_1$	Unity-gain bandwidth	$R_L = 10 \text{ k}\Omega,$		25°C		1		MHz
$r_1$	Input resistance			25°C		$10^{12}$		$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0,$	$R_S = 50 \Omega$	25°C	70	86		dB
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V},$ $V_O = 0,$	$R_S = 50 \Omega$	25°C	70	95		dB
$P_D$	Total power dissipation (each amplifier)	$V_O = 0,$	No load	25°C		6	7.5	mW
$I_{CC}$	Supply current (each amplifier)	$V_O = 0,$	No load	25°C		200	250	$\mu\text{A}$
$V_{O1}/V_{O2}$	Crosstalk attenuation	$A_{VD} = 100$		25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

‡ Full range is 0°C to 70°C.

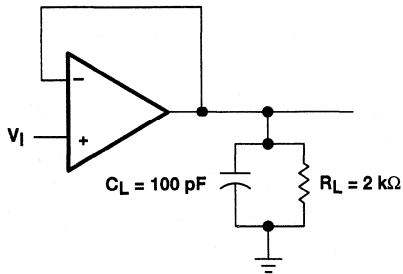
§ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 13. Pulse techniques must be used that maintain the junction temperature as close to the ambient temperature as possible.

**operating characteristics,  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ\text{C}$**

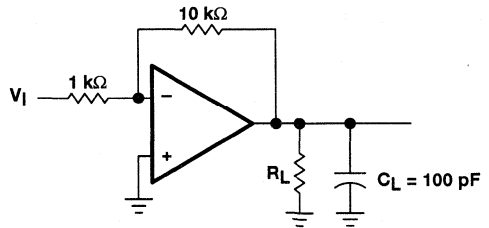
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10 \text{ mV},$ $C_L = 100 \text{ pF},$	$R_L = 10 \text{ k}\Omega,$ See Figure 1	1.5	3.5		$\text{V}/\mu\text{s}$
$t_r$	Rise time	$V_I = 20 \text{ V},$ $C_L = 100 \text{ pF},$ See Figure 1	$R_L = 10 \text{ k}\Omega,$ See Figure 1		0.2		$\mu\text{s}$
	Overshoot factor				10%		
$V_n$	Equivalent input noise voltage	$R_S = 20 \Omega,$	$f = 1 \text{ kHz}$		42		$\text{nV}/\sqrt{\text{Hz}}$



**PARAMETER MEASUREMENT INFORMATION**



**Figure 1. Unity-Gain Amplifier**



**Figure 2. Gain-of-10 Inverting Amplifier**

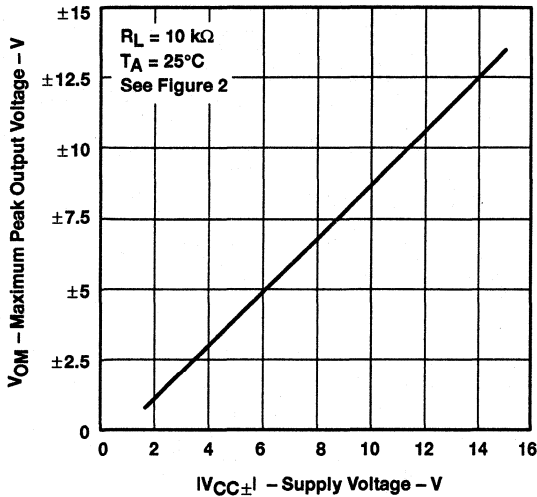
**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		FIGURE		
V <sub>OM</sub>	Maximum peak output voltage	vs Supply voltage	3	
		vs Free-air temperature	4	
		vs Load resistance	5	
		vs Frequency	6	
A <sub>VD</sub>	Differential voltage amplification	vs Free-air temperature	7	
A <sub>VD</sub>	Large-signal differential voltage amplification	vs Frequency	8	
I <sub>CC</sub>	Supply current	vs Supply voltage	9	
		vs Free-air temperature	10	
P <sub>D</sub>	Total power dissipation	vs Free-air temperature	11	
		Normalized unity-gain bandwidth	vs Free-air temperature	12
		Normalized slew rate	vs Free-air temperature	12
I <sub>IB</sub>	Input bias current	vs Free-air temperature	13	
		Pulse response	Large signal	14
V <sub>O</sub>	Output voltage	vs Time	15	
V <sub>n</sub>	Equivalent input noise voltage	vs Frequency	16	
		Normalized phase shift	vs Free-air temperature	12

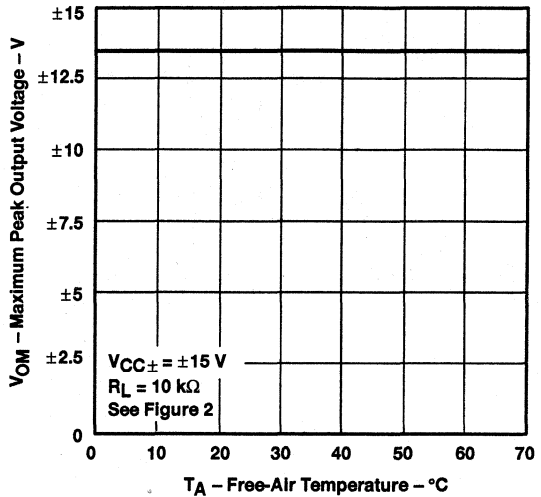
**TYPICAL CHARACTERISTICS**

**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE**



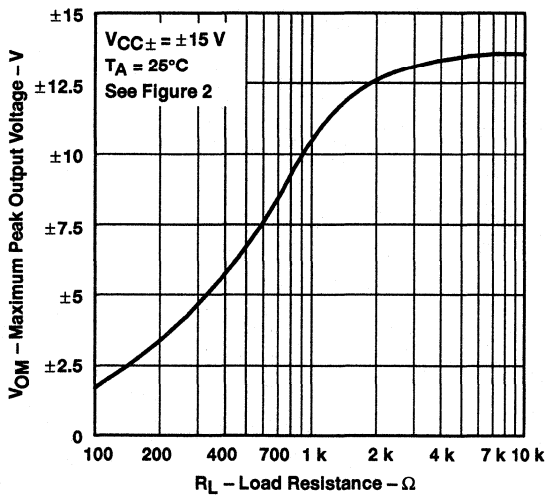
**Figure 3**

**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE**



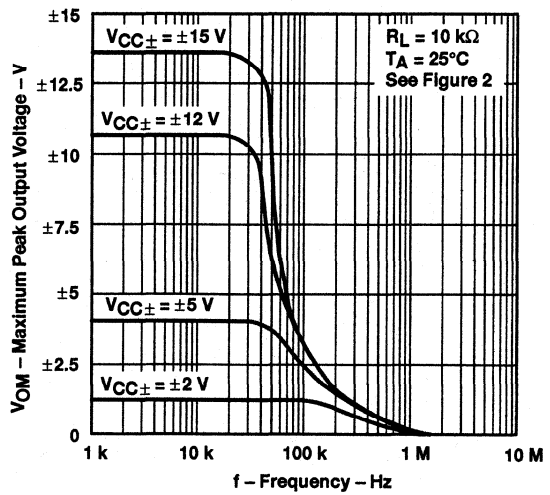
**Figure 4**

**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 LOAD RESISTANCE**



**Figure 5**

**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY**

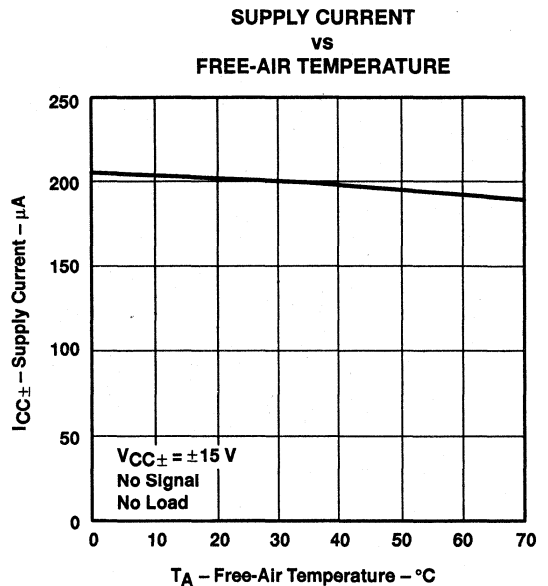
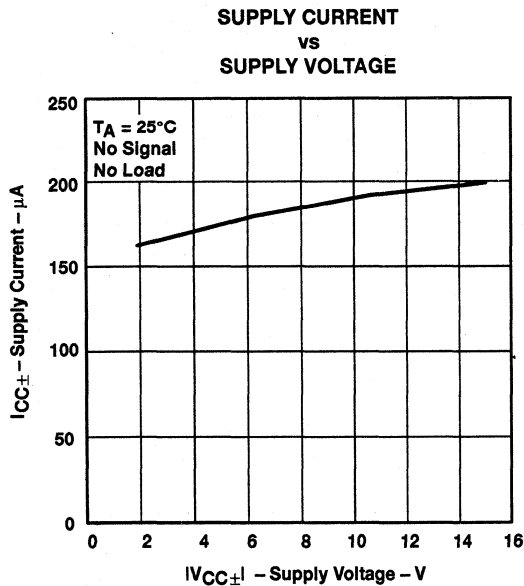
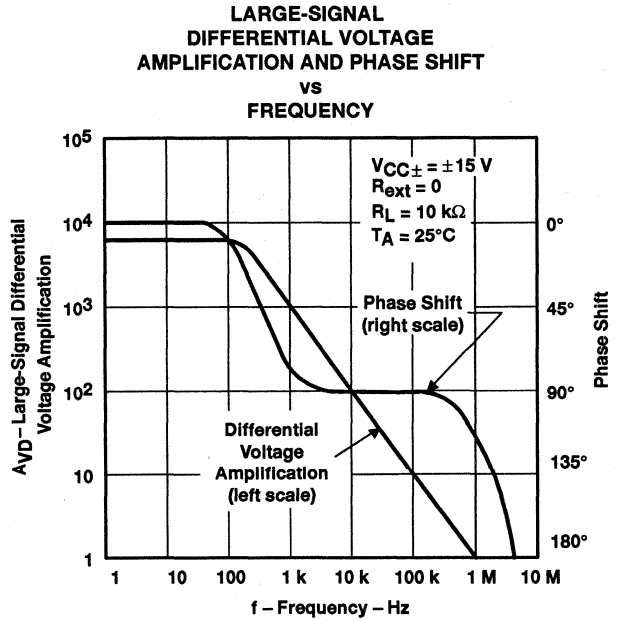
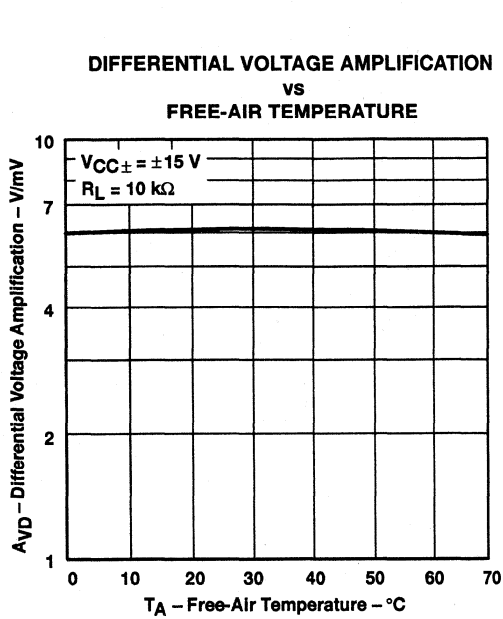


**Figure 6**





**TYPICAL CHARACTERISTICS**

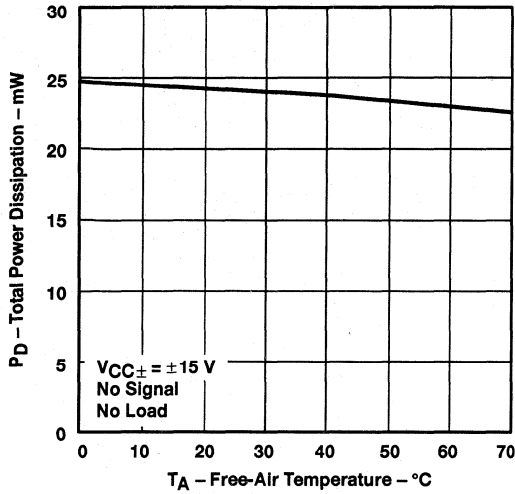


**TL064x2**  
**LOW-POWER JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**

SLOS134 – APRIL 1994

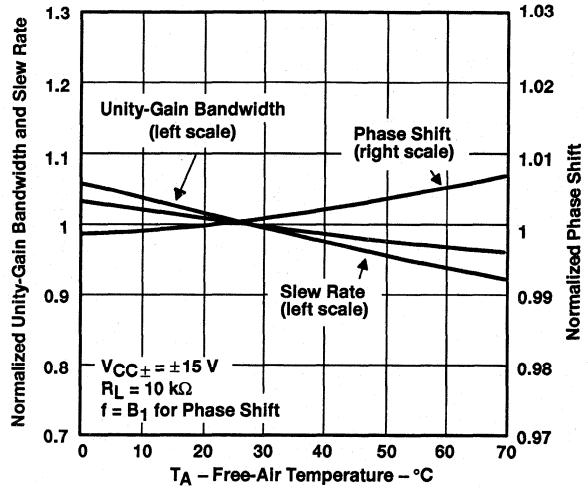
**TYPICAL CHARACTERISTICS**

**TOTAL POWER DISSIPATION**  
**vs**  
**FREE-AIR TEMPERATURE**



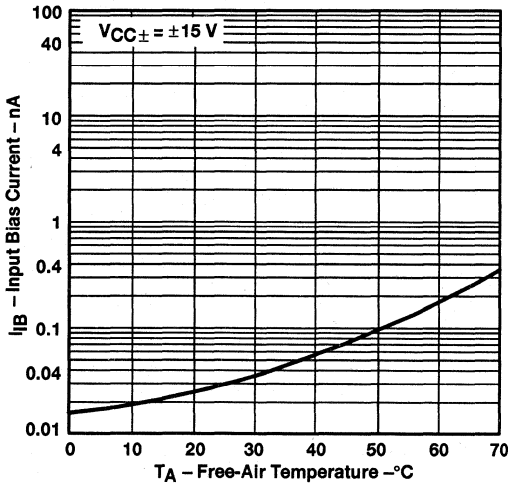
**Figure 11**

**NORMALIZED UNITY-GAIN BANDWIDTH,**  
**NORMALIZED SLEW RATE, AND**  
**NORMALIZED PHASE SHIFT**  
**vs**  
**FREE-AIR TEMPERATURE**



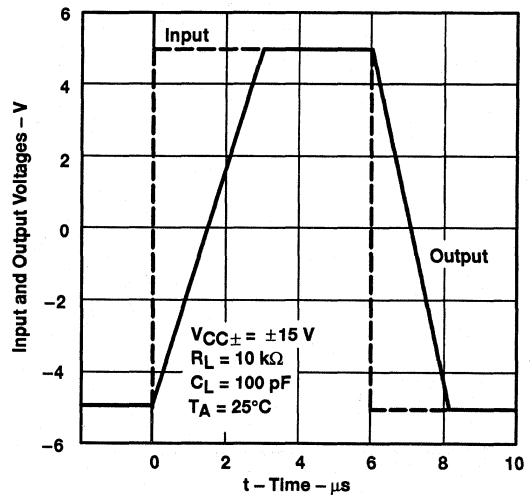
**Figure 12**

**INPUT BIAS CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 13**

**VOLTAGE FOLLOWER**  
**LARGE SIGNAL PULSE RESPONSE**



**Figure 14**



**TYPICAL CHARACTERISTICS**

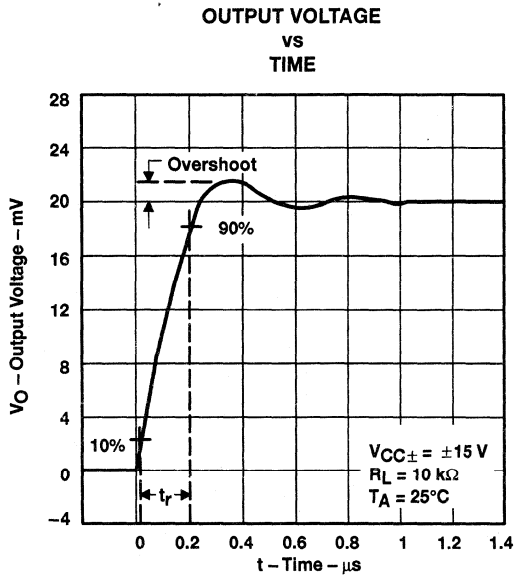


Figure 15

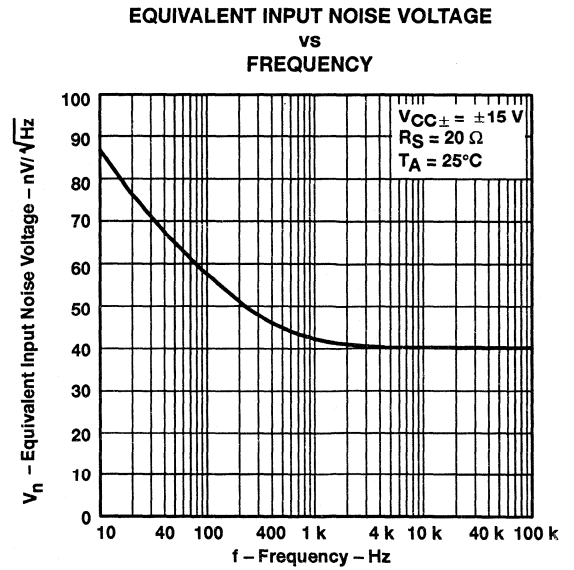


Figure 16



# TL070 LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIER

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

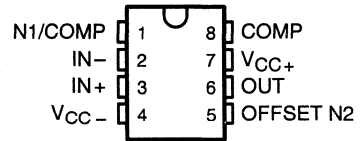
- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion  
0.003% Typ
- Low Noise  
 $V_n = 18 \text{ nV}/\sqrt{\text{Hz}}$  Typ at  $f = 1 \text{ kHz}$
- High Input Impedance . . . JFET Input Stage
- Common-Mode Input Voltage Range  
Includes  $V_{CC+}$
- Latch-Up-Free Operation
- High Slew Rate . . .  $13 \text{ V}/\mu\text{s}$  Typ

## description

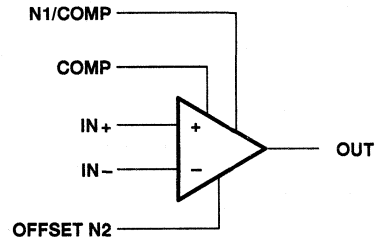
The JFET-input TL070 operational amplifier is designed as the low-noise version of the TL080 amplifier with low input bias and offset currents and fast slew rate. The low harmonic distortion and low noise make the TL070 ideally suited for high-fidelity and audio preamplifier applications. This amplifier features JFET inputs (for high input impedance) coupled with bipolar output stages integrated on a single monolithic chip.

The TL070C device is characterized for operation from 0°C to 70°C. The TL070I device is characterized for operation from -40°C to 85°C. The TL070M device is characterized for operation from -55°C to 125°C.

D, P, OR PW PACKAGE  
(TOP VIEW)



symbol



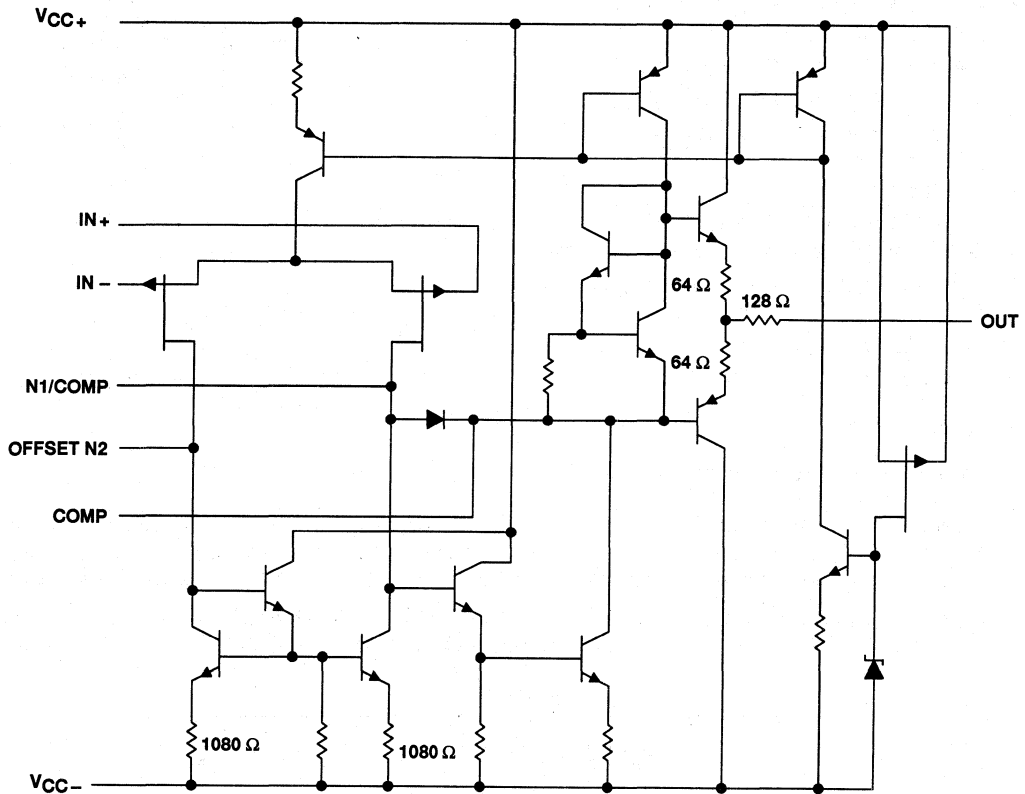
AVAILABLE OPTIONS

$T_A$	$V_{IOmax}$ AT 25°C	PACKAGE		
		SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)
0°C to 70°C	10 mV	TL070CD	TL070CP	TL070CPW
-40°C to 85°C	10 mV	TL070ID	TL070IP	—
-55°C to 125°C	10 mV	TL070MD	TL070MP	—

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**schematic**



All component values shown are nominal.

COMPONENT COUNT †	
Transistors	13
Diodes	2
Resistors	10
epi-FET	1
JFET	2

† Includes all bias and trim circuitry

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$	-18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$ V
Input voltage, $V_I$ (see Notes 1 and 3)	$\pm 15$ V
Duration of short-circuit current (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
  3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
  4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	145 mW
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	200 mW
PW	525 mW	4.2 mW/°C	70°C	336 mW	N/A	N/A



**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL070C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0, R_S = 50 \Omega$	25°C		3	10	mV
		Full range			13	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_O = 0, R_S = 50 \Omega$	Full range		18		$\mu V/^\circ C$
$I_{IO}$ Input offset current	$V_O = 0$	25°C		5	100	pA
		Full range			10	nA
$I_{IB}$ Input bias current‡	$V_O = 0$	25°C		65	200	pA
		Full range			7	nA
$V_{ICR}$ Common-mode input voltage range		25°C	$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10 k\Omega$	25°C	$\pm 12$	$\pm 13.5$		V
	$R_L \geq 10 k\Omega$	Full range	$\pm 12$			
	$R_L \geq 2 k\Omega$		$\pm 10$			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10$ V, $R_L \geq 2 k\Omega$	25°C	25	200		V/mV
		Full range	15			
$B_1$ Unity-gain bandwidth		25°C		3		MHz
$r_i$ Input resistance		25°C		$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0, R_S = 50 \Omega$	25°C	70	100		dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9$ V to $\pm 15$ V, $V_O = 0, R_S = 50 \Omega$	25°C	70	100		dB
$I_{CC}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.4	2.5	mA
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$	25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for  $T_A$  is 0°C to 70°C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 5. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.





**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL070I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0, R_S = 50 \Omega$	25°C		3	10	mV
		Full range			13	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_O = 0, R_S = 50 \Omega$	Full range		18		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current	$V_O = 0$	25°C		5	100	pA
		Full range			10	nA
$I_{IB}$ Input bias current‡	$V_O = 0$	25°C		65	200	pA
		Full range			20	nA
$V_{ICR}$ Common-mode input voltage range		25°C	$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10 \text{ k}\Omega$	25°C	$\pm 12$	$\pm 13.5$		V
	$R_L \geq 10 \text{ k}\Omega$	Full range	$\pm 12$			
	$R_L \geq 2 \text{ k}\Omega$		$\pm 10$			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10$ V, $R_L \geq 2 \text{ k}\Omega$	25°C	25	200		V/mV
		Full range	15			
$B_1$ Unity-gain bandwidth		25°C		3		MHz
$r_i$ Input resistance		25°C		$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0, R_S = 50 \Omega$	25°C	70	100		dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9$ V to $\pm 15$ V, $V_O = 0, R_S = 50 \Omega$	25°C	70	100		dB
$I_{CC}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.4	2.5	mA
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$	25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for  $T_A$  is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 5. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL070M			UNIT
			MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C		3	10	mV
		Full range			13	
αV <sub>IO</sub> Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	Full range		18		μV/°C
I <sub>IO</sub> Input offset current	V <sub>O</sub> = 0	25°C		5	100	pA
		Full range			20	nA
I <sub>IB</sub> Input bias current‡	V <sub>O</sub> = 0	25°C		65	200	pA
		Full range			50	nA
V <sub>ICR</sub> Common-mode input voltage range		25°C	±11	-12 to 15		V
V <sub>OM</sub> Maximum peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	±12	±13.5		V
	R <sub>L</sub> ≥ 10 kΩ	Full range	±12			
	R <sub>L</sub> ≥ 2 kΩ		±10			
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	25°C	25	200		V/mV
		Full range	15			
B <sub>1</sub> Unity-gain bandwidth		25°C		3		MHz
r <sub>i</sub> Input resistance		25°C		10 <sup>12</sup>		Ω
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70	100		dB
k <sub>SVR</sub> Supply voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC</sub> = ±9 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	70	100		dB
I <sub>CC</sub> Supply current	V <sub>O</sub> = 0, No load	25°C	1.4	2.5		mA
V <sub>O1</sub> /V <sub>O2</sub> Crosstalk attenuation	A <sub>VD</sub> = 100	25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T<sub>A</sub> is -55°C to 125°C.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 5. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.



operating characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10\text{ V}$ , $C_L = 100\text{ pF}$ , See Figure 1	8	13		$\text{V}/\mu\text{s}$
$t_r$	Rise time overshoot factor	$V_I = 20\text{ mV}$ , $C_L = 100\text{ pF}$ , See Figure 1		0.1		$\mu\text{s}$
				20		%
$V_n$	Equivalent input noise voltage	$R_S = 20\ \Omega$	$f = 1\text{ kHz}$	18		$\text{nV}/\sqrt{\text{Hz}}$
			$f = 10\text{ Hz to } 10\text{ kHz}$	4		$\mu\text{V}$
$I_n$	Equivalent input noise current	$R_S = 20\ \Omega$ , $f = 1\text{ kHz}$	0.01			$\text{pA}/\sqrt{\text{Hz}}$
THD	Total harmonic distortion	$V_{O(\text{rms})} = 10\text{ V}$ , $R_L \geq 2\text{ k}\Omega$ , $f = 1\text{ kHz}$	0.003			%

PARAMETER MEASUREMENT INFORMATION

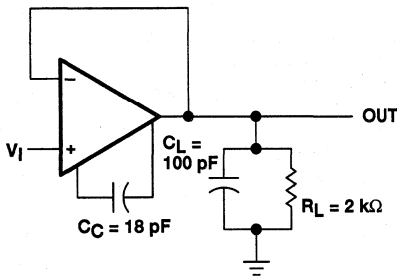


Figure 1. Unity-Gain Amplifier

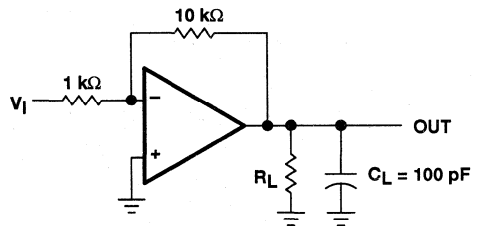


Figure 2. Gain-of-10 Inverting Amplifier

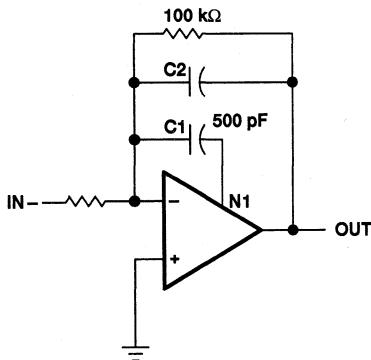


Figure 3. Feed-Forward Compensation

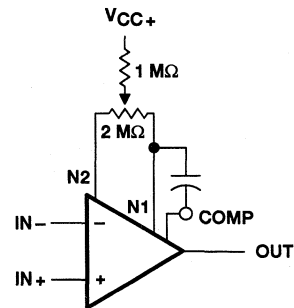


Figure 4. Input Offset Voltage Null Circuit

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$I_{IB}$	Input bias current	vs Free-air temperature	5
$V_{OM}$	Maximum output voltage	vs Frequency	6, 7, 8
		vs Free-air temperature	9
		vs Load resistance	10
		vs Supply voltage	11
$A_{VD}$	Large-signal differential voltage amplification	vs Free-air temperature	12
		vs Frequency	14
$A_{VD}$	Differential voltage amplification	vs Frequency	13
	Phase shift	vs Frequency	14
	Normalized unity-gain bandwidth	vs Free-air temperature	15
	Normalized phase shift	vs Free-air temperature	15
$CMRR$	Common-mode rejection ratio	vs Free-air temperature	16
$I_{CC}$	Supply current	vs Supply voltage	17
		vs Free-air temperature	18
$P_D$	Total power dissipation	vs Free-air temperature	19
		Normalized slew rate	vs Free-air temperature
$V_n$	Equivalent input noise voltage	vs Frequency	21
THD	Total harmonic distortion	vs Frequency	22
		Large-signal pulse response	vs Time
$V_O$	Output voltage	vs Elapsed time	24



TYPICAL CHARACTERISTICS†

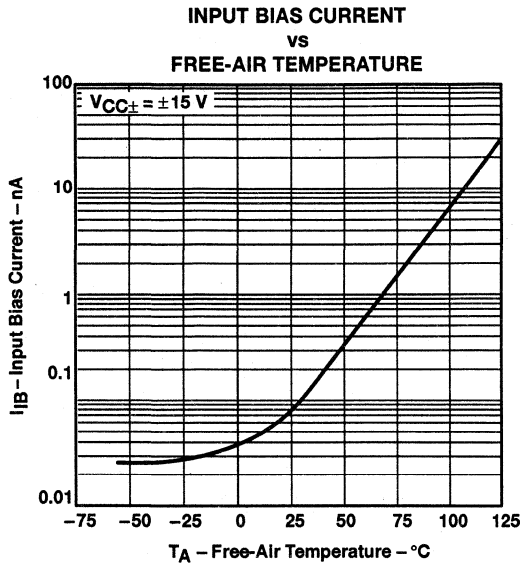


Figure 5

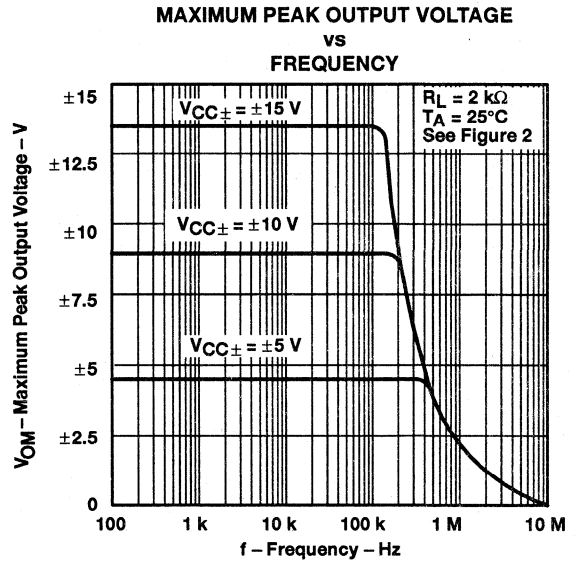


Figure 6

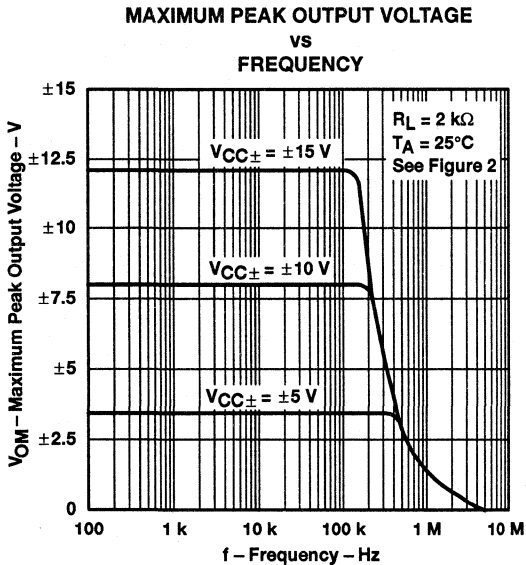


Figure 7

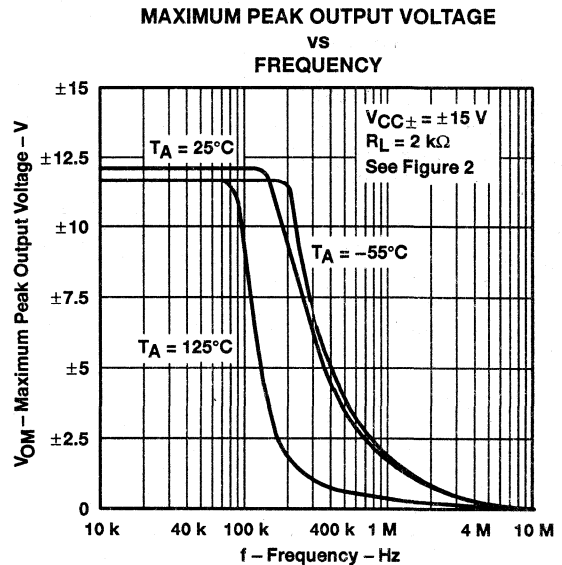


Figure 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used.

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A - NOVEMBER 1993 - REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

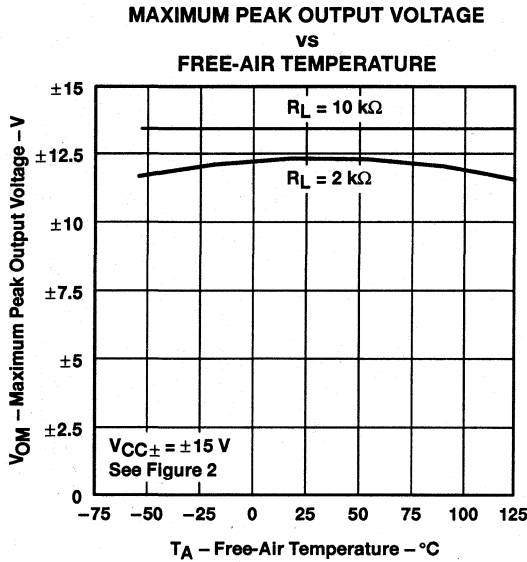


Figure 9

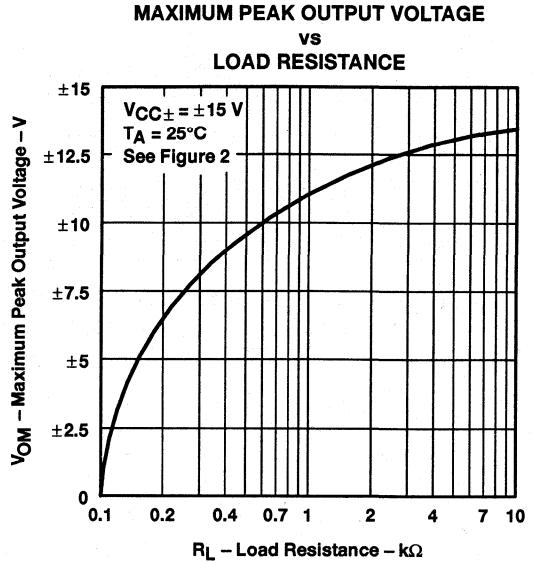


Figure 10

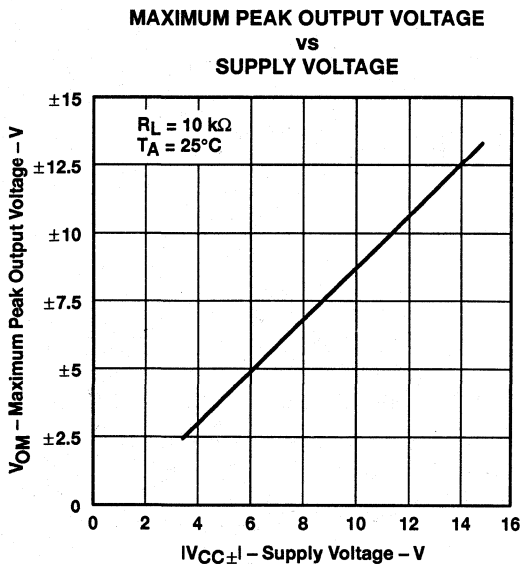


Figure 11

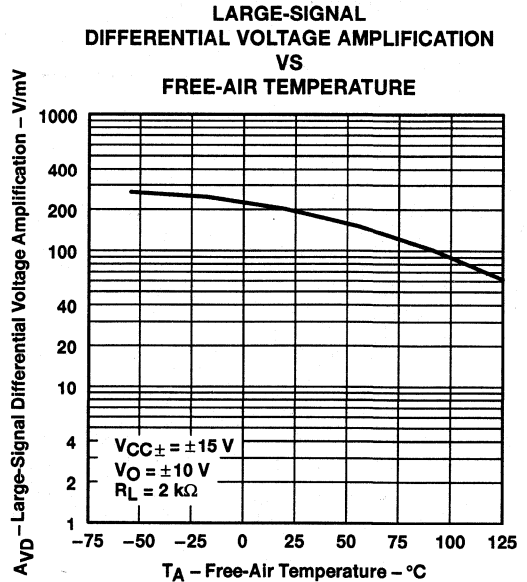


Figure 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used.

TYPICAL CHARACTERISTICS†

DIFFERENTIAL VOLTAGE AMPLIFICATION  
 VS  
 FREQUENCY WITH FEED-FORWARD  
 COMPENSATION

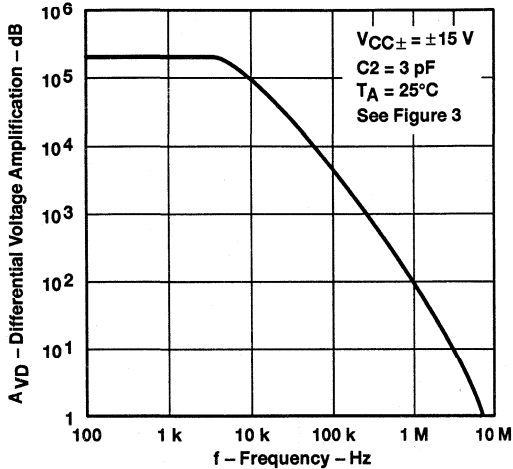


Figure 13

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 VS  
 FREQUENCY

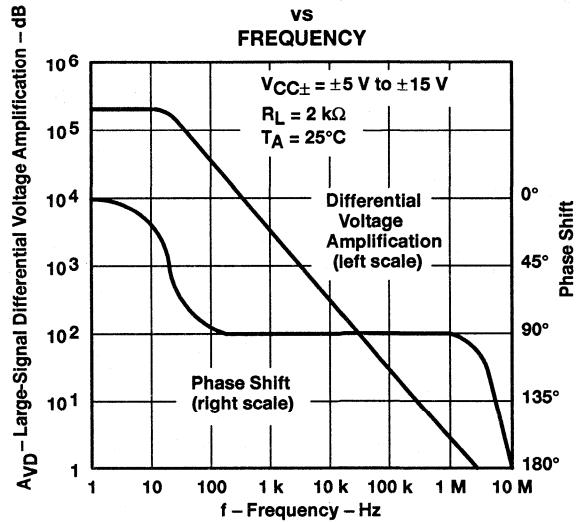


Figure 14

NORMALIZED UNITY-GAIN BANDWIDTH  
 AND PHASE SHIFT  
 VS  
 FREE-AIR TEMPERATURE

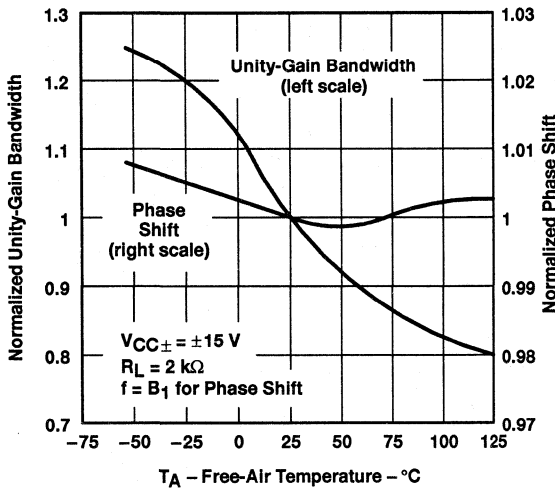


Figure 15

COMMON-MODE REJECTION RATIO  
 VS  
 FREE-AIR TEMPERATURE

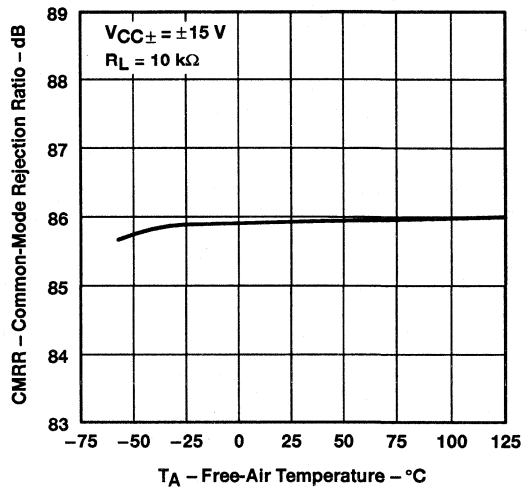


Figure 16

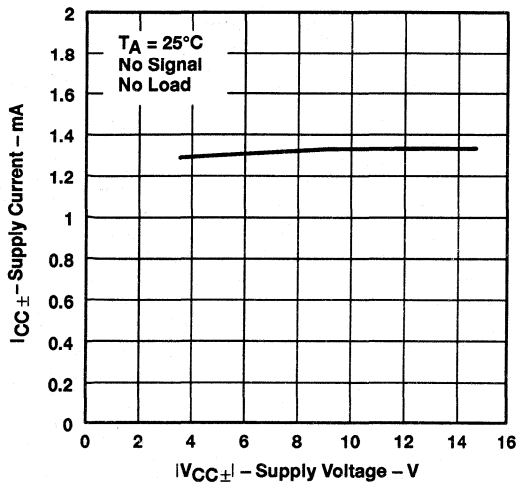
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used.

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A – NOVEMBER 1993 – REVISED AUGUST 1994

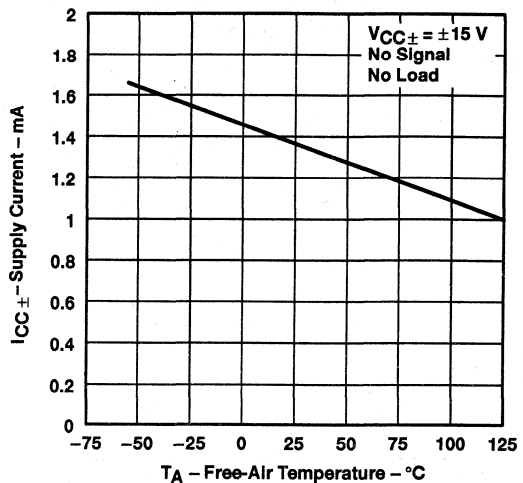
**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



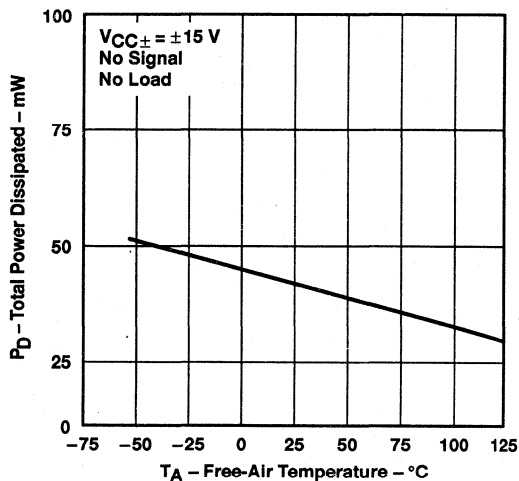
**Figure 17**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



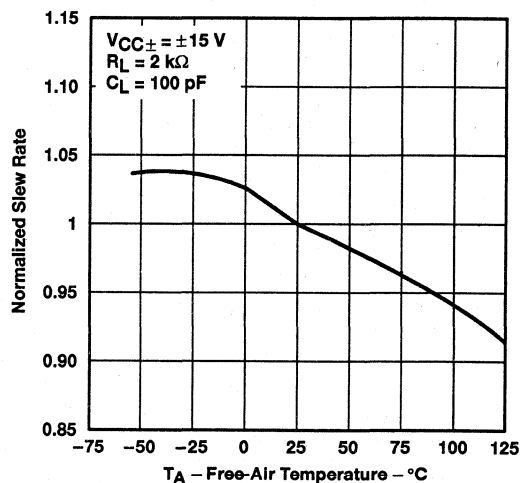
**Figure 18**

**TOTAL POWER DISSIPATED**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 19**

**NORMALIZED SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 20**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices. An 18-pF compensation capacitor is used.





TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

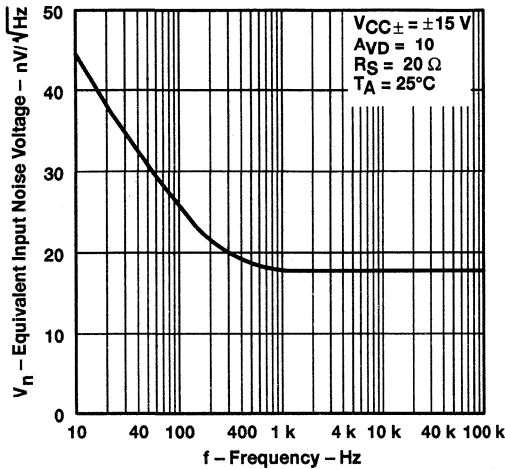


Figure 21

TOTAL HARMONIC DISTORTION  
 vs  
 FREQUENCY

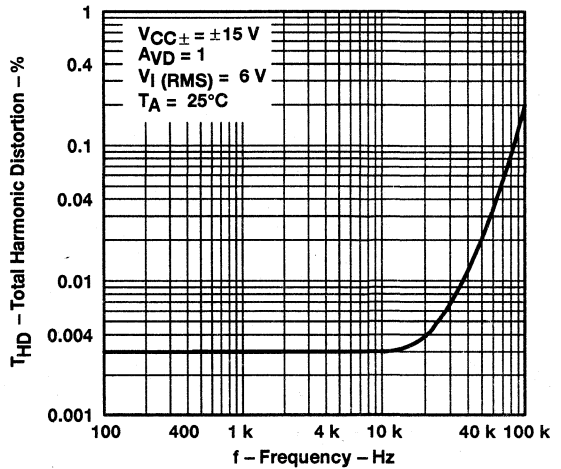


Figure 22

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL PULSE RESPONSE

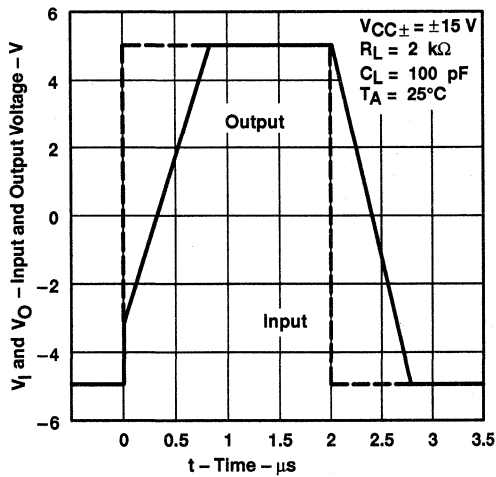


Figure 23

OUTPUT VOLTAGE  
 vs  
 ELAPSED TIME

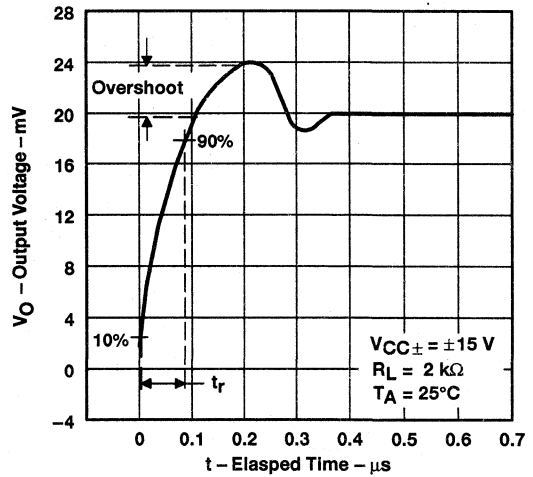


Figure 24

**TL070**  
**LOW-NOISE JFET-INPUT**  
**OPERATIONAL AMPLIFIER**

SLOS121A - NOVEMBER 1993 - REVISED AUGUST 1994

**APPLICATION INFORMATION**

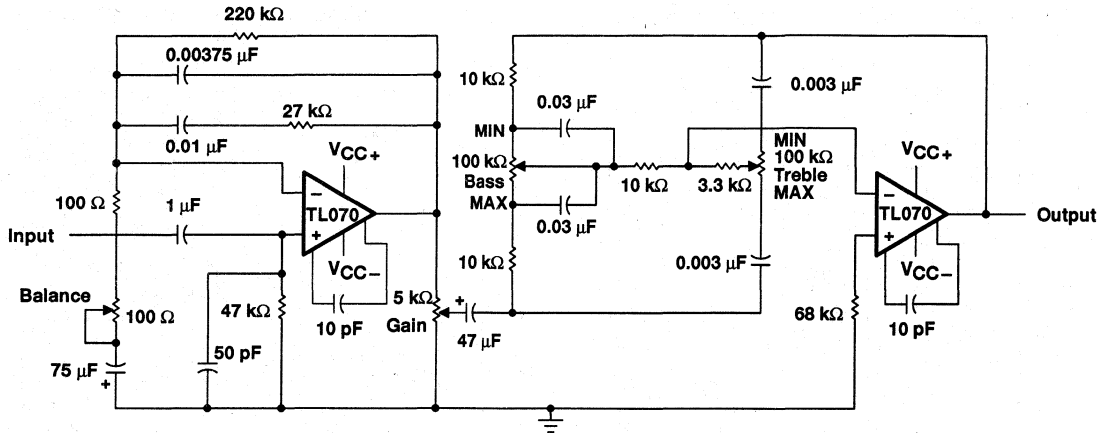


Figure 25. IC Preamplifier

**IC PREAMPLIFIER**  
**RESPONSE CHARACTERISTICS**

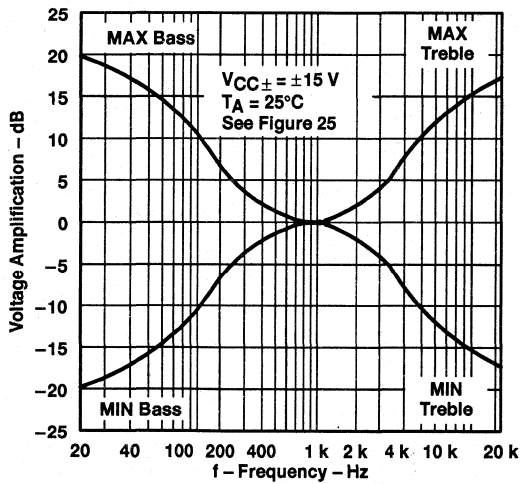


Figure 26



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL071, TL071A, TL071B, TL072**  
**TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**15 DEVICES COVER COMMERCIAL, INDUSTRIAL,  
AND MILITARY TEMPERATURE RANGES**

- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion  
0.003% Typ
- Low Noise  
 $V_n = 18 \text{ nV}/\sqrt{\text{Hz}}$  Typ at  $f = 1 \text{ kHz}$
- High Input Impedance . . . JFET Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . .  $13 \text{ V}/\mu\text{s}$  Typ
- Common-Mode Input Voltage Range  
Includes  $V_{CC+}$

**description**

The JFET-input operational amplifiers in the TL07\_ series are designed as low-noise versions of the TL08\_ series amplifiers with low input bias and offset currents and fast slew rate. The low harmonic distortion and low noise make the TL07\_ series ideally suited for high-fidelity and audio preamplifier applications. Each amplifier features JFET inputs (for high input impedance) coupled with bipolar output stages integrated on a single monolithic chip.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

**AVAILABLE OPTIONS**

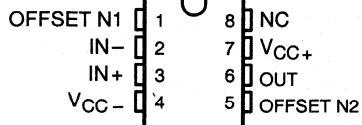
TA	V <sub>IO</sub> max AT 25°C	PACKAGE							
		SMALL OUTLINE (D)†	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	PLASTIC DIP (N)	PLASTIC DIP (P)	TSSOP PACKAGE (PW)	FLAT PACKAGE (W)
0°C to 70°C	10 mV 6 mV 3 mV	TL071CD TL071ACD TL071BCD	—	—	—	—	TL071CP TL071ACP TL071BCP	TL071CPWLE — —	—
	10 mV 6 mV 3 mV	TL072CD TL072ACD TL072BCD	—	—	—	—	TL072CP TL072ACP TL072BCP	TL072CPWLE — —	—
	10 mV 6 mV 3 mV	TL074CD TL074ACD TL074BCD	—	—	—	TL074CN TL074ACN TL074BCN	— — —	TL074CPWLE — —	—
-40°C to 85°C	6 mV	TL071ID TL072ID TL074ID	—	—	—	— — TL074IN	TL071IP TL072P —	— — —	—
-55°C to 125°C	6 mV 6 mV 9 mV	—	TL071MFK TL072MFK TL074MFK	— — TL074MJ	TL071MJG TL072MJG —	— — —	— — —	— — —	— — TL074MW

† The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL071CDR). The PW package is only available left-ended taped and reeled (e.g., TL072CPWLE).

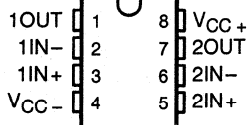
# TL071, TL071A, TL071B, TL072 TL072A, TL072B, TL074, TL074A, TL074B LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS080C - SEPTEMBER 1978 - REVISED AUGUST 1994

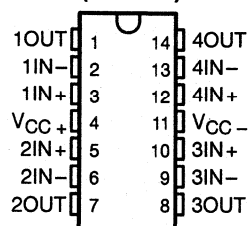
TL071, TL071A, TL071B  
D, JG, P, OR PW PACKAGE  
(TOP VIEW)



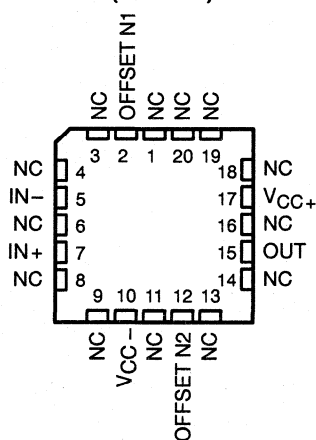
TL072, TL072A, TL072B  
D, JG, P, OR PW PACKAGE  
(TOP VIEW)



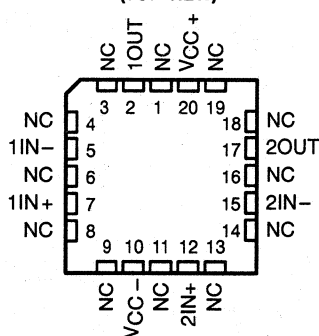
TL074, TL074A, TL074B  
D, J, N, OR PW PACKAGE  
TL074...W PACKAGE  
(TOP VIEW)



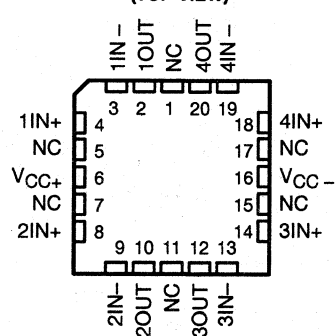
TL071  
FK PACKAGE  
(TOP VIEW)



TL072  
FK PACKAGE  
(TOP VIEW)

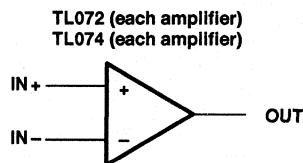
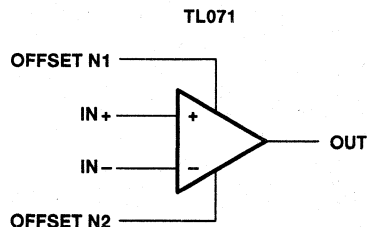


TL074  
FK PACKAGE  
(TOP VIEW)



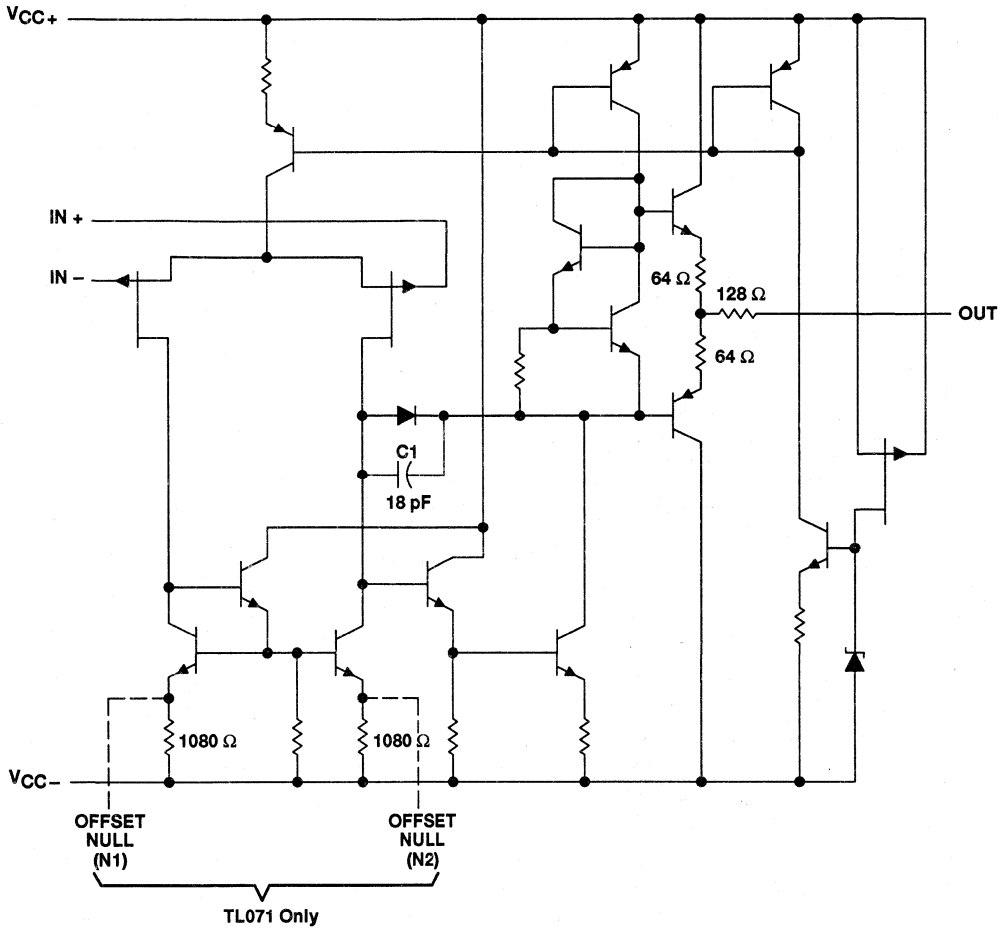
NC - No internal connection

## symbols



**TL071, TL071A, TL071B, TL072**  
**TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**  
SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**schematic (each amplifier)**



COMPONENT COUNT†			
COMPONENT TYPE	TL071	TL072	TL074
Resistors	11	22	44
Transistors	14	28	56
JFET	2	4	6
Diodes	1	2	4
Capacitors	1	2	4
epi-FET	1	2	4

† Includes bias and trim circuitry

**TL071, TL071A, TL071B, TL072  
TL072A, TL072B, TL074, TL074A, TL074B  
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1) .....	18 V
Supply voltage, $V_{CC-}$ (see Note 1) .....	-18 V
Differential input voltage, $V_{ID}$ (see Note 2) .....	$\pm 30$ V
Input voltage, $V_I$ (see Notes 1 and 3) .....	$\pm 15$ V
Duration of output short-circuit (see Note 4) .....	unlimited
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix .....	0°C to 70°C
I suffix .....	-40°C to 85°C
M suffix .....	-55°C to 125°C
Storage temperature range .....	-65°C to 150°C
Case temperature for 60 seconds: FK package .....	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: J, JG, or W package .....	300°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, P, or PW package .....	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
  3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
  4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D (8 pin)	680 mW	5.8 mW/°C	33°C	464 mW	377 mW	N/A
D (14 pin)	680 mW	7.6 mW/°C	60°C	608 mW	494 mW	N/A
FK	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
J	680 mW	11.0 mW/°C	88°C	680 mW	680 mW	275 mW
JG	680 mW	8.4 mW/°C	69°C	672 mW	546 mW	210 mW
N	680 mW	9.2 mW/°C	76°C	680 mW	598 mW	N/A
P	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	N/A
PW (8 pin)	525 mW	4.2 mW/°C	70°C	525 mW	N/A	N/A
PW (14 pin)	700 mW	5.6 mW/°C	70°C	700 mW	N/A	N/A
W	680 mW	8.0 mW/°C	65°C	640 mW	520 mW	200 mW

**TL071, TL071A, TL071B, TL072**  
**TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITION <sup>†</sup>	T <sub>A</sub> <sup>‡</sup>	TL071C TL072C TL074C			TL071AC TL072AC TL074AC			TL071BC TL072BC TL074BC			TL071I TL072I TL074I			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	3	10	3	6	2	3	6	2	3	6	mV		
		Full range		13		7.5		5		8		8			
α <sub>VIO</sub>	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	Full range	18		18		18		18		18		μV/°C		
I <sub>IO</sub>	V <sub>O</sub> = 0	25°C	5	100	5	100	5	100	5	100	5	100	pA		
		Full range		10		2		2		2		2	nA		
I <sub>IB</sub>	V <sub>O</sub> = 0	25°C	65	200	65	200	65	200	65	200	65	200	pA		
		Full range		7		7		7		7		20	nA		
V <sub>ICR</sub>	Common-mode input voltage range	25°C	-12 to 15		-12 to 15		-12 to 15		-12 to 15		-12 to 15		V		
		25°C	±11		±11		±11		±11		±11				
		Full range		±12		±12		±12		±12		±12			
		Full range		±10		±10		±10		±10		±10			
V <sub>OM</sub>	Maximum peak output voltage swing	25°C	R <sub>L</sub> = 10 kΩ	±12 to ±13.5	R <sub>L</sub> = 10 kΩ	±12 to ±13.5	R <sub>L</sub> = 10 kΩ	±12 to ±13.5	R <sub>L</sub> = 10 kΩ	±12 to ±13.5	R <sub>L</sub> = 10 kΩ	±12 to ±13.5	V		
		25°C	R <sub>L</sub> ≥ 2 kΩ	±12 to ±13.5	R <sub>L</sub> ≥ 2 kΩ	±12 to ±13.5	R <sub>L</sub> ≥ 2 kΩ	±12 to ±13.5	R <sub>L</sub> ≥ 2 kΩ	±12 to ±13.5	R <sub>L</sub> ≥ 2 kΩ	±12 to ±13.5			
		Full range		±10		±10		±10		±10		±10			
A <sub>VD</sub>	Large-signal differential voltage amplification	25°C	V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	25	200	25	200	25	200	25	200	25	200	V/mV	
		25°C		15		25		25		25		25			
		Full range		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>			
B <sub>1</sub>	Unity-gain bandwidth	25°C		3		3		3		3		3	MHz		
r <sub>i</sub>	Input resistance	25°C		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>		10 <sup>12</sup>	Ω		
CMRR	Common-mode rejection ratio	25°C	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	70	100	75	100	75	100	75	100	75	100	dB	
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	25°C	V <sub>CC</sub> = ±9 V to ±15 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	70	100	80	100	80	100	80	100	80	100	dB	
I <sub>CC</sub>	Supply current (each amplifier)	25°C	V <sub>O</sub> = 0, No load	1.4	2.5	1.4	2.5	1.4	2.5	1.4	2.5	1.4	2.5	mA	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	25°C	A <sub>VD</sub> = 100	120		120		120		120		120	dB		

<sup>†</sup> All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

<sup>‡</sup> Full range is T<sub>A</sub> = 0°C to 70°C for TL07\_C, TL07\_AC, TL07\_BC and is T<sub>A</sub> = -40°C to 85°C for TL07\_I.

<sup>§</sup> Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 4. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**TL071, TL071A, TL071B, TL072  
TL072A, TL072B, TL074, TL074A, TL074B  
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITION†	$T_A$ ‡	TL071M TL072M			TL074M			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0, R_S = 50 \Omega$	25°C	3 6			3 9			mV
		Full range	9			15			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_O = 0, R_S = 50 \Omega$	Full range	18			18			$\mu V/^\circ C$
$I_{IO}$ Input offset current	$V_O = 0$	25°C	5 100			5 100			pA
		Full range	20			20			nA
$I_{IB}$ Input bias current‡	$V_O = 0$	25°C	65 200			65 200			pA
			50			50			nA
$V_{ICR}$ Common-mode input voltage range		25°C	±11 to 15			±11 to 15			V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10 k\Omega$	25°C	±12 ±13.5			±12 ±13.5			V
	$R_L \geq 10 k\Omega$	Full range	±12			±12			
	$R_L \geq 2 k\Omega$		±10			±10			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10$ V, $R_L \geq 2 k\Omega$	25°C	35 200			35 200			V/mV
			15			15			
$B_1$ Unity-gain bandwidth	$T_A = 25^\circ C$		3			3			MHz
$r_i$ Input resistance	$T_A = 25^\circ C$		$10^{12}$			$10^{12}$			$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0, R_S = 50 \Omega$	25°C	80	86		80	86		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9$ V to $\pm 15$ V, $V_O = 0, R_S = 50 \Omega$	25°C	80	86		80	86		dB
$I_{CC}$ Supply current (each amplifier)	$V_O = 0, \text{ No load}$	25°C	1.4 2.5			1.4 2.5			mA
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$	25°C	120			120			dB

† Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 4. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

‡ All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range is  $T_A = -55^\circ C$  to  $125^\circ C$ .





**TL071, TL071A, TL071B, TL072**  
**TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

operating characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TL07xM			ALL OTHERS			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_I = 10\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 2\text{ k}\Omega$ , See Figure 1	5	13		8	13		$\text{V}/\mu\text{s}$
$t_r$	Rise time overshoot factor $V_I = 20\text{ mV}$ , $C_L = 100\text{ pF}$ , $R_L = 2\text{ k}\Omega$ , See Figure 1	0.1			0.1			$\mu\text{s}$
		20%			20%			
$V_n$	Equivalent input noise voltage $R_S = 20\ \Omega$	$f = 1\text{ kHz}$			18			$\text{nV}/\sqrt{\text{Hz}}$
		$f = 10\text{ Hz to } 10\text{ kHz}$			4			$\mu\text{V}$
$I_n$	Equivalent input noise current $R_S = 20\ \Omega$ , $f = 1\text{ kHz}$	0.01			0.01			$\text{pA}/\sqrt{\text{Hz}}$
THD	Total harmonic distortion $V_O(\text{RMS}) = 10\text{ V}$ , $R_L \geq 2\text{ k}\Omega$ , $R_S \leq 1\text{ k}\Omega$ , $f = 1\text{ kHz}$	0.003%			0.003%			

**PARAMETER MEASUREMENT INFORMATION**

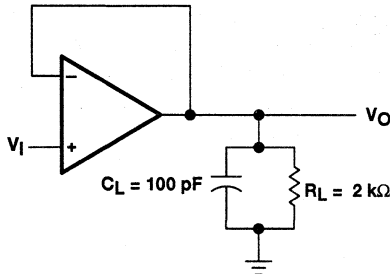


Figure 1. Unity-Gain Amplifier

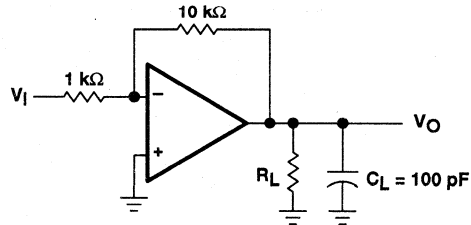


Figure 2. Gain-of-10 Inverting Amplifier

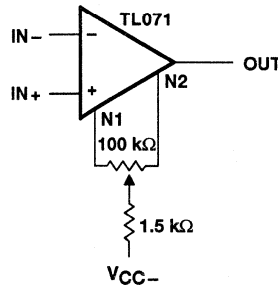


Figure 3. Input Offset Voltage Null Circuit

**TL071, TL071A, TL071B, TL072  
 TL072A, TL072B, TL074, TL074A, TL074B  
 LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**  
 SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$I_B$	Input bias current	vs Free-air temperature	4
$V_{OM}$	Maximum output voltage	vs Frequency	5, 6, 7
		vs Free-air temperature	8
		vs Load resistance	9
		vs Supply voltage	10
$A_{VD}$	Large-signal differential voltage amplification	vs Free-air temperature	11
		vs Frequency	12
	Phase shift	vs Frequency	12
	Normalized unity-gain bandwidth	vs Free-air temperature	13
	Normalized phase shift	vs Free-air temperature	13
$CMRR$	Common-mode rejection ratio	vs Free-air temperature	14
$I_{CC}$	Supply current	vs Supply voltage	15
		vs Free-air temperature	16
$P_D$	Total power dissipation	vs Free-air temperature	17
		Normalized slew rate	vs Free-air temperature
$V_n$	Equivalent input noise voltage	vs Frequency	19
THD	Total harmonic distortion	vs Frequency	20
		Large-signal pulse response	vs Time
$V_O$	Output voltage	vs Time	22

**TL071, TL071A, TL071B, TL072  
TL072A, TL072B, TL074, TL074A, TL074B**  
**LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**  
SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

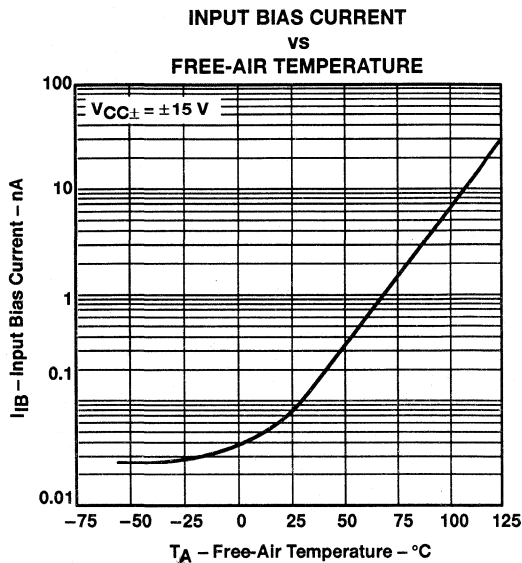


Figure 4

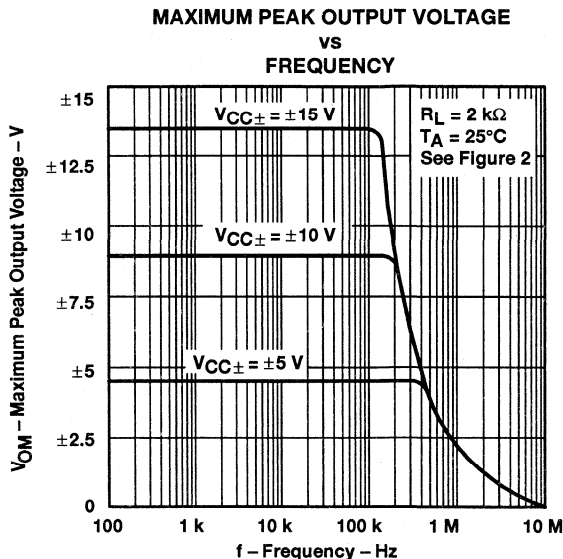


Figure 5

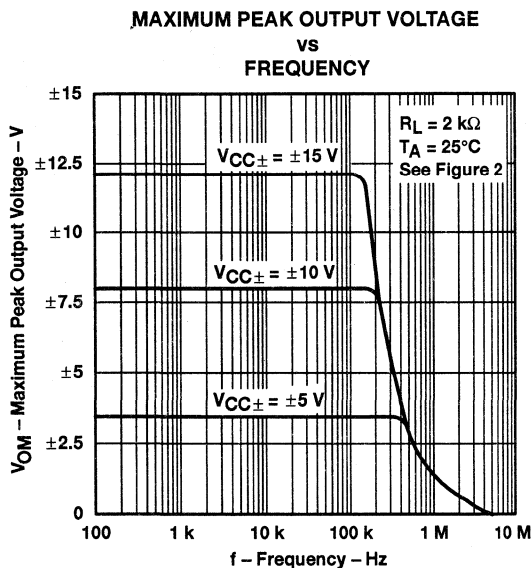


Figure 6

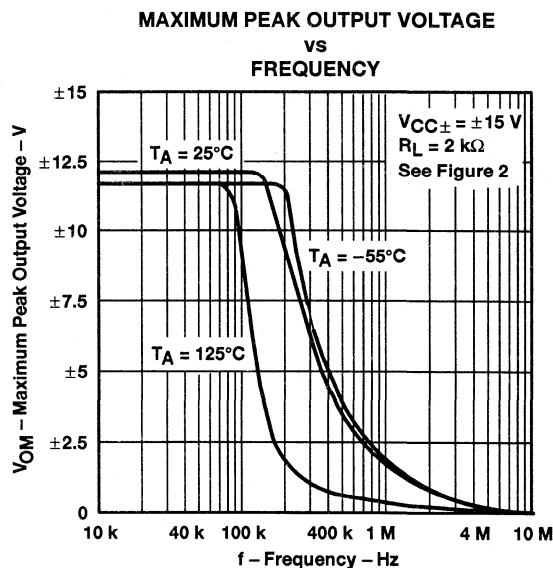


Figure 7

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

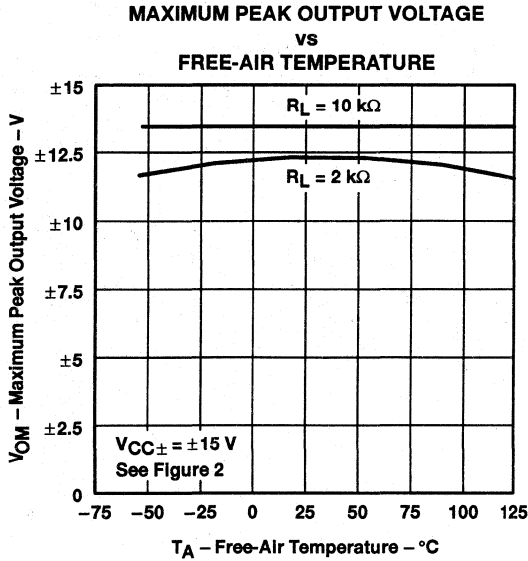


Figure 8

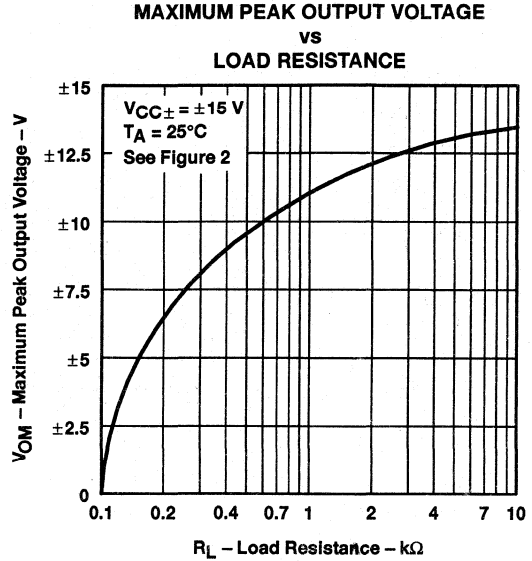


Figure 9

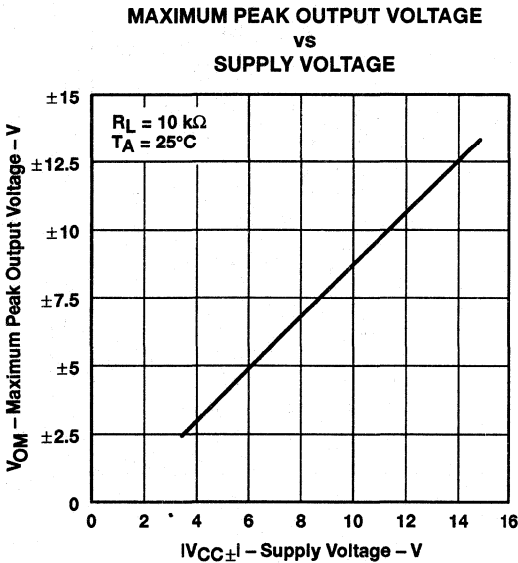


Figure 10

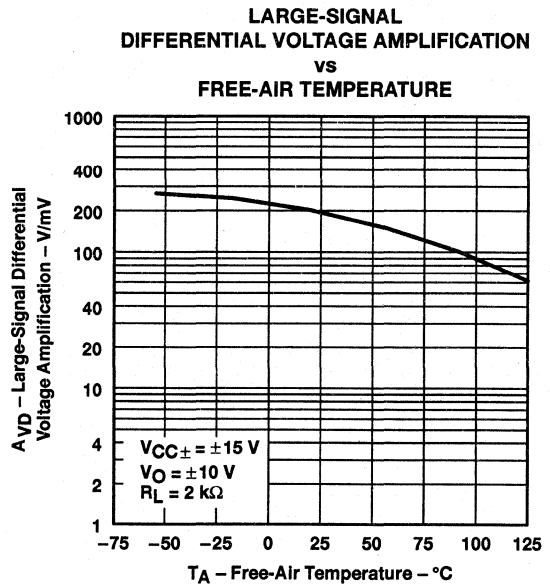


Figure 11

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 VS  
 FREQUENCY**

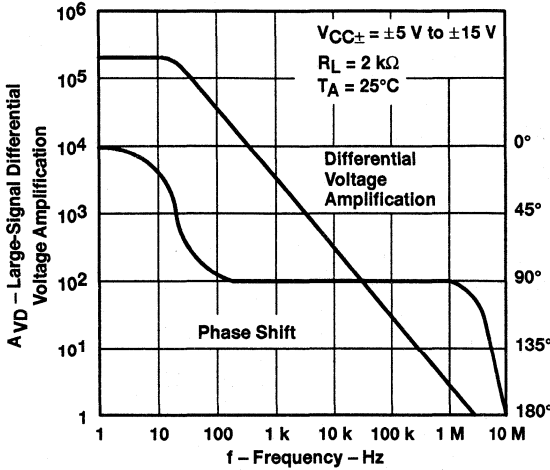


Figure 12

**NORMALIZED UNITY-GAIN BANDWIDTH  
 AND PHASE SHIFT  
 VS  
 FREE-AIR TEMPERATURE**

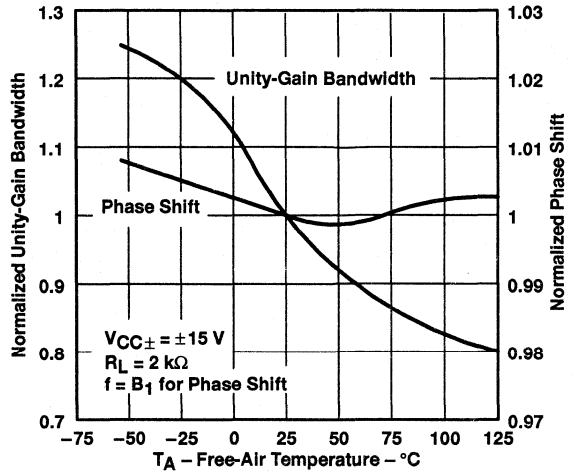


Figure 13

**COMMON-MODE REJECTION RATIO  
 VS  
 FREE-AIR TEMPERATURE**

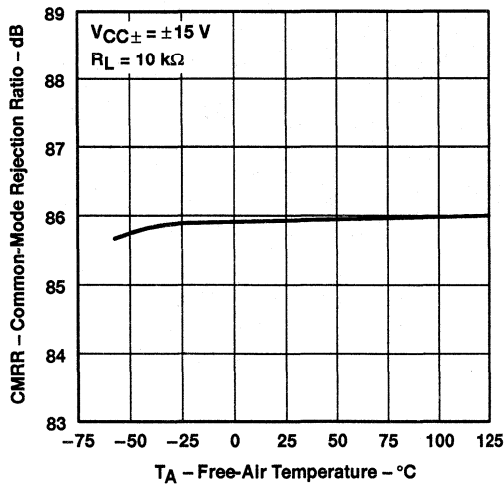


Figure 14

**SUPPLY CURRENT PER AMPLIFIER  
 VS  
 SUPPLY VOLTAGE**

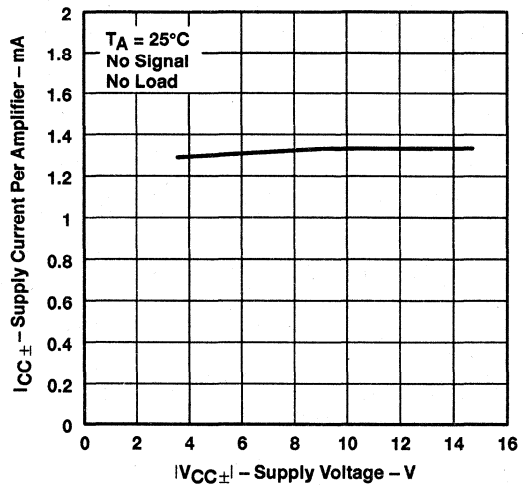


Figure 15

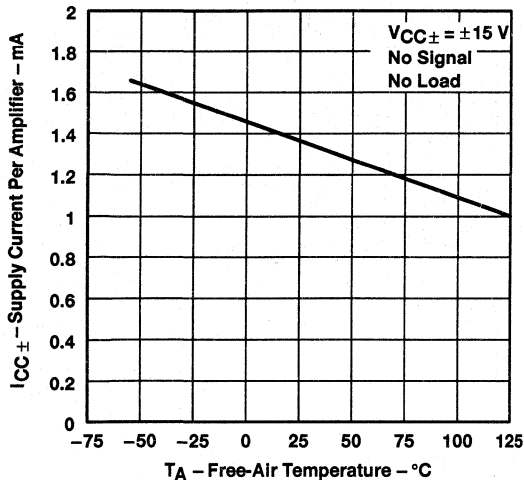
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TL071, TL071A, TL071B, TL072  
TL072A, TL072B, TL074, TL074A, TL074B  
LOW-NOISE JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS080C – SEPTEMBER 1978 – REVISED AUGUST 1994

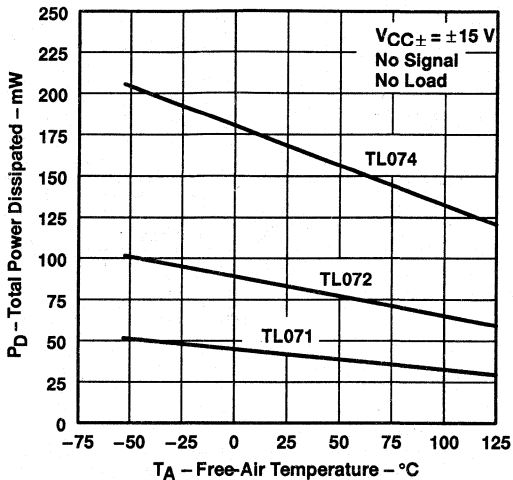
**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT PER AMPLIFIER  
vs  
FREE-AIR TEMPERATURE**



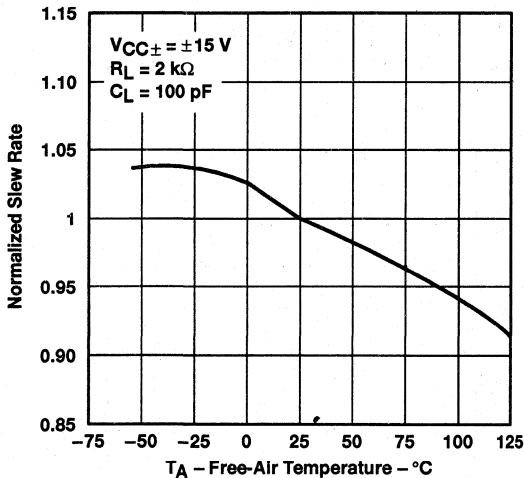
**Figure 16**

**TOTAL POWER DISSIPATED  
vs  
FREE-AIR TEMPERATURE**



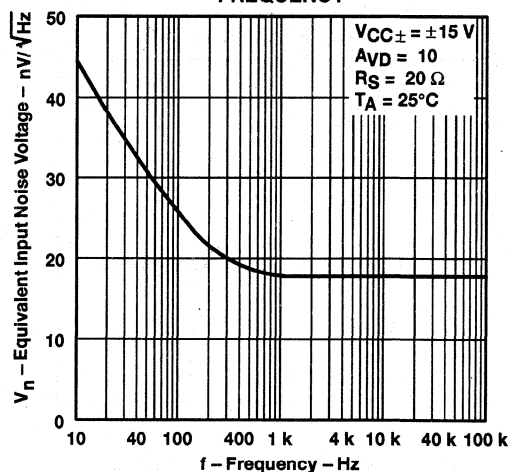
**Figure 17**

**NORMALIZED SLEW RATE  
vs  
FREE-AIR TEMPERATURE**



**Figure 18**

**EQUIVALENT INPUT NOISE VOLTAGE  
vs  
FREQUENCY**



**Figure 19**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

TOTAL HARMONIC DISTORTION  
 VS  
 FREQUENCY

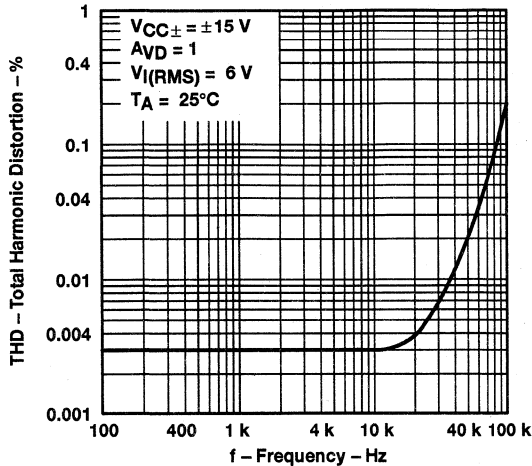


Figure 20

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL PULSE RESPONSE

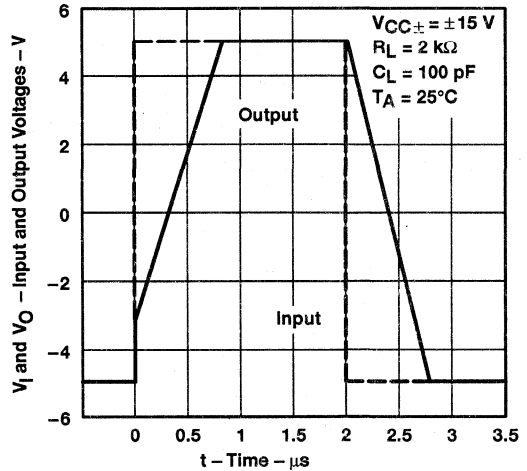


Figure 21

OUTPUT VOLTAGE  
 VS  
 ELAPSED TIME

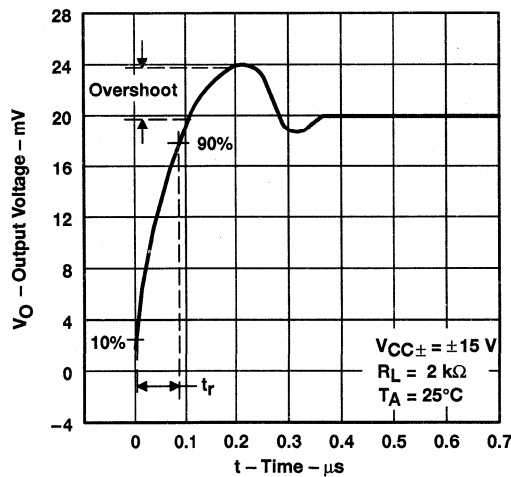
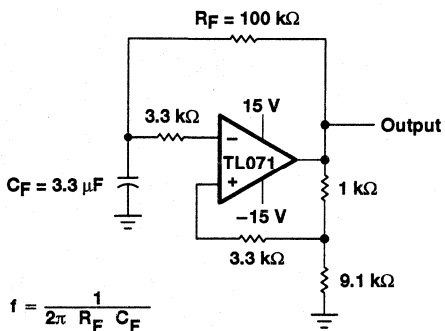


Figure 22

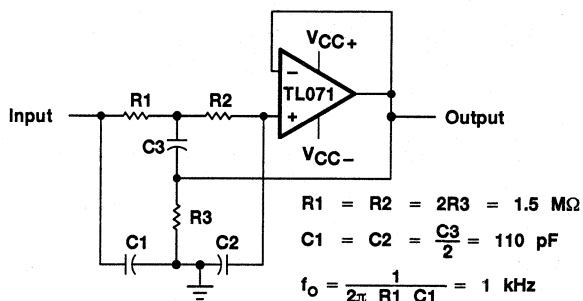
**APPLICATION INFORMATION**

**Table of Application Diagrams**

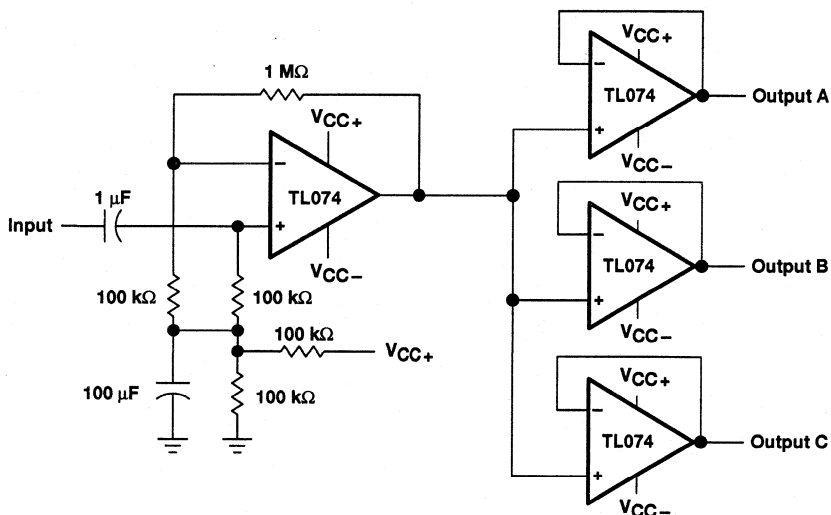
APPLICATION DIAGRAM	PART NUMBER	FIGURE
0.5-Hz square-wave oscillator	TL071	23
High-Q notch filter	TL071	24
Audio-distribution amplifier	TL074	25
100-kHz quadrature oscillator	TL072	26
AC amplifier	TL071	27



**Figure 23. 0.5-Hz Square-Wave Oscillator**



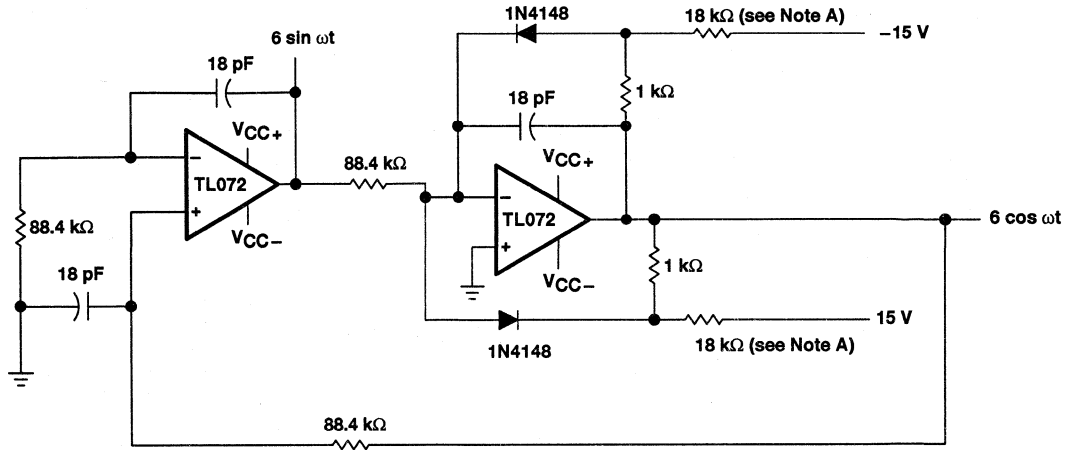
**Figure 24. High-Q Notch Filter**



**Figure 25. Audio-Distribution Amplifier**



APPLICATION INFORMATION



NOTE A: These resistor values may be adjusted for a symmetrical output.

Figure 26. 100-kHz Quadrature Oscillator

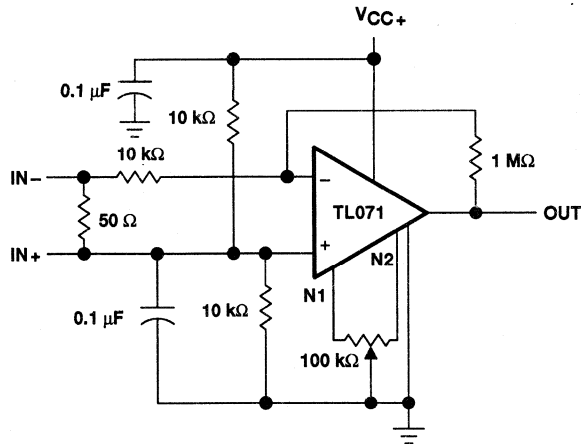


Figure 27. AC Amplifier



# TL074x2 LOW-NOISE JFET-INPUT OCTAL OPERATIONAL AMPLIFIER

SLOS135 – APRIL 1994

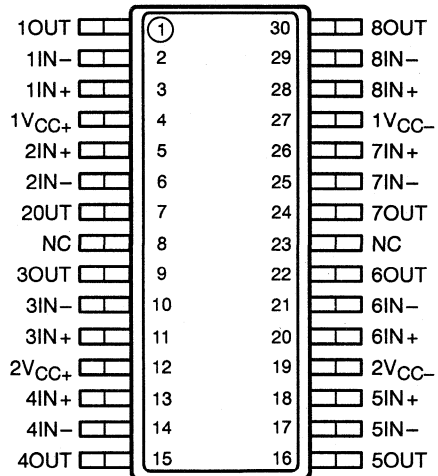
- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion  
0.003% Typ
- Low Noise  
 $V_n = 18 \text{ nV}/\sqrt{\text{Hz}}$  Typ at  $f = 1 \text{ kHz}$
- High Input Impedance . . . JFET Input Stage
- Internal Frequency Compensation
- Latch-Up-Free Operation
- High Slew Rate . . .  $13 \text{ V}/\mu\text{s}$  Typ
- Common-Mode Input Voltage Range  
Includes  $V_{CC+}$

## description

The TL074x2 JFET-input operational amplifier is designed as a low-noise version of the TL084x2 amplifier with low input bias and offset currents and fast slew rate. The low harmonic distortion and low noise make the TL074x2 ideally suited for high-fidelity and audio-preamplifier applications. Each amplifier features JFET inputs (for high input impedance) coupled with bipolar output stages integrated on a single monolithic chip.

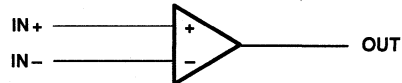
The TL074x2 is characterized for operation from 0°C to 70°C.

**DB PACKAGE  
(TOP VIEW)**



NC – No internal connection

## symbol (each amplifier)



## AVAILABLE OPTION

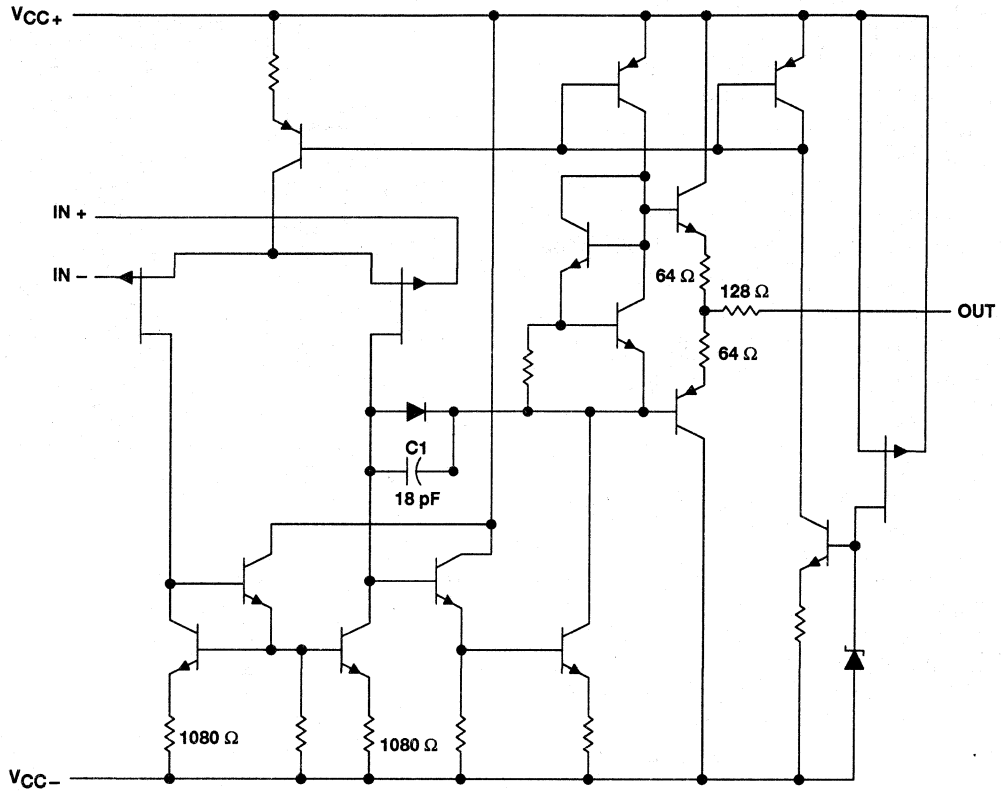
T <sub>A</sub>	V <sub>IQ</sub> max AT 25°C	PACKAGE
		SMALL OUTLINE (DB)†
0°C to 70°C	10 mV	TL074x2DBLE

† The DB package is only available left-taped and reeled.

**TL074x2**  
**LOW-NOISE JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**

SLOS135 - APRIL 1994

**schematic (each amplifier)**



All component values shown are nominal.

COMPONENT COUNT †	
Resistors	88
Transistors	112
JFET	20
Diodes	12
Capacitors	8

† Includes bias and trim circuitry

**TL074x2**  
**LOW-NOISE JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**  
SLOS135 – APRIL 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	18 V
Supply voltage, $V_{CC-}$ (see Note 1)	-18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$ V
Input voltage range, $V_I$ (see Notes 1 and 3)	$\pm 15$ V
Duration of output short circuit (see Note 4)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
  3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
  4. The output can be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DB	1024 mW	8.2 mW/°C	655 mW

**TL074x2**  
**LOW-NOISE JFET-INPUT**  
**OCTAL OPERATIONAL AMPLIFIER**

SLOS135 – APRIL 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†		$T_A$ ‡	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 0$ , $R_S = 50 \Omega$		25°C		3	10	mV
				Full range			13	
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	$V_O = 0$ , $R_S = 50 \Omega$		Full range		18		$\mu V/^\circ C$
$I_{IO}$	Input offset current	$V_O = 0$		25°C		5	100	pA
				Full range			10	nA
$I_{IB}$	Input bias current§	$V_O = 0$		25°C		65	200	pA
				Full range			7	nA
$V_{ICR}$	Common-mode input voltage range			25°C	$\pm 11$	-12 to 15		V
$V_{OM}$	Maximum peak output voltage swing	$R_L = 10 k\Omega$		25°C	$\pm 12$	$\pm 13.5$		V
		$R_L \geq 10 k\Omega$		Full range	$\pm 12$			
		$R_L \geq 2 k\Omega$			$\pm 10$			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 10$ V, $R_L \geq 2 k\Omega$		25°C	25	200		V/mV
				Full range	15			
$B_1$	Unity-gain bandwidth			25°C		3		MHz
$r_i$	Input resistance			25°C		$10^{12}$		$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$ , $V_O = 0$ , $R_S = 50 \Omega$		25°C	70	100		dB
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC} = \pm 9$ V to $\pm 15$ V, $V_O = 0$ , $R_S = 50 \Omega$		25°C	70	100		dB
$I_{CC}$	Supply current (each amplifier)	$V_O = 0$ , No load		25°C		1.4	2.5	mA
$V_{O1}/V_{O2}$	Crosstalk attenuation	$A_{VD} = 100$		25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

‡ Full range is  $T_A = 0^\circ C$  to  $70^\circ C$ .

§ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 2. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**operating characteristics,  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ C$**

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10$ V, $C_L = 100$ pF,	$R_L = 2 k\Omega$ , See Figure 1	8	13		V/ $\mu s$
$t_r$	Overshoot factor rise time	$V_I = 20$ mV, $C_L = 100$ pF,	$R_L = 2 k\Omega$ , See Figure 1		0.1		$\mu s$
					20%		
$V_n$	Equivalent input noise voltage	$R_S = 20 \Omega$	$f = 1$ kHz		18		$nV/\sqrt{Hz}$
			$f = 10$ Hz to 10 kHz		4		$\mu V$
$I_n$	Equivalent input noise current	$R_S = 20 \Omega$ ,	$f = 1$ kHz		0.01		$pA/\sqrt{Hz}$
THD	Total harmonic distortion	$V_{Orms} = 10$ V, $R_L \geq 2 k\Omega$ ,	$R_S \leq 1 k\Omega$ , $f = 1$ kHz		0.003%		



**PARAMETER MEASUREMENT INFORMATION**

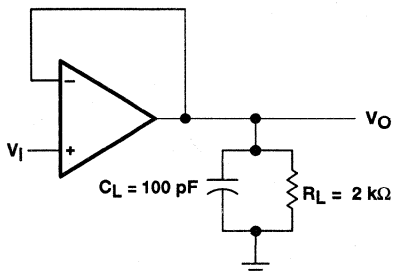


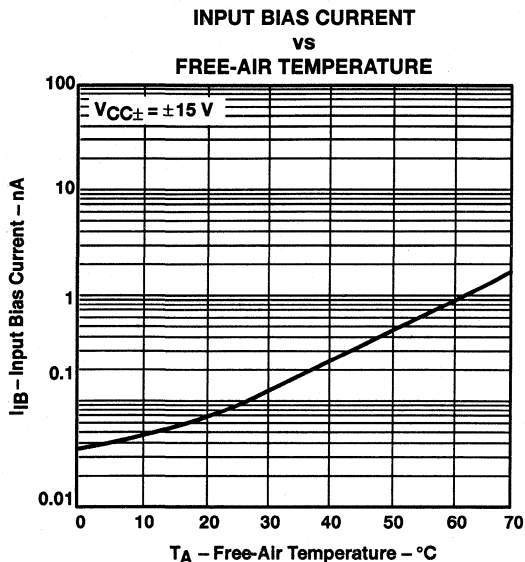
Figure 1. Unity-Gain Amplifier

**TYPICAL CHARACTERISTICS**

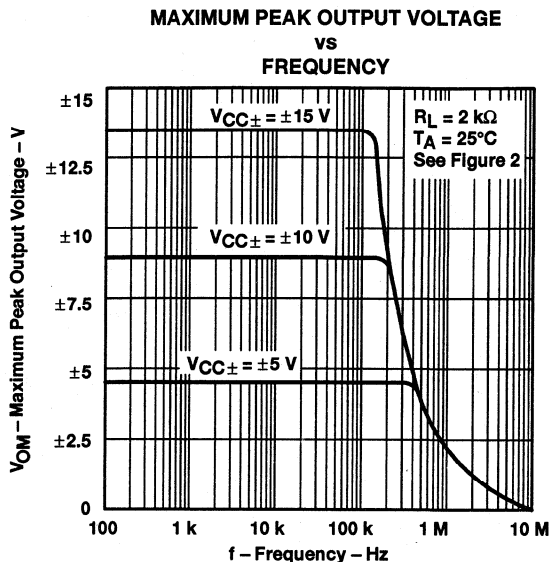
**Table of Graphs**

			FIGURE
$I_{IB}$	Input bias current	vs Free-air temperature	2
$V_{OM}$	Maximum peak output voltage	vs Frequency	3, 4, 5
		vs Free-air temperature	6
		vs Load resistance	7
		vs Supply voltage	8
$A_{VD}$	Large-signal differential voltage amplification	vs Free-air temperature	9
		vs Frequency	10
	Normalized unity-gain bandwidth	vs Free-air temperature	11
$CMRR$	Common-mode rejection ratio	vs Free-air temperature	12
$I_{CC}$	Supply current	vs Supply voltage	13
		vs Free-air temperature	14
$P_D$	Total power dissipation	vs Free-air temperature	15
		Normalized slew rate	16
$V_n$	Equivalent input noise voltage	vs Frequency	17
$THD$	Total harmonic distortion	vs Frequency	18
		Pulse response	Large signal
$V_O$	Output voltage	vs Time	20
		Normalized phase shift	vs Free-air temperature
			11

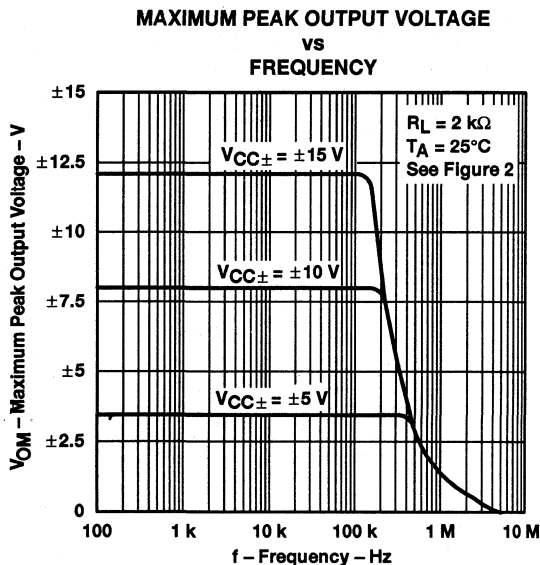
**TYPICAL CHARACTERISTICS**



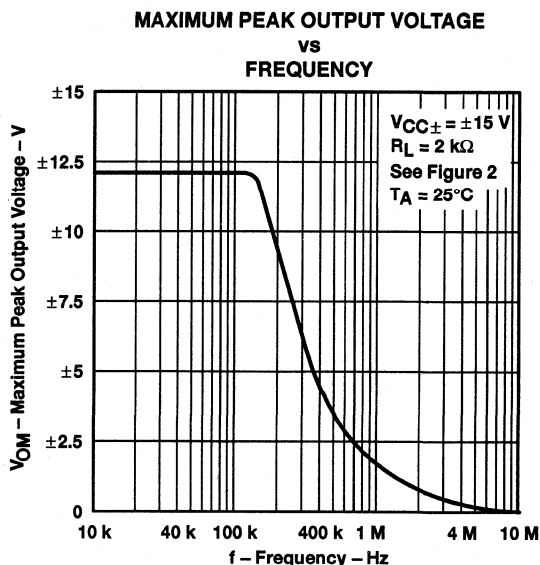
**Figure 2**



**Figure 3**



**Figure 4**



**Figure 5**



**TYPICAL CHARACTERISTICS**

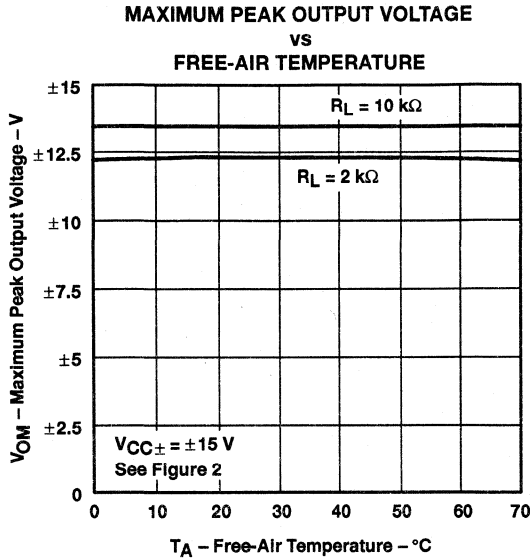


Figure 6

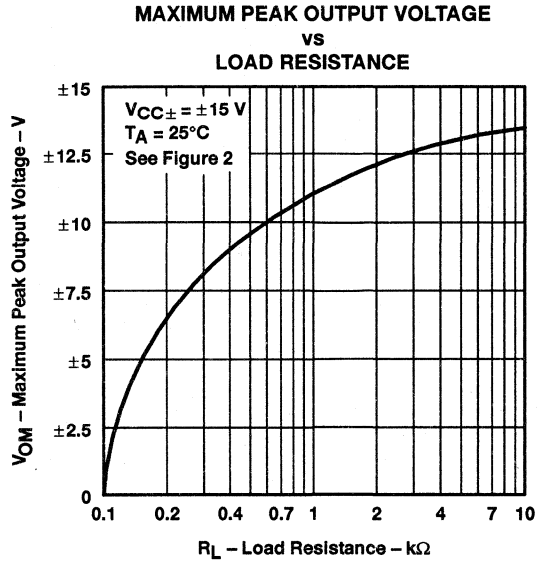


Figure 7

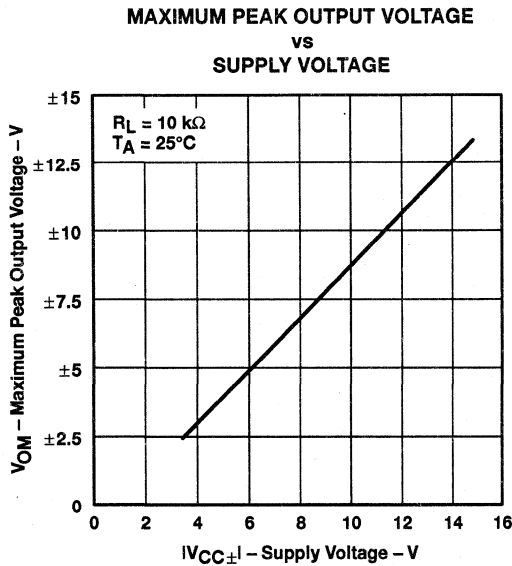


Figure 8

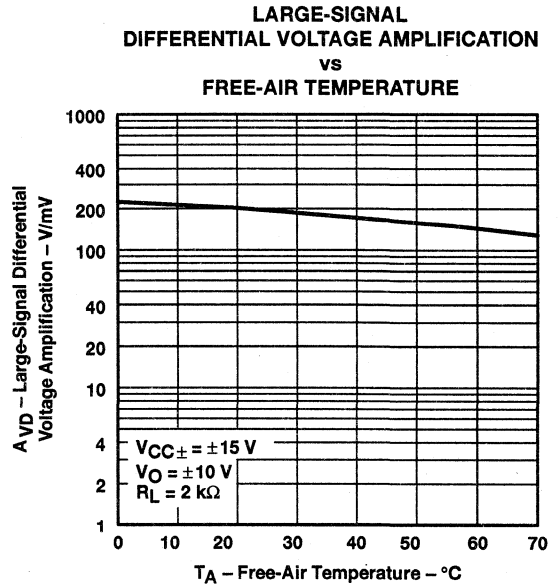
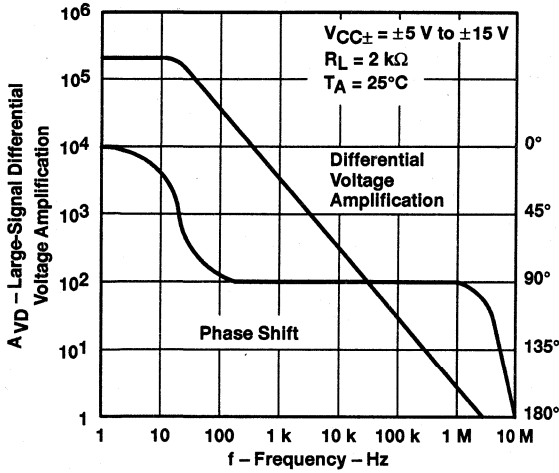


Figure 9

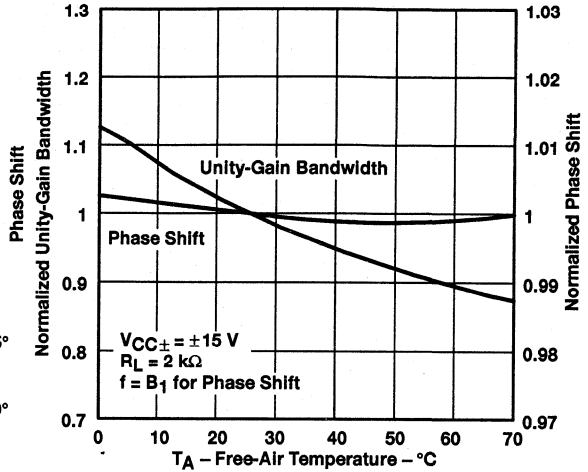
**TYPICAL CHARACTERISTICS**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY**



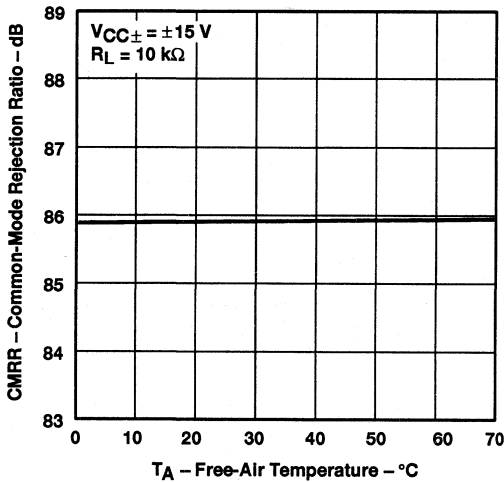
**Figure 10**

**NORMALIZED UNITY-GAIN BANDWIDTH AND PHASE SHIFT VS FREE-AIR TEMPERATURE**



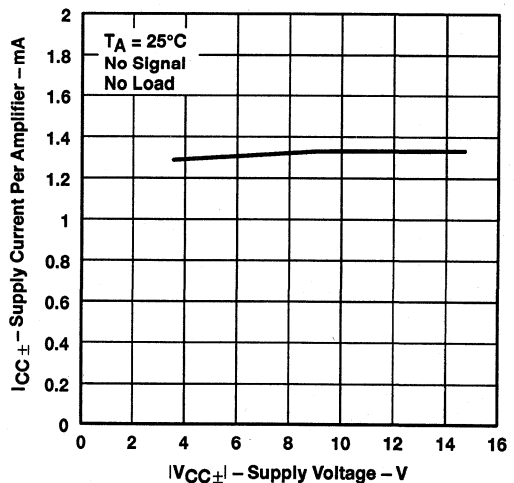
**Figure 11**

**COMMON-MODE REJECTION RATIO VS FREE-AIR TEMPERATURE**



**Figure 12**

**SUPPLY CURRENT PER AMPLIFIER VS SUPPLY VOLTAGE**



**Figure 13**

**TYPICAL CHARACTERISTICS**

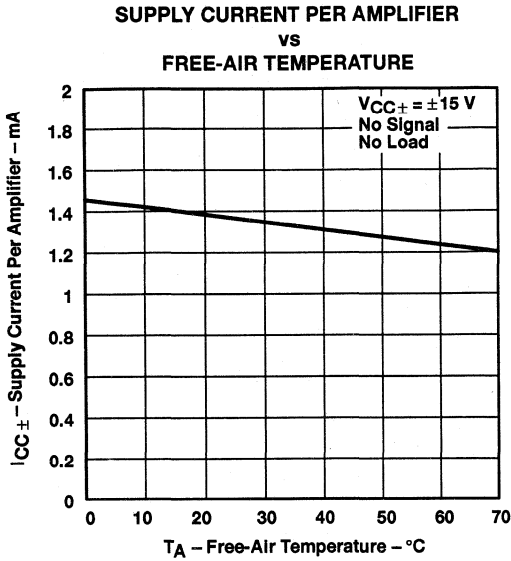


Figure 14

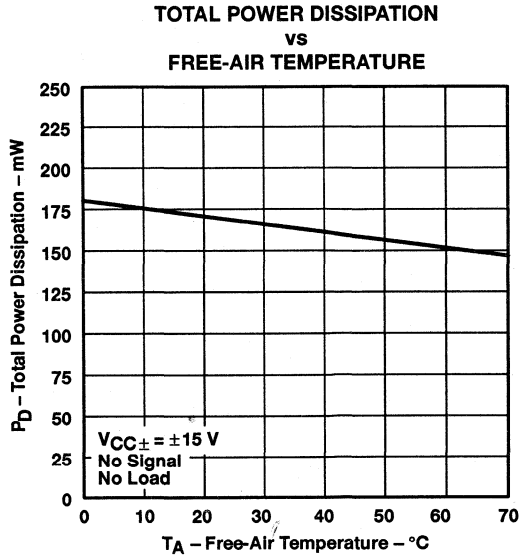


Figure 15

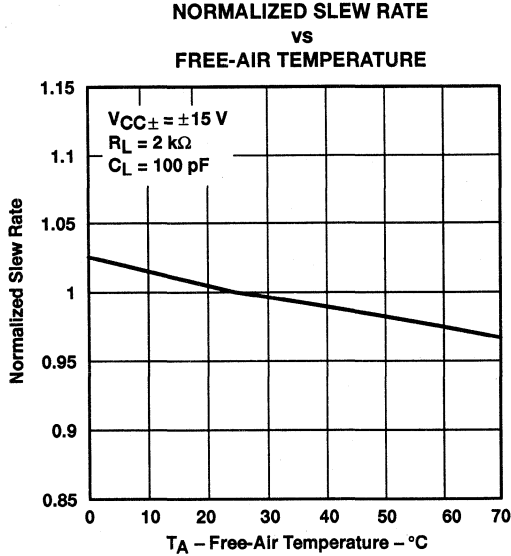


Figure 16

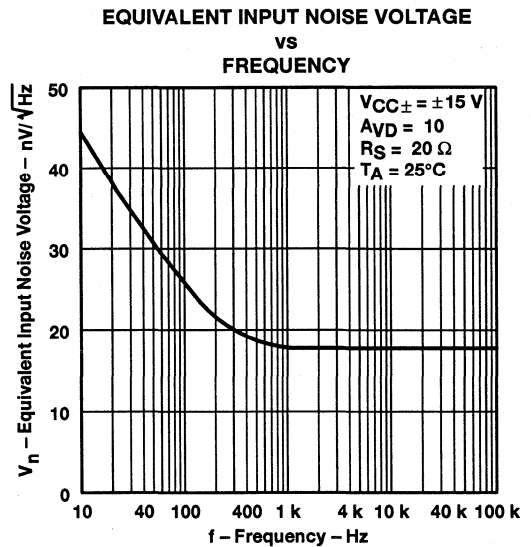
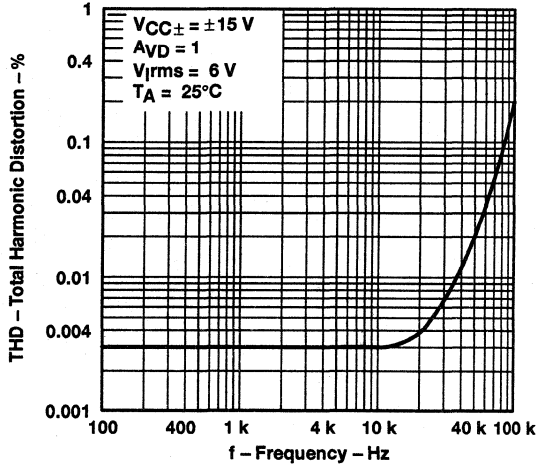


Figure 17

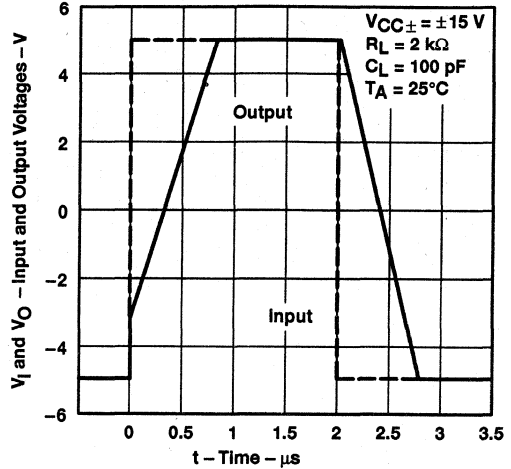
**TYPICAL CHARACTERISTICS**

**TOTAL HARMONIC DISTORTION  
 vs  
 FREQUENCY**



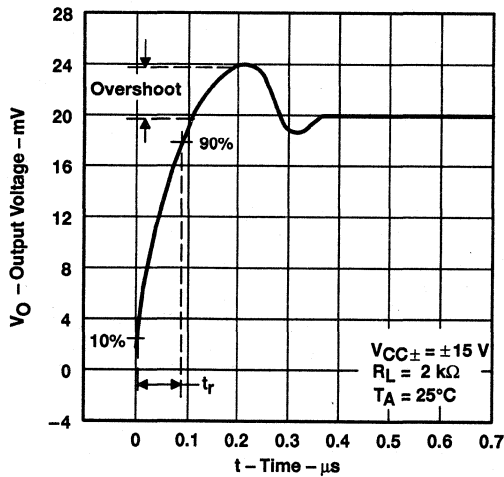
**Figure 18**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL PULSE RESPONSE**



**Figure 19**

**OUTPUT VOLTAGE  
 vs  
 TIME**



**Figure 20**

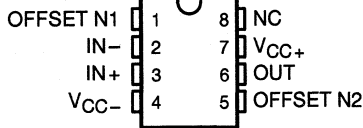
# TL081, TL081A, TL081B, TL082, TL082A, TL082B TL082Y, TL084, TL084A, TL084B, TL084Y JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

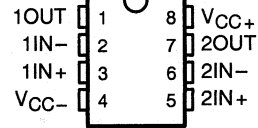
## 24 DEVICES COVER COMMERCIAL, INDUSTRIAL, AND MILITARY TEMPERATURE RANGES

- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion . . . 0.003% Typ
- High Input Impedance . . . JFET-Input Stage
- Latch-Up-Free Operation
- High Slew Rate . . . 13 V/ $\mu$ s Typ
- Common-Mode Input Voltage Range Includes  $V_{CC+}$

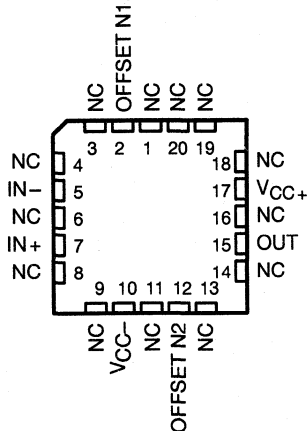
TL081, TL081A, TL081B  
D, JG, P, PW, OR U PACKAGE  
(TOP VIEW)



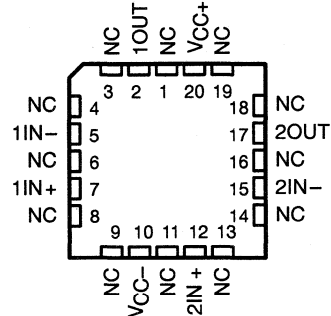
TL082, TL082A, TL082B  
D, JG, P, PW, OR U PACKAGE  
(TOP VIEW)



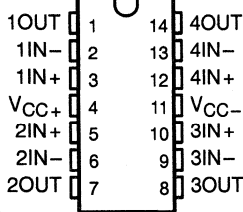
TL081M . . . FK PACKAGE  
(TOP VIEW)



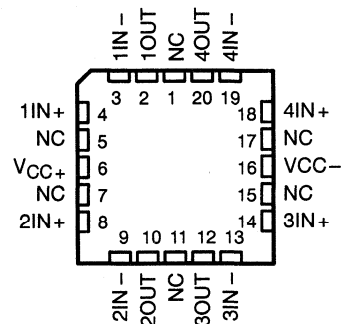
TL082M . . . FK PACKAGE  
(TOP VIEW)



TL084, TL084A, TL084B  
D, J, N, PW, OR U PACKAGE  
(TOP VIEW)



TL084M . . . FK PACKAGE  
(TOP VIEW)



NC—No internal connection

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

 **TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

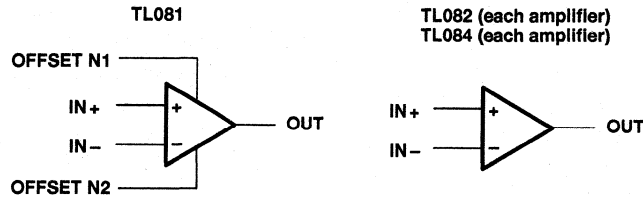
On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

2-361

**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

**symbols**



**description**

The TL08\_ JFET-input operational amplifier family is designed to offer a wider selection than any previously developed operational amplifier family. Each of these JFET-input operational amplifiers incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. The devices feature high slew rates, low input bias and offset currents, and low offset voltage temperature coefficient. Offset adjustment and external compensation options are available within the TL08\_ family.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.

TL081, TL081A, TL081B, TL082, TL082A, TL082B  
 TL082Y, TL084, TL084A, TL084B, TL084Y  
 JFET-INPUT OPERATIONAL AMPLIFIERS

SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

AVAILABLE OPTIONS

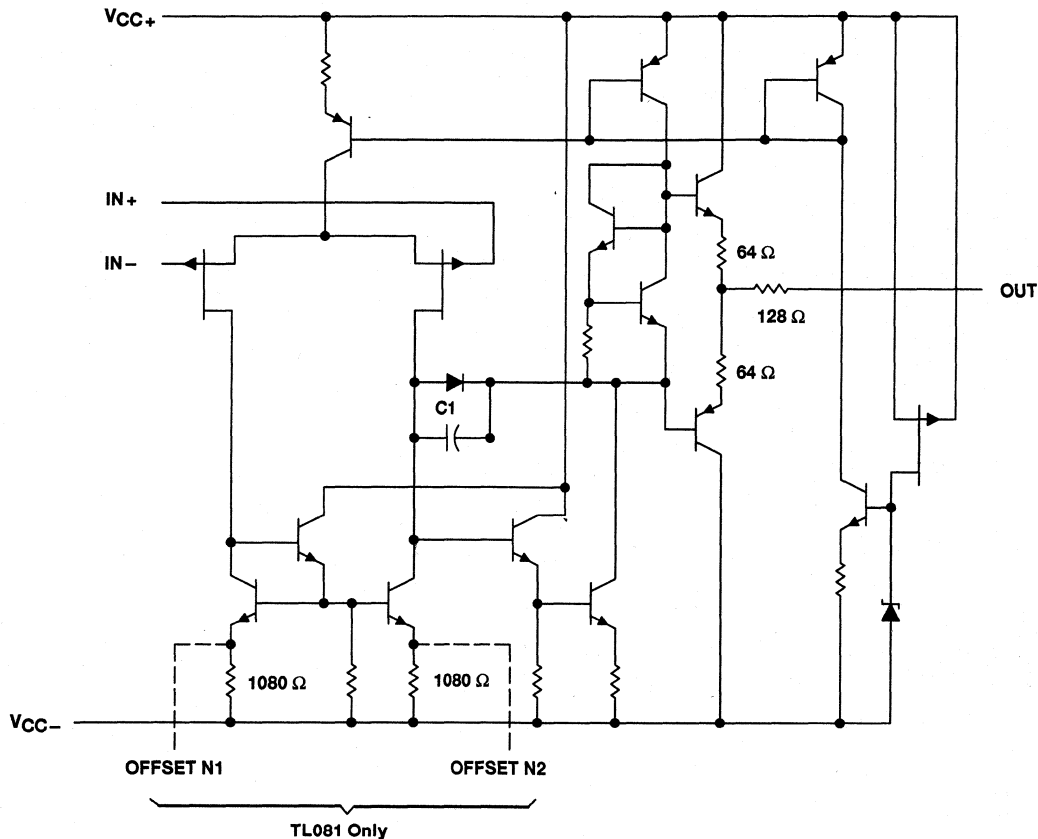
T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES											CHIP FORM FORM (Y)				
		SMALL OUTLINE (D008)	SMALL OUTLINE (D014)	CHIP CARRIER (FK)	CERAMIC DIP (J)	CERAMIC DIP (JG)	PLASTIC DIP (N)	PLASTIC DIP (P)	TSSOP (PW)	FLAT PACK (U)	FLAT PACK (W)						
0°C to 70°C	15 mV	TL081CD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	TL081ACD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	3 mV	TL081BCD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
—40°C to 85°C	15 mV	TL082CD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	TL082ACD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	3 mV	TL082BCD	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
—55°C to 125°C	15 mV	—	TL084CD	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	—	TL084ACD	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	3 mV	—	TL084BCD	—	—	—	—	—	—	—	—	—	—	—	—	—	—
—40°C to 85°C	6 mV	TL081ID	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	TL082ID	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	TL084ID	—	—	—	—	—	—	—	—	—	—	—	—	—	—	—
—55°C to 125°C	6 mV	—	—	TL081MFK	—	—	—	—	—	—	—	—	—	—	—	—	—
	6 mV	—	—	TL082MFK	—	—	—	—	—	—	—	—	—	—	—	—	—
	9 mV	—	—	TL084MFK	—	—	—	—	—	—	—	—	—	—	—	—	—
																	TL084MW

The D package is available taped and reeled. Add R suffix to device type (e.g., TL081CDR).



**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
 TL082Y, TL084, TL084A, TL084B, TL084Y  
 JFET-INPUT OPERATIONAL AMPLIFIERS**  
 SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

**schematic (each amplifier)**



Component values shown are nominal.

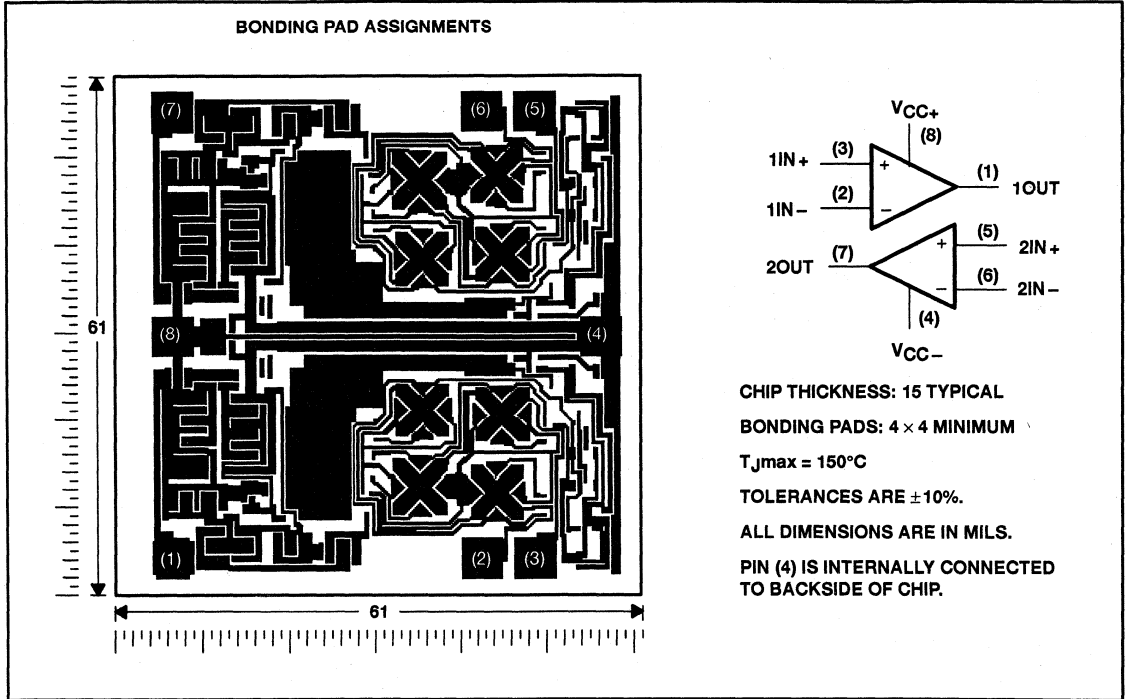


TL081, TL081A, TL081B, TL082, TL082A, TL082B  
 TL082Y, TL084, TL084A, TL084B, TL084Y  
**JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

**TL082Y chip information**

These chips, when properly assembled, display characteristics similar to the TL082. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.

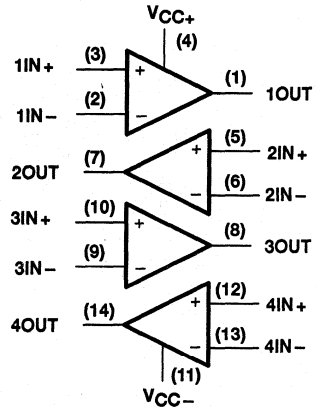
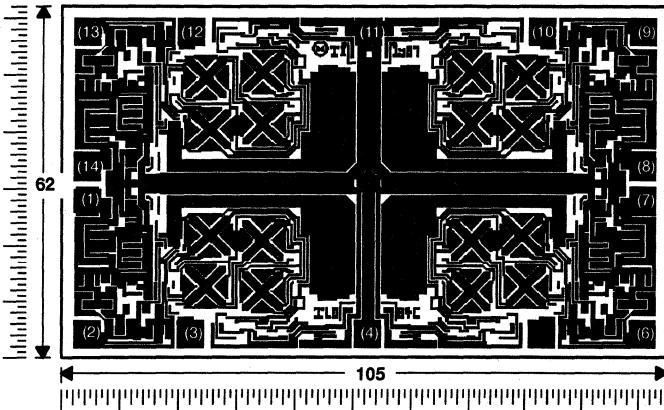


**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
 TL082Y, TL084, TL084A, TL084B, TL084Y  
 JFET-INPUT OPERATIONAL AMPLIFIERS**  
 SLOS081B – FEBRUARY 1977 – REVISED AUGUST 1994

**TL084Y chip information**

These chips, when properly assembled, display characteristics similar to the TL084. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.

**BONDING PAD ASSIGNMENTS**



**CHIP THICKNESS: 15 TYPICAL**  
**BONDING PADS: 4 × 4 MINIMUM**  
 $T_{jmax} = 150^{\circ}C$   
**TOLERANCES ARE  $\pm 10\%$ .**  
**ALL DIMENSIONS ARE IN MILS.**  
**PIN (11) IS INTERNALLY CONNECTED TO BACKSIDE OF CHIP.**

**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B – FEBRUARY 1977 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

	TL08_C TL08_AC TL08_BC	TL08_I	TL08_M	UNIT
Supply voltage, $V_{CC+}$ (see Note 1)	18	18	18	V
Supply voltage $V_{CC-}$ (see Note 1)	-18	-18	-18	V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 30$	$\pm 30$	$\pm 30$	V
Input voltage, $V_I$ (see Notes 1 and 3)	$\pm 15$	$\pm 15$	$\pm 15$	V
Duration of output short circuit (see Note 4)	unlimited	unlimited	unlimited	
Continuous total dissipation	See Dissipation Rating Table			
Operating free-air temperature range, $T_A$	0 to 70	-40 to 85	-55 to 125	$^{\circ}\text{C}$
Storage temperature range	-65 to 150	-65 to 150	-65 to 150	$^{\circ}\text{C}$
Case temperature for 60 seconds	FK package			260
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds	J or JG package			300
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	D, N, P, or PW package			260

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values except differential voltages are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
  2. Differential voltages are at  $\text{IN}+$  with respect to  $\text{IN}-$ .
  3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
  4. The output may be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^{\circ}\text{C}$ POWER RATING	DERATING FACTOR	DERATE ABOVE $T_A$	$T_A = 70^{\circ}\text{C}$ POWER RATING	$T_A = 85^{\circ}\text{C}$ POWER RATING	$T_A = 125^{\circ}\text{C}$ POWER RATING
D (8 pin)	680 mW	5.8 mW/ $^{\circ}\text{C}$	32 $^{\circ}\text{C}$	464 mW	377 mW	N/A
D (14 pin)	680 mW	7.6 mW/ $^{\circ}\text{C}$	60 $^{\circ}\text{C}$	608 mW	494 mW	N/A
FK	680 mW	11.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	680 mW	680 mW	275 mW
J	680 mW	11.0 mW/ $^{\circ}\text{C}$	88 $^{\circ}\text{C}$	680 mW	680 mW	275 mW
JG	680 mW	8.4 mW/ $^{\circ}\text{C}$	69 $^{\circ}\text{C}$	672 mW	546 mW	210 mW
N	680 mW	9.2 mW/ $^{\circ}\text{C}$	76 $^{\circ}\text{C}$	680 mW	598 mW	N/A
P	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	520 mW	N/A
PW (8 pin)	525 mW	4.2 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	336 mW	N/A	N/A
PW (14 pin)	700 mW	5.6 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	448 mW	N/A	N/A
U	675 mW	5.4 mW/ $^{\circ}\text{C}$	25 $^{\circ}\text{C}$	432 mW	351 mW	135 mW
W	680 mW	8.0 mW/ $^{\circ}\text{C}$	65 $^{\circ}\text{C}$	640 mW	520 mW	200 mW

**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B, FEBRUARY 1977 - REVISED AUGUST 1994

**electrical characteristics,  $V_{CC} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TL081C TL082C TL084C			TL081AC TL082AC TL084AC			TL081BC TL082BC TL084BC			TL081I TL082I TL084I			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C Full range	3	15	20	3	6	7.5	2	3	3	6	9	mV	
α <sub>VIO</sub>	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	Full range	18			18			18					μV/°C	
I <sub>IO</sub>	V <sub>O</sub> = 0	25°C Full range	5	200	2	5	100	2	5	100	5	100	10	pA	
I <sub>IB</sub>	V <sub>O</sub> = 0	25°C Full range	30	400	10	30	200	7	30	200	30	200	20	nA	
V <sub>ICR</sub>	Common-mode input voltage range	25°C	±11	-12 to 15		±11	-12 to 15		±11	-12 to 15		±11 to 15		V	
V <sub>OM</sub>	Maximum peak output voltage swing	25°C	±12	±13.5		±12	±13.5		±12	±13.5		±12	±13.5	V	
		Full range	±10	±12		±10	±12		±10	±12		±10	±12		
			25	200		25	200		25	200		25	200		
A <sub>VD</sub>	Large-signal differential voltage amplification	25°C Full range	15			15			15			25		V/mV	
B <sub>1</sub>	Unity-gain bandwidth	25°C	3			3			3			3		MHz	
r <sub>i</sub>	Input resistance	25°C	10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω	
CMRR	Common-mode rejection ratio	25°C	70	86		75	86		75	86		75	86	dB	
k <sub>SVR</sub>	Supply voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	25°C	70	86		80	86		80	86		80	86	dB	
I <sub>CC</sub>	Supply current (per amplifier)	25°C	1.4	2.8		1.4	2.8		1.4	2.8		1.4	2.8	mA	
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	25°C	120			120			120			120		dB	

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified. Full range for T<sub>A</sub> is 0°C to 70°C for TL08\_C, TL08\_AC, TL08\_BC and -40°C to 85°C for TL08\_I.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 18. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.



**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B – FEBRUARY 1977 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC} \pm = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITION†	T <sub>A</sub>	TL081M, TL082M			TL084M			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	UNIT
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			3 6			mV
			-55°C to 125°C			9 15			mV
α <sub>VIO</sub>	Temperature coefficient of input offset voltage	V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	-55°C to 125°C			18 18			μV/°C
I <sub>IO</sub>	Input offset current‡	V <sub>O</sub> = 0	25°C			5 100			pA
			125°C			20 20			nA
I <sub>IB</sub>	Input bias current‡	V <sub>O</sub> = 0	25°C			30 200			pA
			125°C			50 50			nA
V <sub>ICR</sub>	Common-mode input voltage range		25°C			±11 -12 to 15			V
V <sub>OM</sub>	Maximum peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C			±12 ±13.5			V
		R <sub>L</sub> ≥ 10 kΩ	-55°C to 125°C			±12			V
		R <sub>L</sub> ≥ 2 kΩ				±10 ±12			V
A <sub>VD</sub>	Large-signal differential voltage amplification	V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	25°C			25 200			V/mV
		V <sub>O</sub> = ±10 V, R <sub>L</sub> ≥ 2 kΩ	-55°C to 125°C			15 15			V/mV
B <sub>1</sub>	Unity-gain bandwidth		25°C			3 3			MHz
r <sub>i</sub>	Input resistance		25°C			10 <sup>12</sup> 10 <sup>12</sup>			Ω
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			80 86			dB
k <sub>SVR</sub>	Supply voltage rejection ratio (ΔV <sub>CC±</sub> /ΔV <sub>IO</sub> )	V <sub>CC</sub> = ±15 V to ±9 V, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C			80 86			dB
I <sub>CC</sub>	Supply current (per amplifier)	V <sub>O</sub> = 0, No load	25°C			1.4 2.8			mA
V <sub>O1</sub> /V <sub>O2</sub>	Crosstalk attenuation	A <sub>VD</sub> = 100	25°C			120 120			dB

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.

‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 18. Pulse techniques must be used that will maintain the junction temperatures as close to the ambient temperature as is possible.

**operating characteristics,  $V_{CC} \pm = \pm 15$  V, T<sub>A</sub> = 25°C (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
SR	V <sub>I</sub> = 10 V, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, See Figure 1	8*	13		V/μs	
	V <sub>I</sub> = 10 V, R <sub>L</sub> = 2 kΩ, C <sub>L</sub> = 100 pF, T <sub>A</sub> = -55°C to 125°C, See Figure 1	5*				
t <sub>r</sub>	Rise time		0.05		μs	
	Overshoot factor		20%			
V <sub>n</sub>	Equivalent input noise voltage	R <sub>S</sub> = 20 Ω	f = 1 kHz		18	nV/√Hz
			f = 10 Hz to 10 kHz		4	μV
I <sub>n</sub>	Equivalent input noise current	R <sub>S</sub> = 20 Ω, f = 1 kHz	0.01		pA/√Hz	
THD	Total harmonic distortion	V <sub>O(rms)</sub> = 10 V, R <sub>S</sub> ≤ 1 kΩ, R <sub>L</sub> ≥ 2 kΩ, f = 1 kHz			0.003%	

\*On products compliant to MIL-STD-883, Class B, this parameter is not production tested.



**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B – FEBRUARY 1977 – REVISED AUGUST 1994

**electrical characteristics,  $V_{CC\pm} = \pm 15\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS†	TL082Y, TL084Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$		3	15	mV
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_O = 0$ , $R_S = 50\ \Omega$		18		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current‡	$V_O = 0$		5	200	pA
$I_{IB}$ Input bias current‡	$V_O = 0$		30	400	pA
$V_{ICR}$ Common-mode input voltage range		$\pm 11$	-12 to 15		V
$V_{OM}$ Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega$	$\pm 12$	$\pm 13.5$		V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V}$ , $R_L \geq 2\ \text{k}\Omega$	25	200		V/mV
$B_1$ Unity-gain bandwidth			3		MHz
$r_i$ Input resistance			$10^{12}$		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$ , $V_O = 0$ , $R_S = 50\ \Omega$	70	86		dB
		70	86		dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{CC\pm} / \Delta V_{IO}$ )	$V_{CC} = \pm 15\ \text{V}$ to $\pm 9\ \text{V}$ , $V_O = 0$ , $R_S = 50\ \Omega$	70	86		dB
		70	86		dB
$I_{CC}$ Supply current (per amplifier)	$V_O = 0$ , No load		1.4	2.8	mA
$V_{O1}/V_{O2}$ Crosstalk attenuation	$A_{VD} = 100$		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

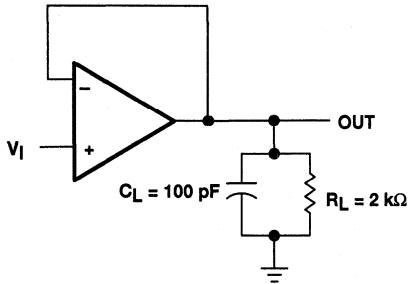
‡ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 18. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

**operating characteristics,  $V_{CC\pm} = \pm 15\ \text{V}$ ,  $T_A = 25^\circ\text{C}$**

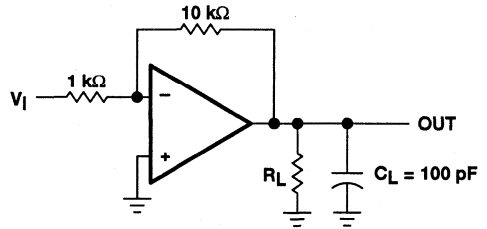
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR Slew rate at unity gain	$V_I = 10\ \text{V}$ , $R_L = 2\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$ , See Figure 1	8	13		V/ $\mu\text{s}$
$t_r$ Rise time	$V_I = 20\ \text{mV}$ , $R_L = 2\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$ , See Figure 1		0.05		$\mu\text{s}$
Overshoot factor			20%		
$V_n$ Equivalent input noise voltage	$R_S = 20\ \Omega$	$f = 1\ \text{kHz}$		18	nV/ $\sqrt{\text{Hz}}$
		$f = 10\ \text{Hz}$ to $10\ \text{kHz}$		4	$\mu\text{V}$
$I_n$ Equivalent input noise current	$R_S = 20\ \Omega$ , $f = 1\ \text{kHz}$		0.01		pA/ $\sqrt{\text{Hz}}$
THD Total harmonic distortion	$V_{O(rms)} = 10\ \text{V}$ , $R_S \leq 1\ \text{k}\Omega$ , $R_L \geq 2\ \text{k}\Omega$ , $f = 1\ \text{kHz}$		0.003%		



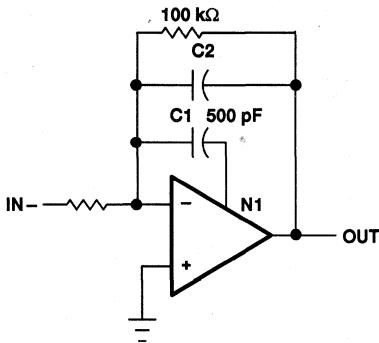
**PARAMETER MEASUREMENT INFORMATION**



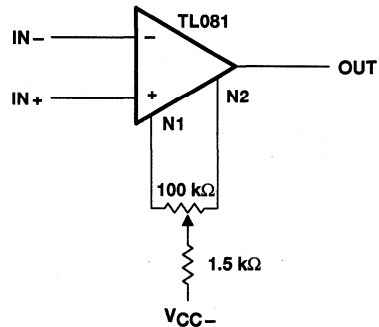
**Figure 1. Unity-Gain Amplifier**



**Figure 2. Gain-of-10 Inverting Amplifier**



**Figure 3. Feed-Forward Compensation**



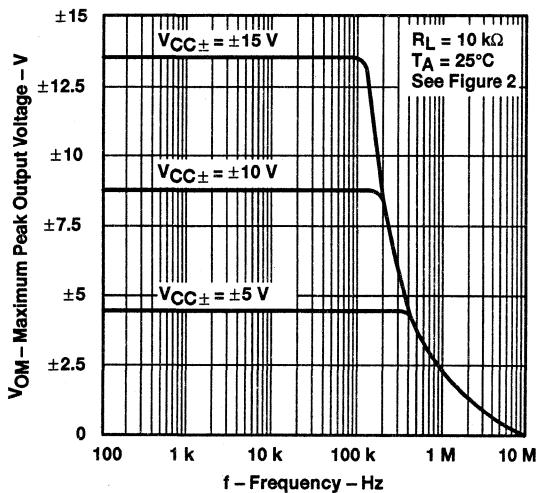
**Figure 4. TL081 Input Offset Voltage Null Circuit**

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

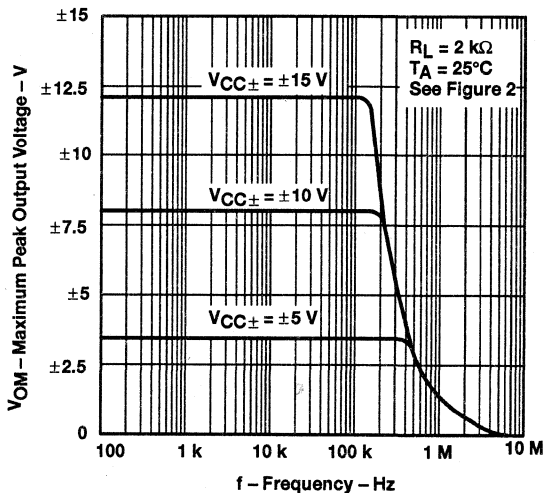
		FIGURE
$V_{OM}$	Maximum peak output voltage	vs Frequency
		vs Free-air temperature
		vs Load resistance
		vs Supply voltage
AVD	Large-signal differential voltage amplification	vs Free-air temperature
		vs Frequency
	Differential voltage amplification	vs Frequency with feed-forward compensation
$P_D$	Total power dissipation	vs Free-air temperature
$I_{CC}$	Supply current	vs Free-air temperature
		vs Supply voltage
$I_{IB}$	Input bias current	vs Free-air temperature
	Large-signal pulse response	vs Time
$V_O$	Output voltage	vs Elapsed time
CMRR	Common-mode rejection ratio	vs Free-air temperature
$V_n$	Equivalent input noise voltage	vs Frequency
THD	Total harmonic distortion	vs Frequency

**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY**



**Figure 5**

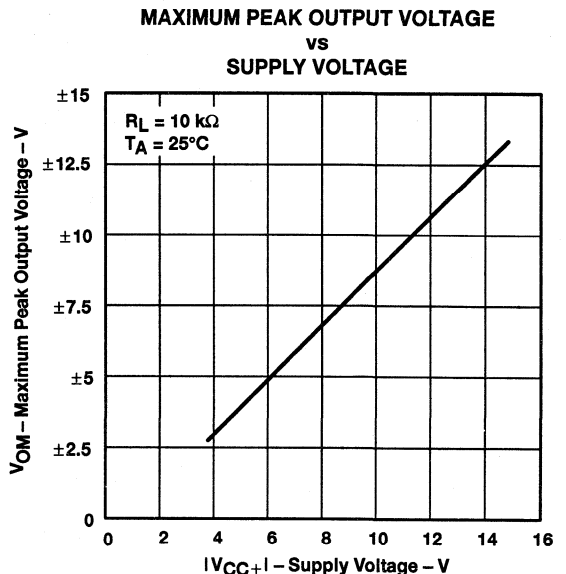
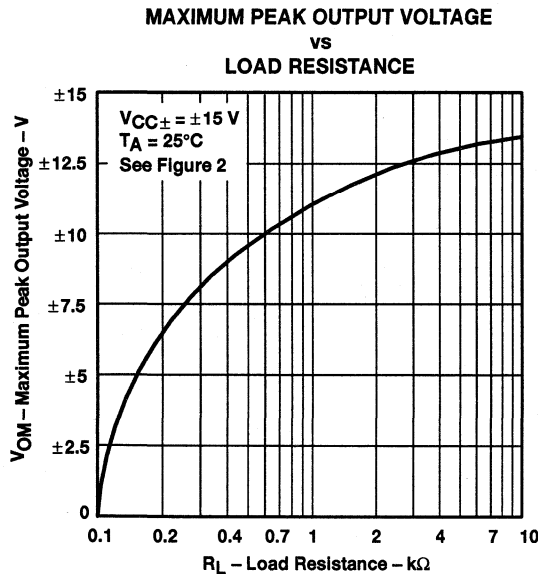
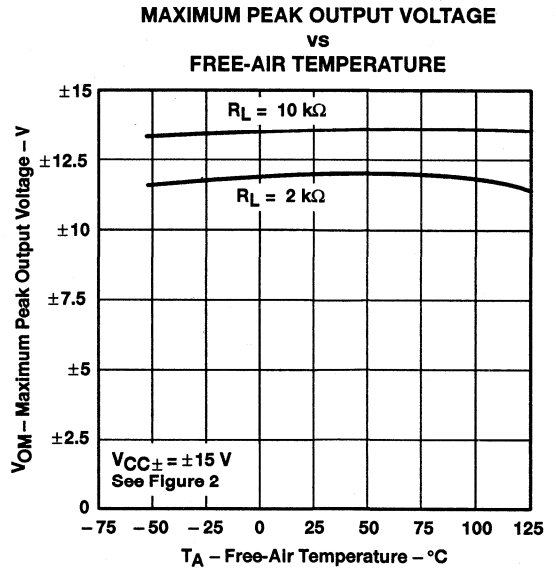
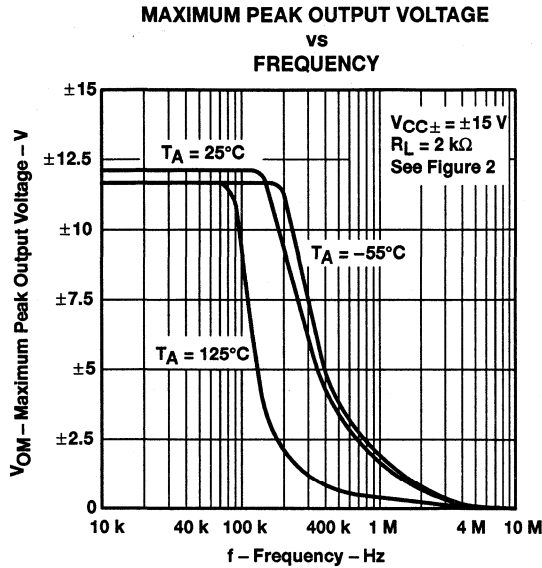
**MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY**



**Figure 6**



TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE**

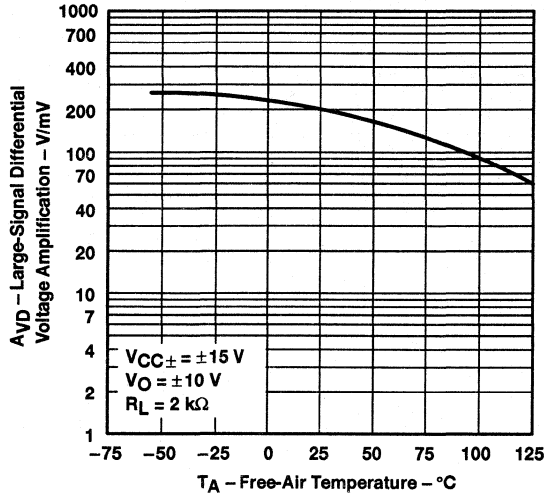


Figure 11

**LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREQUENCY**

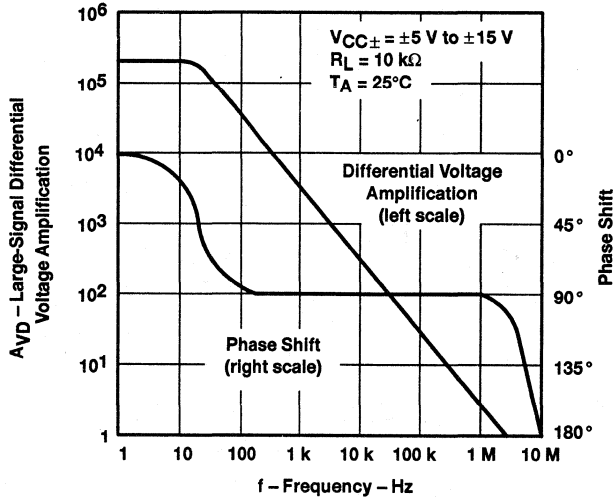


Figure 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

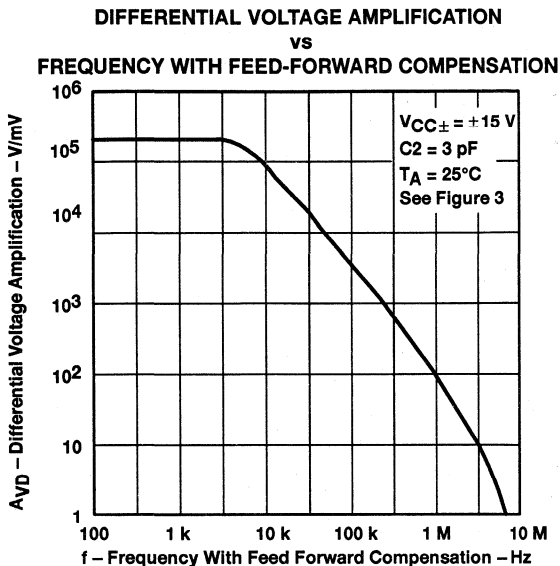


Figure 13

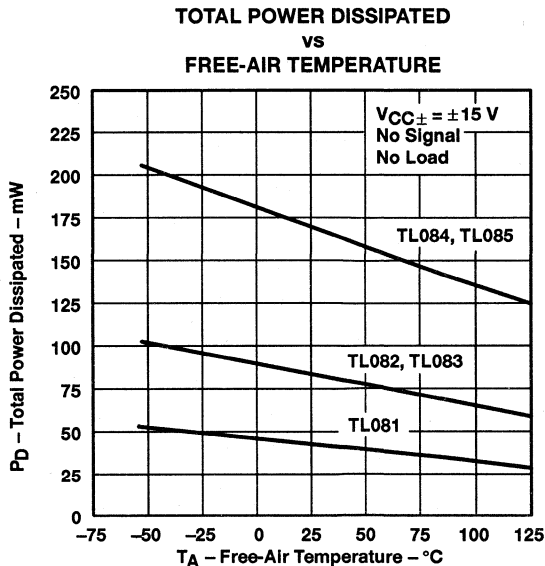


Figure 14

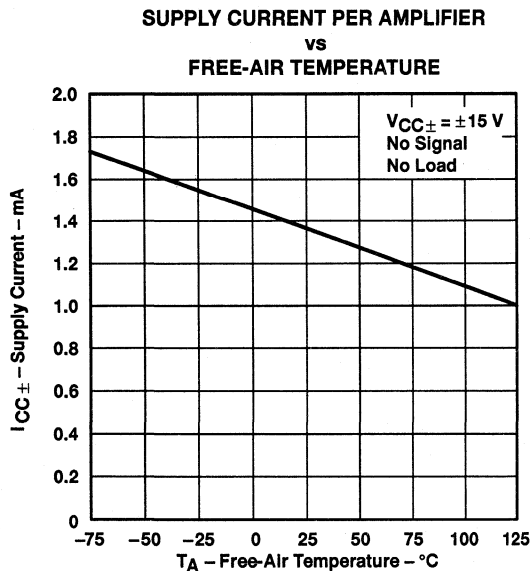


Figure 15

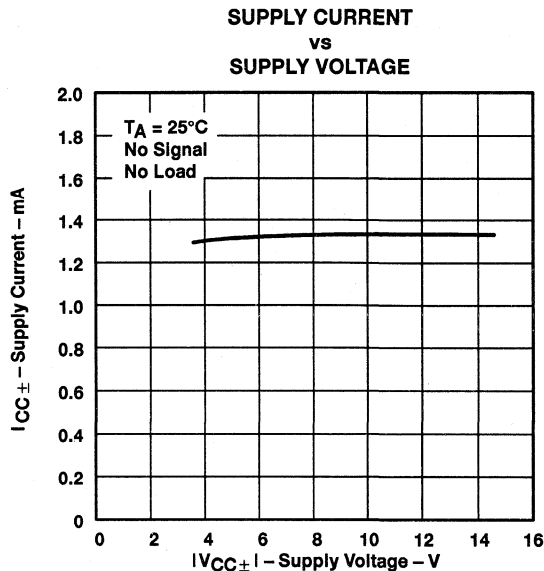
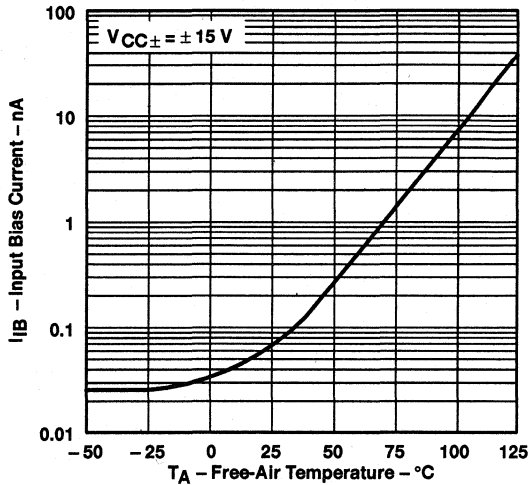


Figure 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

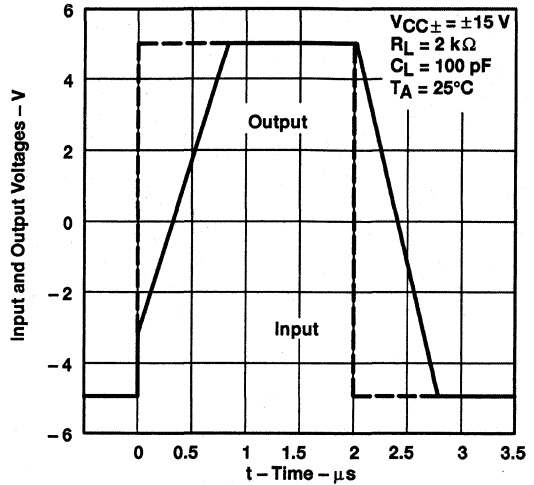
**TYPICAL CHARACTERISTICS†**

**INPUT BIAS CURRENT  
 vs  
 FREE-AIR TEMPERATURE**



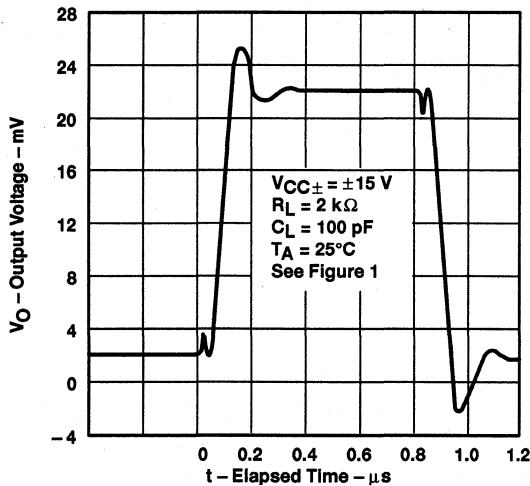
**Figure 17**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL PULSE RESPONSE**



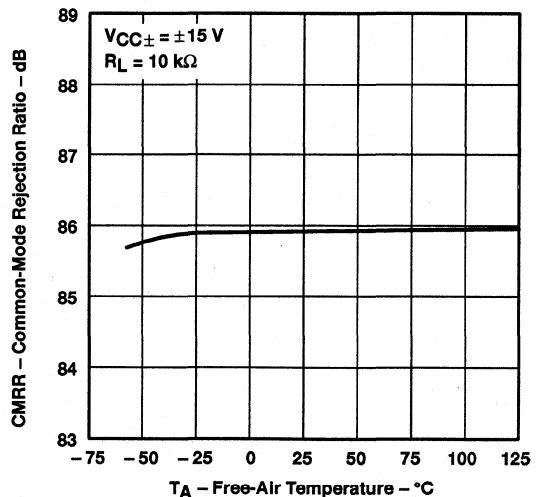
**Figure 18**

**OUTPUT VOLTAGE  
 vs  
 ELAPSED TIME**



**Figure 19**

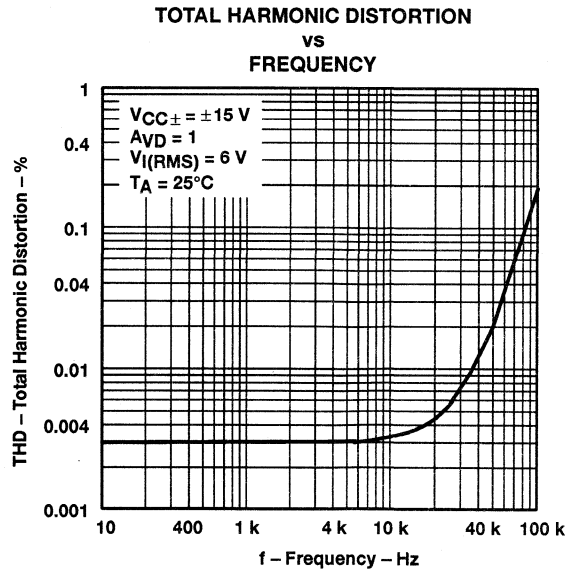
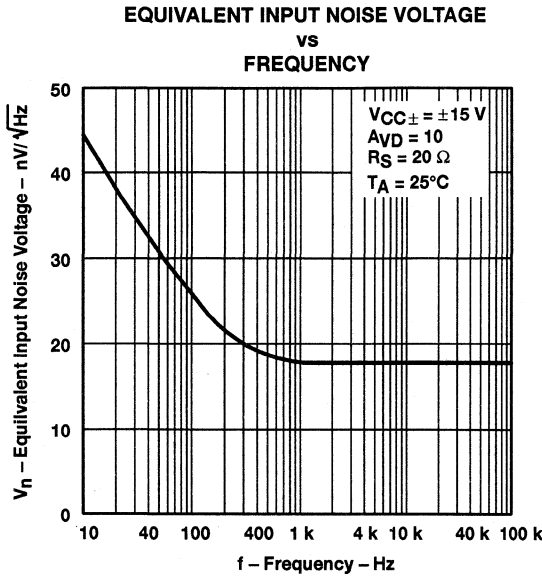
**COMMON-MODE REJECTION RATIO  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 20**

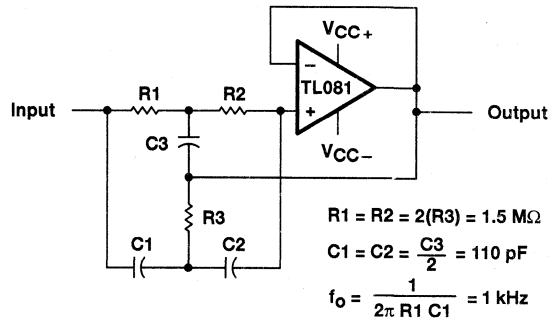
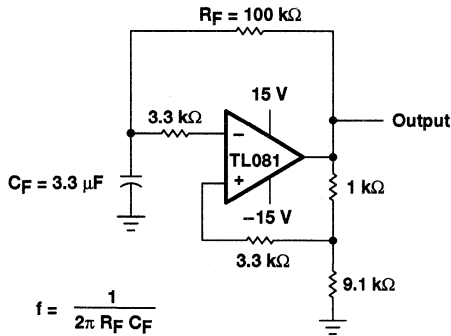
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

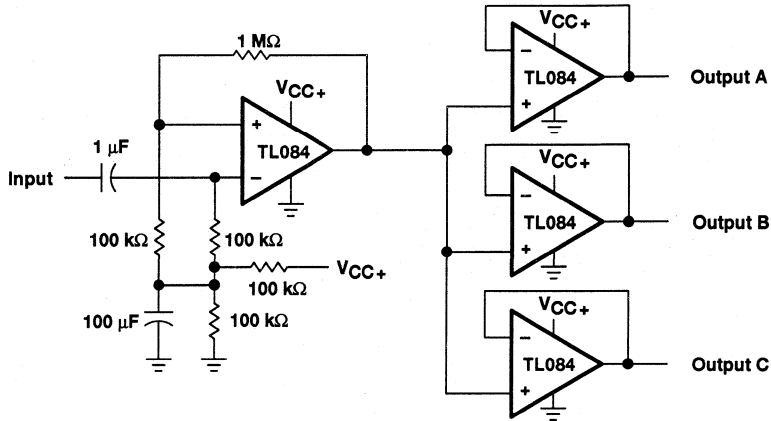
APPLICATION INFORMATION



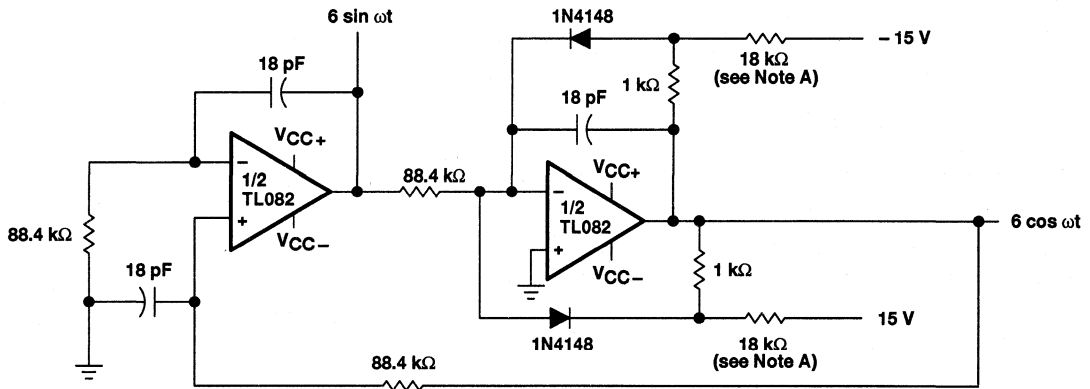
**TL081, TL081A, TL081B, TL082, TL082A, TL082B  
TL082Y, TL084, TL084A, TL084B, TL084Y  
JFET-INPUT OPERATIONAL AMPLIFIERS**

SLOS081B - FEBRUARY 1977 - REVISED AUGUST 1994

**APPLICATION INFORMATION**



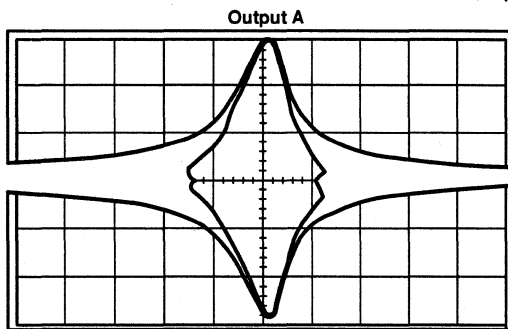
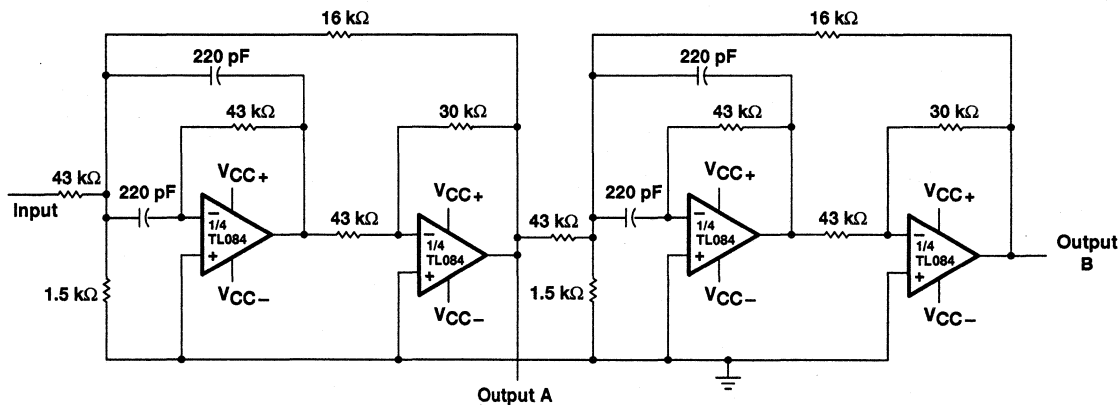
**Figure 25. Audio-Distribution Amplifier**



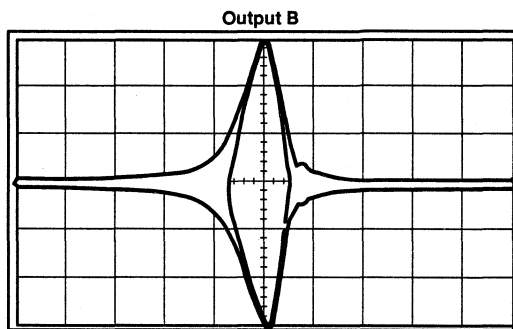
NOTE A: These resistor values may be adjusted for a symmetrical output.

**Figure 26. 100-KHz Quadrature Oscillator**

APPLICATION INFORMATION



2 kHz/div  
 Second-Order Bandpass Filter  
 $f_o = 100 \text{ kHz}$ ,  $Q = 30$ ,  $GAIN = 4$



2 kHz/div  
 Cascaded Bandpass Filter  
 $f_o = 100 \text{ kHz}$ ,  $Q = 69$ ,  $GAIN = 16$

Figure 27. Positive-Feedback Bandpass Filter





# TL084x2 JFET-INPUT OCTAL OPERATIONAL AMPLIFIER

SLOS136 – APRIL 1994

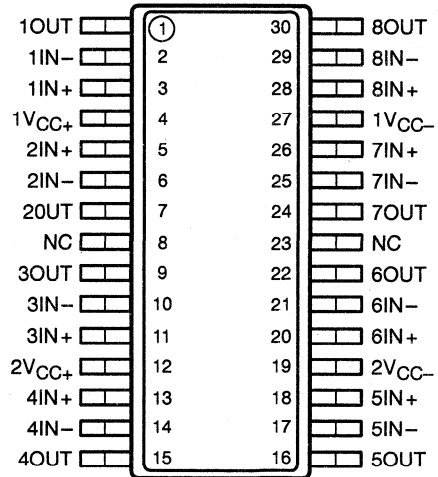
- Low Power Consumption
- Wide Common-Mode and Differential Voltage Ranges
- Low Input Bias and Offset Currents
- Output Short-Circuit Protection
- Low Total Harmonic Distortion . . . 0.003% Typ
- High Input Impedance . . . JFET-Input Stage
- Latch-Up-Free Operation
- High Slew Rate . . . 13 V/ $\mu$ s Typ
- Common-Mode Input Voltage Range Includes  $V_{CC+}$

### description

The TL084x2 JFET-input operational amplifier incorporates well-matched, high-voltage JFET and bipolar transistors in a monolithic integrated circuit. The device features high slew rates, low input bias and offset currents, and low offset voltage temperature coefficient.

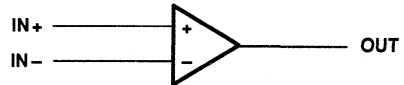
The TL084x2 is characterized for operation from 0°C to 70°C.

**DB PACKAGE  
(TOP VIEW)**



NC – No internal connection

### symbol (each amplifier)



### AVAILABLE OPTION

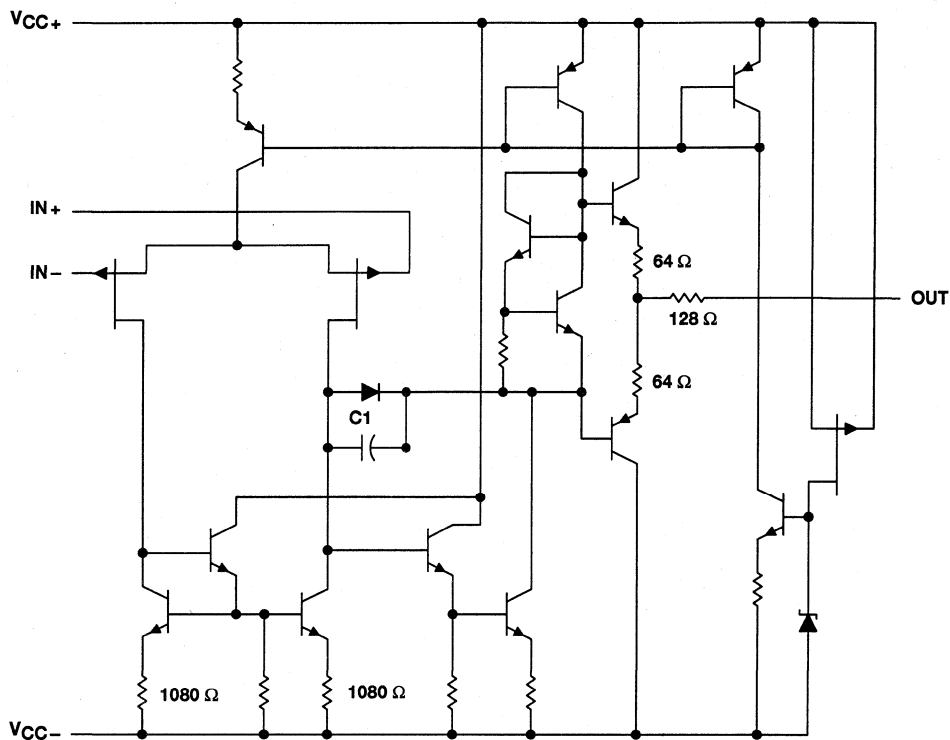
T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE
		SMALL OUTLINE (DB)†
0°C to 70°C	15 mV	TL084x2DBLE

† The DB package is only available left-end taped and reeled.

# TL084x2 JFET-INPUT OCTAL OPERATIONAL AMPLIFIER

SLOS136 – APRIL 1994

## schematic (each amplifier)



All component values shown are nominal.

COMPONENT COUNT	
Resistors	76
Transistors	120
JFET	20
Diodes	12
Capacitors	8

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1) .....	18 V
Supply voltage, $V_{CC-}$ (see Note 1) .....	-18 V
Differential input voltage, $V_{ID}$ (see Note 2) .....	$\pm 30$ V
Input voltage, $V_I$ (any input) (see Notes 1 and 3) .....	$\pm 15$ V
Duration of output short circuit to ground (see Note 4) .....	unlimited
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages and  $V_{CC}$  specified for the measurement of  $I_{OS}$ , are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
4. The output can be shorted to ground or to either supply. Temperature and/or supply voltages must be limited to ensure that the dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DB	1024 mW	8.2 mW/°C	655 mW

# TL084x2 JFET-INPUT OCTAL OPERATIONAL AMPLIFIER

SLOS136 – APRIL 1994

## electrical characteristics, $V_{CC\pm} = \pm 15\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS†	$T_A$ ‡	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 0, R_S = 50\ \Omega$	25°C		5	15	mV
			Full range			20	
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	$V_O = 0, R_S = 50\ \Omega$	Full range		10		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current	$V_O = 0$	25°C		5	200	pA
			Full range			5	nA
$I_{IB}$	Input bias current§	$V_O = 0$	25°C		30	400	pA
			Full range			10	nA
$V_{ICR}$	Common-mode input voltage range		25°C	$\pm 10$	$\pm 11$		V
$V_{OM}$	Maximum peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	$\pm 12$	$\pm 13.5$		V
		$R_L \geq 10\ \text{k}\Omega$	Full range	$\pm 12$			
		$R_L \geq 2\ \text{k}\Omega$		$\pm 10$	$\pm 12$		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 10\ \text{V}, R_L \geq 2\ \text{k}\Omega$	25°C	25	200		V/mV
		$V_O = \pm 10\ \text{V}, R_L \geq 2\ \text{k}\Omega$	Full range	15			
$B_1$	Unity-gain bandwidth		25°C		3		MHz
$r_i$	Input resistance		25°C		10 <sup>12</sup>		$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	70	76		dB
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{CC\pm} / \Delta V_{IO}$ )	$V_{CC} = \pm 15\ \text{V to } \pm 9\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	70	76		dB
$I_{CC}$	Supply current (per amplifier)	$V_O = 0, \text{No load}$	25°C		1.4	2.8	mA
$V_{O1}/V_{O2}$	Crosstalk attenuation	$A_{VD} = 100$	25°C		120		dB

† All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

‡ Full range is 0°C to 70°C.

§ Input bias currents of a FET-input operational amplifier are normal junction reverse currents, which are temperature sensitive as shown in Figure 14. Pulse techniques must be used that will maintain the junction temperature as close to the ambient temperature as possible.

## operating characteristics, $V_{CC\pm} = \pm 15\ \text{V}, T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$V_I = 10\ \text{V}, R_L = 2\ \text{k}\Omega, C_L = 100\ \text{pF}, \text{ See Figure 1}$		13		V/ $\mu\text{s}$
$t_r$	Rise time	$V_I = 20\ \text{mV}, R_L = 2\ \text{k}\Omega, C_L = 100\ \text{pF}, \text{ See Figure 1}$		0.05		$\mu\text{s}$
	Overshoot factor			20%		
$V_n$	Equivalent input noise voltage	$R_S = 20\ \Omega, f = 1\ \text{kHz}$		18		nV/ $\sqrt{\text{Hz}}$



PARAMETER MEASUREMENT INFORMATION

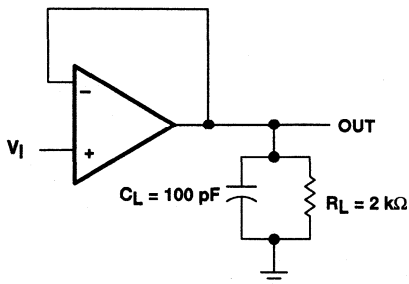


Figure 1. Unity-Gain Amplifier

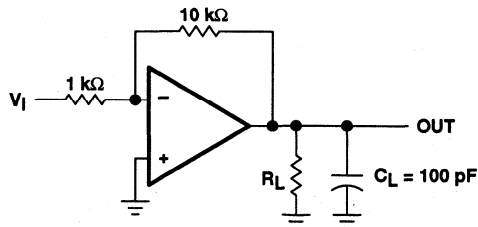


Figure 2. Gain-of-10 Inverting Amplifier

TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE	
V <sub>OM</sub>	Maximum peak output voltage	vs Frequency	3, 4, 5
		vs Free-air temperature	6
		vs Load resistance	7
		vs Supply voltage	8
A <sub>VD</sub>	Large-signal differential voltage amplification	vs Free-air temperature vs Frequency	9 10
P <sub>D</sub>	Total power dissipation	vs Free-air temperature	11
I <sub>CC</sub>	Supply current	vs Free-air temperature	12
		vs Supply voltage	13
I <sub>I(B)</sub>	Input bias current	vs Free-air temperature	14
		Pulse response	Large signal
V <sub>O</sub>	Output voltage	vs Elapsed time	16
CMRR	Common-mode rejection ratio	vs Free-air temperature	17
V <sub>n</sub>	Equivalent input noise voltage	vs Frequency	18
THD	Total harmonic distortion	vs Frequency	19
		Phase shift	vs Free-air temperature

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY

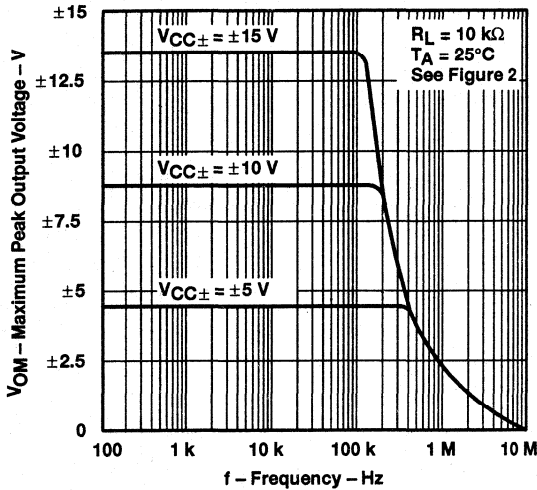


Figure 3

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY

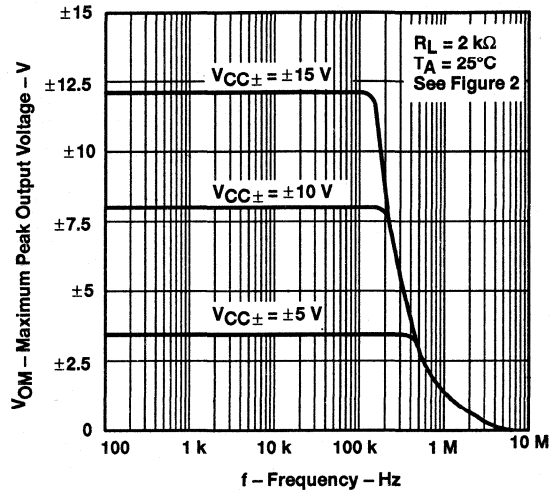


Figure 4

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY

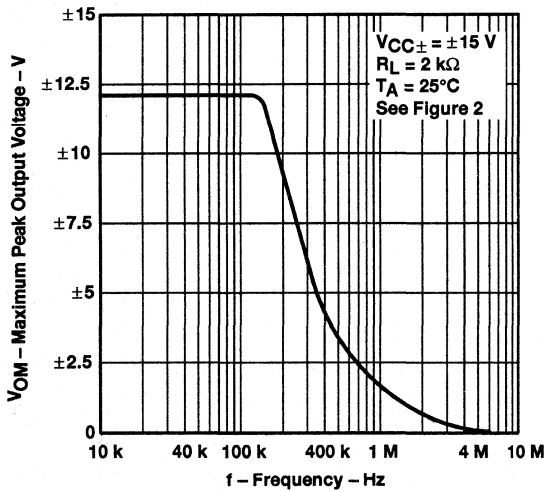


Figure 5

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

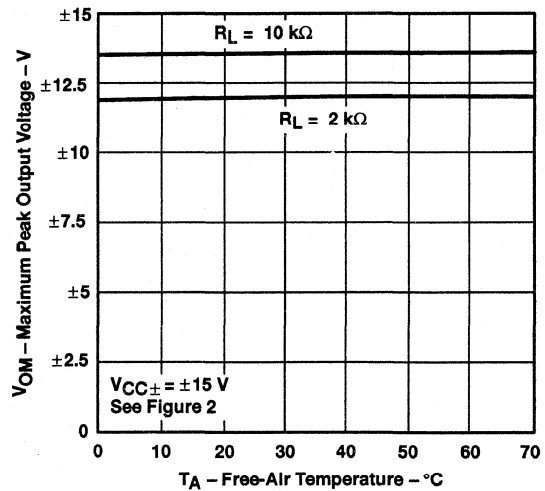


Figure 6

TYPICAL CHARACTERISTICS

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 LOAD RESISTANCE

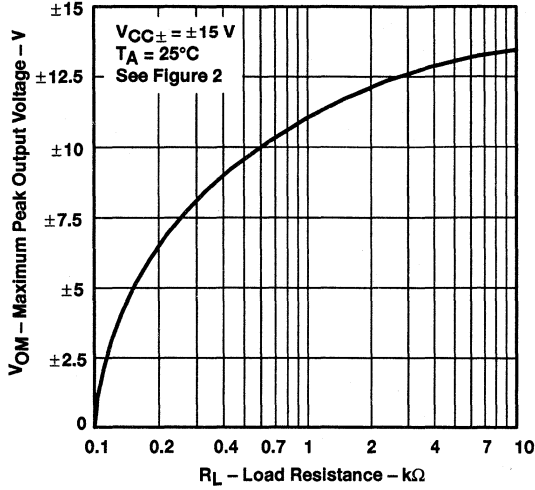


Figure 7

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE

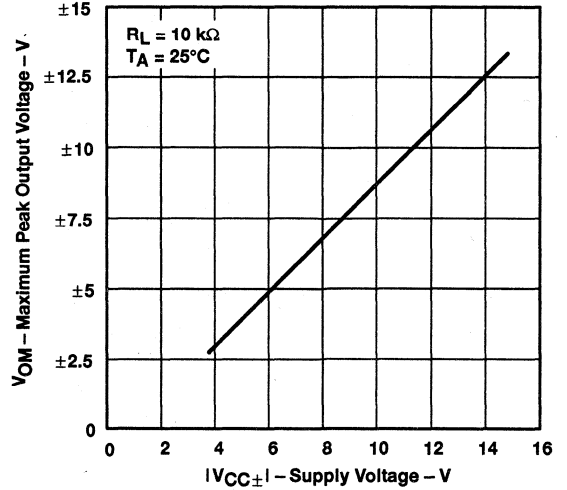


Figure 8

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

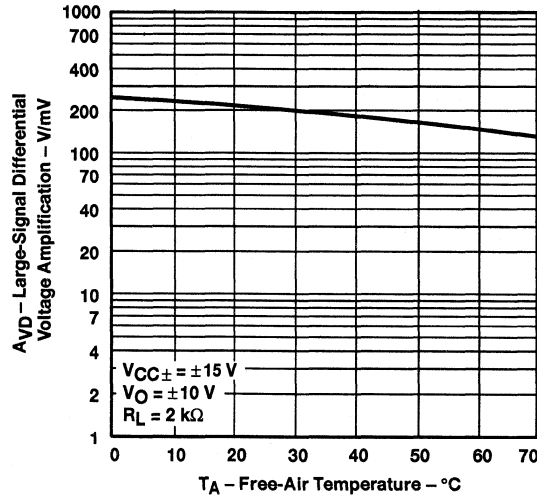


Figure 9

TYPICAL CHARACTERISTICS

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 VS  
 FREQUENCY

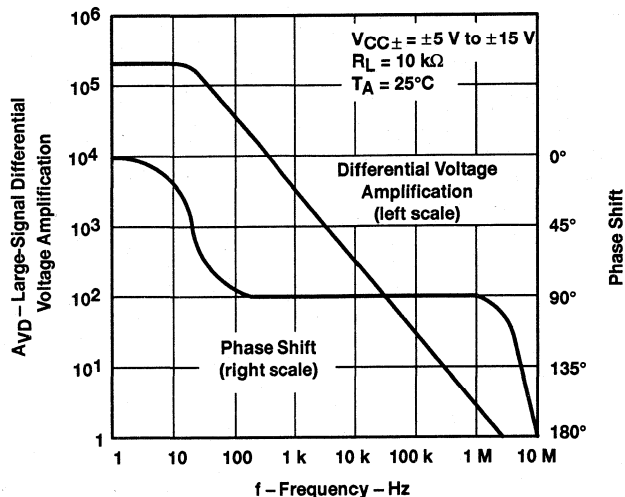


Figure 10

TOTAL POWER DISSIPATION  
 VS  
 FREE-AIR TEMPERATURE

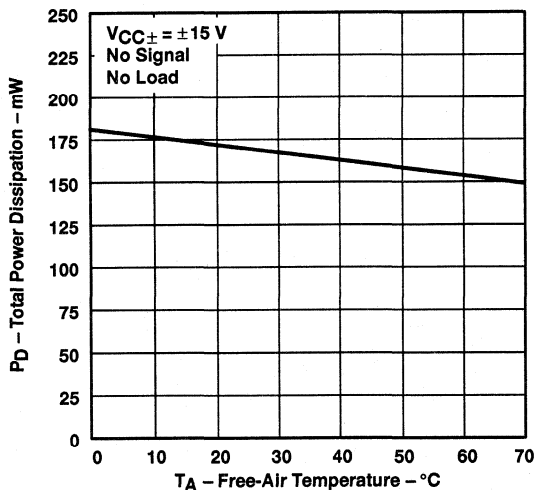


Figure 11

SUPPLY CURRENT (PER AMPLIFIER)  
 VS  
 FREE-AIR TEMPERATURE

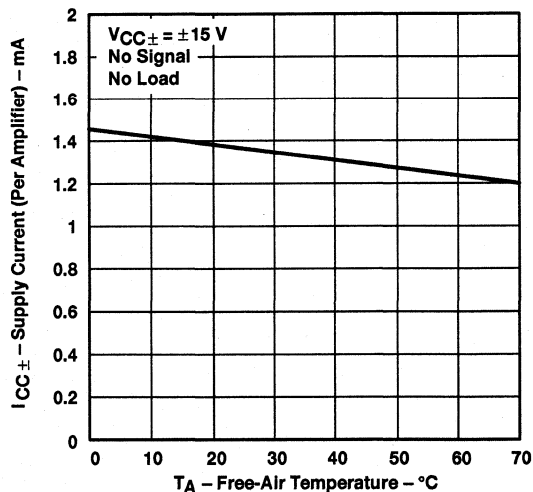


Figure 12



TYPICAL CHARACTERISTICS

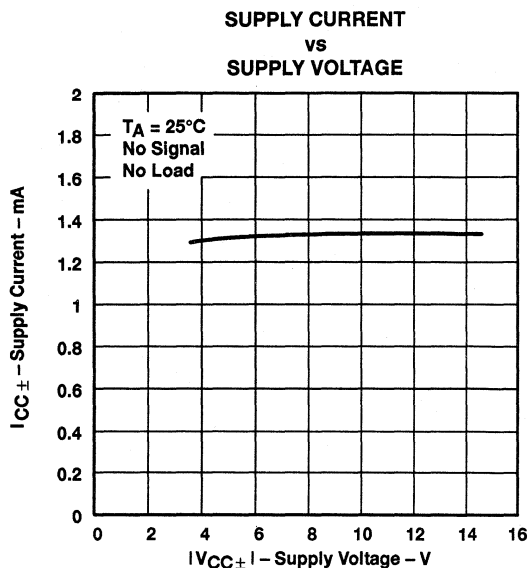


Figure 13

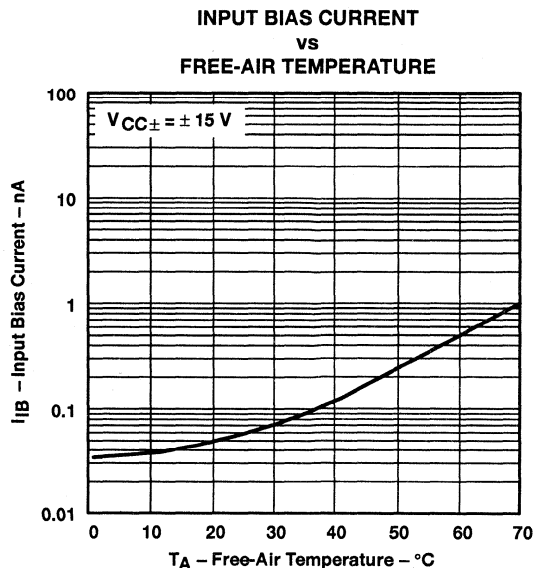


Figure 14

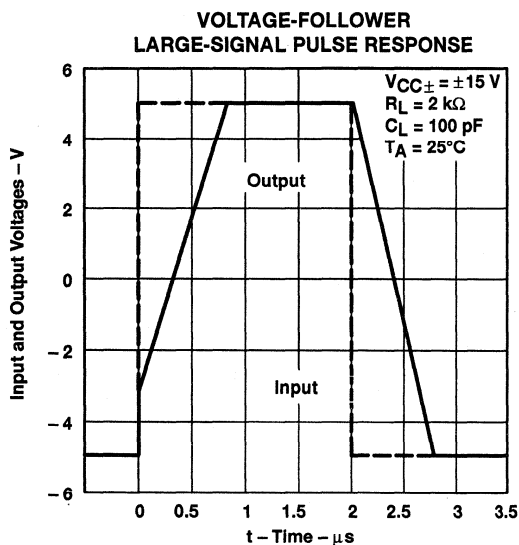


Figure 15

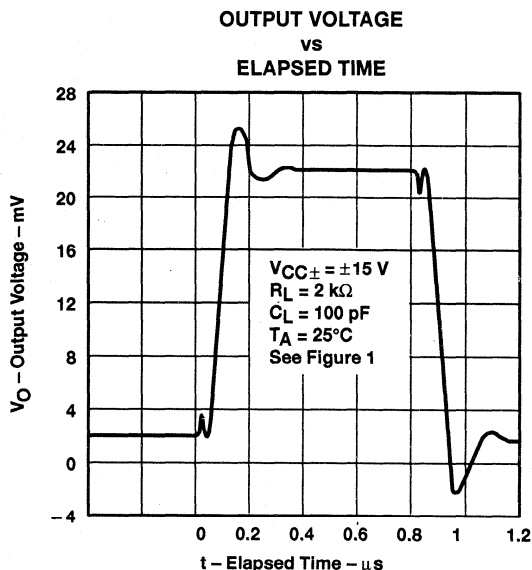


Figure 16

**TL084x2**  
**JFET-INPUT OCTAL OPERATIONAL AMPLIFIER**

SLOS136 – APRIL 1994

**TYPICAL CHARACTERISTICS**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**

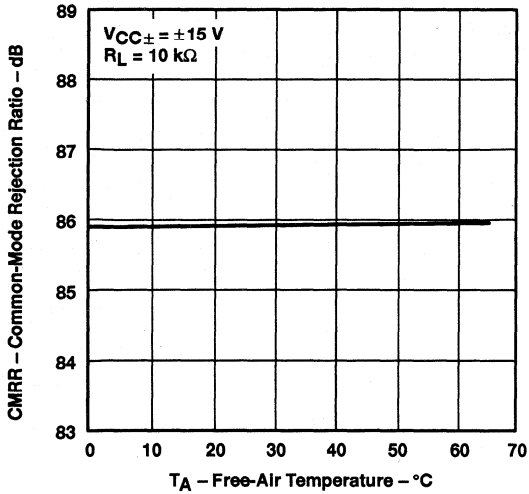


Figure 17

**EQUIVALENT INPUT NOISE VOLTAGE**  
**vs**  
**FREQUENCY**

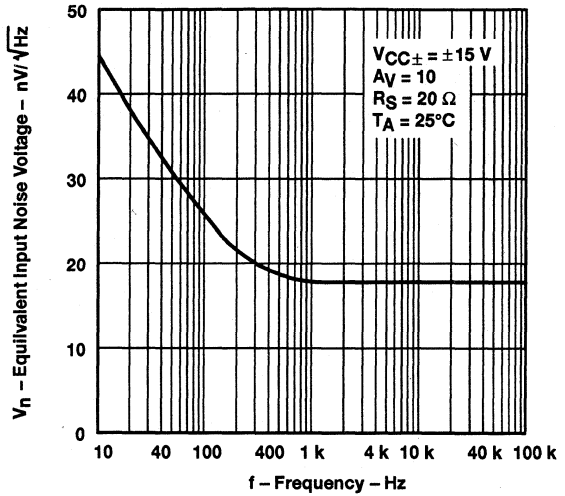


Figure 18

**TOTAL HARMONIC DISTORTION**  
**vs**  
**FREQUENCY**

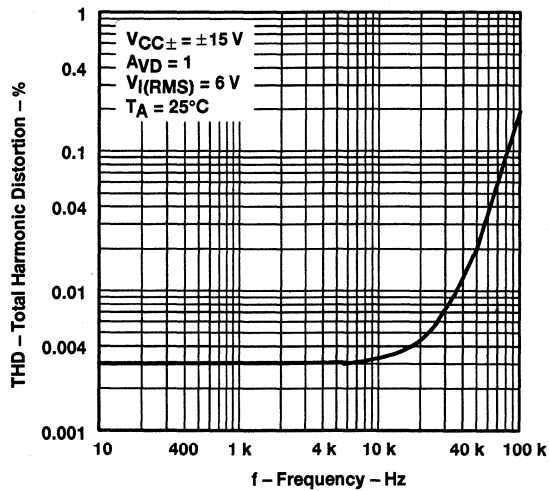


Figure 19

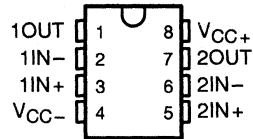


# TL2828Z, TL2828Y HIGH-TEMPERATURE DUAL OPERATIONAL AMPLIFIERS

SLOS104 – D3971, DECEMBER 1991

- **Operating Free-Air Temperature Range**  
–40°C to 150°C
- **Wide Range of Supply Voltages:**  
Single Supply  
or Dual Supply . . . 4 V to 30 V
- **Low Supply Current Drain Independent of Supply Voltage** . . . 0.7 mA Typ
- **Internal Frequency Compensation**
- **Low Input Bias and Offset Parameters**  
Input Offset Voltage . . . 3 mV Typ  
Input Offset Current . . . 2 nA Typ  
Input Bias Current . . . 15 nA Typ
- **Differential Input Voltage Range Equal to Maximum-Rated Supply Voltage** . . . 30 V
- **Open-Loop Differential Voltage Amplification** . . . 100 V/mV Typ

TL2828Z . . . D OR P PACKAGE  
(TOP VIEW)



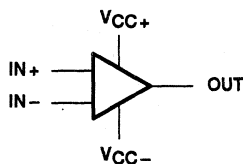
## description

The TL2828Z and TL2828Y devices consist of two independent high-gain frequency-compensated operational amplifiers that are designed specifically to operate over a wide range of voltages from a single supply. Operation from split supplies is also possible as long as the difference between the two supplies is 4 V to 30 V, and  $V_{CC}$  is at least 1.5 V more positive than the common-mode input voltage. The low supply current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational amplifier circuits that now can be implemented more easily in single-supply voltage systems. For example, the TL2828Z can be operated on automotive engine blocks directly off the standard 12-V supply with minimal electrical protection.

The TL2828Z is characterized for operation over the extended temperature range of –40°C to 150°C.

## symbol (each amplifier)



## AVAILABLE OPTIONS

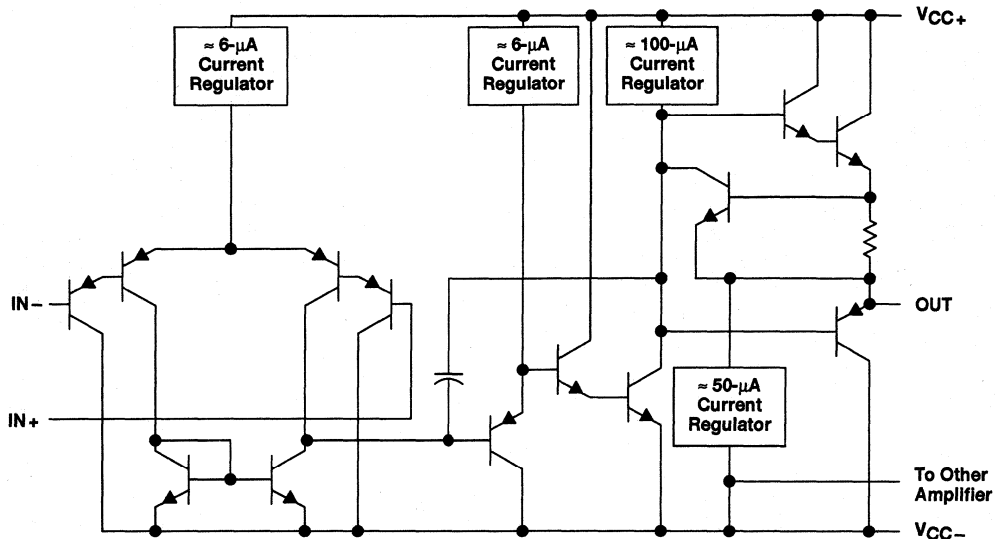
T <sub>A</sub>	V <sub>IO</sub> max at 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
–40°C to 150°C	7 mV	TL2828ZD	TL2828ZP	TL2828Y

The D packages are available taped and reeled. Add R suffix to device type (i.e., TL2828ZDR).  
The chip form is tested at T<sub>A</sub> = 25°C.

**TL2828Z, TL2828Y**  
**HIGH-TEMPERATURE DUAL**  
**OPERATIONAL AMPLIFIERS**

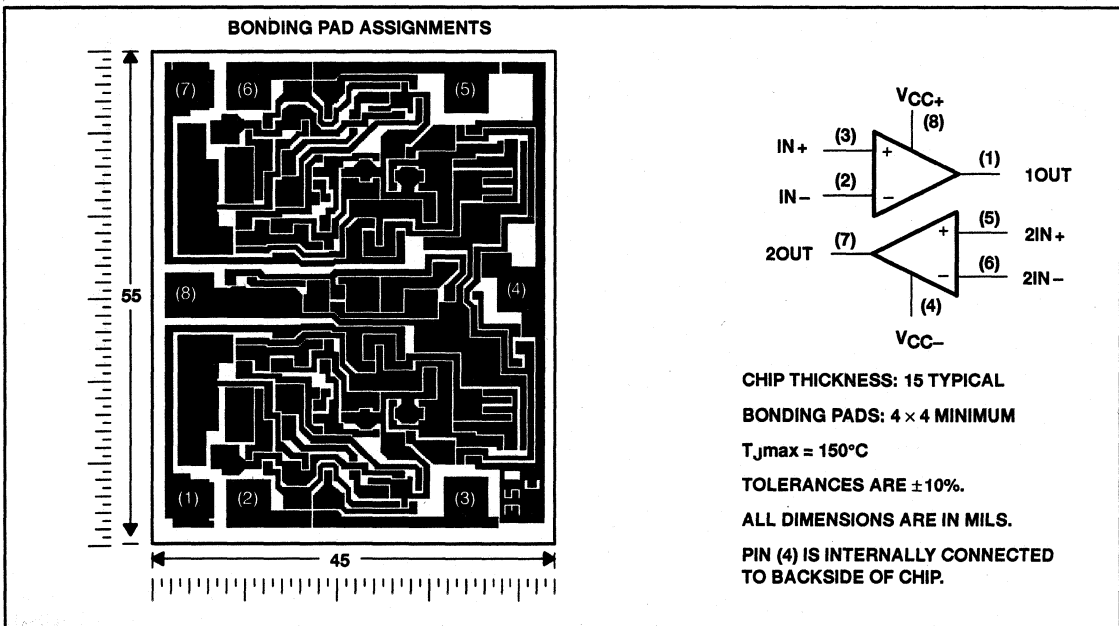
SLOS104 – D3971, DECEMBER 1991

equivalent schematic (each amplifier)



**TL2828Y chip information**

This chip, when properly assembled, displays characteristics similar to the TL2828Z. Thermal compression bonding may be used on the gold bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TL2828Z, TL2828Y HIGH-TEMPERATURE DUAL OPERATIONAL AMPLIFIERS

SLOS104 – D3971, DECEMBER 1991

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{CC+}$ (see Note 1)	16 V
Supply voltage, $V_{CC-}$	-16 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 32$ V
Input voltage range, $V_I$ (any input)	-16 V to 16 V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$	$\pm 40$ mA
Total current into $V_{CC+}$	60 mA
Total current out of $V_{CC-}$	60 mA
Duration of short-circuit at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	-40°C to 150°C
Storage temperature range	-65°C to 165°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$  when dual supplies are specified (e.g.,  $V_{CC\pm} = \pm 15$  V) and with respect to  $V_{CC-}$  when a single supply is specified (e.g.,  $V_{CC} = 5$  V).
  2. Differential voltages are at the noninverting input with respect to the noninverting input. Excessive current will flow if the input is below  $V_{CC-}$ .
  3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 105^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING	$T_A = 150^\circ\text{C}$ POWER RATING
D	812 mW	5.8 mW/°C	551 mW	348 mW	232 mW	87 mW
P	1120 mW	8.0 mW/°C	760 mW	480 mW	320 mW	120 mW

## recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, $V_{CC\pm}$	$\pm 2$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 2.5$ V	-2.5	0.5
	$V_{CC\pm} = \pm 15$ V	-15	13
Input voltage range, $V_I$	$V_{CC\pm} = \pm 2.5$ V	-2.5	0.5
	$V_{CC\pm} = \pm 15$ V	-15	13
Operating free-air temperature, $T_A$	-40	150	°C

**TL2828Z, TL2828Y**  
**HIGH-TEMPERATURE DUAL**  
**OPERATIONAL AMPLIFIERS**

SLOS104 – D3971, DECEMBER 1991

**electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2828Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $R_S = 50\ \Omega$	25°C	3		7	mV
		Full range			10	
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range	15			$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current		25°C	2		30	nA
		Full range			200	
$I_{IB}$ Input bias current		25°C	-15		-100	nA
	Full range			-500		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	25°C	0 to 3.5	0 to 3.5	V	
		Full range	0 to 3			
$V_{OH}$ High-level output voltage	$I_{OH} = 0.1\text{ mA}$	25°C	3.3	3.7	V	
		Full range	3.2			
	$I_{OH} = 1\text{ mA}$	25°C	3.3	3.6		
		Full range	3.2			
$V_{OL}$ Low-level output voltage	$I_{OL} = 0.1\text{ mA}$	25°C	0.8	0.6	V	
		Full range	1			
	$I_{OL} = 1\text{ mA}$	25°C	0.9	0.7		
		Full range	1.1			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to } 3.5\text{ V},$ $R_L = 2\text{ k}\Omega$	25°C	25	100	V/mV	
		Full range	0.7			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $R_S = 50\ \Omega$	25°C	65	80	dB	
		Full range	45			
kSVR Supply-voltage rejection ratio	$V_{CC} = 5\text{ V to } 30\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	65	100	dB	
		Full range	65			
$I_{CC}$ Supply current (total package)	$V_{IC} = 0,$ No load	25°C	0.7	1.2	mA	
		Full range	1.2			
$\Delta I_{CC}$ Supply current change over operating temperature range		Full range	140		$\mu\text{A}$	

† Full range is  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .



**TL2828Z, TL2828Y**  
**HIGH-TEMPERATURE DUAL**  
**OPERATIONAL AMPLIFIERS**

SLOS104 – D3971, DECEMBER 1991

**electrical characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2828Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $R_S = 50 \Omega$ $V_O = 0,$	25°C	3		7	mV
		Full range	10			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	15			$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current		25°C	2	30		nA
		Full range	200			
$I_{IB}$ Input bias current		25°C	-15	-100		nA
	Full range	-500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	25°C	-15 to 13.5		V	
		Full range	-15 to 13			
$V_{OM+}$ Maximum positive peak output voltage swing	$I_O = -0.1$ mA	25°C	13.2	14.1		V
		Full range	13.1			
	$I_O = -1$ mA	25°C	13.1	14		
		Full range	13			
	$I_O = -10$ mA	25°C	12.8	-13.6		
		Full range	12.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$I_O = 0.1$ mA	25°C	-13.7	-14.4		V
		Full range	-13.1			
	$I_O = 1$ mA	25°C	-13.6	-14.3		
		Full range	-13			
	$I_O = 7$ mA	25°C	-12.9	-13.8		
		Full range	-12.5			
$A_{VD}$ Large-signal differential voltage amplification	$R_L = 2$ k $\Omega$ , $V_O = -5$ V to 5 V	25°C	25	100		V/mV
		Full range	0.8			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $R_S = 50 \Omega$ $V_O = 1.4$ V,	25°C	65	75		dB
		Full range	50			
$k_{SVR}$ Supply-voltage rejection ratio	$V_{CC} = 5$ V to 30 V, $R_L = 50$ k $\Omega$	25°C	65	100		dB
		Full range	65			
$I_{CC}$ Supply current (total package)	$V_{IC} = 0,$ No load $V_O = 0,$	25°C	0.7		2	mA
		Full range	2			
$\Delta I_{CC}$ Supply current change over operating temperature range		Full range	140		$\mu\text{A}$	

† Full range is -40°C to 150°C.

**TL2828Z, TL2828Y**  
**HIGH-TEMPERATURE DUAL**  
**OPERATIONAL AMPLIFIERS**

SLOS104 – D3971, DECEMBER 1991

**operating characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15$  V**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2828Z			UNIT
			MIN	TYP	MAX	
SR+ Positive slew rate	$V_O = 1$ V to 4.5 V, $R_L = 2$ k $\Omega$ ‡, $A_{VD} = 1$ , $C_L = 100$ pF	25°C	0.15			V/ $\mu$ s
			Full range 0.1			
SR- Negative slew rate		25°C	0.15			
			Full range 0.1			
$V_n$ Equivalent input noise voltage	$f = 10$ Hz	25°C	39			nV/ $\sqrt{\text{Hz}}$
			$f = 10$ kHz	23		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1$ Hz to 10 Hz	25°C	0.9			$\mu$ V
$B_1$ Unity-gain bandwidth	$R_L = 10$ k $\Omega$ ‡, $C_L = 100$ pF	25°C	400			kHz
$\phi_m$ Phase margin	$R_L = 10$ k $\Omega$ ‡, $C_L = 100$ pF	25°C	60°			

† Full range is -40°C to 150°C.

‡  $R_L$  terminates at 0 V.

**electrical characteristics at  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TL2828Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $V_O = 0$ , $R_S = 50$ $\Omega$	3 7			mV
$I_{IO}$ Input offset current		2 30			nA
$I_{IB}$ Input bias current		-15 -100			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50$ $\Omega$	-15 to 13.5			V
$V_{OM+}$ Maximum positive peak output voltage swing	$I_O = -0.1$ mA	13.2	14.1		V
	$I_O = -1$ mA	13.1	14		
	$I_O = -10$ mA	12.8	13.6		
$V_{OM-}$ Maximum negative peak output voltage swing	$I_O = 0.1$ mA	-13.7	-14.4		V
	$I_O = 1$ mA	-13.6	-14.3		
	$I_O = 10$ mA	-12.9	-13.8		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1$ V to -1.5 V, $R_L = 2$ k $\Omega$	25	100		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = 0$ to 28 V, $V_O = 1.4$ V, $R_S = 50$ $\Omega$	65	75		dB
$k_{SVR}$ Supply-voltage rejection ratio	$V_{CC} = 5$ V to 30 V, $V_O = 1.4$ V, $R_L = 10$ k $\Omega$	65	100		dB
$I_{CC}$ Supply-current (total package)	$V_{IC} = 0$ , $V_O = 0$ , No load	0.7 2			mA



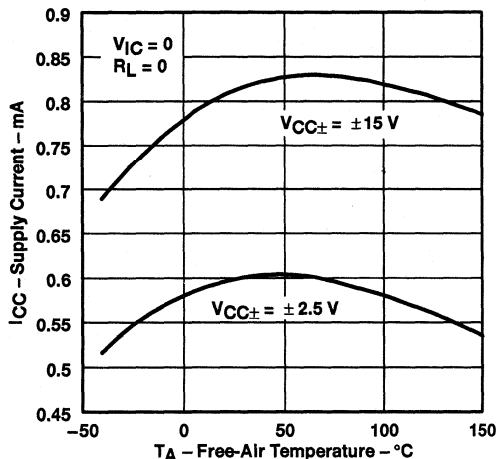


# TL2829Z, TL2829Y HIGH-TEMPERATURE QUADRUPLE OPERATIONAL AMPLIFIERS

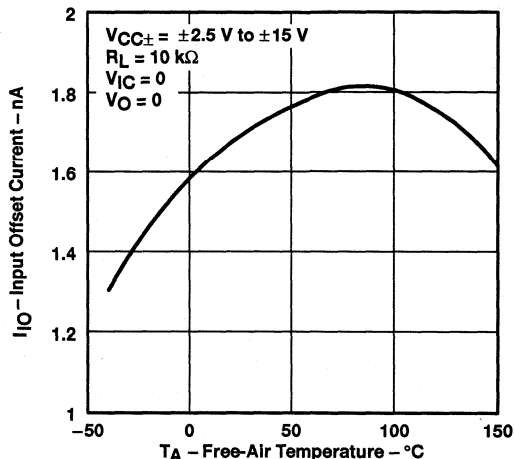
SLOS067A – APRIL 1991 – REVISED MARCH 1993

- Free-Air Operating Temperature Range  
–40°C to 150°C
- Wide Range of Supply Voltages:  
Single Supply . . . 4 V to 30 V  
or Dual Supplies
- Low Supply Current Drain independent of  
Supply Voltage . . . 0.8 mA
- Internal Frequency Compensation
- Low Input Bias and Offset Parameters at  
25°C  
Input Offset Voltage . . . 3 mV Typ  
Input Offset Current . . . 2 nA Typ  
Input Bias Current . . . 15 nA Typ
- Differential Input Voltage Range Equal to  
Maximum-Rated Supply Voltage . . . 30 V
- Open-Loop Differential Voltage  
Amplification . . . 100 V/mV Typ at 25°C

**SUPPLY CURRENT  
vs  
FREE-AIR TEMPERATURE**



**INPUT OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE**



## description

These devices consist of four independent, high-gain frequency-compensated operational amplifiers that are designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is also possible as long as the difference between the two supplies is 4 V to 30 V, and  $V_{CC}$  is at least 1.5 V more positive than the input common-mode voltage. The low supply current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational amplifier circuits that now can be implemented more easily in single-supply-voltage systems. For example, the TL2829 can be operated on automotive engine blocks directly off the standard 12-V supply with minimal electrical protection.

The TL2829 is characterized for operation over the extended temperature range of –40°C to 150°C.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (N)	
–40°C to 150°C	7 mV	TL2829ZD	TL2829ZN	TL2829Y

The D packages are available taped and reeled. Add R suffix to device type (i.e., TL2829ZDR).

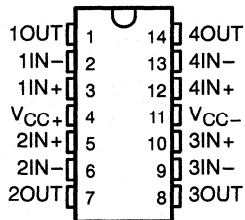
PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



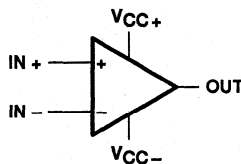
# TL2829Z, TL2829Y HIGH-TEMPERATURE QUADRUPLE OPERATIONAL AMPLIFIERS

SLOS067A - APRIL 1991 - REVISED MARCH 1993

TL2829Z . . . D OR N PACKAGE  
(TOP VIEW)

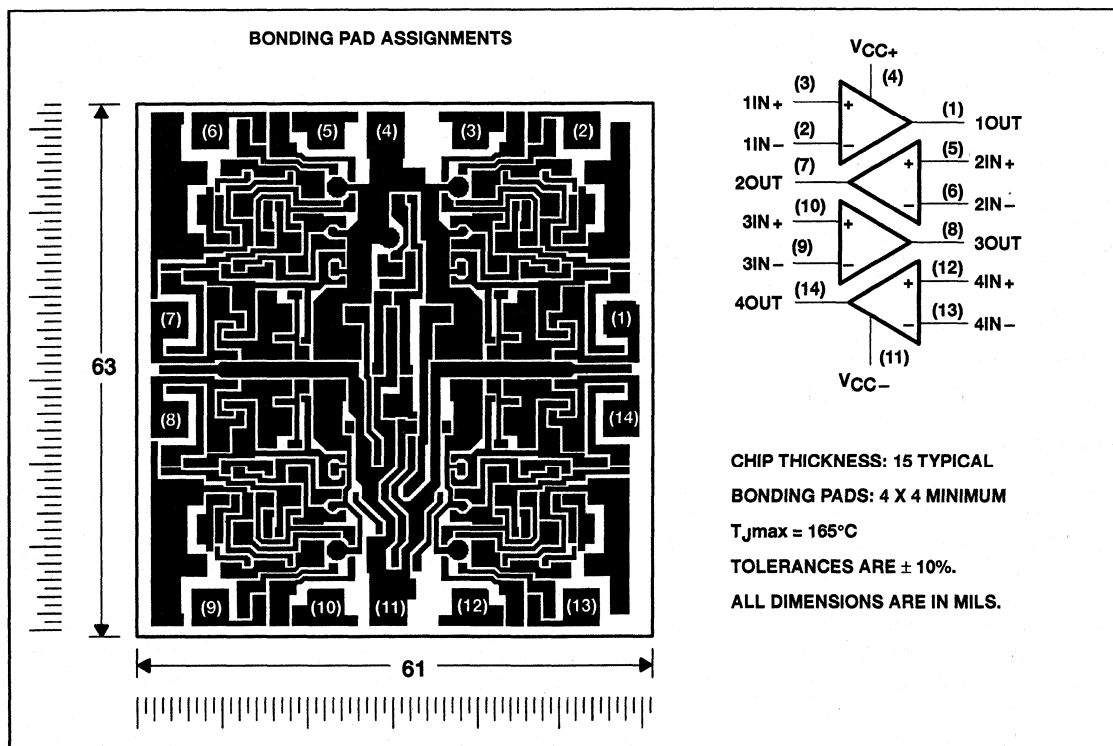


symbol (each amplifier)

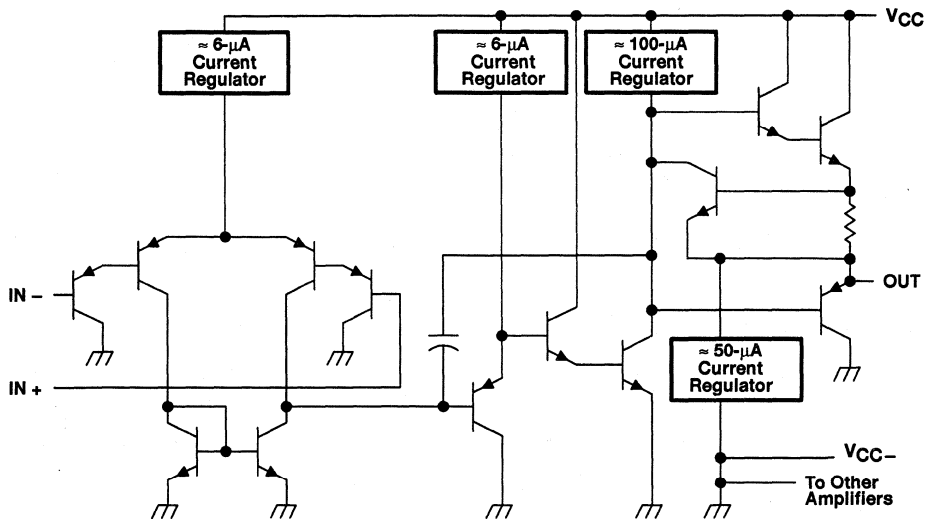


## TL2829Y chip information

This chip, properly assembled, displays characteristics similar to the TL2829. Thermal compression bonding may be used on the gold bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



equivalent schematic (each amplifier)



COMPONENT COUNT (total device)	
Epi-FET	1
Diodes	4
Resistors	11
Transistors	95
Capacitors	4

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{CC+}$ (see Note 1)	16 V
Supply voltage, $V_{CC-}$ (see Note 1)	-16 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 32$ V
Input voltage range, $V_I$ (any input)	-16 to 16 V
Input current, $I_I$ (each input)	$\pm 1$ mA
Output current, $I_O$	$\pm 40$ mA
Total current into $V_{CC+}$	60 mA
Total current out of $V_{CC-}$	60 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	-40°C to 150°C
Storage temperature range	-65°C to 165°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{CC+}$  and  $V_{CC-}$  when dual supplies are specified (e.g.,  $V_{CC\pm} = \pm 15$  V) and with respect to  $V_{CC-}$  when a single supply is specified (e.g.,  $V_{CC} = 5$  V).  
 2. Differential voltages are at the noninverting input with respect to the inverting input. Excessive current will flow if input is brought below  $V_{CC}$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 100^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING	$T_A = 150^\circ\text{C}$ POWER RATING
D	1064 mW	7.6 mW/°C	722 mW	494 mW	304 mW	114 mW
N	1764 mW	12.6 mW/°C	1197 mW	819 mW	504 mW	189 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{CC\pm}$	$\pm 2$	$\pm 15$	V
Common-mode input voltage, $V_{IC}$	$V_{CC\pm} = \pm 2.5$ V	-2.5	0.5
	$V_{CC\pm} = \pm 15$ V	-15	13
Input voltage range, $V_I$	$V_{CC\pm} = \pm 2.5$ V	-2.5	0.5
	$V_{CC\pm} = \pm 15$ V	-15	13
Operating free-air temperature, $T_A$	-40	150	°C



**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

**electrical characteristics at specified free-air temperature,  $V_{CC} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2829Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega$ $V_{IC} = 0,$	25°C		3	7	mV
		Full range			10	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		15		$\mu\text{V}/^\circ\text{C}$
		25°C		2.0	30	nA
$I_{IO}$ Input offset current		Full range			200	
		$I_{IB}$ Input bias current	25°C		-12	-100
Full range					-500	
$V_{ICR}$ Common-mode input voltage range		$R_S = 50\ \Omega$	25°C	0 to 3.5	0 to 3.5	V
	Full range		0 to 3			
$V_{OH}$ High-level output voltage	$I_{OH} = 0.1\text{ mA}$	25°C	3.3	3.7	V	
		Full range	3.2			
	$I_{OH} = 1\text{ mA}$	25°C	3.3	3.6		
		Full range	3.2			
$V_{OL}$ Low-level output voltage	$I_{OL} = 0.1\text{ mA}$	25°C	0.8	0.6	V	
		Full range	1			
	$I_{OL} = 1\text{ mA}$	25°C	0.9	0.7		
		Full range	1.1			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to } 3.5\text{ V},$ $R_L = 2\text{ k}\Omega$	25°C	25	60	V/mV	
		Full range	0.8			
$CMRR$ Common-mode rejection ratio	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega$ $V_{IC} = V_{ICRmin},$	25°C	65	81	dB	
		Full range	50			
$k_{SVR}$ Supply-voltage rejection ratio	$V_{CC} = 5\text{ V to } 30\text{ V},$ $V_O = 1.4\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	65	103	dB	
		Full range	65			
$I_{CC}$ Supply current (total package)	$V_O = 2.5\text{ V},$ No load $V_{IC} = 0,$	25°C		0.6	1.2	mA
		Full range			1.2	
$\Delta I_{CC}$ Supply current change over operating temperature range		Full range		140	$\mu\text{A}$	

† Full range is  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

**electrical characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2829Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0,$ $V_{IC} = 0,$ $R_S = 50 \Omega,$	25°C	3		7	mV
		Full range			10	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	15			$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current		25°C	2		30	nA
		Full range			200	
$I_{IB}$ Input bias current		25°C	-15		-100	nA
		Full range			-500	
$V_{ICR}$ Common-mode input voltage range		$R_S = 50 \Omega$	25°C	-15 to 13.5		V
	Full range		-15 to 13			
$V_{OM+}$ Maximum positive peak output voltage range	$I_O = -0.1$ mA	25°C	13.2	14.1	V	
		Full range	13.1			
	$I_O = -1$ mA	25°C	13.1	14		
		Full range	13			
	$I_O = -10$ mA	25°C	12.8	13.6		
		Full range	12.7			
$V_{OM-}$ Maximum negative peak output voltage range	$I_O = 0.1$ mA	25°C	-13.7	-14.4	V	
		Full range	-13.1			
	$I_O = 1$ mA	25°C	-13	-14.3		
		Full range	-13			
	$I_O = 10$ mA	25°C	-12.9	-13.8		
		Full range	-12.9			
$A_{VD}$ Large-signal differential voltage amplification	$R_L = 2$ k $\Omega,$ $V_O = -5$ V to 5 V	25°C	25	210	V/mV	
		Full range	5			
CMRR Common-mode rejection ratio	$V_O = 1.4$ V, $R_S = 50 \Omega,$ $V_{IC} = V_{ICRmin}$	25°C	65	75	dB	
		Full range	50			
$k_{SVR}$ Supply-voltage rejection ratio	$V_{CC} = 5$ V to 30 V, $R_L = 10$ k $\Omega,$ $V_O = 1.4$ V	25°C	65	103	dB	
		Full range	65			
$I_{CC}$ Supply current (total package)	$V_O = 0,$ No load $V_{IC} = 0,$	25°C	0.8	3	mA	
		Full range	3			
$\Delta I_{CC}$ Supply current change over operating temperature range		Full range	140		$\mu\text{A}$	

† Full range is -40°C to 150°C.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

**operating characteristics at specified free-air temperature,  $V_{CC\pm} = \pm 15$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TL2829Z			UNIT
			MIN	TYP	MAX	
SR+ Positive slew rate	$V_O = 1$ V to 4.5 V, $A_{VD} = 1$ , $R_L = 2$ k $\Omega$ ‡, $C_L = 100$ pF	25°C	0.2			V/ $\mu$ s
		Full range	0.1			
25°C		0.25				
Full range		0.2				
SR- Negative slew rate						
$V_n$ Equivalent input noise voltage	$f = 10$ Hz	25°C	39			nV/ $\sqrt{\text{Hz}}$
	$f = 10$ kHz	25°C	23			
$V_n(\text{PP})$ Peak-to-peak equivalent input noise voltage	$f = 0.1$ Hz to 10 Hz	25°C	0.9			$\mu$ V
$B_n$ Unity-gain bandwidth	$R_L = 10$ k $\Omega$ ‡, $C_L = 100$ pF	25°C	400			kHz
$\phi_m$ Phase margin at unity gain	$R_L = 10$ k $\Omega$ ‡, $C_L = 100$ pF	25°C	60°			

† Full range is  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .

‡  $R_L$  terminates at 0 V.

**electrical characteristics at  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TL2829Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 0$ , $V_{IC} = 0$ , $R_S = 50$ $\Omega$	3 7			mV
$I_{IO}$ Input offset current		2 30			nA
$I_{IB}$ Input bias current		-15 -100			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50$ $\Omega$	-15 to 13.5			V
$V_{OM+}$ Maximum positive peak output voltage range	$I_O = -0.1$ mA	13.2	14.1		V
	$I_O = -1$ mA	13.1	14		
	$I_O = -10$ mA	12.8	13.6		
$V_{OM-}$ Maximum negative peak output voltage range	$I_O = 0.1$ mA	-13.7	-14.4		V
	$I_O = 1$ mA	-13.6	-14.3		
	$I_O = 10$ mA	-12.9	-13.8		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1$ V to $-1.5$ V, $R_L = 2$ k $\Omega$	25	210		V/mV
CMRR Common-mode rejection ratio	$V_O = 1.4$ V, $V_{IC} = 0$ V to 28 V, $R_S = 50$ $\Omega$	65	75		dB
$k_{SVR}$ Supply-voltage rejection ratio	$V_{CC} = 5$ V to 30 V, $V_O = 1.4$ V, $R_L = 10$ k $\Omega$	65	103		dB
$I_{CC}$ Supply current (total package)	$V_O = 0$ , No load, $V_{IC} = 0$ ,	0.8 3			mA

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A—APRIL 1991—REVISED MARCH 1993

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			FIGURE
$I_{IO}$	Input offset current	vs Free-air temperature	1
$I_{IB}$	Input bias current	vs Free-air temperature ( $V_{CC} = \pm 2.5$ V)	2
		vs Free-air temperature ( $V_{CC} = \pm 15$ V)	3
$V_{OM+}$	Maximum positive peak output voltage swing	vs Free-air temperature ( $V_{CC} = \pm 2.5$ V)	4
		vs Free-air temperature ( $V_{CC} = \pm 15$ V)	5
$V_{OM-}$	Maximum negative peak output voltage swing	vs Free-air temperature ( $V_{CC} = \pm 2.5$ V)	6
		vs Free-air temperature ( $V_{CC} = \pm 15$ V)	7
$I_{OS}$	Short-circuit output current	vs Free-air temperature ( $V_{ID} = 1$ V)	8
		vs Free-air temperature ( $V_{ID} = -1$ V)	9
$A_{VD}$	Differential voltage amplification	vs Free-air temperature	10
$CMRR$	Common-mode rejection ratio	vs Free-air temperature	11
$k_{SVR}$	Supply-voltage rejection ratio	vs Free-air temperature	12
$I_{CC}$	Supply current	vs Free-air temperature	13
$SR+$	Positive slew rate	vs Free-air temperature	14
$SR-$	Negative slew rate	vs Free-air temperature	15
	Equivalent input noise voltage	Over a 10-second period	16



TYPICAL CHARACTERISTICS

INPUT OFFSET CURRENT  
 vs  
 FREE-AIR TEMPERATURE

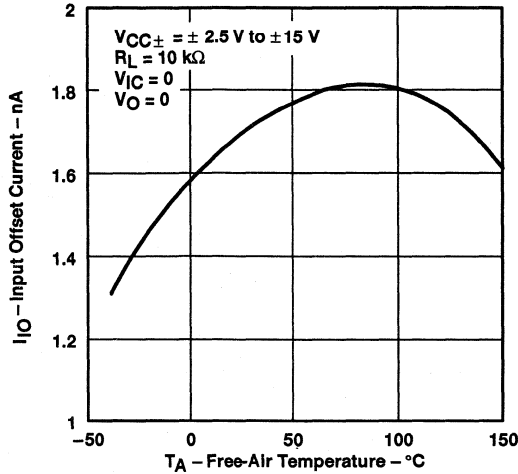


Figure 1

INPUT BIAS CURRENT  
 vs  
 FREE-AIR TEMPERATURE

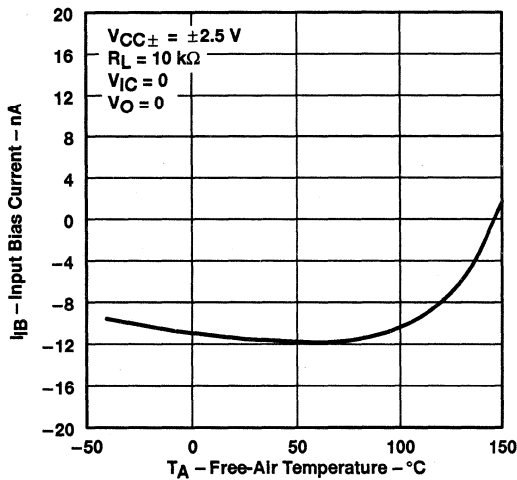


Figure 2

INPUT BIAS CURRENT  
 vs  
 FREE-AIR TEMPERATURE

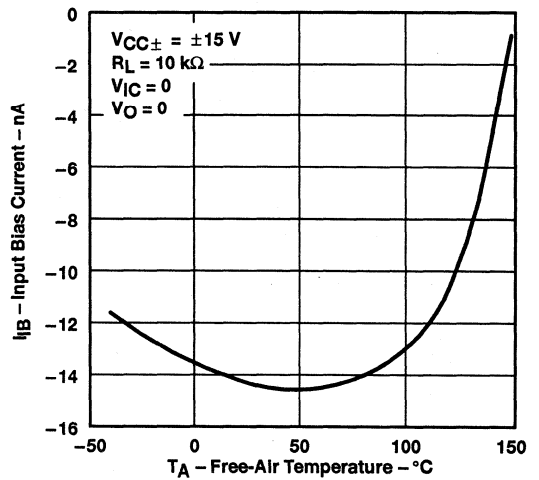


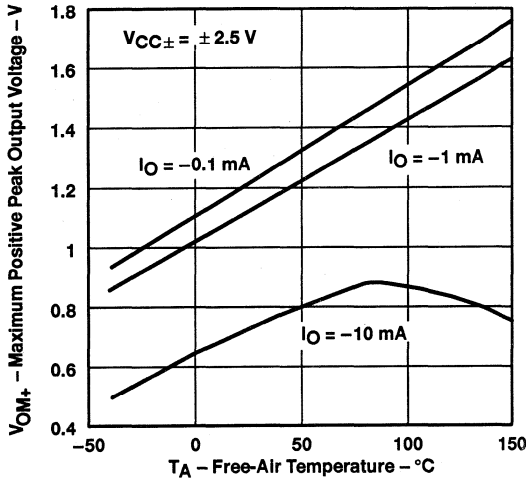
Figure 3

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

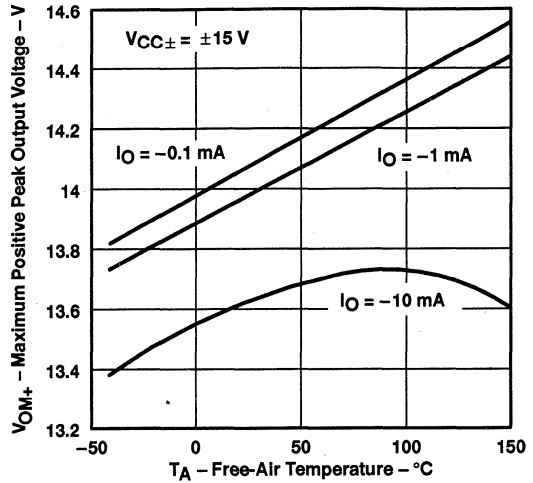
**TYPICAL CHARACTERISTICS**

**MAXIMUM POSITIVE PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



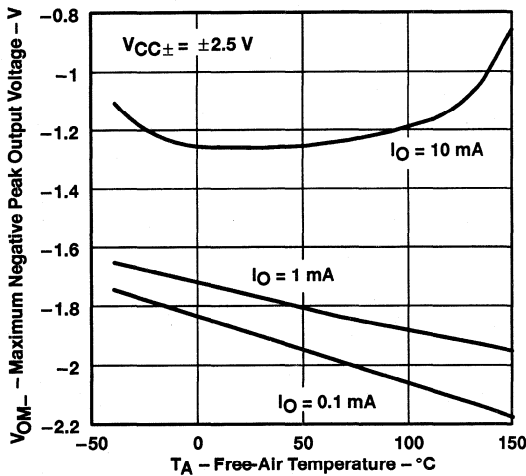
**Figure 4**

**MAXIMUM POSITIVE PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



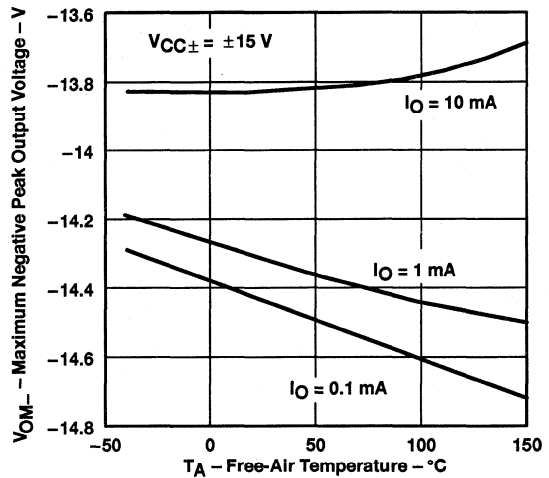
**Figure 5**

**MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 6**

**MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 7**



TYPICAL CHARACTERISTICS

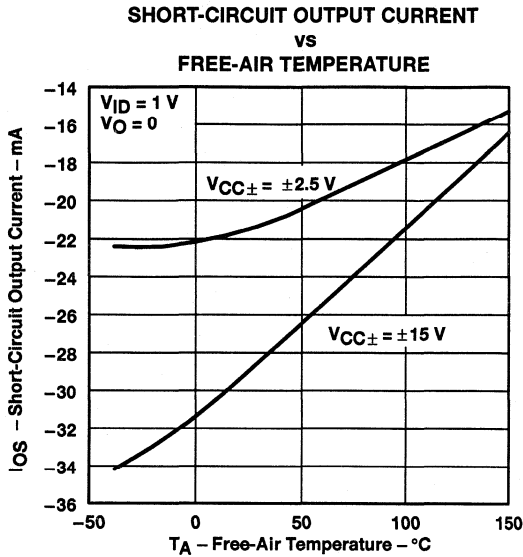


Figure 8

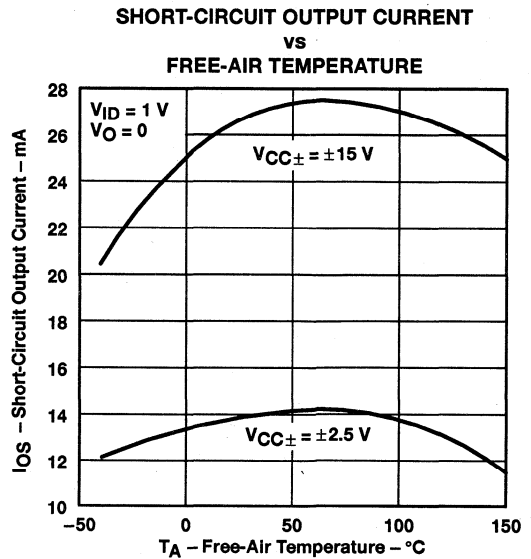


Figure 9

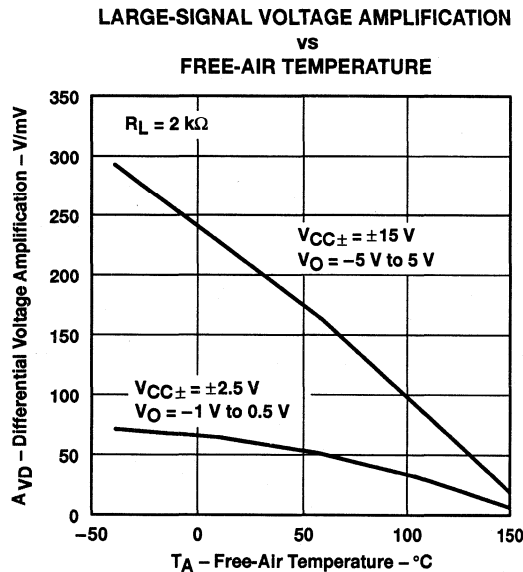


Figure 10

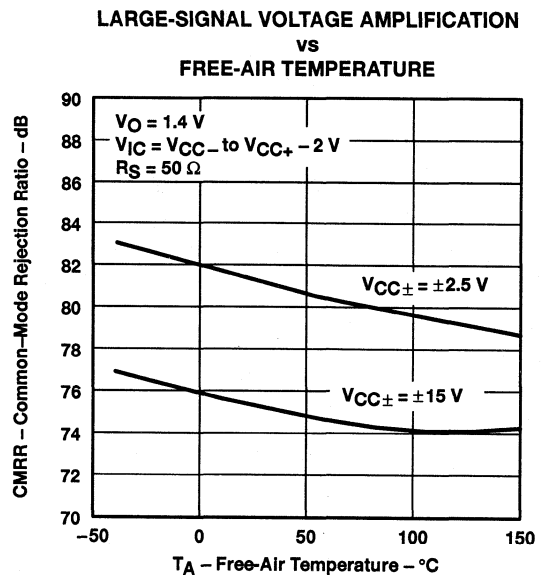


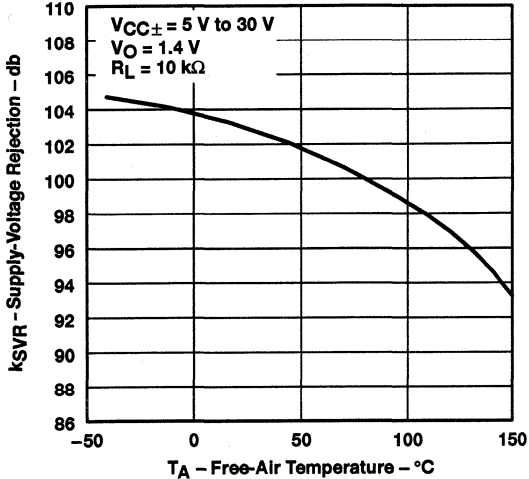
Figure 11

**TL2829Z, TL2829Y**  
**HIGH-TEMPERATURE QUADRUPLE**  
**OPERATIONAL AMPLIFIERS**

SLOS067A – APRIL 1991 – REVISED MARCH 1993

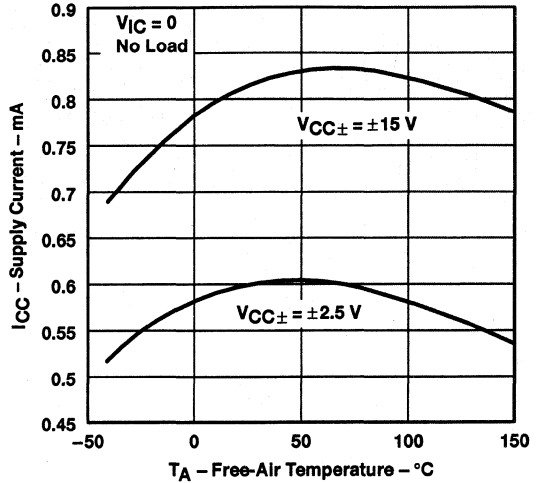
**TYPICAL CHARACTERISTICS**

**SUPPLY-VOLTAGE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



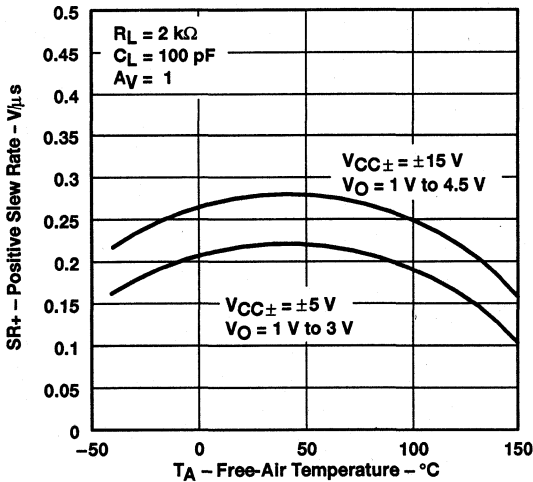
**Figure 12**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



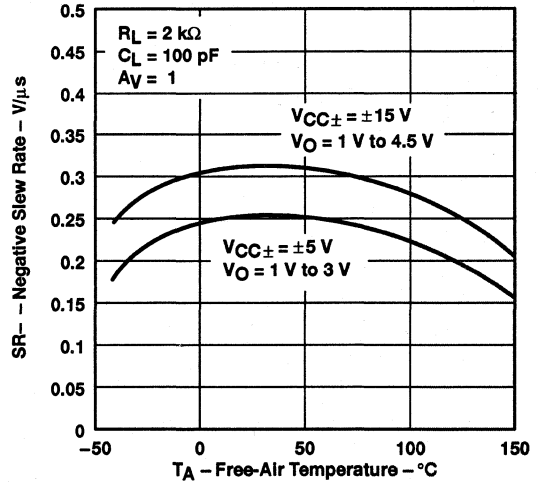
**Figure 13**

**POSITIVE SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 14**

**NEGATIVE SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 15**



TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE  
OVER A 10-SECOND PERIOD

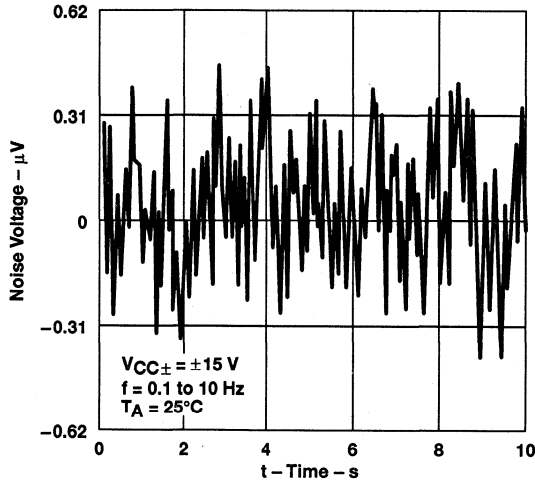


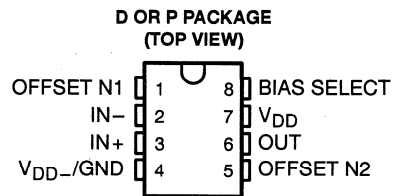
Figure 16



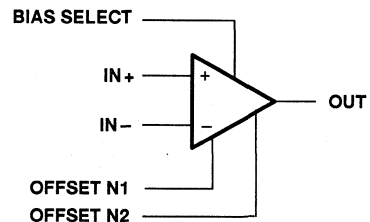
# TLC251, TLC251A, TLC251B, TLC251Y LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS001E – JULY 1983 – REVISED AUGUST 1994

- **Wide Range of Supply Voltages**  
1.4 V to 16 V
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Range**  
Includes the Negative Rail
- **Low Noise . . . 30 nV/√Hz Typ at 1 kHz**  
(High Bias)
- **ESD Protection Exceeds 2000 V Per**  
MIL-STD-883C, Method 3015.1



symbol



## description

The TLC251C, TLC251AC, and TLC251BC are low-cost, low-power programmable operational amplifiers designed to operate with single or dual supplies. Unlike traditional metal-gate CMOS operational amplifiers, these devices utilize Texas Instruments silicon-gate LinCMOS™ process, giving them stable input offset voltages without sacrificing the advantages of metal-gate CMOS.

This series of parts is available in selected grades of input offset voltage and can be nulled with one external potentiometer. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this family is ideally suited for battery-powered or energy-conserving applications. A bias-select pin can be used to program one of three ac performance and power-dissipation levels to suit the application. The series features operation down to a 1.4-V supply and is stable at unity gain.

These devices have internal electrostatic-discharge (ESD) protection circuits that prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC251C series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS™ operational amplifiers without the power penalties of traditional bipolar devices. Remote and inaccessible equipment applications are possible using the low-voltage and low-power capabilities of the TLC251C series.

In addition, by driving the bias-select input with a logic signal from a microprocessor, these operational amplifiers can have software-controlled performance and power consumption. The TLC251C series is well suited to solve the difficult problems associated with single battery and solar cell-powered applications.

The TLC251C series is characterized for operation from 0°C to 70°C.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
0°C to 70°C	10 mV	TLC251CD	TLC251CP	TLC251Y
	5 mV	TLC251ACD	TLC251ACP	—
	2 mV	TLC251BCD	TLC251BCP	—

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC251CDR). Chips are tested at 25°C.

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



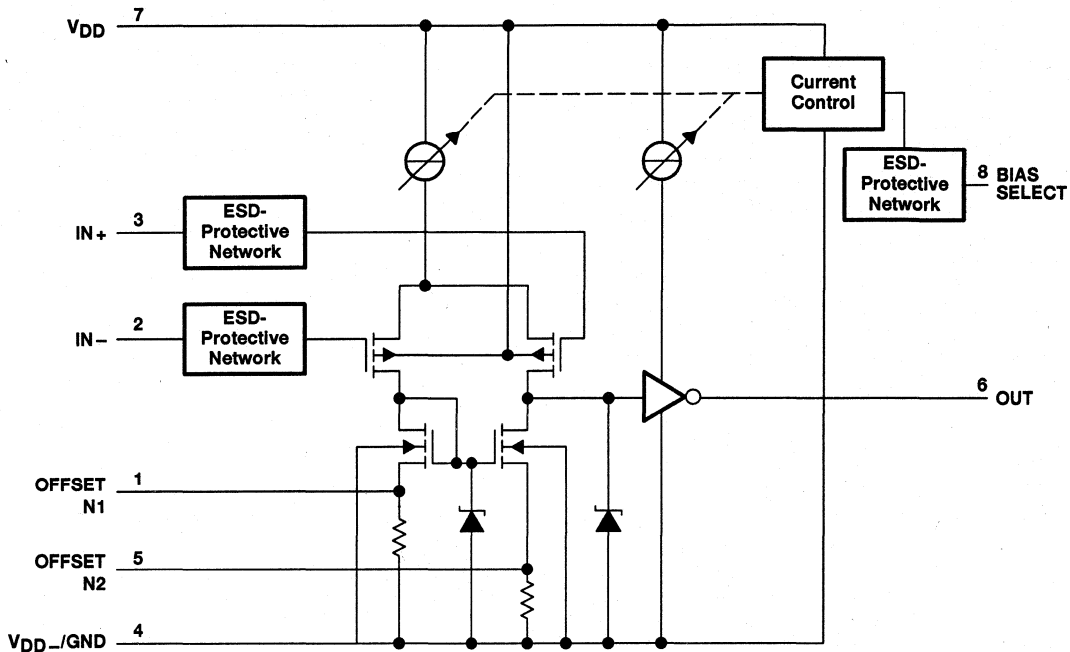
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2-411

**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**  
SLOS001E – JULY 1983 – REVISED AUGUST 1994

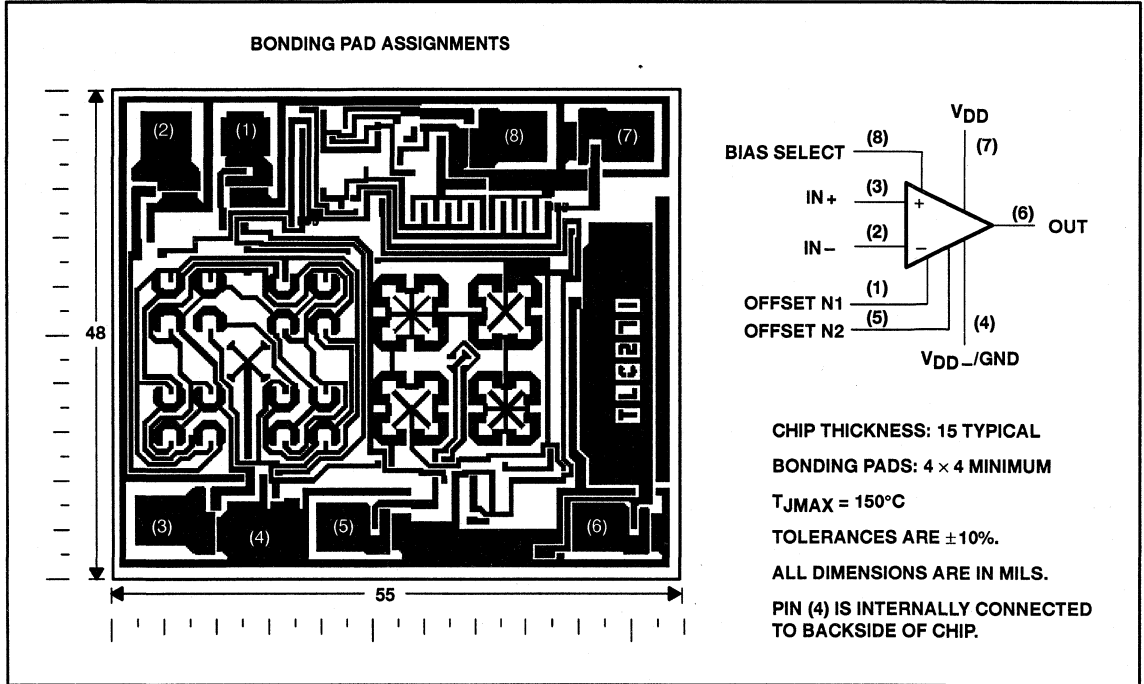
**schematic**





**TLC251Y chip information**

These chips, properly assembled, display characteristics similar to the TLC251C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 18$ V
Input voltage range, $V_I$ (any input)	-0.3 V to 18 V
Duration of short circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to  $V_{DD-}/GND$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD}$	1.4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 1.4$ V	0	0.2
	$V_{DD} = 5$ V	-0.2	4
	$V_{DD} = 10$ V	-0.2	9
	$V_{DD} = 16$ V	-0.2	14
Operating free-air temperature, $T_A$	0	70	°C
Bias-select voltage	See Application Information		



**HIGH-BIAS MODE**

**electrical characteristics at specified free-air temperature**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC251C, TLC251AC, TLC251BC						UNIT	
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V				
				MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub>	Input offset voltage	TLC251C  TLC251AC  TLC251BC	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 10 kΩ	25°C	1.1		10	1.1		10	mV
				Full range			12			12	
				25°C	0.9		5	0.9		5	
				Full range			6.5			6.5	
				25°C	0.34		2	0.39		2	
Full range			3			3					
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.8			2			μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1			pA	
			70°C	7		300	7		300		
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7			pA	
			70°C	40		600	50		600		
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V	
			Full range	-0.2 to 3.5			-0.2 to 8.5			V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 10 kΩ	25°C	3.2	3.8		8	8.5		V	
			0°C	3	3.8		7.8	8.5			
			70°C	3	3.8		7.8	8.4			
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50	0		50	mV	
			0°C	0		50	0		50		
			70°C	0		50	0		50		
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 6	25°C	5	23		10	36		V/mV	
			0°C	4	27		7.5	42			
			70°C	4	20		7.5	32			
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	80		65	85		dB	
			0°C	60	84		60	88			
			70°C	60	85		60	88			
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C	65	95		65	95		dB	
			0°C	60	94		60	94			
			70°C	60	96		60	96			
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = 0	25°C	-1.4			-1.9			μA	
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	675	1600		950	2000		μA	
			0°C	775	1800		1125	2200			
			70°C	575	1300		750	1700			

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.

**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS		T <sub>A</sub>	TLC251C, TLC251AC, TLC251BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 20 pF	V <sub>I(PP)</sub> = 1 V	25°C	3.6		V/μs	
			0°C	4			
			70°C	3			
		V <sub>I(PP)</sub> = 2.5 V	25°C	2.9			
			0°C	3.1			
			70°C	2.5			
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 Ω	25°C	25		nV/√Hz		
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>OH</sub> , C <sub>L</sub> = 20 pF, R <sub>L</sub> = 10 kΩ	25°C	320		kHz		
		0°C	340				
		70°C	260				
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF	25°C	1.7		MHz		
		0°C	2				
		70°C	1.3				
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, f = B <sub>1</sub> , C <sub>L</sub> = 20 pF	25°C	46°				
		0°C	47°				
		70°C	44°				

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS		T <sub>A</sub>	TLC251C, TLC251AC, TLC251BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 20 pF	V <sub>I(PP)</sub> = 1 V	25°C	5.3		V/μs	
			0°C	5.9			
			70°C	4.3			
		V <sub>I(PP)</sub> = 5.5 V	25°C	4.6			
			0°C	5.1			
			70°C	3.8			
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 Ω	25°C	25		nV/√Hz		
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>OH</sub> , C <sub>L</sub> = 20 pF, R <sub>L</sub> = 10 kΩ	25°C	200		kHz		
		0°C	220				
		70°C	140				
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF	25°C	2.2		MHz		
		0°C	2.5				
		70°C	1.8				
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, f = B <sub>1</sub> , C <sub>L</sub> = 20 pF	25°C	49°				
		0°C	50°				
		70°C	46°				



**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**  
SLOS001E – JULY 1983 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**electrical characteristics at specified free-air temperature**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC251C, TLC251AC, TLC251BC						UNIT
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	TLC251C V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 10 kΩ	25°C	1.1		10	1.1		10	mV
			Full range	12			12			
			25°C	0.9		5	0.9		5	
			Full range	6.5			6.5			
			25°C	0.34		2	0.39		2	
			Full range	3			3			
αV <sub>IO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.7			2.1			μV/°C
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1			pA
			70°C	7			300	7		
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7			pA
			70°C	40			600	50		
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V
			Full range	-0.2 to 3.5			-0.2 to 8.5			V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 10 kΩ	25°C	3.2	3.9		8	8.7		V
			0°C	3	3.9		7.8	8.7		
			70°C	3	4		7.8	8.7		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50	0		50	mV
			0°C	0		50	0		50	
			70°C	0		50	0		50	
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 6	25°C	25	170		25	275		V/mV
			0°C	15	200		15	320		
			70°C	15	140		15	230		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	91		65	94		dB
			0°C	60	91		60	94		
			70°C	60	92		60	94		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C	70	93		70	93		dB
			0°C	60	92		60	92		
			70°C	60	94		60	94		
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub> /2	25°C	-130			-160			nA
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	105		280	143		300	μA
			0°C	125		320	173		400	
			70°C	85		220	110		280	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC251C, TLC251AC, TLC251BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		$V/\mu\text{s}$
			0°C	0.46		
			70°C	0.36		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40		
			0°C	0.43		
			70°C	0.34		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	25°C	32		$\text{nV}/\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $C_L = 20\text{ pF}$ , $R_L = 100\text{ k}\Omega$	25°C	55		kHz	
		0°C	60			
		70°C	50			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	25°C	525		kHz	
		0°C	600			
		70°C	400			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , $C_L = 20\text{ pF}$	25°C	40°			
		0°C	41°			
		70°C	39°			

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC251C, TLC251AC, TLC251BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		$V/\mu\text{s}$
			0°C	0.67		
			70°C	0.51		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56		
			0°C	0.61		
			70°C	0.46		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	25°C	32		$\text{nV}/\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $C_L = 20\text{ pF}$ , $R_L = 100\text{ k}\Omega$	25°C	35		kHz	
		0°C	40			
		70°C	30			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	25°C	635		kHz	
		0°C	710			
		70°C	510			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , $C_L = 20\text{ pF}$	25°C	43°			
		0°C	44°			
		70°C	42°			



**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**  
SLOS001E – JULY 1983 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**electrical characteristics at specified free-air temperature**

PARAMETER		TEST CONDITIONS	TA†	TLC251C, TLC251AC, TLC251BC						UNIT
				VDD = 5 V			VDD = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 10 MΩ	25°C	1.1		10	1.1		10	mV
			Full range			12			12	
			25°C	0.9		5	0.9		5	
			Full range			6.5			6.5	
			25°C	0.24		2	0.26		2	
			Full range			3			3	
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.1			1		μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1		pA	
			70°C	7		300	7			300
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7		pA	
			70°C	40		600	50			600
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5			-0.2 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 MΩ	25°C	3.2	4.1		8	8.9	V	
			0°C	3	4.1		7.8	8.9		
			70°C	3	4.2		7.8	8.9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50	0		mV	
			0°C	0		50	0			50
			70°C	0		50	0			50
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 MΩ, See Note 6	25°C	50	520		50	870	V/mV	
			0°C	50	700		50	1030		
			70°C	50	380		50	660		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	94		65	97	dB	
			0°C	60	95		60	97		
			70°C	60	95		60	97		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C	70	97		70	97	dB	
			0°C	60	97		60	97		
			70°C	60	98		60	98		
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub>	25°C	65			95		nA	
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	10		17	14		23	μA
			0°C	12		21	18		33	
			70°C	8		14	11		20	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.

**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC251C, TLC251AC, TLC251BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	0.03		V/ $\mu$ s
			0°C	0.04		
			70°C	0.03		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.03		
			0°C	0.03		
			70°C	0.02		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $C_L = 20\text{ pF}$ , $R_L = 1\text{ M}\Omega$	25°C	5		kHz	
		0°C	6			
		70°C	4.5			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	25°C	85		kHz	
		0°C	100			
		70°C	65			
		$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , $C_L = 20\text{ pF}$	25°C		34°
0°C	36°					
70°C	30°					

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC251C, TLC251AC, TLC251BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	0.05		V/ $\mu$ s
			0°C	0.05		
			70°C	0.04		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.04		
			0°C	0.05		
			70°C	0.04		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $C_L = 20\text{ pF}$ , $R_L = 1\text{ M}\Omega$	25°C	1		kHz	
		0°C	1.3			
		70°C	0.9			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	25°C	110		kHz	
		0°C	125			
		70°C	90			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , $C_L = 20\text{ pF}$	25°C	38°			
		0°C	40°			
		70°C	34°			





**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**  
SLOS001E – JULY 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 1.4\text{ V}$**

PARAMETER		TEST CONDITIONST	$T_A$ ‡	BIAS	TLC251C, TLC251AC, TLC251BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 0.2\text{ V}, R_S = 50\ \Omega$	25°C	Any	10			mV
					Full range			
			25°C	Any	5			
					Full range			
			25°C	Any	2			
					Full range			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage		25°C to 70°C	Any	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current	$V_O = 0.2\text{ V}$	25°C	Any	1		pA	
					Full range			300
$I_{IB}$	Input bias current	$V_O = 0.2\text{ V}$	25°C	Any	1		pA	
					Full range			600
$V_{ICR}$	Common-mode input voltage range		25°C	Any	0 to 0.2		V	
$V_{OM}$	Peak output voltage swing§	$V_{ID} = 100\text{ mV}$	25°C	Any	450	700	mV	
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV}, R_S = 50\ \Omega$	25°C	Low	20			
				High	10			
$CMRR$	Common-mode rejection ratio	$R_S = 50\ \Omega, V_O = 0.2\text{ V}, V_{IC} = V_{ICRmin}$	25°C	Any	60	77	dB	
$I_{DD}$	Supply current	$V_O = 0.2\text{ V}, \text{ No load}$	25°C	Low	5	17	$\mu\text{A}$	
				High	150	190		

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following values: for low bias,  $R_L = 1\text{ M}\Omega$ , for medium bias,  $R_L = 100\text{ k}\Omega$ , and for high bias,  $R_L = 10\text{ k}\Omega$ .

‡ Full range is 0°C to 70°C.

§ The output swings to the potential of  $V_{DD-}/\text{GND}$ .

**operating characteristics,  $V_{DD} = 1.4\text{ V}, T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	BIAS	TLC251C, TLC251AC, TLC251BC			UNIT
				MIN	TYP	MAX	
$B_1$	Unity-gain bandwidth	$C_L = 100\text{ pF}$	Low	12		kHz	
			High	12			
SR	Slew rate at unity gain	See Figure 1	Low	0.001		$\text{V}/\mu\text{s}$	
			High	0.1			
Overshoot factor	See Figure 1	Low	35%				
		High	30%				

**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**electrical characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC251Y									UNIT
		HIGH-BIAS MODE			MEDIUM-BIAS MODE			LOW-BIAS MODE			
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $V_{IC} = 0\text{ V}$ , $R_S = 50\ \Omega$ , $R_L^\dagger$		1.1	10		1.1	10		1.1	10	mV
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage			1.8			1.7			1.1		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.1			0.1			0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.6			0.6			0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L^\dagger$	3.2	3.8		3.2	3.9		3.2	4.1		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50		0	50		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V}$ , $R_L^\dagger$	5	23		25	170		50	480		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	80		65	91		65	94		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		70	93		70	97		dB
$I_{I(SEL)}$ Input current (BIAS SELECT)	$V_{I(SEL)} = V_{DD}/2$		-1.4			-0.13			0.065		$\mu\text{A}$
$I_{DD}$ Supply current	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$ , No load		675	1600		105	280		10	17	$\mu\text{A}$

$^\dagger$  For high-bias mode,  $R_L = 10\text{ k}\Omega$ ; for medium-bias mode,  $R_L = 100\text{ k}\Omega$ ; and for low-bias mode,  $R_L = 1\text{ M}\Omega$ .

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC251Y									UNIT
		HIGH-BIAS MODE			MEDIUM-BIAS MODE			LOW-BIAS MODE			
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $R_L \dagger$ , $C_L = 20\text{ pF}$	$V_I(\text{PP}) = 1\text{ V}$			3.6			0.43			$\text{V}/\mu\text{s}$
		$V_I(\text{PP}) = 2.5\text{ V}$			2.9			0.40			
$V_n$	Equivalent input noise voltage $f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	25			32			68			$\text{nV}/\sqrt{\text{Hz}}$
BOM	Maximum output swing bandwidth $V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	320			55			4.5			kHz
$B_1$	Unity-gain bandwidth $V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	1700			525			65			kHz
$\phi_m$	Phase margin $f = B_1$ , $C_L = 20\text{ pF}$	$V_I = 10\text{ mV}$ , 46°			40°			34°			

† For high-bias mode,  $R_L = 10\text{ k}\Omega$ ; for medium-bias mode,  $R_L = 100\text{ k}\Omega$ ; and for low-bias mode,  $R_L = 1\text{ M}\Omega$ .

### PARAMETER MEASUREMENT INFORMATION

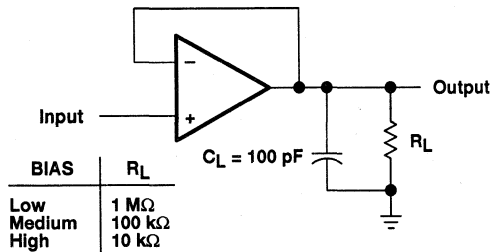


Figure 1. Unity-Gain Amplifier

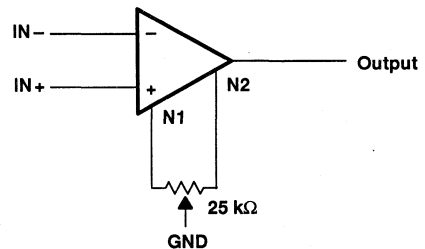


Figure 2. Input Offset Voltage Null Circuit

### TYPICAL CHARACTERISTICS

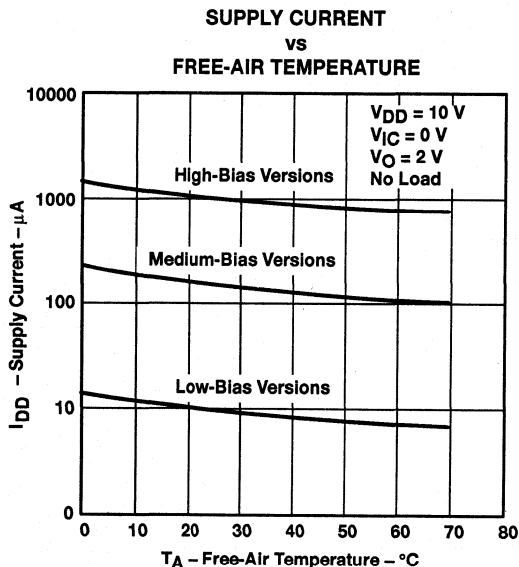
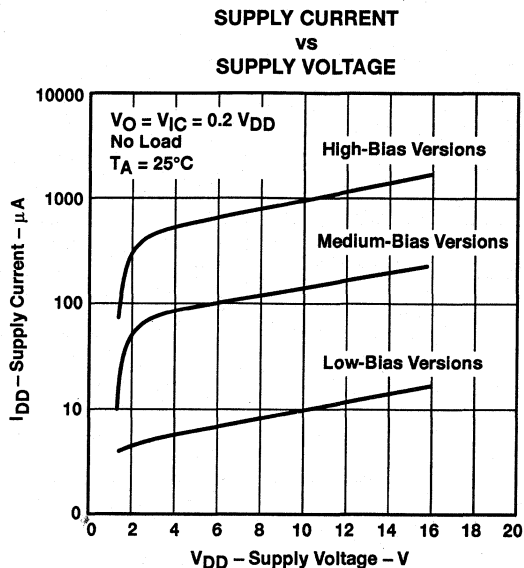
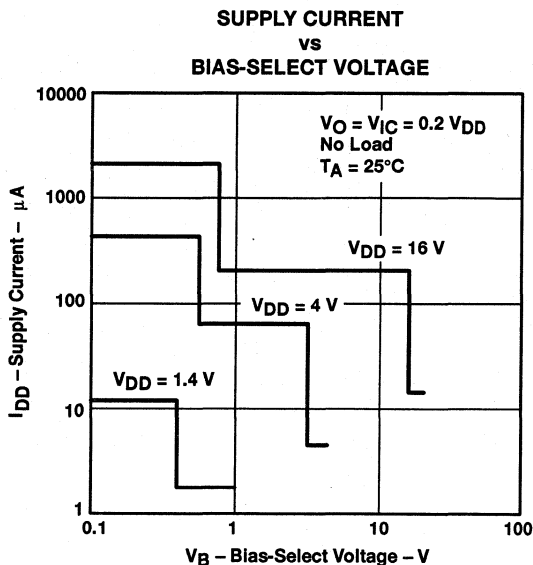
#### Table of Graphs

		FIGURE	
$I_{DD}$	Supply current	vs Bias-select voltage	3
		vs Supply voltage	4
		vs Free-air temperature	5
$A_{VD}$	Large-signal differential voltage amplification	Low bias vs Frequency	6
		Medium bias vs Frequency	7
		High bias vs Frequency	8
Phase shift		Low bias vs Frequency	6
		Medium bias vs Frequency	7
		High bias vs Frequency	8

**TLC251, TLC251A, TLC251B, TLC251Y**  
**LinCMOS™ PROGRAMMABLE**  
**LOW-POWER OPERATIONAL AMPLIFIERS**

SLOS001E – JULY 1983 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**



TYPICAL CHARACTERISTICS

LOW-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 VS  
 FREQUENCY

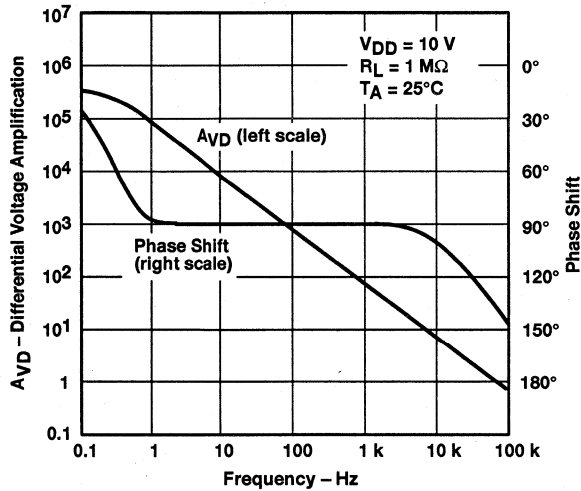


Figure 6

MEDIUM-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 VS  
 FREQUENCY

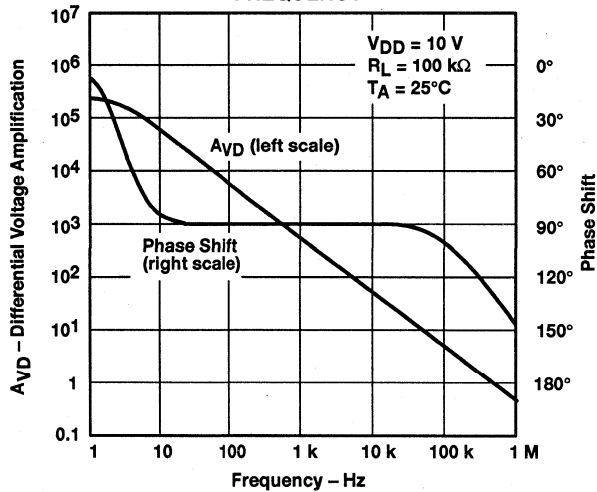


Figure 7

TLC251, TLC251A, TLC251B, TLC251Y  
LinCMOS™ PROGRAMMABLE  
LOW-POWER OPERATIONAL AMPLIFIERS

SLOS001E – JULY 1983 – REVISED AUGUST 1994

TYPICAL CHARACTERISTICS

HIGH-BIAS LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
AND PHASE SHIFT  
vs  
FREQUENCY

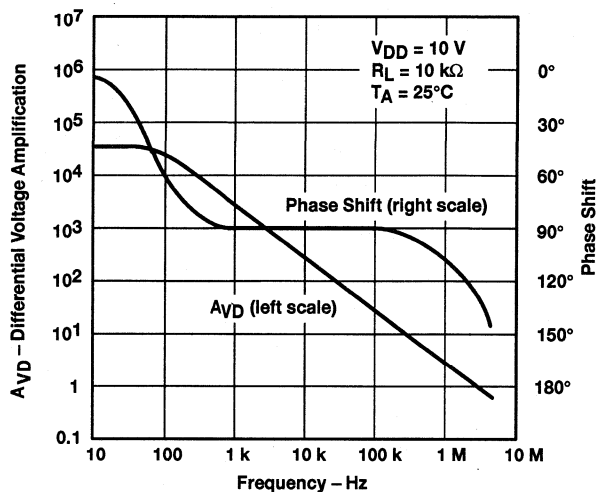


Figure 8

APPLICATION INFORMATION

**latch-up avoidance**

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the operational amplifier supplies should be applied simultaneously with, or before, application of any input signals.



## APPLICATION INFORMATION

### using BIAS SELECT

The TLC251 has a terminal called BIAS SELECT that allows the selection of one of three  $I_{DD}$  conditions (10, 150, and 1000  $\mu\text{A}$  typical). This allows the user to trade-off power and ac performance. As shown in the typical supply current ( $I_{DD}$ ) versus supply voltage ( $V_{DD}$ ) curves (Figure 4), the  $I_{DD}$  varies only slightly from 4 V to 16 V. Below 4 V, the  $I_{DD}$  varies more significantly. Note that the  $I_{DD}$  values in the medium- and low-bias modes at  $V_{DD} = 1.4$  V are typically 2  $\mu\text{A}$ , and in the high mode are typically 12  $\mu\text{A}$ . The following table shows the recommended BIAS SELECT connections at  $V_{DD} = 10$  V.

BIAS MODE	AC PERFORMANCE	BIAS SELECT CONNECTION†	TYPICAL $I_{DD}$ ‡
Low	Low	$V_{DD}$	10 $\mu\text{A}$
Medium	Medium	0.8 V to 9.2 V	150 $\mu\text{A}$
High	High	Ground pin	1000 $\mu\text{A}$

† Bias selection may also be controlled by external circuitry to conserve power, etc. For information regarding BIAS SELECT, see Figure 3 in the typical characteristics curves.

‡ For  $I_{DD}$  characteristics at voltages other than 10 V, see Figure 4 in the typical characteristics curves.

### output stage considerations

The amplifier's output stage consists of a source-follower-connected pullup transistor and an open-drain pulldown transistor. The high-level output voltage ( $V_{OH}$ ) is virtually independent of the  $I_{DD}$  selection and increases with higher values of  $V_{DD}$  and reduced output loading. The low-level output voltage ( $V_{OL}$ ) decreases with reduced output current and higher input common-mode voltage. With no load,  $V_{OL}$  is essentially equal to the potential of  $V_{DD-}/\text{GND}$ .

### input offset nulling

The TLC251C series offers external offset null control. Nulling may be achieved by adjusting a 25-k $\Omega$  potentiometer connected between the offset null terminals with the wiper connected to the device  $V_{DD-}/\text{GND}$  pin as shown in Figure 2. The amount of nulling range varies with the bias selection. At an  $I_{DD}$  setting of 1000  $\mu\text{A}$  (high bias), the nulling range allows the maximum offset specified to be trimmed to zero. In low or medium bias or when the amplifier is used below 4 V, total nulling may not be possible for all units.

### supply configurations

Even though the TLC251C series is characterized for single-supply operation, it can be used effectively in a split-supply configuration when the input common-mode voltage ( $V_{ICR}$ ), output swing ( $V_{OL}$  and  $V_{OH}$ ), and supply voltage limits are not exceeded.

### circuit layout precautions

The user is cautioned that whenever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup, as well as excessive dc leakages.

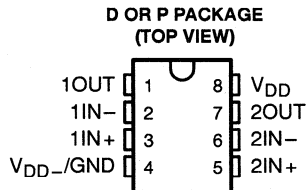




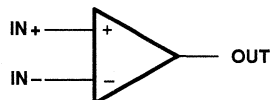
# TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

- A-Suffix Versions Offer 5-mV  $V_{IO}$
- B-Suffix Versions Offer 2-mV  $V_{IO}$
- Wide Range of Supply Voltages  
1.4 V to 16 V
- True Single-Supply Operation
- Common-Mode Input Voltage Includes the Negative Rail
- Low Noise . . . 30 nV/ $\sqrt{\text{Hz}}$  Typ at  $f = 1$  kHz (High-Bias Versions)



symbol (each amplifier)



## description

The TLC252, TLC25L2, and TLC25M2 are low-cost, low-power dual operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments

silicon gate LinCMOS™ process, giving them stable input offset voltages that are available in selected grades of 2, 5, or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The series offers operation down to a 1.4-V supply, is stable at unity gain, and has excellent noise characteristics.

These devices have internal electrostatic-discharge (ESD) protection circuits that prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in a degradation of the device parametric performance.

### AVAILABLE OPTIONS

$T_A$	$V_{IOmax}$ AT 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
0°C to 70°C	10 mV	TLC252CD	TLC252CP	TLC252Y
	5 mV	TLC252ACD	TLC252ACP	—
	2 mV	TLC252BCD	TLC252BCP	—
	10 mV	TLC25L2CD	TLC25L2CP	TLC25L2Y
	5 mV	TLC25L2ACD	TLC25L2ACP	—
	2 mV	TLC25L2BCD	TLC25L2BCP	—
	10 mV	TLC25M2CD	TLC25M2CP	TLC25M2Y
	5 mV	TLC25M2ACD	TLC25M2ACP	—
	2 mV	TLC25M2BCD	TLC25M2BCP	—

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC252CDR). Chips are tested at 25°C.

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



Copyright © 1994, Texas Instruments Incorporated

TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

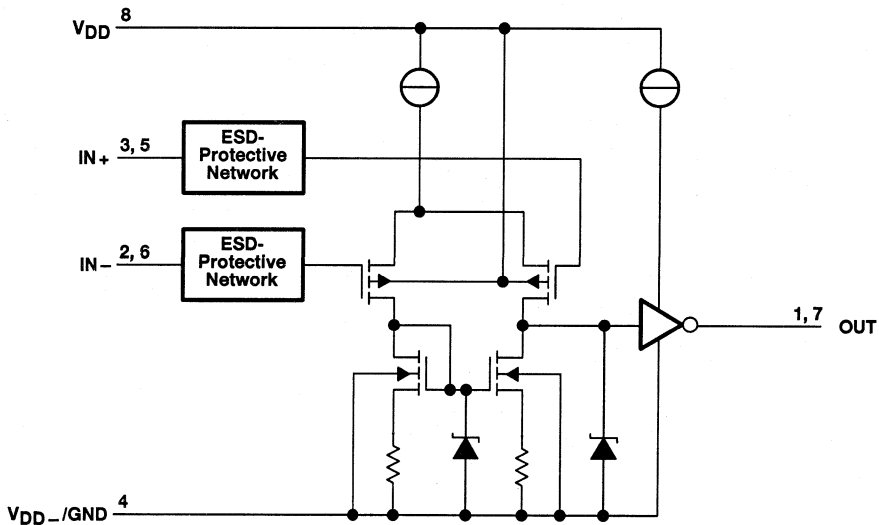
SLOS002F - JUNE 1983 - REVISED AUGUST 1994

**description (continued)**

Because of the extremely high input impedance and low input bias and offset currents, applications for the TLC252/25\_2 series include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS™ operational amplifiers without the power penalties of traditional bipolar devices. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC252/25\_2 series devices. Remote and inaccessible equipment applications are possible using their low-voltage and low-power capabilities. The TLC252/25\_2 series is well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for the commercial temperature range and are available in 8-pin plastic dip and the small-outline package. The device is also available in chip form.

The TLC252/25\_2 series is characterized for operation from 0°C to 70°C.

**equivalent schematic (each amplifier)**

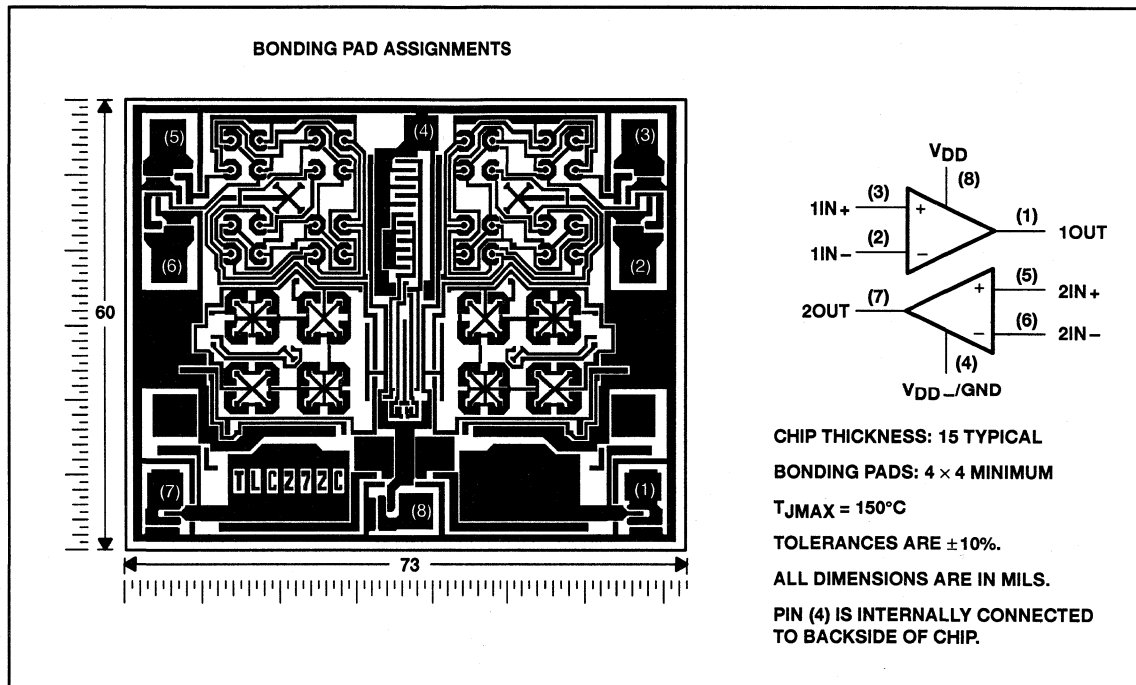


TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
 TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
 LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**TLC252Y, TLC25L2Y, and TLC25M2Y chip information**

These chips, properly assembled, display characteristics similar to the TLC252/25\_2. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOSTM DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 18$ V
Input voltage range, $V_I$ (any input)	-0.3 V to 18 V
Duration of short circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to  $V_{DD-}/GND$ .  
2. Differential voltages are at  $IN+$ , with respect to  $IN-$ .  
3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD}$	1.4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 1.4$ V	0	0.2
	$V_{DD} = 5$ V	-0.2	4
	$V_{DD} = 10$ V	-0.2	9
	$V_{DD} = 16$ V	-0.2	14
Operating free-air temperature, $T_A$	0	70	°C



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 1.4\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS†	TLC252_C			TLC25L2_C			TLC25M2_C			UNIT			
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX				
$V_{IO}$	Input offset voltage	$V_O = 0.2\text{ V}$ , $R_S = 50\ \Omega$	25°C	10			10			10			mV		
				TLC25_2C	0°C to 70°C	12			12			12			
					TLC25_2AC	25°C	5			5				5	
				TLC25_2BC		0°C to 70°C	6.5			6.5				6.5	
					TLC25_2BC	25°C	2			2				2	
				TLC25_2BC		0°C to 70°C	3			3				3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage	25°C to 70°C	1			1			1			$\mu\text{V}/^\circ\text{C}$			
$I_{IB}$	Input bias current	$V_O = 0.2\text{ V}$	25°C	1			1			1			pA		
			0°C to 70°C	300			300			300					
$V_{ICR}$	Common-mode input voltage range	$V_O = 0.2\text{ V}$	25°C	1			1			1			pA		
			0°C to 70°C	600			600			600					
$V_{ICR}$	Common-mode input voltage range		25°C	0 to 0.2	0 to 0.2			0 to 0.2			V				
$V_{OM}$	Peak output voltage swing‡	$V_{ID} = 100\text{ mV}$	25°C	450	700	450	700	450	700	mV					
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV}$ , $R_S = 50\ \Omega$	25°C	10			20			20			V/mV		
CMRR	Common-mode rejection ratio	$V_O = 0.2\text{ V}$ , $V_{IC} = V_{ICRmin}$	25°C	60	77	60	77	60	77	60	77	dB			
$I_{DD}$	Supply current	$V_O = 0.2\text{ V}$ , No load	25°C	300	375	25	34	200	250	$\mu\text{A}$					

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following value: for low bias  $R_L = 1\text{ M}\Omega$ , for medium bias  $R_L = 100\text{ k}\Omega$ , and for high bias  $R_L = 10\text{ k}\Omega$ .

‡ The output swings to the potential of  $V_{DD-}/\text{GND}$ .

**operating characteristics,  $V_{DD} = 1.4\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER		TEST CONDITIONS	TLC252_C			TLC25L2_C			TLC25M2_C			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$B_1$	Unity-gain bandwidth	$A_V = 40\text{ dB}$ , $C_L = 10\text{ pF}$ , $R_S = 50\ \Omega$	12			12			12			kHz
SR	Slew rate at unity gain	See Figure 1	0.1			0.001			0.01			V/ $\mu\text{s}$
	Overshoot factor	See Figure 1	30%			35%			35%			



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	$T_A$ †	TLC252C, TLC252AC, TLC252BC			UNIT
				MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	1.1	10	mV
				Full range		12	
		$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	0.9	5	
				Full range		6.5	
		$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	0.23	2	
				Full range		3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage		25°C to 70°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
			70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
			70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	V	
			Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 10\ \text{k}\Omega$	25°C	3.2	3.8	V	
			0°C	3	3.8		
			70°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
			0°C	0	50		
			70°C	0	50		
$AVD$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ , $R_L = 10\ \text{k}\Omega$	25°C	5	23	V/mV	
			0°C	4	27		
			70°C	4	20		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	65	80	dB	
			0°C	60	84		
			70°C	60	85		
$KSVR$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{DD}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	65	95	dB	
			0°C	60	94		
			70°C	60	96		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	1.4	3.2	mA
				0°C	1.6	3.6	
				70°C	1.2	2.6	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC252C, TLC252AC, TLC252BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC252C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC252AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC252BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.29	2	
					Full range		3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				0°C	8	8.5		
				70°C	7.8	8.4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				0°C	7.5	42		
				70°C	7.5	32		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85	dB	
				0°C	60	88		
				70°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{DD}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	1.9	4	mA	
				0°C	2.3	4.4		
				70°C	1.6	3.4		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC252C, TLC252AC, TLC252BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ $\mu\text{s}$	
				0°C	4			
				70°C	3			
				$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9		
					0°C	3.1		
					70°C	2.5		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure	$C_L = 20\text{ pF}$	$R_L = 100\text{ k}\Omega$ ,	25°C	320		kHz	
				0°C	340			
				70°C	260			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 3	25°C	1.7		MHz	
				0°C	2			
				70°C	1.3			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	25°C	46°			
				0°C	47°			
				70°C	43°			

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC252C, TLC252AC, TLC252BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ $\mu\text{s}$	
				0°C	5.9			
				70°C	4.3			
			$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6			
				0°C	5.1			
				70°C	3.8			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$	$R_L = 100\text{ k}\Omega$ ,	25°C	200		kHz	
				0°C	220			
				70°C	140			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 3	25°C	2.2		MHz	
				0°C	2.5			
				70°C	1.8			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	25°C	49°			
				0°C	50°			
				70°C	46°			





**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25L2C TLC25L2AC TLC25L2BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC252C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1		10
					Full range			12
		TLC252AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9		5
					Full range			6.5
		TLC252BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.204		2
					Full range			3
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7 300			
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	50 600			
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				0°C	3	4.1		
				70°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0 50		mV	
				0°C	0 50			
				70°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	700	V/mV	
				0°C	50	700		
				70°C	50	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	95		
				70°C	60	95		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{DD}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	70	97	dB	
				0°C	60	97		
				70°C	60	98		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	20	34	mA	
				0°C	24	42		
				70°C	16	28		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F - JUNE 1983 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25L2C TLC25L2AC TLC25L2BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC252C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1 10		mV
					Full range	12		
		TLC252AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9 5		
					Full range	6.5		
		TLC252BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.235 2		
					Full range	3		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	8 300			
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50 600			
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9 -0.3 to 9.2		V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 1\text{ M}\Omega$	25°C	8 8.9		V	
				0°C	7.8 8.9			
				70°C	7.8 8.9			
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0 50		mV	
				0°C	0 50			
				70°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 1\text{ M}\Omega$	25°C	50 860		V/mV	
				0°C	50 1025			
				70°C	50 660			
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65 97		dB	
				0°C	60 97			
				70°C	60 97			
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{DD}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	70 97		dB	
				0°C	60 97			
				70°C	60 98			
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	29 46		mA	
				0°C	36 66			
				70°C	22 40			

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC25L2C TLC25L2AC TLC25L2BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_{I(PP)} = 1\text{ V}$	25°C		0.03	V/ $\mu$ s	
				0°C		0.04		
				70°C		0.03		
			$V_{I(PP)} = 2.5\text{ V}$	25°C		0.03		
				0°C		0.03		
				70°C		0.02		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C		68	nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure	$C_L = 20\text{ pF}$ ,	$R_L = 1\text{ M}\Omega$ ,	25°C		5	kHz	
				0°C		6		
				70°C		4.5		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 3	25°C		85	MHz	
				0°C		100		
				70°C		65		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	25°C		34°		
				0°C		36°		
				70°C		30°		

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC25L2C TLC25L2AC TLC25L2BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_{I(PP)} = 1\text{ V}$	25°C		0.05	V/ $\mu$ s	
				0°C		0.05		
				70°C		0.04		
			$V_{I(PP)} = 5.5\text{ V}$	25°C		0.04		
				0°C		0.05		
				70°C		0.04		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C		68	nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$R_L = 1\text{ M}\Omega$ ,	25°C		1	kHz	
				0°C		1.3		
				70°C		0.9		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 3	25°C		110	MHz	
				0°C		125		
				70°C		90		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	25°C		38°		
				0°C		40°		
				70°C		34°		



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	$T_A$ †	TLC25M2C TLC25M2AC TLC25M2BC			UNIT
				MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
				Full range		12	
		$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
				Full range		6.5	
		$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.22	2	
				Full range		3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage		25°C to 70°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
			70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
			70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	V	
			Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
			0°C	3	3.9		
			70°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
			0°C	0	50		
			70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ , $R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
			0°C	15	200		
			70°C	15	140		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	65	91	dB	
			0°C	60	91		
			70°C	60	92		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{DD}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	70	93	dB	
			0°C	60	92		
			70°C	60	94		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	210	560	mA
				0°C	250	640	
				70°C	170	440	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC25M2C TLC25M2AC TLC25M2BC			UNIT
				MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω, V <sub>IC</sub> = 0, R <sub>L</sub> = 100 kΩ	25°C	1.1		10	mV
			Full range	12			
		V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω, V <sub>IC</sub> = 0, R <sub>L</sub> = 100 kΩ	25°C	0.9		5	
			Full range	6.5			
		V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω, V <sub>IC</sub> = 0, R <sub>L</sub> = 100 kΩ	25°C	0.224		2	
			Full range	3			
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	2.1		μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = 5 V, V <sub>IC</sub> = 5 V	25°C	0.1		pA	
			70°C	7 300			
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = 5 V, V <sub>IC</sub> = 5 V	25°C	0.7		pA	
			70°C	50 600			
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 100 kΩ	25°C	8 8.7		V	
			0°C	7.8 8.7			
			70°C	7.8 8.7			
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0 50		mV	
			0°C	0 50			
			70°C	0 50			
A <sub>VD</sub>	Large-signal differential voltage amplification	V <sub>O</sub> = 1 V to 6 V, R <sub>L</sub> = 100 kΩ	25°C	25 275		V/mV	
			0°C	15 320			
			70°C	15 230			
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65 94		dB	
			0°C	60 94			
			70°C	60 94			
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>DD</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C	70 93		dB	
			0°C	60 92			
			70°C	60 94			
I <sub>DD</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 5 V, No load V <sub>IC</sub> = 5 V,	25°C	285 600		mA	
			0°C	345 800			
			70°C	220 560			

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B

TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y

LinCMOS™ DUAL OPERATIONAL AMPLIFIERS

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

operating characteristics,  $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS			$T_A$	TLC25M2C TLC25M2AC TLC25M2BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		$V_I(PP) = 1\text{ V}$	25°C	0.43		V/ $\mu$ s	
				0°C	0.46			
				70°C	0.36			
				$V_I(PP) = 2.5\text{ V}$	25°C	0.40		
					0°C	0.43		
					70°C	0.34		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	See Figure 2	25°C	32		nV/ $\sqrt{\text{Hz}}$		
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure	$C_L = 20\text{ pF}$ , $R_L = 100\text{ k}\Omega$	25°C	55		kHz		
			0°C	60				
			70°C	50				
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	See Figure 3	25°C	525		MHz		
			0°C	600				
			70°C	400				
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ , $C_L = 20\text{ pF}$	25°C	40°				
			0°C	41°				
			70°C	39°				

operating characteristics,  $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS			$T_A$	TLC25M2C TLC25M2AC TLC25M2BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		$V_I(PP) = 1\text{ V}$	25°C	0.62		V/ $\mu$ s	
				0°C	0.67			
				70°C	0.51			
			$V_I(PP) = 5.5\text{ V}$	25°C	0.56			
				0°C	0.61			
				70°C	0.46			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$	See Figure 2	25°C	32		nV/ $\sqrt{\text{Hz}}$		
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ , $R_L = 100\text{ k}\Omega$	25°C	35		kHz		
			0°C	40				
			70°C	30				
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	See Figure 3	25°C	635		MHz		
			0°C	710				
			70°C	510				
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ , $C_L = 20\text{ pF}$	25°C	43°				
			0°C	44°				
			70°C	42°				



**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F - JUNE 1983 - REVISED AUGUST 1994

**electrical characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC252Y			TLC25L2Y			TLC25M2Y			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $V_{IC} = 0\text{ V}$ , $R_S = 50\ \Omega$ , See Note 6		1.1	10		1.1	10		1.1	10	mV
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage			1.8			1.1			1.7		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.1			0.1			0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.6			0.6			0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , See Note 6	3.2	3.8		3.2	4.1		3.2	3.9		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50		0	50		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V}$ , See Note 6	5	23		50	700		25	170		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	80		65	94		65	91		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V}$ to $10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		70	97		70	93		dB
$I_{DD}$ Supply current	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$ , No load		1.4	3.2		0.02	0.034		0.21	0.56	mA

**operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

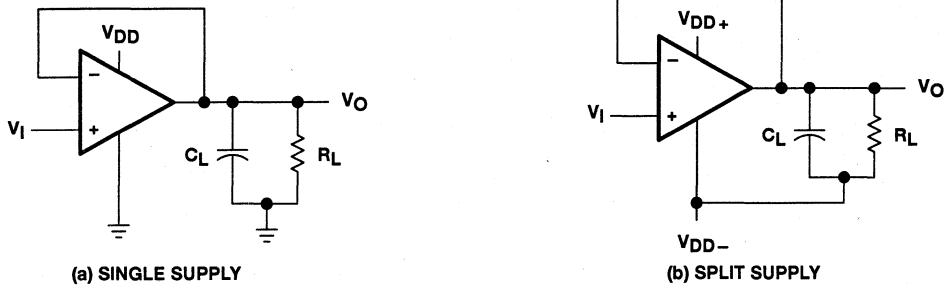
PARAMETER	TEST CONDITIONS		TLC252Y			TLC25L2Y			TLC25M2Y			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Slew rate at unity gain	$C_L = 20\text{ pF}$ , See Note 6	$V_{I(pp)} = 1\text{ V}$ $V_{I(pp)} = 2.5\text{ V}$	3.6			0.03			0.43			V/ $\mu\text{s}$
			2.9			0.03			0.40			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$	2.5			68			32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ ,	$C_L = 20\text{ pF}$ , $R_L = 10\text{ k}\Omega$	320			5			55			kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$	1.7			0.085			0.525			MHz
$\phi_m$ Phase margin	$f = B_1$ ,	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	46°			34°			40°			

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. For low-bias mode,  $R_L = 1\text{ M}\Omega$ ; for medium-bias mode,  $R_L = 100\text{ k}\Omega$ , and for high-bias mode,  $R_L = 10\text{ k}\Omega$ .

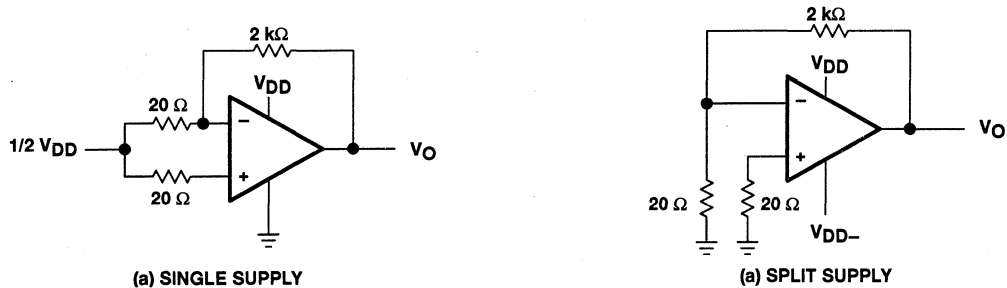
**PARAMETER MEASUREMENT INFORMATION**

**single-supply versus split-supply test circuits**

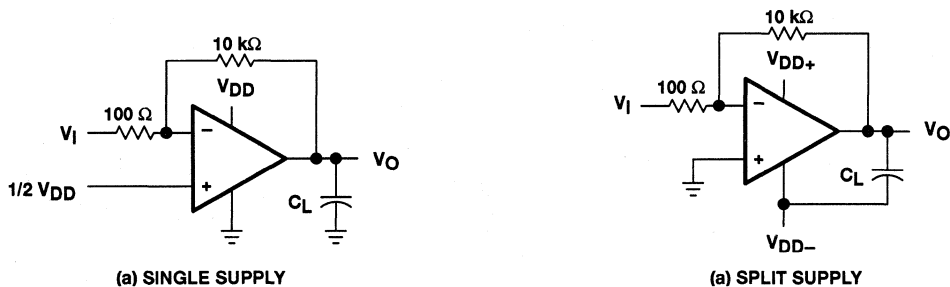
Because the TLC252, TLC25L2, and TLC25M2 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.



**Figure 1. Unity-Gain Amplifier**



**Figure 2. Noise-Test Circuit**



**Figure 3. Gain-of-100 Inverting Amplifier**



TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
$I_{DD}$	Supply current	vs Supply voltage	4
		vs Free-air temperature	5
$A_{VD}$	Large-signal differential voltage amplification	Low bias vs Frequency	6
		Medium bias vs Frequency	7
		High bias vs Frequency	8
	Phase shift	Low bias vs Frequency	6
		Medium bias vs Frequency	7
		High bias vs Frequency	8

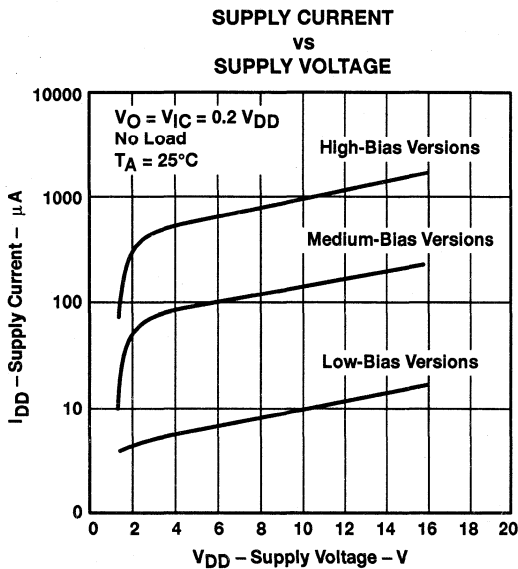


Figure 4

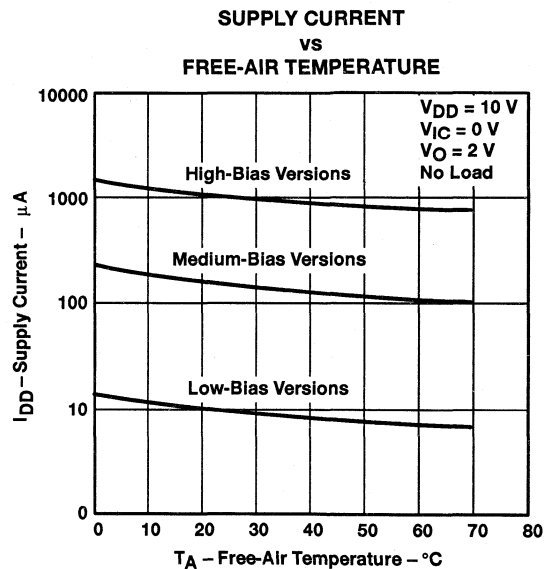


Figure 5

TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
 TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
**LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**LOW-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 vs  
 FREQUENCY**

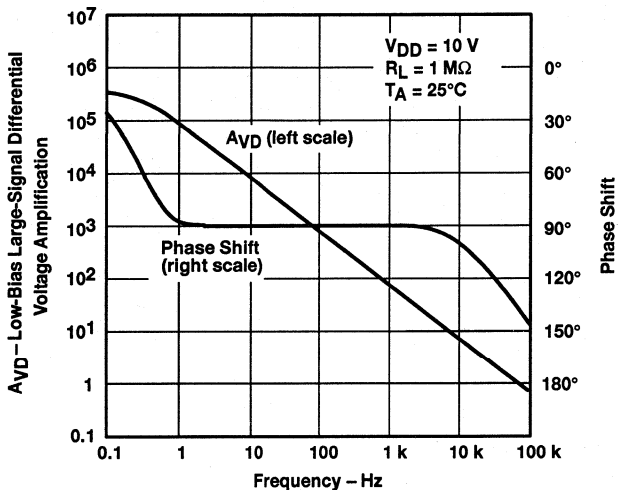


Figure 6

**MEDIUM-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 vs  
 FREQUENCY**

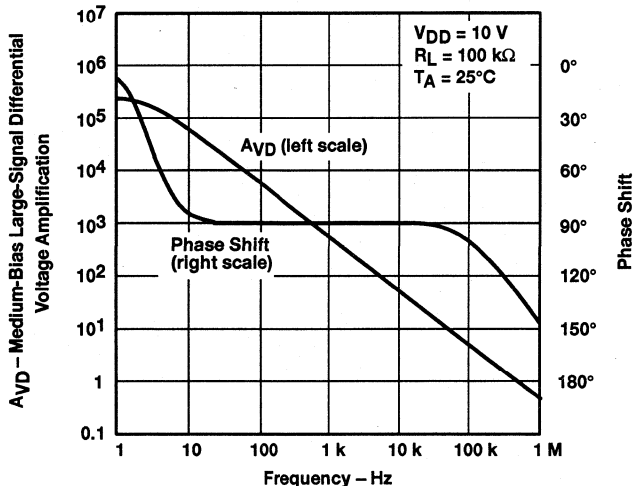


Figure 7



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

TYPICAL CHARACTERISTICS  
 HIGH-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 vs  
 FREQUENCY

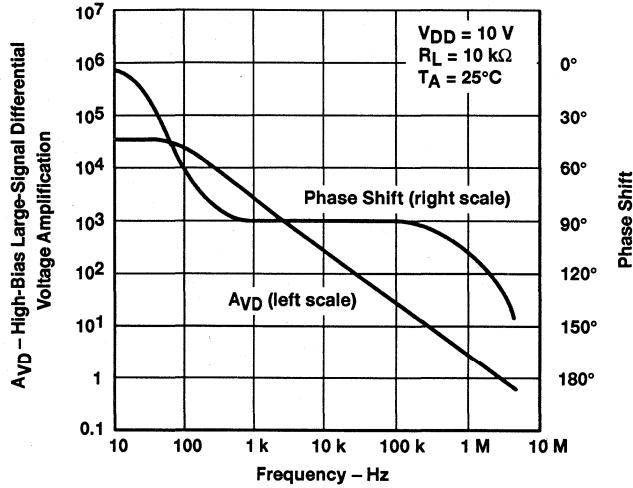


Figure 8

**TLC252, TLC252A, TLC252B, TLC252Y, TLC25L2, TLC25L2A, TLC25L2B  
TLC25L2Y, TLC25M2, TLC25M2A, TLC25M2B, TLC25M2Y  
LinCMOS™ DUAL OPERATIONAL AMPLIFIERS**

SLOS002F – JUNE 1983 – REVISED AUGUST 1994

---

**APPLICATION INFORMATION**

**latch-up avoidance**

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the operational amplifier supplies should be applied simultaneously with, or before, application of any input signals.

**output stage considerations**

The amplifier's output stage consists of a source-follower-connected pullup transistor and an open-drain pulldown transistor. The high-level output voltage ( $V_{OH}$ ) is virtually independent of the  $I_{DD}$  selection and increases with higher values of  $V_{DD}$  and reduced output loading. The low-level output voltage ( $V_{OL}$ ) decreases with reduced output current and higher input common-mode voltage. With no load,  $V_{OL}$  is essentially equal to the potential of  $V_{DD-}/GND$ .

**supply configurations**

Even though the TLC252/25\_2C series is characterized for single-supply operation, it can be used effectively in a split-supply configuration if the input common-mode voltage ( $V_{ICR}$ ), output swing ( $V_{OL}$  and  $V_{OH}$ ), and supply voltage limits are not exceeded.

**circuit layout precautions**

The user is cautioned that whenever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup, as well as excessive dc leakages.

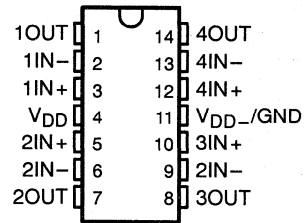


# TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

- **A-Suffix Versions Offer 5-mV  $V_{IO}$**
- **B-Suffix Versions Offer 2-mV  $V_{IO}$**
- **Wide Range of Supply Voltages**  
1.4 V to 16 V
- **True Single-Supply Operation**
- **Common-Mode Input Voltage Includes the Negative Rail**
- **Low Noise . . . 25 nV/ $\sqrt{\text{Hz}}$  Typ at  $f = 1 \text{ kHz}$  (High-Bias Version)**

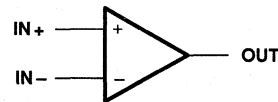
**D, N, OR PW PACKAGE  
(TOP VIEW)**



## description

The TLC254, TLC254A, TLC254B, TLC25L4, TLC254L4A, TLC254L4B, TLC25M4, TLC25M4A and TLC25M4B are low-cost, low-power quad operational amplifiers designed to operate with single or dual supplies. These devices utilize the Texas Instruments silicon gate LinCMOS™ process, giving them stable input-offset voltages that are available in selected grades of 2, 5, or 10 mV maximum, very high input impedances, and extremely low input offset and bias currents. Because the input common-mode range extends to the negative rail and the power consumption is extremely low, this series is ideally suited for battery-powered or energy-conserving applications. The series offers operation down to a 1.4-V supply, is stable at unity gain, and has excellent noise characteristics.

## symbol (each amplifier)



These devices have internal electrostatic-discharge (ESD) protection circuits that prevent catastrophic failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.1. However, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

Because of the extremely high input impedance and low input bias and offset currents, applications for these devices include many areas that have previously been limited to BIFET and NFET product types. Any circuit using high-impedance elements and requiring small offset errors is a good candidate for cost-effective use of these devices. Many features associated with bipolar technology are available with LinCMOS operational amplifiers without the power penalties of traditional bipolar devices.

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES			CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	10 mV	TLC254CD	TLC254CN	TLC254CPW	TLC254Y
	5 mV	TLC254ACD	TLC254ACN	—	—
	2 mV	TLC254BCD	TLC254BCN	—	—
	10 mV	TLC25L4CD	TLC25L4CN	TLC25L4CPW	TLC25L4Y
	5 mV	TLC25L4ACD	TLC25L4ACN	—	—
	2 mV	TLC25L2BCD	TLC25L4BCN	—	—
	10 mV	TLC25M4CD	TLC25M4CN	TLC25M4CPW	TLC25M4Y
	5 mV	TLC25M4ACD	TLC25M4ACN	—	—
	2 mV	TLC25M4BCD	TLC25M4BCN	—	—

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC254CDR). Chips are tested at 25°C.

LinBICMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

 **TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**description (continued)**

General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with these devices. Remote and inaccessible equipment applications are possible using their low-voltage and low-power capabilities. These devices are well suited to solve the difficult problems associated with single-battery and solar-cell-powered applications. This series includes devices that are characterized for the commercial temperature range and are available in 14-pin plastic dip and the small-outline packages. The device is also available in chip form.

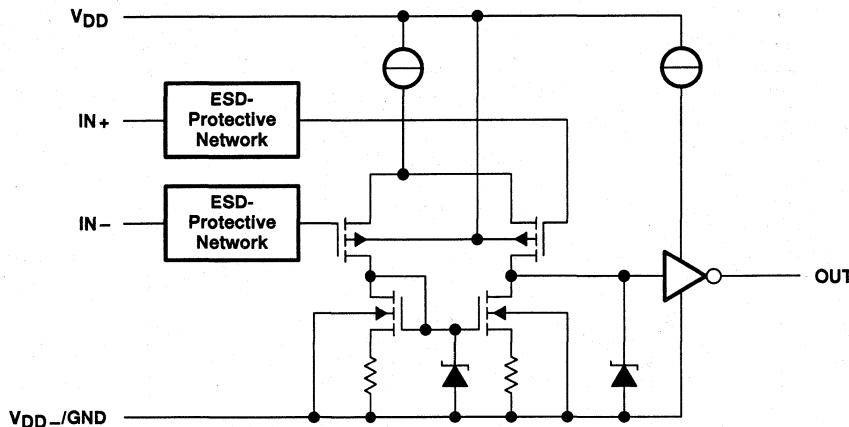
These devices are characterized for operation from 0°C to 70°C.

**DEVICE FEATURES**

PARAMETER	TLC25L4_C (LOW BIAS)	TLC25M4_C (MEDIUM BIAS)	TLC254_C (HIGH BIAS)
Supply current (Typ)	40 $\mu$ A	600 $\mu$ A	4000 $\mu$ A
Slew rate (Typ)	0.04 V/ $\mu$ A	0.6 V/ $\mu$ A	4.5 V/ $\mu$ A
Input offset voltage (Max) TLC254C, TLC25L4C, TLC25M4C TLC254AC, TLC25L4AC, TLC25M4AC TLC254BC, TLC25L4BC, TLC25M4BC	10 mV 5 mV 2 mV	10 mV 5 mV 2 mV	10 mV 5 mV 2 mV
Offset voltage drift (Typ)	0.1 $\mu$ V/month†	0.1 $\mu$ V/month†	0.1 $\mu$ V/month†
Offset voltage temperature coefficient (Typ)	0.7 $\mu$ V/°C	2 $\mu$ V/°C	5 $\mu$ V/°C
Input bias current (Typ)	1 pA	1 pA	1 pA
Input offset current (Typ)	1 pA	1 pA	1 pA

† The long-term drift value applies after the first month.

**equivalent schematic (each amplifier)**

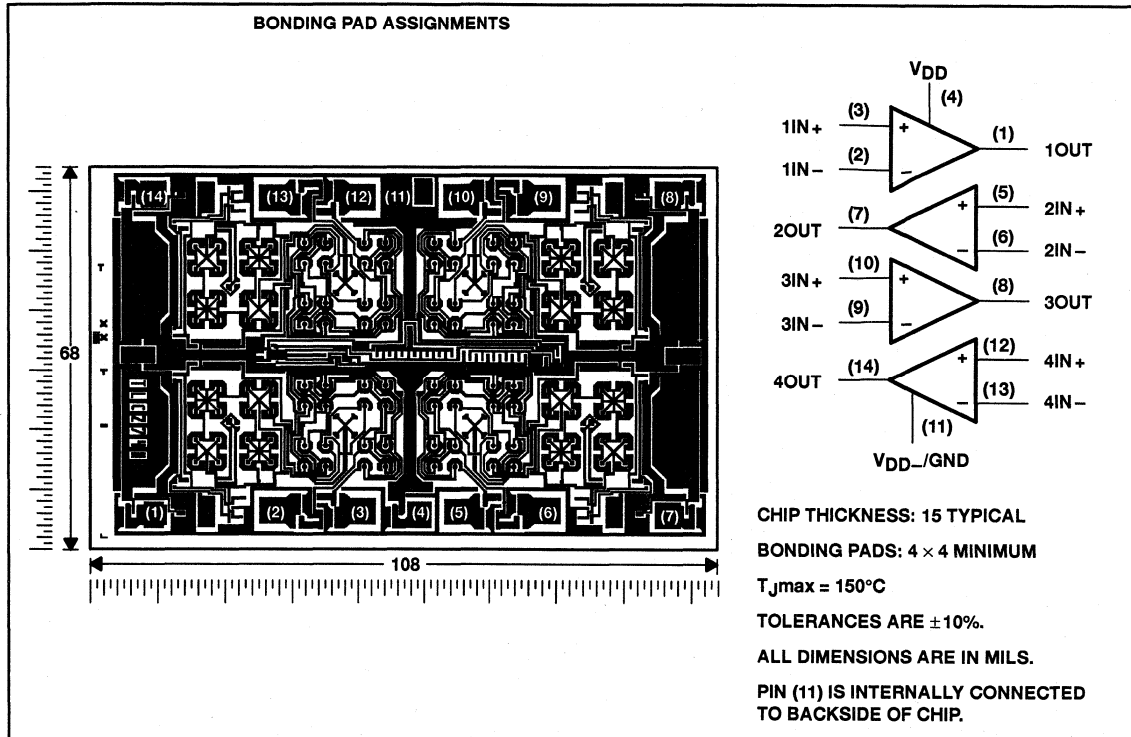


TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
 TLC25L4Y, TLC25M4, RLC25M4A, TLC25M4B, TLC25M4Y  
 LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

SLOS003F - JUNE 1983 - REVISED AUGUST 1994

chip information

These chips, when properly assembled, display characteristics similar to the TLC25\_4C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage (see Note 2)	$\pm 18$ V
Input voltage range (any input)	-0.3 V to 18 V
Duration of short-circuit at (or below) 25°C free-air temperature (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to  $V_{DD-}/GND$ .
  2. Differential voltages are at  $IN+$ , with respect to  $IN-$ .
  3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
N	1050 mW	9.2 mW/°C	736 mW
PW	700 mW	5.6 mW/°C	448 mW

**recommended operating conditions**

		MIN	MAX	UNIT
Supply voltage, $V_{DD}$		1.4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 1.4$ V	0	0.2	V
	$V_{DD} = 5$ V	-0.2	4	
	$V_{DD} = 10$ V	-0.2	9	
	$V_{DD} = 16$ V	-0.2	14	
Operating free-air temperature, $T_A$		0	70	°C





TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
 TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
 LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

SLOS003F - JUNE 1983 - REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 1.4\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	TA	TLC254_C		TLC25L4_C		TLC25M4_C		UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	TLC25_4C	25°C	10		10		10		10
		0°C to 70°C	12		12		12		12
	TLC25_4AC	25°C	5		5		5		5
		0°C to 70°C	6.5		6.5		6.5		6.5
Average temperature coefficient of input offset voltage	TLC25_4BC	25°C	2		2		2		2
		0°C to 70°C	3		3		3		3
		25°C to 70°C	1		1		1		1
$I_{IO}$ Input offset current	$V_O = 0.2\text{ V}$	25°C	1		1		1		1
$I_{IB}$ Input bias current	$V_O = 0.2\text{ V}$	0°C to 70°C	300		300		300		300
		25°C	1		1		1		1
$V_{ICR}$ Common-mode input voltage range	$V_O = 0.2\text{ V}$	0°C to 70°C	600		600		600		600
		25°C	0 to 0.2		0 to 0.2		0 to 0.2		0 to 0.2
$V_{OM}$ Peak output voltage swing†	$V_{ID} = 100\text{ mV}$	25°C	450	700	450	700	450	700	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 100\text{ to }300\text{ mV}$ , $R_S = 50\ \Omega$	25°C	10		20		20		V/mV
		25°C	60	77	60	77	60	77	dB
$I_{DD}$ Supply current	$V_O = 0.2\text{ V}$ , No load	25°C	600	750	600	750	600	500	$\mu\text{A}$

† All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified. Unless otherwise noted, an output load resistor is connected from the output to ground and has the following values: for low bias,  $R_L = 1\text{ M}\Omega$ , for medium bias  $R_L = 100\text{ k}\Omega$ , and for high bias  $R_L = 10\text{ k}\Omega$ .

‡ The output swings to the potential of  $V_{DD-}/\text{GND}$ .

operating characteristics,  $V_{DD} = 1.4\text{ V}$ ,  $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC254_C		TLC25L4_C		TLC25M4_C		UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	See Figure 1	0.1		0.001		0.01		V/ $\mu\text{s}$
$B_1$ Unity-gain bandwidth	$A_V = 40\text{ dB}$ , $R_S = 50\ \Omega$ , See Figure 1	12		12		12		KHz
Overshoot factor	See Figure 1	30%		35%		35%		



**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC254, TLC254AC, TLC254BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC254C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC254AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC254BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.34	2	
					Full range		3	
$\alpha V_{IO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5			
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	0°C	3	3.8	V	
				25°C	3.2	3.8		
				70°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	0°C	0	50	mV	
				25°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 10\text{ k}\Omega$	0°C	4	27	V/mV	
				25°C	5	23		
				70°C	4	20		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60	84	dB	
				25°C	65	80		
				70°C	60	85		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	0°C	60	94	dB	
				25°C	65	95		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	0°C	3.1	7.2	mA	
				25°C	2.7	6.4		
				70°C	2.3	5.2		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F - JUNE 1983 - REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC254C, TLC254AC, TLC254BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC254C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC254AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC254BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.39	2	
					Full range		3	
$^{\circ}V_{IO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2		$\mu\text{V}/^{\circ}\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5			
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\text{ k}\Omega$	0°C	7.8	8.5	V	
				25°C	8	8.5		
				70°C	7.8	8.4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	0°C	0	50	mV	
				25°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 10\text{ k}\Omega$	0°C	7.5	42	V/mV	
				25°C	10	36		
				70°C	7.5	32		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60	88	dB	
				25°C	65	85		
				70°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	0°C	60	94	dB	
				25°C	65	95		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	0°C	4.5	8.8	mA	
				25°C	3.8	8		
				70°C	3.2	6.8		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC254C, TLC254AC, TLC254BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_I(PP) = 1\text{ V}$	0°C	4		V/ $\mu\text{s}$	
				25°C	3.6			
				70°C	3			
				0°C	3.1			
				25°C	2.9			
				70°C	2.5			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$R_L = 10\text{ k}\Omega$ ,	0°C	340		kHz	
				25°C	320			
				70°C	260			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 1	0°C	2		MHz	
				25°C	1.7			
				70°C	1.3			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	0°C	47°			
				25°C	46°			
				70°C	43°			

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC254C, TLC254AC, TLC254BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_I(PP) = 1\text{ V}$	0°C	5.9		V/ $\mu\text{s}$	
				25°C	5.3			
				70°C	4.3			
			$V_I(PP) = 5.5\text{ V}$	0°C	5.1			
				25°C	4.6			
				70°C	3.8			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$R_L = 10\text{ k}\Omega$ ,	0°C	220		kHz	
				25°C	200			
				70°C	140			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 1	0°C	2.5		MHz	
				25°C	2.2			
				70°C	1.8			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	0°C	50°			
				25°C	49°			
				70°C	46°			



**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25L4C TLC25L4AC TLC25L4BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC25L4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1 10		mV
					Full range	12		
		TLC25L4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9 5		
					Full range	6.5		
		TLC25L4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.24 2		
					Full range	3		
$\infty V_{IO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7 300			
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40 600			
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	0°C	3 4.1		V	
				25°C	3.2 4.1			
				70°C	3 4.2			
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	0°C	0 50		mV	
				25°C	0 50			
				70°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 1\text{ M}\Omega$	0°C	50 680		V/mV	
				25°C	50 520			
				70°C	50 380			
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60 95		dB	
				25°C	65 94			
				70°C	60 95			
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	0°C	60 97		dB	
				25°C	70 98			
				70°C	60 97			
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	0°C	48 84		$\mu\text{A}$	
				25°C	40 68			
				70°C	31 56			

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25L4C TLC25L4AC TLC25L4BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC25L4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC25L4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC25L4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.26	2	
					Full range		3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = .5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	0°C	7.8	8.9	V	
				25°C	8	8.9		
				70°C	7.8	8.9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	0°C	0	50	mV	
				25°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	0°C	50	1025	V/mV	
				25°C	50	870		
				70°C	50	660		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60	97	dB	
				25°C	65	97		
				70°C	60	97		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	0°C	60	97	dB	
				25°C	70	97		
				70°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	0°C	72	132	$\mu\text{A}$	
				25°C	57	92		
				70°C	44	80		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F - JUNE 1983 - REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC25L4C TLC25L4AC TLC25L4BC			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_I(\text{PP}) = 1\text{ V}$	0°C	0.04		V/ $\mu\text{s}$
				25°C	0.03		
				70°C	0.03		
			$V_I(\text{PP}) = 2.5\text{ V}$	0°C	0.03		
				25°C	0.03		
				70°C	0.02		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	70		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$R_L = 1\text{ M}\Omega$ ,	0°C	6		kHz
				25°C	5		
				70°C	4.5		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 1	0°C	100		kHz
				25°C	85		
				70°C	65		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	0°C	36°		
				25°C	34°		
				70°C	30°		

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC25L4C TLC25L4AC TLC25L4BC			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$V_I(\text{PP}) = 1\text{ V}$	0°C	0.05		V/ $\mu\text{s}$
				25°C	0.05		
				70°C	0.04		
			$V_I(\text{PP}) = 5.5\text{ V}$	0°C	0.05		
				25°C	0.04		
				70°C	0.04		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ ,	$R_S = 20\ \Omega$ ,	See Figure 2	25°C	70		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ ,	$R_L = 1\text{ M}\Omega$ ,	0°C	1.3		kHz
				25°C	1		
				70°C	0.9		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ ,	$C_L = 20\text{ pF}$ ,	See Figure 1	0°C	125		kHz
				25°C	110		
				70°C	90		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ ,	$C_L = 20\text{ pF}$ ,	0°C	40°		
				25°C	38°		
				70°C	34°		

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25M4C TLC25M4AC TLC25M4BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC25M4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	1.1 10		mV
					Full range	12		
		TLC25M4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	0.9 5		
					Full range	6.5		
		TLC25M4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	0.25 2		
					Full range	3		
$\approx V_{IO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7 300			
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40 600			
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\ \text{k}\Omega$	0°C	3 3.9		V	
				2°C	3.2 3.9			
				70°C	3 4			
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	0°C	0 50		mV	
				25°C	0 50			
				70°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 100\ \text{k}\Omega$	0°C	15 200		V/mV	
				25°C	25 170			
				70°C	15 140			
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60 91		dB	
				25°C	65 91			
				70°C	60 92			
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	0°C	60 92		dB	
				25°C	70 93			
				70°C	60 94			
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	0°C	500 1280		$\mu\text{A}$	
				25°C	420 1120			
				70°C	340 880			

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC25M4C TLC25M4AC TLC25M4BC			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC25M4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC25M4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC25M4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	0.26	2	
					Full range		3	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\ \text{k}\Omega$	0°C	7.8	8.7	V	
				25°C	8	8.7		
				70°C	7.8	8.7		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	0°C	0	50	mV	
				25°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 100\ \text{k}\Omega$	0°C	15	320	V/mV	
				25°C	25	275		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		0°C	60	94	dB	
				25°C	65	94		
				70°C	60	94		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	0°C	60	92	dB	
				25°C	70	93		
				70°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	0°C	690	1600	$\mu\text{A}$	
				25°C	570	1200		
				70°C	440	1120		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC25M4C TLC25M4AC TLC25M4BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ See Figure 1	$C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	0°C	0.46		$\text{V}/\mu\text{s}$	
				25°C	0.43		$\text{V}/\mu\text{s}$	
				70°C	0.36		$\text{V}/\mu\text{s}$	
				0°C	0.43			
				25°C	0.40			
				70°C	0.34			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$	$R_S = 20\ \Omega$	See Figure 2	25°C	32		$\text{nV}/\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ See Figure 1	$C_L = 20\text{ pF}$	$R_L = 100\text{ k}\Omega$	0°C	60		kHz	
				25°C	55			
				70°C	50			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$	$C_L = 20\text{ pF}$	See Figure 1	0°C	610		kHz	
				25°C	525			
				70°C	400			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ See Figure 3	$f = B_1$	$C_L = 20\text{ pF}$	0°C	41°			
				25°C	40°			
				70°C	39°			

**operating characteristics,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS			$T_A$	TLC25M4C TLC25M4AC TLC25M4BC			UNIT
					MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ See Figure 1	$C_L = 20\text{ pF}$	$V_{I(PP)} = 1\text{ V}$	0°C	0.67		$\text{V}/\mu\text{s}$	
				25°C	0.62			
				70°C	0.51			
			$V_{I(PP)} = 5.5\text{ V}$	0°C	0.61			
				25°C	0.56			
				70°C	0.46			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$	$R_S = 20\ \Omega$	See Figure 2	25°C	32		$\text{nV}/\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ See Figure 1	$C_L = 20\text{ pF}$	$R_L = 100\text{ k}\Omega$	0°C	40		kHz	
				25°C	35			
				70°C	30			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$	$C_L = 20\text{ pF}$	See Figure 1	0°C	710		kHz	
				25°C	635			
				70°C	510			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ See Figure 3	$f = B_1$	$C_L = 20\text{ pF}$	0°C	44°			
				25°C	43°			
				70°C	42°			



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

**electrical characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC254Y			TLC25L4Y			TLC25M4Y			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$	Input offset voltage $V_O = 1.4\text{ V}$ , $V_{IC} = 0\text{ V}$ , $R_S = 50\ \Omega$ , See Note 6		1.1	10		1.1	10		1.1	10	mV
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage		1.8			1.1			1.7		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current (see Note 4) $V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.1			0.1			0.1		pA
$I_{IB}$	Input bias current (see Note 4) $V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$		0.6			0.6			0.6		pA
$V_{ICR}$	Common-mode input voltage range (see Note 5)	-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$	High-level output voltage $V_{ID} = 100\text{ mV}$ , $R_L = 100\text{ k}\Omega$	3.2	3.8		3.2	4.1		3.2	3.9		V
$V_{OL}$	Low-level output voltage $V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50		0	50		0	50	mV
$A_{VD}$	Large-signal differential voltage amplification $V_O = 0.25\text{ V}$ , See Note 6	5	23		50	520		25	170		V/mV
CMRR	Common-mode rejection ratio $V_{IC} = V_{ICRmin}$	65	80		65	94		65	91		dB
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ ) $V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		70	97		70	93		dB
$I_{DD}$	Supply current $V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$ , No load		2.7	6.4		0.04	0.068		0.42	1.12	mA

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. For low-bias mode,  $R_L = 1\text{ M}\Omega$ , for medium-bias mode,  $R_L = 100\text{ k}\Omega$ , and for high-bias mode,  $R_L = 10\text{ k}\Omega$ .

**operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC254Y			TLC25L4Y			TLC25M4Y			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $C_L = 20\text{ pF}$ , See Note 6		$V_I(PP) = 1\text{ V}$ 3.6			0.03			0.43		$\text{V}/\mu\text{s}$
			$V_I(PP) = 2.5\text{ V}$ 2.9			0.03			0.40		
$V_n$	Equivalent input noise voltage $f = 1\text{ kHz}$ , $R_S = 20\ \Omega$		2.5			70			32		$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$	Maximum output-swing bandwidth $V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$		320			5			55		kHz
$B_1$	Unity-gain bandwidth $V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$		1.7			0.085			0.525		MHz
$\phi_m$	Phase margin $f = B_1$ , $C_L = 20\text{ pF}$		46°			34°			40°		

NOTE 6: For low-bias mode,  $R_L = 1\text{ M}\Omega$ , for medium-bias mode,  $R_L = 100\text{ k}\Omega$ , and for high-bias mode,  $R_L = 10\text{ k}\Omega$ .



PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC25\_4, TLC25\_4A, and TLC25\_4B are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

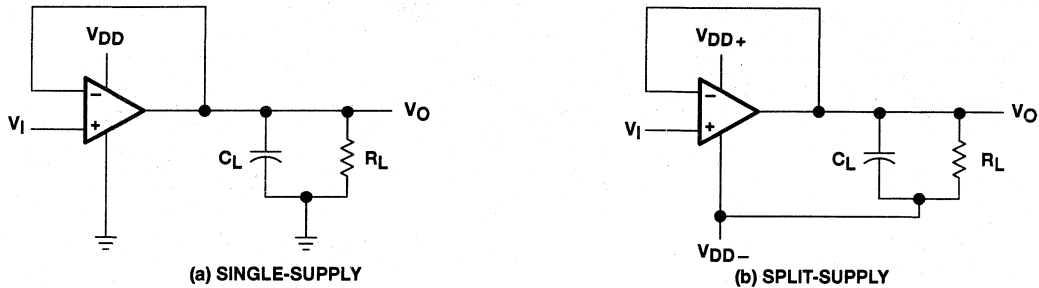


Figure 1. Unity-Gain Amplifier

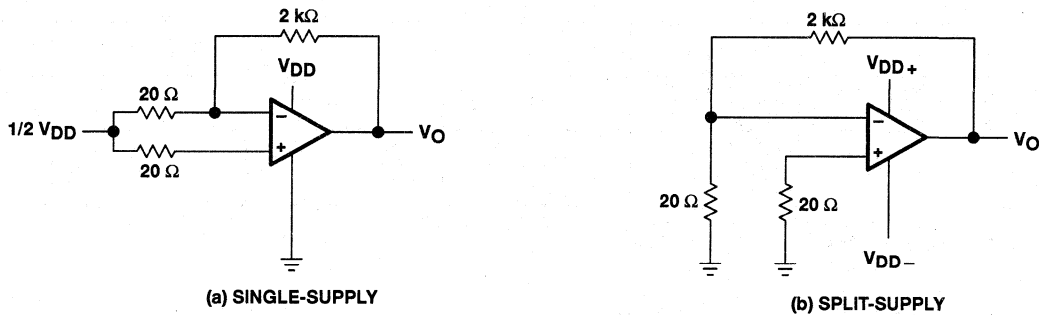


Figure 2. Noise-Test Circuit

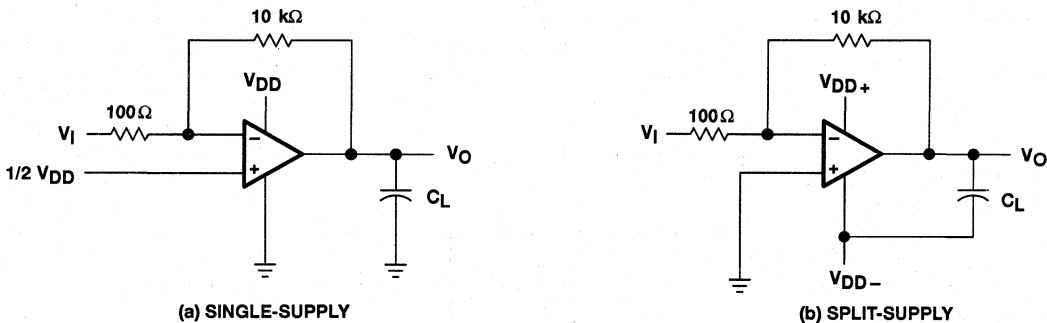


Figure 3. Gain-of-100 Inverting Amplifier

TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
 TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
 LinCMOS™ QUAD OPERATIONAL AMPLIFIERS

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

TYPICAL CHARACTERISTICS

Table of Graphs

				FIGURE
I <sub>DD</sub>	Supply current		vs Supply voltage	4
			vs Free-air temperature	5
A <sub>VD</sub>	Large-signal differential voltage amplification	Low bias	vs Frequency	6
		Medium bias	vs Frequency	7
		High bias	vs Frequency	8
	Phase shift	Low bias	vs Frequency	6
		Medium bias	vs Frequency	7
		High bias	vs Frequency	8

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

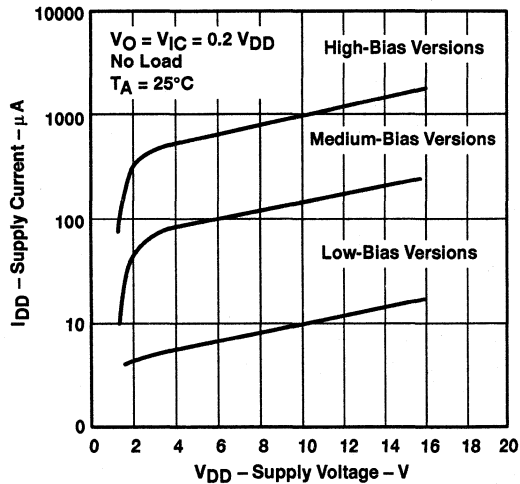


Figure 4

SUPPLY CURRENT  
 vs  
 FREE-AIR TEMPERATURE

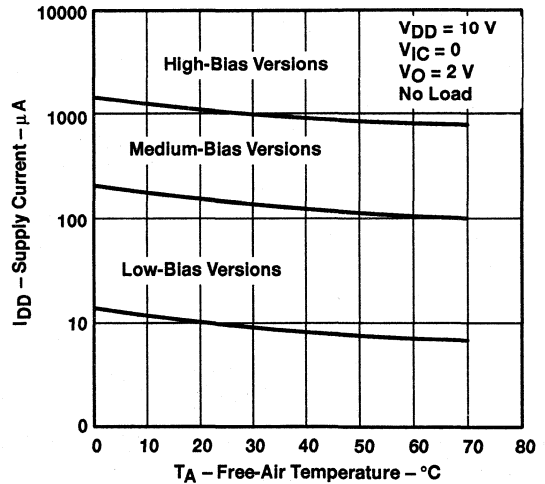


Figure 5

TYPICAL CHARACTERISTICS

LOW-BIAS LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

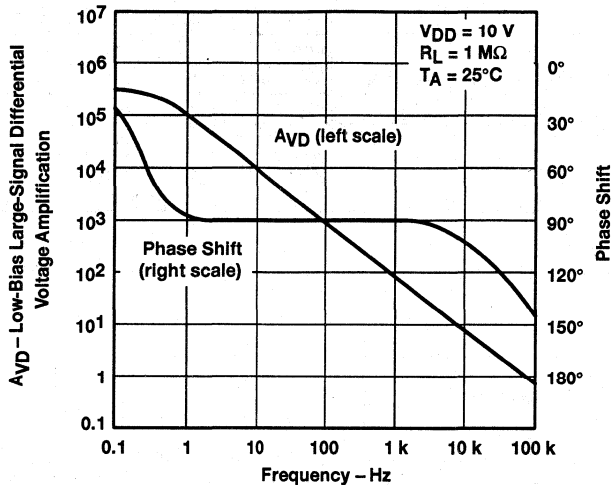


Figure 6

MEDIUM-BIAS LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 AND PHASE SHIFT  
 vs  
 FREQUENCY

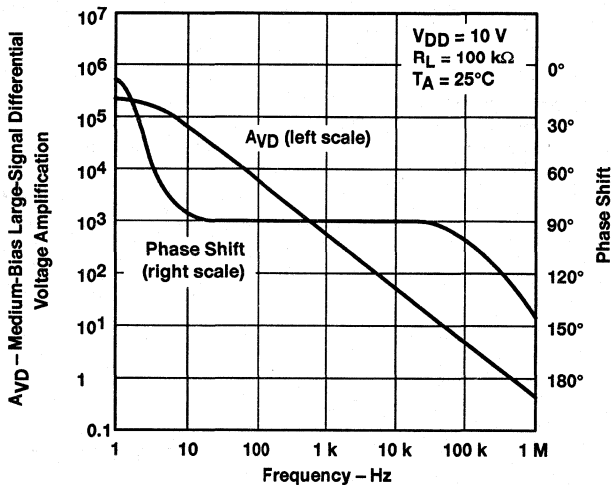


Figure 7

TYPICAL CHARACTERISTICS

HIGH-BIAS LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
AND PHASE SHIFT  
VS  
FREQUENCY

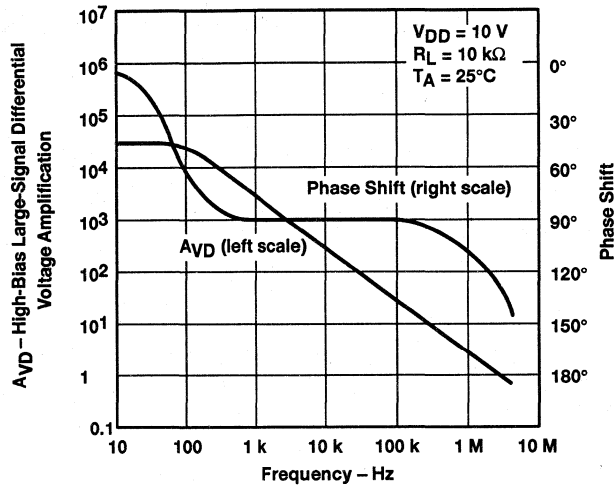


Figure 8

**TLC254, TLC254A, TLC254B, TLC254Y, TLC25L4, TLC25L4A, TLC25L4B  
TLC25L4Y, TLC25M4, TLC25M4A, TLC25M4B, TLC25M4Y  
LinCMOS™ QUAD OPERATIONAL AMPLIFIERS**

SLOS003F – JUNE 1983 – REVISED AUGUST 1994

---

## APPLICATION INFORMATION

### **latch-up avoidance**

Junction-isolated CMOS circuits have an inherent parasitic PNP structure that can function as an SCR. Under certain conditions, this SCR may be triggered into a low-impedance state, resulting in excessive supply current. To avoid such conditions, no voltage greater than 0.3 V beyond the supply rails should be applied to any pin. In general, the operational amplifiers supplies should be established simultaneously with, or before, application of any input signals.

### **output stage considerations**

The amplifier's output stage consists of a source-follower-connected pullup transistor and an open-drain pulldown transistor. The high-level output voltage ( $V_{OH}$ ) is virtually independent of the  $I_{DD}$  selection and increases with higher values of  $V_{DD}$  and reduced output loading. The low-level output voltage ( $V_{OL}$ ) decreases with reduced output current and higher input common-mode voltage. With no load,  $V_{OL}$  is essentially equal to the potential of  $V_{DD-}/GND$ .

### **supply configurations**

Even though the TLC25\_4C series is characterized for single-supply operation, they can be used effectively in a split-supply configuration if the input common-mode voltage ( $V_{ICR}$ ), output swing ( $V_{OL}$  and  $V_{OH}$ ), and supply voltage limits are not exceeded.

### **circuit layout precautions**

Whenever extremely high circuit impedances are used, care must be exercised in layout, construction, board cleanliness, and supply filtering to avoid hum and noise pickup as well as excessive dc leakages.





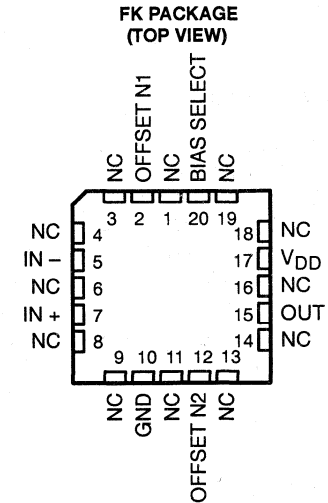
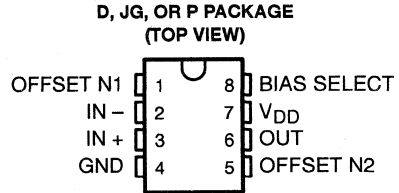
# TLC271, TLC271A, TLC271B LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

- Input Offset Voltage Drift . . . Typically 0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- Wide Range of Supply Voltages Over Specified Temperature Range:
  - 0°C to 70°C . . . 3 V to 16 V
  - 40°C to 85°C . . . 4 V to 16 V
  - 55°C to 125°C . . . 5 V to 16 V
- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix and I-Suffix Types)
- Low Noise . . . 25 nV/ $\sqrt{\text{Hz}}$  Typically at  $f = 1 \text{ kHz}$  (High-Bias Mode)
- Output Voltage Range includes Negative Rail
- High Input Impedance . . .  $10^{12} \Omega$  Typ
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latch-Up Immunity

## description

The TLC271 operational amplifier combines a wide range of input offset voltage grades with low offset voltage drift and high input impedance. In addition, the TLC271 offers a bias-select mode that allows the user to select the best combination of power dissipation and ac performance for a particular application. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.



NC – No internal connection

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IOmax</sub> AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	2 mV 5 mV 10 mV	TLC271BCD TLC271ACD TLC271CD	—	—	TLC271BCP TLC271ACP TLC271CP
-40°C to 85°C	2 mV 5 mV 10 mV	TLC271BID TLC271AID TLC271ID	—	—	TLC271BIP TLC271AIP TLC271IP
-55°C to 125°C	10 mV	TLC271MD	TLC271MFK	TLC271MJG	TLC271MP

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC271BCDR).

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TLC271, TLC271A, TLC271B LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

## DEVICE FEATURES

PARAMETER†	BIAS-SELECT MODE			UNIT
	HIGH	MEDIUM	LOW	
$P_D$	3375	525	50	$\mu W$
SR	3.6	0.4	0.03	V/ $\mu s$
$V_n$	25	32	68	nV/ $\sqrt{Hz}$
$B_1$	1.7	0.5	0.09	MHz
$A_{vD}$	23	170	480	V/mV

† Typical at  $V_{DD} = 5 V$ ,  $T_A = 25^\circ C$

## description (continued)

Using the bias-select option, these cost-effective devices can be programmed to span a wide range of applications that previously required BiFET, NFET or bipolar technology. Three offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC271 (10 mV) to the TLC271B (2 mV) low-offset version. The extremely high input impedance and low bias currents, in conjunction with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available in LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC271. The devices also exhibit low-voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and output are designed to withstand  $-100$ -mA surge currents without sustaining latch-up.

The TLC271 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from  $0^\circ C$  to  $70^\circ C$ . The I-suffix devices are characterized for operation from  $-40^\circ C$  to  $85^\circ C$ . The M-suffix devices are characterized for operation over the full military temperature range of  $-55^\circ C$  to  $125^\circ C$ .

## bias-select feature

The TLC271 offers a bias-select feature that allows the user to select any one of three bias levels depending on the level of performance desired. The tradeoffs between bias levels involve ac performance and power dissipation (see Table 1).



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**bias-select feature (continued)**

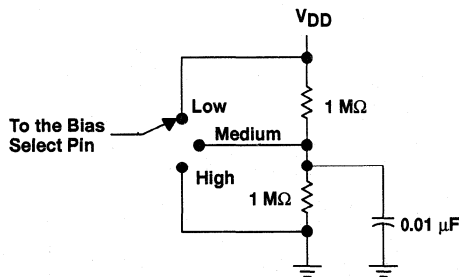
**Table 1. Effect of Bias Selection on Performance**

TYPICAL PARAMETER VALUES T <sub>A</sub> = 25°C, V <sub>DD</sub> = 5 V		MODE			UNIT
		HIGH BIAS R <sub>L</sub> = 10 kΩ	MEDIUM BIAS R <sub>L</sub> = 100 kΩ	LOW BIAS R <sub>L</sub> = 1 MΩ	
P <sub>D</sub>	Power dissipation	3.4	0.5	0.05	mW
SR	Slew rate	3.6	0.4	0.03	V/μs
V <sub>n</sub>	Equivalent input noise voltage at f = 1 kHz	25	32	68	nV/√Hz
B <sub>1</sub>	Unity-gain bandwidth	1.7	0.5	0.09	MHz
φ <sub>m</sub>	Phase margin	46°	40°	34°	
A <sub>VD</sub>	Large-signal differential voltage amplification	23	170	480	V/mV

**bias selection**

Bias selection is achieved by connecting the bias select pin to one of three voltage levels (see Figure 1). For medium-bias applications, it is recommended that the bias select pin be connected to the midpoint between the supply rails. This procedure is simple in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode necessitates using a voltage divider as indicated in Figure 1. The use of large-value resistors in the voltage divider reduces the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor require significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the midpoint can be used if it is within the voltages specified in Figure 1.

**bias selection (continued)**



BIAS MODE	BIAS-SELECT VOLTAGE (single supply)
Low	V <sub>DD</sub>
Medium	1 V to V <sub>DD</sub> - 1 V
High	GND

**Figure 1. Bias Selection for Single-Supply Applications**

**high-bias mode**

In the high-bias mode, the TLC271 series features low offset voltage drift, high input impedance, and low noise. Speed in this mode approaches that of BiFET devices but at only a fraction of the power dissipation. Unity-gain bandwidth is typically greater than 1 MHz.

**medium-bias mode**

The TLC271 in the medium-bias mode features low offset voltage drift, high input impedance, and low noise. Speed in this mode is similar to general-purpose bipolar devices but power dissipation is only a fraction of that consumed by bipolar devices.

# TLC271, TLC271A, TLC271B

## LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A - NOVEMBER 1987 - REVISED AUGUST 1994

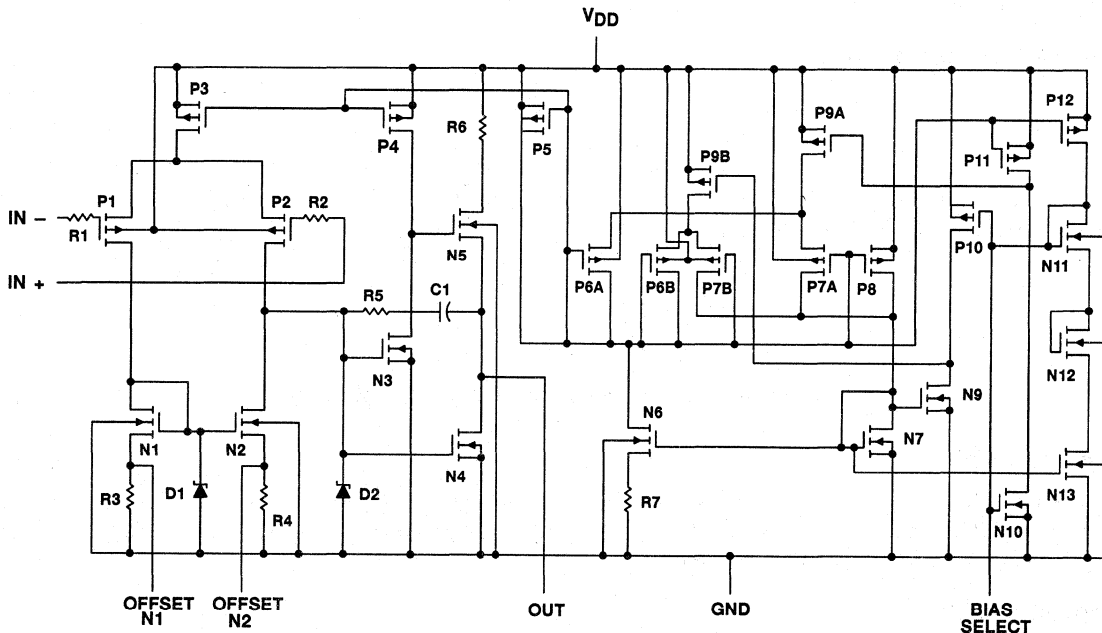
### low-bias mode

In the low-bias mode, the TLC271 features low offset voltage drift, high input impedance, extremely low power consumption, and high differential voltage gain.

#### ORDER OF CONTENTS

TOPIC	BIAS MODE
schematic	all
absolute maximum ratings	all
recommended operating conditions	all
electrical characteristics operating characteristics typical characteristics	high (Figures 2 - 33)
electrical characteristics operating characteristics typical characteristics	medium (Figures 34 - 65)
electrical characteristics operating characteristics typical characteristics	low (Figures 66 - 97)
parameter measurement information	all
application information	all

### equivalent schematic



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	–0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$	$\pm 30$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	–40°C to 85°C
M suffix	–55°C to 125°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	5	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	–0.2	3.5	–0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	–0.2	8.5	–0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	–40	85	–55	125	°C



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC271C, TLC271AC, TLC271BC						UNIT		
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V					
				MIN	TYP	MAX	MIN	TYP	MAX			
V <sub>IO</sub>	Input offset voltage	TLC271C V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 10 kΩ	25°C	1.1		10		1.1		10		mV
			Full range			12				12		
			25°C	0.9		5		0.9		5		
			Full range			6.5				6.5		
			25°C	0.34		2		0.39		2		
			Full range			3				3		
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.8				2		μV/°C		
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1				pA		
			70°C	7		300		7			300	
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6		0.7				pA		
			70°C	40		600		50			600	
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9		-0.3 to 9.2		V		
			Full range	-0.2 to 3.5		-0.2 to 8.5				V		
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 10 kΩ	25°C	3.2	3.8	8		8.5		V		
			0°C	3	3.8	7.8		8.5				
			70°C	3	3.8	7.8		8.4				
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50		0		50		mV
			0°C	0		50		0		50		
			70°C	0		50		0		50		
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 6	25°C	5	23	10		36		V/mV		
			0°C	4	27	7.5		42				
			70°C	4	20	7.5		32				
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	80	65		85		dB		
			0°C	60	84	60		88				
			70°C	60	85	60		88				
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	65	95	65		95		dB		
			0°C	60	94	60		94				
			70°C	60	96	60		96				
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = 0	25°C	-1.4		-1.9		μA				
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	675	1600	950		2000		μA		
			0°C	775	1800	1125		2200				
			70°C	575	1300	750		1700				

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	TA†	TLC271I, TLC271AI, TLC271BI						UNIT
				VDD = 5 V			VDD = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
VIO	Input offset voltage	VO = 1.4 V, VIC = 0 V, RS = 50 Ω, RL = 10 kΩ	25°C	1.1		10	1.1		10	mV
			Full range			13			13	
			25°C	0.9		5	0.9		5	
			Full range			7			7	
			25°C	0.34		2	0.39		2	
			Full range			3.5			3.5	
αVIO	Average temperature coefficient of input offset voltage		25°C to 85°C	1.8			2		μV/°C	
IIO	Input offset current (see Note 4)	VO = VDD/2, VIC = VDD/2	25°C	0.1			0.1		pA	
			85°C	24	1000	26	1000			
IIB	Input bias current (see Note 4)	VO = VDD/2, VIC = VDD/2	25°C	0.6			0.7		pA	
			85°C	200	2000	220	2000			
VICR	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5			-0.2 to 8.5		V	
VOH	High-level output voltage	VID = 100 mV, RL = 10 kΩ	25°C	3.2	3.8		8	8.5	V	
			-40°C	3	3.8		7.8	8.5		
			85°C	3	3.8		7.8	8.5		
VOL	Low-level output voltage	VID = -100 mV, IOL = 0	25°C	0		50	0		mV	
			-40°C	0		50	0			
			85°C	0		50	0			
AVD	Large-signal differential voltage amplification	RL = 10 kΩ, See Note 6	25°C	5	23		10	36	V/mV	
			-40°C	3.5	32		7	46		
			85°C	3.5	19		7	31		
CMRR	Common-mode rejection ratio	VIC = VICRmin	25°C	65	80		65	85	dB	
			-40°C	60	81		60	87		
			85°C	60	86		60	88		
kSVR	Supply-voltage rejection ratio (ΔVDD/ΔVIO)	VO = 1.4 V	25°C	65	95		65	95	dB	
			-40°C	60	92		60	92		
			85°C	60	96		60	96		
II(SEL)	Input current (BIAS SELECT)	VI(SEL) = 0	25°C	-1.4			-1.9		μA	
IDD	Supply current	VO = VDD/2, VIC = VDD/2, No load	25°C	675	1600		950	2000	μA	
			-40°C	950	2200		1375	2500		
			85°C	525	1200		725	1600		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At VDD = 5 V, VO = 0.25 V to 2 V; at VDD = 10 V, VO = 1 V to 6 V.



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TA†	TLC271M						UNIT
			VDD = 5 V			VDD = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 10 kΩ	25°C		1.1	10		1.1	10	mV
		Full range						12	
α <sub>VIO</sub> Average temperature coefficient of input offset voltage		25°C to 125°C		2.1			2.2		μV/°C
I <sub>IO</sub> Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C		0.1			0.1		pA
		125°C		1.4	15		1.8	15	nA
I <sub>IB</sub> Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C		0.6			0.7		pA
		125°C		9	35		10	35	nA
V <sub>ICR</sub> Common-mode input voltage range (see Note 5)		25°C	0 to 4	-0.3 to 4.2			0 to 9	-0.3 to 9.2	V
		Full range	0 to 3.5				0 to 8.5		V
V <sub>OH</sub> High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 10 kΩ	25°C	3.2	3.8		8	8.5		V
		-55°C	3	3.8		7.8	8.5		
		125°C	3	3.8		7.8	8.4		
V <sub>OL</sub> Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C		0	50		0	50	mV
		-55°C		0	50		0	50	
		125°C		0	50		0	50	
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ, See Note 6	25°C	5	23		10	36		V/mV
		-55°C	3.5	35		7	50		
		125°C	3.5	16		7	27		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	80		65	85		dB
		-55°C	60	81		60	87		
		125°C	60	84		60	86		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	65	95		65	95		dB
		-55°C	60	90		60	90		
		125°C	60	97		60	97		
I <sub>I(SEL)</sub> Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = 0	25°C		-1.4			-1.9		μA
I <sub>DD</sub> Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C		675	1600		950	2000	μA
		-55°C		1000	2500		1475	3000	
		125°C		475	1100		625	1400	

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s	
			0°C	4			
			70°C	3			
		$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9			
			0°C	3.1			
			70°C	2.5			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM      Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 98	25°C	320		kHz	
			0°C	340			
			70°C	260			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	1.7		MHz	
			0°C	2			
			70°C	1.3			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 100	25°C	46°			
			0°C	47°			
			70°C	44°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s	
			0°C	5.9			
			70°C	4.3			
		$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6			
			0°C	5.1			
			70°C	3.8			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$	
BOM      Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 98	25°C	200		kHz	
			0°C	220			
			70°C	140			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	2.2		MHz	
			0°C	2.5			
			70°C	1.8			
$\phi_m$ Phase margin	$f = B_1$ , $C_L = 20\text{ pF}$	$V_I = 10\text{ mV}$ , See Figure 100	25°C	49°			
			0°C	50°			
			70°C	46°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s
			-40°C	4.5		
			85°C	2.8		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9		
			-40°C	3.5		
			85°C	2.3		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	320		kHz	
		-40°C	380			
		85°C	250			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	1.7		MHz	
			-40°C	2.6		
			85°C	1.2		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	46°			
		-40°C	49°			
		85°C	43°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s
			-40°C	6.8		
			85°C	4		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6		
			-40°C	5.8		
			85°C	3.5		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	200		kHz	
		-40°C	260			
		85°C	130			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	2.2		MHz	
			-40°C	3.1		
			85°C	1.7		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	49°			
		-40°C	52°			
		85°C	46°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**HIGH-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271M			UNIT	
			MIN	TYP	MAX		
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s	
			-55°C	4.7			
			125°C	2.3			
		$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9			
			-55°C	3.7			
			125°C	2			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	25			nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	320		kHz		
		-55°C	400				
		125°C	230				
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	1.7		MHz		
			-55°C			2.9	
			125°C			1.1	
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	46°				
		-55°C	49°				
		125°C	41°				

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271M			UNIT	
			MIN	TYP	MAX		
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s	
			-55°C	7.1			
			125°C	3.1			
		$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6			
			-55°C	6.1			
			125°C	2.7			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	25			nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	200		kHz		
		-55°C	280				
		125°C	110				
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	2.2		MHz		
			-55°C			3.4	
			125°C			1.6	
$\phi_m$ Phase margin	$f = B_1$ , $C_L = 20\text{ pF}$ , $V_I = 10\text{ mV}$ , See Figure 100	25°C	49°				
		-55°C	52°				
		125°C	44°				

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	2, 3
$\alpha_{VIO}$	Temperature coefficient	Distribution	4, 5
$V_{OH}$	High-level output voltage	vs High-level output current	6, 7
		vs Supply voltage	8
		vs Free-air temperature	9
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	10, 11
		vs Differential input voltage	12
		vs Free-air temperature	13
		vs Low-level output current	14, 15
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	16
		vs Free-air temperature	17
		vs Frequency	28, 29
$I_{IB}$	Input bias current	vs Free-air temperature	18
$I_{IO}$	Input offset current	vs Free-air temperature	18
$V_{IC}$	Common-mode input voltage	vs Supply voltage	19
$I_{DD}$	Supply current	vs Supply voltage	20
		vs Free-air temperature	21
SR	Slew rate	vs Supply voltage	22
		vs Free-air temperature	23
	Bias-select current	vs Supply voltage	24
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	25
$B_1$	Unity-gain bandwidth	vs Free-air temperature	26
		vs Supply voltage	27
$A_{VD}$	Large-signal differential voltage amplification	vs Frequency	28, 29
$\phi_m$	Phase margin	vs Supply voltage	30
		vs Free-air temperature	31
		vs Load capacitance	32
$V_n$	Equivalent input noise voltage	vs Frequency	33
		Phase shift	28, 29



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

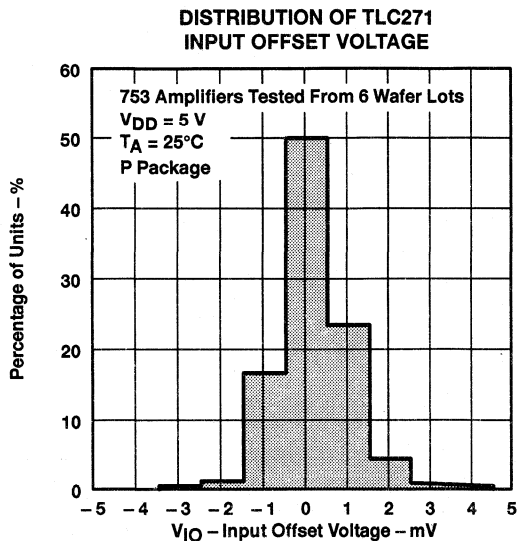


Figure 2

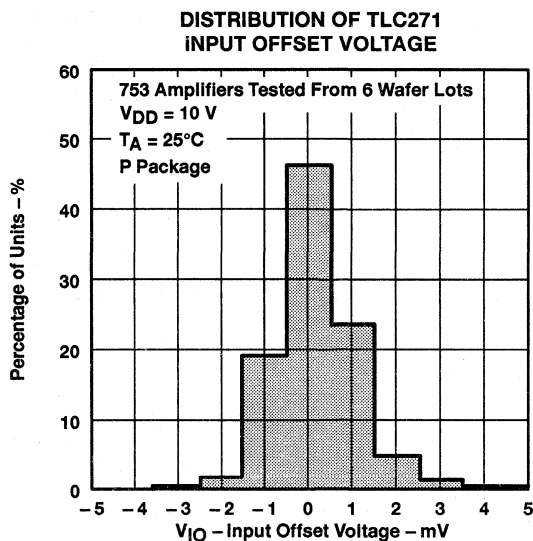


Figure 3

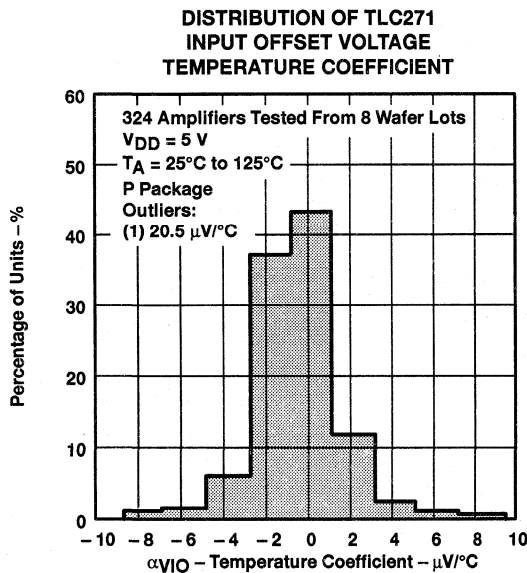


Figure 4

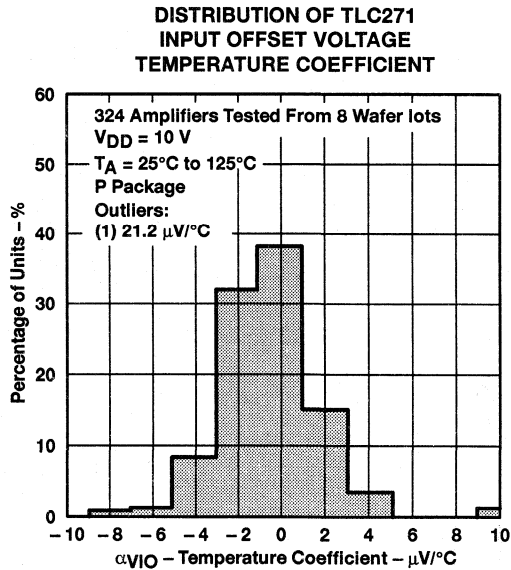


Figure 5

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†**

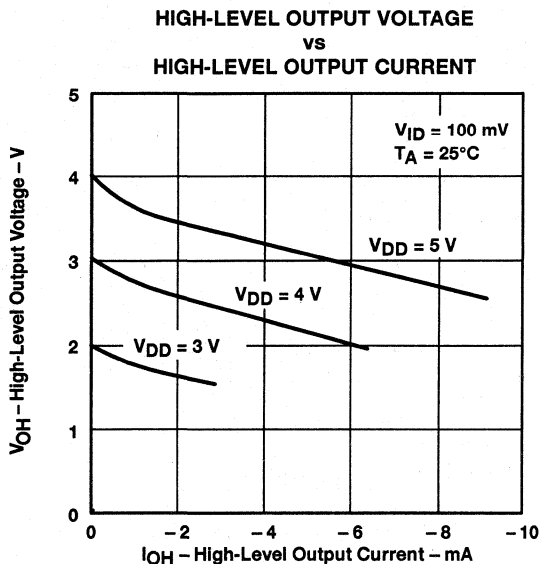


Figure 6

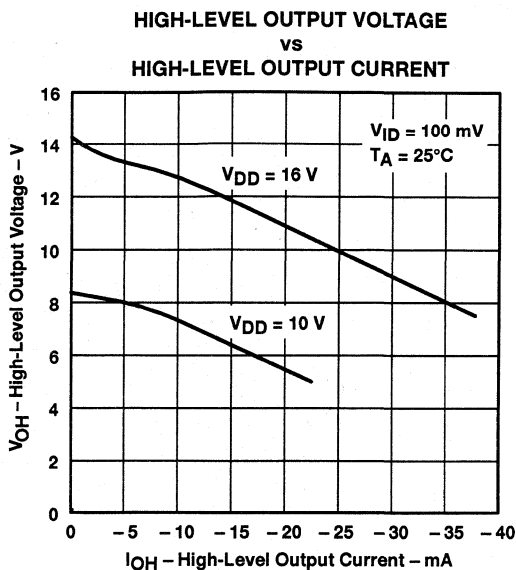


Figure 7

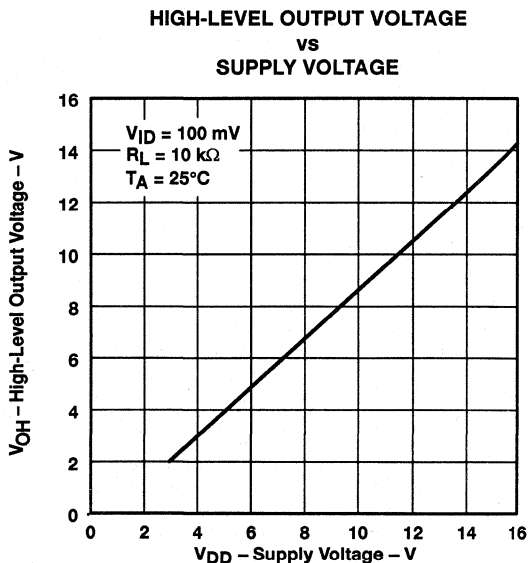


Figure 8

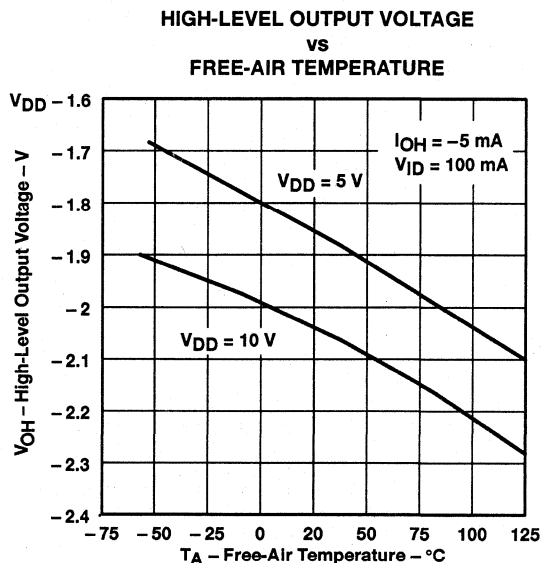


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

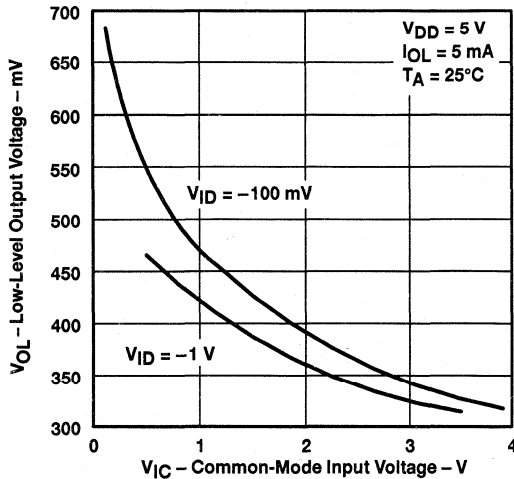


Figure 10

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

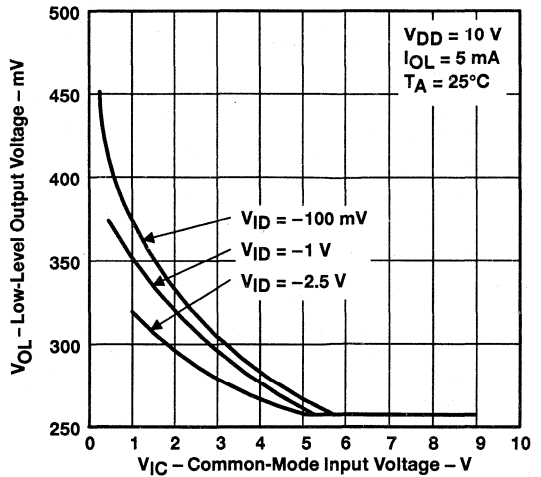


Figure 11

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE

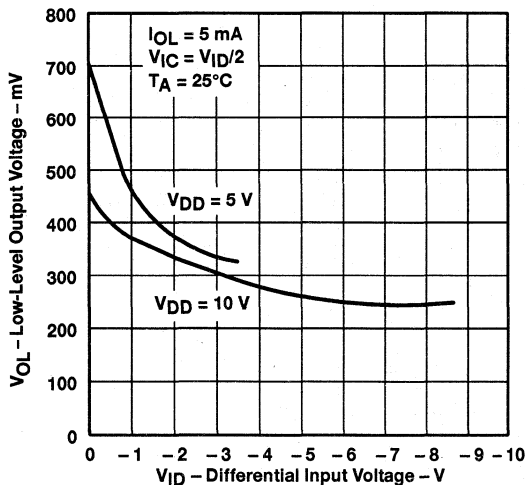


Figure 12

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

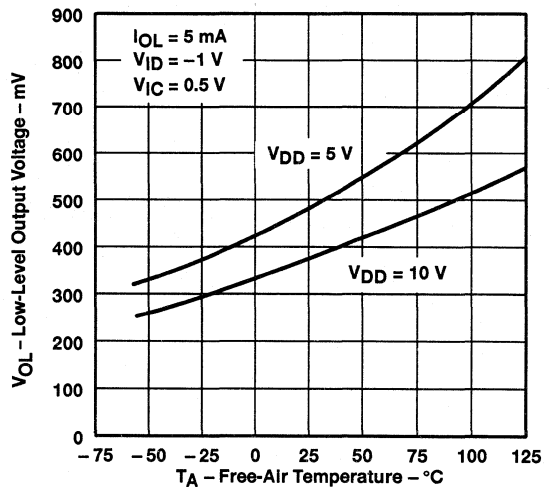


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†**

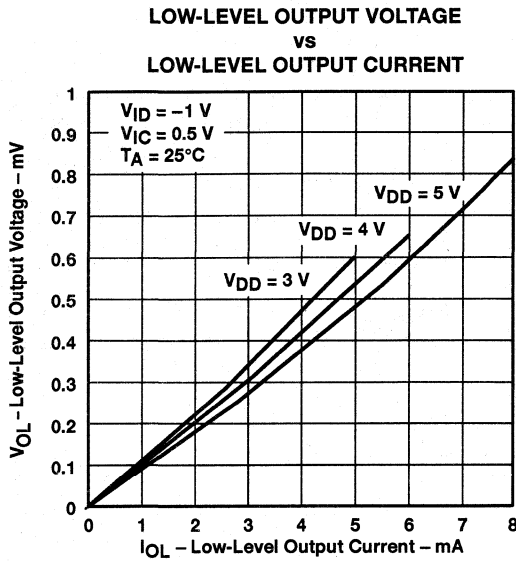


Figure 14

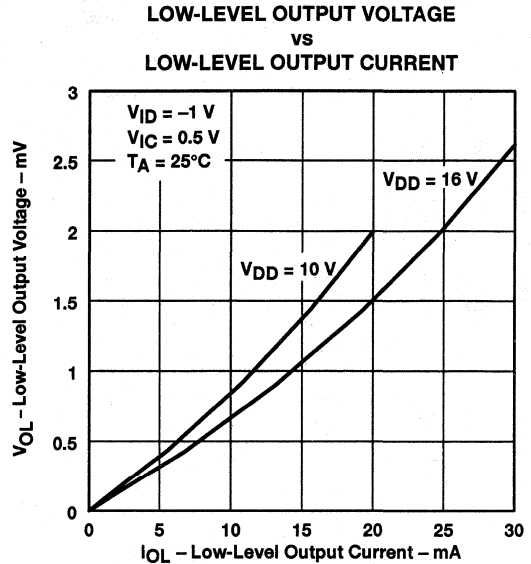


Figure 15

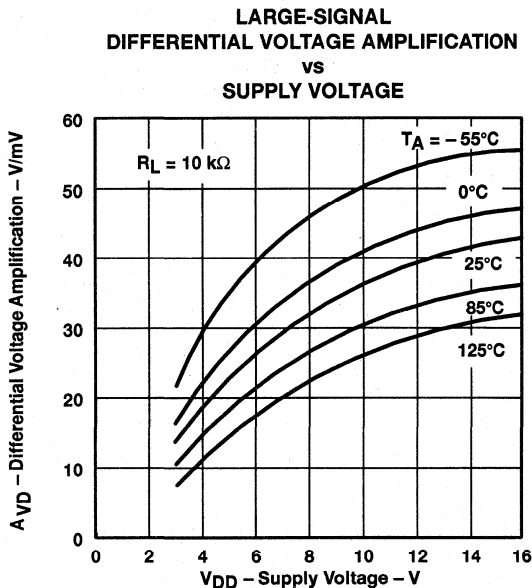


Figure 16

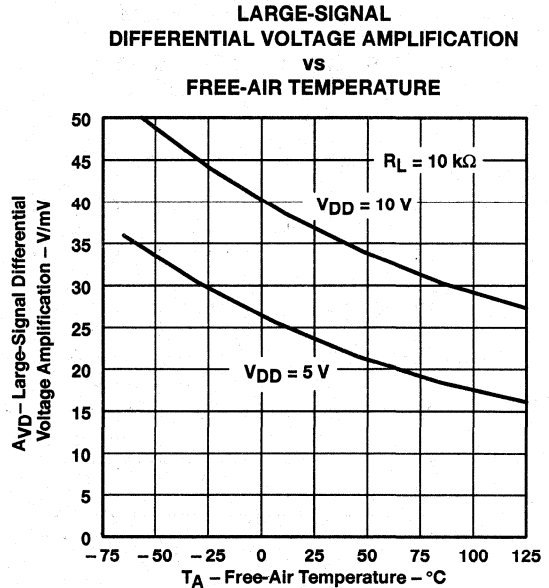


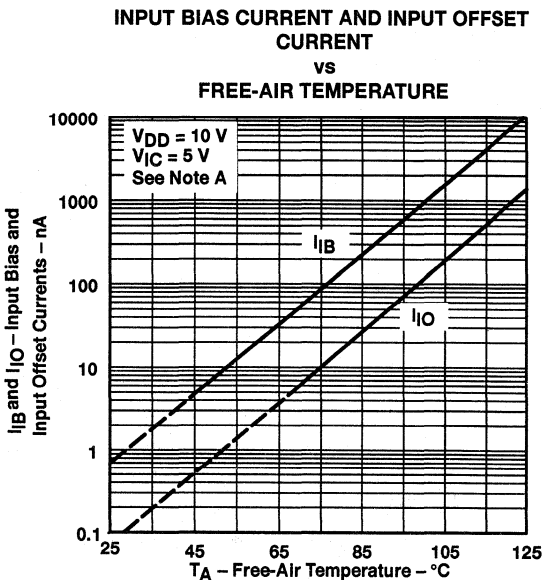
Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 18

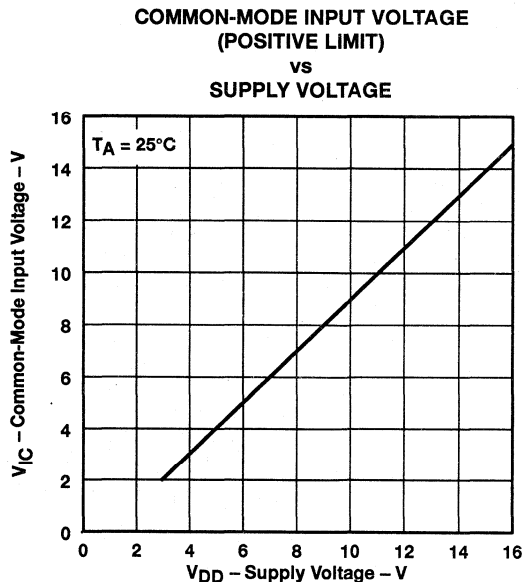


Figure 19

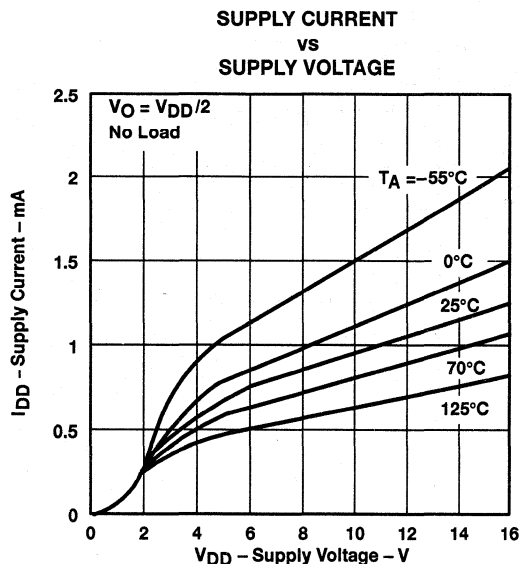


Figure 20

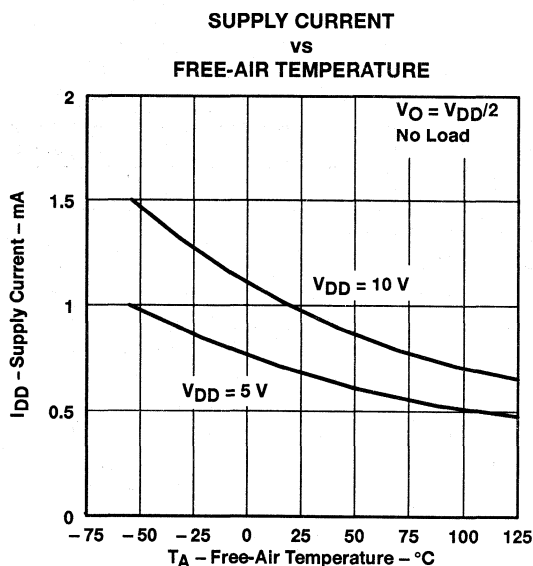


Figure 21

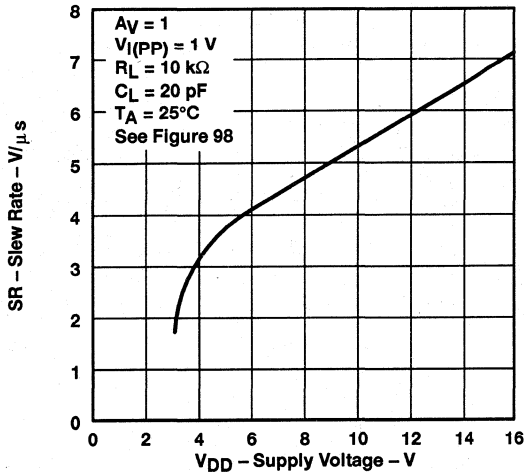
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

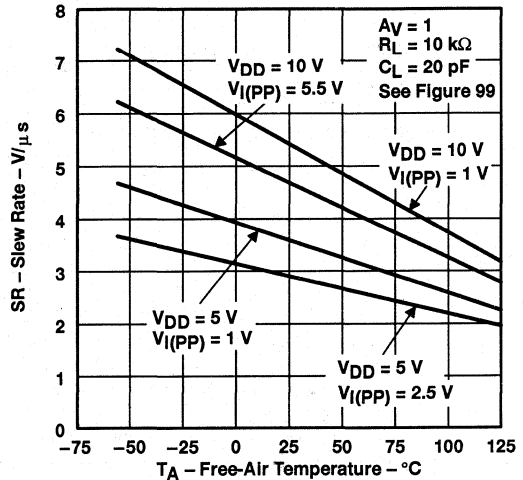
**TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†**

**SLEW RATE**  
**vs**  
**SUPPLY VOLTAGE**



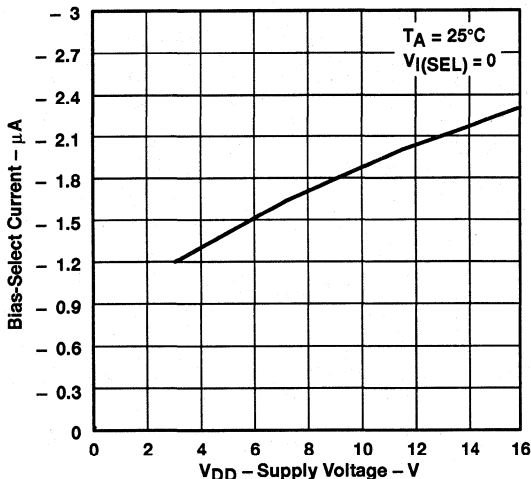
**Figure 22**

**SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**



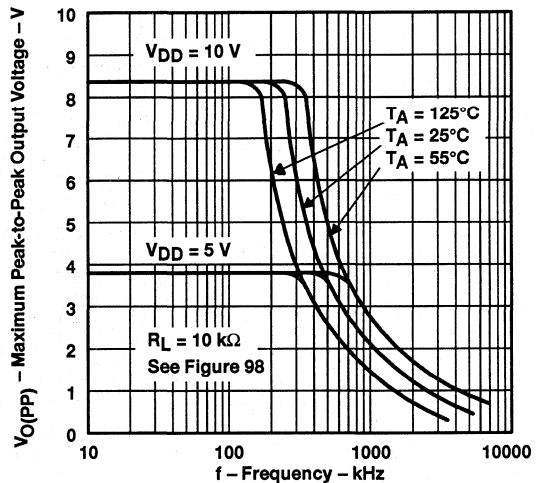
**Figure 23**

**BIAS-SELECT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 24**

**MAXIMUM PEAK-TO-PEAK OUTPUT**  
**VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 25**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

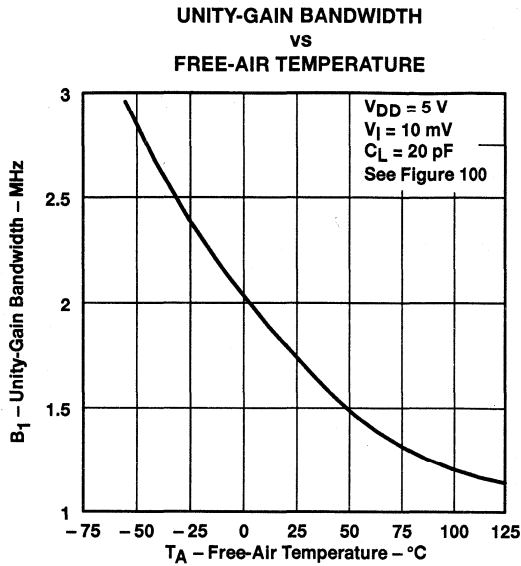


Figure 26

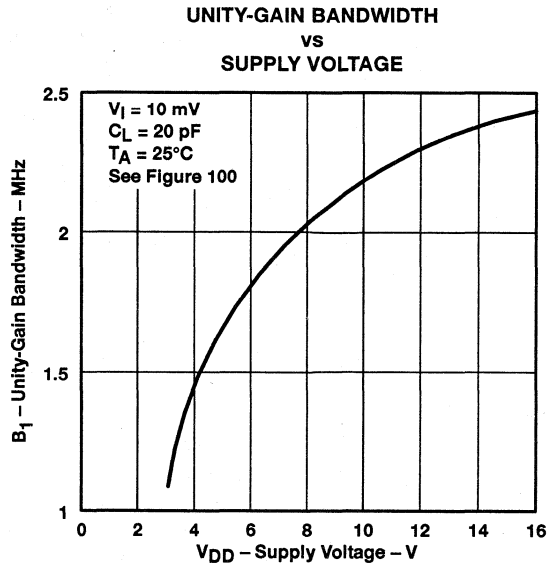


Figure 27

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY

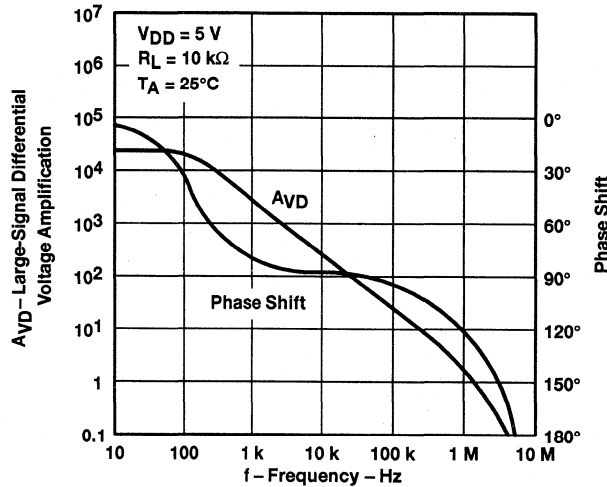


Figure 28

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†**

**LARGE-SCALE DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION AND PHASE SHIFT**

VS  
**FREQUENCY**

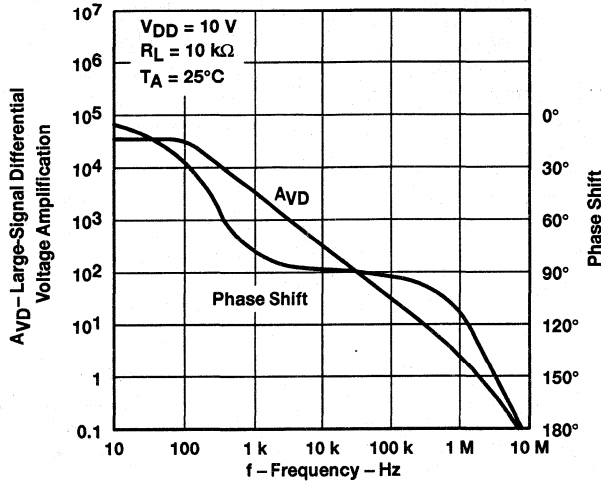


Figure 29

**PHASE MARGIN**  
 VS  
**SUPPLY VOLTAGE**

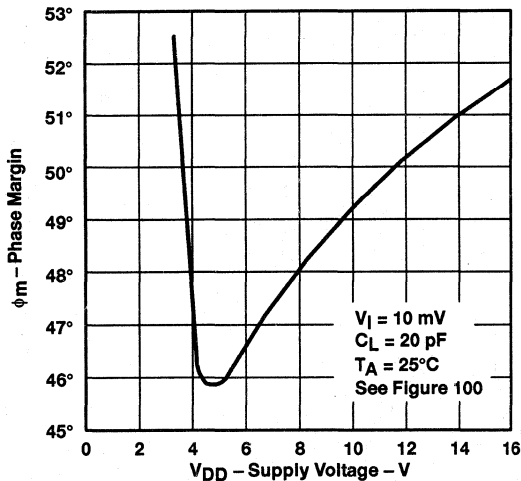


Figure 30

**PHASE MARGIN**  
 VS  
**FREE-AIR TEMPERATURE**

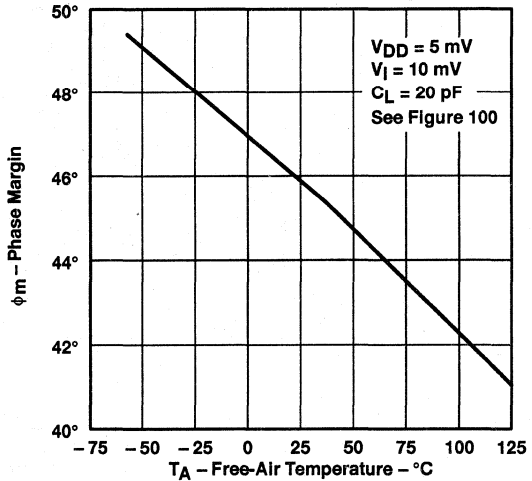


Figure 31

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)†

PHASE MARGIN  
 VS  
 CAPACITIVE LOAD

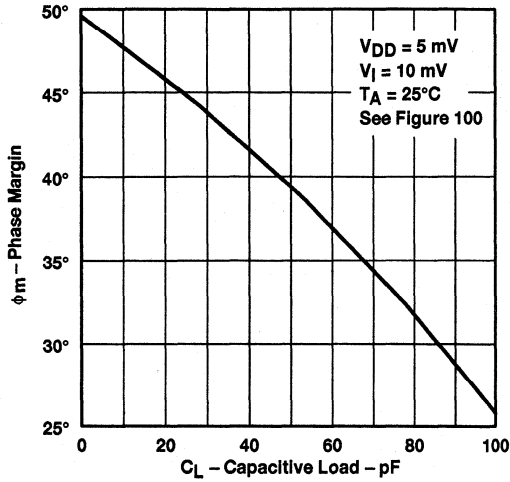


Figure 32

EQUIVALENT NOISE VOLTAGE  
 VS  
 FREQUENCY

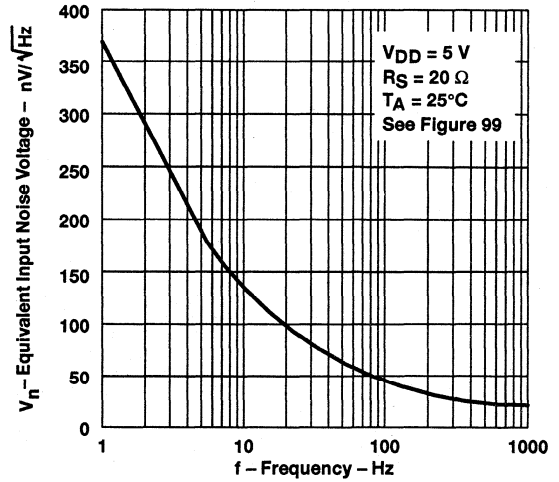


Figure 33

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC271C, TLC271AC, TLC271BC						UNIT
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 R <sub>S</sub> = 50 Ω, R <sub>I</sub> = 100 kΩ	25°C	1.1		10	1.1		10	mV
			Full range			12			12	
			25°C	0.9		5	0.9		5	
			Full range			6.5			6.5	
			25°C	0.25		2	0.26		2	
			Full range			3			3	
αV <sub>IO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.7			2.1			μV/°C
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1			pA
			70°C	7		300	7		300	
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7			pA
			70°C	40		600	50		600	
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V
			Full range	-0.2 to 3.5			-0.2 to 8.5			V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 100 kΩ	25°C	3.2	3.9		8	8.7		V
			0°C	3	3.9		7.8	8.7		
			70°C	3	4		7.8	8.7		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50	0		50	mV
			0°C	0		50	0		50	
			70°C	0		50	0		50	
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 100 kΩ, See Note 6	25°C	25	170		25	275		V/mV
			0°C	15	200		15	320		
			70°C	15	140		15	230		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	91		65	94		dB
			0°C	60	91		60	94		
			70°C	60	92		60	94		
K <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	93		70	93		dB
			0°C	60	92		60	92		
			70°C	60	94		60	94		
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub> /2	25°C	-130			-160			nA
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	105		280	143		300	μA
			0°C	125		320	173		400	
			70°C	85		220	110		280	

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC271I, TLC271AI, TLC271BI						UNIT	
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V				
				MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 100 kΩ	25°C	1.1		10	1.1		10	mV	
			Full range				13				
			25°C	0.9		5	0.9		5		
			Full range				7				
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 85°C	1.7		2.1				μV/°C	
			I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1		pA
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	85°C	24		1000	26		1000	pA	
			I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6		0.7		pA
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		85°C	200		2000	220		2000	pA	
			V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9		-0.3 to 9.2
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 100 kΩ	Full range	-0.2 to 3.5		-0.2 to 8.5				V	
			V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 100 kΩ	25°C	3.2	3.9	8	8.7	V
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	-40°C	3		3.9		7.8		8.7	
			85°C	3		4		7.8		8.7	
			V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50		0
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 100 kΩ, See Note 6	-40°C	0		50		0		50	
			85°C	0		50		0		50	
			A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 100 kΩ, See Note 6	25°C	25	170	25	275	V/mV
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	-40°C	15		270		15		390	
			85°C	15		130		15		220	
			CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	91	65	94	dB
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	-40°C	60		90		60		93	
			85°C	60		90		60		94	
			k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	93	70	93	dB
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub> /2	-40°C	60		91		60		91	
			85°C	60		94		60		94	
			I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub> /2	25°C	-130		-160		nA
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	105		280		143		300	
			-40°C	158		400		225		450	
			I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	85°C	80		200		103

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> <sup>†</sup>	TLC271M						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 100 kΩ	25°C		1.1	10		1.1	10	mV
		Full range			12			12	
α <sub>VIO</sub> Average temperature coefficient of input offset voltage		25°C to 125°C		1.7			2.1		μV/°C
I <sub>IO</sub> Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C		0.1			0.1		pA
		125°C		1.4	15		1.8	15	nA
I <sub>IB</sub> Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C		0.6			0.7		pA
		125°C		9	35		10	35	nA
V <sub>ICR</sub> Common-mode input voltage range (see Note 5)		25°C	0 to 4	-0.3 to 4.2		0 to 9	-0.3 to 9.2		V
		Full range	0 to 3.5			0 to 8.5			V
V <sub>OH</sub> High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 100 kΩ	25°C	3.2	3.9		8	8.7		V
		-55°C	3	3.9		7.8	8.6		
		125°C	3	4		7.8	8.6		
V <sub>OL</sub> Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C		0	50		0	50	mV
		-55°C		0	50		0	50	
		125°C		0	50		0	50	
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 10 kΩ See Note 6	25°C	25	170		25	275		V/mV
		-55°C	15	290		15	420		
		125°C	15	120		15	190		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	91		65	94		dB
		-55°C	60	89		60	93		
		125°C	60	91		60	93		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	93		70	93		dB
		-55°C	60	91		60	91		
		125°C	60	94		60	94		
I <sub>I(SEL)</sub> Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub> /2	25°C		-130			-160		nA
I <sub>DD</sub> Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C		105	280		143	300	μA
		-55°C		170	440		245	500	
		125°C		70	180		90	240	

<sup>†</sup> Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.





**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.43		V/ $\mu$ s	
			0°C	0.46			
			70°C	0.36			
		$V_I(PP) = 2.5\text{ V}$	25°C	0.40			
			0°C	0.43			
			70°C	0.34			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ , See Figure 98	$C_L = 20\text{ pF}$ , See Figure 98	25°C	55		kHz	
			0°C	60			
			70°C	50			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	525		kHz	
			0°C	600			
			70°C	400			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 100	25°C	40°			
			0°C	41°			
			70°C	33°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.62		V/ $\mu$ s	
			0°C	0.67			
			70°C	0.51			
		$V_I(PP) = 5.5\text{ V}$	25°C	0.56			
			0°C	0.61			
			70°C	0.46			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ , See Figure 98	$C_L = 20\text{ pF}$ , See Figure 98	25°C	35		kHz	
			0°C	40			
			70°C	30			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	635		kHz	
			0°C	710			
			70°C	510			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 100	25°C	43°			
			0°C	44°			
			70°C	42°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271I, TLC271AI, TLC271BI			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.43			V/ $\mu$ s
			-40°C	0.51			
			85°C	0.35			
		$V_I(PP) = 2.5\text{ V}$	25°C	0.40			
			-40°C	0.48			
			85°C	0.32			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ , See Figure 98	$C_L = 20\text{ pF}$ , See Figure 98	25°C	55			kHz
			-40°C	75			
			85°C	45			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	525			MHz
			-40°C	770			
			85°C	370			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 100	25°C	40°			
			-40°C	43°			
			85°C	38°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS		$T_A$	TLC271I, TLC271AI, TLC271BI			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.62			V/ $\mu$ s
			-40°C	0.77			
			85°C	0.47			
		$V_I(PP) = 5.5\text{ V}$	25°C	0.56			
			-40°C	0.70			
			85°C	0.44			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH} \cdot 3$ , $R_L = 100\text{ k}\Omega$ , See Figure 98	$C_L = 20\text{ pF}$ , See Figure 98	25°C	35			kHz
			-40°C	45			
			85°C	25			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$	25°C	635			kHz
			-40°C	880			
			85°C	480			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 100	25°C	43°			
			-40°C	46°			
			85°C	41°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**MEDIUM-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 100 kΩ, C <sub>L</sub> = 20 pF, See Figure 98	V <sub>I(PP)</sub> = 1 V	25°C	0.43		V/μs
			-55°C	0.54		
			125°C	0.29		
		V <sub>I(PP)</sub> = 2.5 V	25°C	0.40		
			-55°C	0.50		
			125°C	0.28		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, See Figure 99	R <sub>S</sub> = 20 Ω, 25°C	32		nV/√Hz	
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>O(H)</sub> , R <sub>L</sub> = 100 kΩ, See Figure 98	C <sub>L</sub> = 20 pF, See Figure 98	25°C	55		kHz
			-55°C	80		
			125°C	40		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, See Figure 100	C <sub>L</sub> = 20 pF,	25°C	525		kHz
			-55°C	850		
			125°C	330		
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF,	f = B <sub>1</sub> , See Figure 100	25°C	40°		
			-55°C	43°		
			125°C	36°		

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 100 kΩ, C <sub>L</sub> = 20 pF, See Figure 98	V <sub>I(PP)</sub> = 1 V	25°C	0.62		V/μs
			-55°C	0.81		
			125°C	0.38		
		V <sub>I(PP)</sub> = 5.5 V	25°C	0.56		
			-55°C	0.73		
			125°C	0.35		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, See Figure 99	R <sub>S</sub> = 20 Ω, 25°C	32		nV/√Hz	
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>O(H)</sub> , R <sub>L</sub> = 100 kΩ, See Figure 98	C <sub>L</sub> = 20 pF, See Figure 98	25°C	35		kHz
			-55°C	50		
			125°C	20		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, See Figure 100	C <sub>L</sub> = 20 pF,	25°C	635		kHz
			-55°C	960		
			125°C	440		
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF,	f = B <sub>1</sub> , See Figure 100	25°C	43°		
			-55°C	47°		
			125°C	39°		



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)**

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution	34, 35
$\alpha_{VIO}$	Temperature coefficient	Distribution	36, 37
$V_{OH}$	High-level output voltage	vs High-level output current	38, 39
		vs Supply voltage	40
		vs Free-air temperature	41
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	42, 43
		vs Differential input voltage	44
		vs Free-air temperature	45
		vs Low-level output current	46, 47
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	48
		vs Free-air temperature	49
		vs Frequency	60, 61
$I_{IB}$	Input bias current	vs Free-air temperature	50
$I_{IO}$	Input offset current	vs Free-air temperature	50
$V_I$	Maximum input voltage	vs Supply voltage	51
$I_{DD}$	Supply current	vs Supply voltage	52
		vs Free-air temperature	53
SR	Slew rate	vs Supply voltage	54
		vs Free-air temperature	55
	Bias-select current	vs Supply voltage	56
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	57
$B_1$	Unity-gain bandwidth	vs Free-air temperature	58
		vs Supply voltage	59
$\phi_m$	Phase margin	vs Supply voltage	62
		vs Free-air temperature	63
		vs Load capacitance	64
$V_n$	Equivalent input noise voltage	vs Frequency	65
		Phase shift	vs Frequency

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE

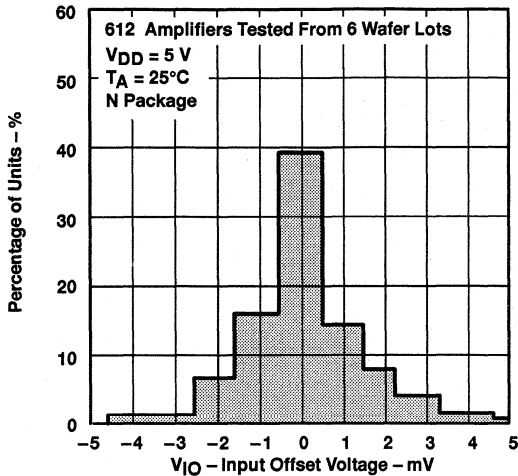


Figure 34

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE

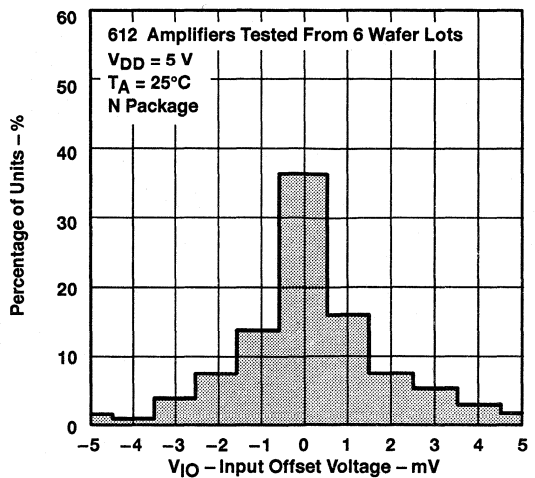


Figure 35

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

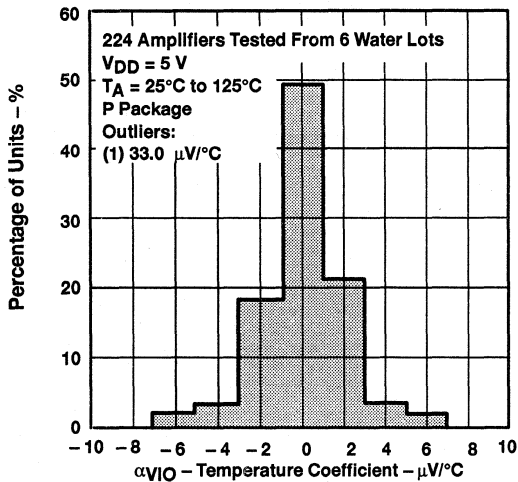


Figure 36

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

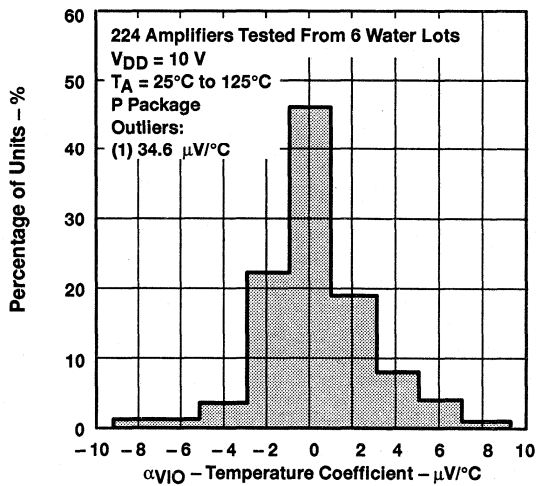


Figure 37

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1984

**TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†**

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**HIGH-LEVEL OUTPUT CURRENT**

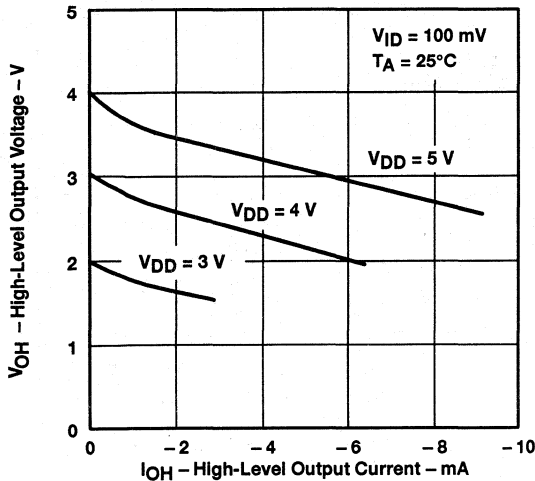


Figure 38

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**HIGH-LEVEL OUTPUT CURRENT**

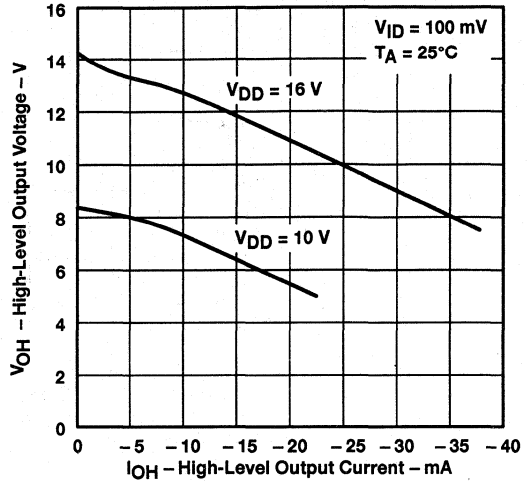


Figure 39

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**SUPPLY VOLTAGE**

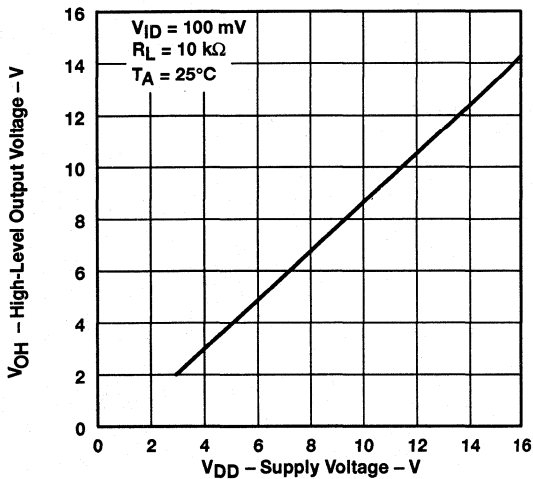


Figure 40

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**

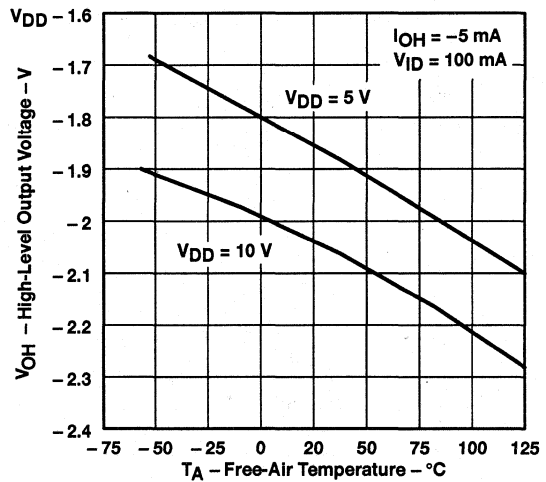


Figure 41

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

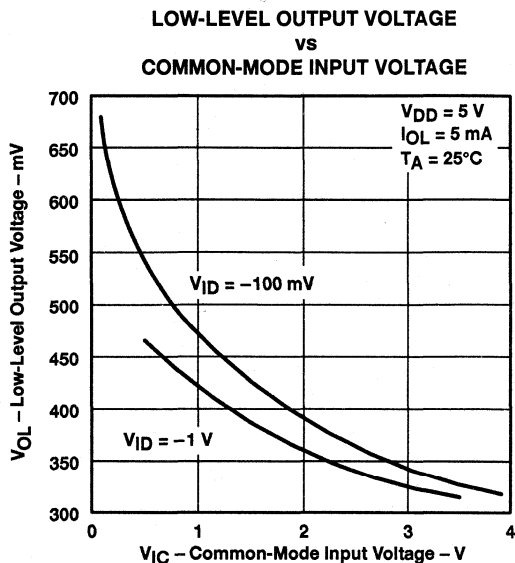


Figure 42

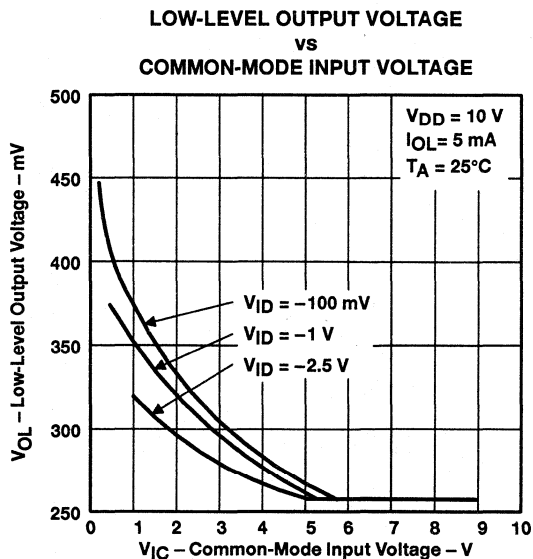


Figure 43

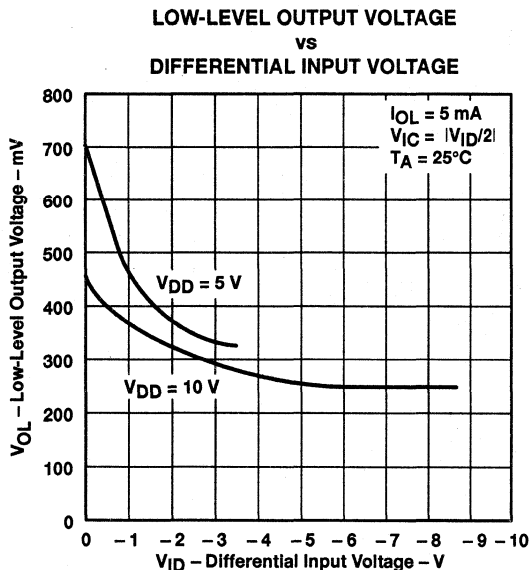


Figure 44

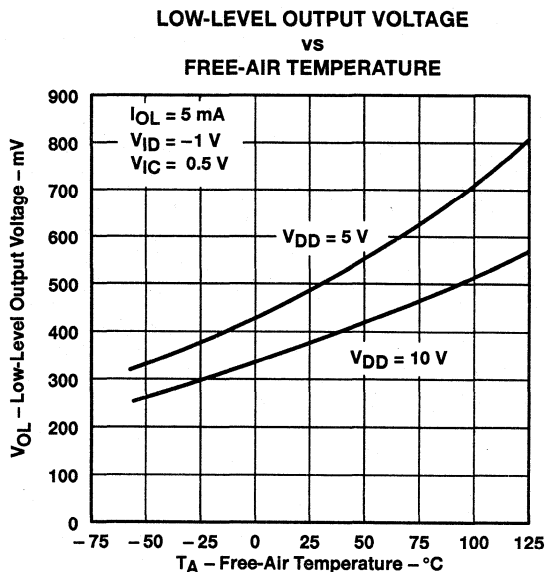


Figure 45

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

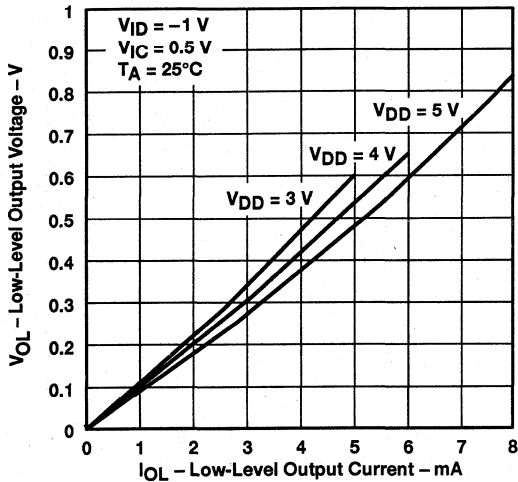


Figure 46

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

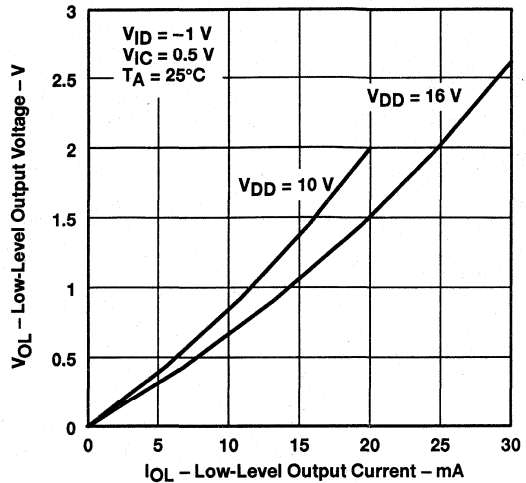


Figure 47

**LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
SUPPLY VOLTAGE**

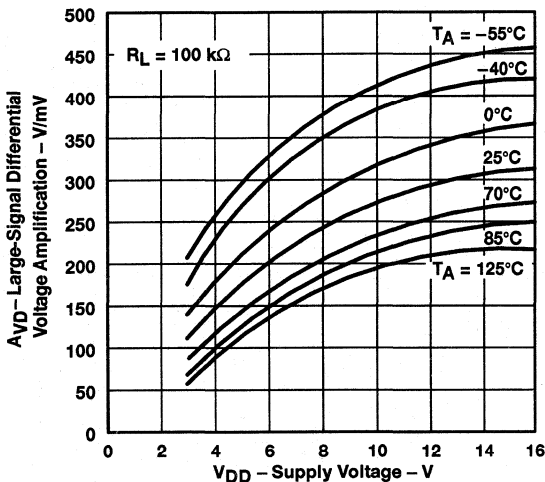


Figure 48

**LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
FREE-AIR TEMPERATURE**

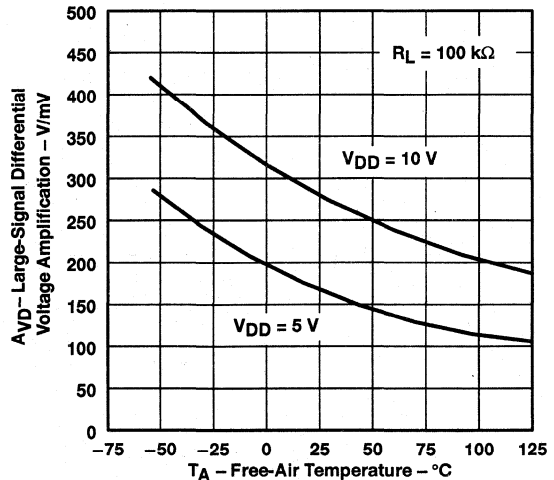


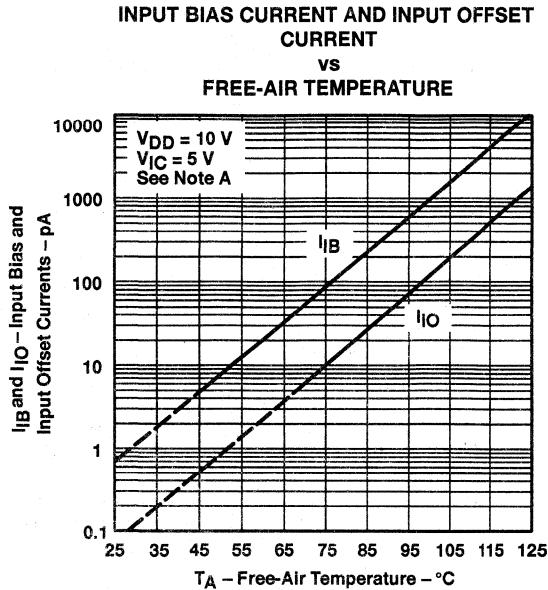
Figure 49

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

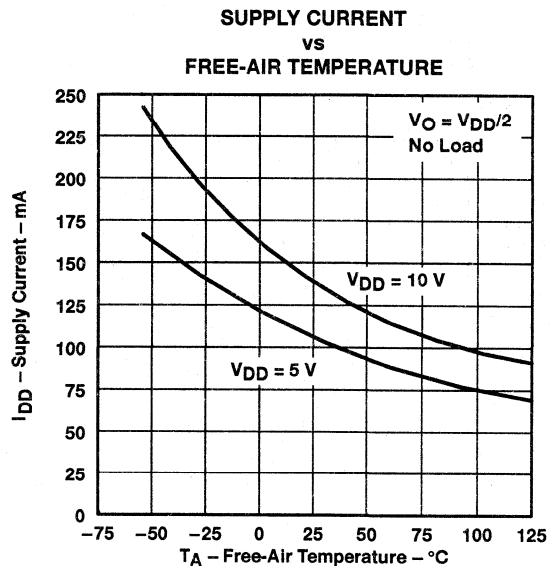
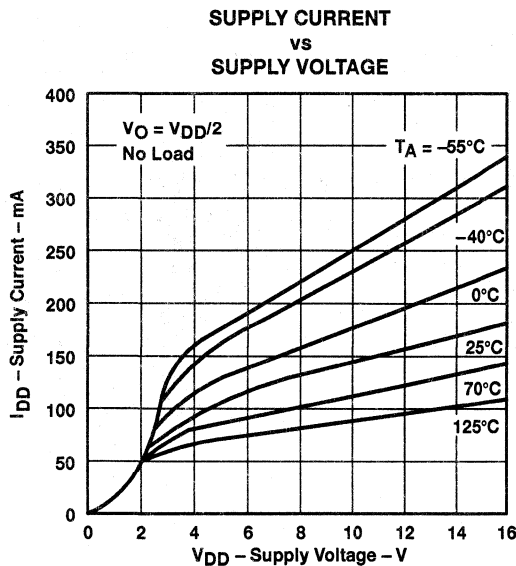
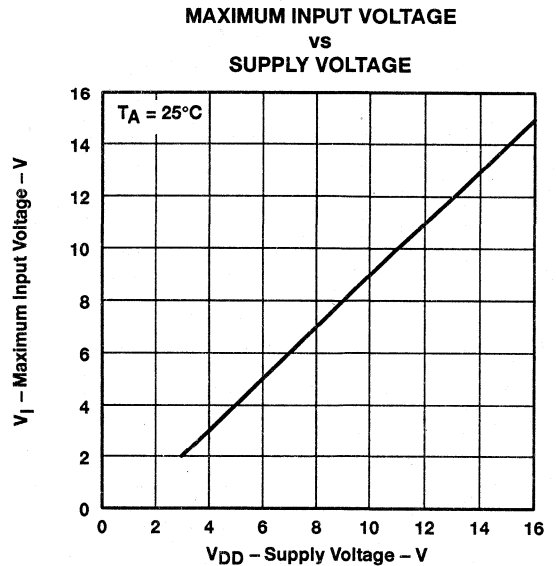




TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†**

**SLEW RATE**  
**vs**  
**SUPPLY VOLTAGE**

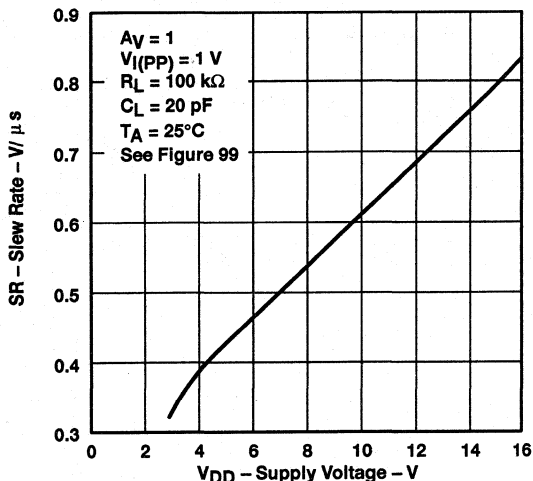


Figure 54

**SLEW RATE**  
**vs**  
**FREE-AIR TEMPERATURE**

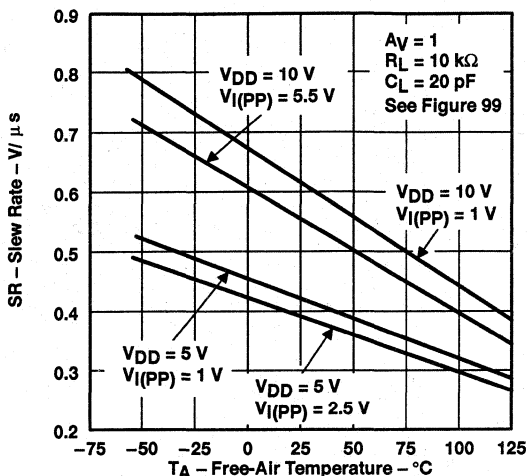


Figure 55

**BIAS-SELECT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**

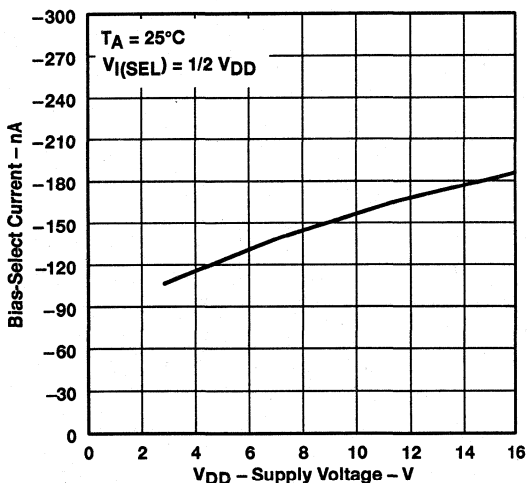


Figure 56

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**

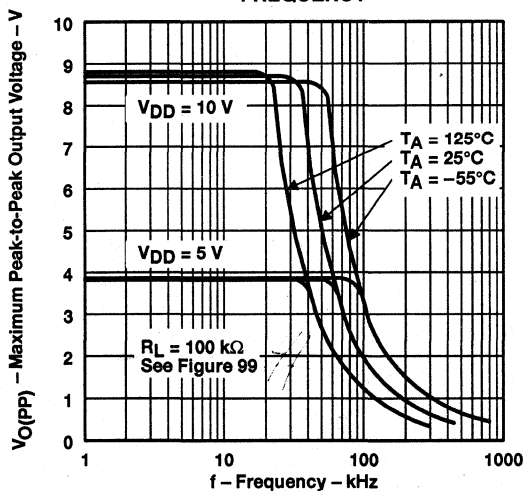


Figure 57

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

UNITY-GAIN BANDWIDTH  
 vs  
 FREE-AIR TEMPERATURE

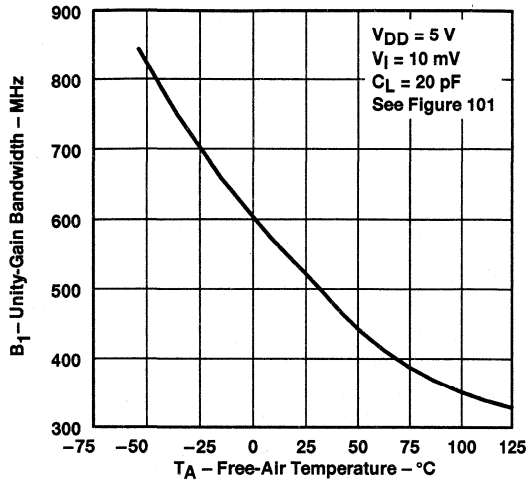


Figure 58

UNITY-GAIN BANDWIDTH  
 vs  
 SUPPLY VOLTAGE

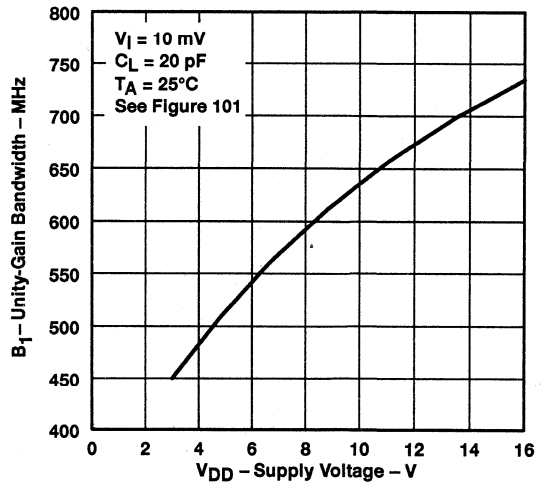


Figure 59

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

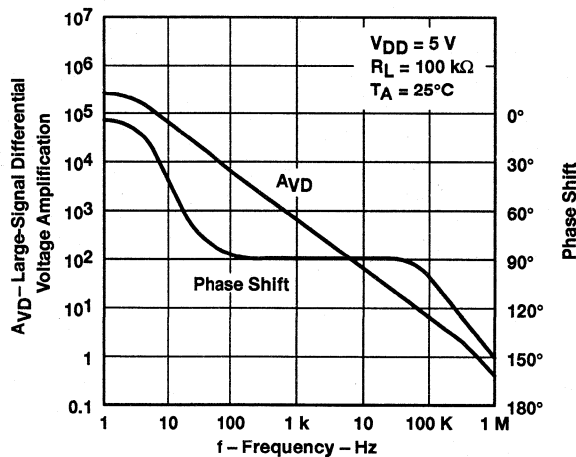


Figure 60

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT vs FREQUENCY**

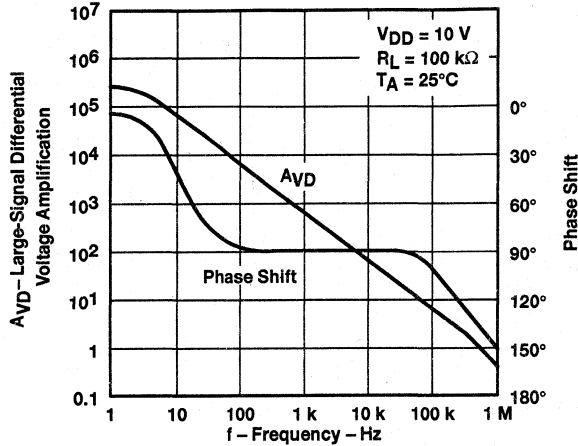


Figure 61

**PHASE MARGIN vs SUPPLY VOLTAGE**

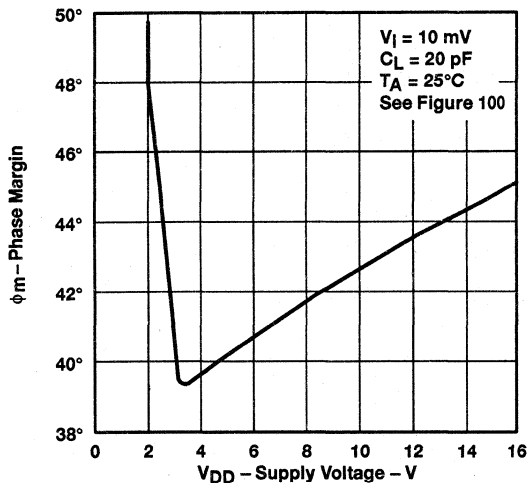


Figure 62

**PHASE MARGIN vs FREE-AIR TEMPERATURE**

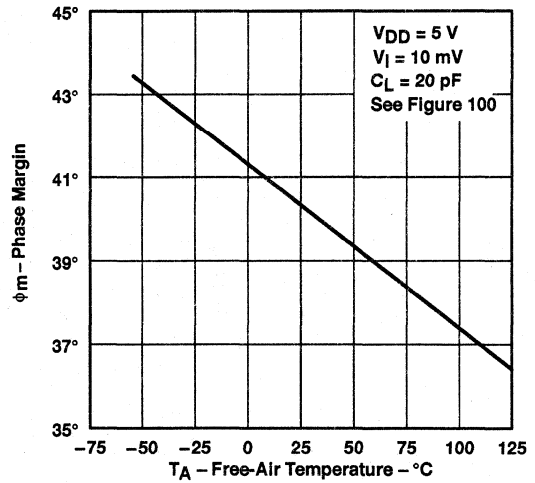


Figure 63

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)†

PHASE MARGIN  
 vs  
 CAPACITIVE LOAD

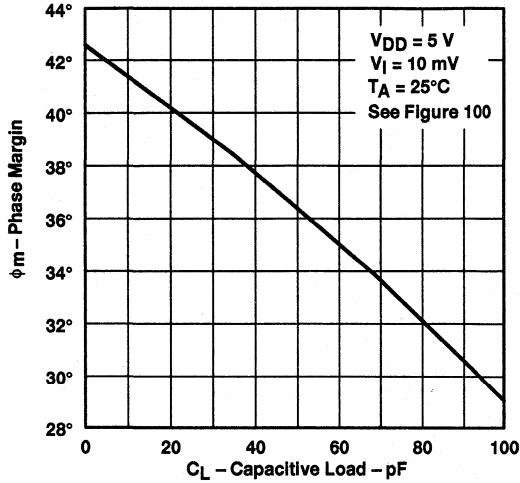


Figure 64

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

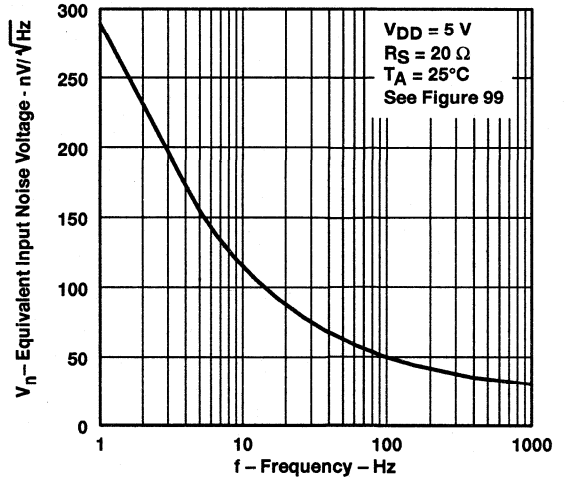


Figure 65

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC271, TLC271A, TLC271B

## LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

### LOW-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC271C, TLC271AC, TLC271BC						UNIT
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>I</sub> = 1 MΩ	25°C	1.1		10	1.1		10	mV
			Full range			12			12	
			25°C	0.9		5	0.9		5	
			Full range			6.5			6.5	
			25°C	0.24		2	0.26		2	
			Full range			3			3	
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 70°C	1.1			1			μV/°C
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1			pA
			70°C	7		300	8		300	
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7			pA
			70°C	40		600	50		600	
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2		-0.2 to 9	-0.3 to 9.2		V
			Full range	-0.2 to 3.5			-0.2 to 8.5			V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 MΩ	25°C	3.2	4.1		8	8.9		V
			0°C	3	4.1		7.8	8.9		
			70°C	3	4.2		7.8	8.9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0		50	0		50	mV
			0°C	0		50	0		50	
			70°C	0		50	0		50	
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 MΩ, See Note 6	25°C	50	520		50	870		V/mV
			0°C	50	700		50	1030		
			70°C	50	380		50	660		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	94		65	97		dB
			0°C	60	95		60	97		
			70°C	60	95		60	97		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	97		70	97		dB
			0°C	60	97		60	97		
			70°C	60	98		60	98		
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub>	25°C	65			95			nA
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	10		17	14		23	μA
			0°C	12		21	18		33	
			70°C	8		14	11		20	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



### LOW-BIAS MODE

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	T <sub>A</sub> †	TLC271I, TLC271AI, TLC271BI						UNIT
				V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 1 MΩ	25°C	1.1	10		1.1	10	mV	
			Full range		13		13			
			25°C	0.9	5		0.9	5		
			Full range		7		7			
			25°C	0.24	2		0.26	2		
			Full range		3.5		3.5			
α <sub>VIO</sub>	Average temperature coefficient of input offset voltage		25°C to 85°C	1.1		1		μV/°C		
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1		pA		
			85°C	24	1000	26	1000			
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6		0.7		pA		
			85°C	200	2000	220	2000			
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V		
			Full range	-0.2 to 3.5		-0.2 to 8.5		V		
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 MΩ	25°C	3	4.1	8	8.9	V		
			-40°C	3	4.1	7.8	8.9			
			85°C	3	4.2	7.8	8.9			
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0	50	0	50	mV		
			-40°C	0	50	0	50			
			85°C	0	50	0	50			
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 MΩ See Note 6	25°C	50	520	50	870	V/mV		
			-40°C	50	900	50	1550			
			85°C	50	330	50	585			
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	94	65	97	dB		
			-40°C	60	95	60	97			
			85°C	60	95	60	98			
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	97	70	97	dB		
			-40°C	60	97	60	97			
			85°C	60	98	60	98			
I <sub>I(SEL)</sub>	Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub>	25°C	65		95	nA			
I <sub>DD</sub>	Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	10	17	14	23	μA		
			-40°C	16	27	25	43			
			85°C	17	13	10	18			

† Full range is -40 to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**electrical characteristics at specified free-air temperature (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC271M						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0 V, R <sub>S</sub> = 50 Ω, R <sub>L</sub> = 1 MΩ	25°C	1.1		10	1.1		10	mV
		Full range			12			12	
α <sub>VIO</sub> Average temperature coefficient of input offset voltage		25°C to 125°C	1.4			1.4			μV/°C
I <sub>IO</sub> Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1			0.1			pA
		125°C	1.4		15	1.8		15	nA
I <sub>IB</sub> Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6			0.7			pA
		125°C	9		35	10		35	nA
V <sub>ICR</sub> Common-mode input voltage range (see Note 5)		25°C	0 to 4	-0.3 to 4.2		0 to 9	-0.3 to 9.2		V
		Full range	0 to 3.5			0 to 8.5			V
V <sub>OH</sub> High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 MΩ	25°C	3.2 4.1			8 8.9			V
		-55°C	3 4.1			7.8 8.8			
		125°C	3 4.2			7.8 9			
V <sub>OL</sub> Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0 50			0 50			mV
		-55°C	0 50			0 50			
		125°C	0 50			0 50			
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 1 MΩ, See Note 6	25°C	50	520		50	870		V/mV
		-55°C	25	1000		25	1775		
		125°C	25	200		25	380		
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	65	94		65	97		dB
		-55°C	60	95		60	97		
		125°C	60	85		60	91		
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	70	97		70	97		dB
		-55°C	60	97		60	97		
		125°C	60	98		60	98		
I <sub>I(SEL)</sub> Input current (BIAS SELECT)	V <sub>I(SEL)</sub> = V <sub>DD</sub>	25°C	65			95			nA
I <sub>DD</sub> Supply current	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	10 17			14 23			μA
		-55°C	17 30			28 48			
		125°C	7 12			9 15			

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.





**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.03		V/ $\mu$ s
			0°C	0.04		
			70°C	0.03		
		$V_I(PP) = 2.5\text{ V}$	25°C	0.03		
			0°C	0.03		
			70°C	0.02		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	5		kHz	
		0°C	6			
		70°C	4.5			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	85		kHz	
		0°C	100			
		70°C	65			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	34°			
		0°C	36°			
		70°C	30°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_I(PP) = 1\text{ V}$	25°C	0.05		V/ $\mu$ s
			0°C	0.05		
			70°C	0.04		
		$V_I(PP) = 5.5\text{ V}$	25°C	0.04		
			0°C	0.05		
			70°C	0.04		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	1		kHz	
		0°C	1.3			
		70°C	0.9			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	110		kHz	
		0°C	125			
		70°C	90			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	38°			
		0°C	40°			
		70°C	34°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.03		V/ $\mu\text{s}$
			-40°C	0.04		
			85°C	0.03		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.03		
			-40°C	0.04		
			85°C	0.02		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	5		kHz	
		-40°C	7			
		85°C	4			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	85		MHz	
			-40°C	130		
			85°C	55		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	34°			
		-40°C	38°			
		85°C	28°			

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.05		V/ $\mu\text{s}$
			-40°C	0.06		
			85°C	0.03		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.04		
			-40°C	0.05		
			85°C	0.03		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 99	$R_S = 20\ \Omega$ , 25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 98	25°C	1		kHz	
		-40°C	1.4			
		85°C	0.8			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 100	$C_L = 20\text{ pF}$ , 25°C	110		MHz	
			-40°C	155		
			85°C	80		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 100	25°C	38°			
		-40°C	42°			
		85°C	32°			



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**LOW-BIAS MODE**

**operating characteristics at specified free-air temperature,  $V_{DD} = 5 V$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC271M			UNIT
			MIN	TYP	MAX	
SR    Slew rate at unity gain	R <sub>L</sub> = 1 MΩ, C <sub>L</sub> = 20 pF, See Figure 98	V <sub>I(PP)</sub> = 1 V	25°C	0.03		V/μs
			-55°C	0.04		
			125°C	0.02		
		V <sub>I(PP)</sub> = 2.5 V	25°C	0.03		
			-55°C	0.04		
			125°C	0.02		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, See Figure 99	R <sub>S</sub> = 20 Ω, 25°C	68			nV/√Hz
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>OH</sub> , R <sub>L</sub> = 1 MΩ,	C <sub>L</sub> = 20 pF, See Figure 98	25°C	5		kHz
			-55°C	8		
			125°C	3		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, See Figure 100	C <sub>L</sub> = 20 pF,	25°C	85		kHz
			-55°C	140		
			125°C	45		
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF,	f = B <sub>1</sub> , See Figure 100	25°C	34°		
			-55°C	39°		
			125°C	25°		

**operating characteristics at specified free-air temperature,  $V_{DD} = 10 V$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC271M			UNIT
			MIN	TYP	MAX	
SR    Slew rate at unity gain	R <sub>L</sub> = 1 MΩ, C <sub>L</sub> = 20 pF, See Figure 98	V <sub>I(PP)</sub> = 1 V	25°C	0.05		V/μs
			-55°C	0.06		
			125°C	0.03		
		V <sub>I(PP)</sub> = 5.5 V	25°C	0.04		
			-55°C	0.06		
			125°C	0.03		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, See Figure 99	R <sub>S</sub> = 20 Ω, 25°C	68			nV/√Hz
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>OH</sub> , R <sub>L</sub> = 1 MΩ,	C <sub>L</sub> = 20 pF, See Figure 98	25°C	1		kHz
			-55°C	1.5		
			125°C	0.7		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, See Figure 100	C <sub>L</sub> = 20 pF,	25°C	110		kHz
			-55°C	165		
			125°C	70		
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF,	f = B <sub>1</sub> , See Figure 100	25°C	38°		
			-55°C	43°		
			125°C	29°		



**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (LOW-BIAS MODE)**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	66, 67
$\alpha_{VIO}$	Temperature coefficient	Distribution	68, 69
$V_{OH}$	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	70, 71 72 73
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	74, 75 76 77 78, 79
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	80 81 92, 93
$I_{IB}$	Input bias current	vs Free-air temperature	82
$I_{IO}$	Input offset current	vs Free-air temperature	82
$V_I$	Maximum input voltage	vs Supply voltage	83
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	84 85
$SR$	Slew rate	vs Supply voltage vs Free-air temperature	86 87
	Bias-select current	vs Supply voltage	88
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	89
$B_1$	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	90 91
$\phi_m$	Phase margin	vs Supply voltage vs Free-air temperature vs Load capacitance	94 95 96
$V_n$	Equivalent input noise voltage	vs Frequency	97
	Phase shift	vs Frequency	92, 93



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE

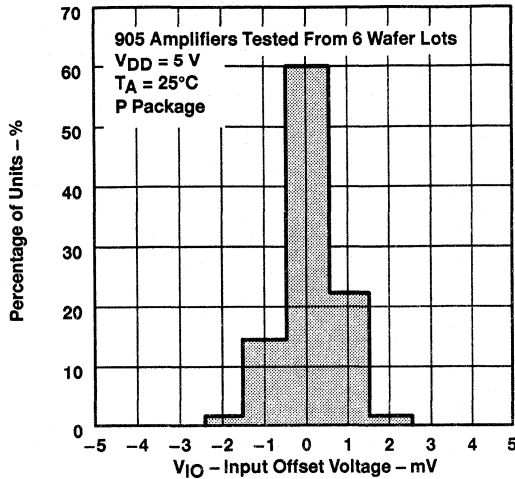


Figure 66

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE

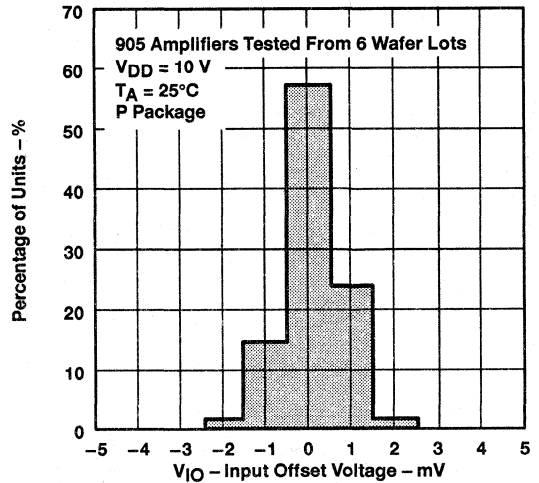


Figure 67

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

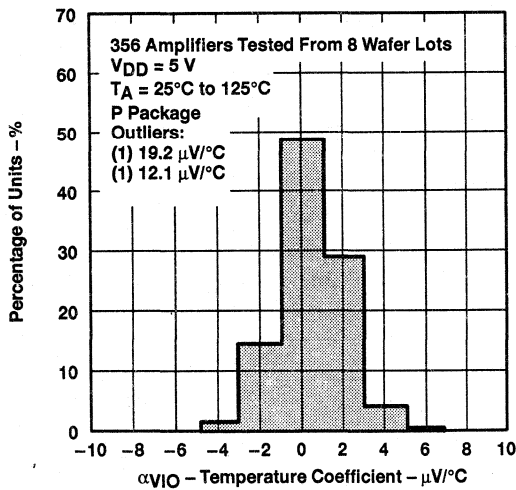


Figure 68

DISTRIBUTION OF TLC271  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

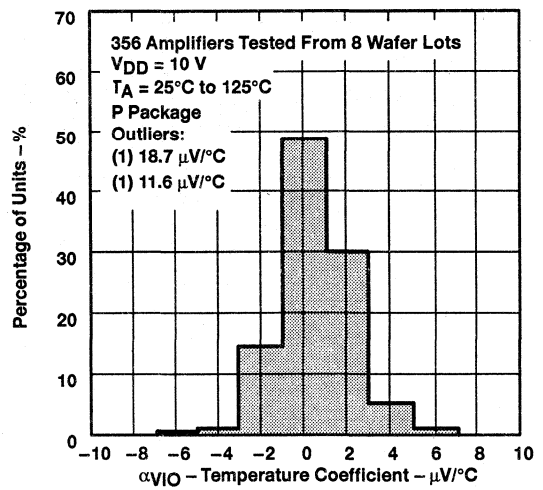


Figure 69

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

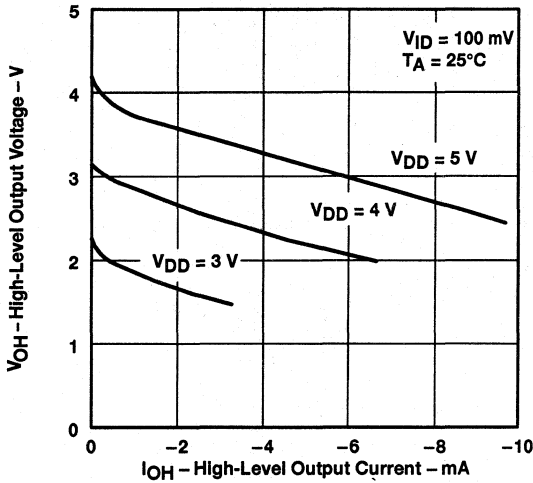


Figure 70

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

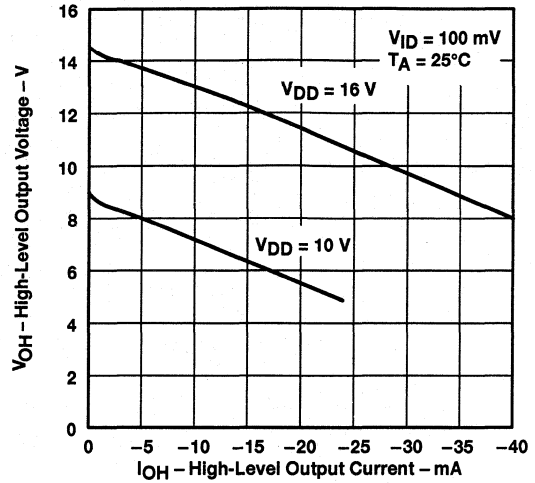


Figure 71

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE

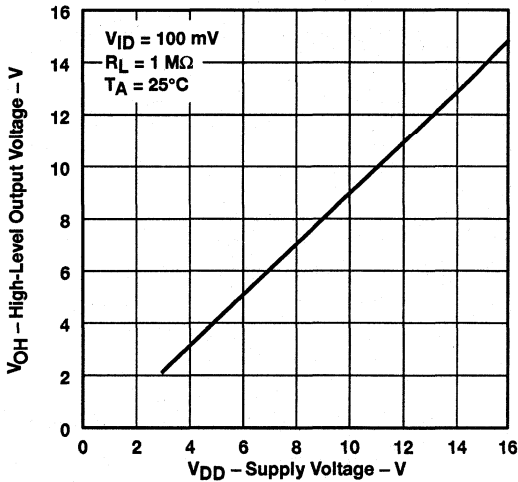


Figure 72

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

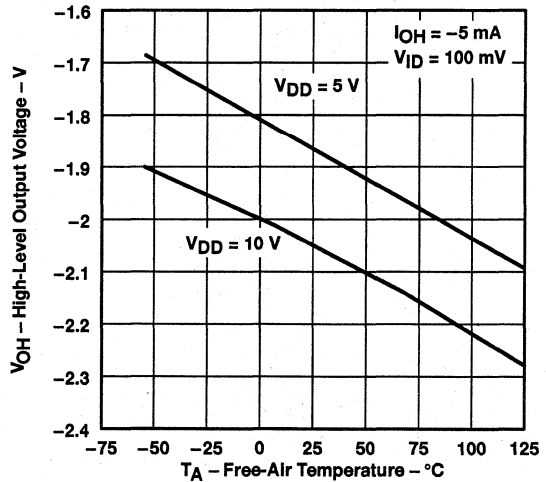


Figure 73

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

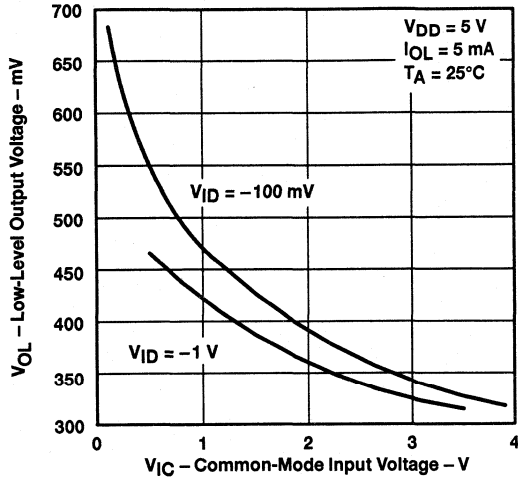


Figure 74

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

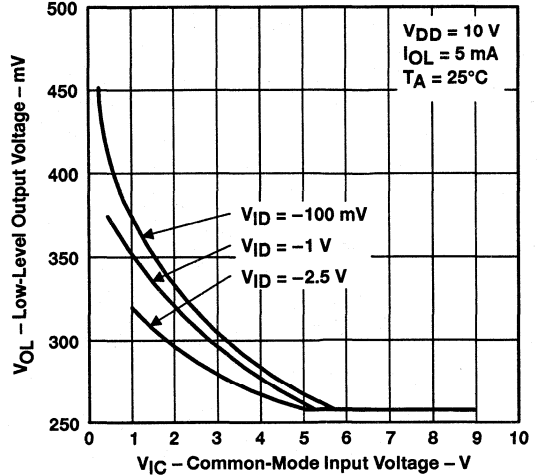


Figure 75

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE

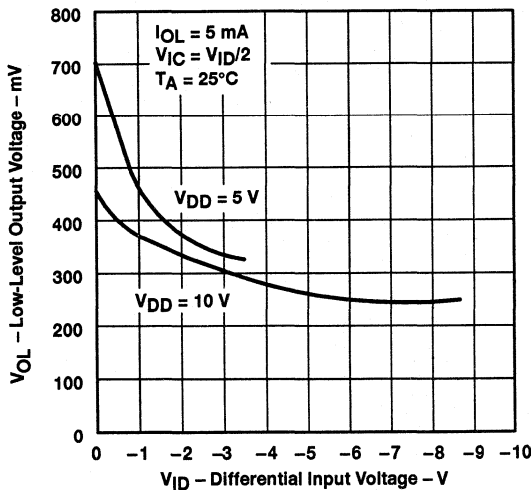


Figure 76

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

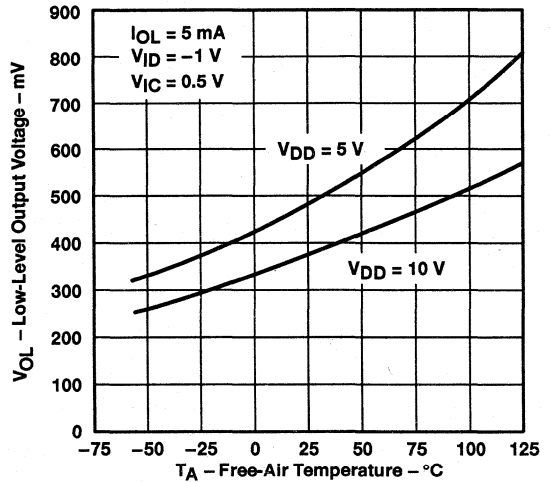


Figure 77

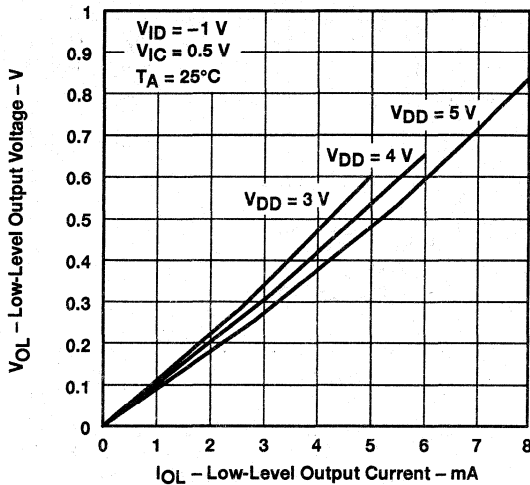
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

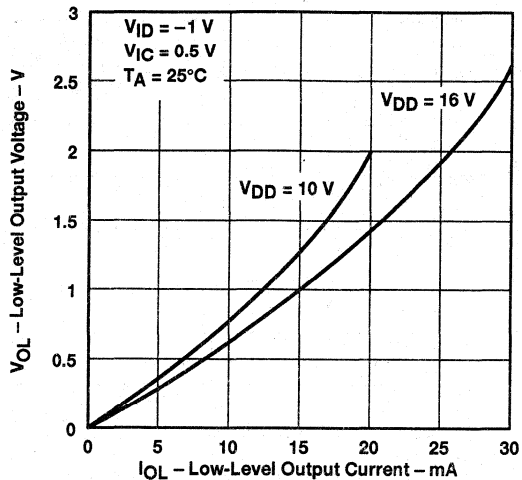
**TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†**

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**LOW-LEVEL OUTPUT CURRENT**



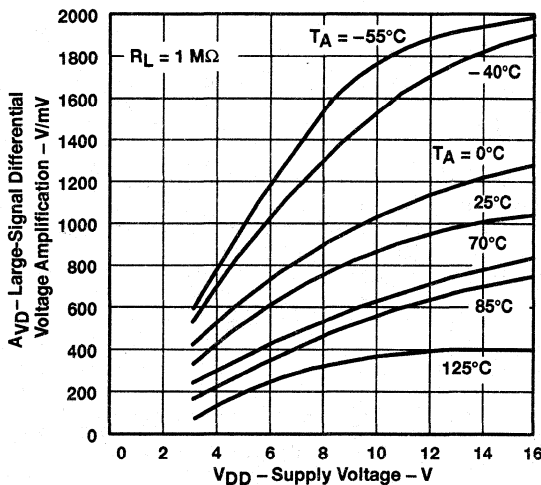
**Figure 78**

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**LOW-LEVEL OUTPUT CURRENT**



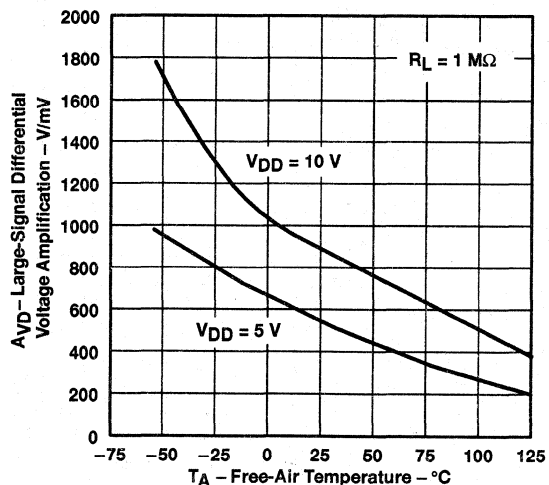
**Figure 79**

**LARGE-SIGNAL**  
**DIFFERENTIAL VOLTAGE AMPLIFICATION**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 80**

**LARGE-SIGNAL**  
**DIFFERENTIAL VOLTAGE AMPLIFICATION**  
**vs**  
**FREE-AIR TEMPERATURE**

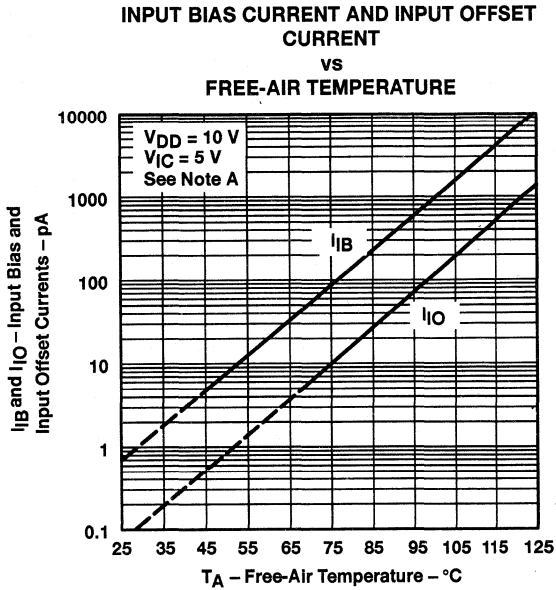


**Figure 81**

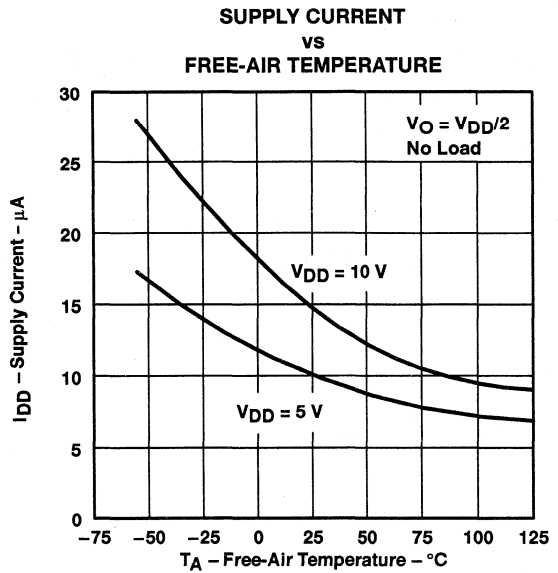
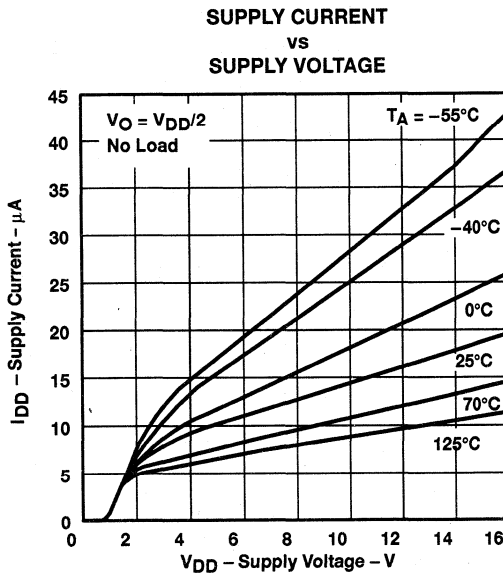
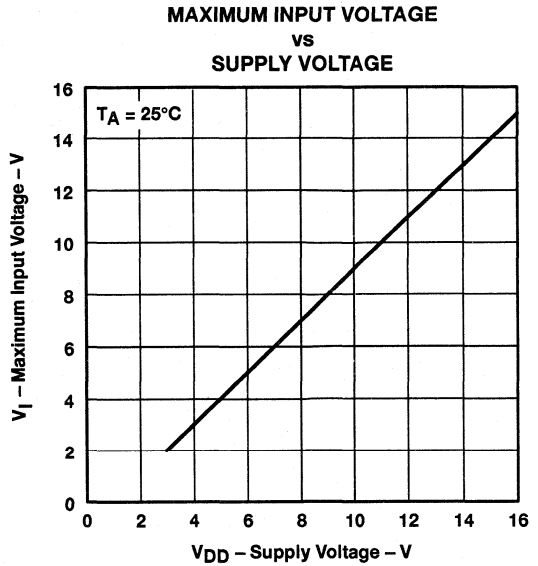
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC271, TLC271A, TLC271B LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

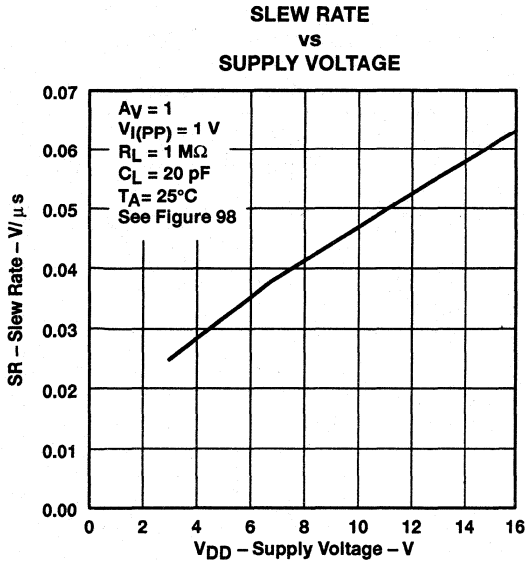


Figure 86

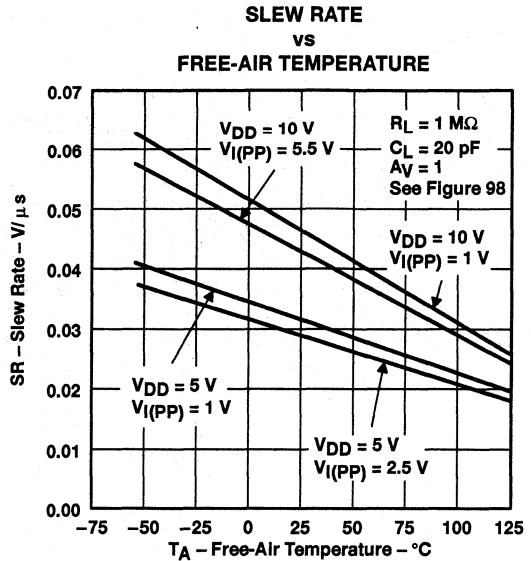


Figure 87

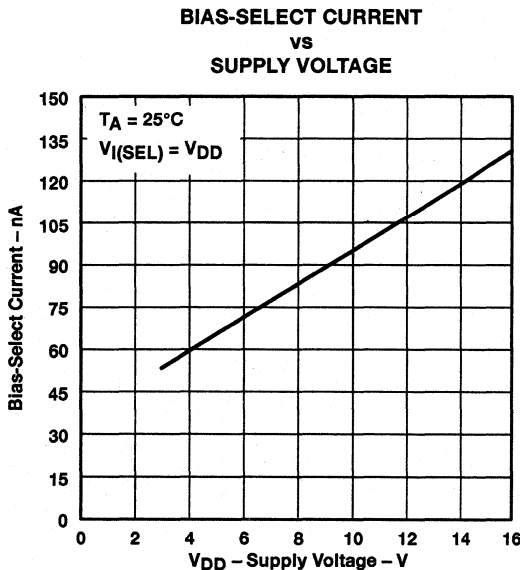


Figure 88

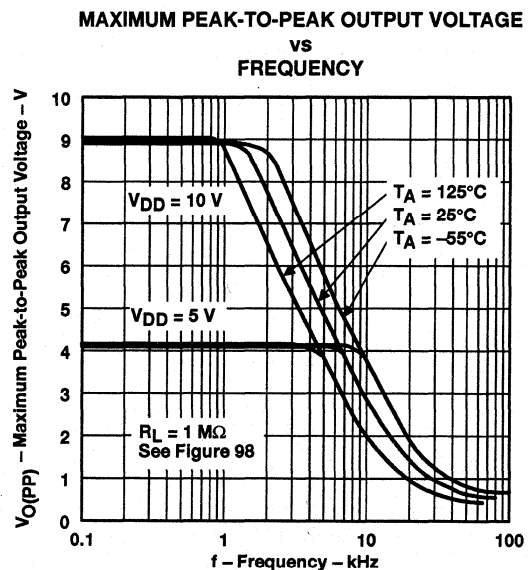


Figure 89

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

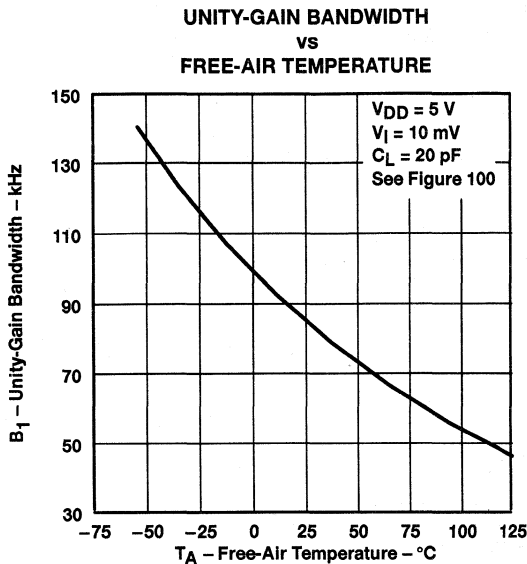


Figure 90

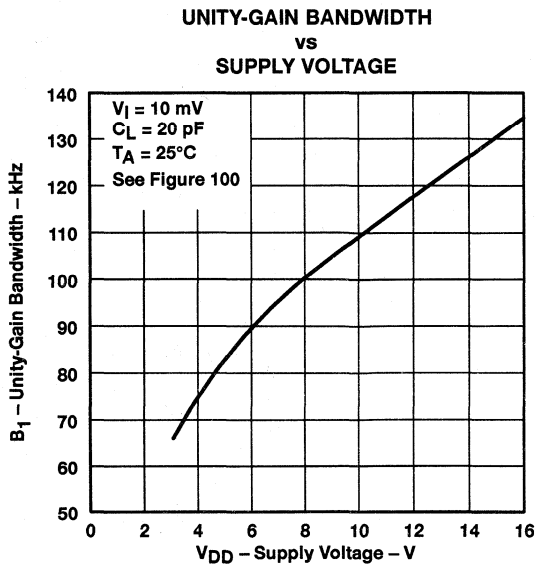


Figure 91

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY**

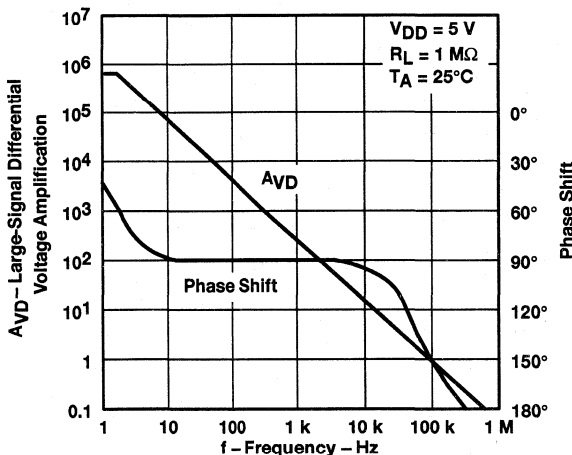


Figure 92

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY**

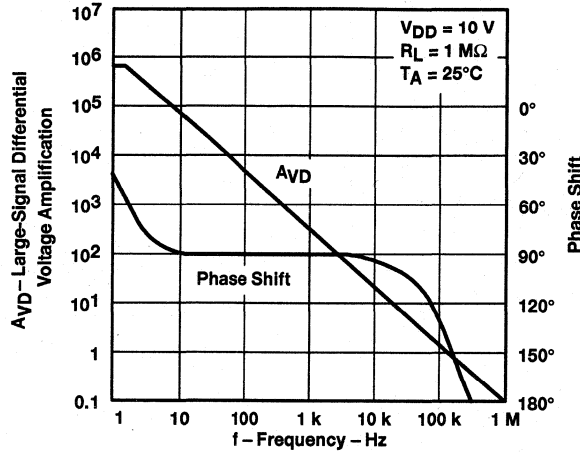


Figure 93

**PHASE MARGIN VS SUPPLY VOLTAGE**

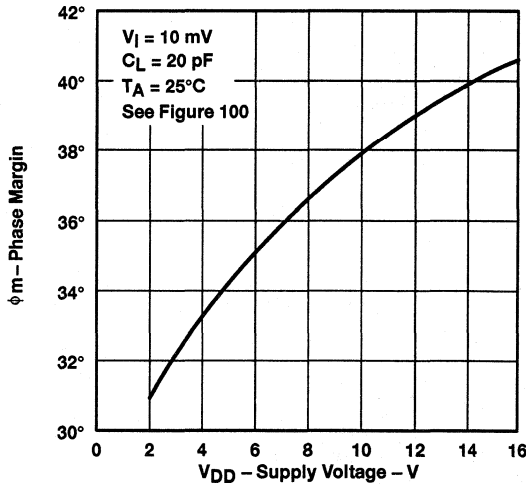


Figure 94

**PHASE MARGIN VS FREE-AIR TEMPERATURE**

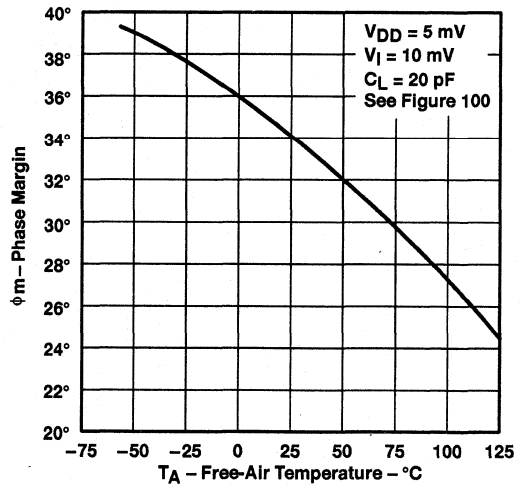


Figure 95

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS (LOW-BIAS MODE)†

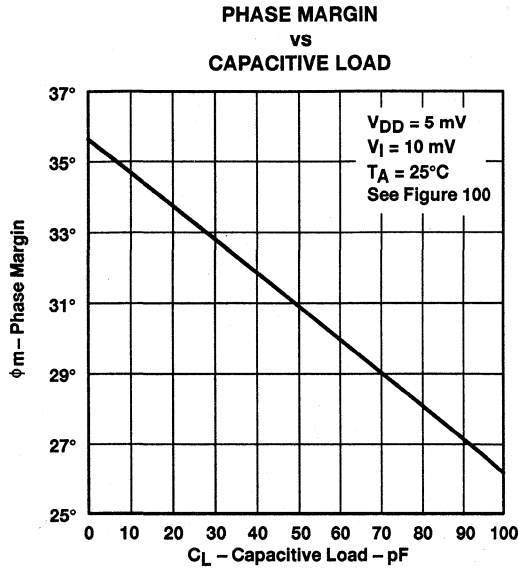


Figure 96

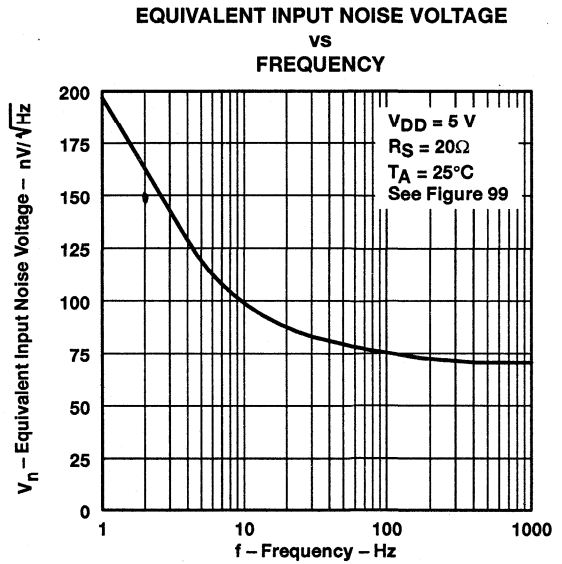


Figure 97

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC271 is optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

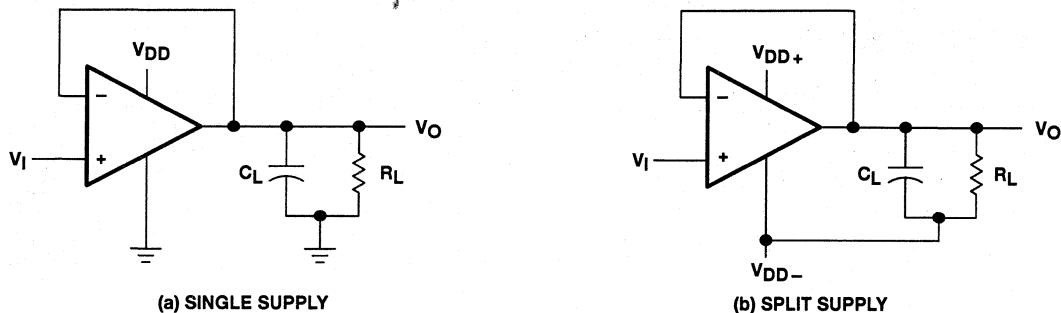


Figure 98. Unity-Gain Amplifier

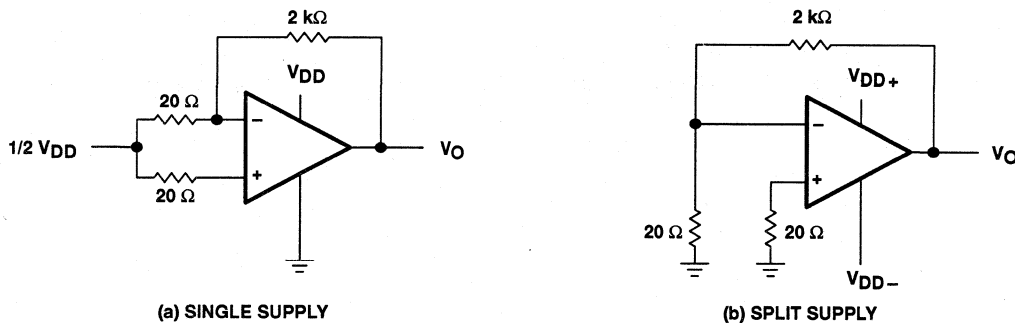


Figure 99. Noise-Test Circuit

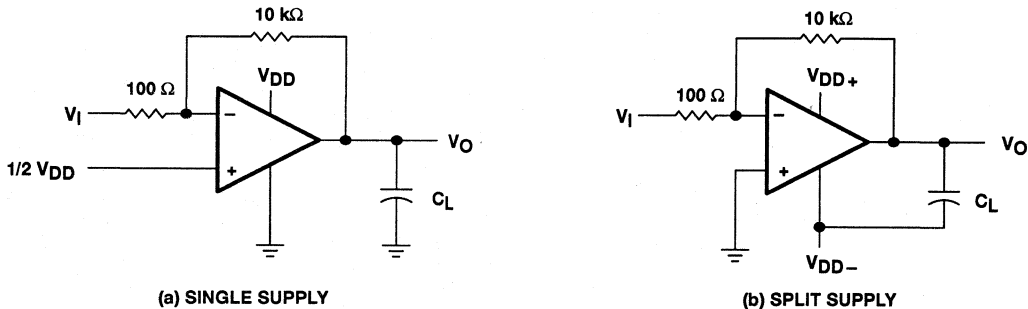


Figure 100. Gain-of-100 Inverting Amplifier

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC271 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 101). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

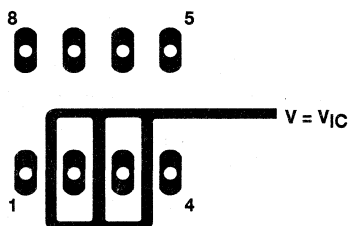


Figure 101. Isolation Metal Around Device inputs (JG and P packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise is necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to the Typical Characteristics section of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

### full-power response

Full-power response, the frequency above which the amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

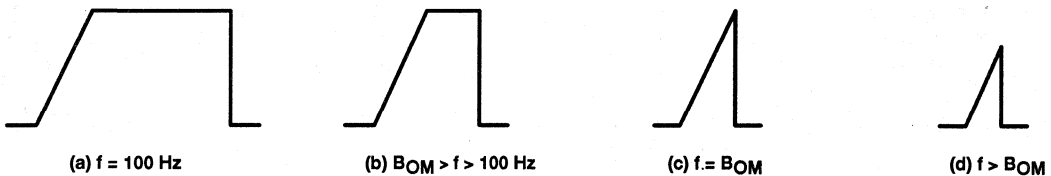
SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**PARAMETER MEASUREMENT INFORMATION**

**full-power response (continued)**

by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 98. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 102). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.



**Figure 102. Full-Power-Response Output Signal**

**test time**

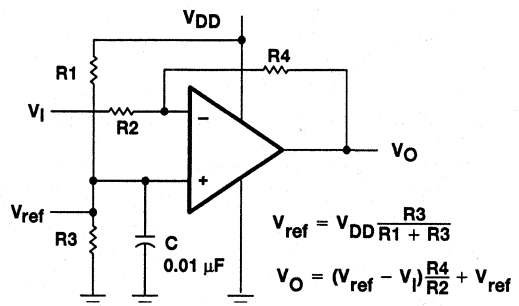
Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices, and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

**APPLICATION INFORMATION**

**single-supply operation**

While the TLC271 performs well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 103). The low input bias current consumption of the TLC271 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.



**Figure 103. Inverting Amplifier With Voltage Reference**

$$V_{ref} = V_{DD} \frac{R3}{R1 + R3}$$

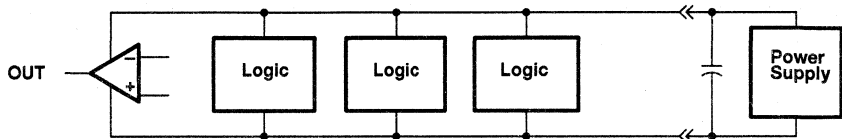
$$V_O = (V_{ref} - V_i) \frac{R4}{R2} + V_{ref}$$



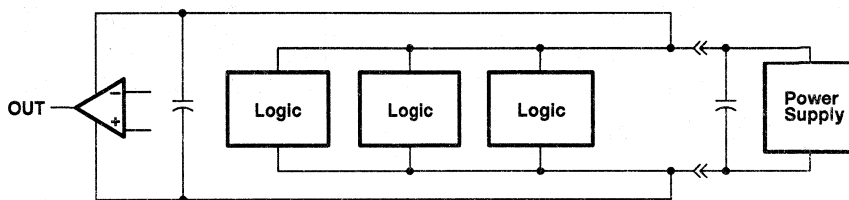
**single-supply operation (continued)**

The TLC271 works well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 104); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, RC decoupling may be necessary in high-frequency applications.



(a) COMMON SUPPLY RAILS

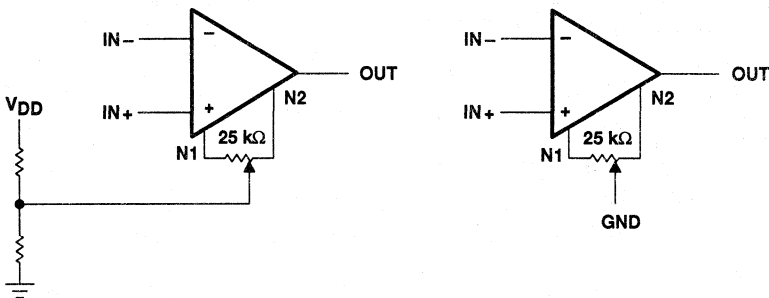


(b) SEPARATE BYPASSED SUPPLY RAILS (preferred)

**Figure 104. Common Versus Separate Supply Rails**

**input offset voltage nulling**

The TLC271 offers external input offset null control. Nulling of the input offset voltage may be achieved by adjusting a 25-k $\Omega$  potentiometer connected between the offset null terminals with the wiper connected as shown in Figure 105. The amount of nulling range varies with the bias selection. In the high-bias mode, the nulling range allows the maximum offset voltage specified to be trimmed to zero. In low-bias and medium-bias modes, total nulling may not be possible.



(a) SINGLE SUPPLY

(b) SPLIT SUPPLY

**Figure 105. Input Offset Voltage Null Circuit**

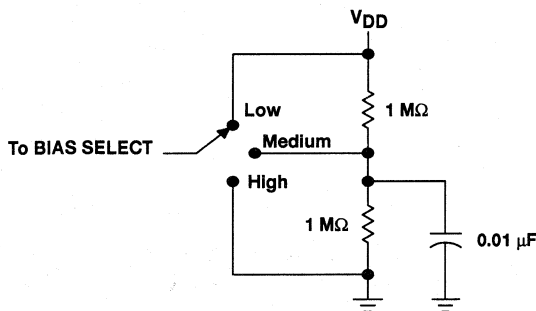
# TLC271, TLC271A, TLC271B LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### bias selection

Bias selection is achieved by connecting the bias select pin to one of the three voltage levels (see Figure 106). For medium-bias applications, it is recommended that the bias select pin be connected to the mid-point between the supply rails. This is a simple procedure in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode necessitates using a voltage divider as indicated. The use of large-value resistors in the voltage divider reduces the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor requires significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the midpoint may be used if it is within the voltages specified in the table of Figure 106.



BIAS MODE	BIAS-SELECT VOLTAGE (single supply)
Low	$V_{DD}$
Medium	1 V to $V_{DD} - 1$ V
High	GND

Figure 106. Bias Selection for Single-Supply Applications

### input characteristics

The TLC271 is specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5$  V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC271 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1 \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC271 is well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 101 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 107).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

### noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC271 results in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50 \text{ k}\Omega$ , since bipolar devices exhibit greater noise currents.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

APPLICATION INFORMATION

noise performance (continued)

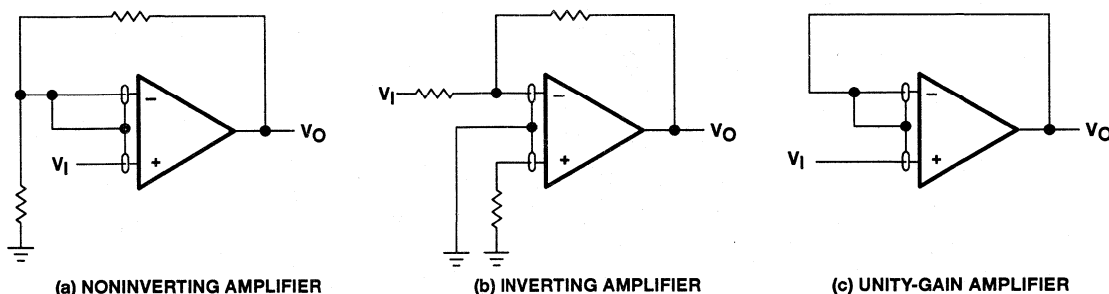


Figure 107. Guard-Ring Schemes

feedback

Operational amplifier circuits almost always employ feedback, and since feedback is the first prerequisite for oscillation, a little caution is appropriate. Most oscillation problems result from driving capacitive loads and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 108). The value of this capacitor is optimized empirically.

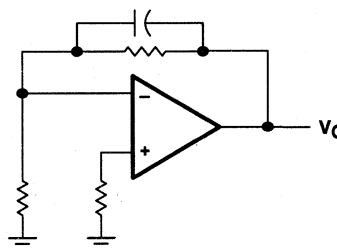


Figure 108. Compensation for Input Capacitance

electrostatic discharge protection

The TLC271 incorporates an internal electrostatic-discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC271 inputs and output were designed to withstand  $-100\text{-mA}$  surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not by design be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ( $0.1\ \mu\text{F}$  typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

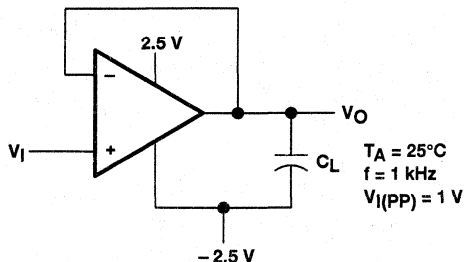
SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

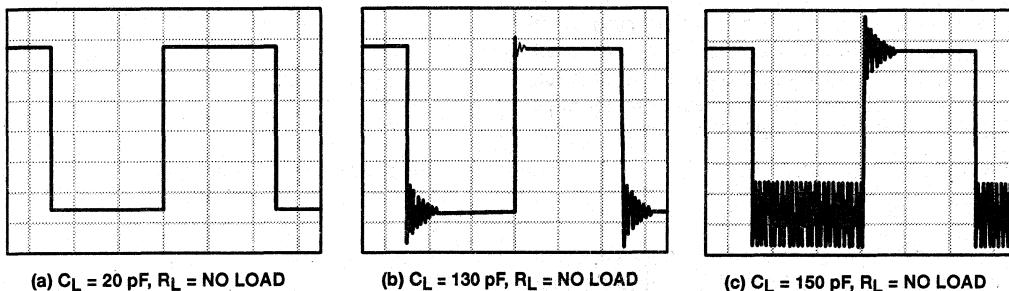
**output characteristics**

The output stage of the TLC271 is designed to sink and source relatively high amounts of current (see Typical Characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

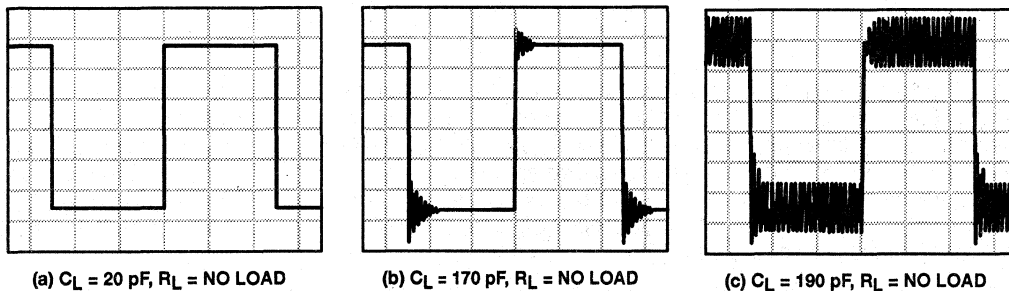
All operating characteristics of the TLC271 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figures 110, 111, and 112). In many cases, adding some compensation in the form of a series resistor in the feedback loop alleviates the problem.



**Figure 109. Test Circuit for Output Characteristics**



**Figure 110. Effect of Capacitive Loads in High-Bias Mode**



**Figure 111. Effect of Capacitive Loads in Medium-Bias Mode**

APPLICATION INFORMATION

output characteristics (continued)

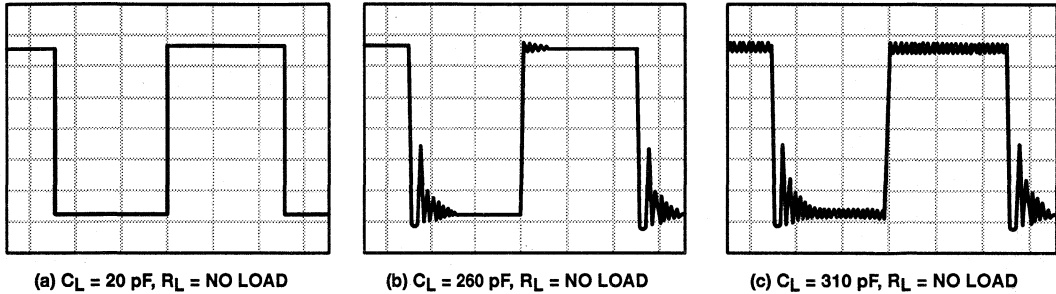


Figure 112. Effect of Capacitive Loads in Low-Bias Mode

Although the TLC271 possesses excellent high-level output voltage and current capability, methods are available for boosting this capability, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 113). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor, N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60 \Omega$  and  $180 \Omega$ , depending on how hard the operational amplifier input is driven. With very low values of  $R_P$ , a voltage offset from  $0 \text{ V}$  at the output occurs. Secondly, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

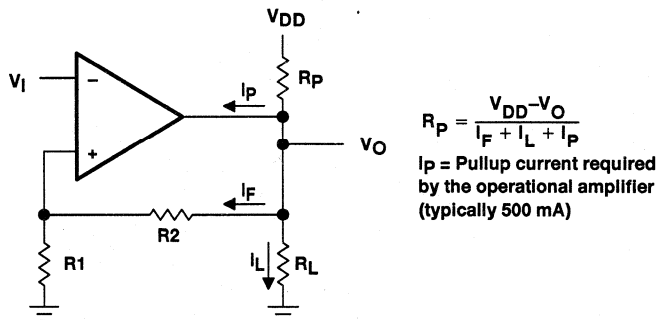


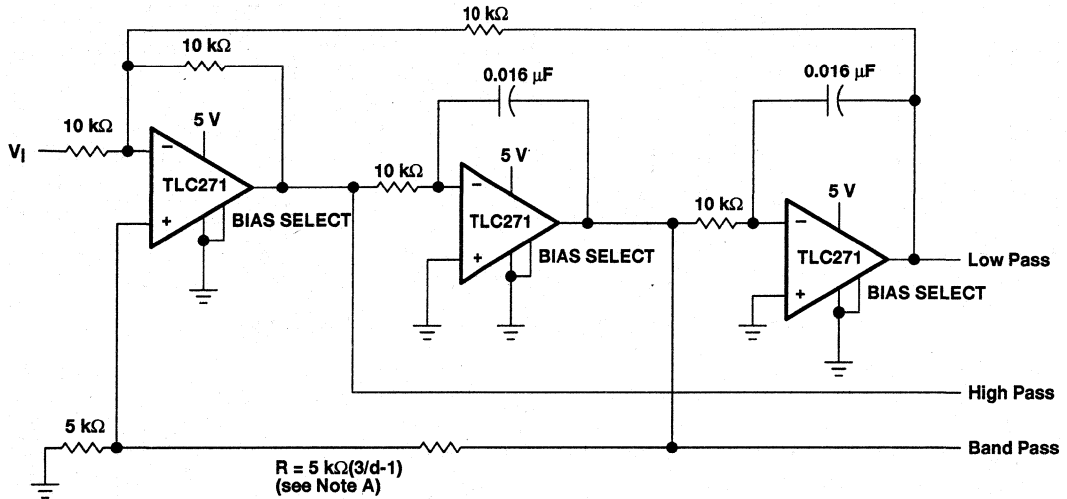
Figure 113. Resistive Pullup to Increase  $V_{OH}$

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

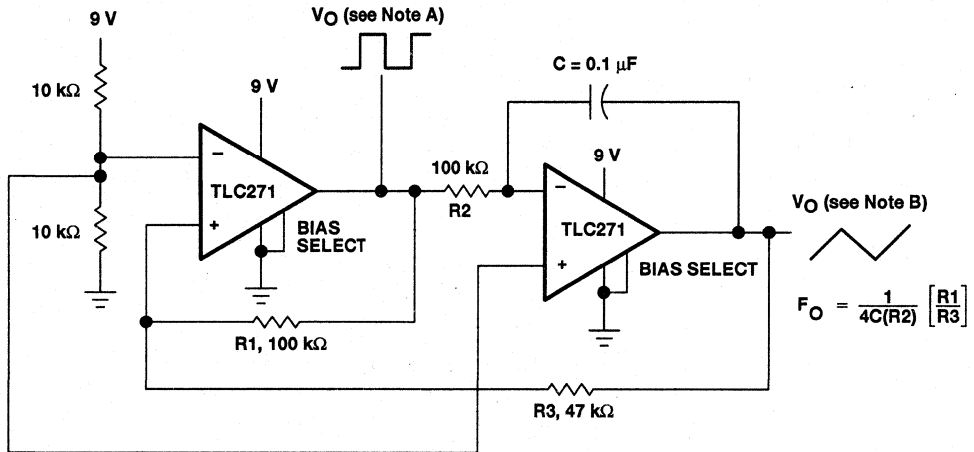
**APPLICATION INFORMATION**

**output characteristics (continued)**



NOTE A: d = damping factor, I/O

**Figure 114. State-Variable Filter**



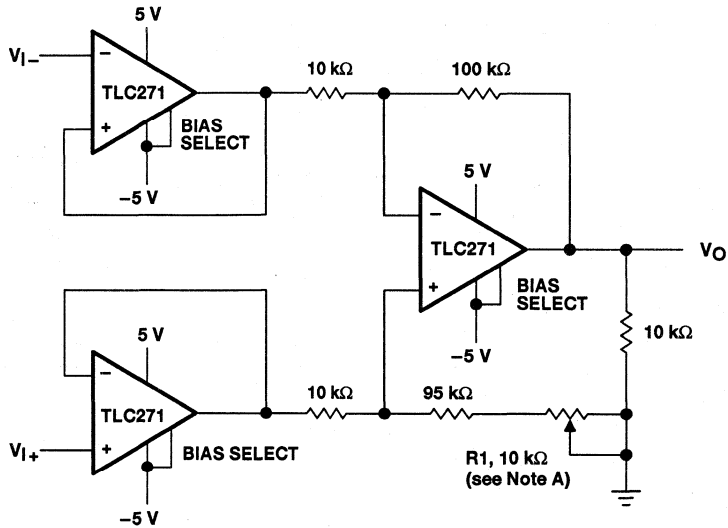
NOTES: A.  $V_{O(PP)} = 8\text{ V}$   
 B.  $V_{O(PP)} = 4\text{ V}$

**Figure 115. Single-Supply Function Generator**



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

APPLICATION INFORMATION (HIGH-BIAS MODE)



NOTE A: CMRR adjustment must be noninductive.

Figure 116. Low-Power Instrumentation Amplifier

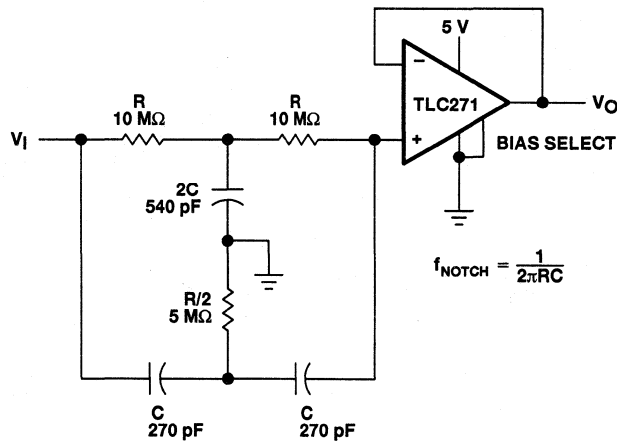
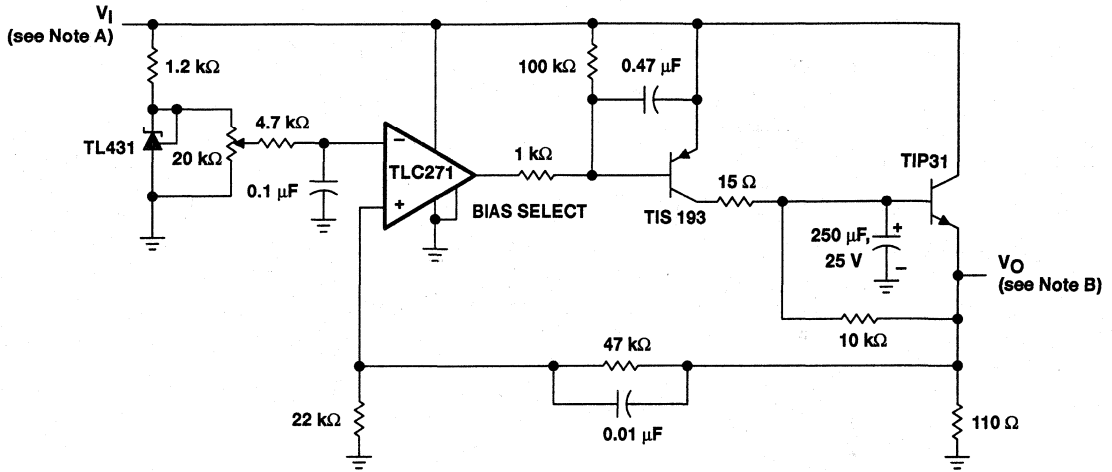


Figure 117. Single-Supply Twin-T Notch Filter

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

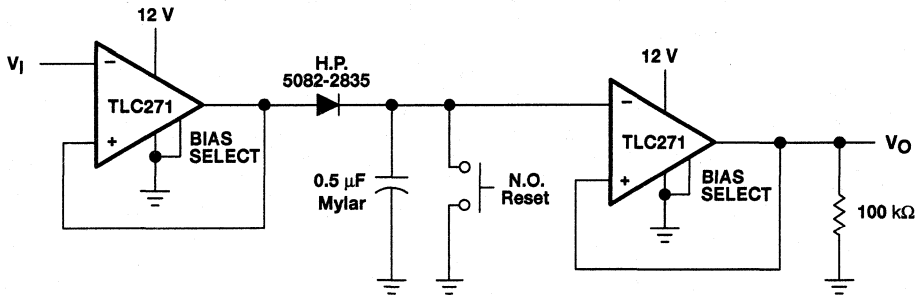
SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION (HIGH-BIAS MODE)**



NOTES: C.  $V_I = 3.5$  to  $15$  V  
 D.  $V_O = 2.0$  V,  $0$  to  $1$  A

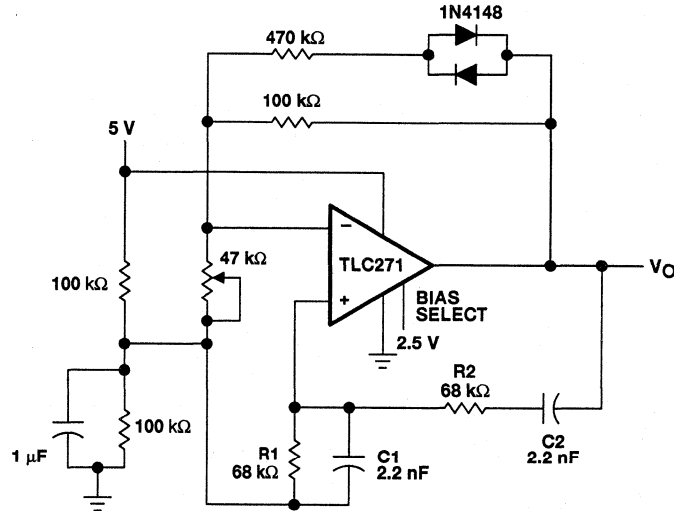
**Figure 118. Logic-Array Power Supply**



**Figure 119. Positive-Peak Detector**



APPLICATION INFORMATION (MEDIUM-BIAS MODE)



NOTES:  $V_{O(PP)} = 2V$

$$f_o = \frac{1}{2\pi\sqrt{R_1 R_2 C_1 C_2}}$$

Figure 120. Wein Oscillator

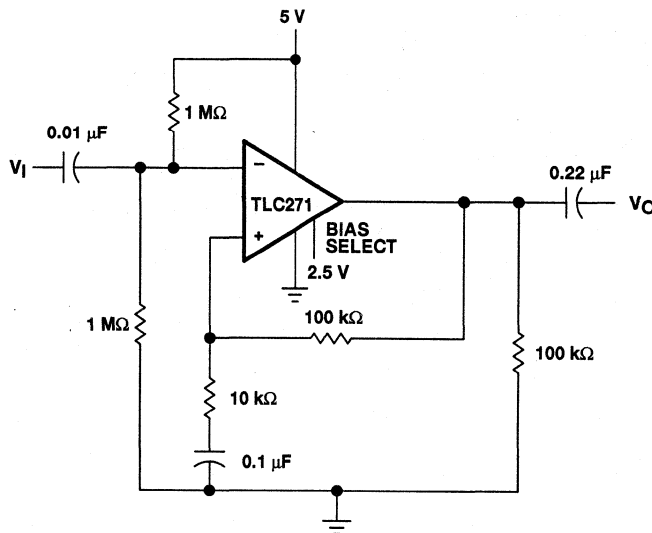
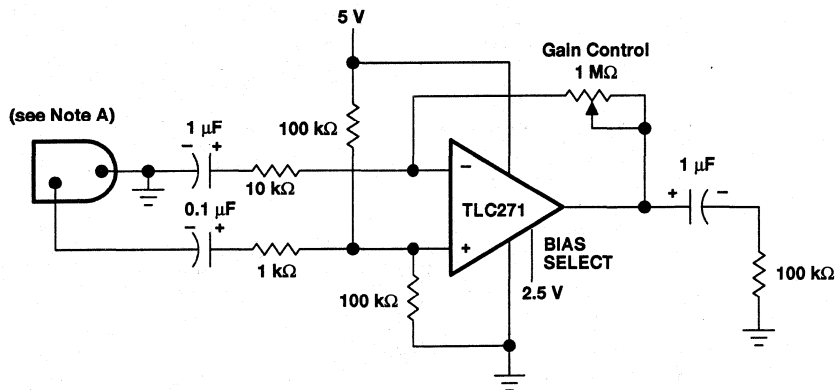


Figure 121. Single-Supply AC Amplifier

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

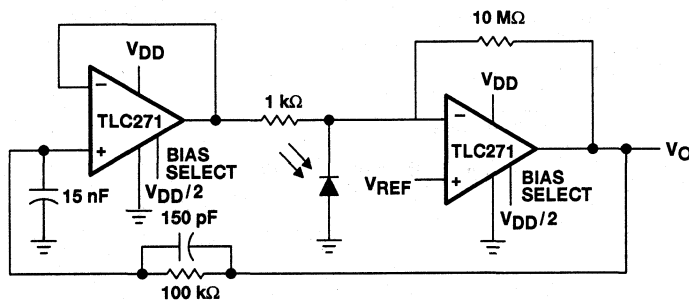
SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION (MEDIUM-BIAS MODE)**



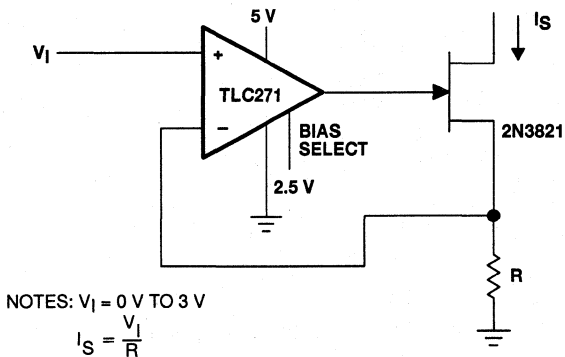
NOTE A: Low to medium impedance dynamic mike

**Figure 122. Microphone Preamp**



NOTES:  $V_{DD} = 4 \text{ V to } 15 \text{ V}$   
 $V_{ref} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

**Figure 123. Photo-Diode Amplifier With Ambient Light Rejection**



NOTES:  $V_I = 0 \text{ V to } 3 \text{ V}$   
 $I_S = \frac{V_I}{R}$

**Figure 124. Precision Low-Current Sink**

APPLICATION INFORMATION (LOW-BIAS MODE)

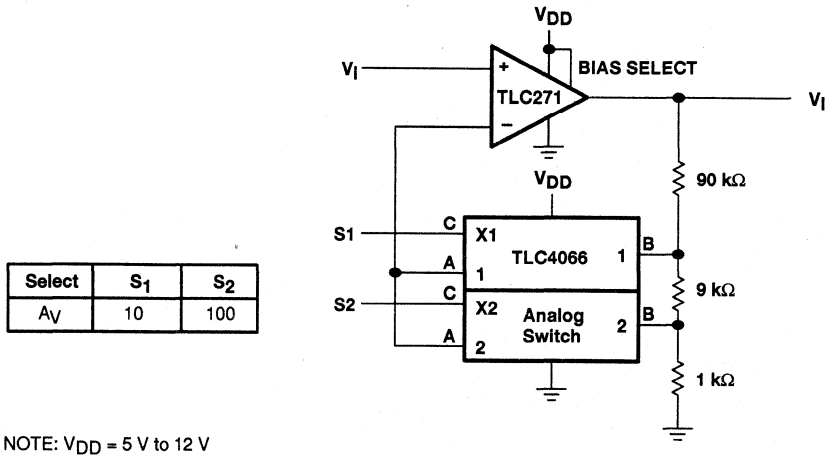


Figure 125. Amplifier With Digital Gain Selection

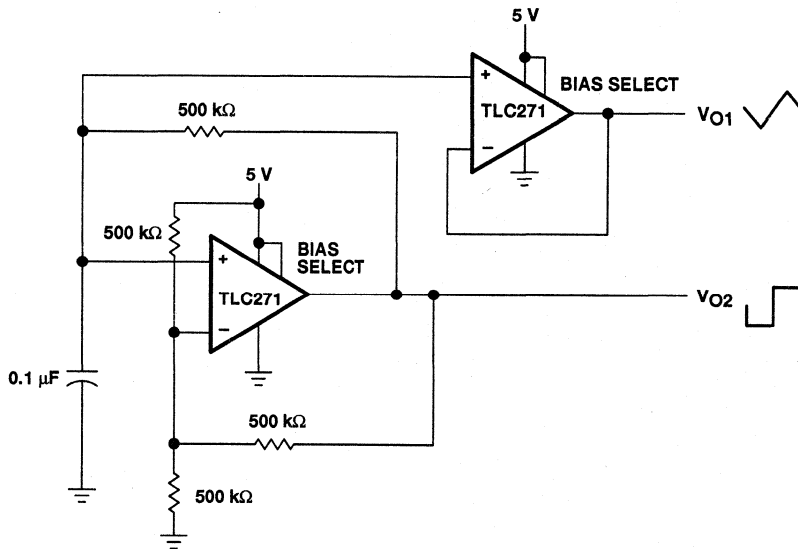
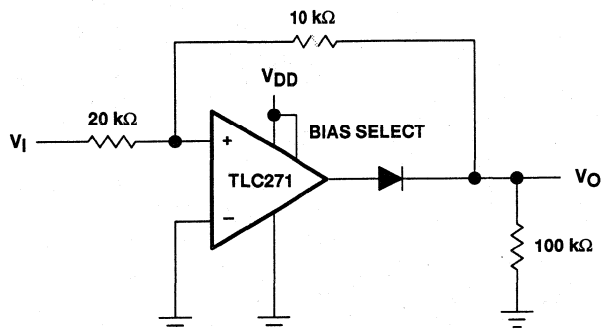


Figure 126. Multivibrator

**TLC271, TLC271A, TLC271B**  
**LinCMOS™ PROGRAMMABLE LOW-POWER**  
**OPERATIONAL AMPLIFIERS**

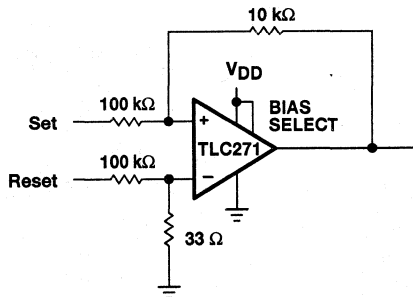
SLOS090A – NOVEMBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION (LOW-BIAS MODE)**



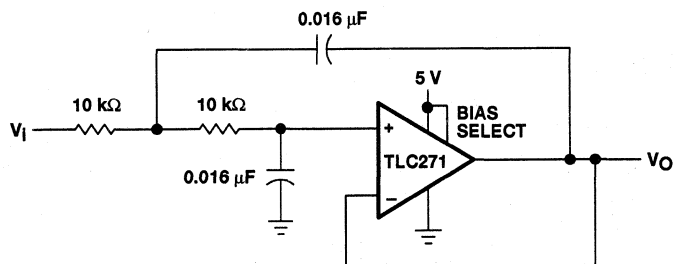
NOTE:  $V_{DD} = 5\text{ V to }16\text{ V}$

**Figure 127. Full-Wave Rectifier**



NOTE:  $V_{DD} = 5\text{ V to }16\text{ V}$

**Figure 128. Set/Reset Flip-Flop**



NOTE: Normalized to  $F_C = 1\text{ kHz}$  and  $R_L = 10\text{ k}\Omega$

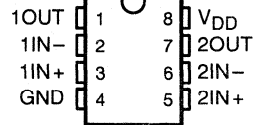
**Figure 129. Two-Pole Low-Pass Butterworth Filter**

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

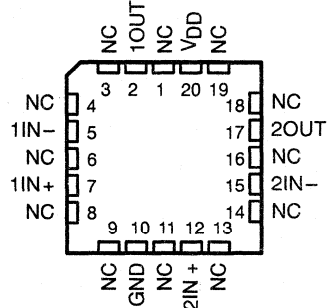
SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC277 . . . 500  $\mu\text{V}$  Max at 25°C,  
 $V_{\text{DD}} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
0°C to 70°C . . . 3 V to 16 V  
–40°C to 85°C . . . 4 V to 16 V  
–55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)**
- **Low Noise . . . Typically 25 nV/ $\sqrt{\text{Hz}}$  at**  
 $f = 1\text{ kHz}$
- **Output Voltage Range Includes Negative Rail**
- **High Input impedance . . . 10<sup>12</sup>  $\Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-in Latch-Up Immunity**

D, JG, P, OR PW PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

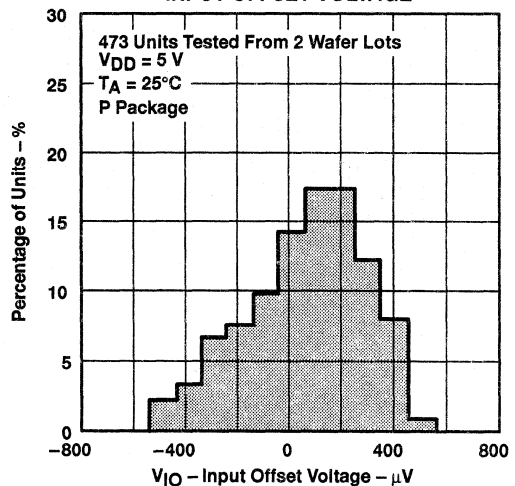
## description

The TLC272 and TLC277 precision dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose BiFET devices.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for BiFET and NFET products. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC272 (10 mV) to the high-precision TLC277 (500  $\mu\text{V}$ ). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

DISTRIBUTION OF TLC277  
INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2–537

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	TSSOP (PW)	
0°C to 70°C	500 μV	TLC277CD	—	—	TLC277CP	—	—
	2 mV	TLC272BCD	—	—	TLC272BCP	—	—
	5 mV	TLC272ACD	—	—	TLC272ACP	—	—
	10 mV	TLC272CD	—	—	TLC272CP	TLC272CPW	TLC272Y
-40°C to 85°C	500 μV	TLC277ID	—	—	TLC277IP	—	—
	2 mV	TLC272BID	—	—	TLC272BIP	—	—
	5 mV	TLC272AID	—	—	TLC272AIP	—	—
	10 mV	TLC272ID	—	—	TLC272IP	—	—
-55°C to 125°C	500 μV	TLC277MD	TLC277MFK	TLC277MJG	TLC277MP	—	—
	10 mV	TLC272MD	TLC272MFK	TLC272MJG	TLC272MP	—	—

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC277CDR).

### description (continued)

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC272 and TLC277. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand -100-mA surge currents without sustaining latch-up.

The TLC272 and TLC277 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

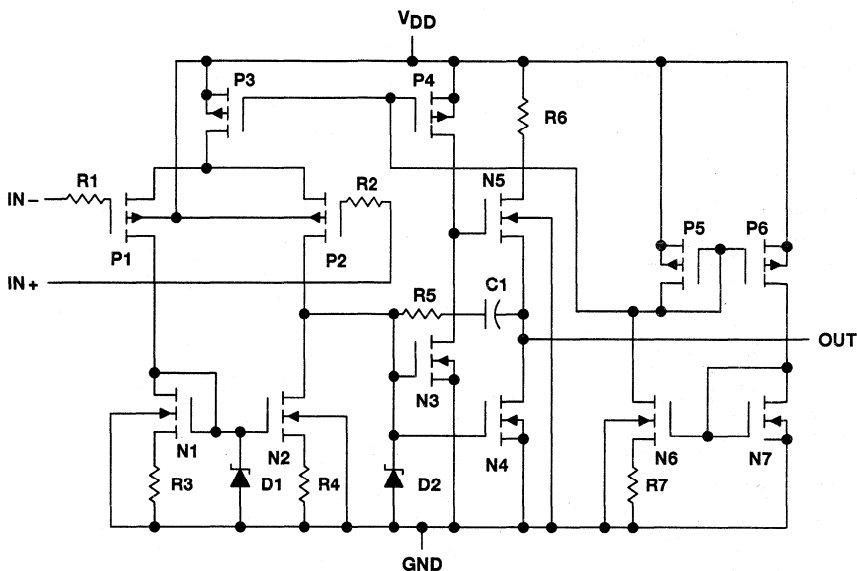
The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

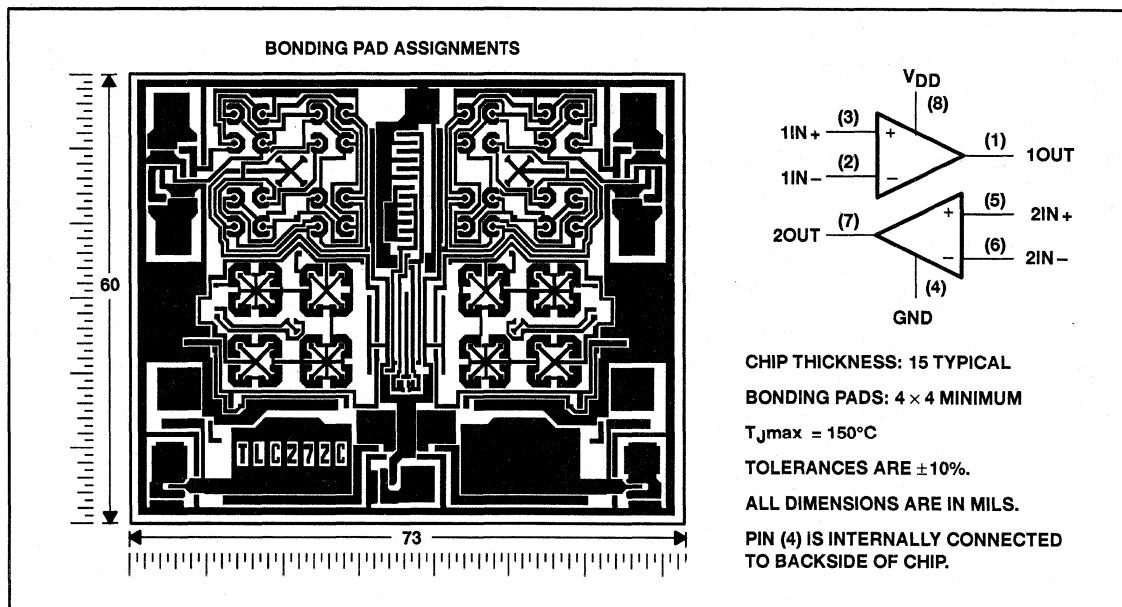
SLOS091B - OCTOBER 1987 - REVISED AUGUST 1994

equivalent schematic (each amplifier)



## TLC272Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC272C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, P, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to network ground.
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
  3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	N/A
FK	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	N/A
PW	525 mW	4.2 mW/°C	336 mW	N/A	N/A

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C





# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC272C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC272AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC272BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	230	2000	$\mu\text{V}$
					Full range		3000	
		TLC277C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	200	500	$\mu\text{V}$
					Full range		1500	
$\alpha_{VIO}$	Temperature coefficient of input offset voltage			25°C to 70°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8	V	
				0°C	3	3.8		
				70°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C		0 50	mV	
				0°C		0 50		
				70°C		0 50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ ,	$R_L = 10\text{ k}\Omega$	25°C	5	23	V/mV	
				0°C	4	27		
				70°C	4	20		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				0°C	60	84		
				70°C	60	85		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	1.4	3.2	mA	
				0°C	1.6	3.6		
				70°C	1.2	2.6		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T <sub>A</sub> †	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
					MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	TLC272C	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	1.1 10		mV
					Full range	12		
		TLC272AC	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	0.9 5		
					Full range	6.5		
	TLC272BC	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	290	2000	μV	
				Full range	3000			
		TLC277C	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	250		800
					Full range	1900		
αV <sub>IO</sub>	Temperature coefficient of input offset voltage			25°C to 70°C	2		μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = 5 V,	V <sub>IC</sub> = 5 V	25°C	0.1		pA	
				70°C	7	300		
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = 5 V,	V <sub>IC</sub> = 5 V	25°C	0.7		pA	
				70°C	50	600		
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV,	R <sub>L</sub> = 10 kΩ	25°C	8	8.5	V	
				0°C	7.8	8.5		
				70°C	7.8	8.4		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV,	I <sub>OL</sub> = 0	25°C	0 50		mV	
				0°C	0 50			
				70°C	0 50			
A <sub>VD</sub>	Large-signal differential voltage amplification	V <sub>O</sub> = 1 V to 6 V,	R <sub>L</sub> = 10 kΩ	25°C	10	36	V/mV	
				0°C	7.5	42		
				70°C	7.5	32		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>		25°C	65	85	dB	
				0°C	60	88		
				70°C	60	88		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V		25°C	65	95	dB	
				0°C	60	94		
				70°C	60	96		
I <sub>DD</sub>	Supply current (two amplifiers)	V <sub>O</sub> = 2.5 V, No load	V <sub>IC</sub> = 5 V,	25°C	1.9	4	mA	
				0°C	2.3	4.4		
				70°C	1.6	3.4		

† Full range is 0°C to 70°C.

NOTES: 4.. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5.. This range also applies to each input individually.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC272I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC272AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		7	
TLC272BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	230	2000	$\mu\text{V}$		
			Full range		3500			
TLC277I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	200	500			
			Full range		2000			
$\alpha_{VIO}$	Temperature coefficient of input offset voltage			25°C to 85°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				85°C	24	15		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				85°C	200	35		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8	V	
				-40°C	3	3.8		
				85°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 10\text{ k}\Omega$	25°C	5	23	V/mV	
				-40°C	3.5	32		
				85°C	3.5	19		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				-40°C	60	81		
				85°C	60	86		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-40°C	60	92		
				85°C	60	96		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	1.4	3.2	mA	
				-40°C	1.9	4.4		
				85°C	1.1	2.4		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC272I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC272AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		7	
	TLC272BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	290	2000	$\mu\text{V}$	
				Full range		3500		
		TLC277I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	250		800
					Full range			2900
$\alpha_{VIO}$	Temperature coefficient of input offset voltage			25°C to 85°C	2		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				-40°C	7.8	8.5		
				85°C	7.8	8.5		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				-40°C	7	46		
				85°C	7	31		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85	dB	
				-40°C	60	87		
				85°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-40°C	60	92		
				85°C	60	96		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	1.4	4	mA	
				-40°C	2.8	5		
				85°C	1.5	3.2		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC272M, TLC277M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC272M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC277M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	200	500	$\mu\text{V}$
					Full range		3750	
$\alpha_{VIO}$	Temperature coefficient of input offset voltage			25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8	V	
				-55°C	3	3.8		
				125°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	5	23	V/mV	
				-55°C	3.5	35		
				125°C	3.5	16		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				-55°C	60	81		
				125°C	60	84		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$ ,	25°C	1.4	3.2	mA	
				-55°C	2	5		
				125°C	1	2.2		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC272M, TLC277M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC272M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range	12		
		TLC277M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	250	800	$\mu\text{V}$
					Full range	4300		
$\alpha_{VIO}$	Temperature coefficient of input offset voltage			25°C to 125°C	2.2		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				-55°C	7.8	8.5		
				125°C	7.8	8.4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0 50		mV	
				-55°C	0 50			
				125°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				-55°C	7	50		
				125°C	7	27		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85	dB	
				-55°C	60	87		
				125°C	60	86		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	1.9	4	mA	
				-55°C	3 6			
				125°C	1.3	2.8		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## electrical characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC272Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Temperature coefficient of input offset voltage			1.8		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 10\text{ k}\Omega$	3.2	3.8		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ , $R_L = 10\text{ k}\Omega$	5	23		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	80		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		dB
$I_{DD}$ Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$		1.4	3.2	mA

## electrical characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC272Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Temperature coefficient of input offset voltage			1.8		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.7		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 9	-0.3 to 9.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 10\text{ k}\Omega$	8	8.5		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ , $R_L = 10\text{ k}\Omega$	10	36		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	85		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		dB
$I_{DD}$ Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load $V_{IC} = 5\text{ V}$		1.9	4	mA

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C	3.6		$V/\mu\text{s}$
			0°C	4		
		70°C	3			
		$V_{Ipp} = 2.5\text{ V}$	25°C	2.9		
			0°C	3.1		
			70°C	2.5		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	320		kHz
			0°C	340		
			70°C	260		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	1.7		MHz
			0°C	2		
			70°C	1.3		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	46°		
			0°C	47°		
			70°C	43°		

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C	5.3		$V/\mu\text{s}$
			0°C	5.9		
			70°C	4.3		
		$V_{Ipp} = 5.5\text{ V}$	25°C	4.6		
			0°C	5.1		
			70°C	3.8		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	200		kHz
			0°C	220		
			70°C	140		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	2.2		MHz
			0°C	2.5		
			70°C	1.8		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	49°		
			0°C	50°		
			70°C	46°		





# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
			MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I_{PP}} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s
			-40°C	4.5		
			85°C	2.8		
	$V_{I_{PP}} = 2.5\text{ V}$	25°C	2.9			
		-40°C	3.5			
		85°C	2.3			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	320		kHz
			-40°C	380		
			85°C	250		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	1.7		MHz
			-40°C	2.6		
			85°C	1.2		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	46°		
			-40°C	49°		
			85°C	43°		

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
			MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I_{PP}} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s
			-40°C	6.8		
			85°C	4		
	$V_{I_{PP}} = 5.5\text{ V}$	25°C	4.6			
		-40°C	5.8			
		85°C	3.5			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	200		kHz
			-40°C	260		
			85°C	130		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	2.2		MHz
			-40°C	3.1		
			85°C	1.7		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	49°		
			-40°C	52°		
			85°C	46°		

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272M, TLC277M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s
			-55°C	4.7		
			125°C	2.3		
		$V_{Ipp} = 2.5\text{ V}$	25°C	2.9		
			-55°C	3.7		
			125°C	2		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ , See Figure 1	25°C	320		kHz
			-55°C	400		
			125°C	230		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ , See Figure 3	25°C	1.7		MHz
			-55°C	2.9		
			125°C	1.1		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 3		25°C	46°		
			-55°C	49°		
			125°C	41°		

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$	TLC272M, TLC277M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s
			-55°C	7.1		
			125°C	3.1		
		$V_{Ipp} = 5.5\text{ V}$	25°C	4.6		
			-55°C	6.1		
			125°C	2.7		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ , See Figure 1	25°C	200		kHz
			-55°C	280		
			125°C	110		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ , See Figure 3	25°C	2.2		MHz
			-55°C	3.4		
			125°C	1.6		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , $f = B_1$ , See Figure 3		25°C	49°		
			-55°C	52°		
			125°C	44°		



# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC272Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	3.6		$\text{V}/\mu\text{s}$
		$V_{I\text{PP}} = 2.5\text{ V}$	2.9		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$ , See Figure 2	25			$\text{nV}/\sqrt{\text{Hz}}$
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $C_L = 20\text{ pF}$ , See Figure 1, $R_L = 10\text{ k}\Omega$	320			$\text{kHz}$
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , See Figure 3	1.7			$\text{MHz}$
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , See Figure 3, $C_L = 20\text{ pF}$	46°			

## operating characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC272Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	5.3		$\text{V}/\mu\text{s}$
		$V_{I\text{PP}} = 5.5\text{ V}$	4.6		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$ , See Figure 2	25			$\text{nV}/\sqrt{\text{Hz}}$
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $C_L = 20\text{ pF}$ , See Figure 1, $R_L = 10\text{ k}\Omega$	200			$\text{kHz}$
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , See Figure 3	2.2			$\text{MHz}$
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $f = B_1$ , See Figure 3, $C_L = 20\text{ pF}$	49°			

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC272 and TLC277 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

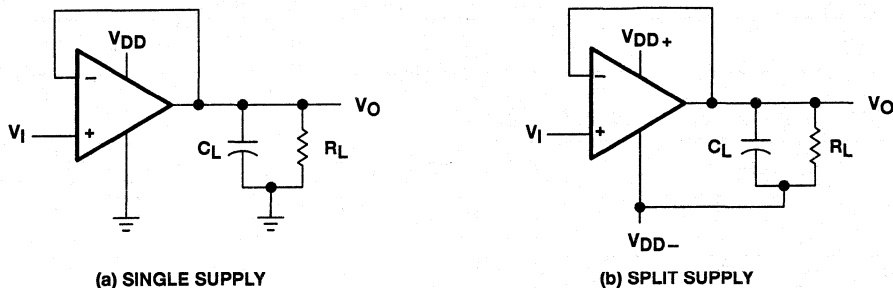


Figure 1. Unity-Gain Amplifier

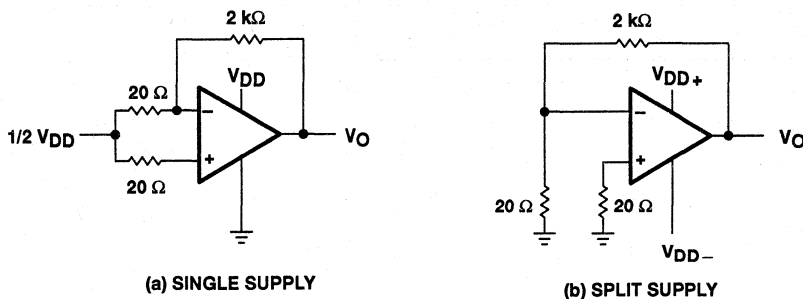


Figure 2. Noise-Test Circuit

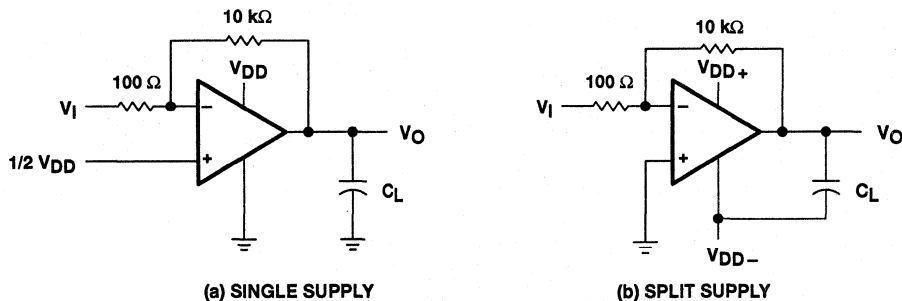


Figure 3. Gain-of-100 Inverting Amplifier

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC272 and TLC277 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

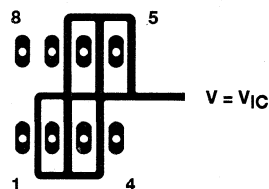


Figure 4. Isolation Metal Around Device Inputs  
 (JG and P packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

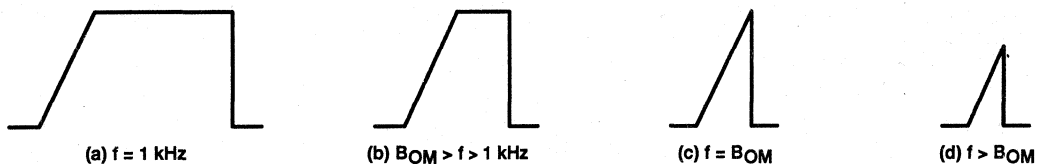


Figure 5. Full-Power-Response Output Signal

### test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current	10, 11
		vs Supply voltage	12
		vs Free-air temperature	13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	14, 15
		vs Differential input voltage	16
		vs Free-air temperature	17
		vs Low-level output current	18, 19
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	20
		vs Free-air temperature	21
		vs Frequency	32, 33
$I_{IB}$	Input bias current	vs Free-air temperature	22
$I_{IO}$	Input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage	24
		vs Free-air temperature	25
SR	Slew rate	vs Supply voltage	26
		vs Free-air temperature	27
		Normalized slew rate	vs Free-air temperature
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature	30
		vs Supply voltage	31
$\phi_m$	Phase margin	vs Supply voltage	34
		vs Free-air temperature	35
		vs Load capacitance	36
$V_n$	Equivalent input noise voltage	vs Frequency	37
		Phase shift	vs Frequency

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS

**DISTRIBUTION OF TLC272  
INPUT OFFSET VOLTAGE**

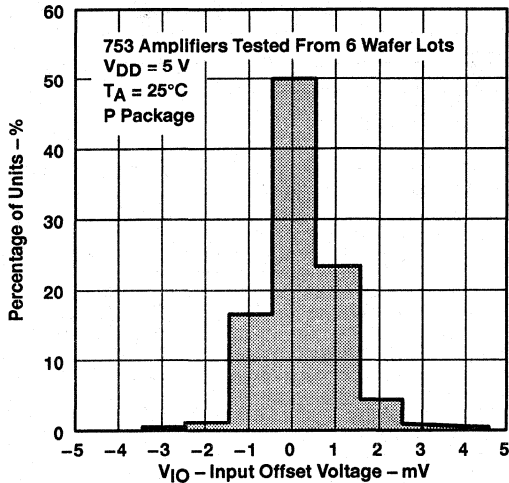


Figure 6

**DISTRIBUTION OF TLC272  
INPUT OFFSET VOLTAGE**

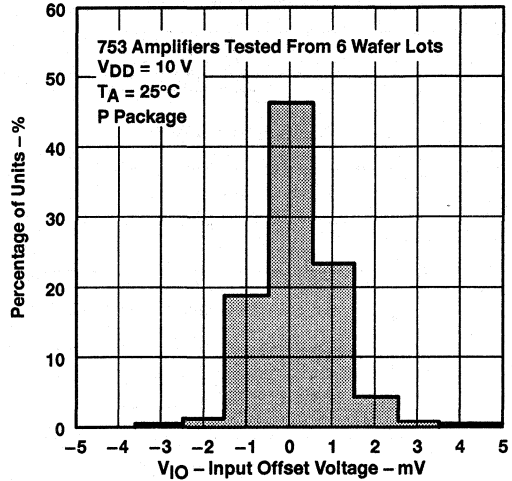


Figure 7

**DISTRIBUTION OF TLC272 AND TLC277  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

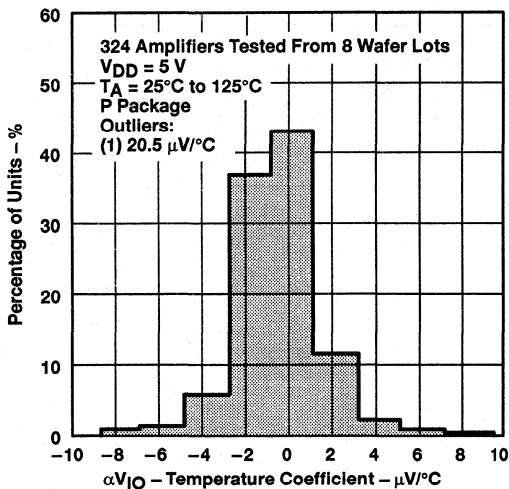


Figure 8

**DISTRIBUTION OF TLC272 AND TLC277  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

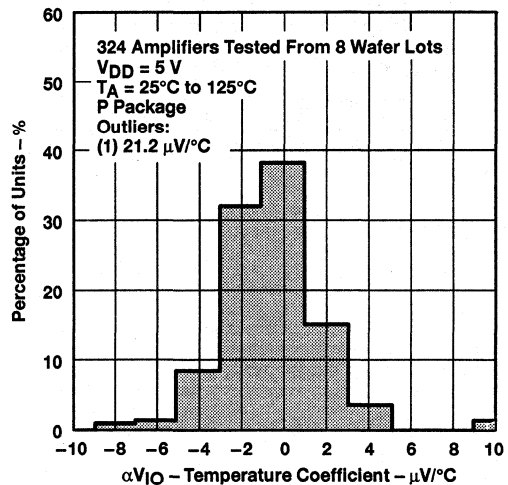


Figure 9



TYPICAL CHARACTERISTICS†

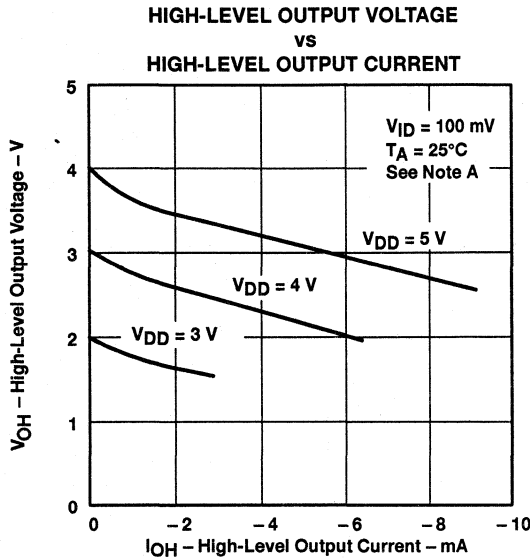


Figure 10

NOTE A: The 3-V curve only applies to the C version.

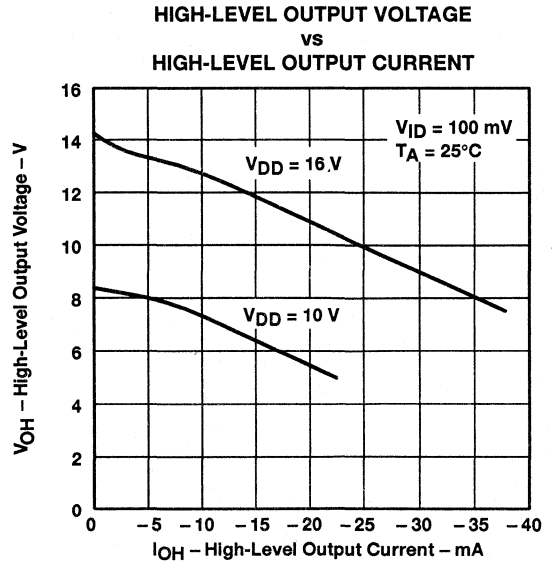


Figure 11

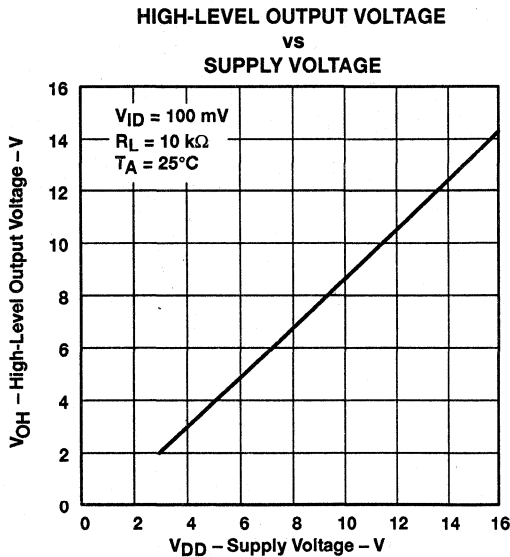


Figure 12

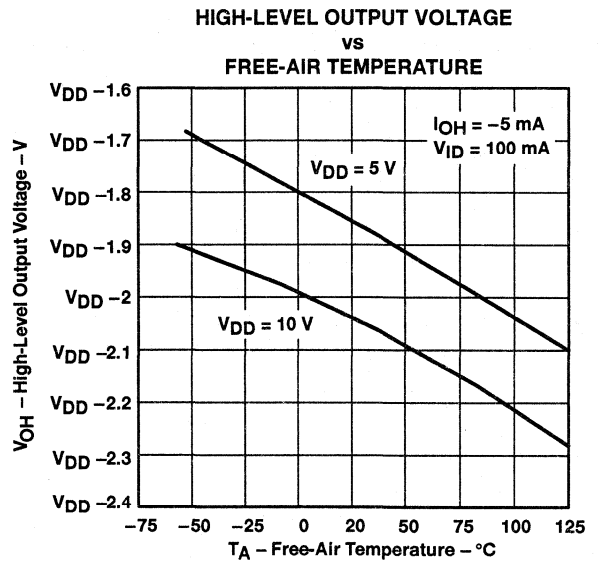


Figure 13

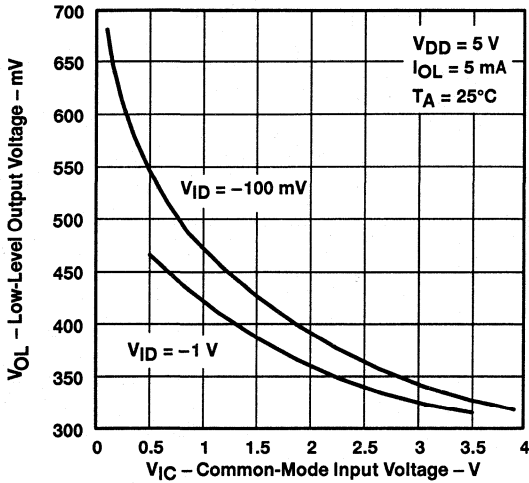
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC272, TLC272A, TLC272B, TLC272Y, TLC277**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

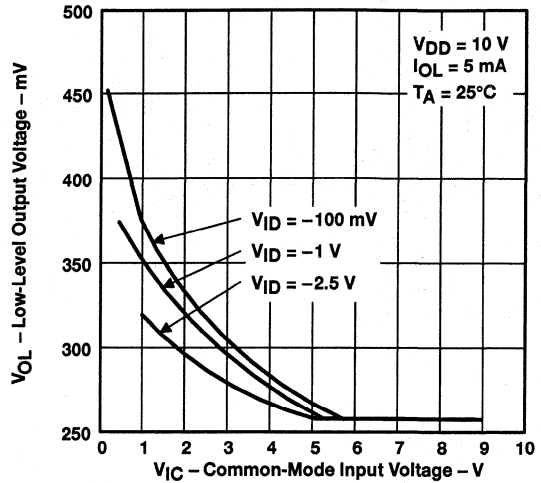
**TYPICAL CHARACTERISTICS†**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE**



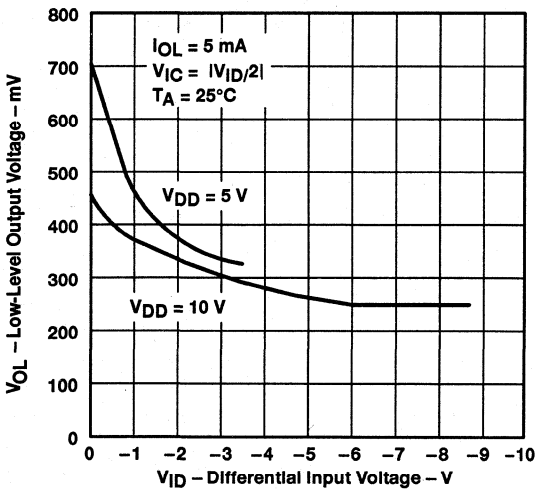
**Figure 14**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE**



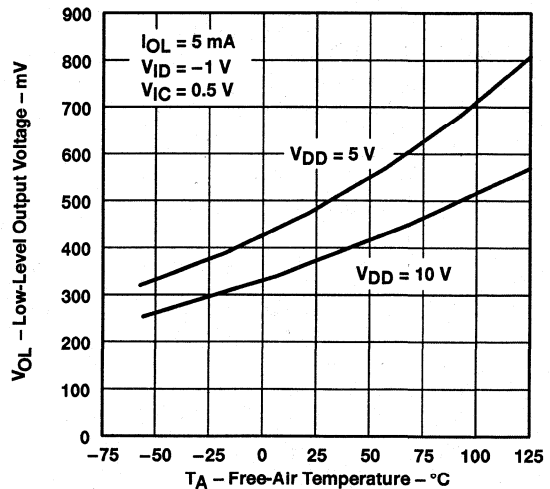
**Figure 15**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
DIFFERENTIAL INPUT VOLTAGE**



**Figure 16**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

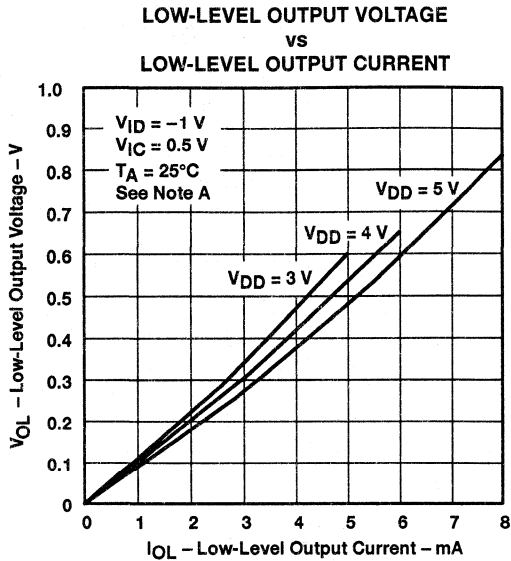


**Figure 17**

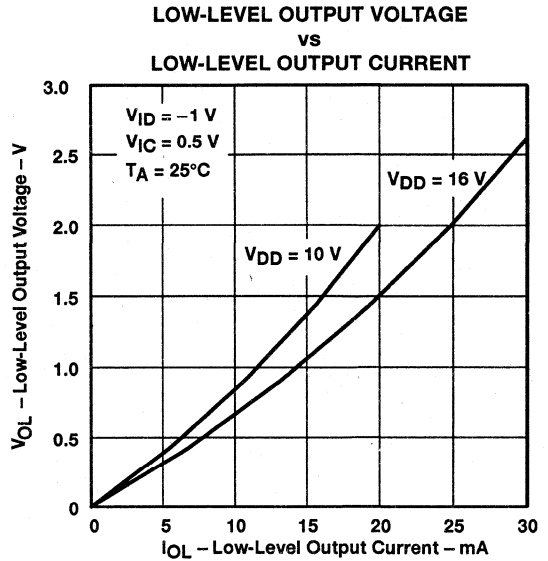
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



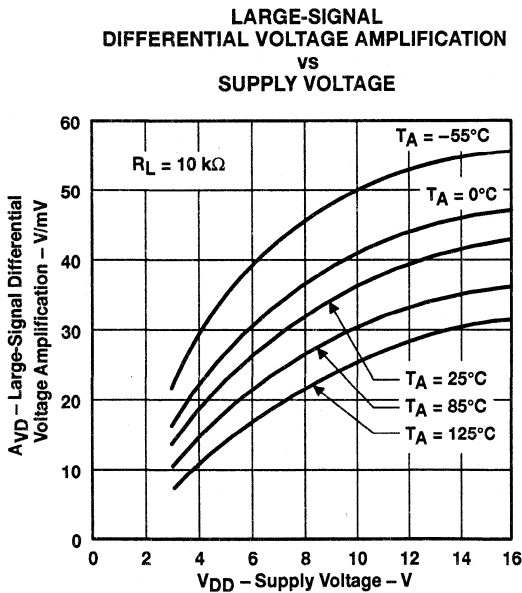
**TYPICAL CHARACTERISTICS†**



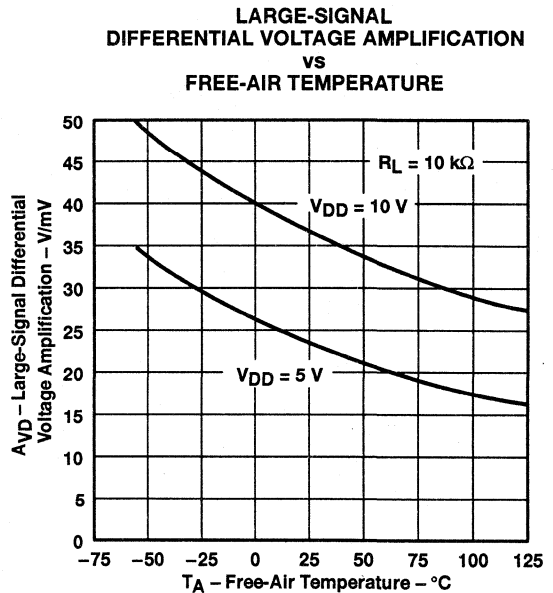
**Figure 18**



**Figure 19**



**Figure 20**



**Figure 21**

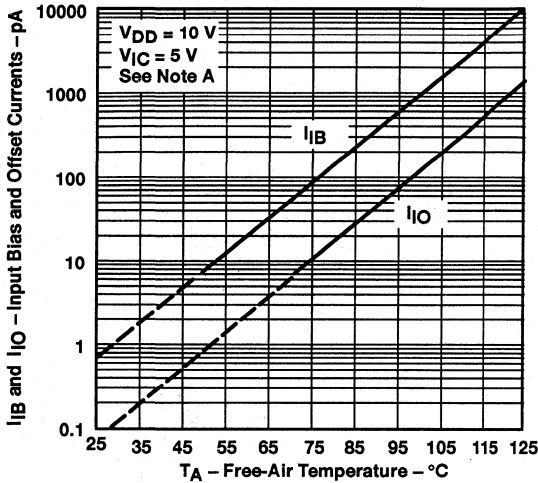
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1984

## TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

COMMON-MODE  
INPUT VOLTAGE POSITIVE LIMIT  
vs  
SUPPLY VOLTAGE

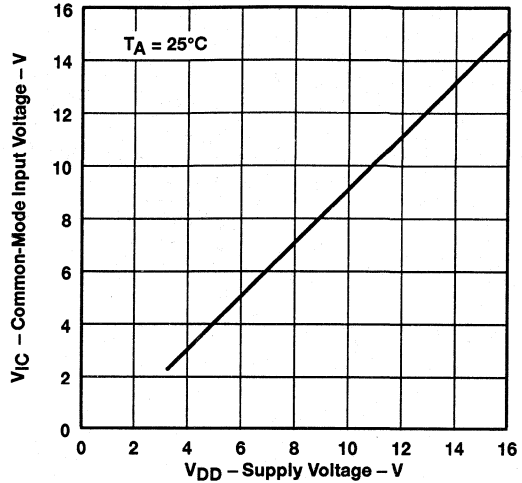


Figure 23

SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

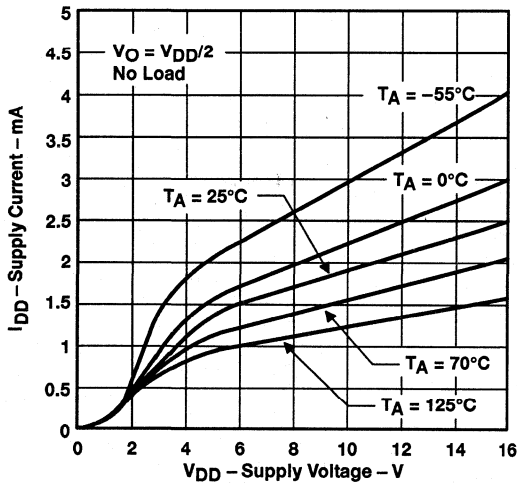


Figure 24

SUPPLY CURRENT  
vs  
FREE-AIR TEMPERATURE

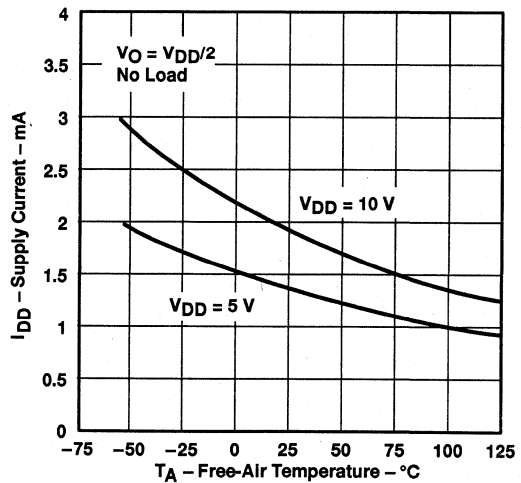


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

TYPICAL CHARACTERISTICS†

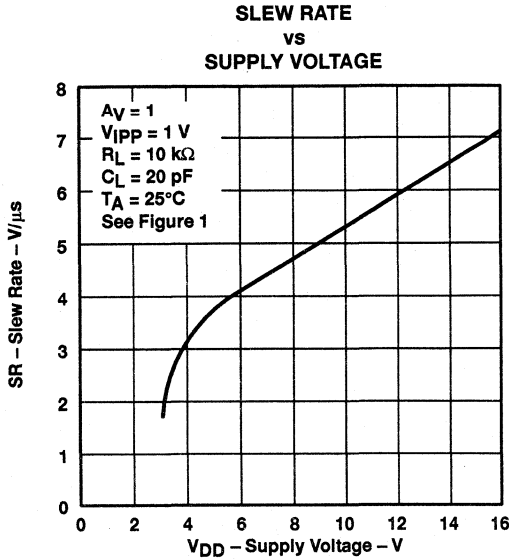


Figure 26

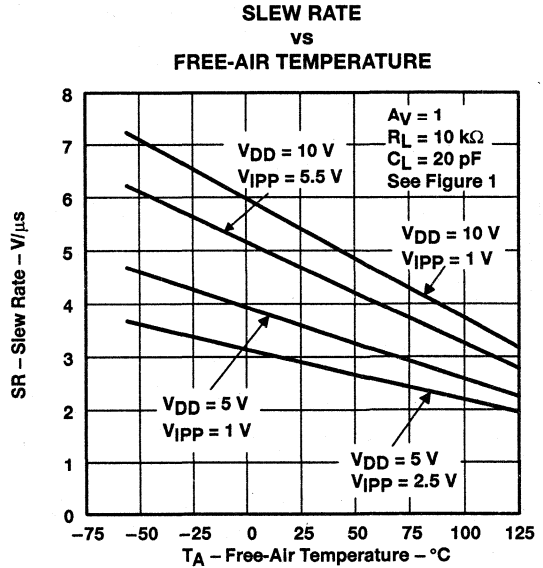


Figure 27

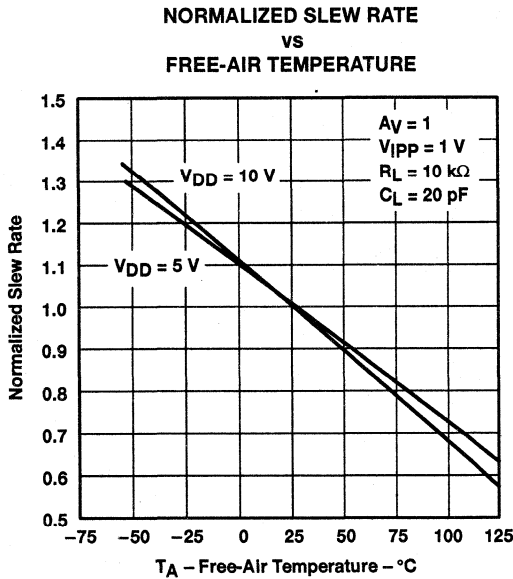


Figure 28

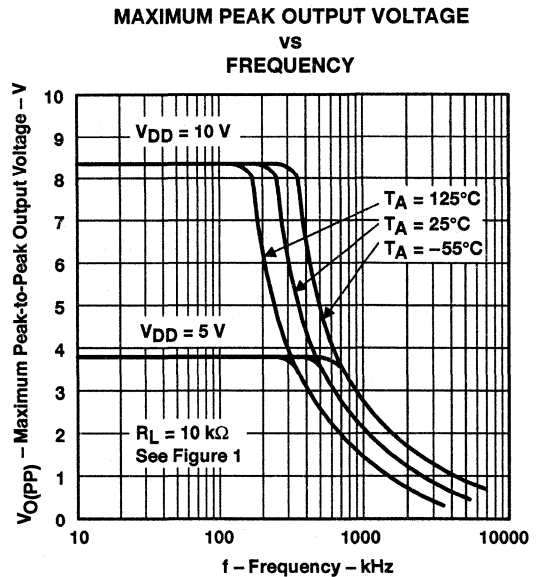


Figure 29

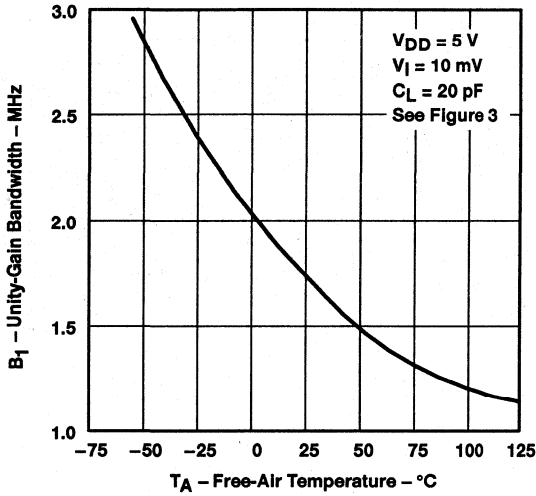
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC272, TLC272A, TLC272B, TLC272Y, TLC277**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

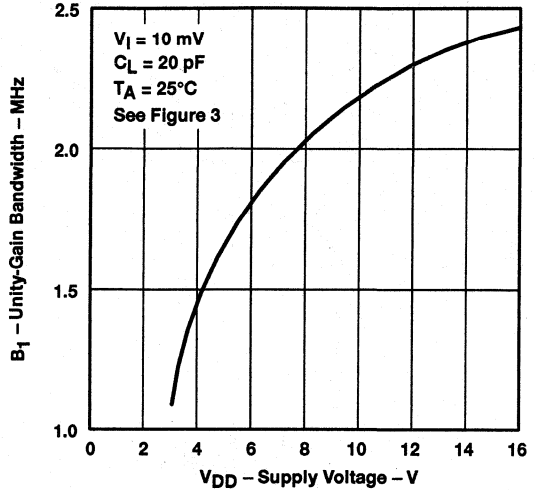
**TYPICAL CHARACTERISTICS†**

**UNITY-GAIN BANDWIDTH  
 vs  
 FREE-AIR TEMPERATURE**



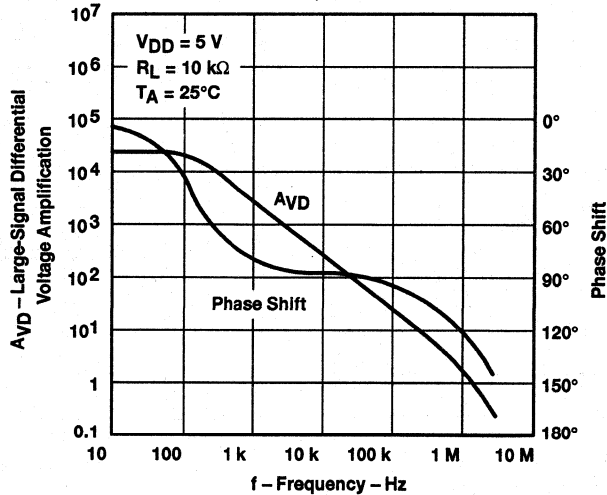
**Figure 30**

**UNITY-GAIN BANDWIDTH  
 vs  
 SUPPLY VOLTAGE**



**Figure 31**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY**



**Figure 32**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

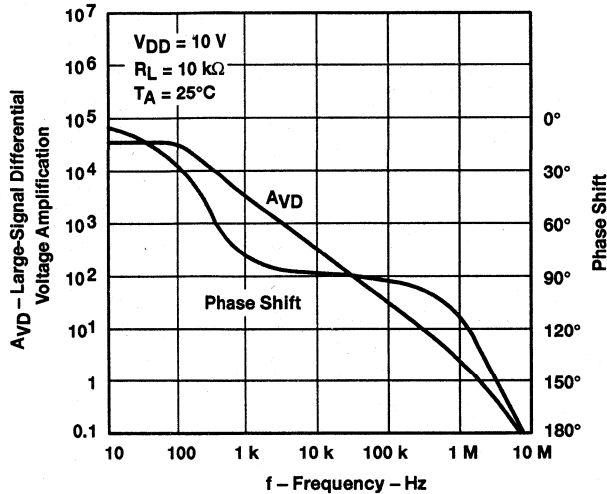


Figure 33

PHASE MARGIN  
 vs  
 SUPPLY VOLTAGE

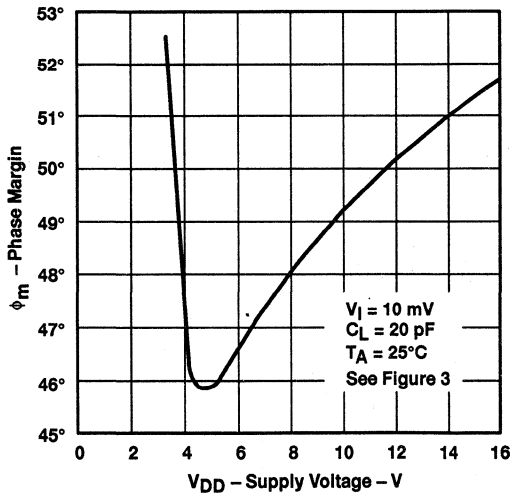


Figure 34

PHASE MARGIN  
 vs  
 FREE-AIR TEMPERATURE

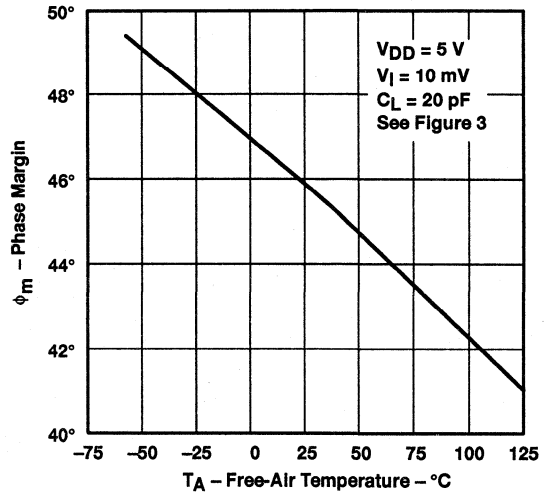


Figure 35

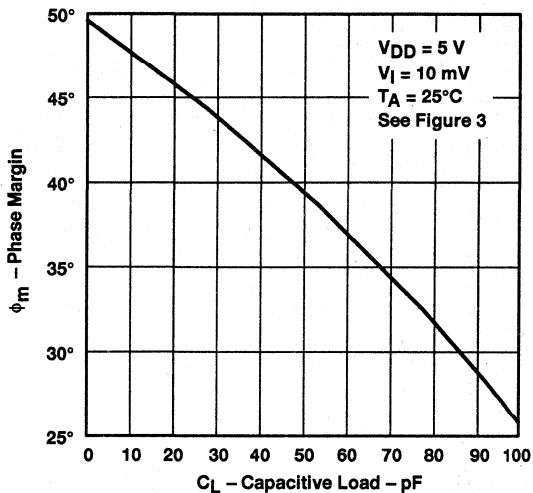
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC272, TLC272A, TLC272B, TLC272Y, TLC277**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

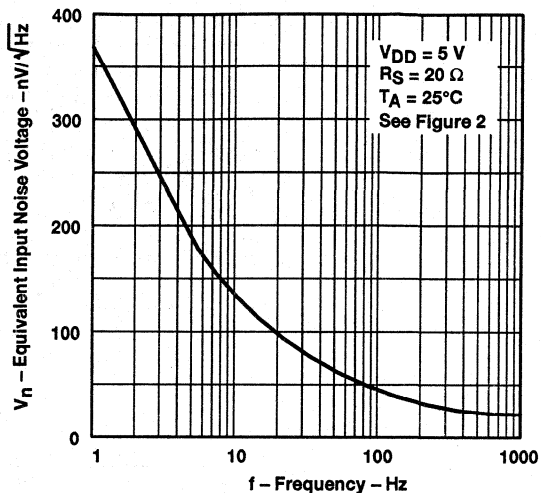
**TYPICAL CHARACTERISTICS**

**PHASE MARGIN  
 VS  
 CAPACITIVE LOAD**



**Figure 36**

**EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY**



**Figure 37**





APPLICATION INFORMATION

single-supply operation

While the TLC272 and TLC277 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC272 and TLC277 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC272 and TLC277 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

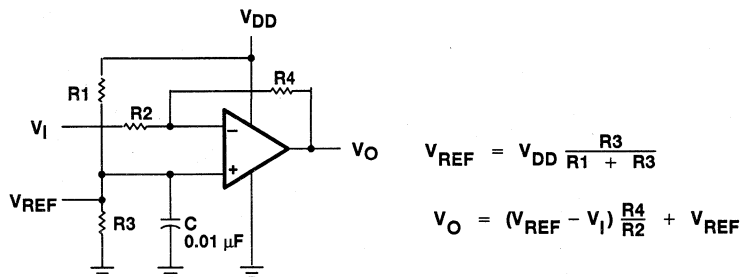
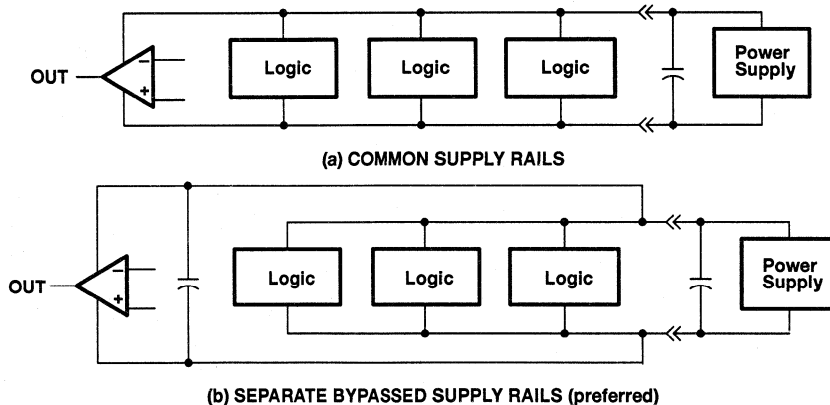


Figure 38. Inverting Amplifier With Voltage Reference



(b) SEPARATE BYPASSED SUPPLY RAILS (preferred)

Figure 39. Common vs Separate Supply Rails

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277

## LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

### APPLICATION INFORMATION

#### input characteristics

The TLC272 and TLC277 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5$  V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC272 and TLC277 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1 \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC272 and TLC277 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

#### noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC272 and TLC277 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50 \text{ k}\Omega$ , since bipolar devices exhibit greater noise currents.

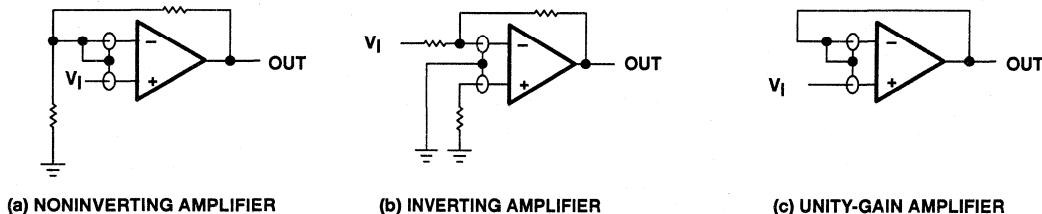


Figure 40. Guard-Ring Schemes

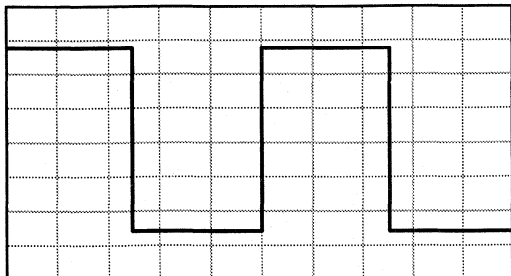
#### output characteristics

The output stage of the TLC272 and TLC277 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

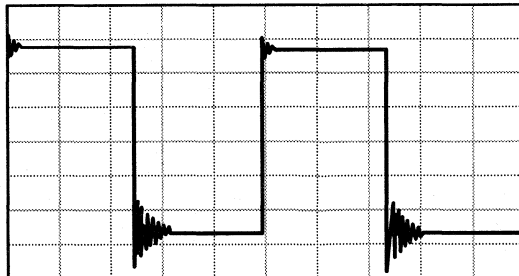
All operating characteristics of the TLC272 and TLC277 are measured using a  $20\text{-pF}$  load. The devices can drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

APPLICATION INFORMATION

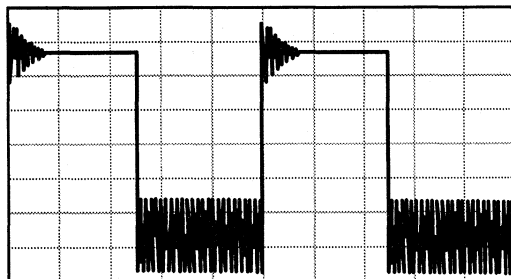
output characteristics (continued)



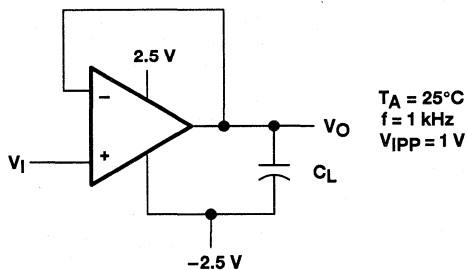
(a)  $C_L = 20 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(b)  $C_L = 130 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(c)  $C_L = 150 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(d) TEST CIRCUIT

Figure 41. Effect of Capacitive Loads and Test Circuit

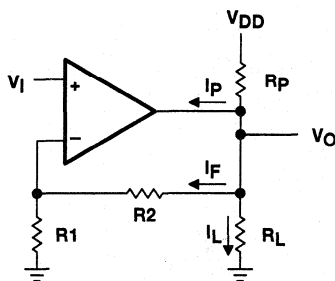
Although the TLC272 and TLC277 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on resistance between approximately  $60 \Omega$  and  $180 \Omega$ , depending on how hard the operational amplifier input is driven. With very low values of  $R_P$ , a voltage offset from 0 V at the output occurs. Second, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### output characteristics (continued)



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

$I_p$  = Pullup current required by the operational amplifier (typically 500  $\mu$ A)

Figure 42. Resistive Pullup to Increase  $V_{OH}$

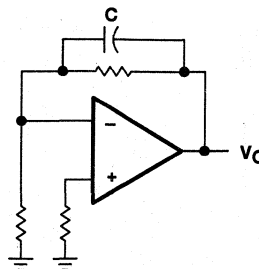


Figure 43. Compensation for Input Capacitance

### feedback

Operational amplifier circuits almost always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

### electrostatic discharge protection

The TLC272 and TLC277 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

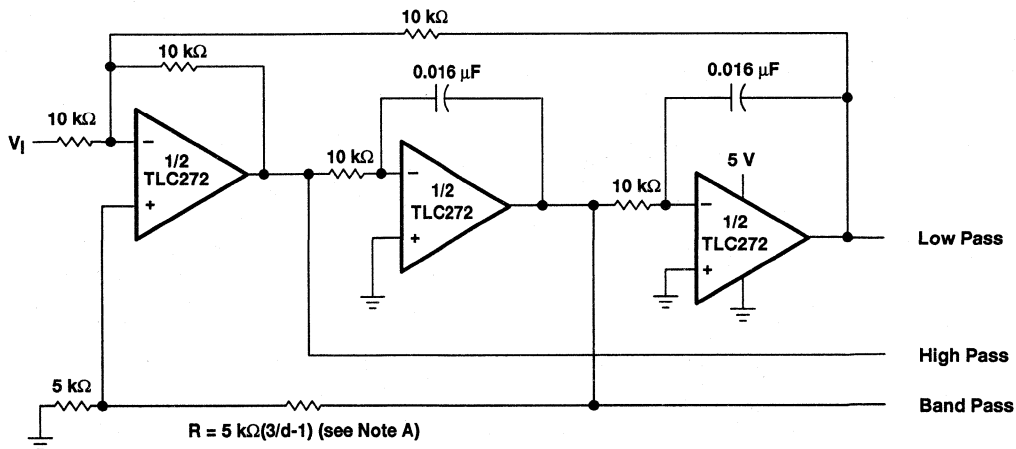
### latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC272 and TLC277 inputs and outputs were designed to withstand  $-100$ -mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1  $\mu$ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

 **TEXAS  
INSTRUMENTS**

APPLICATION INFORMATION



NOTE A: d = damping factor, 1/Q

Figure 44. State-Variable Filter

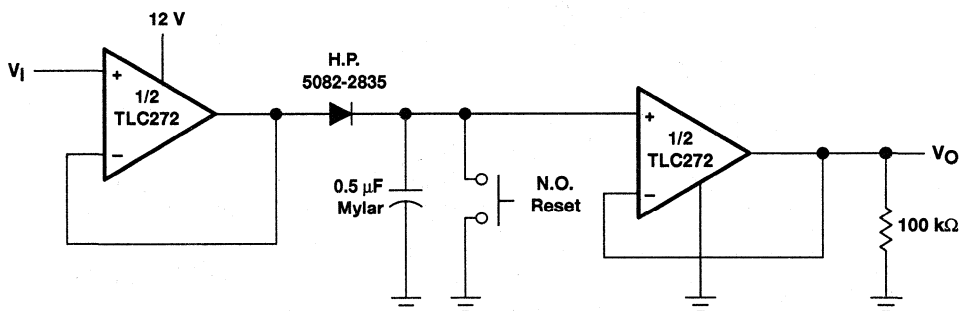
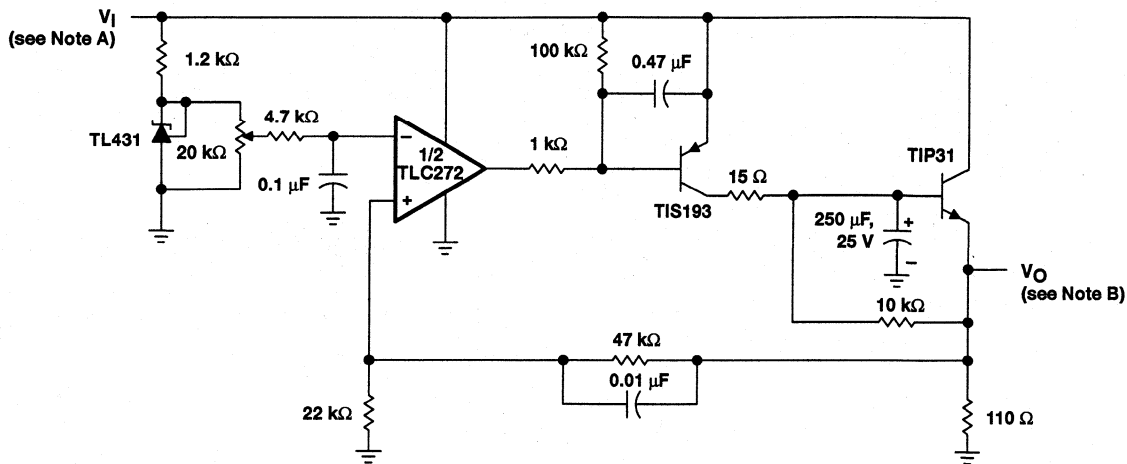


Figure 45. Positive-Peak Detector.

# TLC272, TLC272A, TLC272B, TLC272Y, TLC277 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

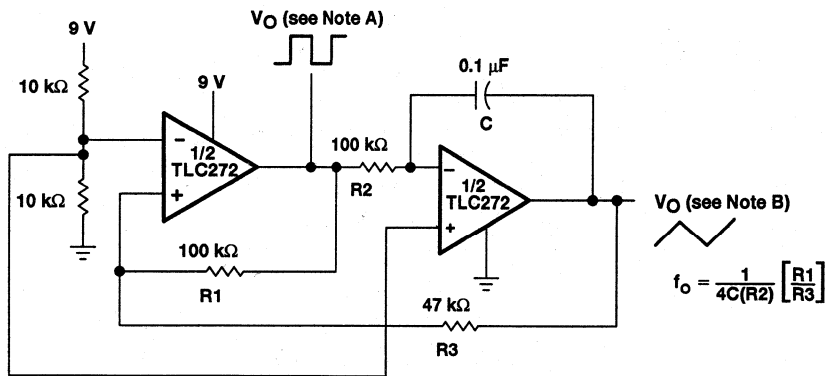
SLOS091B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION



- NOTES: A.  $V_I = 3.5$  to  $15$  V  
B.  $V_O = 2$  V, 0 to  $1$  A

Figure 46. Logic-Array Power Supply

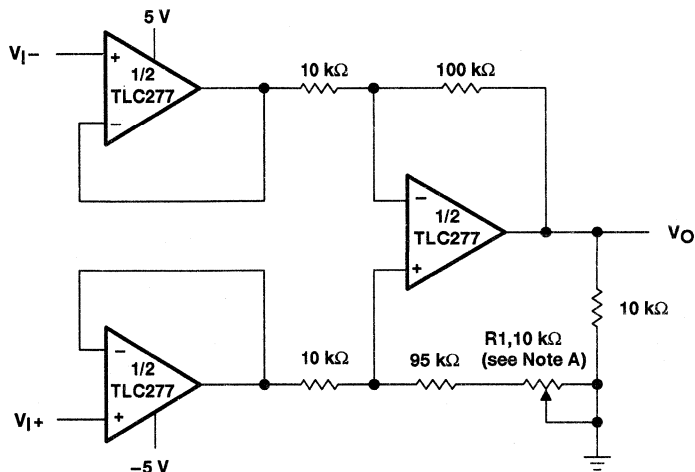


- NOTES: A.  $V_{O(PP)} = 8$  V  
B.  $V_{O(PP)} = 4$  V

Figure 47. Single-Supply Function Generator

$$f_o = \frac{1}{4C(R2)} \left[ \frac{R1}{R3} \right]$$

APPLICATION INFORMATION



NOTE A: CMRR adjustment must be noninductive.

Figure 48. Low-Power Instrumentation Amplifier

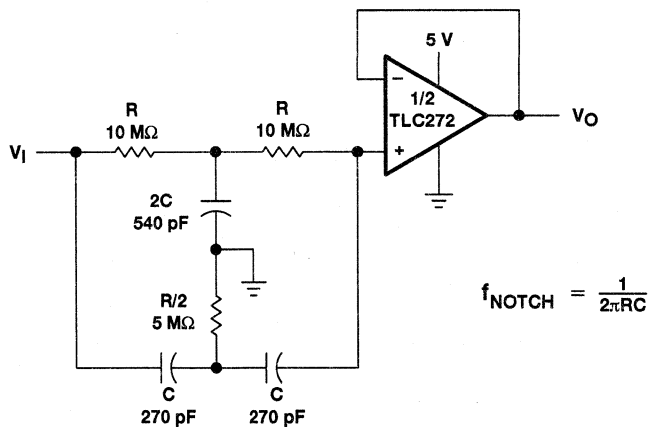


Figure 49. Single-Supply Twin-T Notch Filter



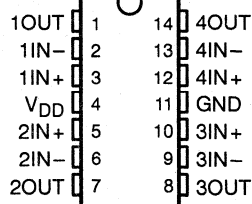


# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

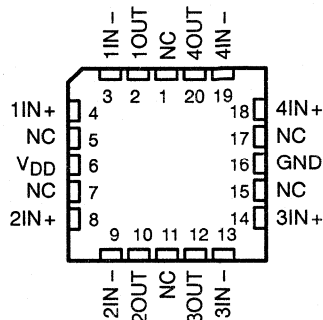
SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC279 . . . 900  $\mu\text{V}$  Max at 25°C,  $V_{\text{DD}} = 5 \text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
0°C to 70°C . . . 3 V to 16 V  
–40°C to 85°C . . . 4 V to 16 V  
–55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix and I-Suffix Versions)**
- **Low Noise . . . Typically 25 nV/ $\sqrt{\text{Hz}}$  at  $f = 1 \text{ kHz}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . .  $10^{12} \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**

D, J, N, OR PW PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

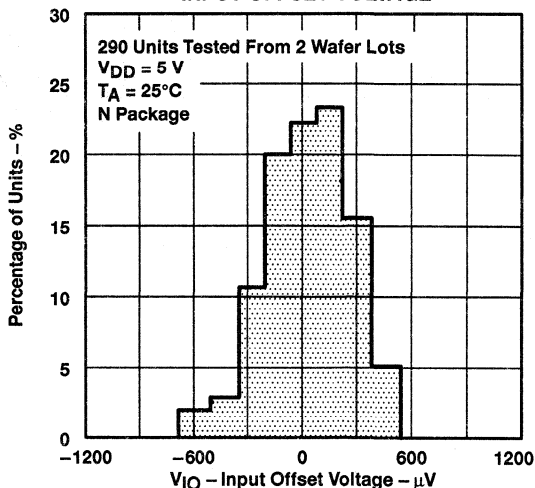
## description

The TLC274 and TLC279 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose BIFET devices.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for BIFET and NFET products. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC274 (10  $\mu\text{V}$ ) to the high-precision TLC279 (900  $\mu\text{V}$ ). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

DISTRIBUTION OF TLC279  
INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2-573

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## description (continued)

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC274 and TLC279. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up.

The TLC274 and TLC279 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	900 μV	TLC279CD	—	—	TLC279CN	—	—
	2 mV	TLC274BCD	—	—	TLC274BCN	—	—
	5 mV	TLC274ACD	—	—	TLC274ACN	—	—
	10 mV	TLC274CD	—	—	TLC274CN	TLC274CPW	TLC274Y
–40°C to 85°C	900 μV	TLC279ID	—	—	TLC279IN	—	—
	2 mV	TLC274BID	—	—	TLC274BIN	—	—
	5 mV	TLC274AID	—	—	TLC274AIN	—	—
	10 mV	TLC274ID	—	—	TLC274IN	—	—
–55°C to 125°C	900 μV	TLC279MD	TLC279MFK	TLC279MJ	TLC279MN	—	—
	10 mV	TLC274MD	TLC274MFK	TLC274MJ	TLC274MN	—	—

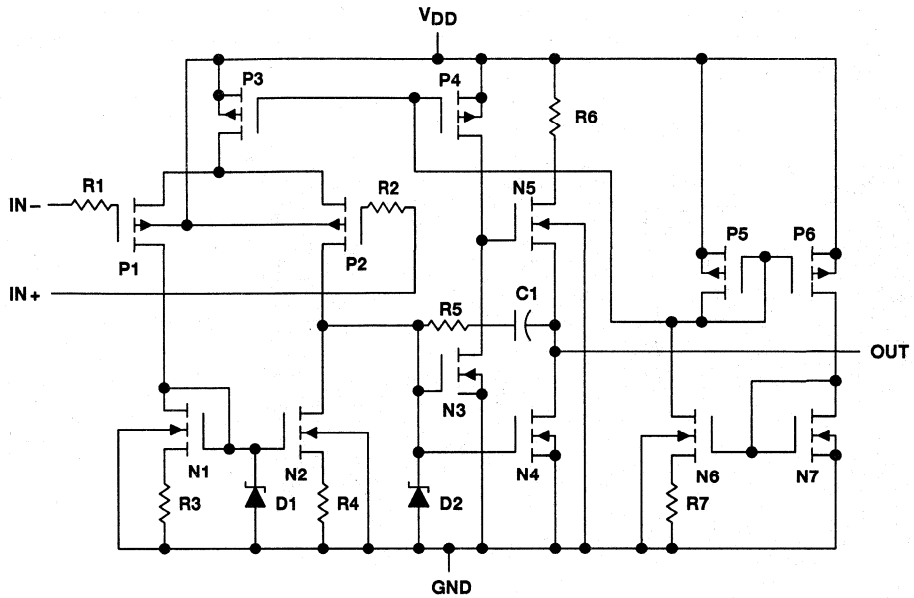
The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC279CDR).



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

equivalent schematic (each amplifier)

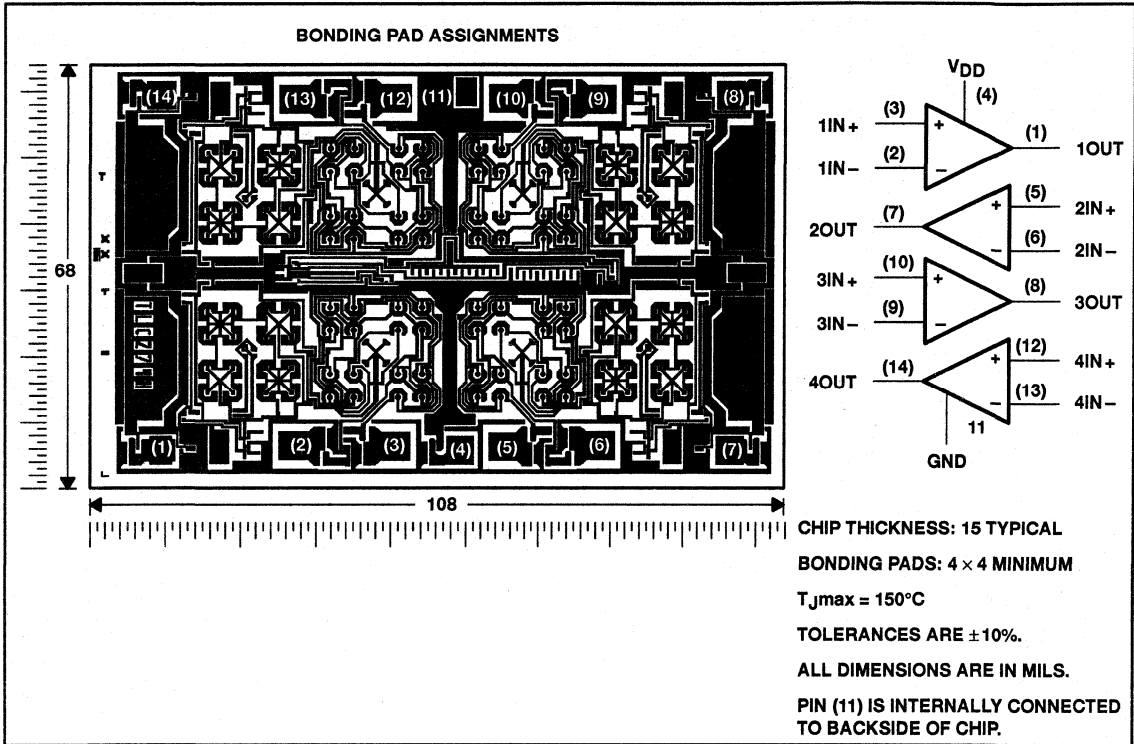


# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## TLC274Y chip information

These chips, when properly assembled, display characteristics similar to the TLC274C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	$-0.3$ V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at the noninverting input with respect to the inverting input.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	—
PW	700 mW	5.6 mW/°C	448 mW	—	—

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC274C, TLC274AC, TLC274BC, TLC279C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC274C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC274AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC274BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	340	2000	$\mu\text{V}$
					Full range		3000	
		TLC279C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	320	900	$\mu\text{V}$
					Full range		1500	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\ \text{k}\Omega$	25°C	3.2	3.8	V	
				0°C	3	3.8		
				70°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 10\ \text{k}\Omega$	25°C	5	23	V/mV	
				0°C	4	27		
				70°C	4	20		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				0°C	60	84		
				70°C	60	85		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	2.7	6.4	mA	
				0°C	3.1	7.2		
				70°C	2.3	5.2		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC274C, TLC274AC, TLC274BC, TLC279C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC274C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC274AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
		TLC274BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	390	2000	$\mu\text{V}$
					Full range		3000	
		TLC279C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	370	1200	
					Full range		1900	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				0°C	7.8	8.5		
				70°C	7.8	8.4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				0°C	7.5	42		
				70°C	7.5	32		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85	dB	
				0°C	60	88		
				70°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	3.8	8	mA	
				0°C	4.5	8.8		
				70°C	3.2	6.8		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC274I, TLC274AI, TLC274BI, TLC279I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC274I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC274AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	0.9	5	mV
					Full range		7	
		TLC274BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	340	2000	$\mu\text{V}$
					Full range		3500	
TLC279I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	320	900	$\mu\text{V}$		
			Full range		2000			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	1.8		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				85°C	24	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				85°C	200	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\ \text{k}\Omega$	25°C	3.2	3.8	V	
				-40°C	3	3.8		
				85°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 10\ \text{k}\Omega$	25°C	5	23	V/mV	
				-40°C	3.5	32		
				85°C	3.5	19		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				-40°C	60	81		
				85°C	60	86		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-40°C	60	92		
				85°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	2.7	6.4	mA	
				-40°C	3.8	8.8		
				85°C	2.1	4.8		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC274I, TLC274AI, TLC274BI, TLC279I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC274I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC274AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		7	
		TLC274BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	390	2000	$\mu\text{V}$
					Full range		3500	
		TLC279I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	370	1200	$\mu\text{V}$
					Full range		2900	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	2		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\text{ k}\Omega$	25°C	8	8.5	V	
				-40°C	7.8	8.5		
				85°C	7.8	8.5		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 10\text{ k}\Omega$	25°C	10	36	V/mV	
				-40°C	7	47		
				85°C	7	31		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85	dB	
				-40°C	60	87		
				85°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	65	95	dB	
				-40°C	60	92		
				85°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	3.8	8	mA	
				-40°C	5.5	10		
				85°C	2.9	6.4		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC274M, TLC279M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC274M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC279M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	320	900	$\mu\text{V}$
					Full range		3750	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8	V	
				-55°C	3	3.8		
				125°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	5	23	V/mV	
				-55°C	3.5	35		
				125°C	3.5	16		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80	dB	
				-55°C	60	81		
				125°C	60	84		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	2.7	6.4	mA	
				-55°C	4	10		
				125°C	1.9	4.4		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless) otherwise noted)**

PARAMETER		TEST CONDITIONS		T <sub>A</sub> †	TLC274M, TLC279M		UNIT	
					MIN	TYP		MAX
V <sub>IO</sub>	Input offset voltage	TLC274M	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	1.1	10	mV
					Full range		12	
		TLC279M	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 Ω,	V <sub>IC</sub> = 0, R <sub>L</sub> = 10 kΩ	25°C	370	1200	μV
					Full range		4300	
αV <sub>IO</sub>	Average temperature coefficient of input offset voltage			25°C to 125°C	2.2		μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = 5 V,	V <sub>IC</sub> = 5 V	25°C	0.1		pA	
				125°C	1.8	15	nA	
I <sub>IB</sub>	Input bias current (see Note 4)	V <sub>O</sub> = 5 V,	V <sub>IC</sub> = 5 V	25°C	0.7		pA	
				125°C	10	35	nA	
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV,	R <sub>L</sub> = 10 kΩ	25°C	8	8.5	V	
				-55°C	7.8	8.5		
				125°C	7.8	8.4		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV,	I <sub>OL</sub> = 0	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
A <sub>VD</sub>	Large-signal differential voltage amplification	V <sub>O</sub> = 1 V to 6 V,	R <sub>L</sub> = 10 kΩ	25°C	10	36	V/mV	
				-55°C	7	50		
				125°C	7	27		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>		25°C	65	85	dB	
				-55°C	60	87		
				125°C	60	86		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V,	V <sub>O</sub> = 1.4 V	25°C	65	95	dB	
				-55°C	60	90		
				125°C	60	97		
I <sub>DD</sub>	Supply current (four amplifiers)	V <sub>O</sub> = 5 V, No load	V <sub>IC</sub> = 5 V,	25°C	3.8	8	mA	
				-55°C	6.0	12		
				125°C	2.5	5.6		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274C, TLC274AC, TLC274BC, TLC279C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	3.6		V/ $\mu\text{s}$	
			0°C	4			
			70°C	3			
		$V_{Ipp} = 2.5\ \text{V}$	25°C	2.9			
			0°C	3.1			
			70°C	2.5			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\ \text{k}\Omega$ ,	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	320		kHz	
			0°C	340			
			70°C	260			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$ ,	25°C	1.7		MHz	
			0°C	2			
			70°C	1.3			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$ ,	$f = B_1$ ,	25°C	46°			
			0°C	47°			
			70°C	44°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\ \text{V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274C, TLC274AC, TLC274BC, TLC279C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	5.3		V/ $\mu\text{s}$	
			0°C	5.9			
			70°C	4.3			
		$V_{Ipp} = 5.5\ \text{V}$	25°C	4.6			
			0°C	5.1			
			70°C	3.8			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\ \text{k}\Omega$ ,	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	200		kHz	
			0°C	220			
			70°C	140			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$ ,	25°C	2.2		MHz	
			0°C	2.5			
			70°C	1.8			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$ ,	$f = B_1$ , See Figure 3	25°C	49°			
			0°C	50°			
			70°C	46°			



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274I, TLC274AI, TLC274BI, TLC279I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	3.6		V/ $\mu$ s	
			-40°C	4.5			
			85°C	2.8			
		$V_{IPP} = 2.5\text{ V}$	25°C	2.9			
			-40°C	3.5			
			85°C	2.3			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	320		kHz	
			-40°C	380			
			85°C	250			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	1.7		MHz	
			-40°C	2.6			
			85°C	1.2			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	46°			
			-40°C	49°			
			85°C	43°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274I, TLC274AI, TLC274BI, TLC279I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	5.3		V/ $\mu$ s	
			-40°C	6.7			
			85°C	4			
		$V_{IPP} = 5.5\text{ V}$	25°C	4.6			
			-40°C	5.8			
			85°C	3.5			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	25		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	200		kHz	
			-40°C	260			
			85°C	130			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	2.2		MHz	
			-40°C	3.1			
			85°C	1.7			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	49°			
			-40°C	52°			
			85°C	46°			

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274M, TLC279M			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		3.6	V/ $\mu$ s	
			-55°C		4.7		
			125°C		2.3		
		$V_{Ipp} = 2.5\text{ V}$	25°C		2.9		
			-55°C		3.7		
			125°C		2		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C		25	nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C		320	kHz	
			-55°C		400		
			125°C		230		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C		1.7	MHz	
			-55°C		2.9		
			125°C		1.1		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C		46°		
			-55°C		49°		
			125°C		41°		

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC274M, TLC279M			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	25°C		5.3	V/ $\mu$ s	
			-55°C		7.1		
			125°C		3.1		
		$V_{Ipp} = 5.5\text{ V}$	25°C		4.6		
			-55°C		6.1		
			125°C		2.7		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C		25	nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C		200	kHz	
			-55°C		280		
			125°C		110		
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C		2.2	MHz	
			-55°C		3.4		
			125°C		1.6		
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C		49°		
			-55°C		52°		
			125°C		44°		



# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## electrical characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC274Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$		1.1	10	mV
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 10\ \text{k}\Omega$	3.2	3.8		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ , $R_L = 10\ \text{k}\Omega$	5	23		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	80		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$		2.7	6.4	mA

## electrical characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC274Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$		1.1	10	mV
$I_{IO}$ Input offset current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.7		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 9	-0.3 to 9.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 10\ \text{k}\Omega$	8	8.5		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ , $R_L = 10\ \text{k}\Omega$	10	36		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	85		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	65	95		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load $V_{IC} = 5\text{ V}$		3.8	8	mA

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

### operating characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC274Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	$V_{IPP} = 1\text{ V}$	3.6	$\text{V}/\mu\text{s}$
			$V_{IPP} = 2.5\text{ V}$	2.9	
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$ , See Figure 2		25	$\text{nV}/\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ , $R_L = 10\text{ k}\Omega$	320	$\text{kHz}$	
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , See Figure 3		1.7	$\text{MHz}$	
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ , $C_L = 20\text{ pF}$	$46^\circ$		

### operating characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC274Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	$V_{IPP} = 1\text{ V}$	5.3	$\text{V}/\mu\text{s}$
			$V_{IPP} = 5.5\text{ V}$	4.6	
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , $R_S = 20\ \Omega$ , See Figure 2		25	$\text{nV}/\sqrt{\text{Hz}}$	
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , See Figure 1	$C_L = 20\text{ pF}$ , $R_L = 10\text{ k}\Omega$	200	$\text{kHz}$	
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ , See Figure 3		2.2	$\text{MHz}$	
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , See Figure 3	$f = B_1$ , $C_L = 20\text{ pF}$	$49^\circ$		



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC274 and TLC279 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

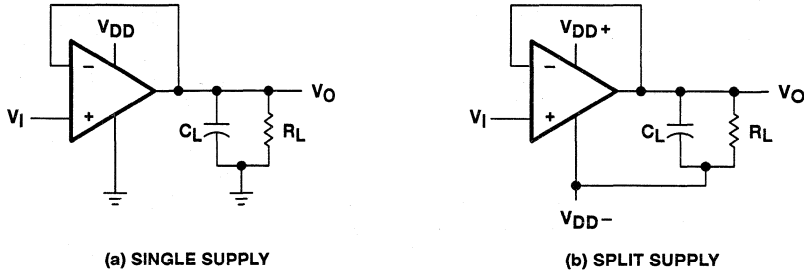


Figure 1. Unity-Gain Amplifier

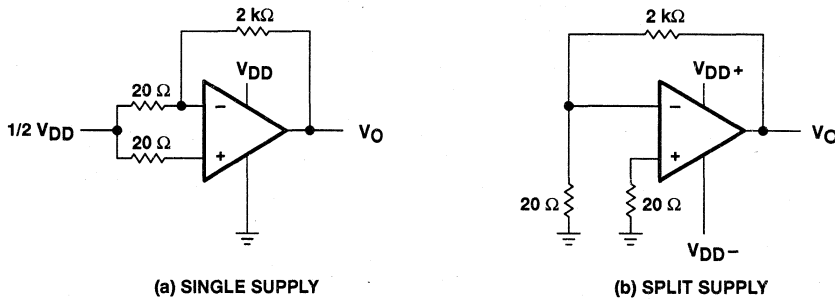


Figure 2. Noise-Test Circuit

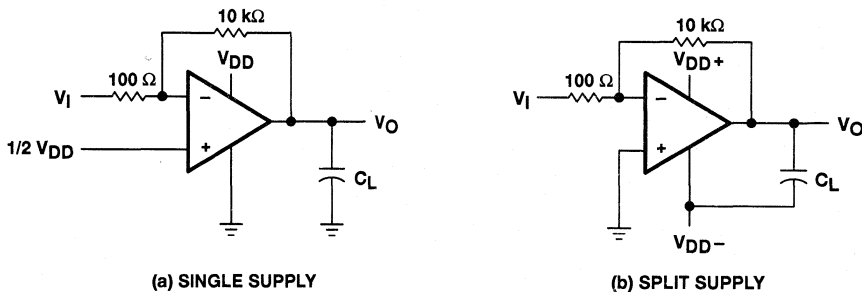


Figure 3. Gain-of-100 Inverting Amplifier

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC274 and TLC279 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

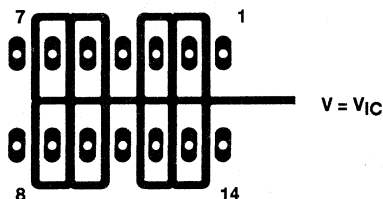


Figure 4. Isolation Metal Around Device Inputs (J and N packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

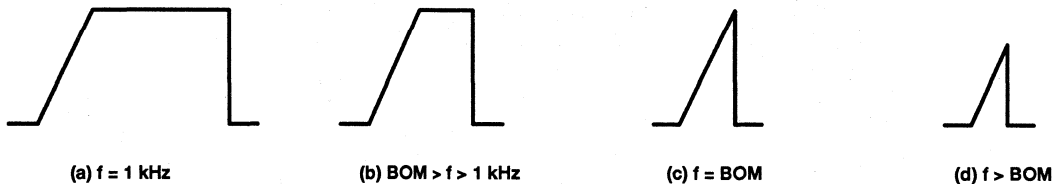


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

**TLC274, TLC274A, TLC274B, TLC274Y, TLC279**  
**LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS**

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	10, 11 12 13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	14, 15 16 17 18, 19
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	20 21 32, 33
$I_{IB}$	Input bias current	vs Free-air temperature	22
$I_{IO}$	Input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	24 25
$SR$	Slew rate	vs Supply voltage vs Free-air temperature	26 27
	Normalized slew rate	vs Free-air temperature	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	30 31
$\phi_m$	Phase margin	vs Supply voltage vs Free-air temperature vs Load capacitance	34 35 36
$V_n$	Equivalent input noise voltage	vs Frequency	37
	Phase shift	vs Frequency	32, 33



TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC274  
 INPUT OFFSET VOLTAGE

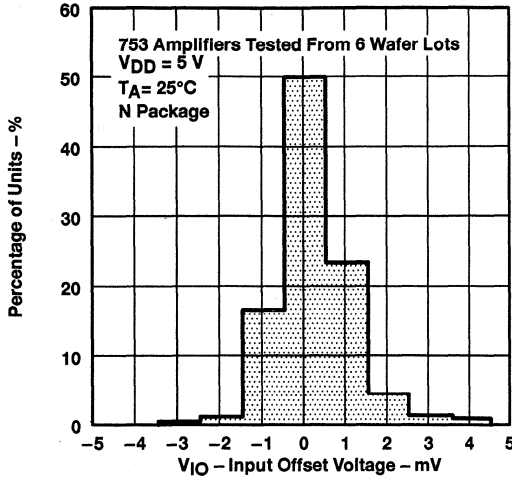


Figure 6

DISTRIBUTION OF TLC274  
 INPUT OFFSET VOLTAGE

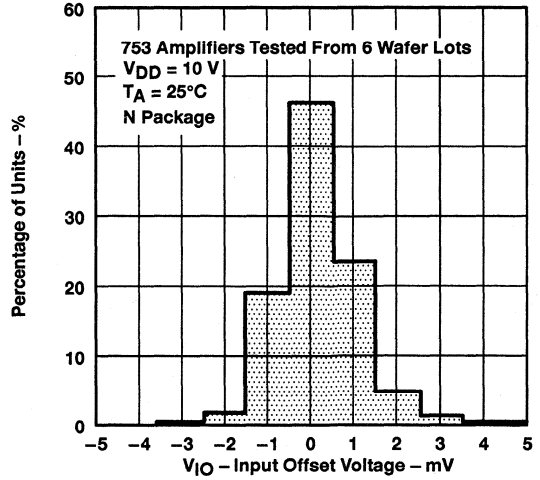


Figure 7

DISTRIBUTION OF TLC274 AND TLC279  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

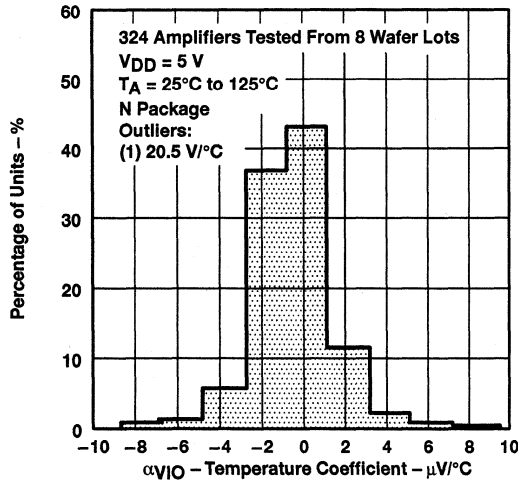


Figure 8

DISTRIBUTION OF TLC274 AND TLC279  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

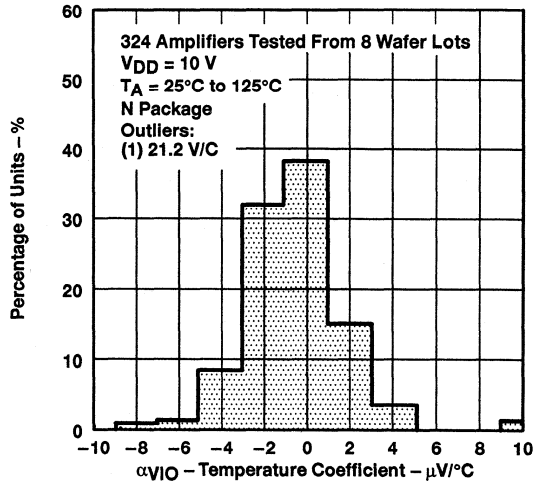


Figure 9

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

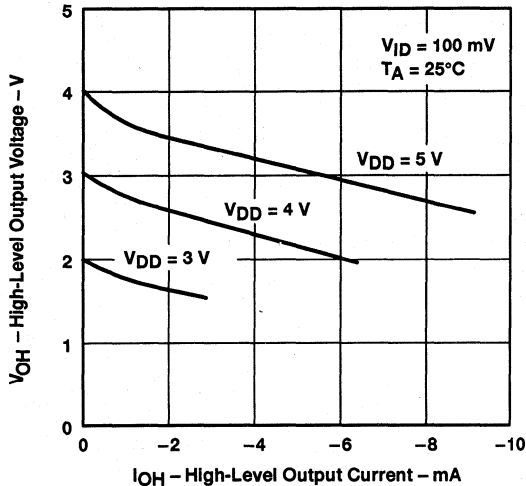


Figure 10

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

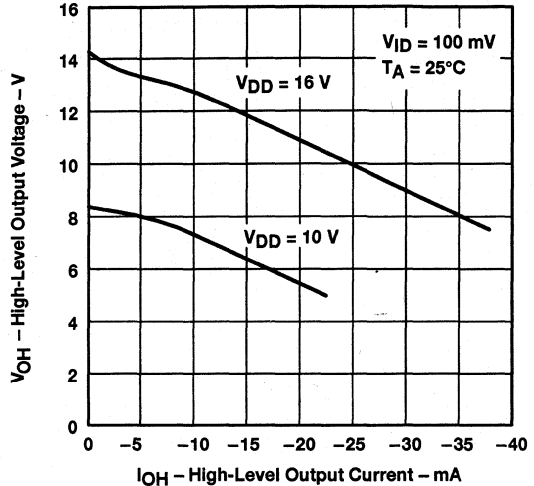


Figure 11

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE**

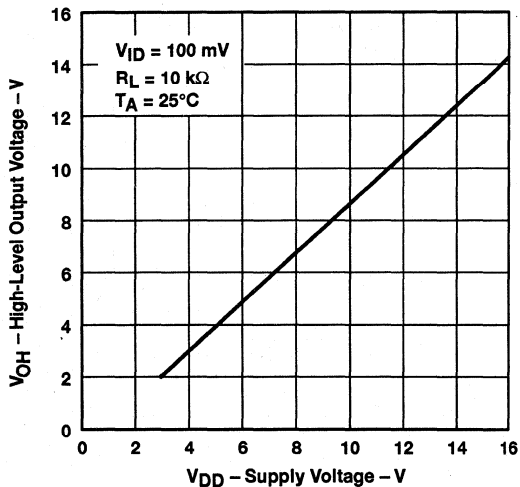


Figure 12

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

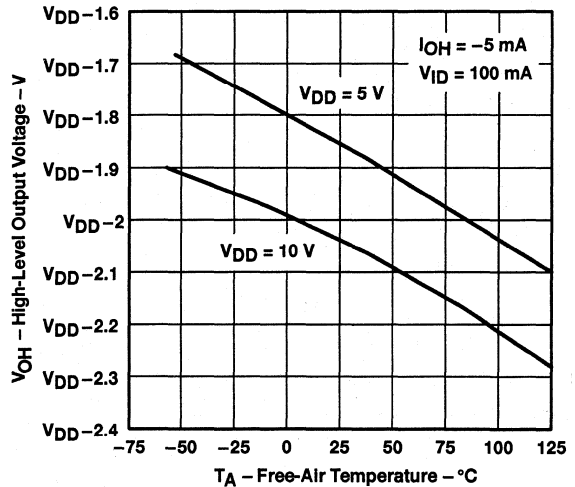


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

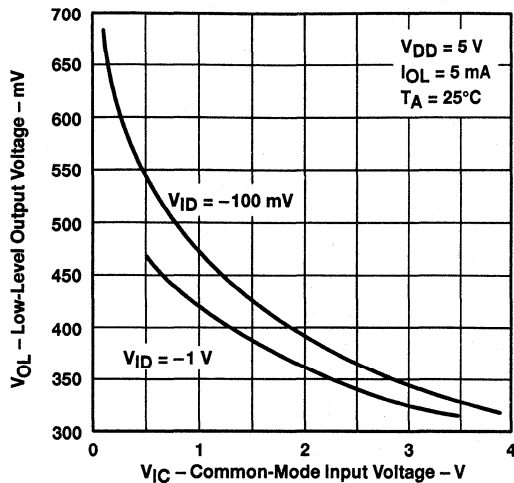


Figure 14

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

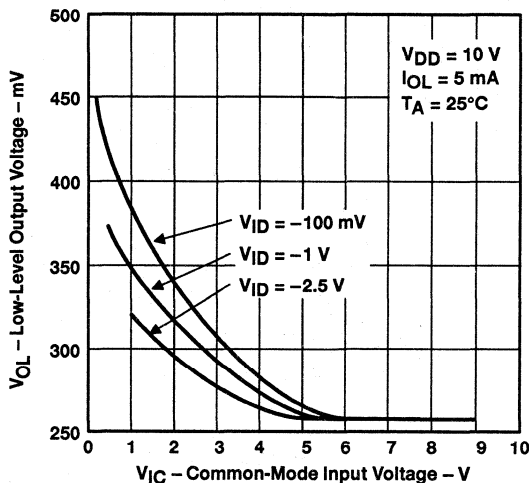


Figure 15

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE

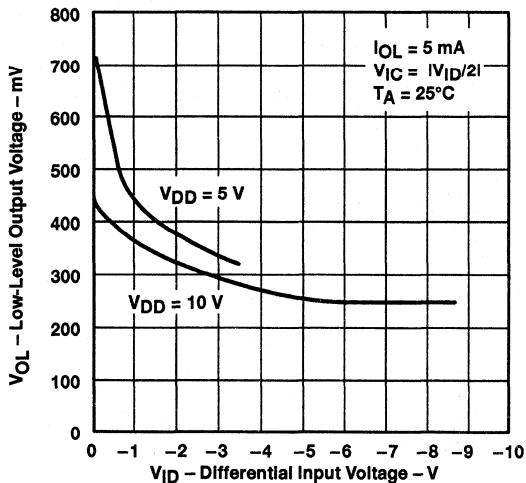


Figure 16

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

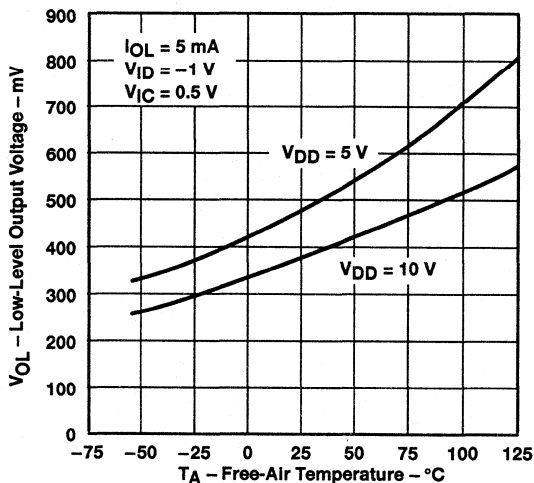


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

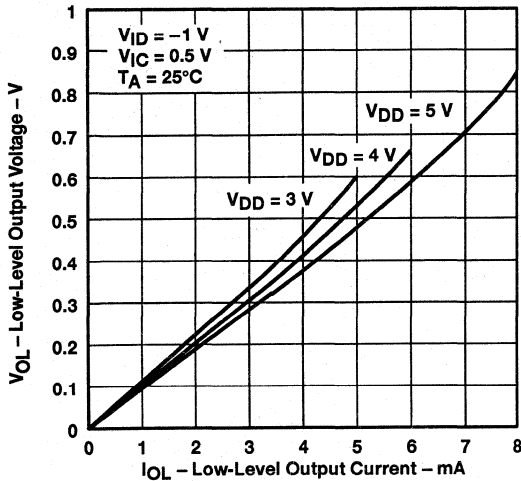


Figure 18

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

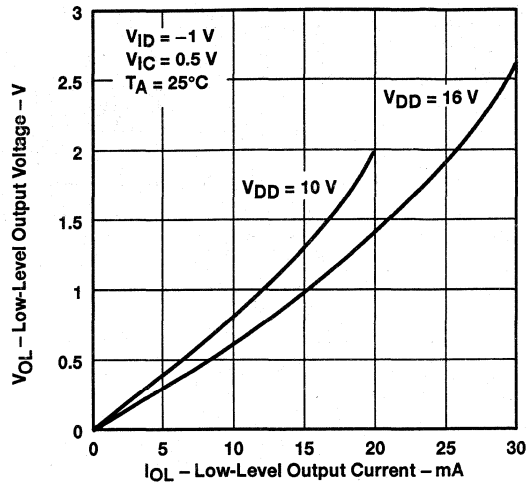


Figure 19

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
SUPPLY VOLTAGE

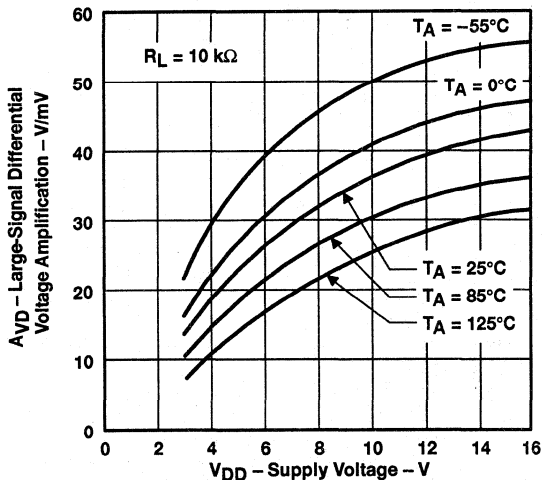


Figure 20

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
FREE-AIR TEMPERATURE

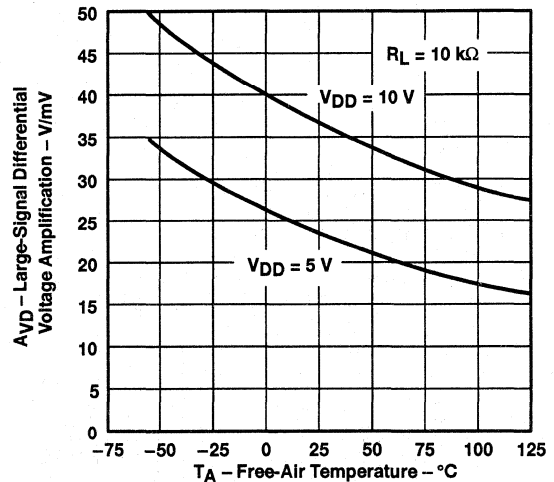


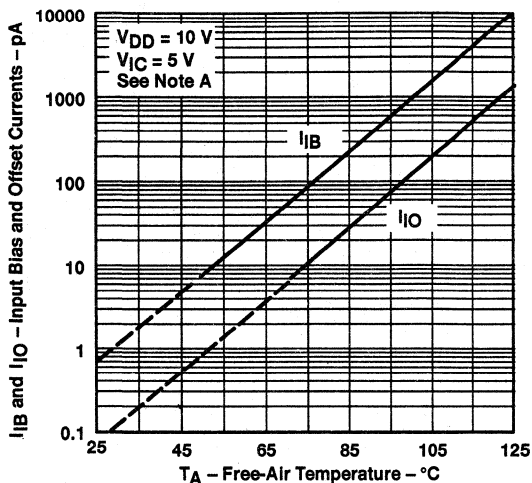
Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT  
 VS  
 FREE-AIR TEMPERATURE



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

COMMON-MODE  
 INPUT VOLTAGE POSITIVE LIMIT  
 VS  
 SUPPLY VOLTAGE

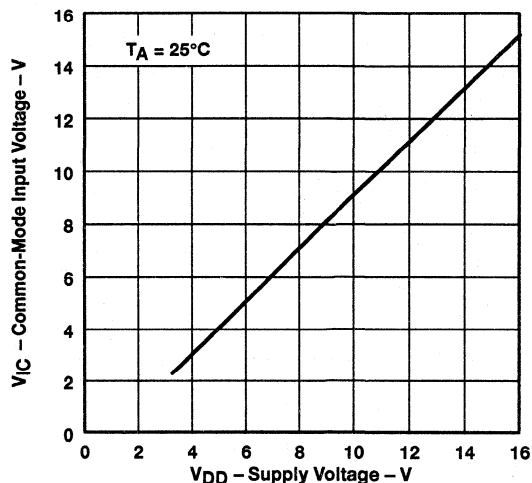


Figure 23

SUPPLY CURRENT  
 VS  
 SUPPLY VOLTAGE

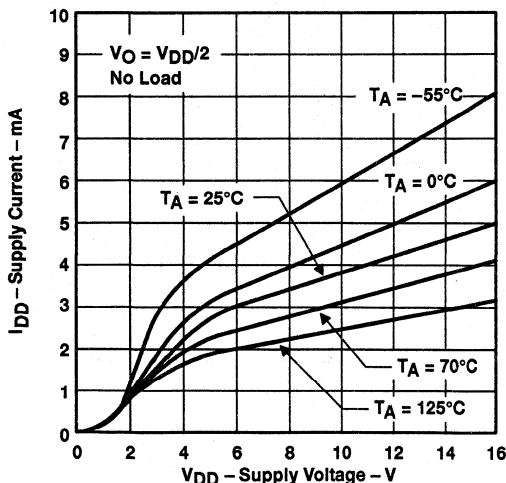


Figure 24

SUPPLY CURRENT  
 VS  
 FREE-AIR TEMPERATURE

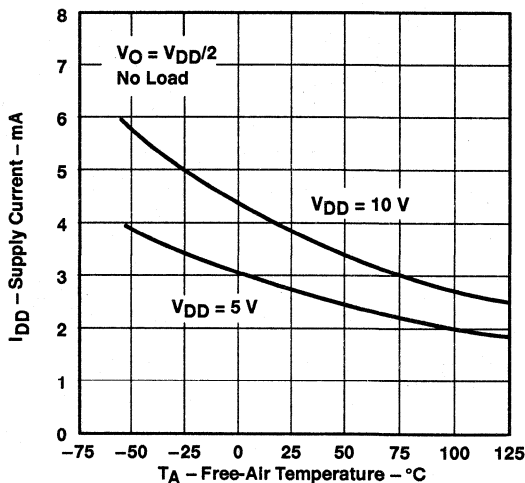


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

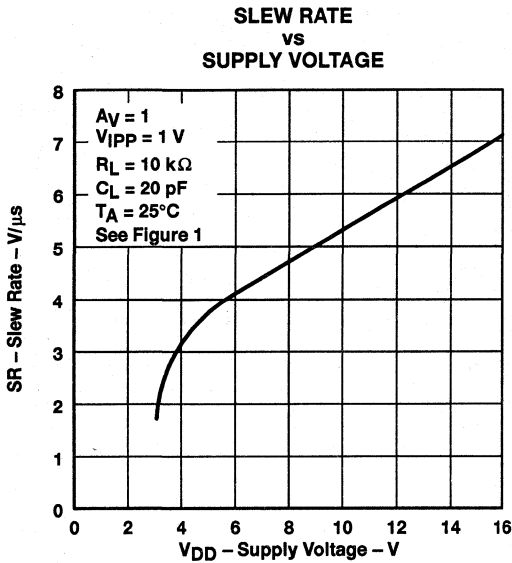


Figure 26

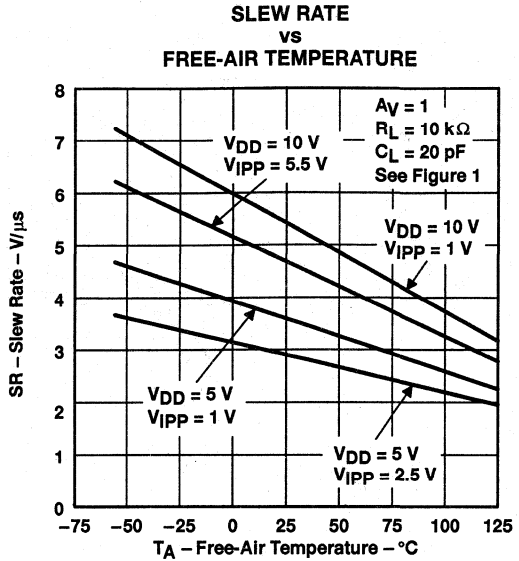


Figure 27

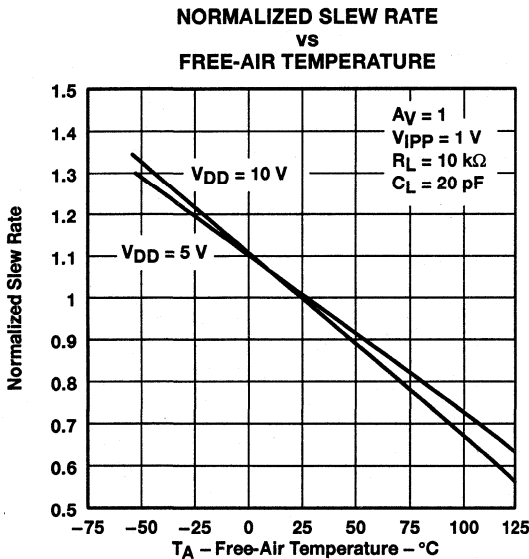


Figure 28

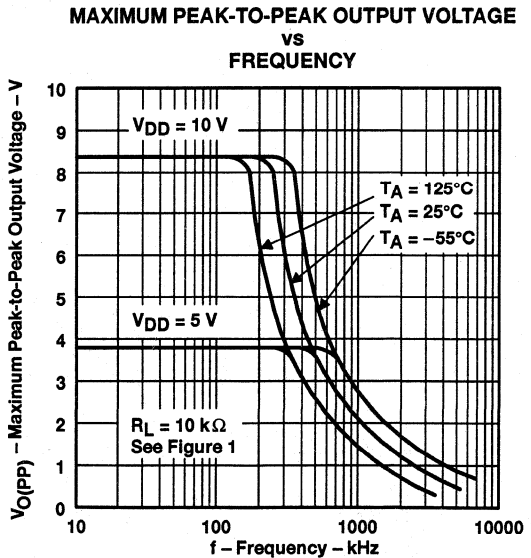


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

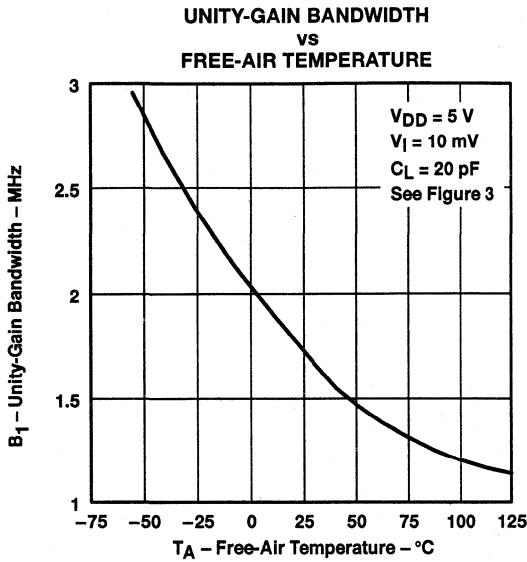


Figure 30

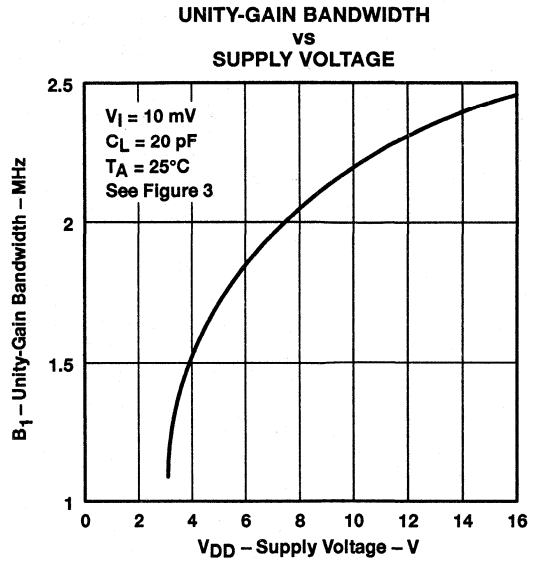


Figure 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY

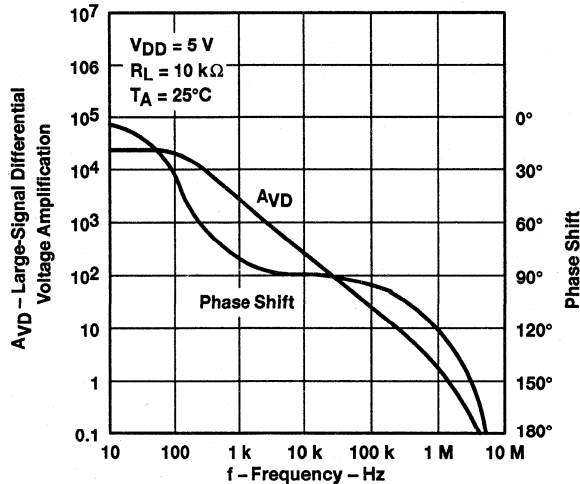


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC274, TLC274A, TLC274B, TLC274Y, TLC279**  
**LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS**

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY**

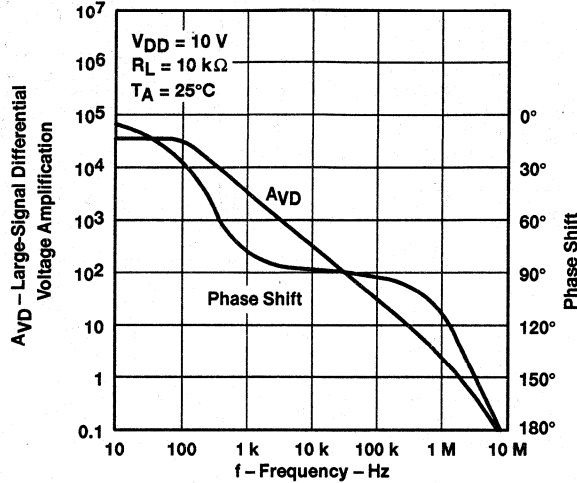


Figure 33

**PHASE MARGIN VS SUPPLY VOLTAGE**

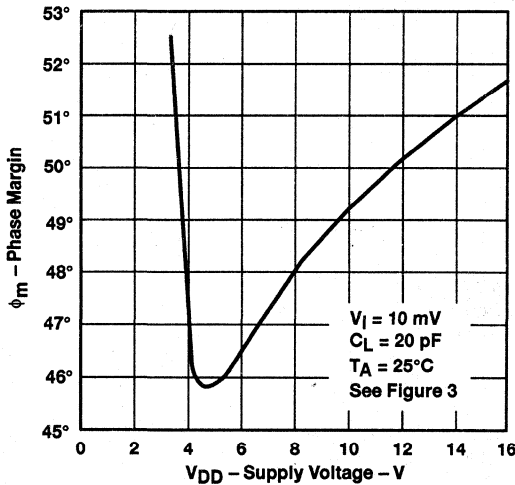


Figure 34

**PHASE MARGIN VS FREE-AIR TEMPERATURE**

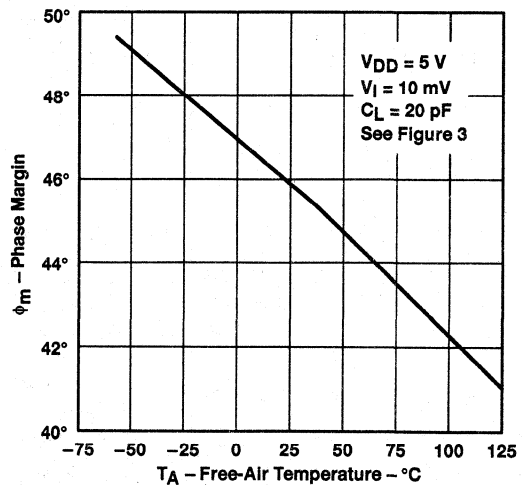


Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TLC274, TLC274A, TLC274B, TLC274Y, TLC279  
LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

TYPICAL CHARACTERISTICS

PHASE MARGIN  
VS  
CAPACITIVE LOAD

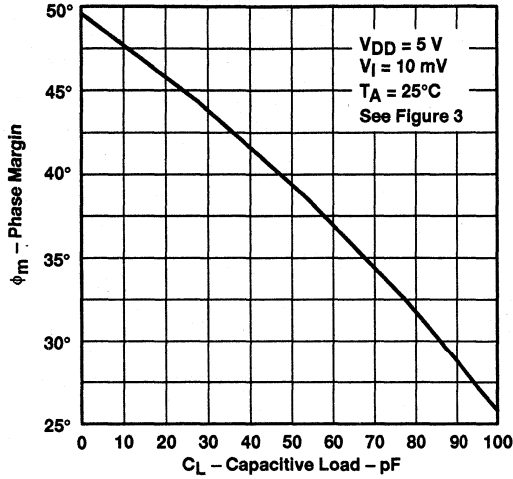


Figure 36

EQUIVALENT INPUT NOISE VOLTAGE  
VS  
FREQUENCY

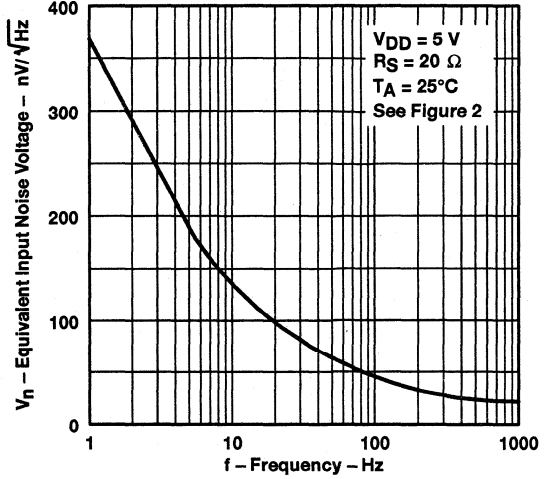


Figure 37

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### single-supply operation

While the TLC274 and TLC279 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC274 and TLC279 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC274 and TLC279 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require  $R_C$  decoupling.

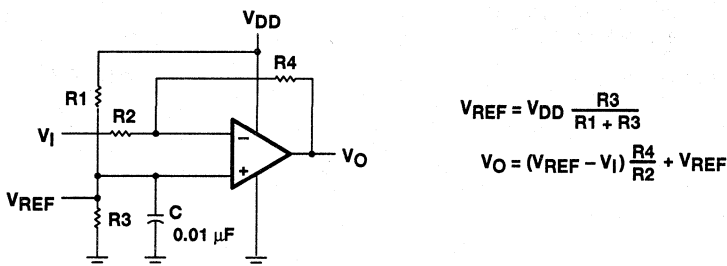


Figure 38. Inverting Amplifier With Voltage Reference

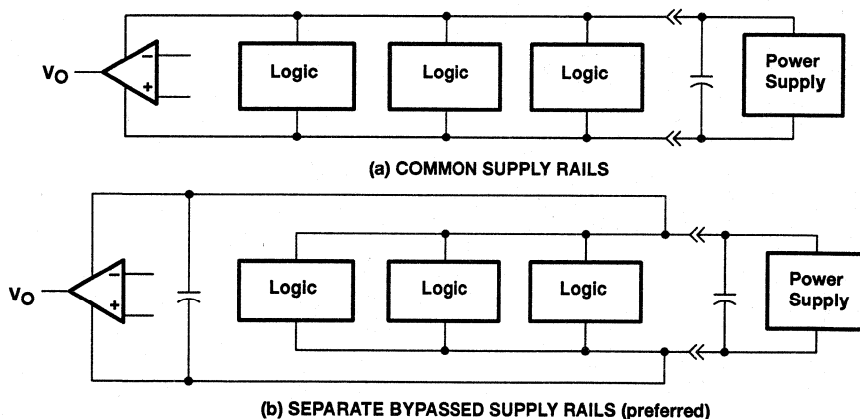


Figure 39. Common Versus Separate Supply Rails

APPLICATION INFORMATION

input characteristics

The TLC274 and TLC279 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1\text{ V}$  at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5\text{ V}$  at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC274 and TLC279 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1\ \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC274 and TLC279 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC274 and TLC279 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50\text{ k}\Omega$ , since bipolar devices exhibit greater noise currents.

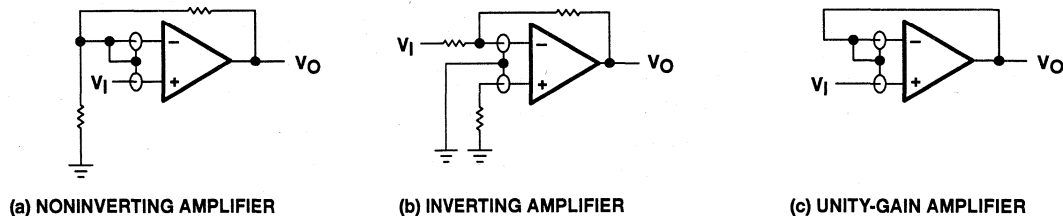


Figure 40. Guard-Ring Schemes

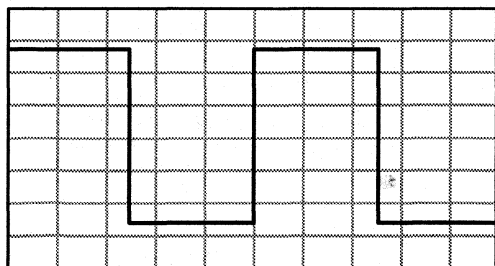
output characteristics

The output stage of the TLC274 and TLC279 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

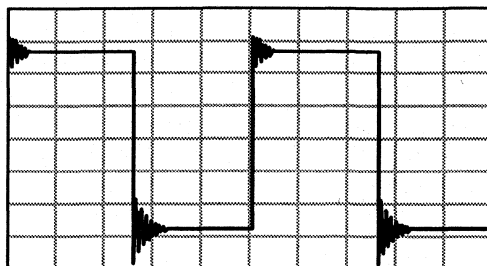
All operating characteristics of the TLC274 and TLC279 were measured using a  $20\text{-pF}$  load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

APPLICATION INFORMATION

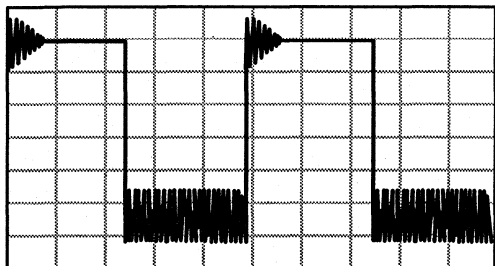
output characteristics (continued)



(a)  $C_L = 20 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(b)  $C_L = 130 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(c)  $C_L = 150 \text{ pF}$ ,  $R_L = \text{NO LOAD}$

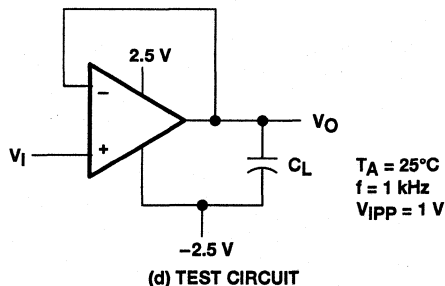


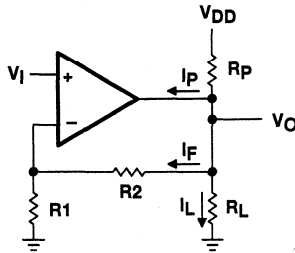
Figure 41. Effect of Capacitive Loads and Test Circuit

Although the TLC274 and TLC279 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60 \Omega$  and  $180 \Omega$ , depending on how hard the op amp input is driven. With very low values of  $R_P$ , a voltage offset from 0 V at the output occurs. Second, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.



APPLICATION INFORMATION

output characteristics (continued)



$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

$I_p$  = Pullup current required  
 by the operational amplifier  
 (typically 500  $\mu$ A)

Figure 42. Resistive Pullup to Increase  $V_{OH}$

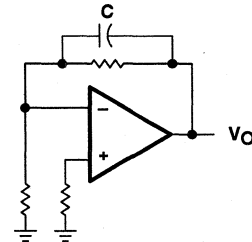


Figure 43. Compensation for  
 Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic discharge protection

The TLC274 and TLC279 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature-dependent and have the characteristics of a reverse-biased diode.

latch-up

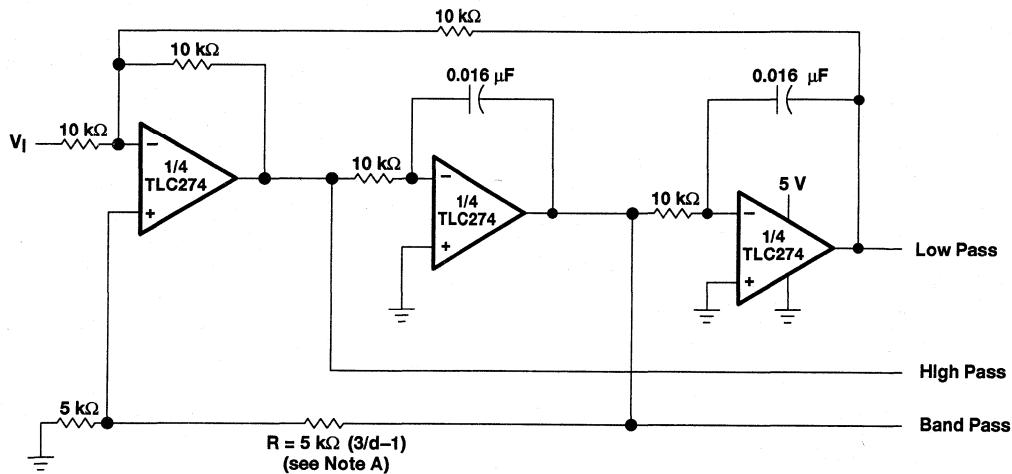
Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC274 and TLC279 inputs and outputs were designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1  $\mu$ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

# TLC274, TLC274A, TLC274B, TLC274Y, TLC279 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS092B – SEPTEMBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION



NOTE A:  $d$  = damping factor,  $1/Q$

Figure 44. State-Variable Filter

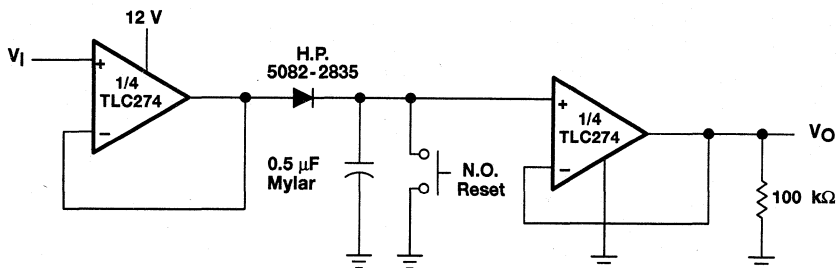
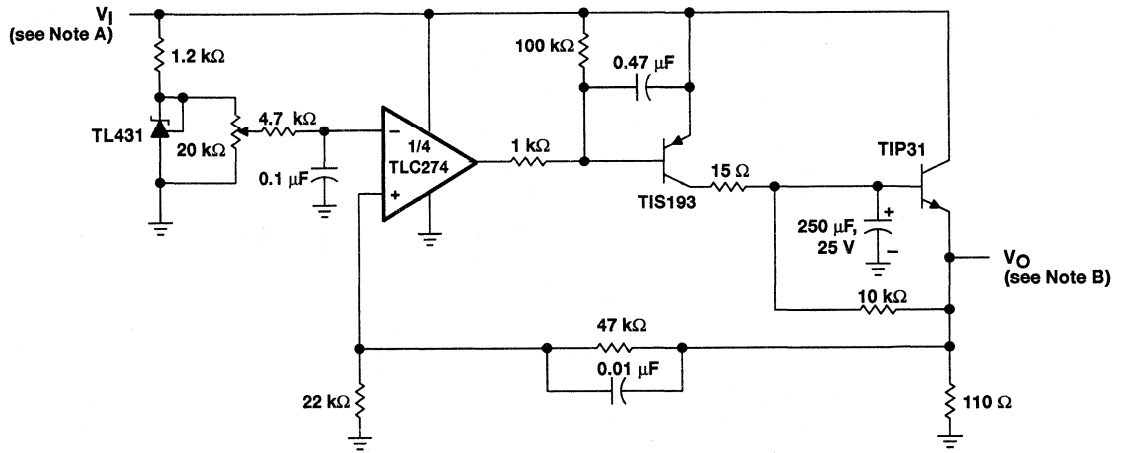


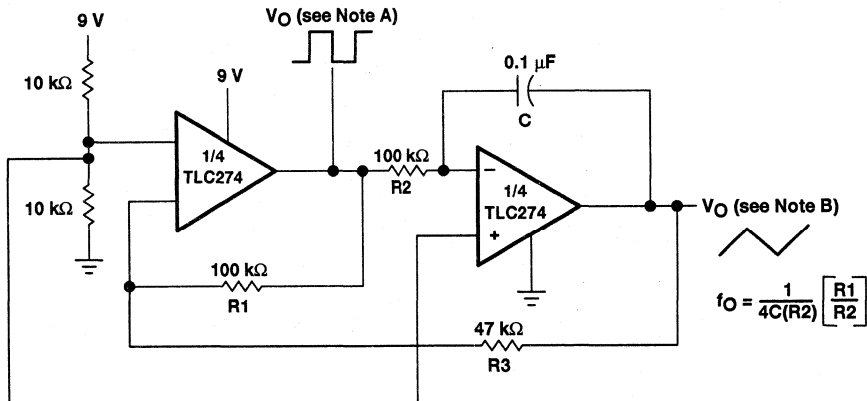
Figure 45. Positive-Peak Detector

APPLICATION INFORMATION



NOTES: B.  $V_I = 3.5 \text{ V to } 15 \text{ V}$   
 C.  $V_O = 2 \text{ V, } 0 \text{ to } 1 \text{ A}$

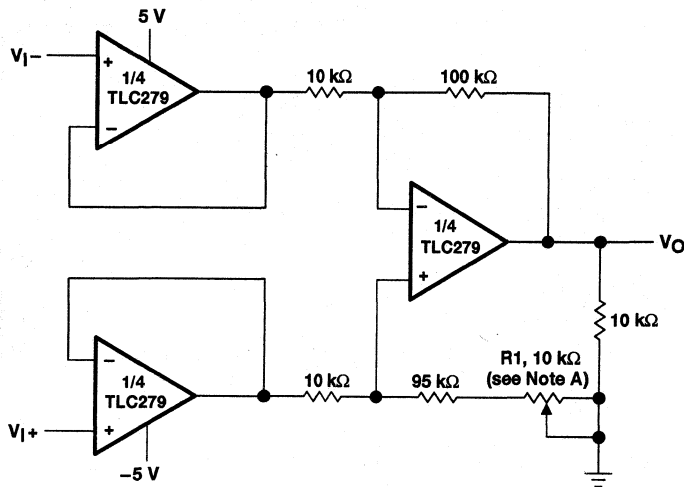
Figure 46. Logic-Array Power Supply



NOTES: A.  $V_{O(pp)} = 8 \text{ V}$   
 B.  $V_{O(pp)} = 4 \text{ V}$

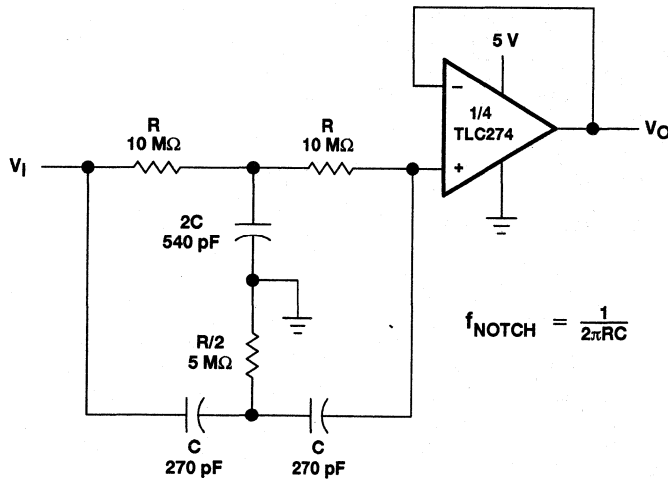
Figure 47. Single-Supply Function Generator

APPLICATION INFORMATION



NOTE A: CMRR adjustment must be noninductive.

Figure 48. Low-Power Instrumentation Amplifier



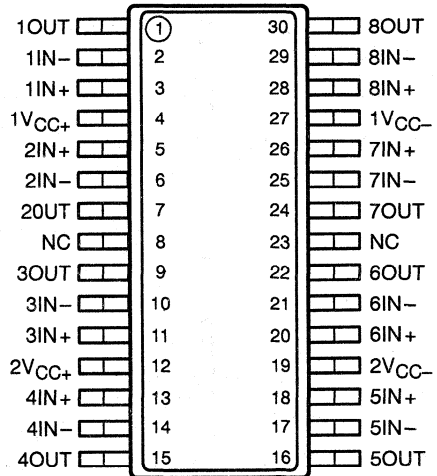
$$f_{\text{NOTCH}} = \frac{1}{2\pi RC}$$

Figure 49. Single-Supply Twin-T Notch Filter

**TLC274x2**  
**LinCMOS™ PRECISION**  
**OCTAL OPERATIONAL AMPLIFIER**  
 SLOS137 – JULY 1994

- **Trimmed Offset Voltage**  
10 mV Max at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages**  
3 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range**  
Extends Below the Negative Rail
- **Low Noise . . . Typically 25 nV/ $\sqrt{\text{Hz}}$**   
at  $f = 1\text{ kHz}$
- **Output Voltage Range Includes Negative Rail**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**

**DB PACKAGE**  
(TOP VIEW)

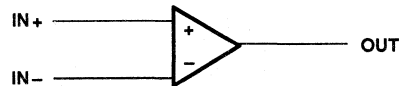


NC – No internal connection

**description**

The TLC274x2 octal operational amplifier incorporates low offset-voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose BIFET devices into a single package. This device uses Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

**symbol (each amplifier)**



The extremely high input impedance, low bias currents, and high slew rates make this a cost-effective device ideal for applications that have previously been reserved for BIFET and NFET products. These advantages, in combination with good common-mode rejection and supply-voltage rejection, make this device a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC274x2. The device also exhibits low-voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail. The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up.

**AVAILABLE OPTION**

$T_A$	$V_{IOmax}$ AT $25^\circ\text{C}$	PACKAGE
		SMALL OUTLINE (DB)†
$0^\circ\text{C}$ to $70^\circ\text{C}$	10 mV	TLC274x2DBLE

† The DB package is only available left-end taped and reeled.

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

**TLC274x2**  
**LinCMOS™ PRECISION**  
**OCTAL OPERATIONAL AMPLIFIER**

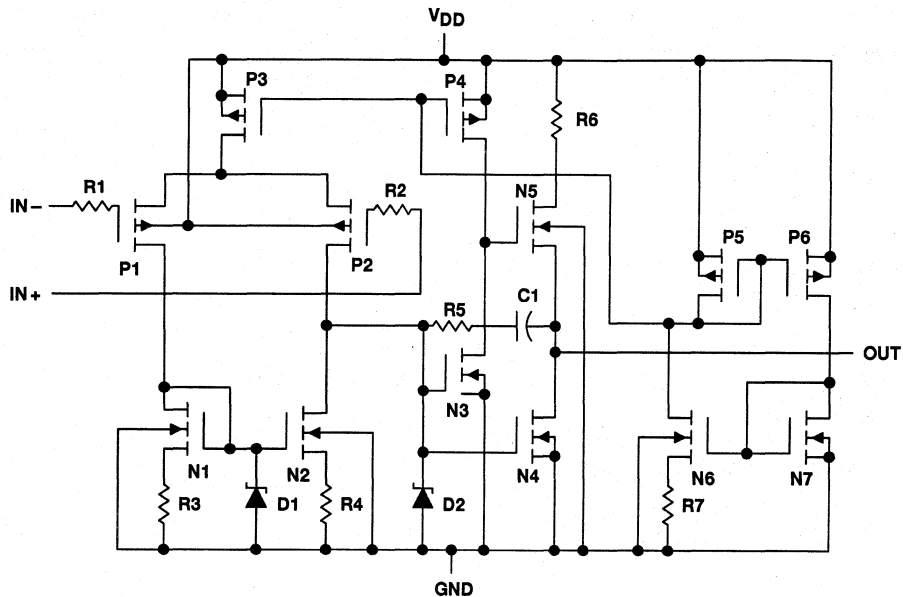
SLOS137 – JULY 1994

**description (continued)**

The TLC274x2 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, exercise care in handling this device as exposure to ESD can result in the degradation of the device parametric performance.

The TLC274x2 is characterized for operation from 0°C to 70°C.

**equivalent schematic (each amplifier)**



COMPONENT COUNT	
Resistors	56
Transistors	80
Diodes	16
Capacitors	8

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$V_{DD} \pm$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$	0°C to 70°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
3. The output can be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
DB	1024 mW	8.2 mW/°C	655 mW

**recommended operating conditions**

		MIN	MAX	UNIT
Supply voltage, $V_{DD}$		3	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	
Operating free-air temperature, $T_A$		0	70	°C

**TLC274x2**  
**LinCMOS™ PRECISION**  
**OCTAL OPERATIONAL AMPLIFIER**  
 SLOS137 – JULY 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A^\dagger$	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 10\text{ k}\Omega$	25°C	1.1	10		mV
				Full range			12	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C		1.8		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.1		pA
				70°C		7	300	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C		0.6		pA
				70°C		40	600	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2		V
				Full range	-0.2 to 3.5			V
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 10\text{ k}\Omega$	25°C	3.2	3.8		V
				0°C	3	3.8		
				70°C	3	3.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	5	23		V/mV
				0°C	4	27		
				70°C	4	20		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	80		dB
				0°C	60	84		
				70°C	60	85		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C		2.7	6.4	mA
				0°C		3.1	7.2	
				70°C		2.3	5.2	

$^\dagger$  Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
 5. This range also applies to each input individually.





**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A^\dagger$	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 10\ \text{k}\Omega$	25°C	1.1		10	mV
				Full range			12	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C		2		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current (see Note 4)	$V_O = 0.5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1			pA
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7			pA
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2		V
				Full range	-0.2 to 8.5			V
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 10\ \text{k}\Omega$	25°C	8	8.5		V
				0°C	7.8	8.5		
				70°C	7.8	8.4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C		0	50	mV
				0°C		0	50	
				70°C		0	50	
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 10\ \text{k}\Omega$	25°C	10	36		V/mV
				0°C	7.5	42		
				70°C	7.5	32		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	85		dB
				0°C	60	88		
				70°C	60	88		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	65	95		dB
				0°C	60	94		
				70°C	60	96		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	3.8	8		mA
				0°C	4.5	8.8		
				70°C	3.2	6.8		

$^\dagger$  Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

**TLC274x2**  
**LinCMOS™ PRECISION**  
**OCTAL OPERATIONAL AMPLIFIER**  
 SLOS137 – JULY 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER		TEST CONDITIONS		$T_A$	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{I(PP)} = 1\ \text{V}$	25°C		3.6		V/ $\mu\text{s}$
				0°C		4		
				70°C		3		
			$V_{I(PP)} = 2.5\ \text{V}$	25°C		2.9		
				0°C		3.1		
				70°C		2.5		
$V_n$	Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C		25		nV/ $\sqrt{\text{Hz}}$
B <sub>OM</sub>	Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\ \text{k}\Omega$ ,	$C_L = 20\ \text{pF}$ , See Figure 1	25°C		320		kHz
				0°C		340		
				70°C		260		
B <sub>1</sub>	Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$ ,	25°C		1.7		MHz
				0°C		2		
				70°C		1.3		
$\phi_m$	Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$ ,	$f = B_1$ ,	25°C		46°		
				0°C		47°		
				70°C		44°		

**operating characteristics at specified free-air temperature,  $V_{DD} = 10\ \text{V}$**

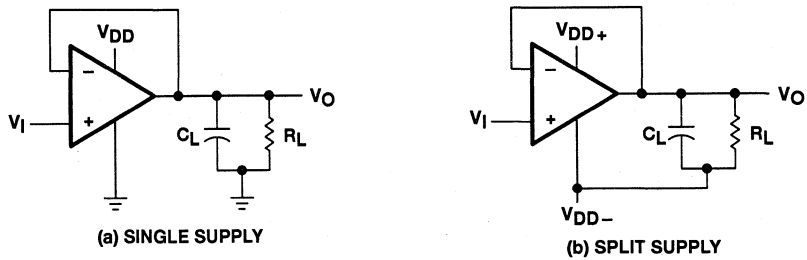
PARAMETER		TEST CONDITIONS		$T_A$	MIN	TYP	MAX	UNIT
SR	Slew rate at unity gain	$R_L = 10\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{I(PP)} = 1\ \text{V}$	25°C		5.3		V/ $\mu\text{s}$
				0°C		5.9		
				70°C		4.3		
			$V_{I(PP)} = 5.5\ \text{V}$	25°C		4.6		
				0°C		5.1		
				70°C		3.8		
$V_n$	Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C		25		nV/ $\sqrt{\text{Hz}}$
B <sub>OM</sub>	Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\ \text{k}\Omega$ ,	$C_L = 20\ \text{pF}$ , See Figure 1	25°C		200		kHz
				0°C		220		
				70°C		140		
B <sub>1</sub>	Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$ ,	25°C		2.2		MHz
				0°C		2.5		
				70°C		1.8		
$\phi_m$	Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$ ,	$f = B_1$ , See Figure 3	25°C		49°		
				0°C		50°		
				70°C		46°		



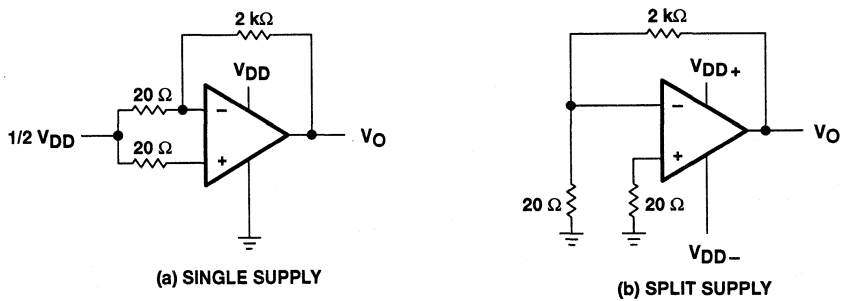
**PARAMETER MEASUREMENT INFORMATION**

**single-supply versus split-supply test circuits**

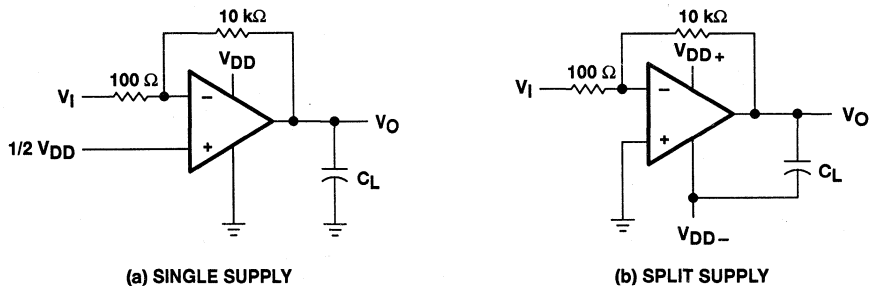
Because the TLC274x2 is optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.



**Figure 1. Unity-Gain Amplifier**



**Figure 2. Noise-Test Circuit**



**Figure 3. Gain-of-100 Inverting Amplifier**

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
V <sub>OH</sub>	High-level output voltage	vs High-level output current	4, 5
		vs Supply voltage	6
		vs Free-air temperature	7
V <sub>OL</sub>	Low-level output voltage	vs Common-mode input voltage	8, 9
		vs Differential input voltage	10
		vs Free-air temperature	11
		vs Low-level output current	12, 13
A <sub>VD</sub>	Large-signal differential voltage amplification	vs Supply voltage	14
		vs Free-air temperature	15
		vs Frequency	26, 27
I <sub>IB</sub>	Input bias current	vs Free-air temperature	16
I <sub>IO</sub>	Input offset current	vs Free-air temperature	16
V <sub>IC</sub>	Common-mode input voltage	vs Supply voltage	17
I <sub>DD</sub>	Supply current	vs Supply voltage	18
		vs Free-air temperature	19
SR	Slew rate	vs Supply voltage	20
		vs Free-air temperature	21
	Normalized slew rate	vs Free-air temperature	22
V <sub>O(PP)</sub>	Maximum peak-to-peak output voltage	vs Frequency	23
B <sub>1</sub>	Unity-gain bandwidth	vs Free-air temperature	24
		vs Supply voltage	25
φ <sub>m</sub>	Phase margin	vs Supply voltage	28
		vs Free-air temperature	29
		vs Load capacitance	30
V <sub>n</sub>	Equivalent input noise voltage	vs Frequency	31
		Phase shift	26, 27

**TYPICAL CHARACTERISTICS**

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

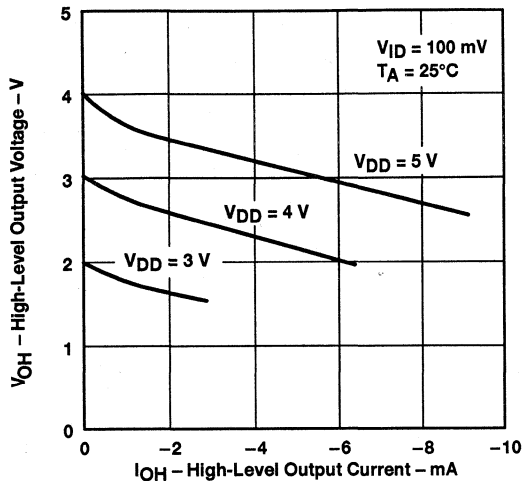


Figure 4

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT**

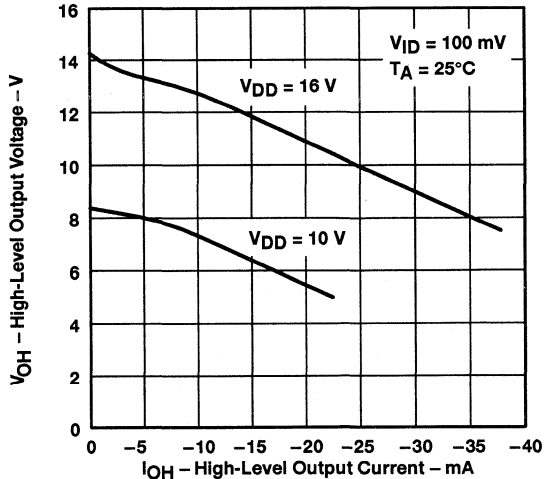


Figure 5

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE**

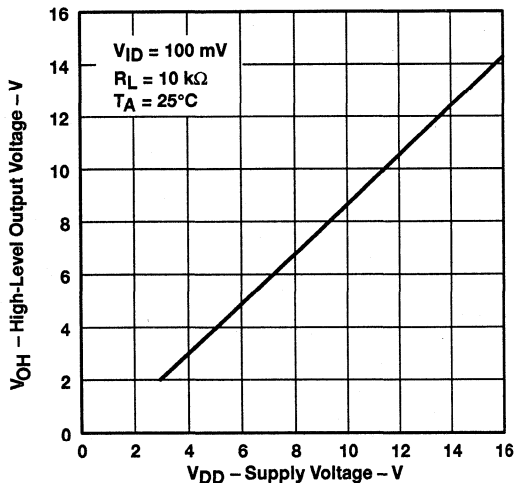


Figure 6

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

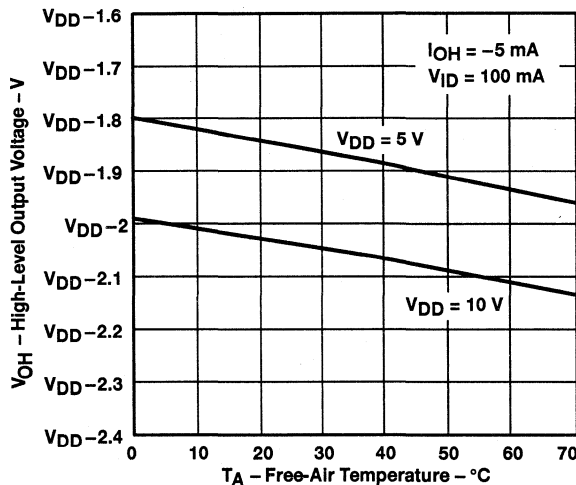


Figure 7

TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

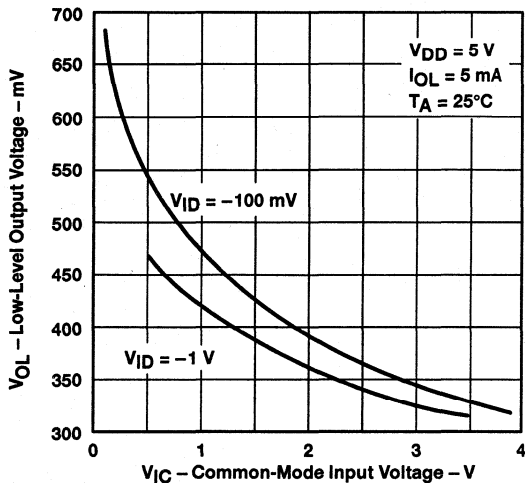


Figure 8

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

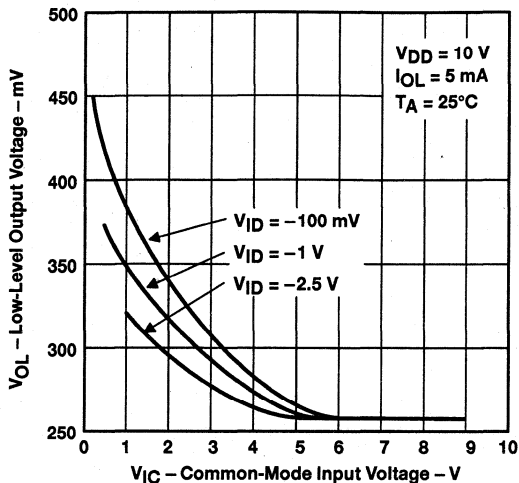


Figure 9

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 DIFFERENTIAL INPUT VOLTAGE

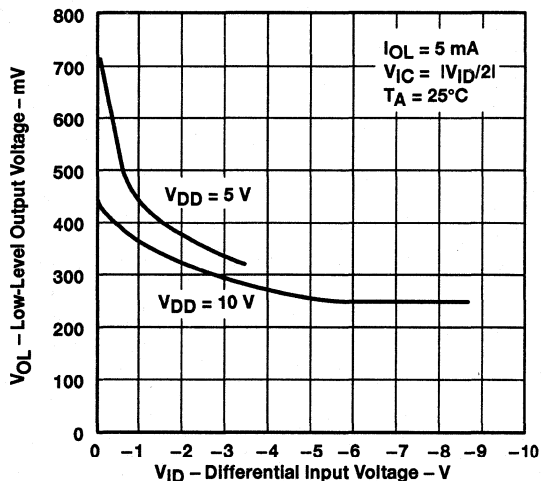


Figure 10

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

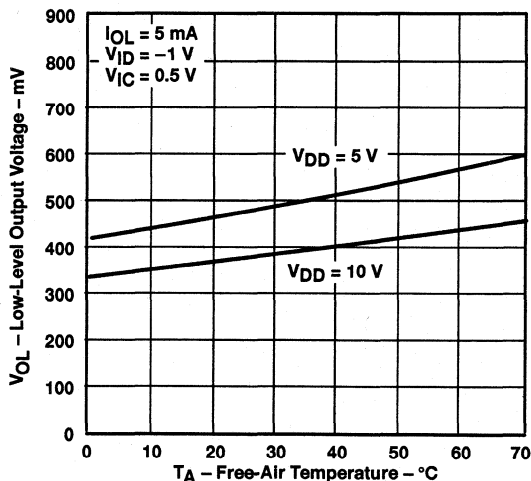


Figure 11

TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

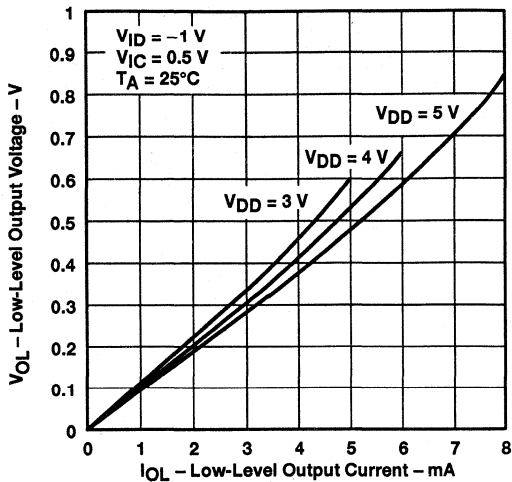


Figure 12

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

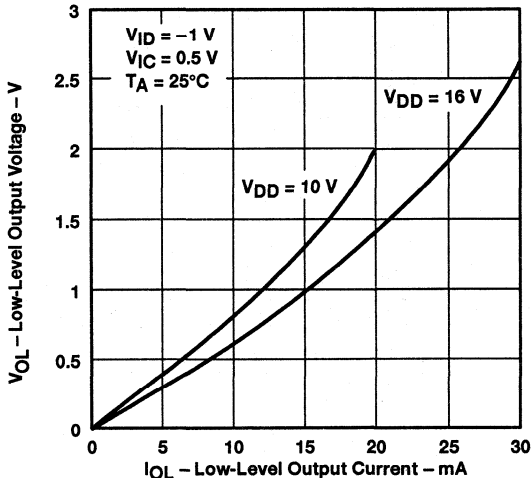


Figure 13

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 SUPPLY VOLTAGE

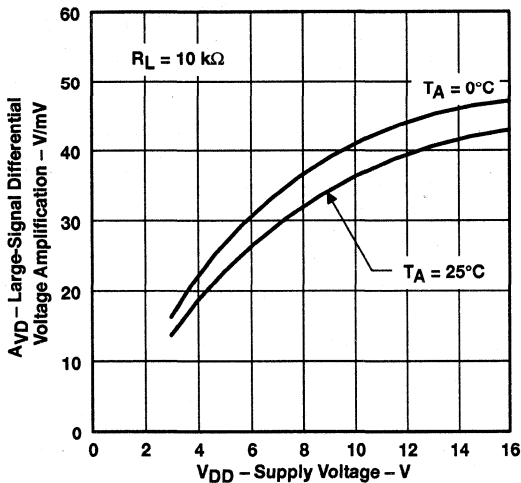


Figure 14

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

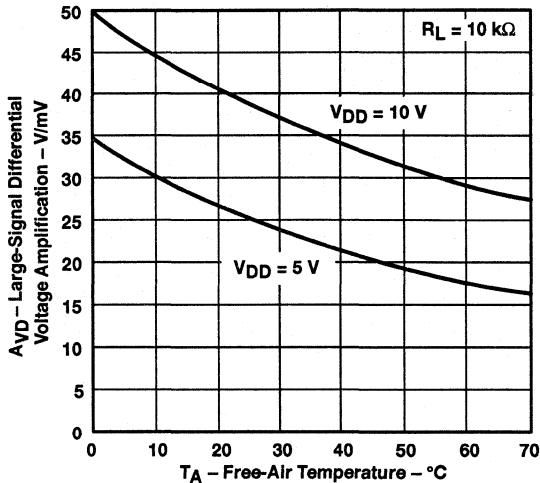
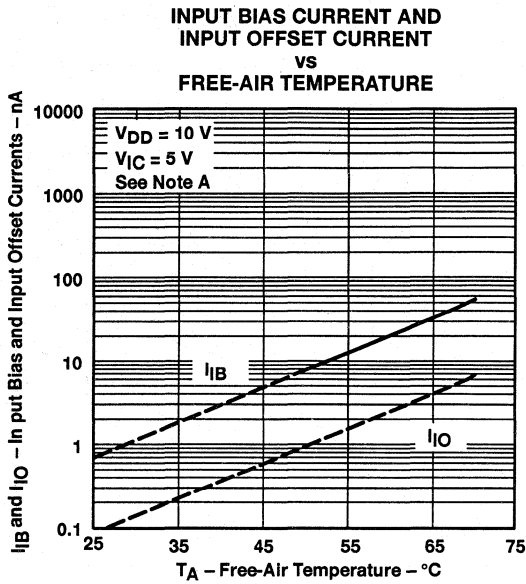


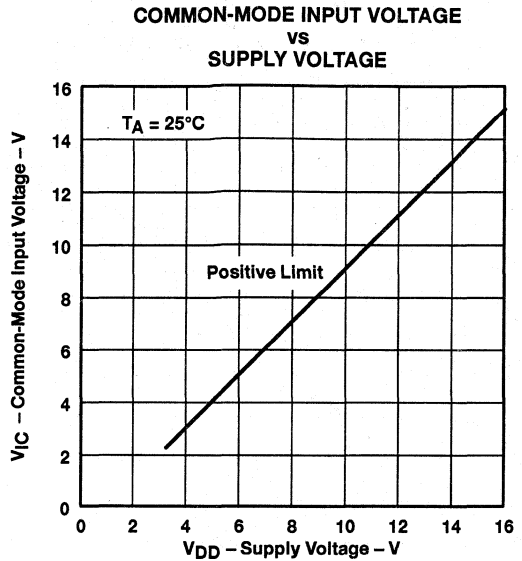
Figure 15

**TYPICAL CHARACTERISTICS**

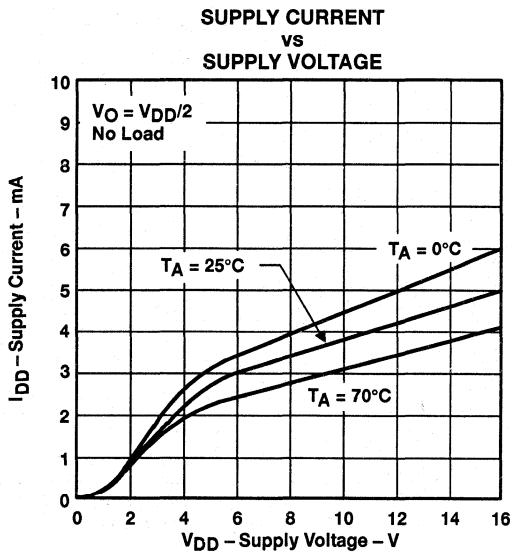


NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

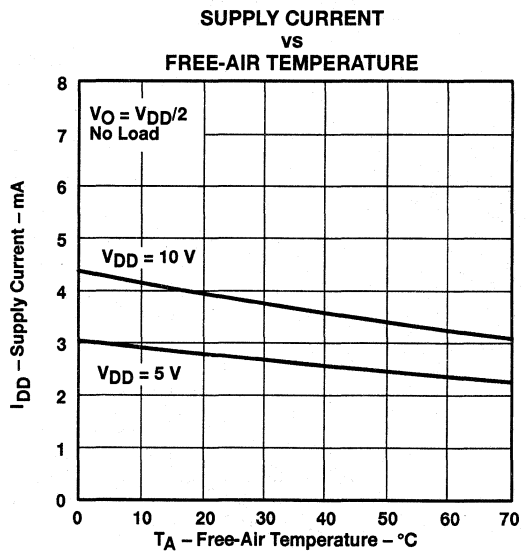
**Figure 16**



**Figure 17**



**Figure 18**



**Figure 19**



TYPICAL CHARACTERISTICS

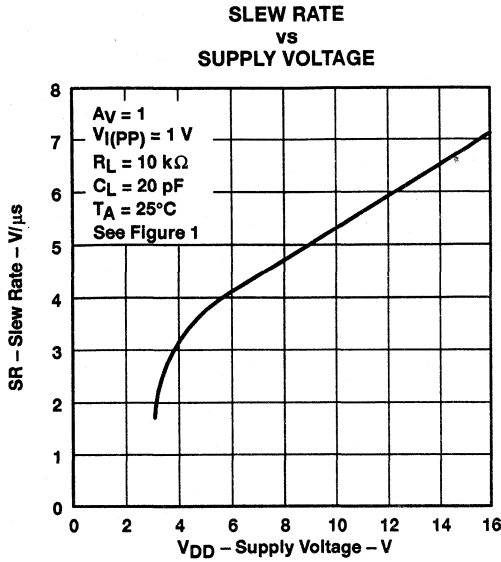


Figure 20

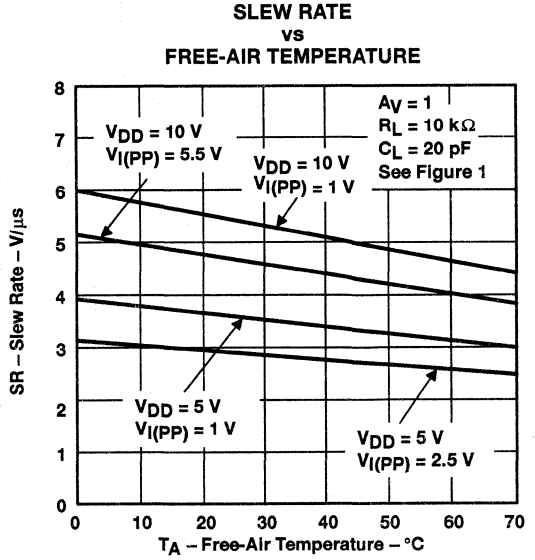


Figure 21

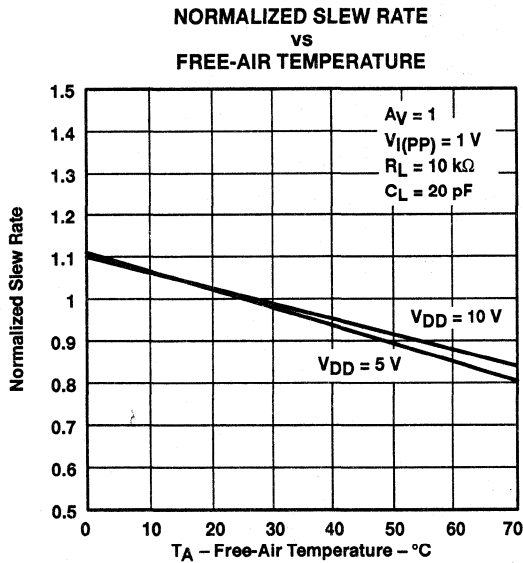


Figure 22

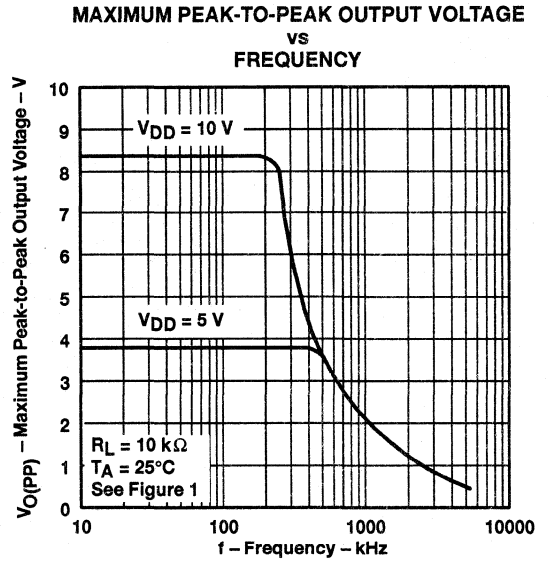
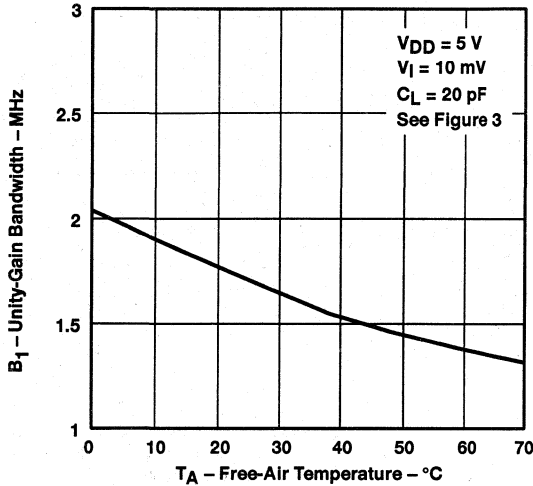


Figure 23

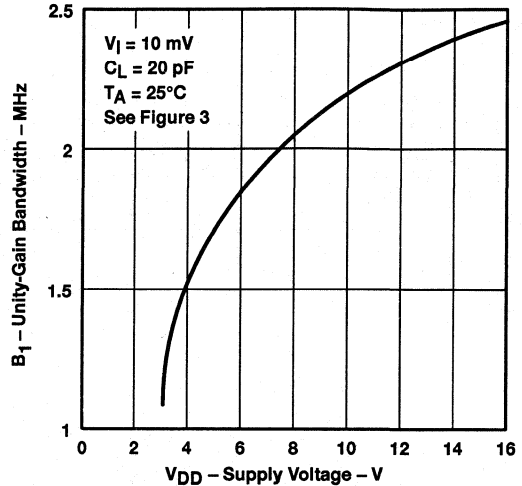
**TYPICAL CHARACTERISTICS**

**UNITY-GAIN BANDWIDTH  
 VS  
 FREE-AIR TEMPERATURE**



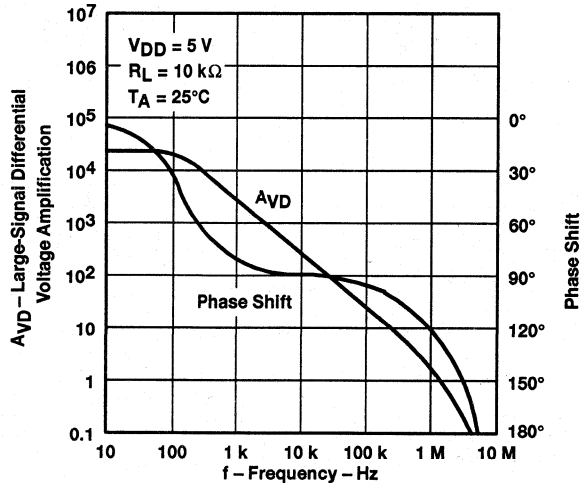
**Figure 24**

**UNITY-GAIN BANDWIDTH  
 VS  
 SUPPLY VOLTAGE**



**Figure 25**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY**



**Figure 26**

TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY

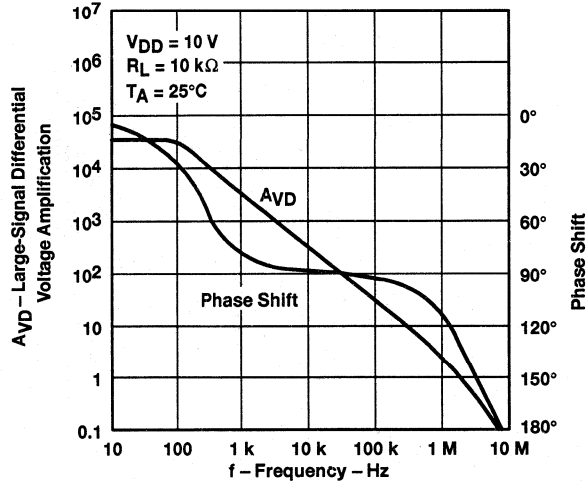


Figure 27

PHASE MARGIN  
 VS  
 SUPPLY VOLTAGE

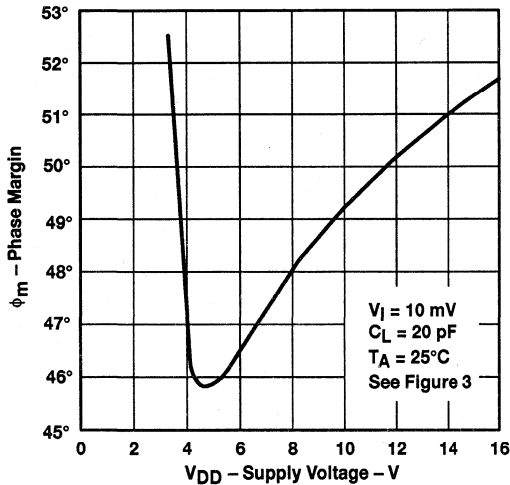


Figure 28

PHASE MARGIN  
 VS  
 FREE-AIR TEMPERATURE

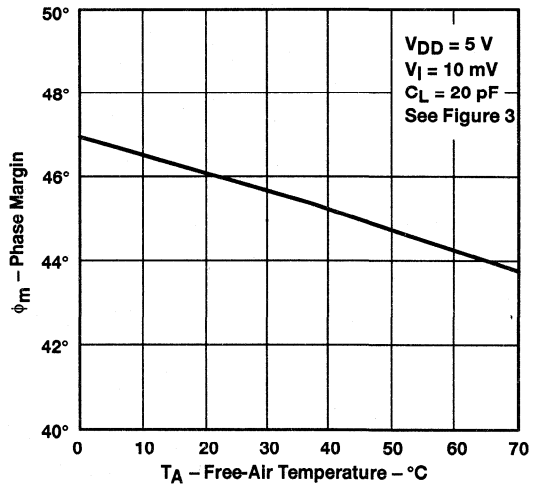


Figure 29

TYPICAL CHARACTERISTICS

PHASE MARGIN  
 vs  
 CAPACITIVE LOAD

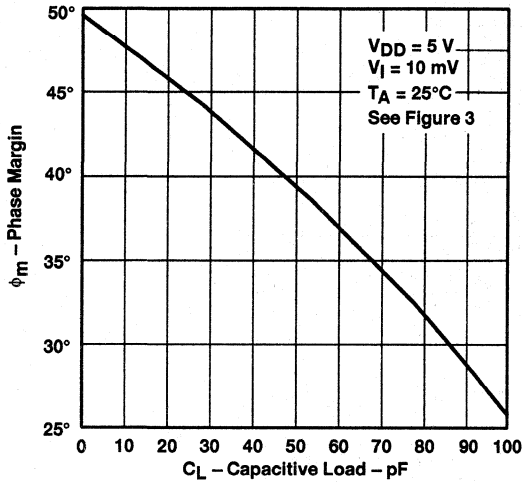


Figure 30

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

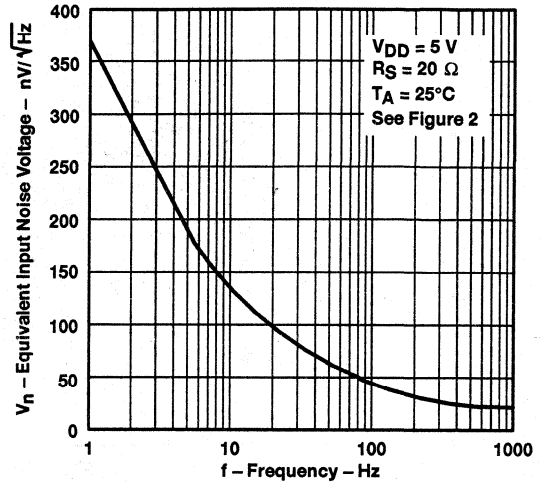


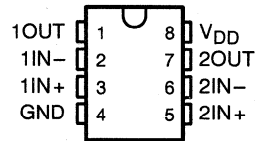
Figure 31

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

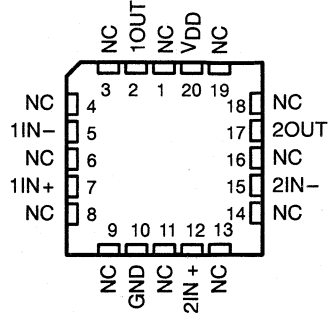
SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC27L7 . . . 500  $\mu\text{V}$  Max at 25°C,  
 $V_{\text{DD}} = 5 \text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
0°C to 70°C . . . 3 V to 16 V  
-40°C to 85°C . . . 4 V to 16 V  
-55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)**
- **Ultra-Low Power . . . Typically 95  $\mu\text{W}$  at 25°C,  $V_{\text{DD}} = 5 \text{ V}$**
- **Output Voltage Range includes Negative Rail**
- **High Input Impedance . . .  $10^{12} \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up immunity**

D, JG, OR P PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

## description

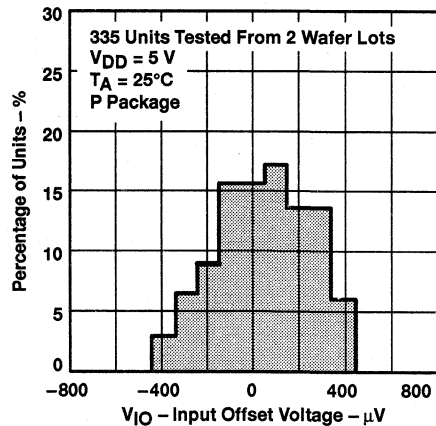
The TLC27L2 and TLC27L7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, extremely low power, and high gain.

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	500 $\mu\text{V}$ 2 mV 5 mV 10 mV	TLC27L7CD TLC27L2BCD TLC27L2ACD TLC27L2CD	—	—	TLC27L7CP TLC27L2BCP TLC27L2ACP TLC27L2CP
-40°C to 85°C	500 $\mu\text{V}$ 2 mV 5 mV 10 mV	TLC27L7ID TLC27L2BID TLC27L2AID TLC27L2ID	—	—	TLC27L7IP TLC27L2BIP TLC27L2AIP TLC27L2IP
-55°C to 125°C	500 $\mu\text{V}$ 10 mV	TLC27L7MD TLC27L2MD	TLC27L7MFK TLC27L2MFK	TLC27L7MJG TLC27L2MJG	TLC27L7MP TLC27L2MP

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC27L7CDR).

DISTRIBUTION OF TLC27L7  
INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

TEXAS  
INSTRUMENTS

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## description (continued)

These devices use Texas instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and low power consumption make these cost-effective devices ideal for high gain, low frequency, low power applications. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27L2 (10 mV) to the high-precision TLC27L7 (500  $\mu$ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available in LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27L2 and TLC27L7. The devices also exhibit low voltage single-supply operation and ultra-low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

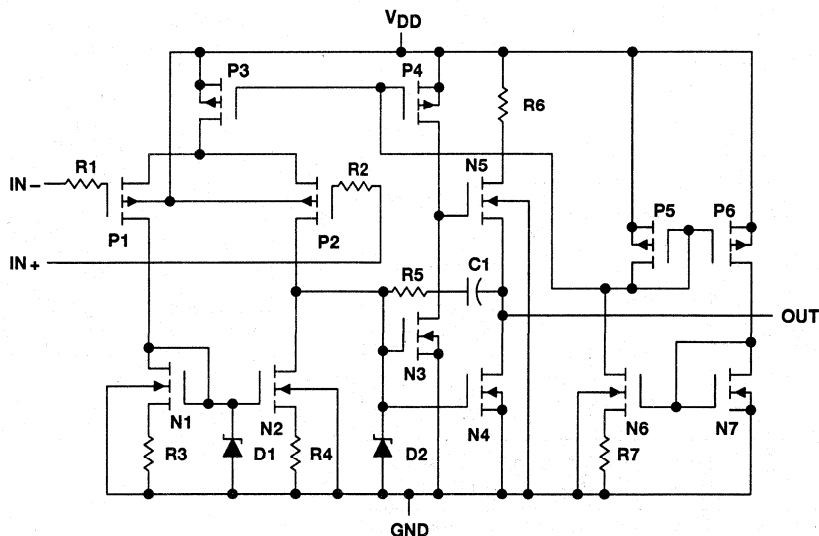
A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand  $\pm 100$ -mA surge currents without sustaining latch-up.

The TLC27L2 and TLC27L7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-Suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from  $-40^{\circ}$ C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of  $-55^{\circ}$ C to 125°C.

## equivalent schematic (each amplifier)



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PROCISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	—

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2C TLC27L2AC TLC27L2BC TLC27L7C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L2AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC27L2BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	204	2000	$\mu\text{V}$
					Full range		3000	
		TLC27L7C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	170	500	$\mu\text{V}$
					Full range		1500	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				0°C	3	4.1		
				70°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ ,	$R_L = 1\text{ M}\Omega$	25°C	50	700	V/mV	
				0°C	50	700		
				70°C	50	380		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	95		
				70°C	60	95		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	70	97	dB	
				0°C	60	97		
				70°C	60	98		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	20	34	$\mu\text{A}$	
				0°C	24	42		
				70°C	16	28		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2C TLC27L2AC TLC27L2BC TLC27L7C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L2AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
TLC27L2BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	235	2000	$\mu\text{V}$		
			Full range		3000			
TLC27L7C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	190	800	$\mu\text{V}$		
			Full range		1900			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	8	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				0°C	7.8	8.9		
				70°C	7.8	8.9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	860	V/mV	
				0°C	50	1025		
				70°C	50	660		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	97	dB	
				0°C	60	97		
				70°C	60	97		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				0°C	60	97		
				70°C	60	98		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	29	46	$\mu\text{A}$	
				0°C	36	66		
				70°C	22	40		

† Full range is 0°C to 70°C.

NOTES: 4 The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5 This range also applies to each input individually.



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 linCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2I TLC27L2AI TLC27L2BI TLC27L7I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27L2AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		7	
	TLC27L2BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	240	2000	$\mu\text{V}$	
				Full range		3500		
	TLC27L7I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	170	500		
				Full range		2000		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				85°C	24	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				85°C	200	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				-40°C	3	4.1		
				85°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	480	V/mV	
				-40°C	50	900		
				85°C	50	330		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-40°C	60	95		
				85°C	60	95		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-40°C	60	97		
				85°C	60	98		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	20	34	$\mu\text{A}$	
				-40°C	31	54		
				85°C	15	26		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2I TLC27L2AI TLC27L2BI TLC27L7I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1 10		mV
					Full range	13		
		TLC27L2AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9 5		
					Full range	7		
	TLC27L2BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	235	2000	$\mu\text{V}$	
				Full range	3500			
	TLC27L7I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	190	800		
				Full range	2900			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)		$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
$I_{IB}$	Input bias current (see Note 4)		$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage		$V_{ID} = 100\text{ mV}$ , $R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				-40°C	7.8	8.9		
				85°C	7.8	8.9		
$V_{OL}$	Low-level output voltage		$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification		$V_O = 1\text{ V to }6\text{ V}$ , $R_L = 1\text{ M}\Omega$	25°C	50	860	V/mV	
				-40°C	50	1550		
				85°C	50	585		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICRmin}$	25°C	65	97	dB	
				-40°C	60	97		
				85°C	60	98		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )		$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-40°C	60	97		
				85°C	60	98		
$I_{DD}$	Supply current (two amplifiers)		$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	29	46	$\mu\text{A}$
					-40°C	49	86	
					85°C	20	36	

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2M TLC27L7M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L7M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	170	500	$\mu\text{V}$
					Full range		3750	
$\alpha_{IO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				-55°C	3	4.1		
				125°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	500	V/mV	
				-55°C	25	1000		
				125°C	25	200		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-55°C	60	95		
				125°C	60	85		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-55°C	60	97		
				125°C	60	98		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	20	34	$\mu\text{A}$	
				-55°C	35	60		
				125°C	14	24		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PROCISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L2M TLC27L7M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L2M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
	TLC27L7M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	190	800	$\mu\text{V}$	
				Full range		4300		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)		$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA
					125°C	1.8	15	nA
$I_{IB}$	Input bias current (see Note 4)		$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA
					125°C	10	35	nA
$V_{ICR}$	Common-mode input voltage range (see Note 5)				25°C	0 to 9	-0.3 to 9.2	V
					Full range	0 to 8.5		V
$V_{OH}$	High-level output voltage		$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V
					-55°C	7.8	8.8	
					125°C	7.8	9	
$V_{OL}$	Low-level output voltage		$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV
					-55°C	0	50	
					125°C	0	50	
$A_{VD}$	Large-signal differential voltage amplification		$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	860	V/mV
					-55°C	25	1750	
					125°C	25	380	
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICRmin}$		25°C	65	97	dB
					-55°C	60	97	
					125°C	60	91	
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )		$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB
					-55°C	60	97	
					125°C	60	98	
$I_{DD}$	Supply current (two amplifiers)		$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	29	46	$\mu\text{A}$
					-55°C	56	96	
					125°C	18	30	

† Full range is -55 °C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2C TLC27L2AC TLC27L2BC TLC27L7C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.03		V/ $\mu\text{s}$	
			0°C	0.04			
			70°C	0.03			
		$V_I(\text{PP}) = 2.5\text{ V}$	25°C	0.03			
			0°C	0.03			
			70°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	5		kHz	
			0°C	6			
			70°C	4.5			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	85		kHz	
			0°C	100			
			70°C	65			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	34°			
			0°C	36°			
			70°C	30°			

## operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2C TLC27L2AC TLC27L2BC TLC27L7C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.05		V/ $\mu\text{s}$	
			0°C	0.05			
			70°C	0.04			
		$V_I(\text{PP}) = 5.5\text{ V}$	25°C	0.04			
			0°C	0.05			
			70°C	0.04			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	1		kHz	
			0°C	1.3			
			70°C	0.9			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	110		kHz	
			0°C	125			
			70°C	90			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	38°			
			0°C	40°			
			70°C	34°			



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PROCISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2I TLC27L2AI TLC27L2BI TLC27L7I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.03		V/ $\mu\text{s}$	
			-40°C	0.04			
			85°C	0.03			
		$V_I(\text{PP}) = 2.5\text{ V}$	25°C	0.03			
			-40°C	0.04			
			85°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	5		kHz	
			-40°C	7			
			85°C	4			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	85		kHz	
			-40°C	130			
			85°C	55			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	34°			
			-40°C	38°			
			85°C	29°			

## operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2I TLC27L2AI TLC27L2BI TLC27L7I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.05		V/ $\mu\text{s}$	
			-40°C	0.06			
			85°C	0.03			
		$V_I(\text{PP}) = 5.5\text{ V}$	25°C	0.04			
			-40°C	0.05			
			85°C	0.03			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	1		kHz	
			-40°C	1.4			
			85°C	0.8			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	110		kHz	
			-40°C	155			
			85°C	80			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	38°			
			-40°C	42°			
			85°C	32°			



# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2M TLC27L7M			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.03		V/ $\mu$ s	
			-55°C	0.04			
			125°C	0.02			
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.03			
			-55°C	0.04			
			125°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ ,	$C_L = 20\text{ pF}$ , See Figure 1	25°C	5		kHz	
			-55°C	8			
			125°C	3			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	85		kHz	
			-55°C	140			
			125°C	45			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	34°			
			-55°C	39°			
			125°C	25°			

## operating characteristics, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L2M TLC27L7M			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.05		V/ $\mu$ s	
			-55°C	0.06			
			125°C	0.03			
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.04			
			-55°C	0.06			
			125°C	0.03			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	68		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ ,	$C_L = 20\text{ pF}$ , See Figure 1	25°C	1		kHz	
			-55°C	1.5			
			125°C	0.7			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	110		kHz	
			-55°C	165			
			125°C	70			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	38°			
			-55°C	43°			
			125°C	29°			





PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27L2 and TLC27L7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

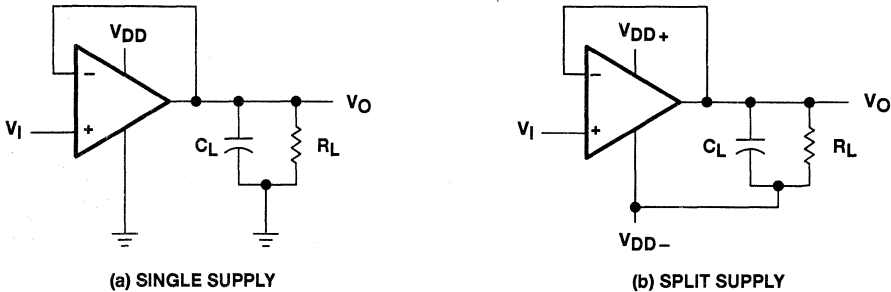


Figure 1. Unity-Gain Amplifier

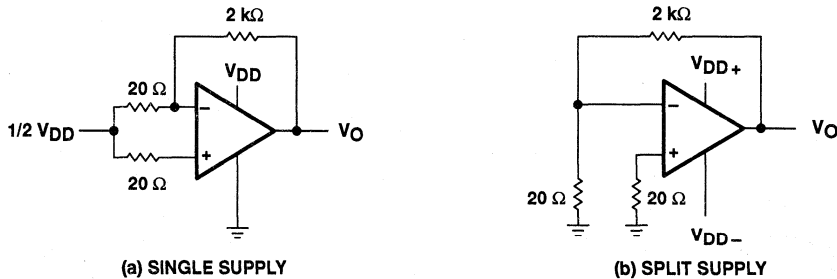


Figure 2. Noise-Test Circuit

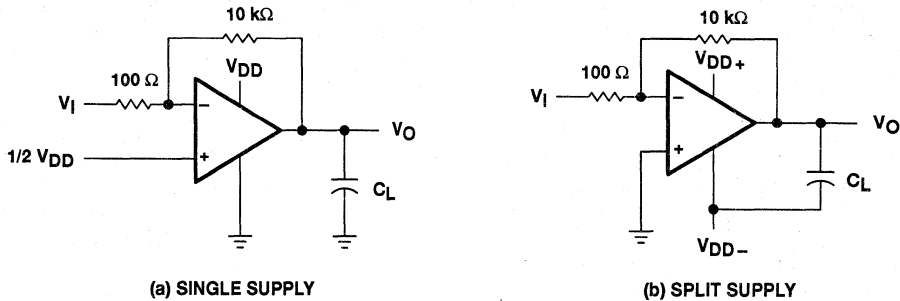


Figure 3. Gain-of-100 Inverting Amplifier

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC27L2 and TLC27L7 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

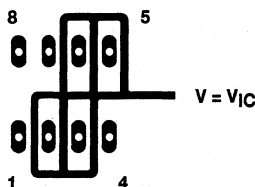


Figure 4. Isolation Metal Around Device Inputs  
(JG and P packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

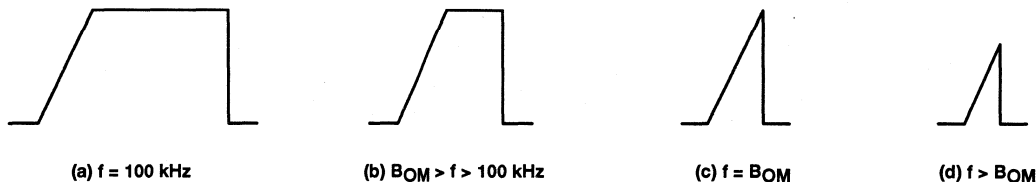


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

**TLC27L2, TLC27L2A, TLC27L2B, TLC27L7**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current	10, 11
		vs Supply voltage	12
		vs Free-air temperature	13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	14, 15
		vs Differential input voltage	16
		vs Free-air temperature	17
		vs Low-level output current	18, 19
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	20
		vs Free-air temperature	21
		vs Frequency	32, 33
$I_{IB}$	Input bias current	vs Free-air temperature	22
$I_{IO}$	Input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage	24
		vs Free-air temperature	25
SR	Slew rate	vs Supply voltage	26
		vs Free-air temperature	27
	Normalized slew rate	vs Free-air temperature	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature	30
		vs Supply voltage	31
$\phi_m$	Phase margin	vs Supply voltage	34
		vs Free-air temperature	35
		vs Load capacitance	36
$V_n$	Equivalent input noise voltage	vs Frequency	37
		Phase shift	vs Frequency



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27L2  
 INPUT OFFSET VOLTAGE

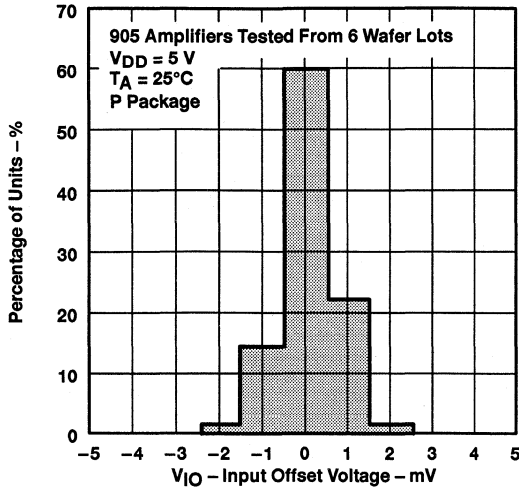


Figure 6

DISTRIBUTION OF TLC27L2  
 INPUT OFFSET VOLTAGE

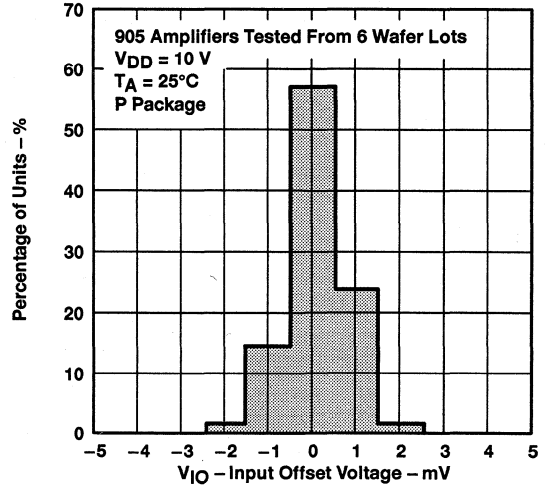


Figure 7

DISTRIBUTION OF TLC27L2C AND TLC27L7  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

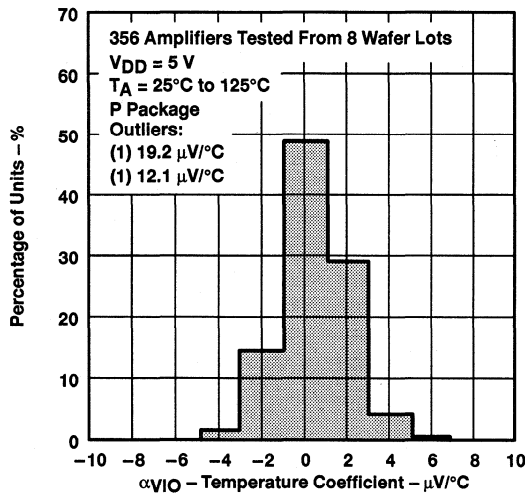


Figure 8

DISTRIBUTION OF TLC27L2C AND TLC27L7  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

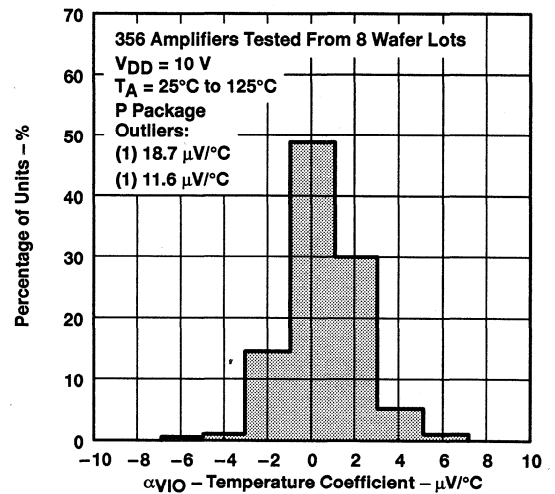


Figure 9

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

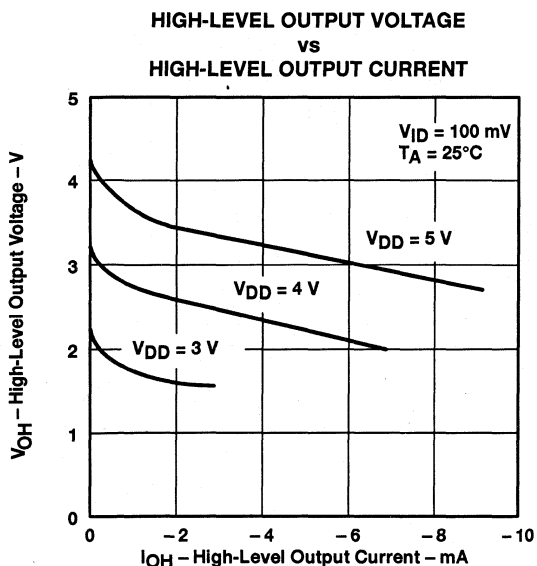


Figure 10

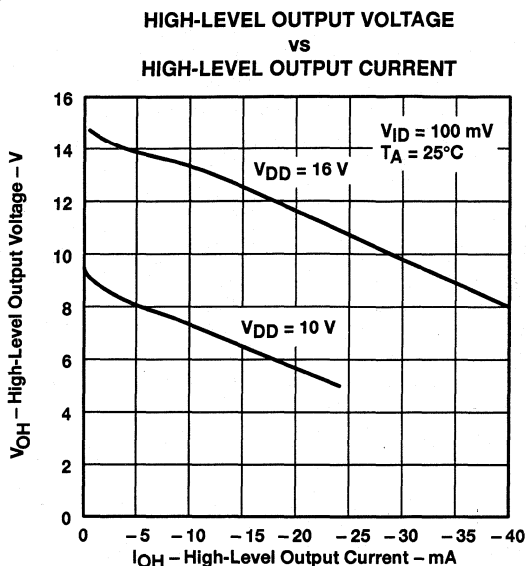


Figure 11

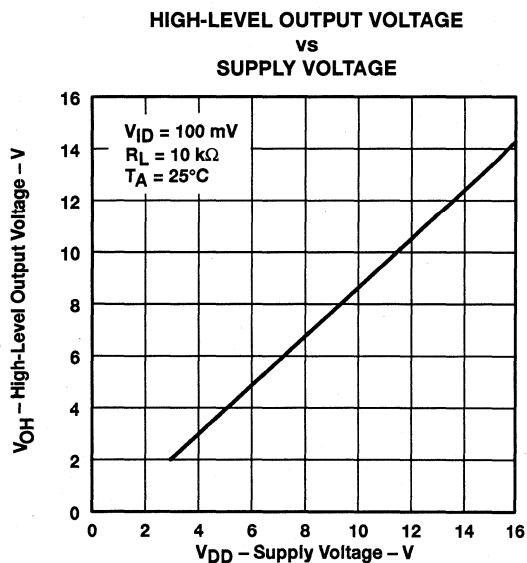


Figure 12

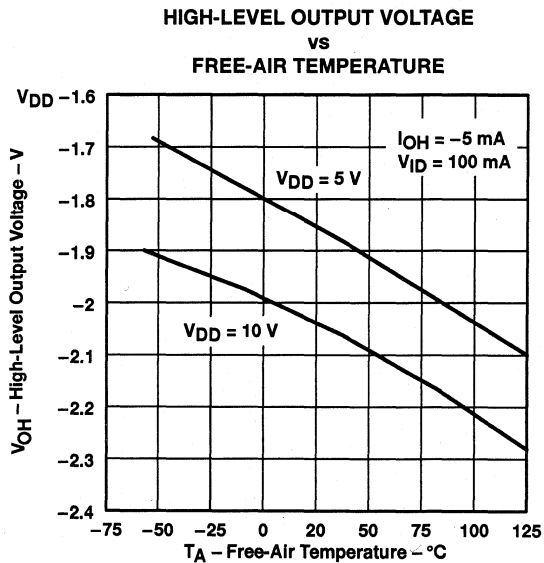


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

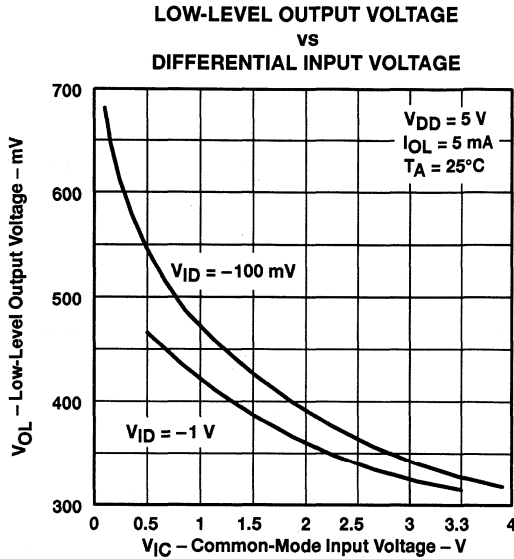


Figure 14

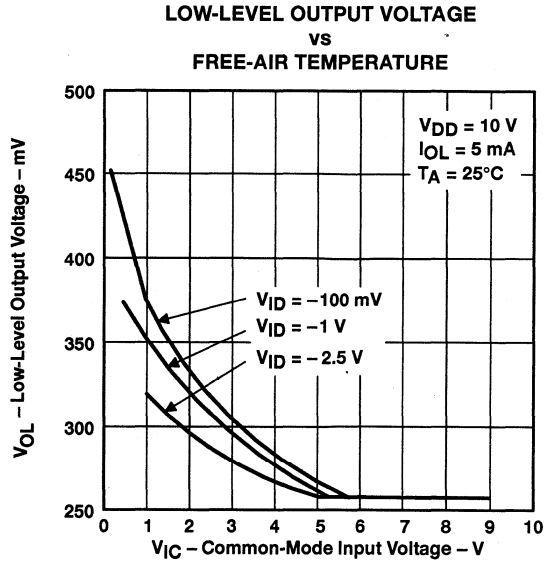


Figure 15

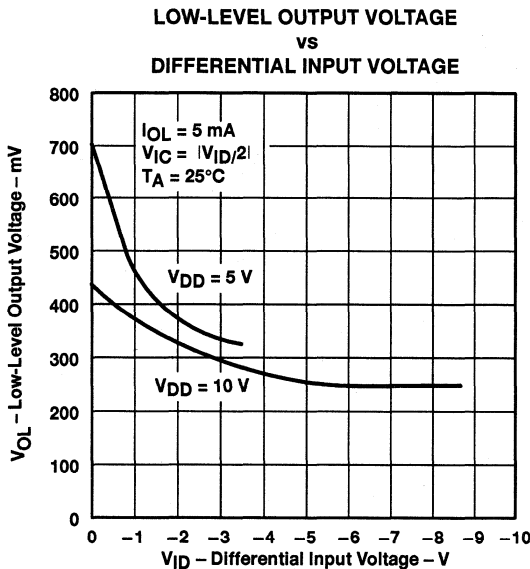


Figure 16

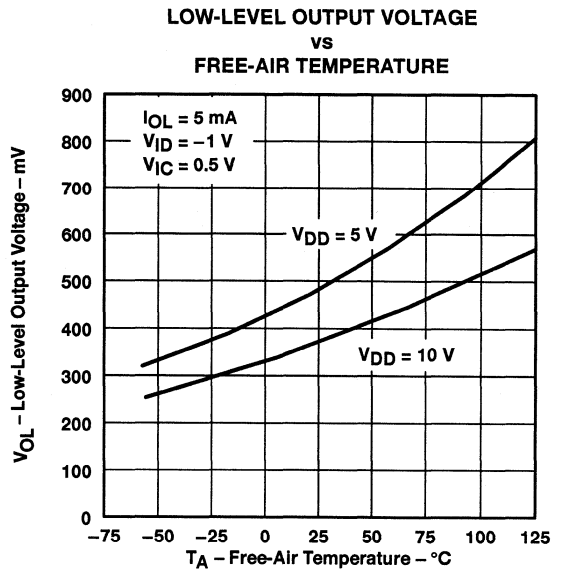


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

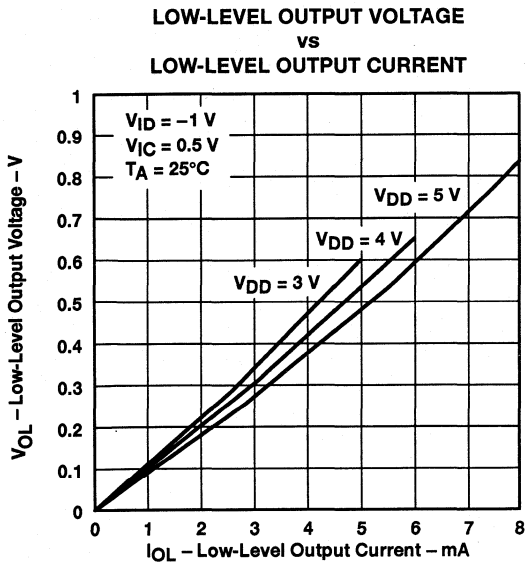


Figure 18

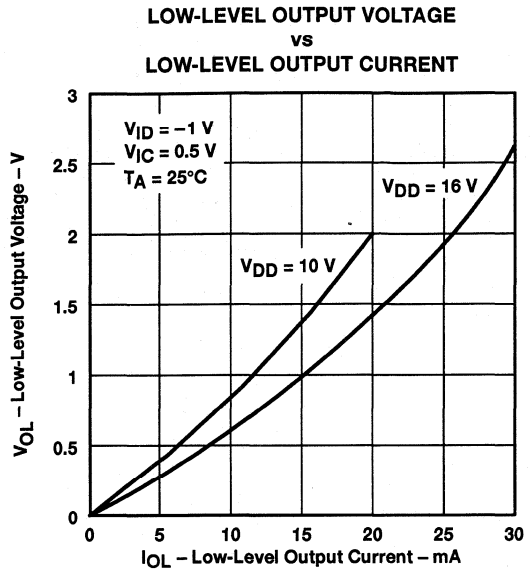


Figure 19

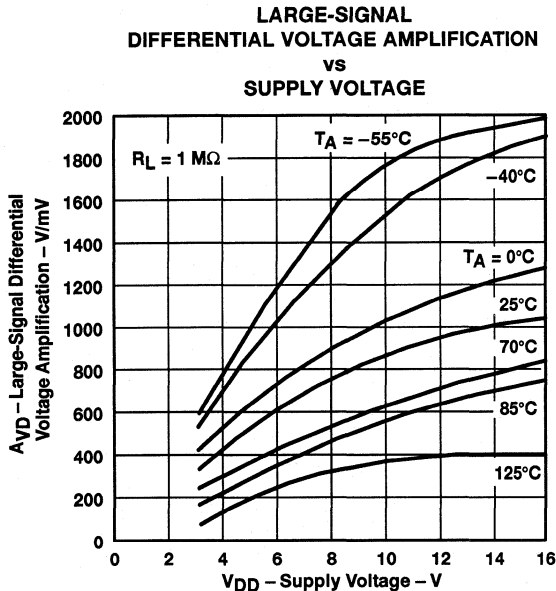


Figure 20

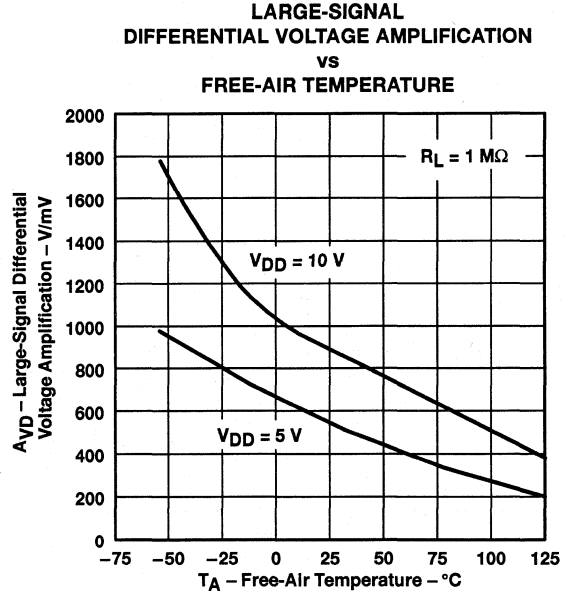


Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

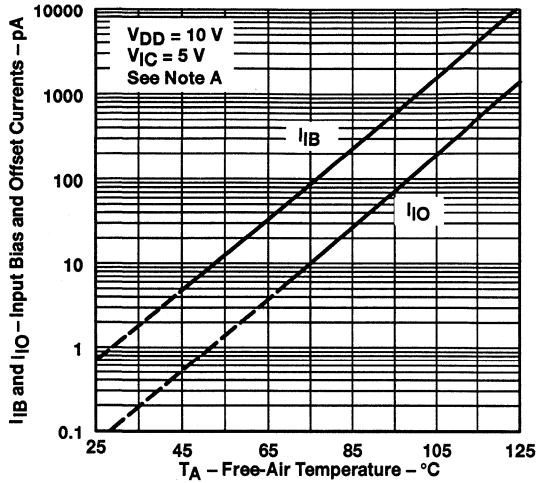


# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PROCISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

**INPUT BIAS CURRENT AND INPUT OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE**



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

**COMMON-MODE  
INPUT VOLTAGE POSITIVE LIMIT  
vs  
SUPPLY VOLTAGE**

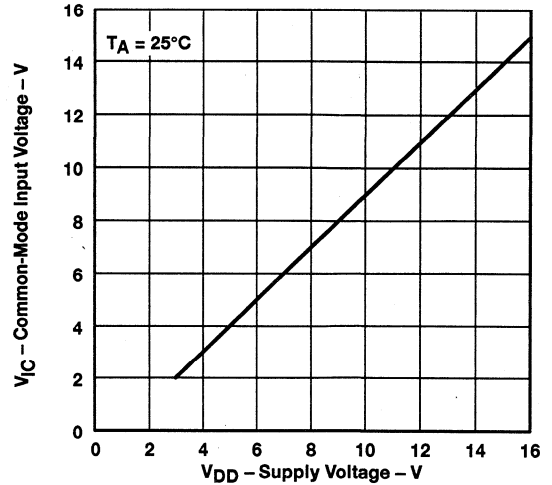


Figure 23

**SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE**

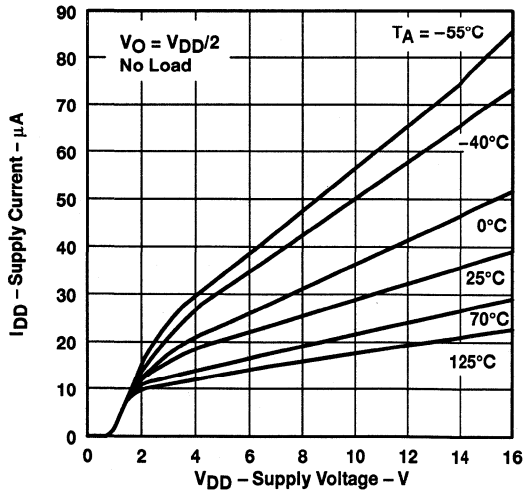


Figure 24

**SUPPLY CURRENT  
vs  
FREE-AIR TEMPERATURE**

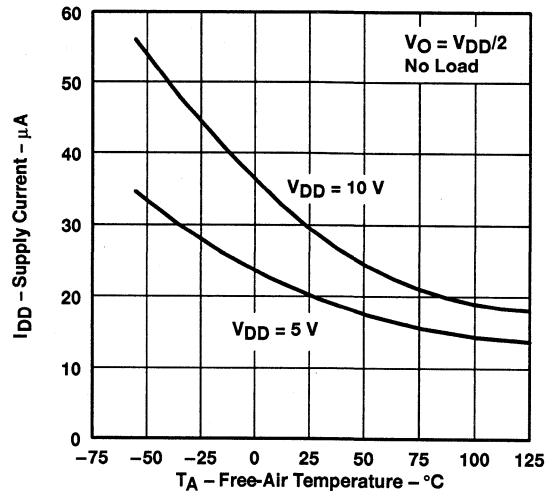


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

**SLEW RATE  
vs  
SUPPLY VOLTAGE**

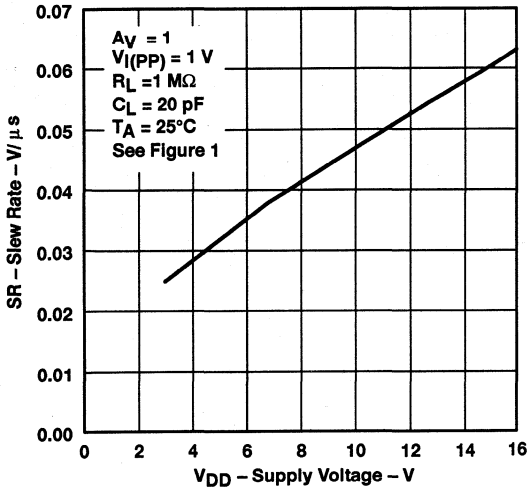


Figure 26

**SLEW RATE  
vs  
FREE-AIR TEMPERATURE**

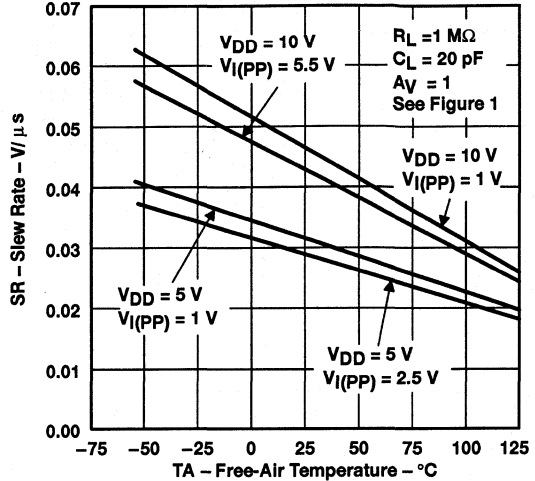


Figure 27

**NORMALIZED SLEW RATE  
vs  
FREE-AIR TEMPERATURE**

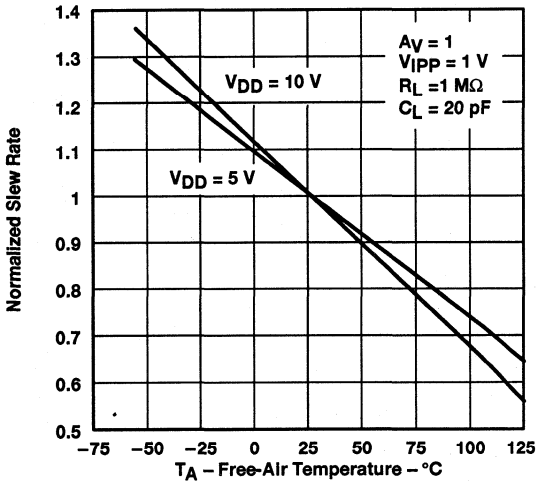


Figure 28

**MAXIMUM-PEAK-TO-PEAK OUTPUT VOLTAGE  
vs  
FREQUENCY**

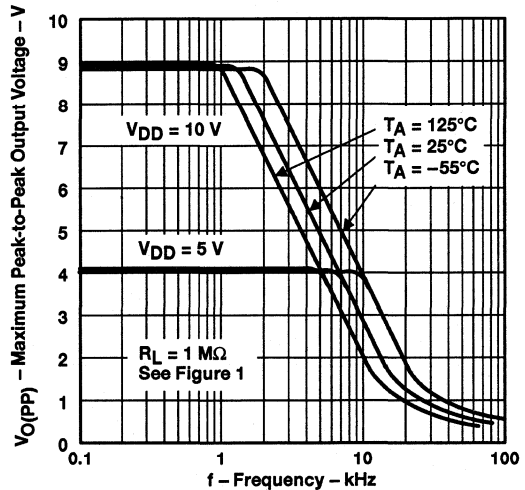


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PROCISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

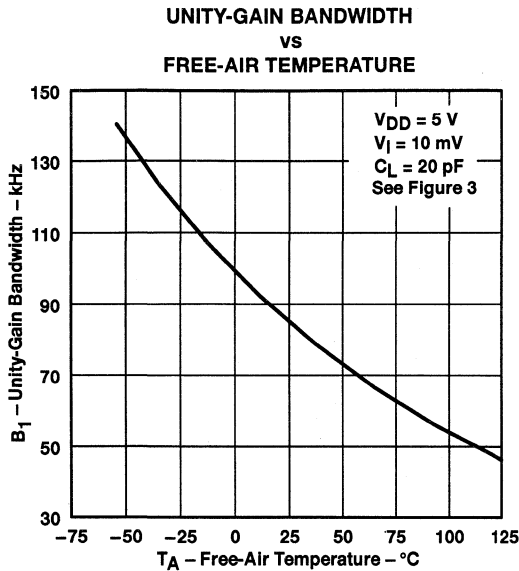


Figure 30

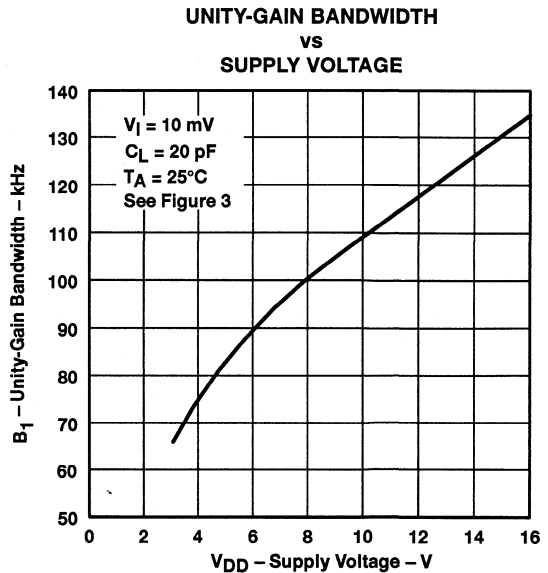


Figure 31

## LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT vs FREQUENCY

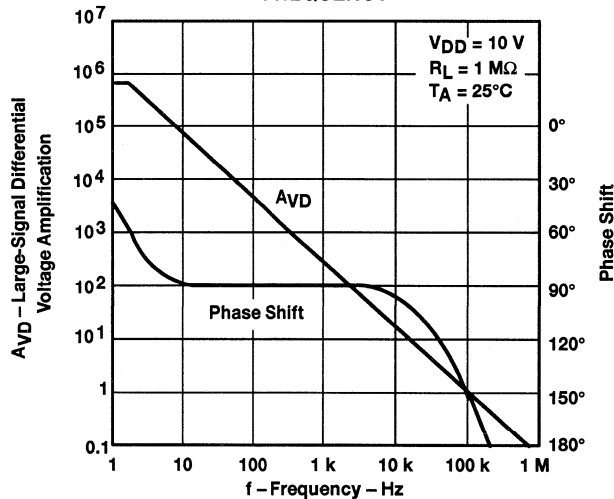


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

### LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

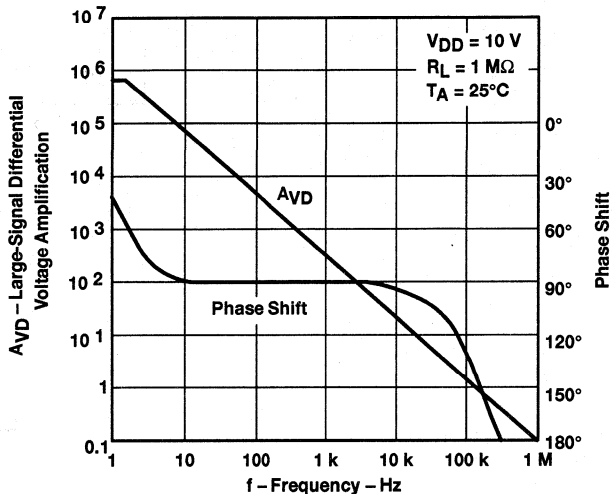


Figure 33

### PHASE MARGIN VS SUPPLY VOLTAGE

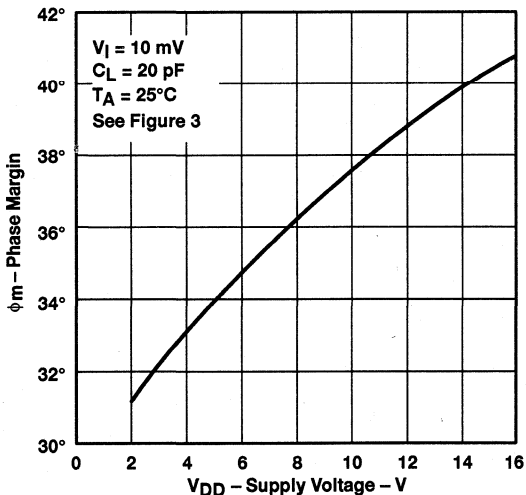


Figure 34

### PHASE MARGIN VS FREE-AIR TEMPERATURE

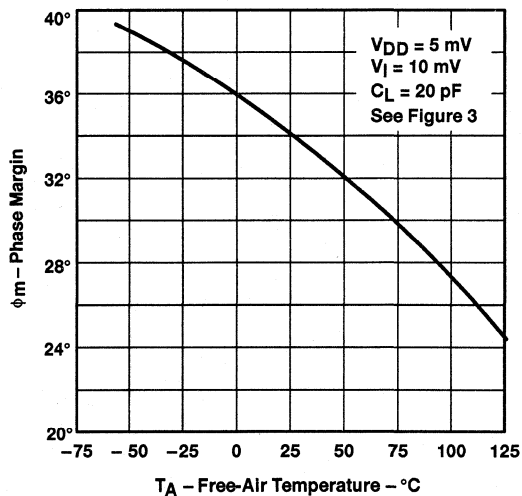


Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

PHASE MARGIN  
 VS  
 CAPACITIVE LOAD

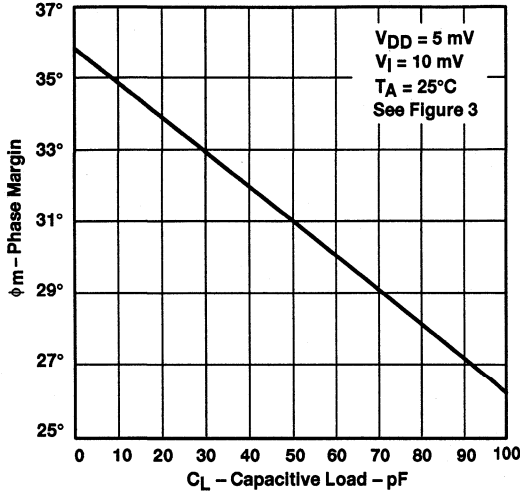


Figure 36

EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY

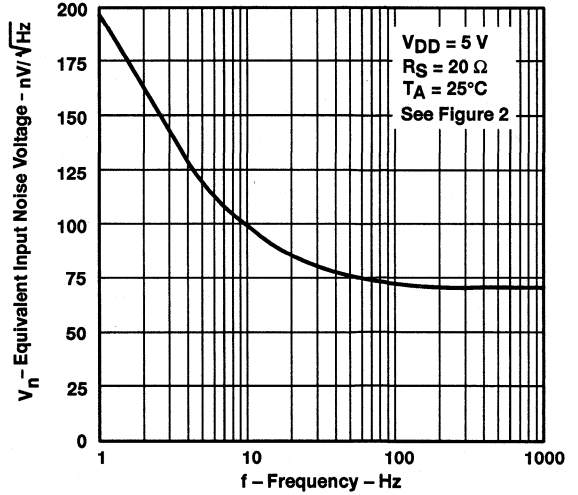


Figure 37

APPLICATION INFORMATION

single-supply operation

While the TLC27L2 and TLC27L7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27L2 and TLC27L7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27L2 and TLC27L7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

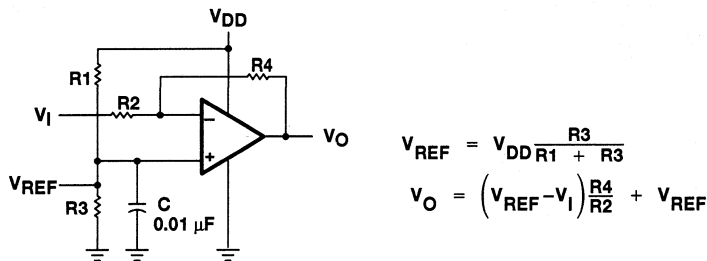


Figure 38. Inverting Amplifier With Voltage Reference

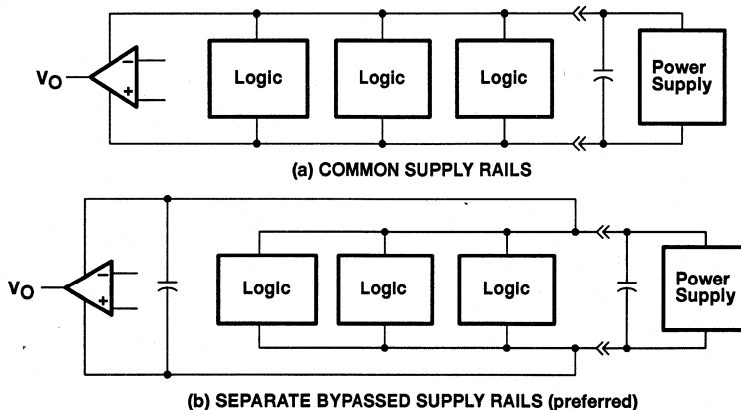


Figure 39. Common Versus Separate Supply Rails

## APPLICATION INFORMATION

### input characteristics

The TLC27L2 and TLC27L7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5$  V at all other temperatures.

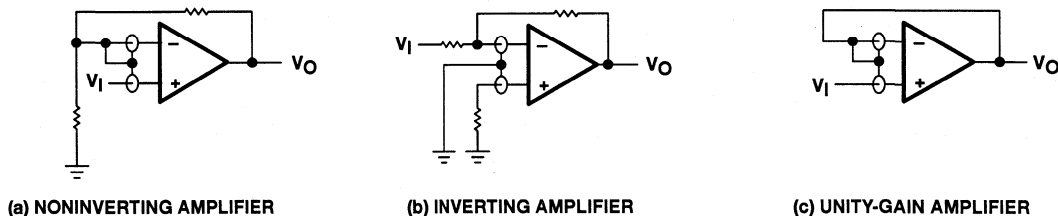
The use of the polysilicon-gate process and the careful input circuit design gives the TLC27L2 and TLC27L7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1 \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27L2 and TLC27L7 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

### noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27L2 and TLC27L7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50 \text{ k}\Omega$ , since bipolar devices exhibit greater noise currents.



**Figure 40. Guard-Ring Schemes**

### output characteristics

The output stage of the TLC27L2 and TLC27L7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

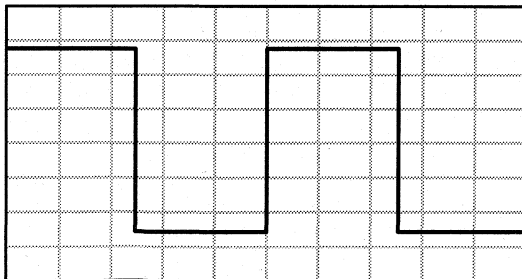
All operating characteristics of the TLC27L2 and TLC27L7 were measured using a  $20\text{-pF}$  load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

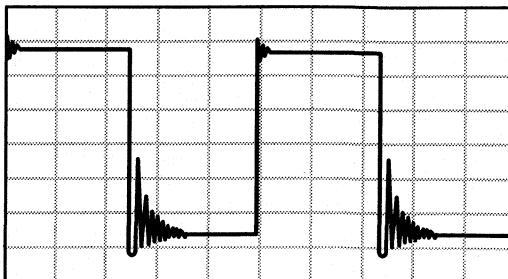
SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### output characteristics (continued)



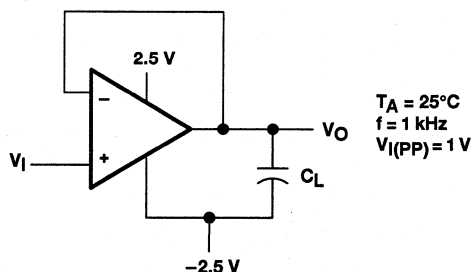
(a)  $C_L = 20 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(b)  $C_L = 260 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(c)  $C_L = 310 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(d) TEST CIRCUIT

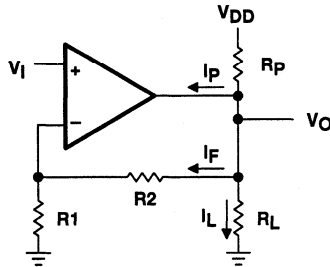
**Figure 41. Effect of Capacitive Loads and Test Circuit**

Although the TLC27L2 and TLC27L7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60 \Omega$  and  $180 \Omega$ , depending on how hard the operational amplifier input is driven. With very low values of  $R_P$ , a voltage offset from  $0 \text{ V}$  at the output occurs. Second, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.



APPLICATION INFORMATION

output characteristics (continued)



$$R_P = \frac{V_{DD} - V_O}{I_F + I_L + I_P}$$

$I_P$  = Pullup current required by the operational amplifier (typically 500  $\mu$ A)

Figure 42. Resistive Pullup to Increase  $V_{OH}$

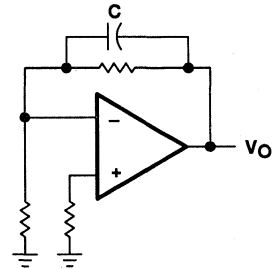


Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic discharge protection

The TLC27L2 and TLC27L7 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

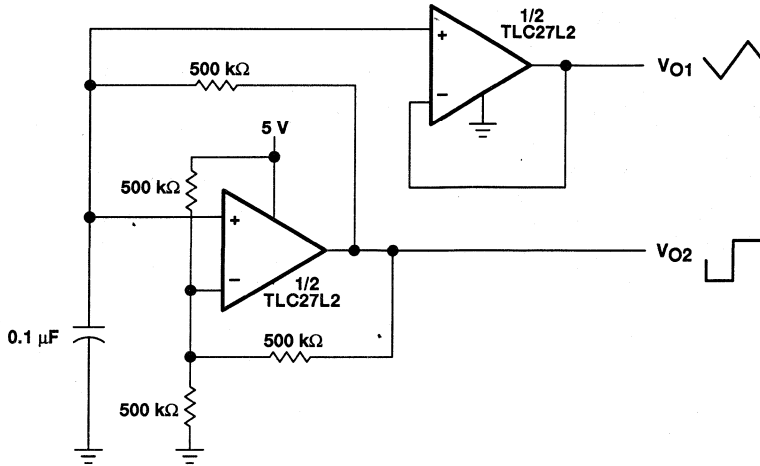
Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27L2 and TLC27L7 inputs and outputs were designed to withstand  $-100$ -mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1  $\mu$ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

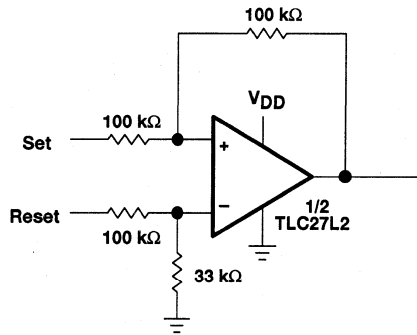
**TLC27L2, TLC27L2A, TLC27L2B, TLC27L7**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION**



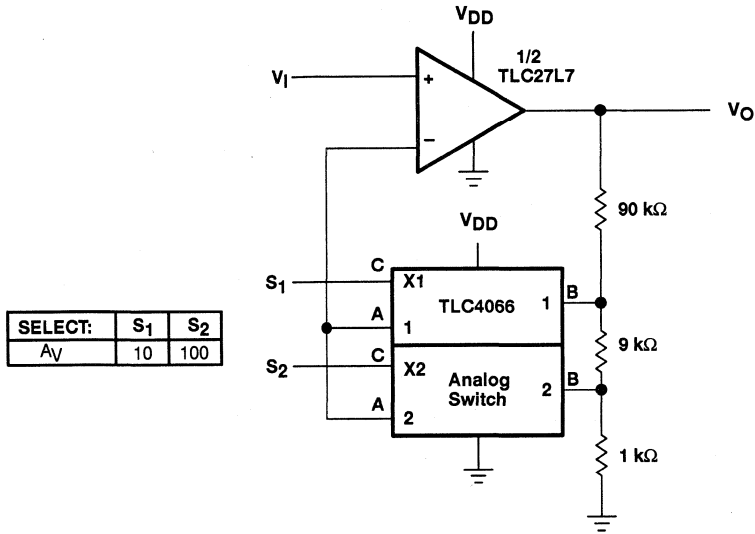
**Figure 44. Multivibrator**



NOTE: V<sub>DD</sub> = 5 V to 16 V

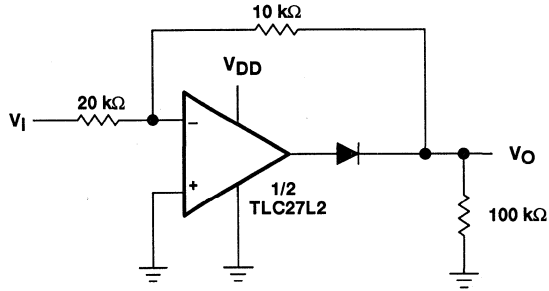
**Figure 45. Set/Reset Flip-Flop**

APPLICATION INFORMATION



NOTE: V<sub>DD</sub> = 5 V to 12 V

Figure 46. Amplifier With Digital Gain Selection



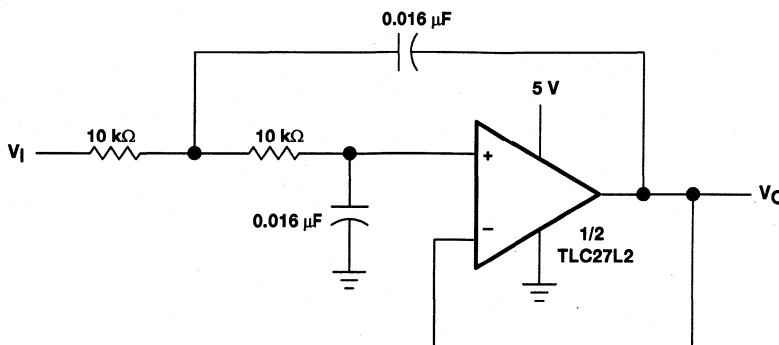
NOTE: V<sub>DD</sub> = 5 V to 16 V

Figure 47. Full-Wave Rectifier

# TLC27L2, TLC27L2A, TLC27L2B, TLC27L7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

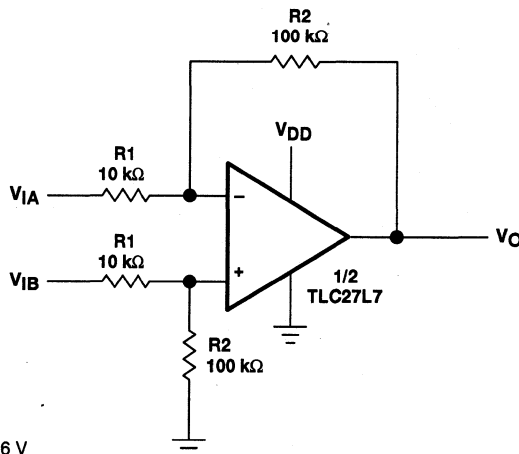
SLOS052B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION



NOTE: Normalized to  $f_c = 1$  kHz and  $R_L = 10$  k $\Omega$

Figure 48. Two-Pole Low-Pass Butterworth Filter



NOTE:  $V_{DD} = 5$  V to 16 V

$$V_O = \frac{R_2}{R_1}(V_{IB} - V_{IA})$$

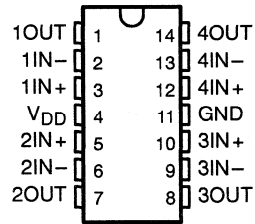
Figure 49. Difference Amplifier

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

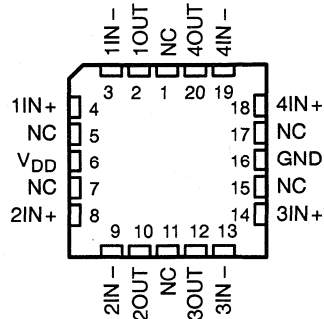
SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC27L9 . . . 900  $\mu\text{V}$  Max at 25°C,  
 $V_{\text{DD}} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
0°C to 70°C . . . 3 V to 16 V  
-40°C to 85°C . . . 4 V to 16 V  
-55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)**
- **Ultra-Low Power . . . Typically 195  $\mu\text{W}$  at 25°C,  $V_{\text{DD}} = 5\text{ V}$**
- **Output Voltage Range includes Negative Rail**
- **High Input Impedance . . .  $10^{12}\ \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**

D, J, N, OR PW PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

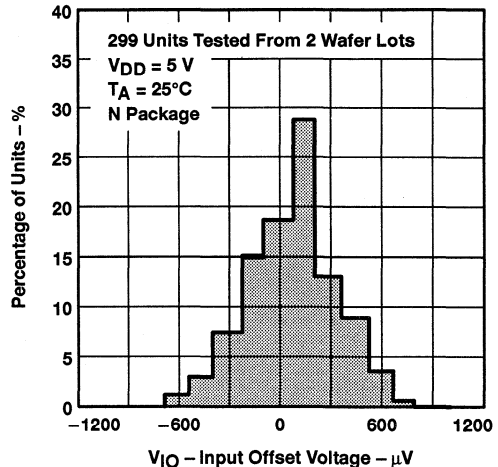
## description

The TLC27L4 and TLC27L9 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, extremely low power, and high gain.

These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and low-power consumption make these cost-effective devices ideal for high-gain, low-frequency, low-power applications. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27L4 (10 mV) to the high-precision TLC27L9 (900  $\mu\text{V}$ ). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

DISTRIBUTION OF TLC27L9  
INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## description (continued)

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27L4 and TLC27L9. The devices also exhibit low voltage single-supply operation and ultra-low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up.

The TLC27L4 and TLC27L9 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices, as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ . The I-suffix devices are characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The M-suffix devices are characterized for operation from  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	900 $\mu\text{V}$	TLC27L9CD	—	—	TLC27L9CN	—	—
	2 mV	TLC27L4BCD	—	—	TLC27L4BCN	—	—
	5 mV	TLC27L4ACD	—	—	TLC27L4ACN	—	—
	10 mV	TLC27L4CD	—	—	TLC27L4CN	TLC27L4CPW	TLC27L4Y
$-40^{\circ}\text{C}$ to $85^{\circ}\text{C}$	900 $\mu\text{V}$	TLC27L9ID	—	—	TLC27L9IN	—	—
	2 mV	TLC27L4BID	—	—	TLC27L4BIN	—	—
	5 mV	TLC27L4AID	—	—	TLC27L4AIN	—	—
	10 mV	TLC27L4ID	—	—	TLC27L4IN	—	—
$-55^{\circ}\text{C}$ to $125^{\circ}\text{C}$	900 $\mu\text{V}$	TLC27L9MD	TLC27L9MFK	TLC27L9MJ	TLC27L9MN	—	—
	10 mV	TLC27L4MD	TLC27L4MFK	TLC27L4MJ	TLC27L4MN	—	—

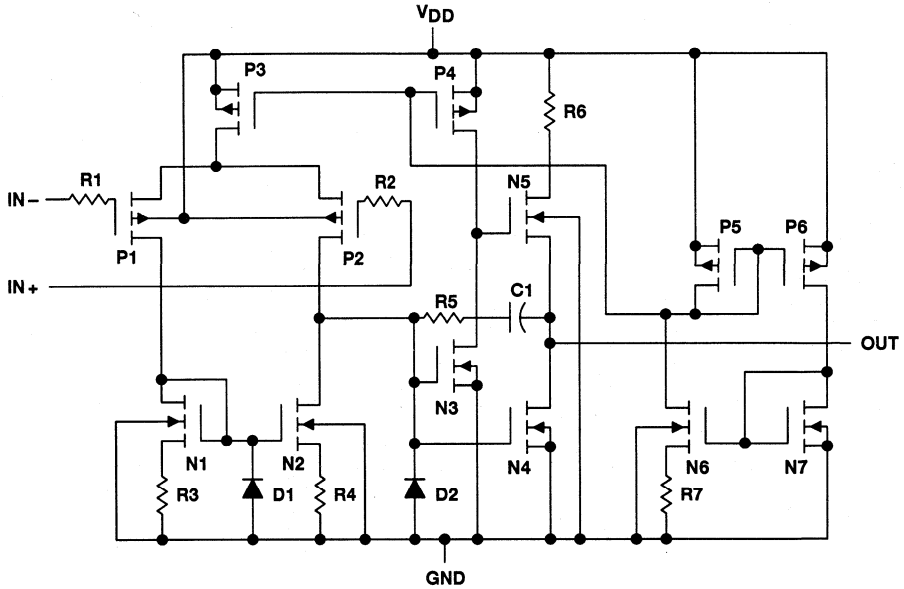
The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC27L9CDR).



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

equivalent schematic (each amplifier)

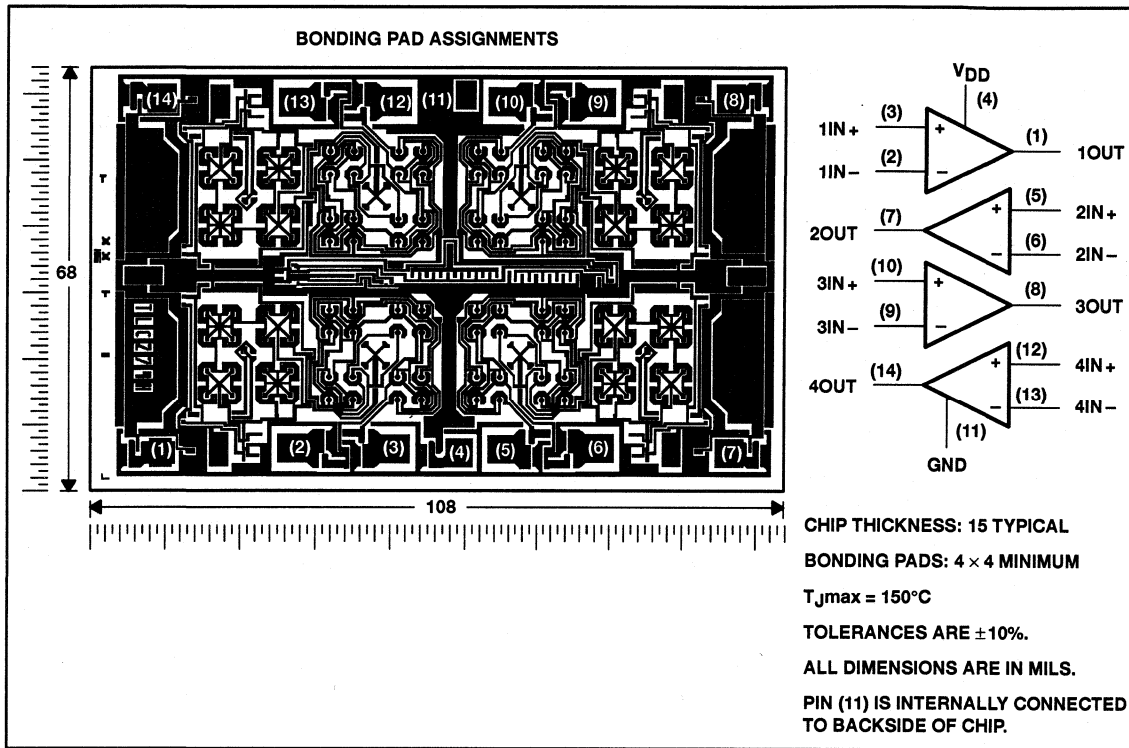


# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TLC27L4Y chip information

These chips, when properly assembled, display characteristics similar to the TLC27L4C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.





# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at  $IN_+$  with respect to  $IN_-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	—
PW	700 mW	5.6 mW/°C	448 mW	—	—

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L4C TLC27L4AC TLC27L4BC TLC27L9C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
	TLC27L4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	240	2000	$\mu\text{V}$	
				Full range		3000		
	TLC27L9C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	200	900	$\mu\text{V}$	
				Full range		1500		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				0°C	3	4.1		
				70°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$AVD$	Large-signal differential voltage amplification	$V_O = 2.5\text{ V to }2\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	520	V/mV	
				0°C	50	680		
				70°C	50	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	95		
				70°C	60	95		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				0°C	60	97		
				70°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	40	68	$\mu\text{A}$	
				0°C	48	84		
				70°C	31	56		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L4C TLC27L4AC TLC27L4BC TLC27L9C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC27L4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	260	2000	$\mu\text{V}$	
				Full range		3000		
	TLC27L9C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	210	1200		
				Full range		1900		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				0°C	7.8	8.9		
				70°C	7.8	8.9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 1\text{ M}\Omega$	25°C	50	870	V/mV	
				0°C	50	1020		
				70°C	50	660		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	97	dB	
				0°C	60	97		
				70°C	60	97		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	70	97	dB	
				0°C	60	97		
				70°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	57	92	$\mu\text{A}$	
				0°C	72	132		
				70°C	44	80		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC27L4I TLC27L4AI TLC27L4BI TLC27L9I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	TLC27L4I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
			Full range		13	
	TLC27L4AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9	5	mV
			Full range		7	
	TLC27L4BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	240	2000	$\mu\text{V}$
			Full range		3500	
	TLC27L9I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	200	900	$\mu\text{V}$
			Full range		2000	
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage		25°C to 85°C	1.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
		85°C	24	1000		
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
		85°C	200	2000		
$V_{ICR}$ Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	V	
		Full range	-0.2 to 3.5		V	
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
		-40°C	3	4.1		
		85°C	3	4.2		
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
		-40°C	0	50		
		85°C	0	50		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ , $R_L = 1\text{ M}\Omega$	25°C	50	480	V/mV	
		-40°C	50	900		
		85°C	50	330		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	65	94	dB	
		-40°C	60	95		
		85°C	60	95		
$K_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	70	97	dB	
		-40°C	60	97		
		85°C	60	98		
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$	25°C	39	68	$\mu\text{A}$	
		-40°C	62	108		
		85°C	29	52		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L4I TLC27L4AI TLC27L4BI TLC27L9I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L4I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1 10		mV
					Full range	13		
		TLC27L4AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	0.9 5		
					Full range	7		
	TLC27L4BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	260	2000	$\mu\text{V}$	
				Full range	3500			
		TLC27L9I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	210		1200
					Full range	2900		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				-40°C	7.8	8.9		
				85°C	7.8	8.9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0 50		mV	
				-40°C	0 50			
				85°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	800	V/mV	
				-40°C	50	1550		
				85°C	50	585		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	97	dB	
				-40°C	60	97		
				85°C	60	98		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-40°C	60	97		
				85°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	57	92	$\mu\text{A}$	
				-40°C	98	172		
				85°C	40	72		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L4M TLC27L9M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L4M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L9M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	200	900	$\mu\text{V}$
					Full range		3750	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ ,	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 1\text{ M}\Omega$	25°C	3.2	4.1	V	
				-55°C	3	4.1		
				125°C	3	4.2		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ ,	$R_L = 1\text{ M}\Omega$	25°C	50	480	V/mV	
				-55°C	25	950		
				125°C	25	200		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-55°C	60	95		
				125°C	60	85		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-55°C	60	97		
				125°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$ ,	25°C	39	68	$\mu\text{A}$	
				-55°C	69	120		
				125°C	27	48		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27L4M TLC27L9M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27L4M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27L9M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$	25°C	210	1200	$\mu\text{V}$
					Full range		4300	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.4		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 1\text{ M}\Omega$	25°C	8	8.9	V	
				-55°C	7.8	8.8		
				125°C	7.8	9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 1\text{ M}\Omega$	25°C	50	800	V/mV	
				-55°C	25	1750		
				125°C	25	380		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	97	dB	
				-55°C	60	97		
				125°C	60	91		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	97	dB	
				-55°C	60	97		
				125°C	60	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	57	92	$\mu\text{A}$	
				-55°C	111	192		
				125°C	35	60		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and Input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27L4Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage	$T_A = 25^\circ\text{C}$ to $70^\circ\text{C}$		1.1		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 1\text{ M}\Omega$	3.2	4.1		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$AV_D$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V}$ to $2\text{ V}$ , $R_L = 1\text{ M}\Omega$	50	520		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	94		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V}$ to $10\text{ V}$ , $V_O = 1.4\text{ V}$	70	97		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$		40	68	$\mu\text{A}$

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27L4Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 1\text{ M}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage	$T_A = 25^\circ\text{C}$ to $70^\circ\text{C}$		1		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.7		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 9	-0.3 to 9.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 1\text{ M}\Omega$	8	8.9		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$AV_D$ Large-signal differential voltage amplification	$V_O = 1\text{ V}$ to $6\text{ V}$ , $R_L = 1\text{ M}\Omega$	50	870		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	97		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V}$ to $10\text{ V}$ , $V_O = 1.4\text{ V}$	70	97		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load $V_{IC} = 5\text{ V}$		57	92	$\mu\text{A}$

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4C TLC27L4AC TLC27L4BC TLC27L9C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	25°C	0.03			V/ $\mu$ s
			0°C	0.04			
			70°C	0.03			
		$V_{I\text{PP}} = 2.5\text{ V}$	25°C	0.03			
			0°C	0.03			
			70°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	70			nV/ $\sqrt{\text{Hz}}$
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ , See Figure 1	25°C	5			kHz
			0°C	6			
			70°C	4.5			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	85			kHz
			0°C	100			
			70°C	65			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	34°			
			0°C	36°			
			70°C	30°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4C TLC27L4AC TLC27L4BC TLC27L9C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	25°C	0.05			V/ $\mu$ s
			0°C	0.05			
			70°C	0.04			
		$V_{I\text{PP}} = 5.5\text{ V}$	25°C	0.04			
			0°C	0.05			
			70°C	0.04			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	70			nV/ $\sqrt{\text{Hz}}$
BOM Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$ , See Figure 1	25°C	1			kHz
			0°C	1.3			
			70°C	0.9			
B <sub>1</sub> Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	110			kHz
			0°C	125			
			70°C	90			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	38°			
			0°C	40°			
			70°C	34°			



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4I TLC27L4AI TLC27L4BI TLC27L9I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	25°C	0.03			V/ $\mu\text{s}$
			-40°C	0.04			
			85°C	0.03			
		$V_{I\text{PP}} = 2.5\text{ V}$	25°C	0.03			
			-40°C	0.04			
			85°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ Hz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	70			nV/ $\sqrt{\text{Hz}}$
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	5			kHz
			-40°C	7			
			85°C	4			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	85			kHz
			-40°C	130			
			85°C	55			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	34°			
			-40°C	38°			
			85°C	28°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4I TLC27L4AI TLC27L4BI TLC27L9I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I\text{PP}} = 1\text{ V}$	25°C	0.05			V/ $\mu\text{s}$
			-40°C	0.06			
			85°C	0.03			
		$V_{I\text{PP}} = 2.5\text{ V}$	25°C	0.04			
			-40°C	0.05			
			85°C	0.03			
$V_n$ Equivalent input noise voltage	$f = 1\text{ Hz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	70			nV/ $\sqrt{\text{Hz}}$
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 1\text{ M}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	1			kHz
			-40°C	1.4			
			85°C	0.8			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	110			kHz
			-40°C	155			
			85°C	80			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	38°			
			-40°C	42°			
			85°C	32°			



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4M TLC27L9M			UNIT
				MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.03		V/ $\mu$ s	
			-55°C	0.04			
			125°C	0.02			
		$V_{IPP} = 2.5\text{ V}$	25°C	0.03			
			-55°C	0.04			
			125°C	0.02			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	70		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	5		kHz	
			-55°C	8			
			125°C	3			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	85		kHz	
			-55°C	140			
			125°C	45			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	34°			
			-55°C	39°			
			125°C	25°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27L4M TLC27L9M			UNIT
				MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.05		V/ $\mu$ s	
			-55°C	0.06			
			125°C	0.03			
		$V_{IPP} = 5.5\text{ V}$	25°C	0.04			
			-55°C	0.06			
			125°C	0.03			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$ ,	25°C	70		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	1		kHz	
			-55°C	1.5			
			125°C	0.7			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	110		kHz	
			-55°C	165			
			125°C	70			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	38°			
			-55°C	43°			
			125°C	29°			



# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

### operating characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC27L4Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	0.03		$\text{V}/\mu\text{s}$
		$V_{Ipp} = 2.5\text{ V}$	0.03		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	70		$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	5		kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	85		kHz
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	34°		

### operating characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC27L4Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	0.05		$\text{V}/\mu\text{s}$
		$V_{Ipp} = 5.5\text{ V}$	0.04		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	70		$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 1\text{ M}\Omega$ , See Figure 1	$C_L = 20\text{ pF}$	1		kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	110		kHz
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	38°		



PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27L4 and TLC27L9 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

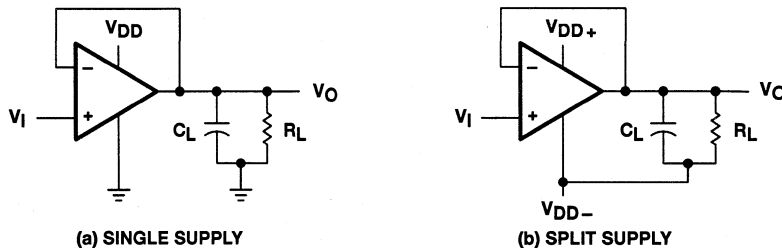


Figure 1. Unity-Gain Amplifier

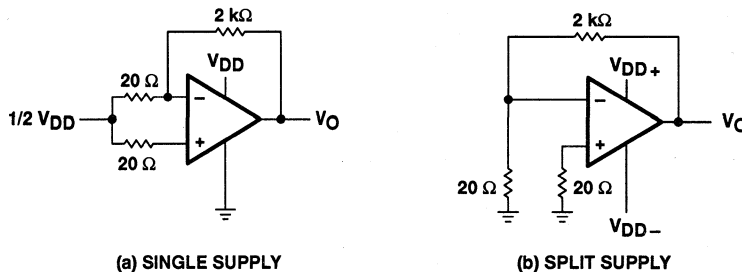


Figure 2. Noise-Test Circuit

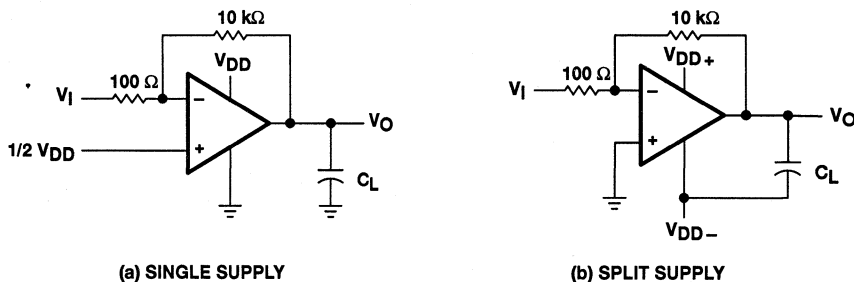


Figure 3. Gain-of-100 Inverting Amplifier

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC27L4 and TLC27L9 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

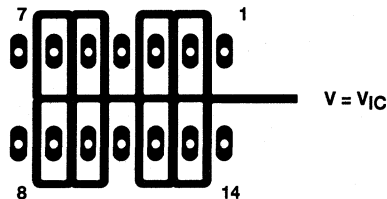


Figure 4. Isolation Metal Around Device Inputs (J and N packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

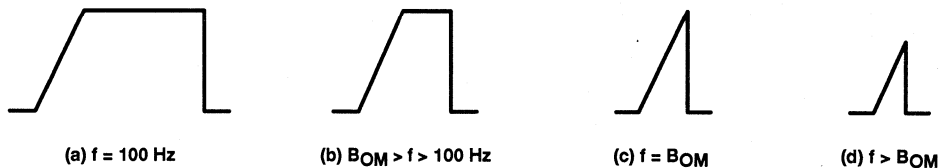


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

### TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha_{VIO}$	Temperature coefficient	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	10, 11 12 13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	14, 15 16 17 18, 19
$A_{VD}$	Differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	20 21 32, 33
$I_{IB}/I_{IO}$	Input bias and input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	24 25
$SR$	Slew rate	vs Supply voltage vs Free-air temperature	26 27
	Normalized slew rate	vs Free-air temperature	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	30 31
$\phi_m$	Phase margin	vs Supply voltage vs Free-air temperature vs Capacitive loads	34 35 36
$V_n$	Equivalent input noise voltage	vs Frequency	37
$\phi$	Phase shift	vs Frequency	32, 33



TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27L4  
 INPUT OFFSET VOLTAGE

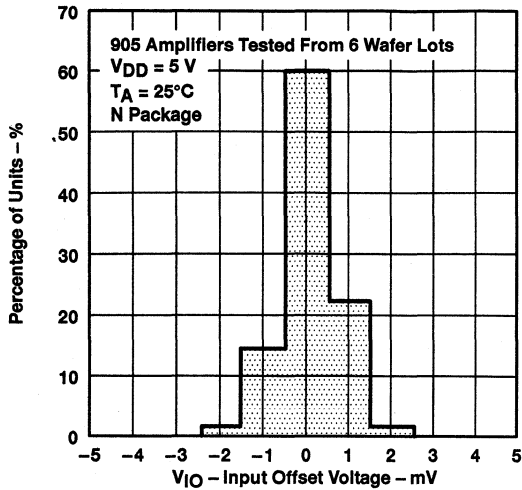


Figure 6

DISTRIBUTION OF TLC27L4  
 INPUT OFFSET VOLTAGE

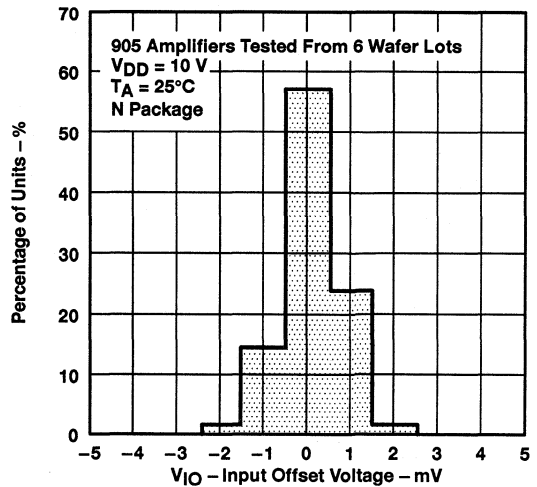


Figure 7

DISTRIBUTION OF TLC27L4 AND TLC27L9  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

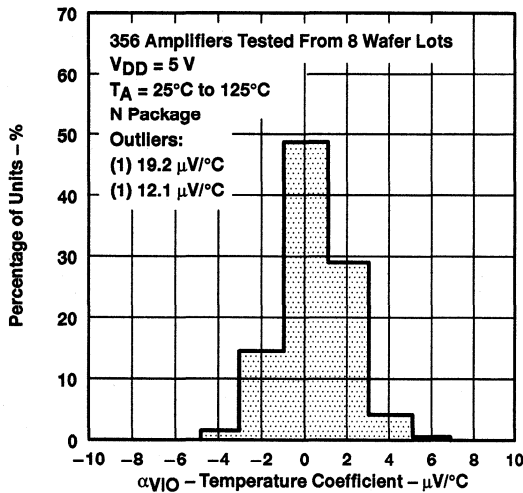


Figure 8

DISTRIBUTION OF TLC27L4 AND TLC27L9  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

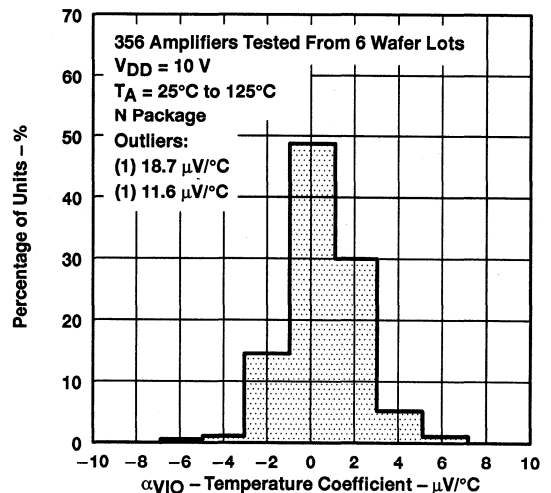


Figure 9

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

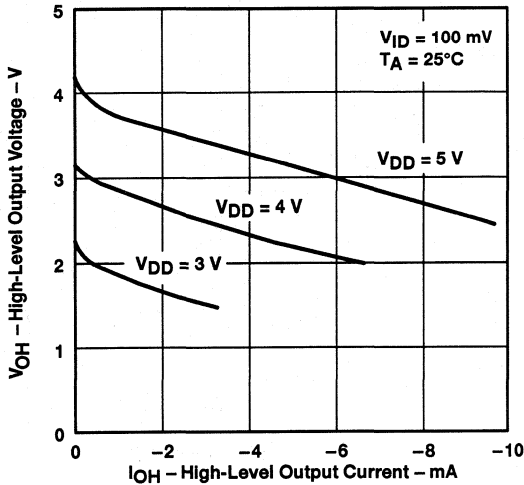


Figure 10

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
HIGH-LEVEL OUTPUT CURRENT

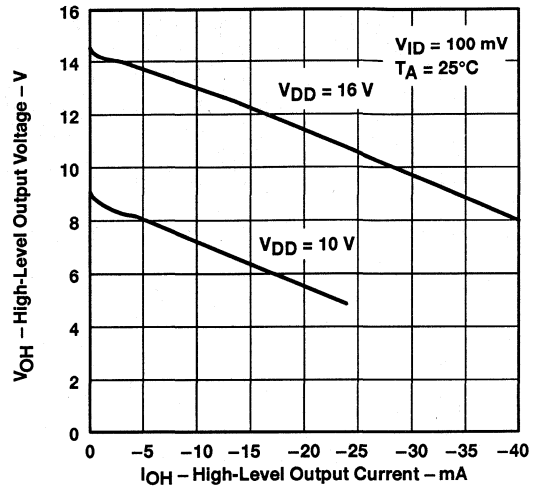


Figure 11

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE

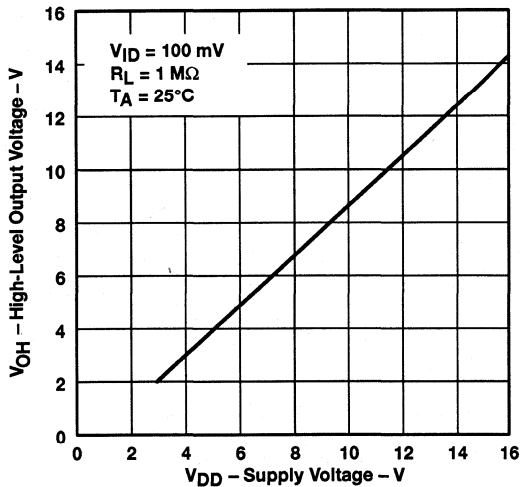


Figure 12

HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

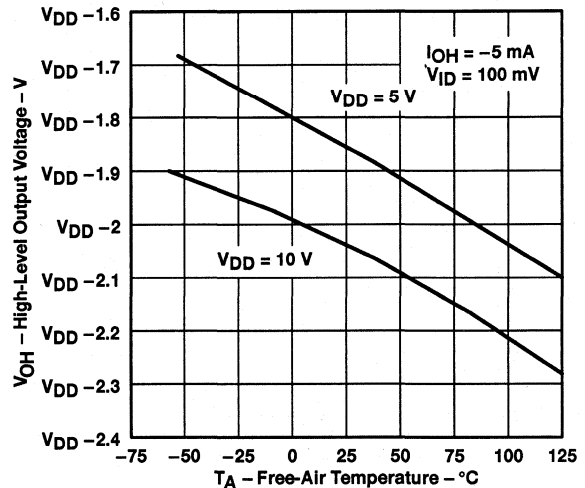


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
VS  
COMMON-MODE INPUT VOLTAGE

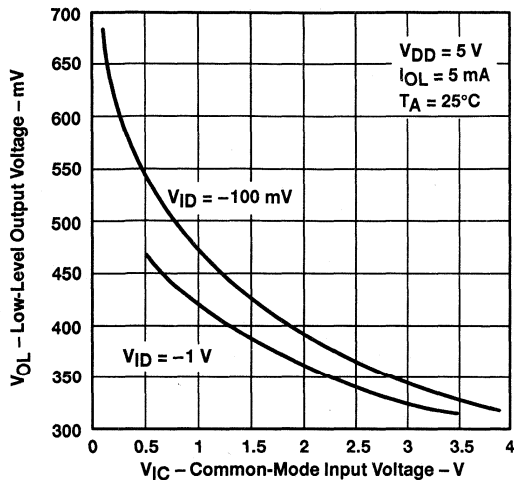


Figure 14

LOW-LEVEL OUTPUT VOLTAGE  
VS  
COMMON-MODE INPUT VOLTAGE

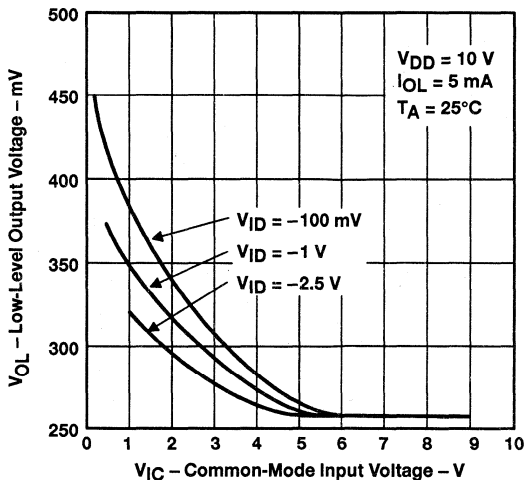


Figure 15

LOW-LEVEL OUTPUT VOLTAGE  
VS  
DIFFERENTIAL INPUT VOLTAGE

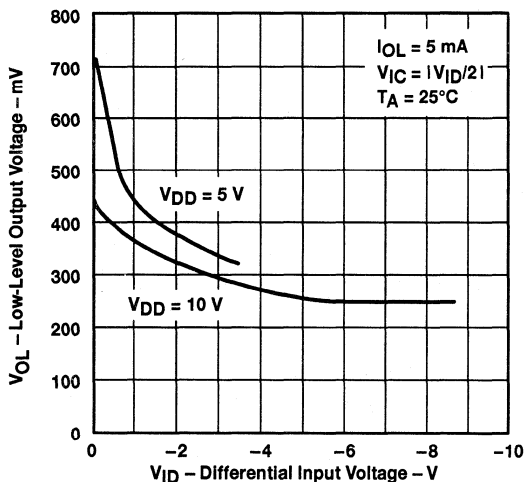


Figure 16

LOW-LEVEL OUTPUT VOLTAGE  
VS  
FREE-AIR TEMPERATURE

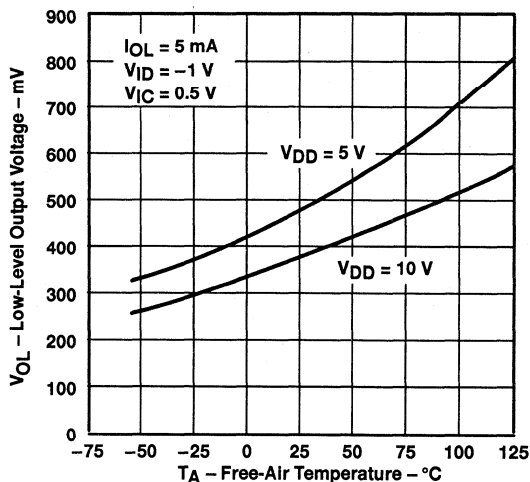


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

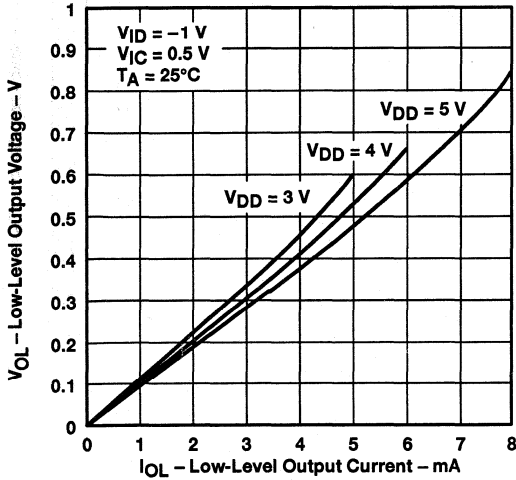


Figure 18

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

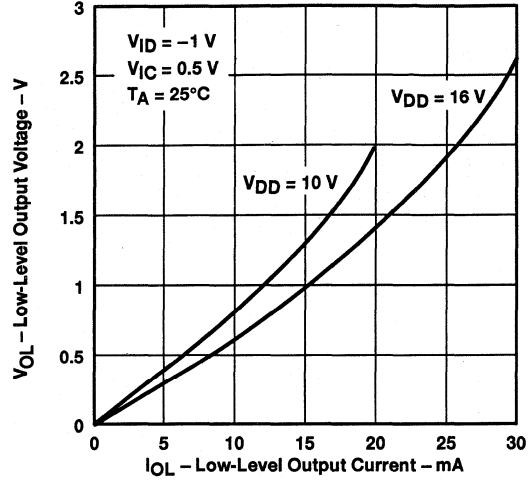


Figure 19

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
SUPPLY VOLTAGE

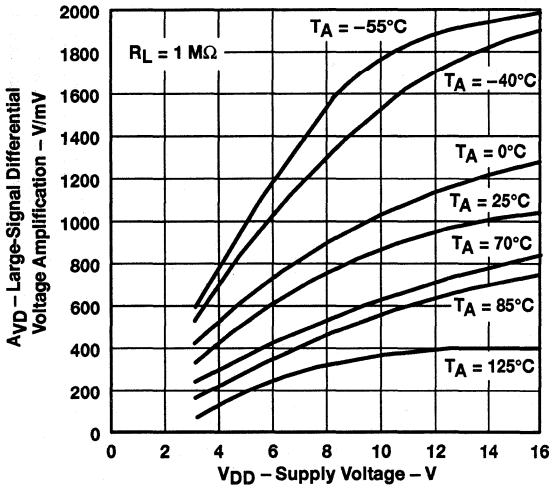


Figure 20

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
FREE-AIR TEMPERATURE

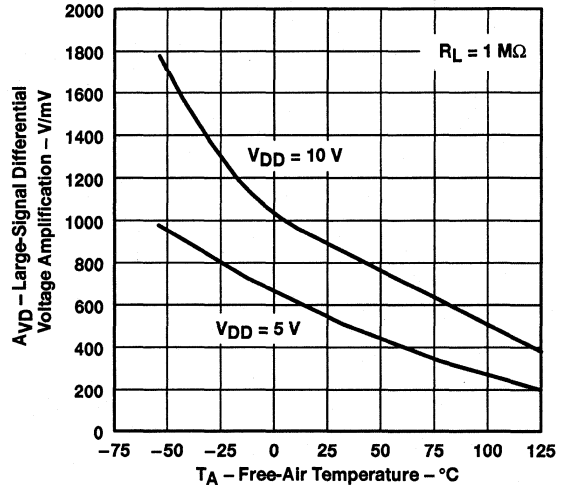
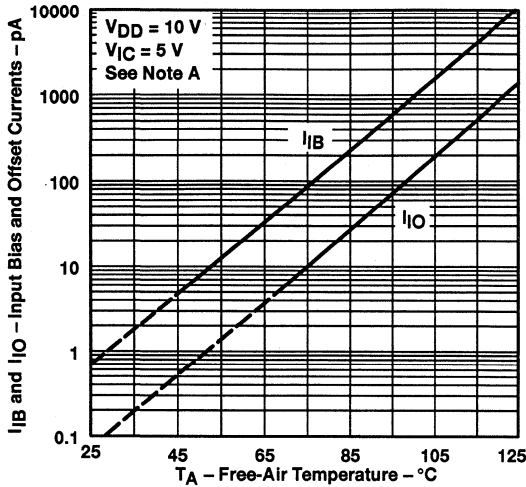


Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT  
 vs  
 FREE-AIR TEMPERATURE



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

COMMON-MODE  
 INPUT VOLTAGE POSITIVE LIMIT  
 vs  
 SUPPLY VOLTAGE

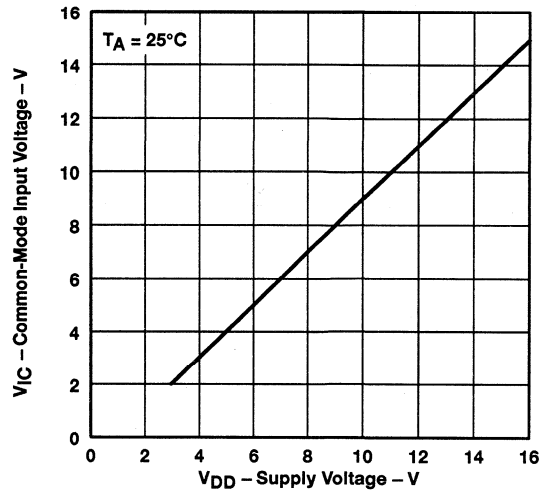


Figure 23

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

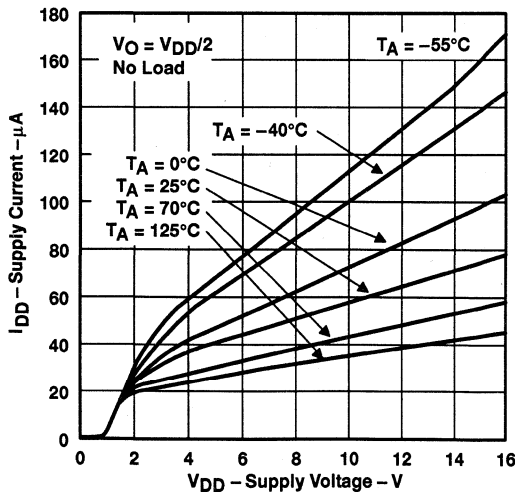


Figure 24

SUPPLY CURRENT  
 vs  
 FREE-AIR TEMPERATURE

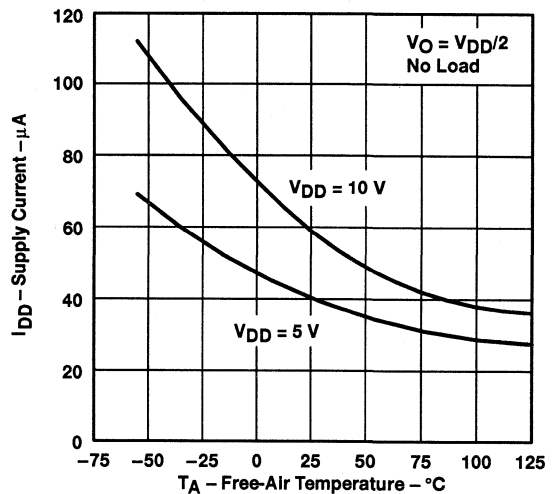


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

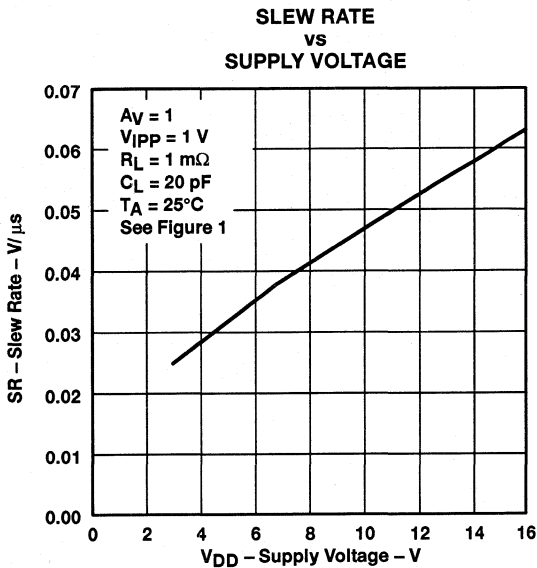


Figure 26

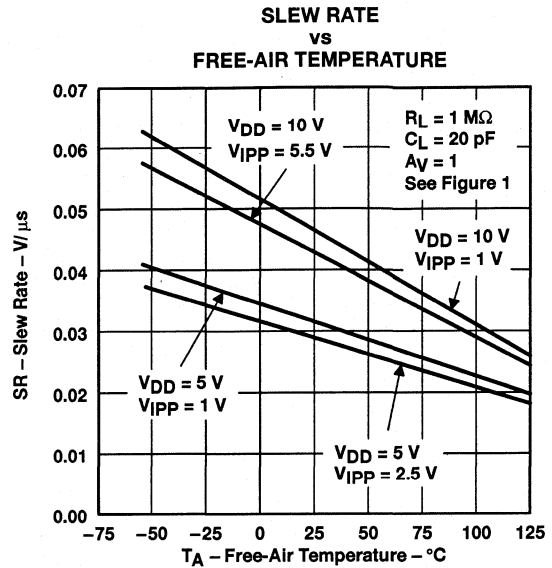


Figure 27

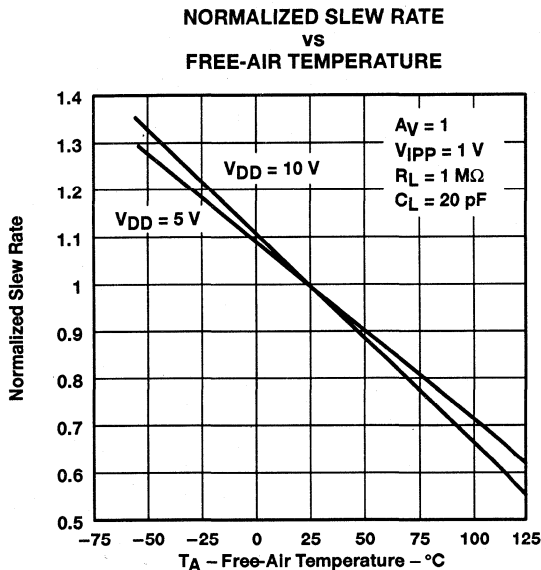


Figure 28

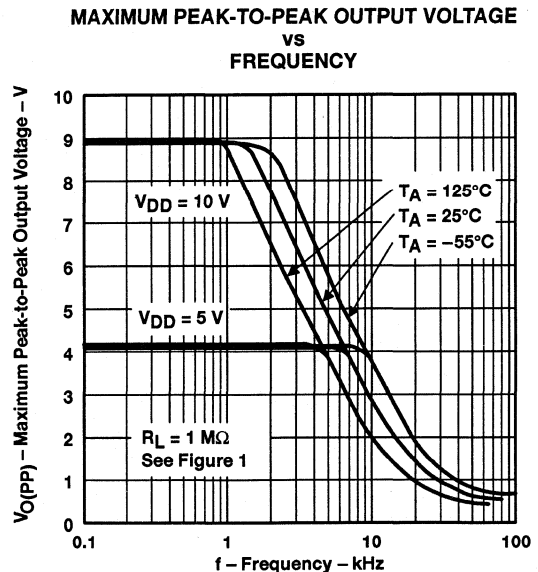


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH  
vs  
FREE-AIR TEMPERATURE

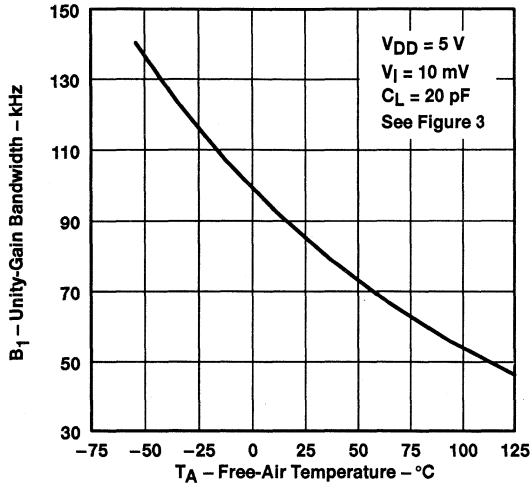


Figure 30

UNITY-GAIN BANDWIDTH  
vs  
SUPPLY VOLTAGE

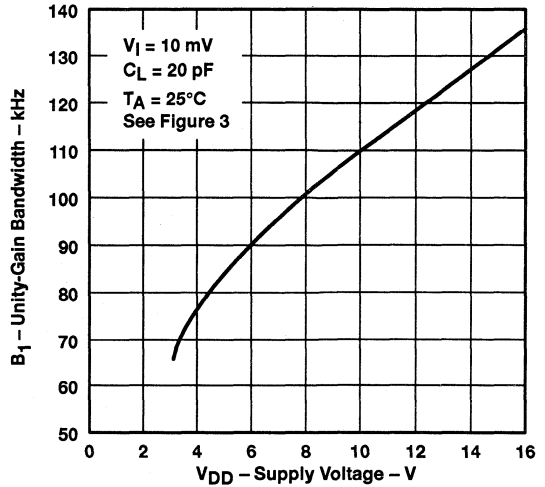


Figure 31

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
AMPLIFICATION AND PHASE SHIFT  
vs  
FREQUENCY

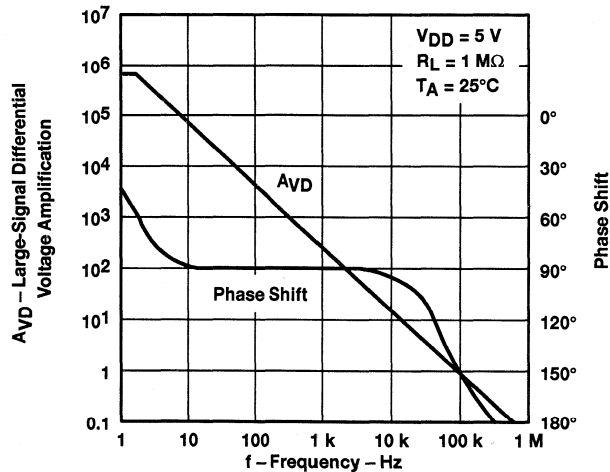


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

### LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

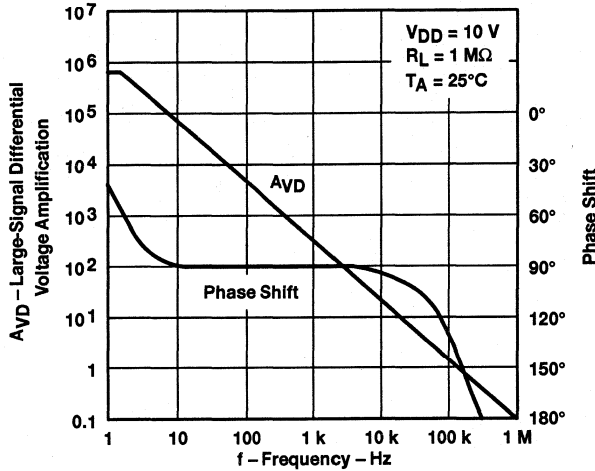


Figure 33

### PHASE MARGIN VS SUPPLY VOLTAGE

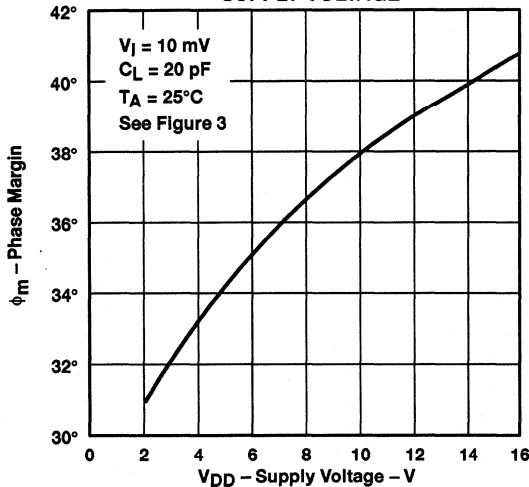


Figure 34

### PHASE MARGIN VS FREE-AIR TEMPERATURE

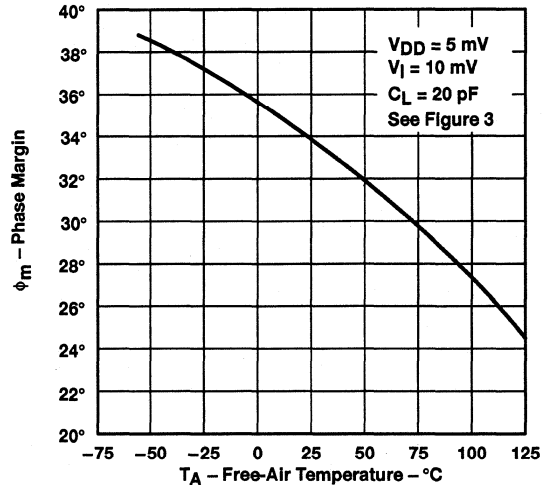


Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

PHASE MARGIN  
 VS  
 CAPACITIVE LOAD

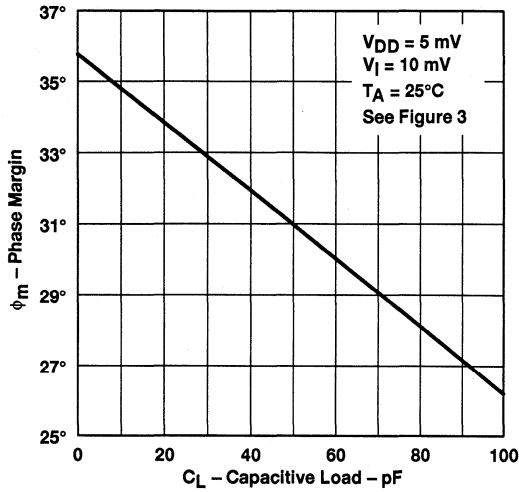


Figure 36

EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY

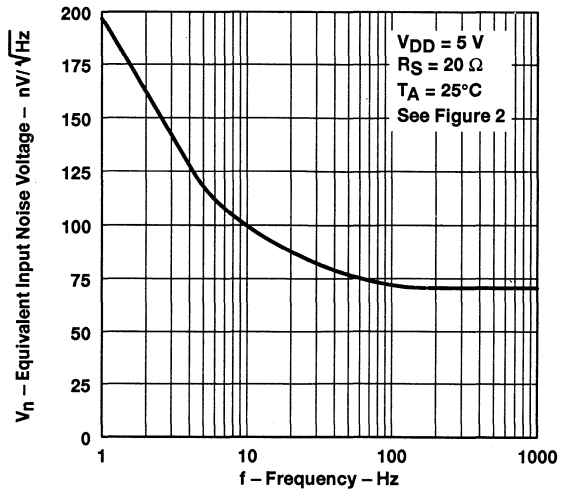


Figure 37

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### single-supply operation

While the TLC27L4 and TLC27L9 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27L4 and TLC27L9 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27L4 and TLC27L9 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

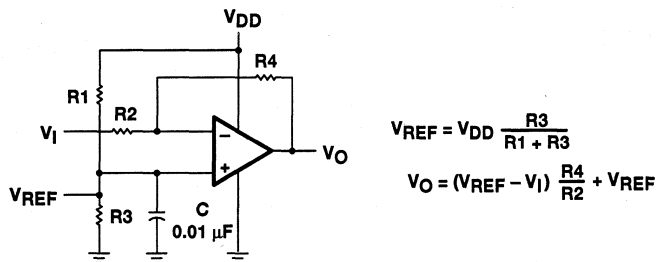


Figure 38. Inverting Amplifier With Voltage Reference

APPLICATION INFORMATION

single-supply operation (continued)

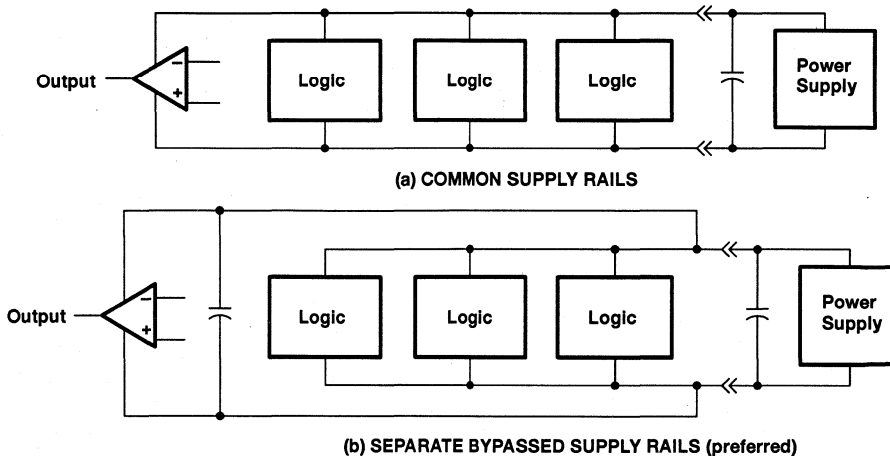


Figure 39. Common Versus Separate Supply Rails

input characteristics

The TLC27L4 and TLC27L9 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5$  V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27L4 and TLC27L9 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1 \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27L4 and TLC27L9 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27L4 and TLC27L9 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50 \text{ k}\Omega$ , since bipolar devices exhibit greater noise currents.

APPLICATION INFORMATION

noise performance (continued)

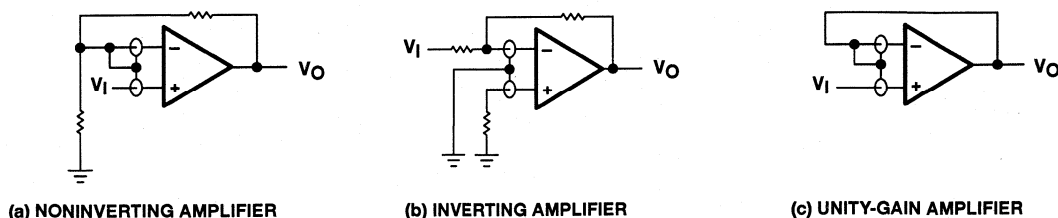


Figure 40. Guard-Ring Schemes

output characteristics

The output stage of the TLC27L4 and TLC27L9 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27L4 and TLC27L9 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

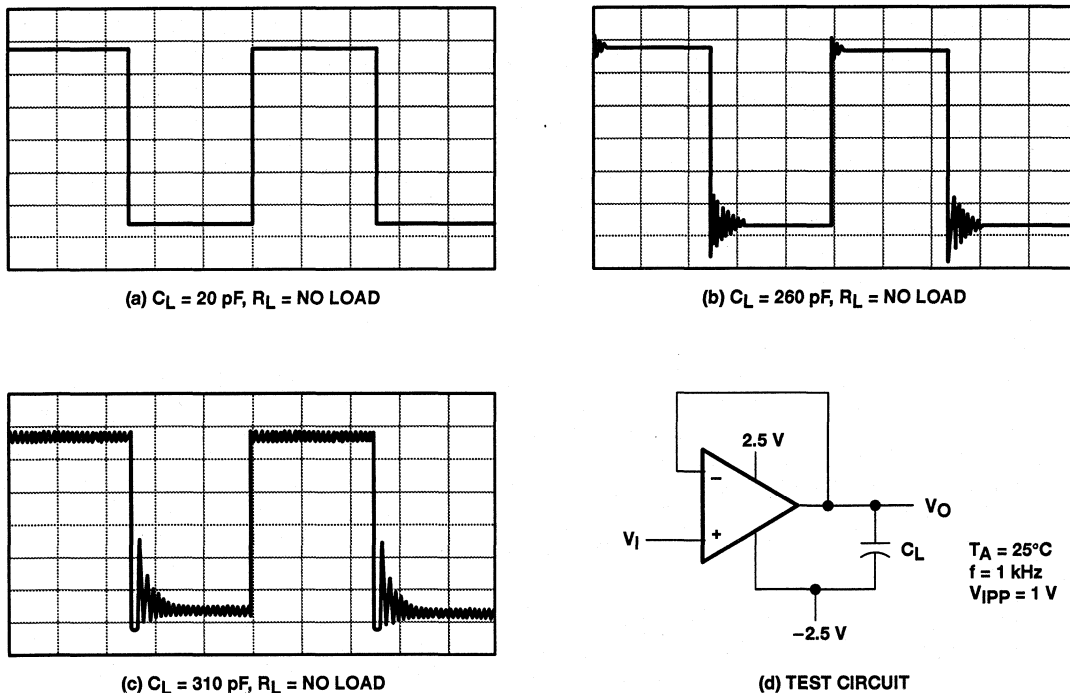


Figure 41. Effect of Capacitive Loads and Test Circuit

APPLICATION INFORMATION

output characteristics (continued)

Although the TLC27L4 and TLC27L9 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_p$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60\ \Omega$  and  $180\ \Omega$ , depending on how hard the operational amplifier input is driven. With very low values of  $R_p$ , a voltage offset from 0 V at the output occurs. Second, pullup resistor  $R_p$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

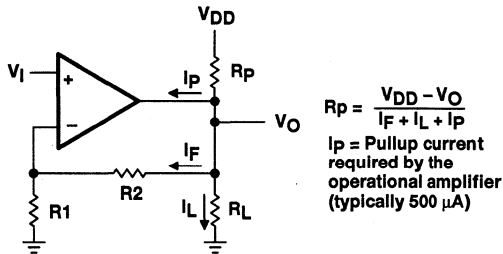


Figure 42. Resistive Pullup to Increase  $V_{OH}$

$$R_p = \frac{V_{DD} - V_O}{I_F + I_L + I_p}$$

$I_p$  = Pullup current required by the operational amplifier (typically  $500\ \mu A$ )

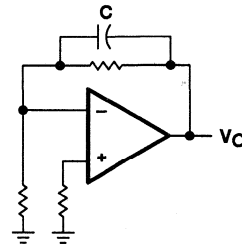


Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic discharge protection

The TLC27L4 and TLC27L9 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27L4 and TLC27L9 inputs and outputs were designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ( $0.1\ \mu F$  typical) located across the supply rails as close to the device as possible.

# TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### latch-up (continued)

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

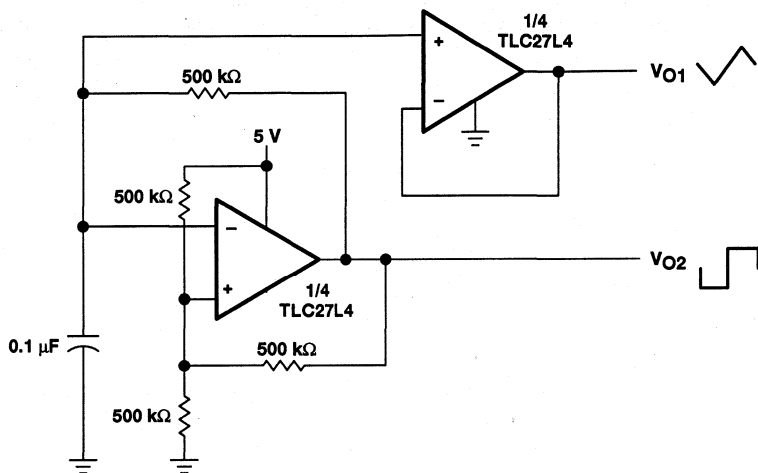
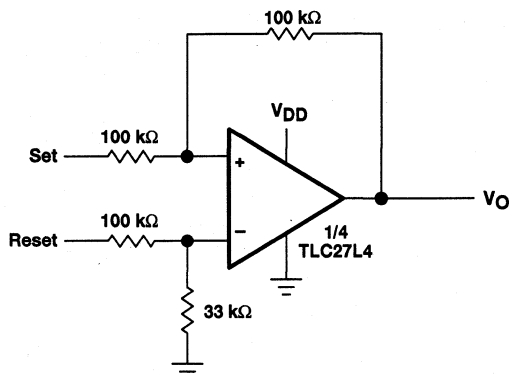


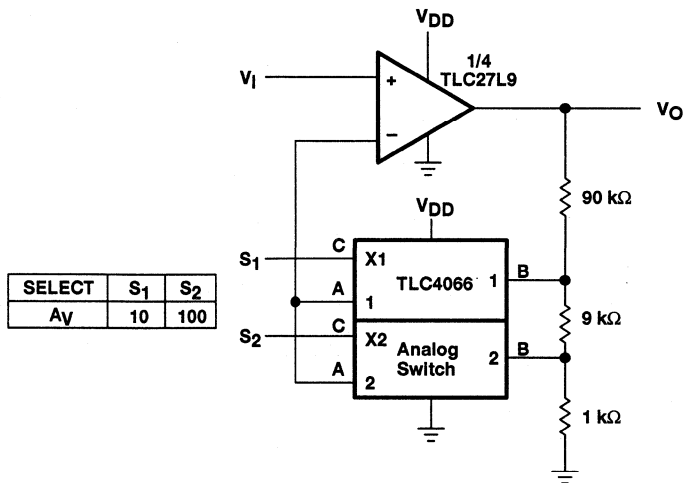
Figure 44. Multivibrator



NOTE:  $V_{DD} = 5\text{ V to }16\text{ V}$

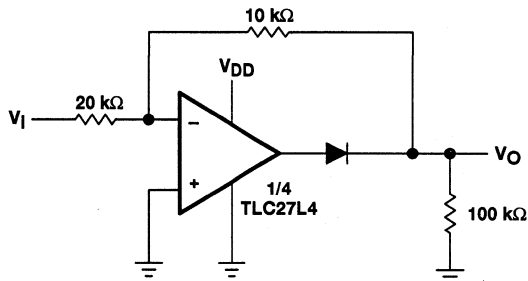
Figure 45. Set/Reset Flip-Flop

APPLICATION INFORMATION



NOTE: V<sub>DD</sub> = 5 V to 12 V

Figure 46. Amplifier With Digital Gain Selection



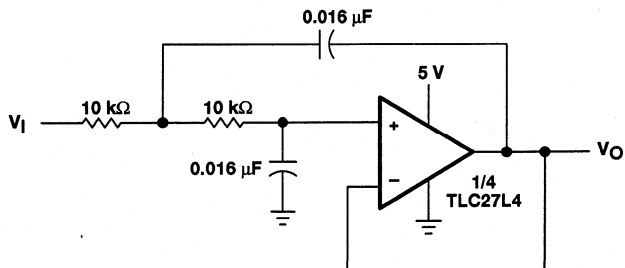
NOTE: V<sub>DD</sub> = 5 V to 16 V

Figure 47. Full-Wave Rectifier

**TLC27L4, TLC27L4A, TLC27L4B, TLC27L4Y, TLC27L9**  
**LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS**

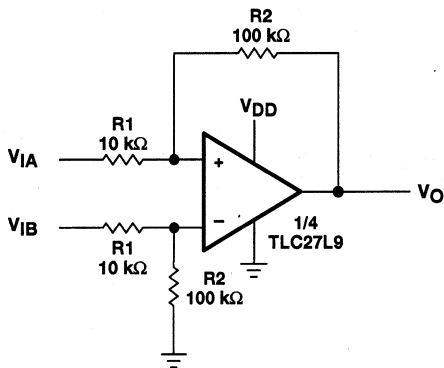
SLOS053C – OCTOBER 1987 – REVISED AUGUST 1994

**APPLICATION INFORMATION**



NOTE: Normalized to  $F_C = 1 \text{ kHz}$  and  $R_L = 10 \text{ k}\Omega$

**Figure 48. Two-Pole Low-Pass Butterworth Filter**



NOTE:  $V_{DD} = 5 \text{ V to } 16 \text{ V}$

$$V_O = \frac{R_2}{R_1}(V_{IB} - V_{IA})$$

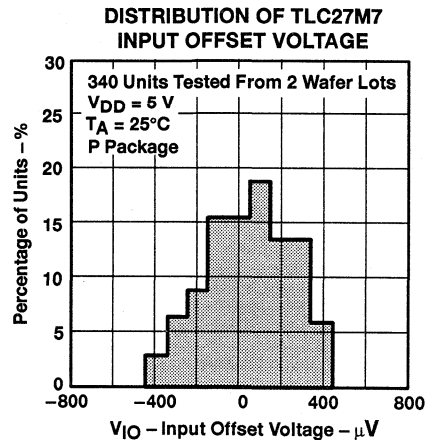
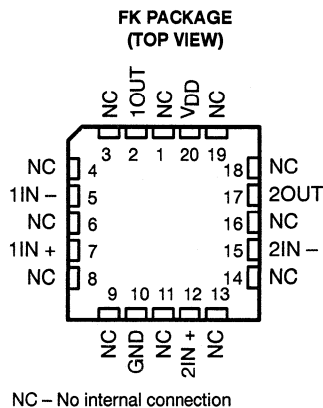
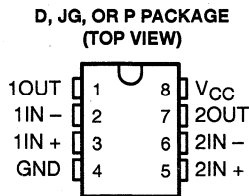
**Figure 49. Difference Amplifier**



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC27M7 . . . 500  $\mu\text{V}$  Max at 25°C,  
 $V_{\text{DD}} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Ranges:**  
0°C to 70°C . . . 3 V to 16 V  
–40°C to 85°C . . . 4 V to 16 V  
–55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)**
- **Low Noise . . . Typically 32 nV/ $\sqrt{\text{Hz}}$  at  $f = 1\text{ kHz}$**
- **Low Power . . . Typically 2.1 mW at 25°C,  $V_{\text{DD}} = 5\text{ V}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . .  $10^{12}\ \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**



## AVAILABLE OPTIONS

$T_A$	$V_{\text{IOmax}}$ AT 25°C	PACKAGE			
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	500 $\mu\text{V}$	TLC27M7CD	—	—	TLC27M7CP
	2 mV	TLC27M2BCD	—	—	TLC27M2BCP
	5 mV	TLC27M2ACD	—	—	TLC27M2ACP
	10 mV	TLC27M2CD	—	—	TLC27M2CP
–40°C to 85°C	500 $\mu\text{V}$	TLC27M7ID	—	—	TLC27M7IP
	2 mV	TLC27M2BID	—	—	TLC27M2BIP
	5 mV	TLC27M2AID	—	—	TLC27M2AIP
	10 mV	TLC27M2ID	—	—	TLC27M2IP
–55°C to 125°C	500 $\mu\text{V}$	TLC27M7MD	TLC27M7MFK	TLC27M7MJG	TLC27M7MP
	10 mV	TLC27M2MD	TLC27M2MFK	TLC27M2MJG	TLC27M2MP

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC27M7CDR).

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2–693

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

---

## description

The TLC27M2 and TLC27M7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M2 (10 mV) to the high-precision TLC27M7 (500  $\mu$ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M2 and TLC27M7. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand  $-100$ -mA surge currents without sustaining latch-up.

The TLC27M2 and TLC27M7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

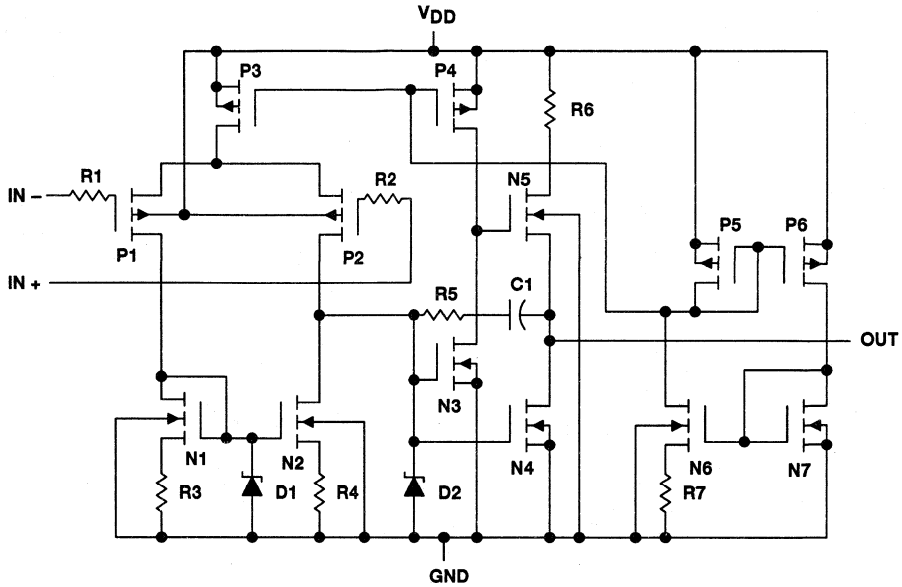
The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from  $-40$ °C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of  $-55$ °C to 125°C.



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

equivalent schematic (each amplifier)



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to network ground.
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
  3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M2C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_I = 100\ \text{k}\Omega$	25°C	1.1 10		mV
					Full range	12		
		TLC27M2AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_I = 100\ \text{k}\Omega$	25°C	0.9 5		
					Full range	6.5		
	TLC27M2BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_I = 100\ \text{k}\Omega$	25°C	220	2000	$\mu\text{V}$	
				Full range	3000			
	TLC27M7C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_I = 100\ \text{k}\Omega$	25°C	185	500		
				Full range	1500			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 100\ \text{k}\Omega$	25°C	3.2	3.9	V	
				0°C	3	3.9		
				70°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0 50		mV	
				0°C	0 50			
				70°C	0 50			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$ ,	$R_L = 100\ \text{k}\Omega$	25°C	25	170	V/mV	
				0°C	15	200		
				70°C	15	140		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				0°C	60	91		
				70°C	60	92		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$ ,	25°C	210	560	$\mu\text{A}$	
				0°C	250	640		
				70°C	170	440		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M2C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1 10		mV
					Full range	12		
		TLC27M2AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9 5		
					Full range	6.5		
	TLC27M2BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	224	2000	$\mu\text{V}$	
				Full range	3000			
	TLC27M7C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	190	800		
				Full range	1900			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				0°C	7.8	8.7		
				70°C	7.8	8.7		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				0°C	15	320		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	94		
				70°C	60	94		
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	285	600	$\mu\text{A}$	
				0°C	345	800		
				70°C	220	560		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	TLC27M2I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
			Full range		13	
	TLC27M2AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	mV
			Full range		7	
	TLC27M2BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	220	2000	$\mu\text{V}$
			Full range		3500	
	TLC27M7I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	185	500	$\mu\text{V}$
			Full range		2000	
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage		25°C to 85°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
		85°C	24	1000		
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
		85°C	200	2000		
$V_{ICR}$ Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	V	
		Full range	-0.2 to 3.5		V	
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
		-40°C	3	3.9		
		85°C	3	4		
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
		-40°C	0	50		
		85°C	0	50		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ , $R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
		-40°C	15	270		
		85°C	15	130		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	65	91	dB	
		-40°C	60	90		
		85°C	60	90		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	70	93	dB	
		-40°C	60	91		
		85°C	60	94		
$I_{DD}$ Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$	25°C	210	560	$\mu\text{A}$	
		-40°C	315	800		
		85°C	160	400		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I		UNIT	
					MIN	TYP		MAX
$V_{IO}$	Input offset voltage	TLC27M2I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M2AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		7	
	TLC27M2BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	224	2000	$\mu\text{V}$	
				Full range		3500		
	TLC27M7I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	190	800		
				Full range		2900		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	200 0		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				-40°C	7.8	8.7		
				85°C	7.8	8.7		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0 50	mV	
				-40°C		0 50		
				85°C		0 50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				-40°C	15	390		
				85°C	15	220		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-40°C	60	93		
				85°C	60	94		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-40°C	60	91		
				85°C	60	94		
$I_{DD}$	Supply current	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	285	600	$\mu\text{A}$	
				-40°C	450	900		
				85°C	205	520		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M2M TLC27M7M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M2M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1		mV
					Full range	12		
		TLC27M7M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	185	500	
					Full range	3750		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		v	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				-55°C	3	3.9		
				125°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ , $R_L = 100\text{ k}\Omega$		25°C	25	170	V/mV	
				-55°C	15	290		
				125°C	15	120		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				-55°C	60	89		
				125°C	60	91		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$ , $V_O = 1.4\text{ V}$		25°C	70	93	dB	
				-55°C	60	91		
				125°C	60	94		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	210	560	$\mu\text{A}$	
				-55°C	340	880		
				125°C	140	360		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M2M TLC27M7M		UNIT	
					MIN	TYP		MAX
$V_{IO}$	Input offset voltage	TLC27M2M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M7M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ ,	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	190	800	
					Full range		4300	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				-55°C	7.8	8.6		
				125°C	7.8	8.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				-55°C	15	420		
				125°C	15	190		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-55°C	60	93		
				125°C	60	93		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-55°C	60	91		
				125°C	60	94		
$I_{DD}$	Supply current (two amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	285	600	$\mu\text{A}$	
				-55°C	490	1000		
				125°C	180	480		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(PP) = 1\text{ V}$	25°C	0.43			V/ $\mu$ s
			0°C	0.46			
			70°C	0.36			
		$V_I(PP) = 2.5\text{ V}$	25°C	0.40			
			0°C	0.43			
			70°C	0.34			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	55			kHz
			0°C	60			
			70°C	50			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	525			kHz
			0°C	600			
			70°C	400			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	40°			
			0°C	41°			
			70°C	39°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_I(PP) = 1\text{ V}$	25°C	0.62			V/ $\mu$ s
			0°C	0.67			
			70°C	0.51			
		$V_I(PP) = 5.5\text{ V}$	25°C	0.56			
			0°C	0.61			
			70°C	0.46			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	25°C	35			kHz
			0°C	40			
			70°C	30			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	635			kHz
			0°C	710			
			70°C	510			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	25°C	43°			
			0°C	44°			
			70°C	42°			

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7

## LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

### operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ $\mu$ s	
			-40°C	0.51			
			85°C	0.35			
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40			
			-40°C	0.48			
			85°C	0.32			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	55		kHz	
			-40°C	75			
			85°C	45			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	525		MHz	
			-40°C	770			
			85°C	370			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	40°			
			-40°C	43°			
			85°C	38°			

### operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ $\mu$ s	
			-40°C	0.77			
			85°C	0.47			
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56			
			-40°C	0.70			
			85°C	0.44			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1		25°C	35		kHz	
			-40°C	45			
			85°C	25			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	25°C	635		MHz	
			-40°C	880			
			85°C	480			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	43°			
			-40°C	46°			
			85°C	41°			



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2M TLC27M7M			UNIT
				MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ $\mu$ s	
			-55°C	0.54			
			125°C	0.29			
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40			
			-55°C	0.49			
			125°C	0.28			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ ,	$C_L = 20\text{ pF}$ , See Figure 1	25°C	55		kHz	
			-55°C	80			
			125°C	40			
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	525		kHz	
			-55°C	850			
			125°C	330			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	40°			
			-55°C	44°			
			125°C	36°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M2M TLC27M7M			UNIT
				MIN	TYP	MAX	
SR    Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ $\mu$ s	
			-55°C	0.81			
			125°C	0.38			
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56			
			-55°C	0.73			
			125°C	0.35			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$ ,	$C_L = 20\text{ pF}$ , See Figure 1	25°C	35		kHz	
			-55°C	50			
			125°C	20			
$B_1$ Unity gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$ ,	25°C	635		kHz	
			-55°C	960			
			125°C	440			
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 3	25°C	43°			
			-55°C	47°			
			125°C	39°			

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## PARAMETER MEASUREMENT INFORMATION

### single-supply versus split-supply test circuits

Because the TLC27M2 and TLC27M7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

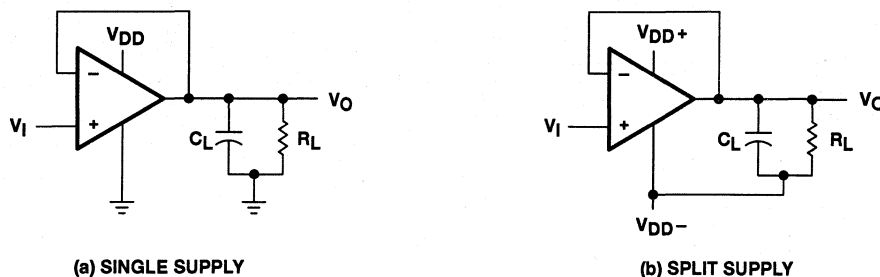


Figure 1. Unity-Gain Amplifier

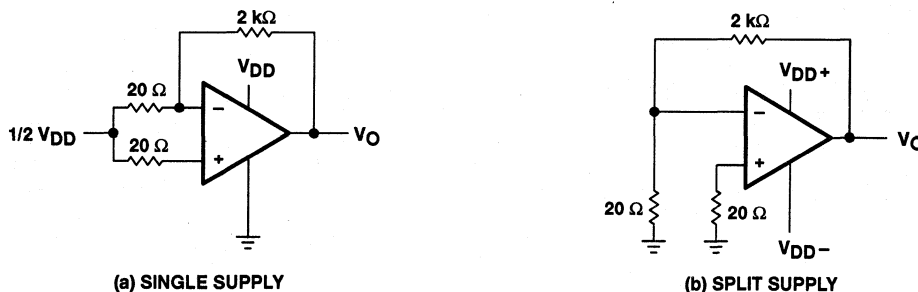


Figure 2. Noise-Test Circuit

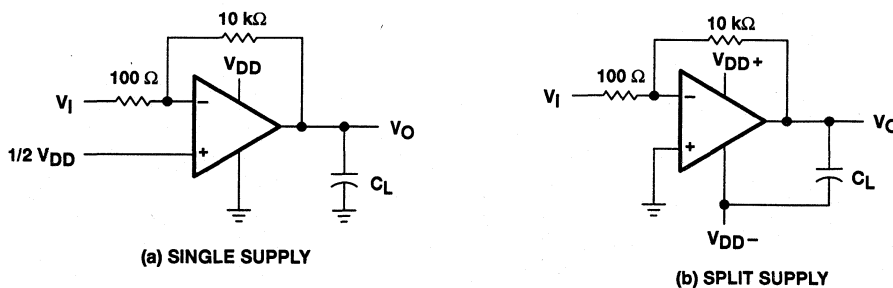


Figure 3. Gain-of-100 Inverting Amplifier

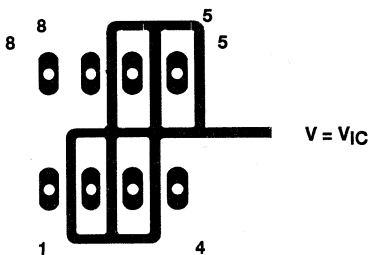
## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC27M2 and TLC27M7 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution ... many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.



**Figure 4. Isolation Metal Around Device Inputs  
(JG and P packages)**

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

## PARAMETER MEASUREMENT INFORMATION

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

### full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

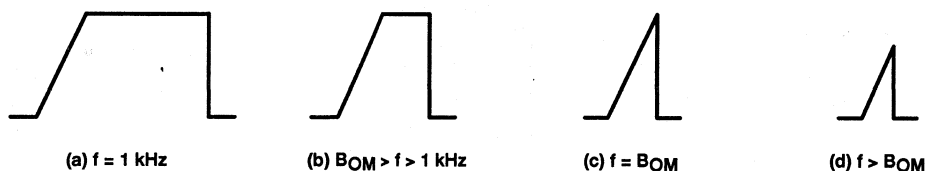


Figure 5. Full-Power-Response Output Signal

### test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS

**Table of Graphs**

			FIGURE
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha_{VIO}$	Temperature coefficient	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current	10, 11
		vs Supply voltage	12
		vs Free-air temperature	13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	14, 15
		vs Differential input voltage	16
		vs Free-air temperature	17
		vs Low-level output current	18, 19
$A_{VD}$	Differential voltage amplification	vs Supply voltage	20
		vs Free-air temperature	21
		vs Frequency	32, 33
$I_{IB}/I_{IO}$	Input bias and input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage	24
		vs Free-air temperature	25
SR	Slew rate	vs Supply voltage	26
		vs Free-air temperature	27
		Normalized slew rate	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature	30
		vs Supply voltage	31
$\phi_m$	Phase margin	vs Supply voltage	34
		vs Free-air temperature	35
		vs Capacitive loads	36
$V_n$	Equivalent input noise voltage	vs Frequency	37
$\phi$	Phase shift	vs Frequency	32, 33

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## TYPICAL CHARACTERISTICS

**DISTRIBUTION OF TLC27M2  
INPUT OFFSET VOLTAGE**

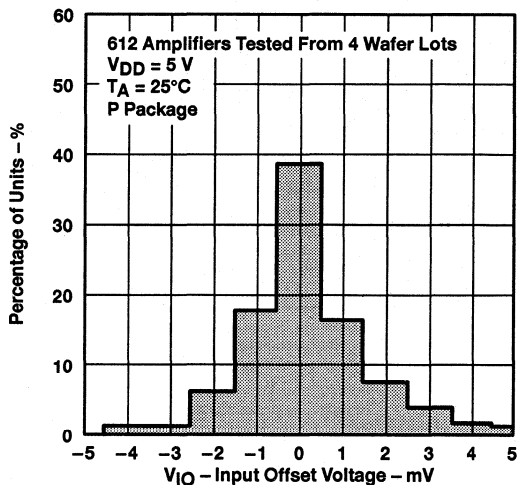


Figure 6

**DISTRIBUTION OF TLC27M2  
INPUT OFFSET VOLTAGE**

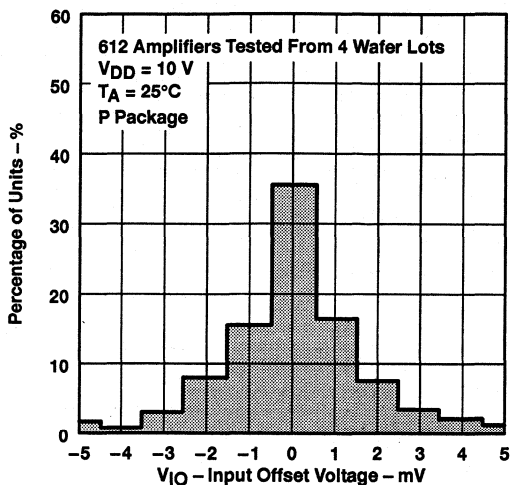


Figure 7

**DISTRIBUTION OF TLC27M2 AND TLC27M7  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

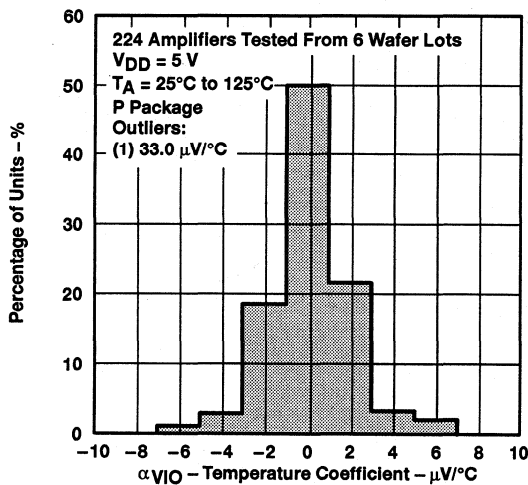


Figure 8

**DISTRIBUTION OF TLC27M2 AND TLC27M7  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

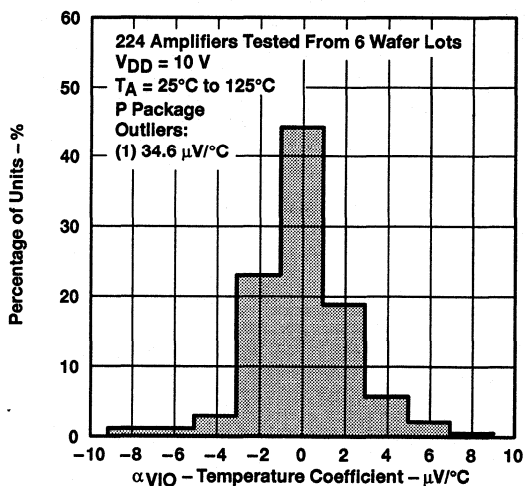


Figure 9



# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

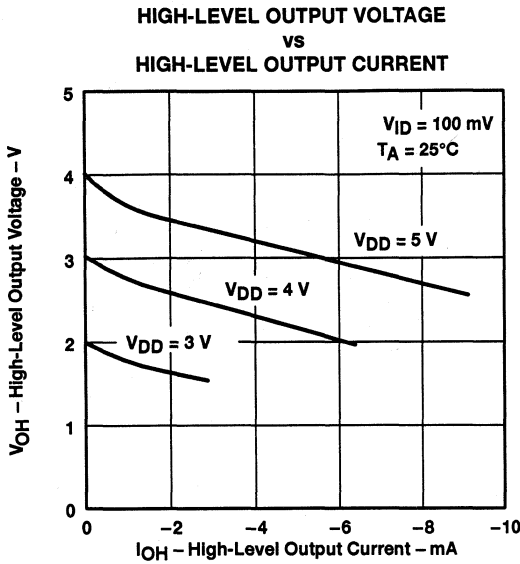


Figure 10

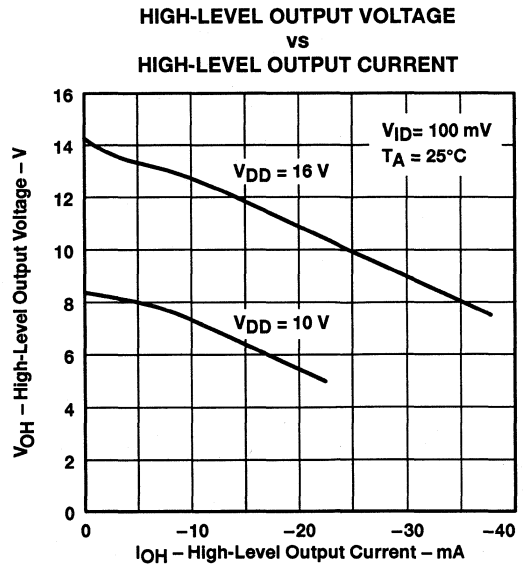


Figure 11

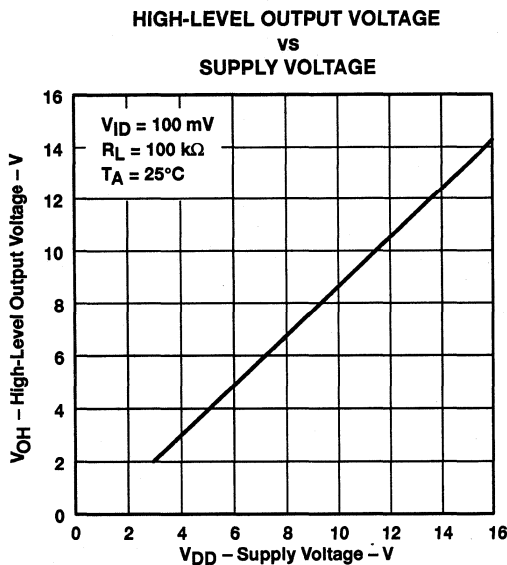


Figure 12

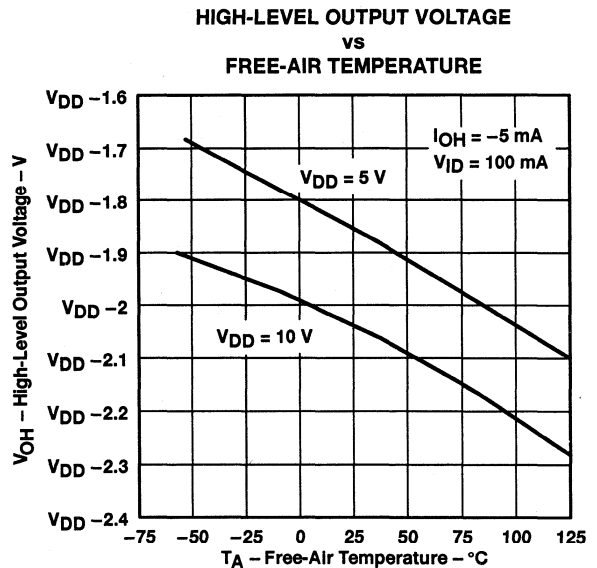


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## TYPICAL CHARACTERISTICS†

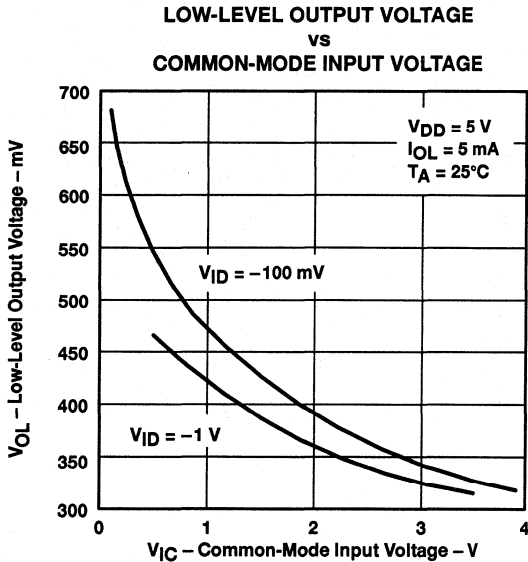


Figure 14

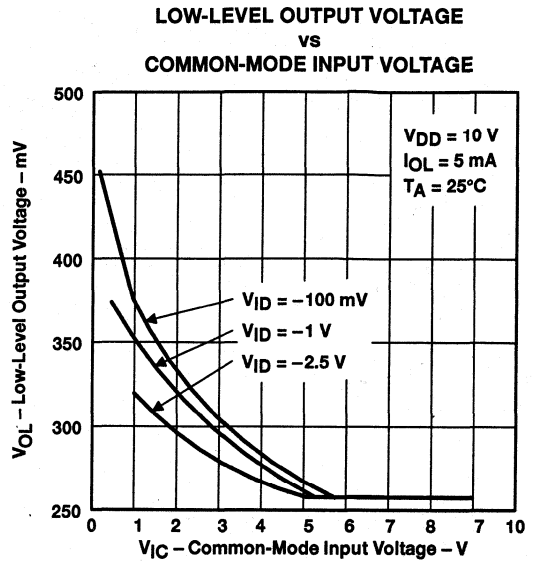


Figure 15

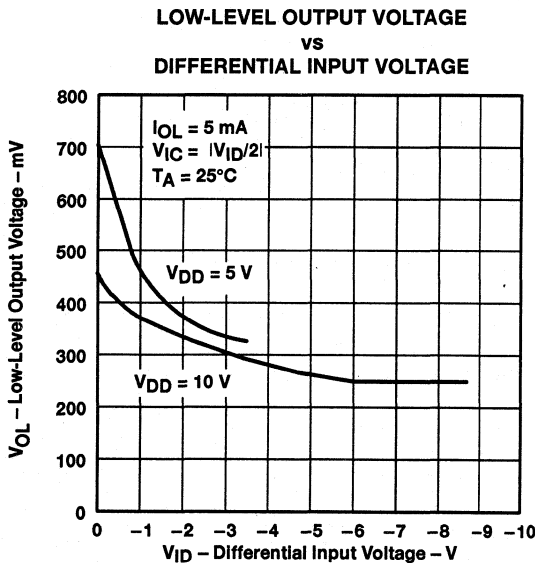


Figure 16

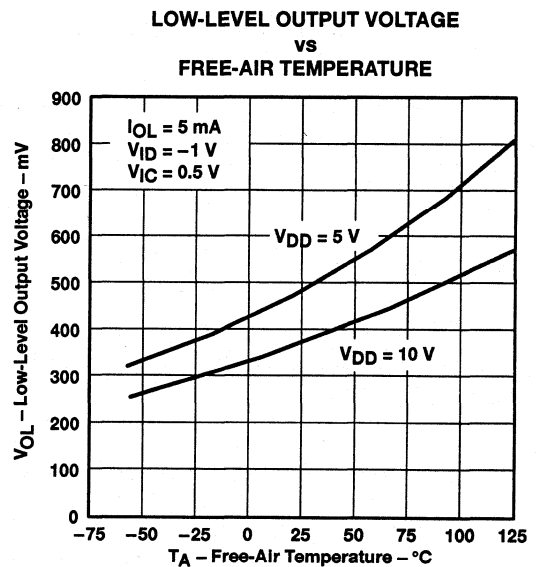


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

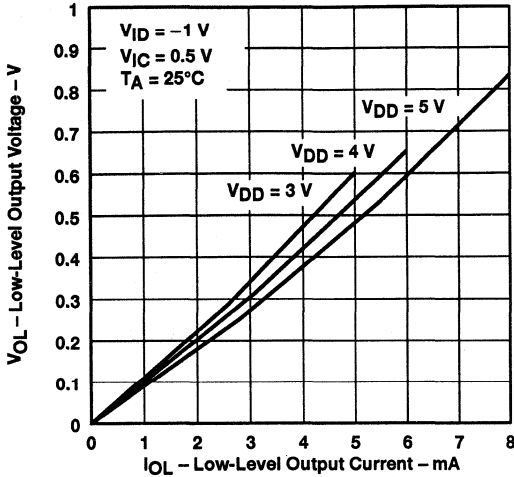


Figure 18

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

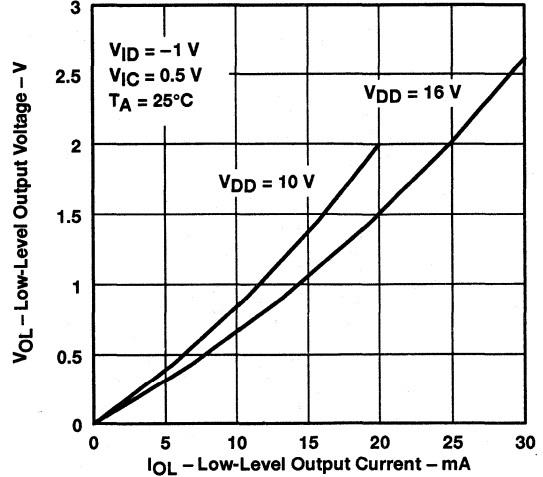


Figure 19

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 SUPPLY VOLTAGE

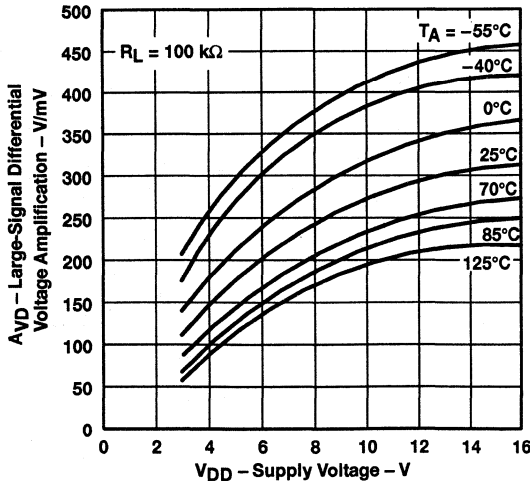


Figure 20

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

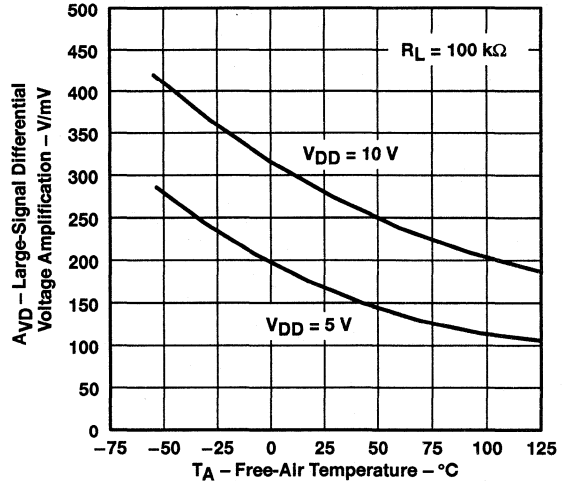


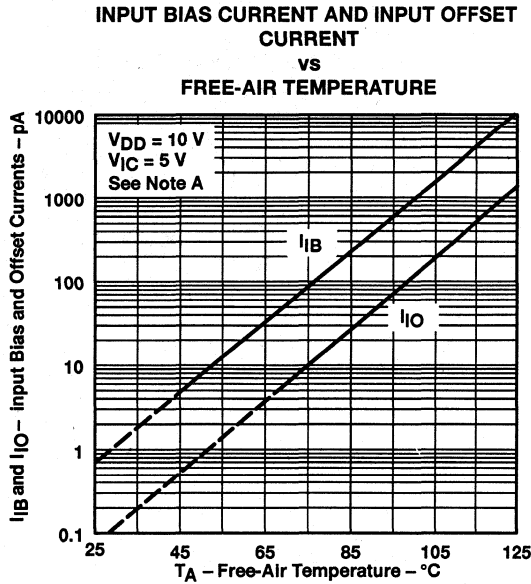
Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## TYPICAL CHARACTERISTICS†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

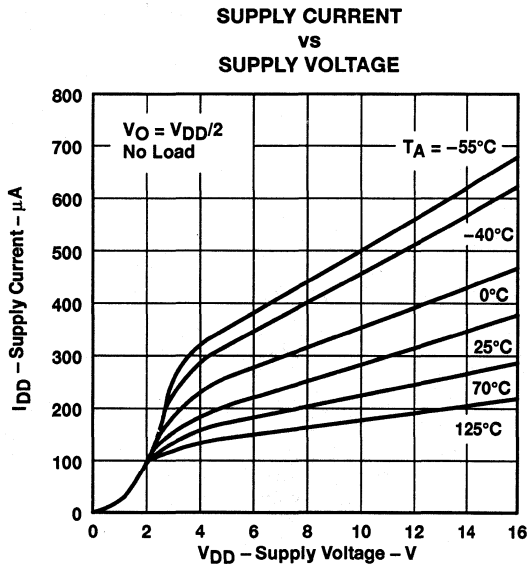


Figure 24

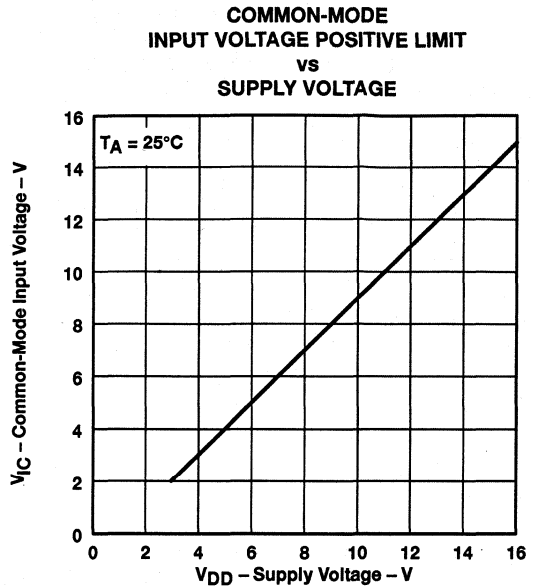


Figure 23

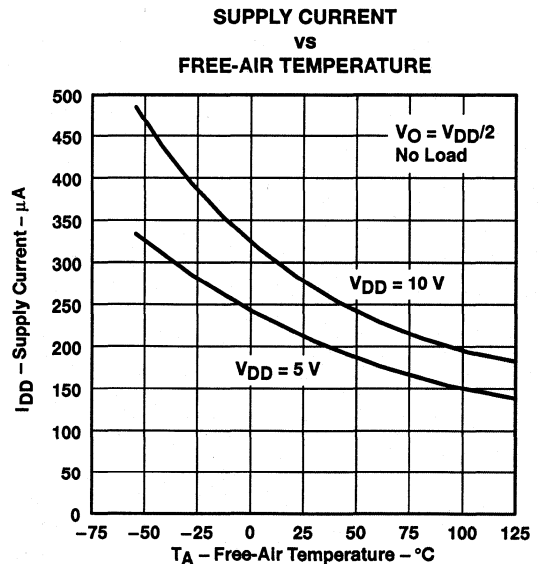


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

SLEW RATE  
 VS  
 SUPPLY VOLTAGE

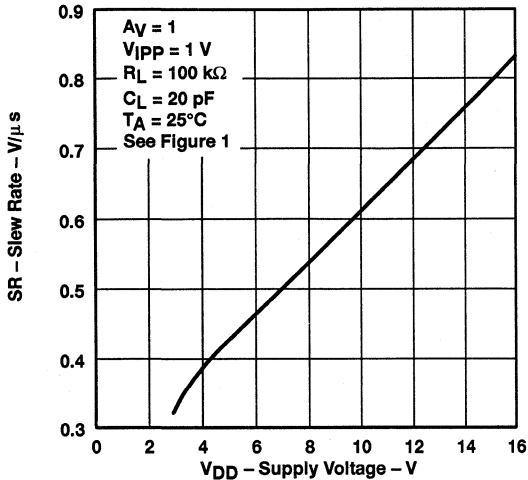


Figure 26

SLEW RATE  
 VS  
 FREE-AIR TEMPERATURE

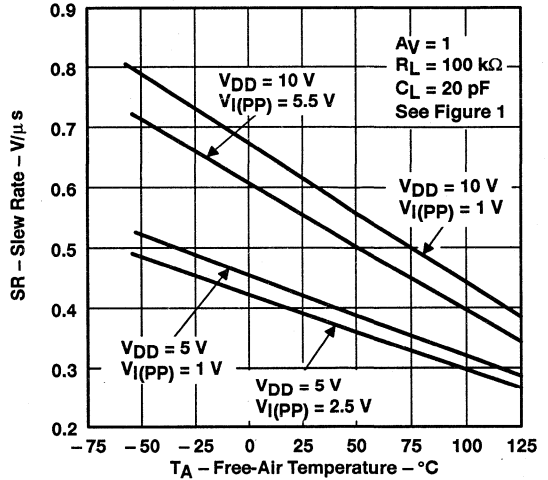


Figure 27

NORMALIZED SLEW RATE  
 VS  
 FREE-AIR TEMPERATURE

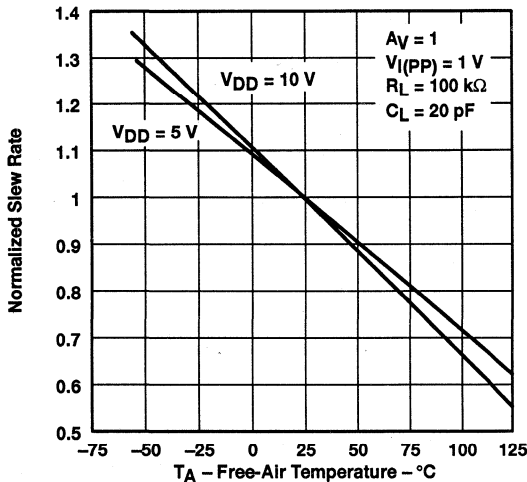


Figure 28

MAXIMUM PEAK-TO-PEAK OUTPUT  
 VOLTAGE  
 VS  
 FREQUENCY

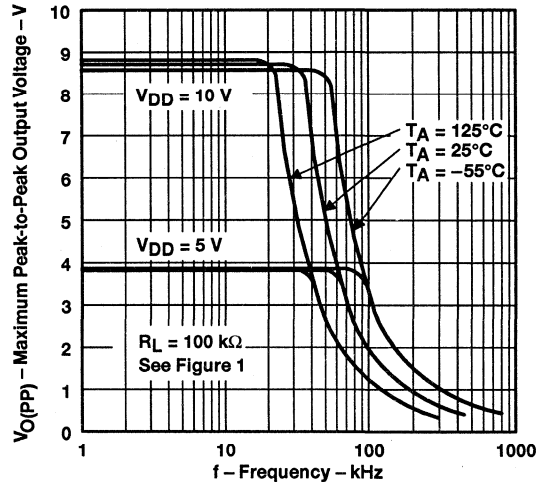


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## TYPICAL CHARACTERISTICS†

UNITY-GAIN BANDWIDTH  
vs  
FREE-AIR TEMPERATURE

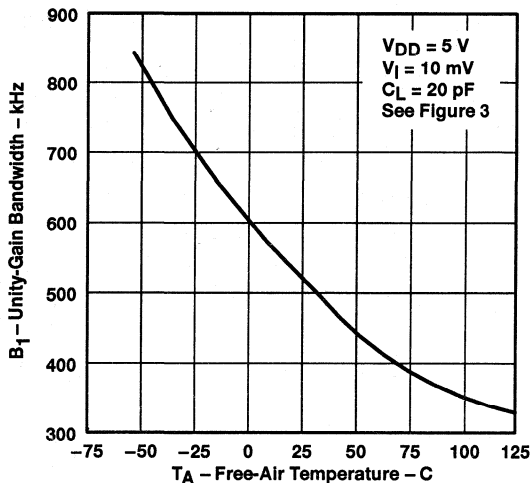


Figure 30

UNITY-GAIN BANDWIDTH  
vs  
SUPPLY VOLTAGE

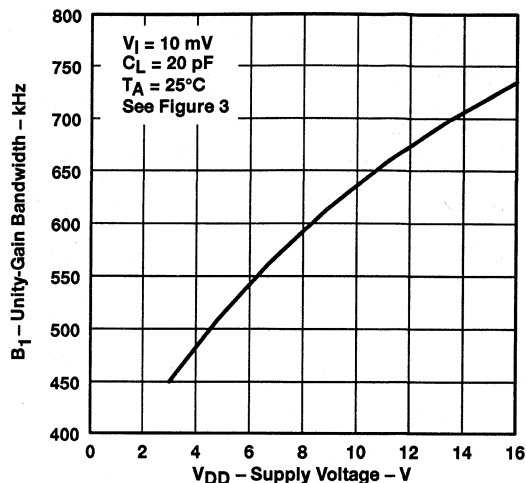


Figure 31

LARGE-SCALE DIFFERENTIAL VOLTAGE  
AMPLIFICATION AND PHASE SHIFT  
vs  
FREQUENCY

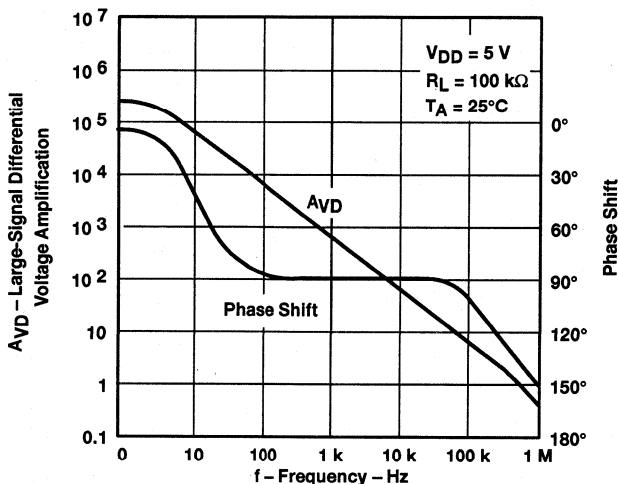


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LARGE-SCALE DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

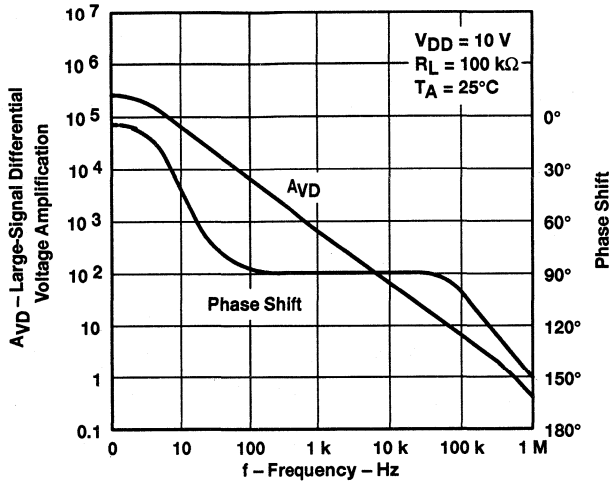


Figure 33

PHASE MARGIN  
 vs  
 SUPPLY VOLTAGE

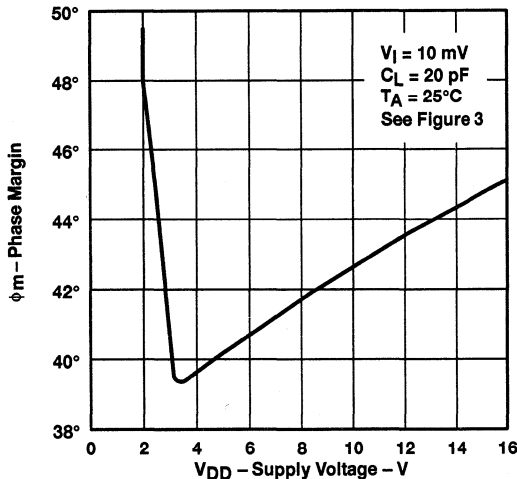


Figure 34

PHASE MARGIN  
 vs  
 FREE-AIR TEMPERATURE

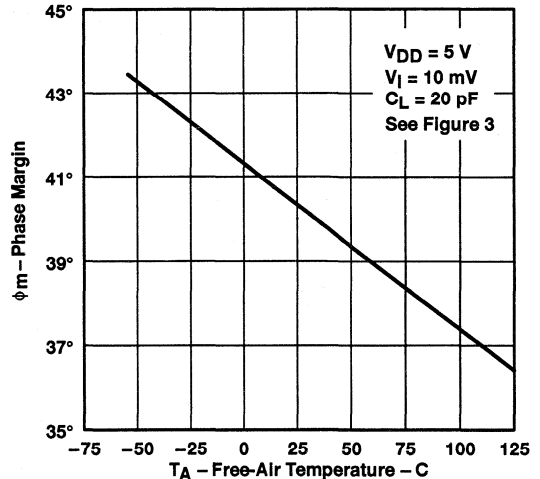


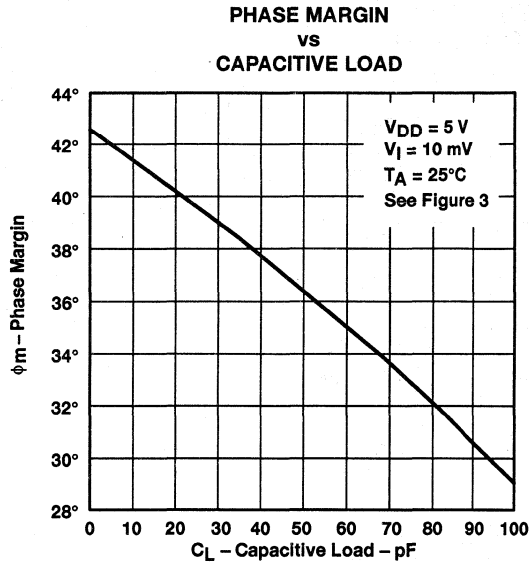
Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

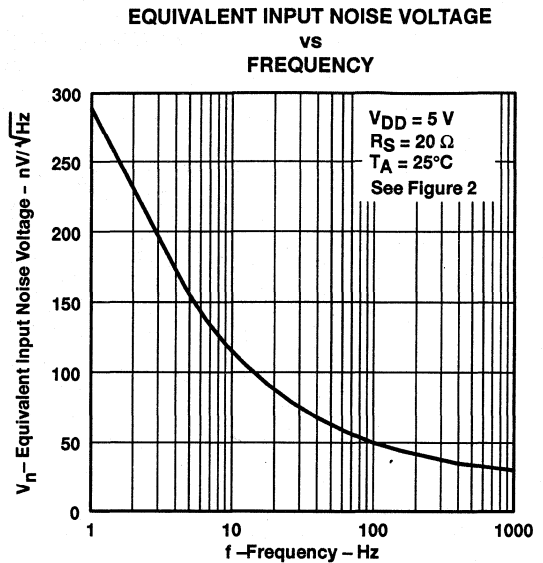
**TLC27M2, TLC27M2A, TLC27M2B, TLC27M7**  
**LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS**

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

**TYPICAL CHARACTERISTICS**



**Figure 36**



**Figure 37**

APPLICATION INFORMATION

single-supply operation

While the TLC27M2 and TLC27M7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M2 and TLC27M7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M2 and TLC27M7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

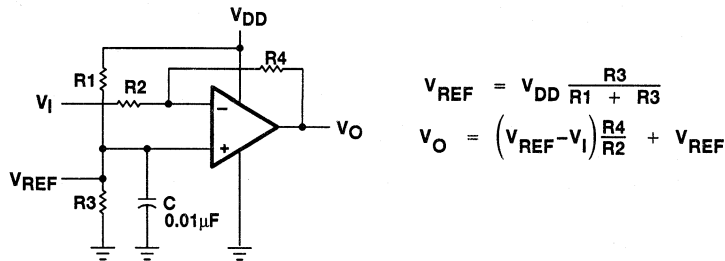


Figure 38. Inverting Amplifier With Voltage Reference

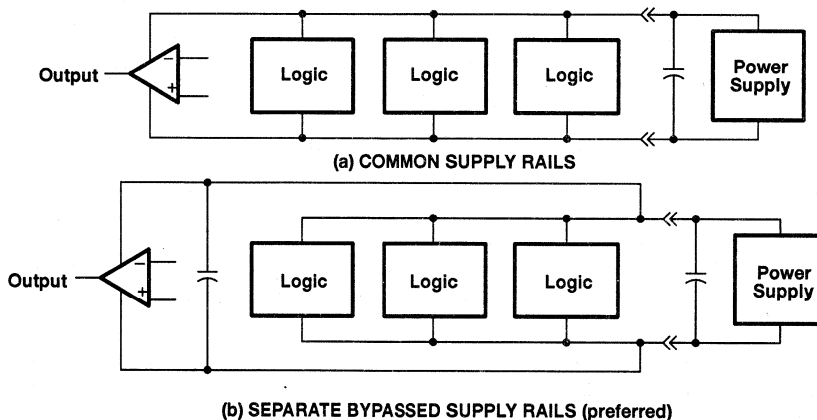


Figure 39. Common Versus Separate Supply Rails

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7

## LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

### APPLICATION INFORMATION

#### input characteristics

The TLC27M2 and TLC27M7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5$  V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M2 and TLC27M7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1\ \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M2 and TLC27M7 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

#### noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M2 and TLC27M7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50\ \text{k}\Omega$ , since bipolar devices exhibit greater noise currents.

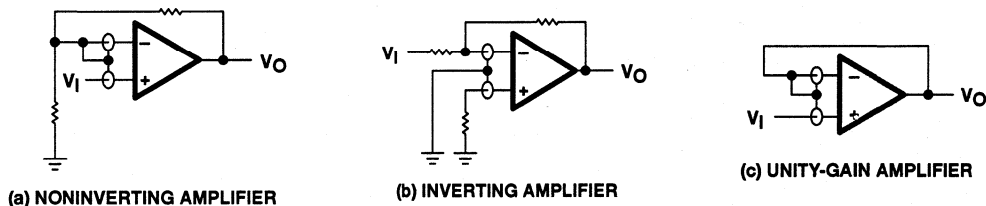


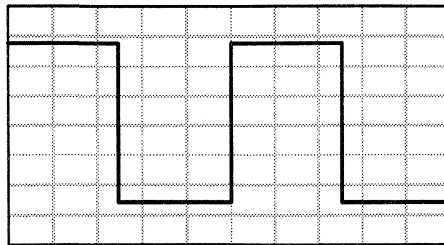
Figure 40. Guard-Ring Schemes

#### output characteristics

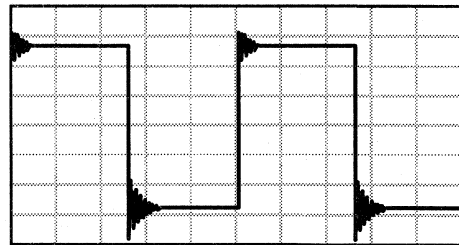
The output stage of the TLC27M2 and TLC27M7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M2 and TLC27M7 were measured using a  $20\text{-pF}$  load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

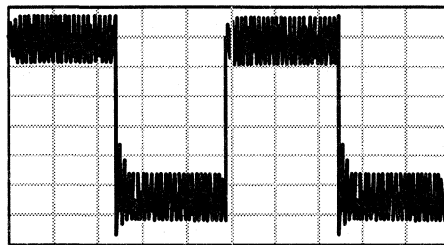
APPLICATION INFORMATION



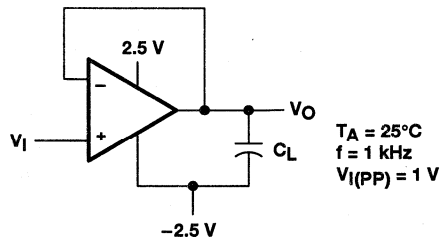
(a)  $C_L = 20 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(b)  $C_L = 170 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(c)  $C_L = 190 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(d) TEST CIRCUIT

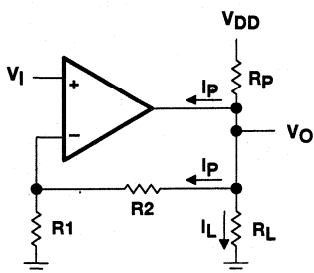
Figure 41. Effect of Capacitive Loads and Test Circuit

output characteristics (continued)

Although the TLC27M2 and TLC27M7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60 \Omega$  and  $180 \Omega$ , depending on how hard the op amp input is driven. With very low values of  $R_P$ , a voltage offset from 0 V at the output occurs. Second, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

APPLICATION INFORMATION

output characteristics (continued)



$$R_P = \frac{V_{DD} - V_O}{I_F + I_L + I_P}$$

$I_P$  = Pullup current required by the operational amplifier (typically 500  $\mu$ A)

Figure 42. Resistive Pullup to Increase  $V_{OH}$

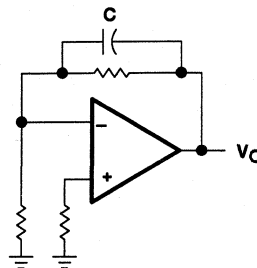


Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic-discharge protection

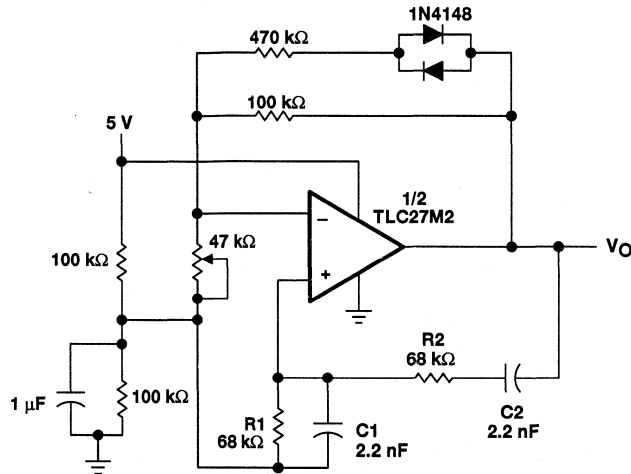
The TLC27M2 and TLC27M7 incorporate an internal electrostatic-discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27M2 and TLC27M7 inputs and outputs were designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1  $\mu$ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

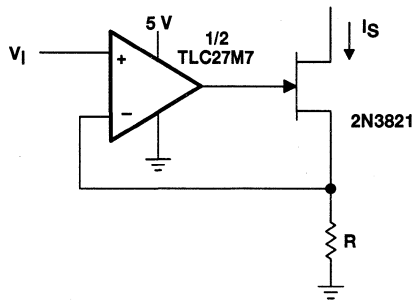
APPLICATION INFORMATION



NOTES:  $V_{O(PP)} \approx 2 V$

$$f_0 = \frac{1}{2\pi\sqrt{R_1 R_2 C_1 C_2}}$$

Figure 44. Wien Oscillator



NOTES:  $V_I = 0 V$  to  $3 V$

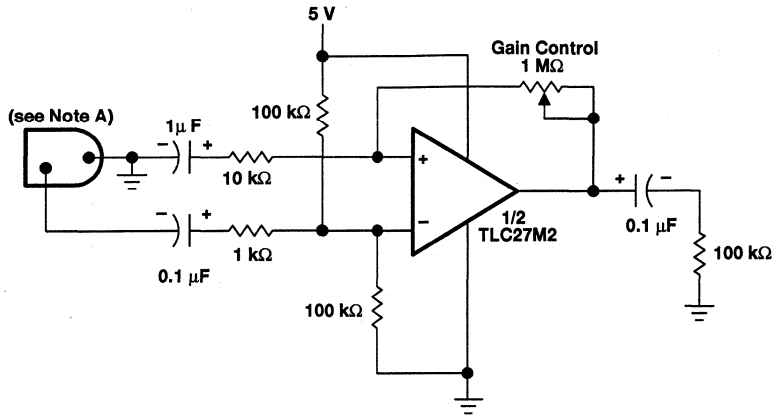
$$I_S = \frac{V_I}{R}$$

Figure 45. Precision Low-Current Sink

# TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

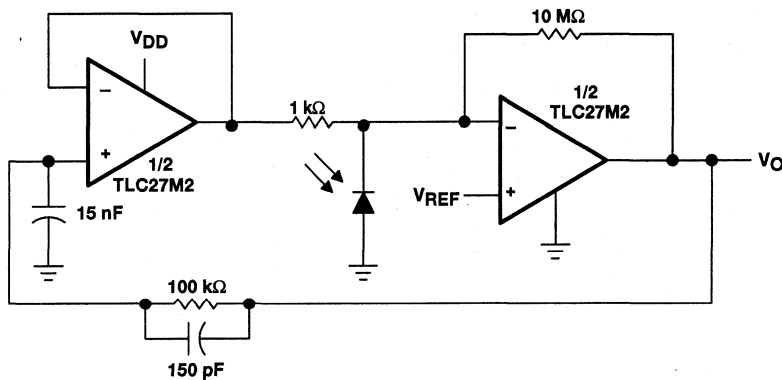
SLOS051B – OCTOBER 1987 – REVISED DECEMBER 1994

## APPLICATION INFORMATION



NOTE A: Low to medium impedance dynamic mike

Figure 46. Microphone Preamp

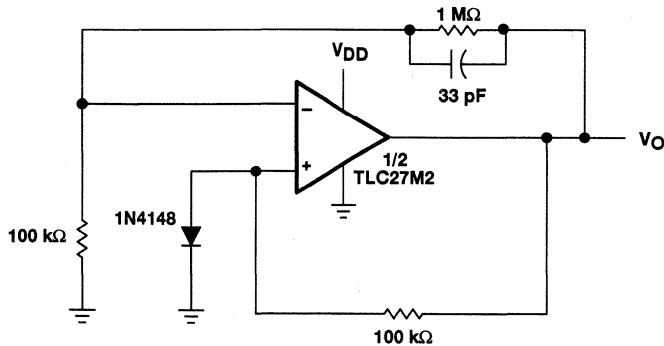


NOTES: V<sub>DD</sub> = 4 V to 15 V  
V<sub>ref</sub> = 0 V to V<sub>DD</sub> - 2 V

Figure 47. Photo-Diode Amplifier With Ambient Light Rejection



APPLICATION INFORMATION



NOTES:  $V_{DD} = 8\text{ V to }16\text{ V}$   
 $V_O = 5\text{ V, }10\text{ mA}$

Figure 48. 5-V Low-Power Voltage Regulator

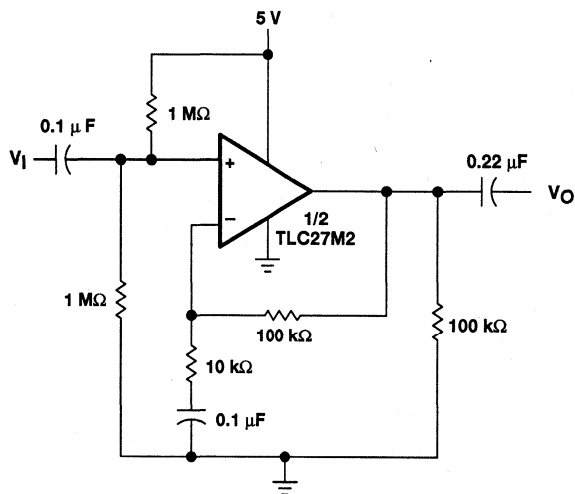


Figure 49. Single-Rail AC Amplifiers

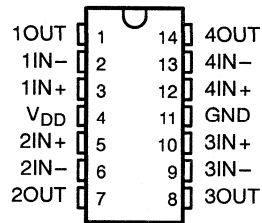


# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

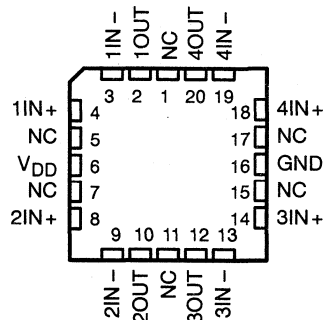
SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

- **Trimmed Offset Voltage:**  
TLC27M9 . . . 900  $\mu\text{V}$  Max at  $T_A = 25^\circ\text{C}$ ,  
 $V_{DD} = 5\text{ V}$
- **Input Offset Voltage Drift . . . Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
0°C to 70°C . . . 3 V to 16 V  
-40°C to 85°C . . . 4 V to 16 V  
-55°C to 125°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)**
- **Low Noise . . . Typically 32 nV/ $\sqrt{\text{Hz}}$  at  $f = 1\text{ kHz}$**
- **Low Power . . . Typically 2.1 mW at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = 5\text{ V}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . .  $10^{12}\ \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**

D, J, N, OR PW PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

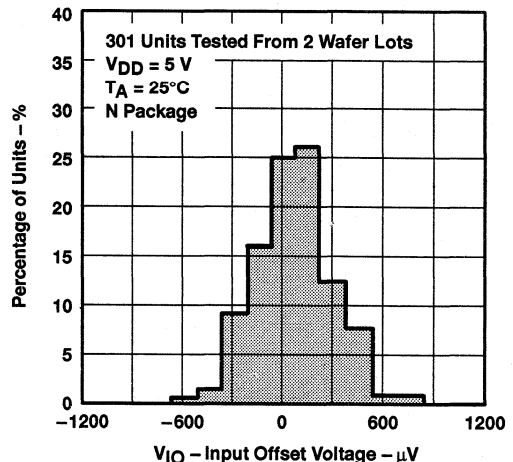
## description

The TLC27M4 and TLC27M9 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds comparable to that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, make these cost-effective devices ideal for applications that have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption.

Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M4 (10 mV) to the high-precision TLC27M9 (900  $\mu\text{V}$ ). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

DISTRIBUTION OF TLC27M9  
INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2-727

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## description (continued)

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M4 and TLC27M9. The devices also exhibit low voltage single-supply operation, and low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up.

The TLC27M4 and TLC27M9 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C.

### AVAILABLE OPTIONS

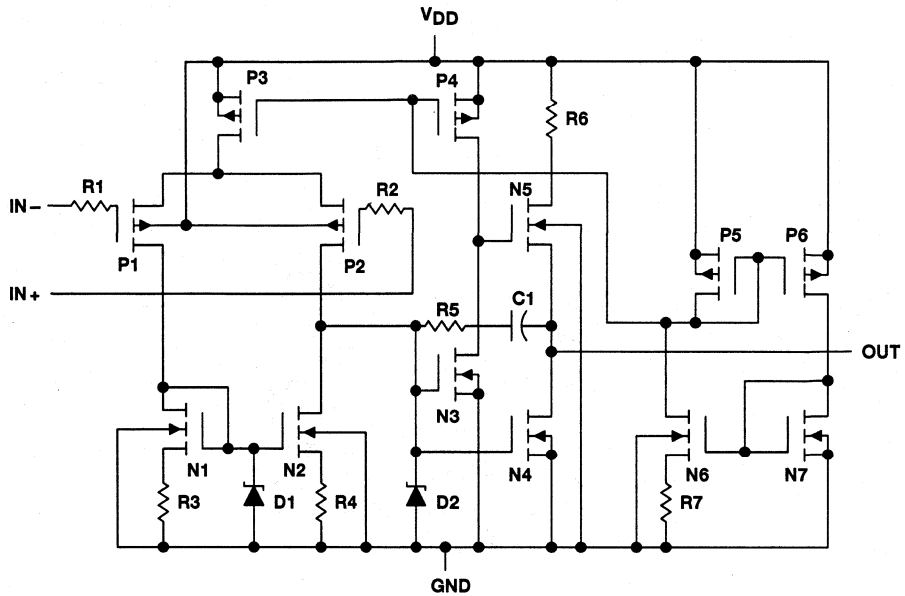
T <sub>A</sub>	V <sub>IOMAX</sub> AT 25°C	PACKAGE					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	900 μV	TLC27M9CD	—	—	TLC27M9CN	—	—
	2 mV	TLC27M4BCD	—	—	TLC27M4BCN	—	—
	5 mV	TLC27M4ACD	—	—	TLC27M4ACN	—	—
	10 mV	TLC27M4CD	—	—	TLC27M4CN	TLC27M4CPW	TLC27M4Y
–40°C to 85°C	900 μV	TLC27M9ID	—	—	TLC27M9IN	—	—
	2 mV	TLC27M4BID	—	—	TLC27M4BIN	—	—
	5 mV	TLC27M4AID	—	—	TLC27M4AIN	—	—
	10 mV	TLC27M4ID	—	—	TLC27M4IN	—	—
–55°C to 125°C	900 μV	TLC27M9MD	TLC27M9MFK	TLC27M9MJ	TLC27M9MN	—	—
	10 mV	TLC27M4MD	TLC27M4MFK	TLC27M4MJ	TLC27M4MN	—	—

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC279CDR).

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B - OCTOBER 1987 - REVISED AUGUST 1994

equivalent schematic (each amplifier)

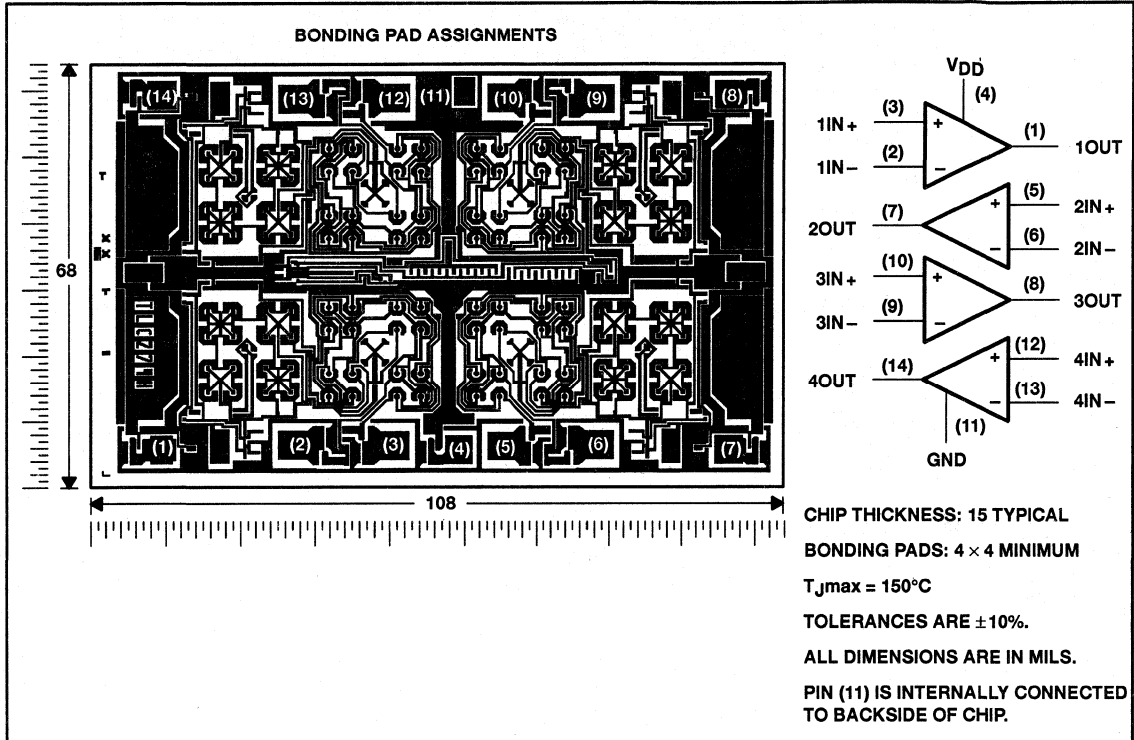


# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TLC27M4Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC27M4C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	—
PW	700 mW	5.6 mW/°C	448 mW	—	—

## recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		3	16	4	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	3.5	-0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	-0.2	8.5	-0.2	8.5	0	8.5	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC274BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	250	2000	$\mu\text{V}$	
				Full range		3000		
	TLC279C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	210	900		
				Full range		1500		
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				0°C	3	3.9		
				70°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
				0°C	15	200		
				70°C	15	140		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				0°C	60	91		
				70°C	60	92		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	420	1120	$\mu\text{A}$	
				0°C	500	1280		
				70°C	340	880		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.





# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M4AC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
TLC27M4BC	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	260	2000	$\mu\text{V}$		
			Full range		3000			
TLC27M9C	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	220	1200	$\mu\text{V}$		
			Full range		1900			
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 70°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$ ,	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ ,	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				0°C	7.8	8.7		
				70°C	7.8	8.7		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ ,	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ ,	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				0°C	15	320		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	94		
				70°C	60	94		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ ,	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$ ,	25°C	570	1200	$\mu\text{A}$	
				0°C	690	1600		
				70°C	440	1120		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M4AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC27M4BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	250	2000	$\mu\text{V}$
					Full range		3000	
		TLC27M9I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	210	900	$\mu\text{V}$
					Full range		2000	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 85°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				85°C	24	1000		
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				85°C	200	2000		
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				-40°C	3	3.9		
				85°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
				-40°C	15	270		
				85°C	15	130		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				-40°C	60	90		
				85°C	60	90		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to } 10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-40°C	60	91		
				85°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	420	1120	$\mu\text{A}$	
				-40°C	630	1600		
				85°C	320	800		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 10\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	1.1	10	mV
			Full range		13	
	TLC27M4AI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	0.9	5	mV
			Full range		7	
	TLC27M4BI	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	260	2000	$\mu\text{V}$
			Full range		3500	
	TLC27M9I	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\ \text{k}\Omega$	25°C	220	1200	$\mu\text{V}$
			Full range		2900	
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage		25°C to 85°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$ Input offset current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$	25°C	0.1		pA	
		85°C	26	1000		
$I_{IB}$ Input bias current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$	25°C	0.7		pA	
		85°C	220	2000		
$V_{ICR}$ Common-mode input voltage range (see Note 5)		25°C	-0.2 to 9	-0.3 to 9.2	V	
		Full range	-0.2 to 8.5		V	
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 100\ \text{k}\Omega$	25°C	8	8.7	V	
		-40°C	7.8	8.7		
		85°C	7.8	8.7		
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C	0	50	mV	
		-40°C	0	50		
		85°C	0	50		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$ , $R_L = 100\ \text{k}\Omega$	25°C	25	275	V/mV	
		-40°C	15	390		
		85°C	15	220		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	65	94	dB	
		-40°C	60	93		
		85°C	60	94		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	70	93	dB	
		-40°C	60	91		
		85°C	60	94		
$I_{DD}$ Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load $V_{IC} = 5\text{ V}$	25°C	570	1200	$\mu\text{A}$	
		-40°C	900	1800		
		85°C	410	1040		

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M4M TLC27M9M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M4M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M9M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	210	900	$\mu\text{V}$
					Full range		3750	
$\alpha V_{IO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	1.7		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				125°C	1.4	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 2.5\text{ V}$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				125°C	9	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 4	-0.3 to 4.2	V	
				Full range	0 to 3.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				-55°C	3	3.9		
				125°C	3	4		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C		0	50	
				-55°C		0	50	
				125°C		0	50	
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
				-55°C	15	290		
				125°C	15	120		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				-55°C	60	89		
				125°C	60	91		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-55°C	60	91		
				125°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	$V_{IC} = 2.5\text{ V}$	25°C	420	1120	$\mu\text{A}$	
				-55°C	680	1760		
				125°C	280	720		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A$ †	TLC27M4M TLC27M9M			UNIT
					MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	TLC27M4M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M9M	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$	25°C	220	1200	$\mu\text{V}$
					Full range		4300	
$\alpha_{VIO}$	Average temperature coefficient of input offset voltage			25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
$I_{IO}$	Input offset current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				125°C	1.8	15	nA	
$I_{IB}$	Input bias current (see Note 4)	$V_O = 5\text{ V}$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				125°C	10	35	nA	
$V_{ICR}$	Common-mode input voltage range (see Note 5)			25°C	0 to 9	-0.3 to 9.2	V	
				Full range	0 to 8.5		V	
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				-55°C	7.8	8.6		
				125°C	7.8	8.8		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$	$I_{OL} = 0$	25°C	0	50	mV	
				-55°C	0	50		
				125°C	0	50		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V}$	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				-55°C	15	420		
				125°C	15	190		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICRmin}$	25°C	65	94	dB	
				-55°C	60	93		
				125°C	60	93		
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-55°C	60	91		
				125°C	60	94		
$I_{DD}$	Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load	$V_{IC} = 5\text{ V}$	25°C	570	1200	$\mu\text{A}$	
				-55°C	980	2000		
				125°C	360	960		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

### electrical characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$T_A = 25^\circ\text{C}$ to $70^\circ\text{C}$		1.7		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 2.5\text{ V}$ , $V_{IC} = 2.5\text{ V}$		0.6		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 100\text{ k}\Omega$	3.2	3.9		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V}$ to $2\text{ V}$ , $R_L = 100\text{ k}\Omega$	25	170		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	91		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V}$ to $10\text{ V}$ , $V_O = 1.4\text{ V}$	70	93		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load $V_{IC} = 2.5\text{ V}$		420	1120	$\mu\text{A}$

### electrical characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1.4\text{ V}$ , $R_S = 50\ \Omega$ , $V_{IC} = 0$ , $R_L = 100\text{ k}\Omega$		1.1	10	mV
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$T_A = 25^\circ\text{C}$ to $70^\circ\text{C}$		2.1		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.1		pA
$I_{IB}$ Input bias current (see Note 4)	$V_O = 5\text{ V}$ , $V_{IC} = 5\text{ V}$		0.7		pA
$V_{ICR}$ Common-mode input voltage range (see Note 5)		-0.2 to 9	-0.3 to 9.2		V
$V_{OH}$ High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 100\text{ k}\Omega$	8	8.7		V
$V_{OL}$ Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V}$ to $6\text{ V}$ , $R_L = 100\text{ k}\Omega$	25	275		V/mV
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	65	94		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V}$ to $10\text{ V}$ , $V_O = 1.4\text{ V}$	70	93		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 5\text{ V}$ , No load $V_{IC} = 5\text{ V}$		570	1200	$\mu\text{A}$

NOTES:4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	0.43		V/ $\mu\text{s}$	
			0°C	0.46			
			70°C	0.36			
		$V_{Ipp} = 2.5\ \text{V}$	25°C	0.40			
			0°C	0.43			
			70°C	0.34			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	55		kHz	
			0°C	60			
			70°C	50			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	525		kHz	
			0°C	610			
			70°C	400			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	40°			
			0°C	41°			
			70°C	39°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\ \text{V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	0.62		V/ $\mu\text{s}$	
			0°C	0.67			
			70°C	0.51			
		$V_{Ipp} = 5.5\ \text{V}$	25°C	0.56			
			0°C	0.61			
			70°C	0.46			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	35		kHz	
			0°C	40			
			70°C	30			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	635		kHz	
			0°C	710			
			70°C	510			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	43°			
			0°C	44°			
			70°C	42°			



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	0.43			V/ $\mu\text{s}$
			-40°C	0.51			
			85°C	0.35			
		$V_{Ipp} = 2.5\ \text{V}$	25°C	0.40			
			-40°C	0.48			
			85°C	0.32			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	55			kHz
			-40°C	75			
			85°C	45			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	525			kHz
			-40°C	770			
			85°C	370			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	40°			
			-40°C	43°			
			85°C	38°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\ \text{V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{Ipp} = 1\ \text{V}$	25°C	0.62			V/ $\mu\text{s}$
			-40°C	0.77			
			85°C	0.47			
		$V_{Ipp} = 5.5\ \text{V}$	25°C	0.56			
			-40°C	0.70			
			85°C	0.44			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32			nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	35			kHz
			-40°C	45			
			85°C	25			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	635			kHz
			-40°C	880			
			85°C	480			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	43°			
			-40°C	46°			
			85°C	41°			





# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4M TLC27M9M			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{I\text{PP}} = 1\ \text{V}$	25°C	0.43		V/ $\mu\text{s}$	
			-55°C	0.54			
			125°C	0.29			
		$V_{I\text{PP}} = 2.5\ \text{V}$	25°C	0.40			
			-55°C	0.50			
			125°C	0.28			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	55		kHz	
			-55°C	80			
			125°C	40			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	525		kHz	
			-55°C	850			
			125°C	330			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	40°			
			-55°C	44°			
			125°C	36°			

## operating characteristics at specified free-air temperature, $V_{DD} = 10\ \text{V}$

PARAMETER	TEST CONDITIONS		$T_A$	TLC27M4M TLC27M9M			UNIT
				MIN	TYP	MAX	
SR      Slew rate at unity gain	$R_L = 100\ \Omega$ , $C_L = 20\ \text{pF}$ , See Figure 1	$V_{I\text{PP}} = 1\ \text{V}$	25°C	0.62		V/ $\mu\text{s}$	
			-55°C	0.81			
			125°C	0.38			
		$V_{I\text{PP}} = 5.5\ \text{V}$	25°C	0.56			
			-55°C	0.73			
			125°C	0.35			
$V_n$ Equivalent input noise voltage	$f = 1\ \text{kHz}$ , See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$	
$B_{\text{OM}}$ Maximum output-swing bandwidth	$V_O = V_{\text{OH}}$ , $R_L = 100\ \text{k}\Omega$	$C_L = 20\ \text{pF}$ , See Figure 1	25°C	35		kHz	
			-55°C	50			
			125°C	20			
$B_1$ Unity-gain bandwidth	$V_I = 10\ \text{mV}$ , See Figure 3	$C_L = 20\ \text{pF}$	25°C	635		kHz	
			-55°C	960			
			125°C	440			
$\phi_m$ Phase margin	$V_I = 10\ \text{mV}$ , $C_L = 20\ \text{pF}$	$f = B_1$ , See Figure 3	25°C	43°			
			-55°C	47°			
			125°C	39°			



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9

## LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

### operating characteristics, $V_{DD} = 5\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		TLC27M4Y			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	0.43			$\text{V}/\mu\text{s}$
		$V_{Ipp} = 2.5\text{ V}$	0.40			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	32			$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	55			kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	525			kHz
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	40°			

### operating characteristics, $V_{DD} = 10\text{ V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		TLC27M4Y			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 1	$V_{Ipp} = 1\text{ V}$	0.62			$\text{V}/\mu\text{s}$
		$V_{Ipp} = 5.5\text{ V}$	0.56			
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 2	$R_S = 20\ \Omega$	32			$\text{nV}/\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 100\text{ k}\Omega$	$C_L = 20\text{ pF}$ , See Figure 1	35			kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 3	$C_L = 20\text{ pF}$	635			kHz
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$	$f = B_1$ , See Figure 3	43°			



# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### single-supply versus split-supply test circuits

Because the TLC27M4 and TLC27M9 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

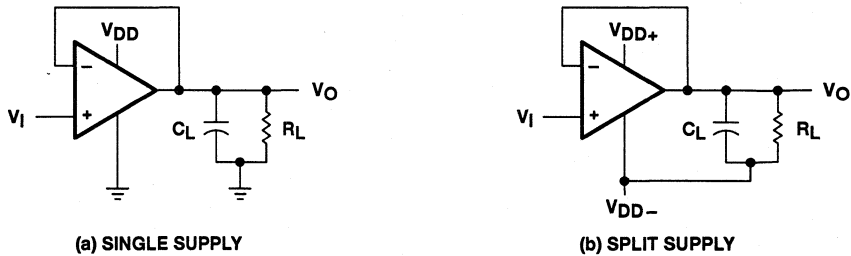


Figure 1. Unity-Gain Amplifier

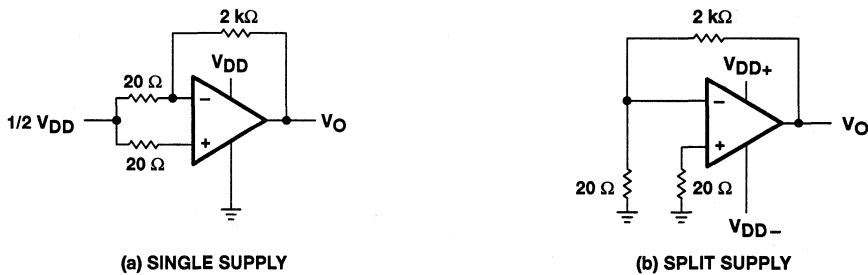


Figure 2. Noise-Test Circuit

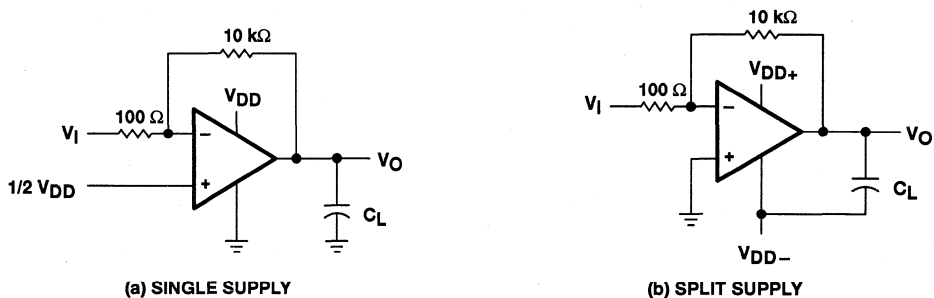


Figure 3. Gain-of-100 Inverting Amplifier

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC27M4 and TLC27M9 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution ... many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

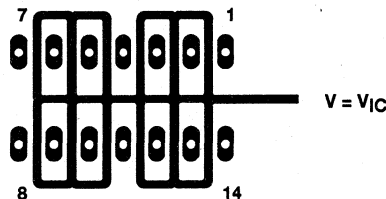


Figure 4. Isolation Metal Around Device Inputs  
(J and N packages)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

**PARAMETER MEASUREMENT INFORMATION**

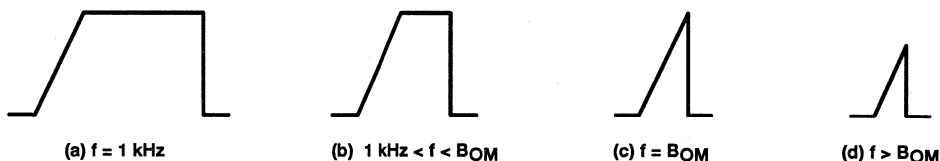
**input offset voltage temperature coefficient**

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

**full-power response**

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.



**Figure 5. Full-Power-Response Output Signal**

**test time**

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

**TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9**  
**LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS**

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	6, 7
$\alpha V_{IO}$	Temperature coefficient of input offset voltage	Distribution	8, 9
$V_{OH}$	High-level output voltage	vs High-level output current	10, 11
		vs Supply voltage	12
		vs Free-air temperature	13
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	14, 15
		vs Differential input voltage	16
		vs Free-air temperature	17
		vs Low-level output current	18, 19
$A_{VD}$	Differential voltage amplification	vs Supply voltage	20
		vs Free-air temperature	21
		vs Frequency	32, 33
$I_{IB}$	Input bias current	vs Free-air temperature	22
$I_{IO}$	Input offset current	vs Free-air temperature	22
$V_{IC}$	Common-mode input voltage	vs Supply voltage	23
$I_{DD}$	Supply current	vs Supply voltage	24
		vs Free-air temperature	25
SR	Slew rate	vs Supply voltage	26
		vs Free-air temperature	27
	Normalized slew rate	vs Free-air temperature	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
$B_1$	Unity-gain bandwidth	vs Free-air temperature	30
		vs Supply voltage	31
	Phase shift	vs Frequency	32, 33
$\phi_m$	Phase margin	vs Supply voltage	34
		vs Free-air temperature	35
		vs Load capacitance	36
$V_n$	Equivalent input noise voltage	vs Frequency	37

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS

**DISTRIBUTION OF TLC27M4  
INPUT OFFSET VOLTAGE**

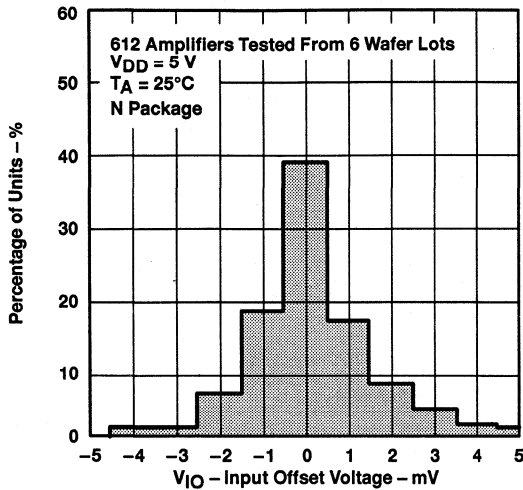


Figure 6

**DISTRIBUTION OF TLC27M4  
INPUT OFFSET VOLTAGE**

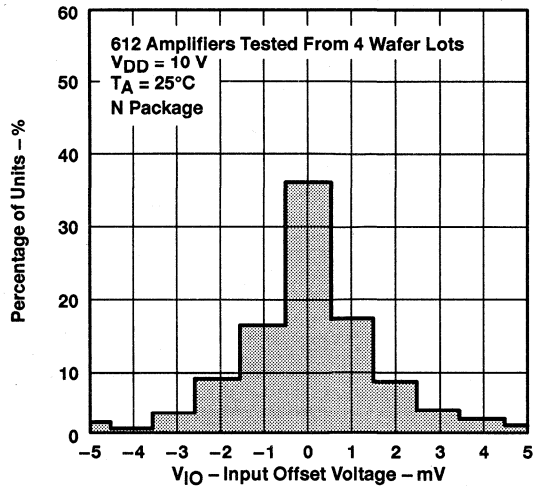


Figure 7

**DISTRIBUTION OF TLC27M4 AND TLC27M9  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

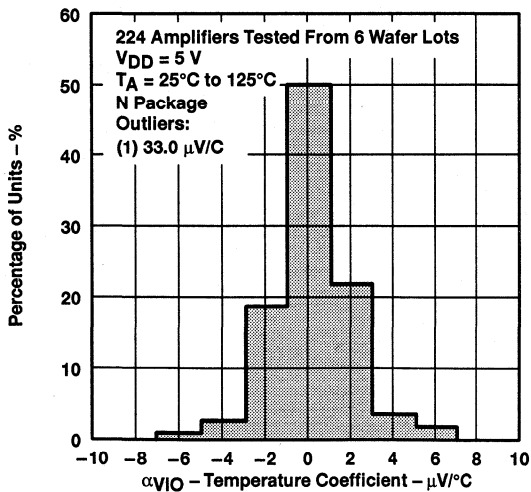


Figure 8

**DISTRIBUTION OF TLC27M4 AND TLC27M9  
INPUT OFFSET VOLTAGE  
TEMPERATURE COEFFICIENT**

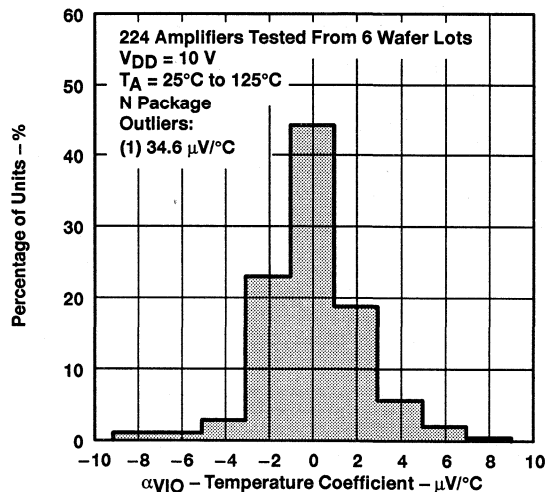


Figure 9

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

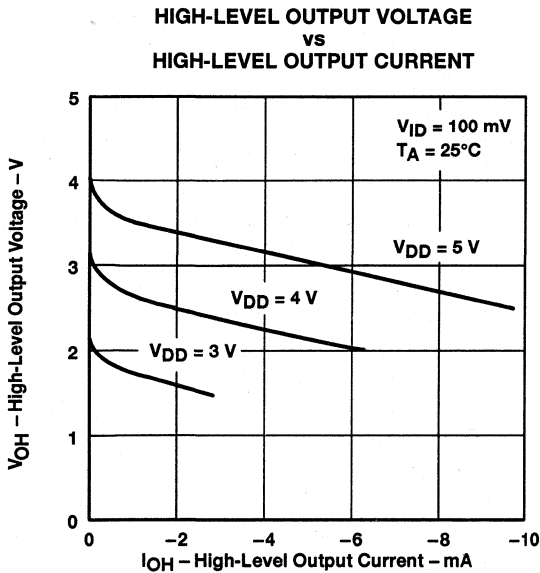


Figure 10

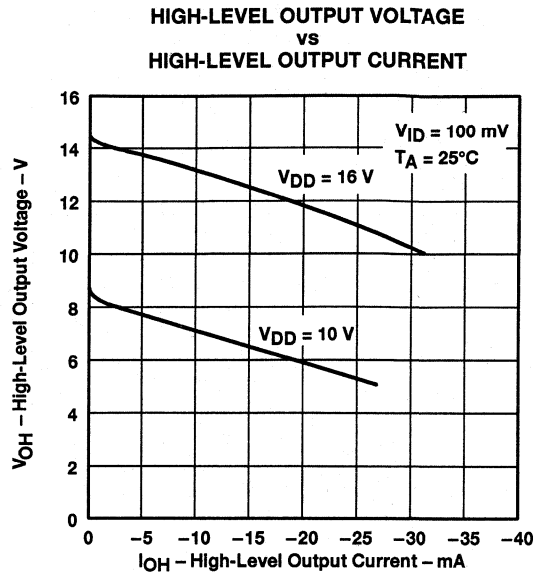


Figure 11

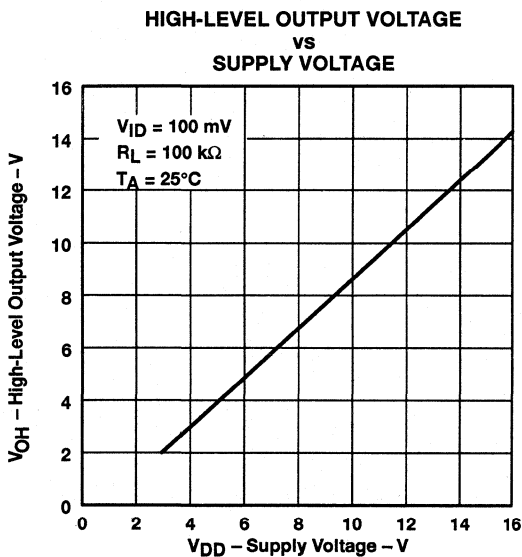


Figure 12

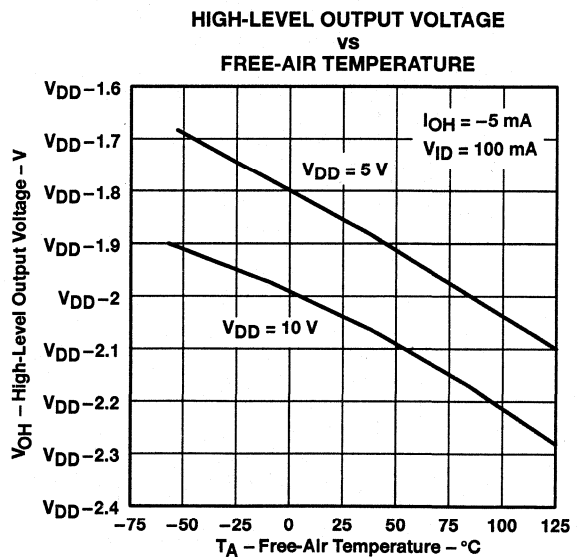


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE

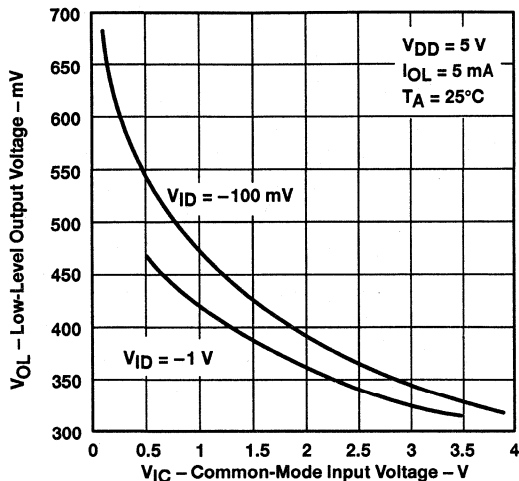


Figure 14

LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE

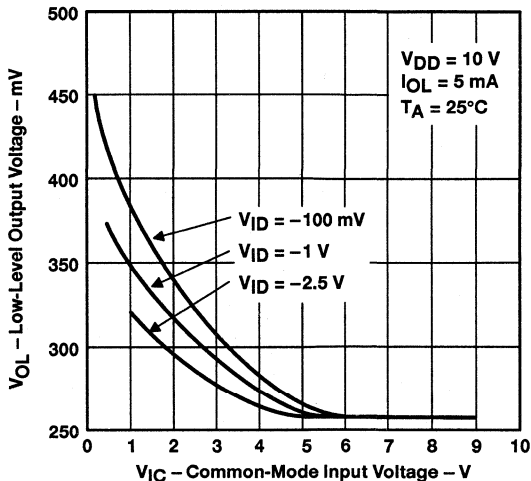


Figure 15

LOW-LEVEL OUTPUT VOLTAGE  
vs  
DIFFERENTIAL INPUT VOLTAGE

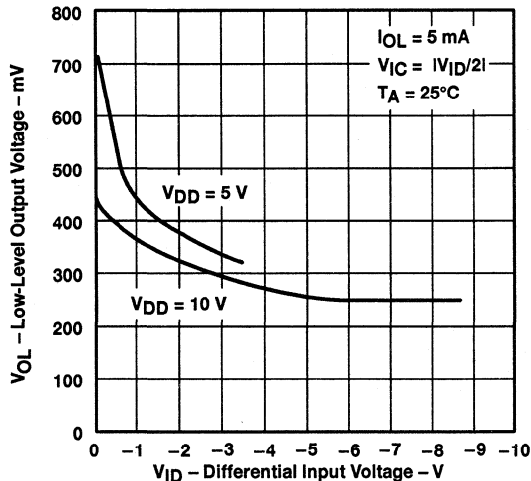


Figure 16

LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE

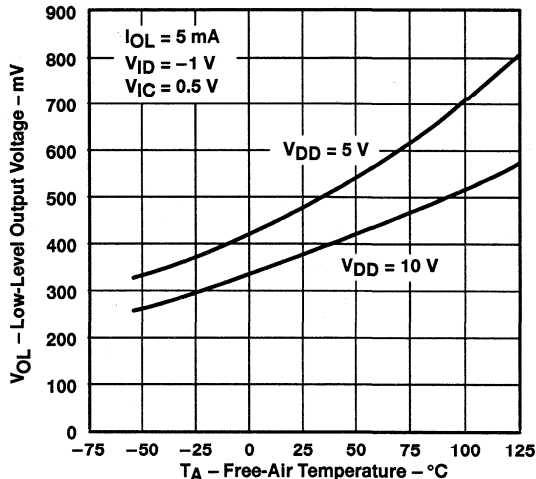


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

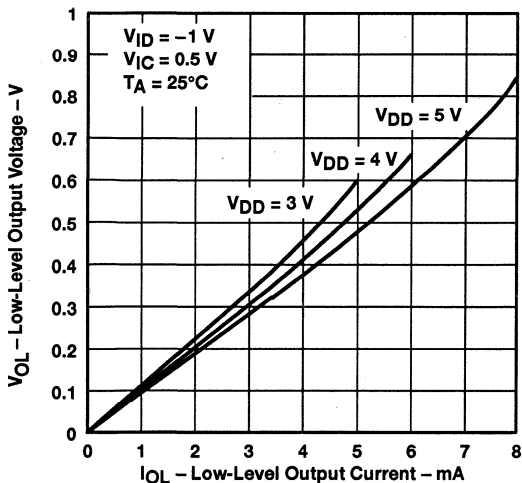


Figure 18

LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT

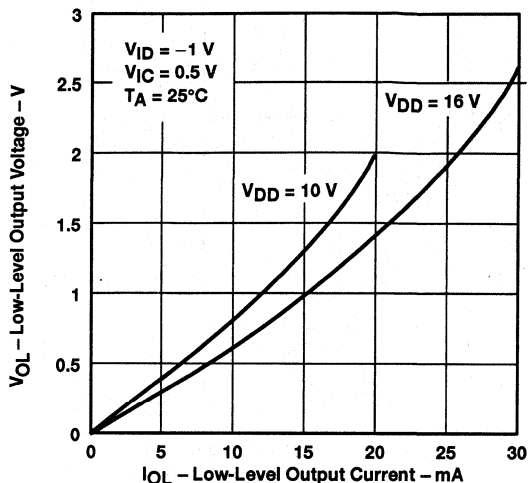


Figure 19

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
SUPPLY VOLTAGE

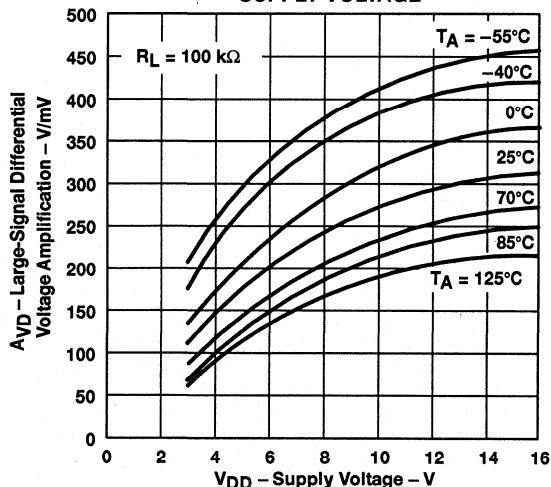


Figure 20

LARGE-SIGNAL  
DIFFERENTIAL VOLTAGE AMPLIFICATION  
vs  
FREE-AIR TEMPERATURE

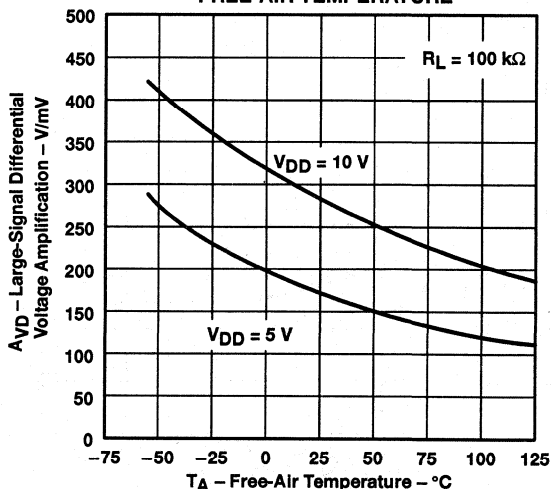
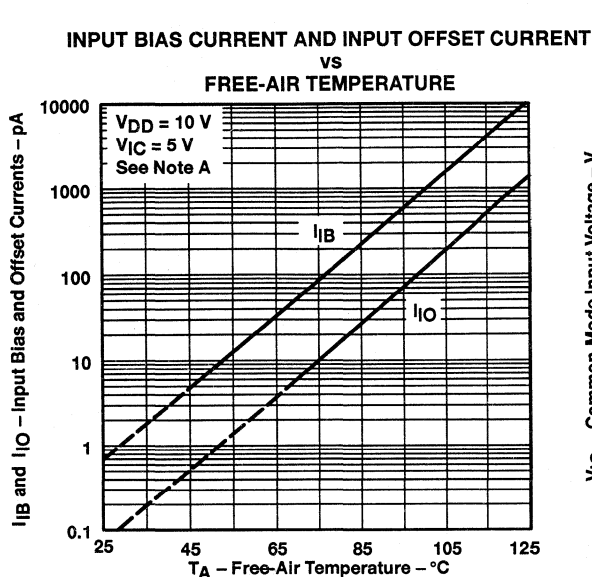


Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

Figure 23

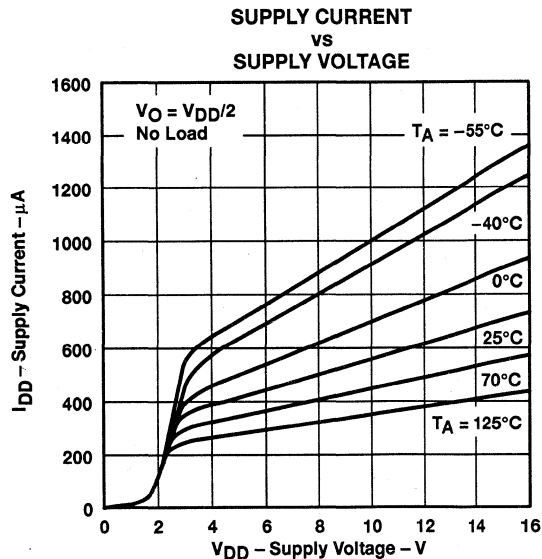


Figure 24

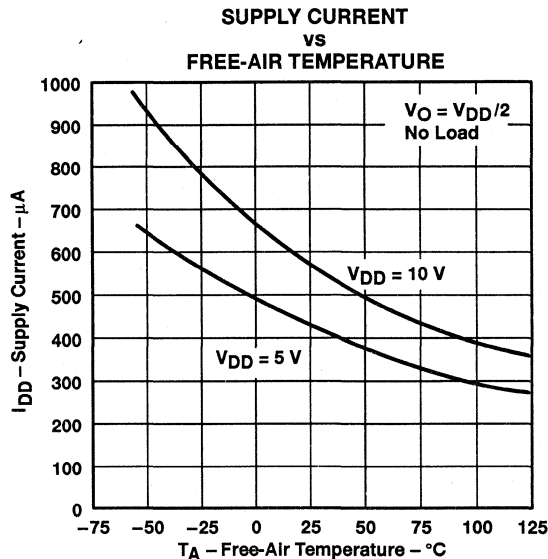


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

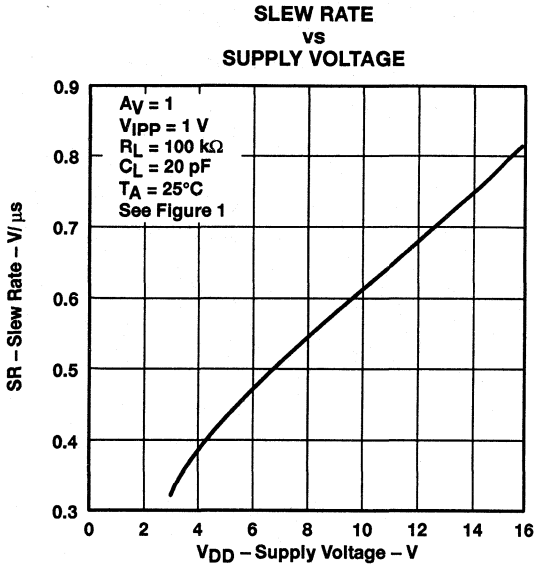


Figure 26

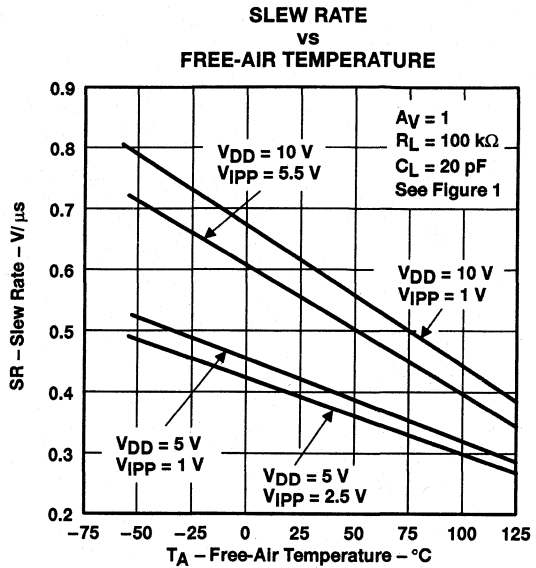


Figure 27

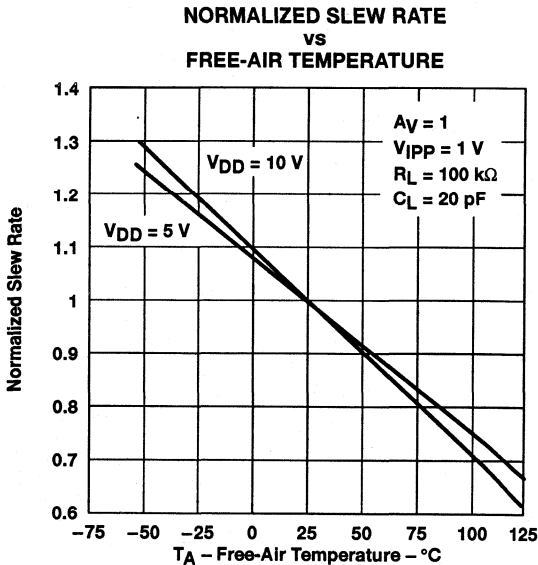


Figure 28

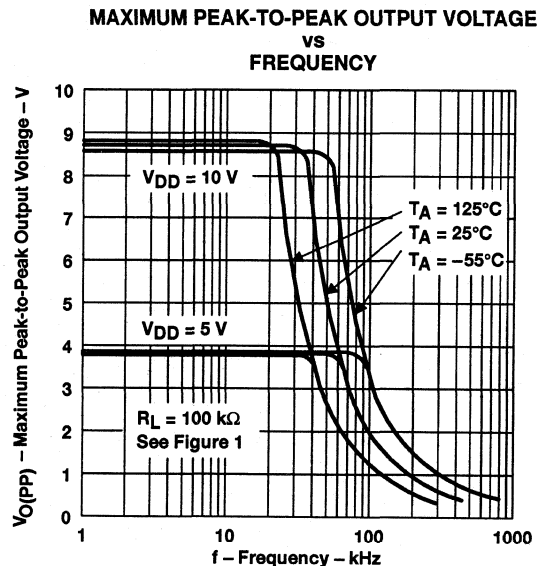


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

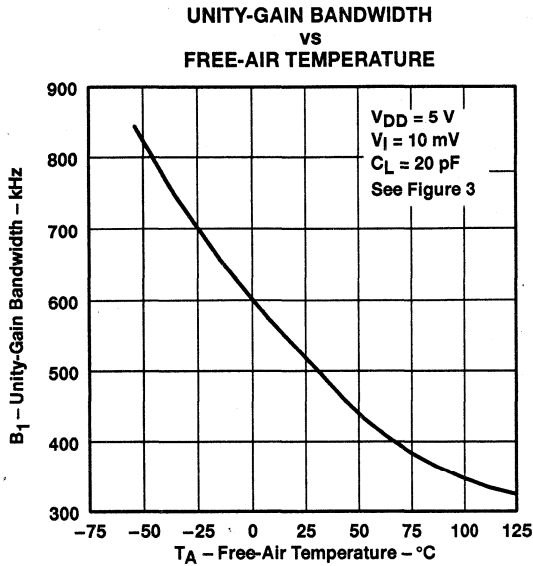


Figure 30

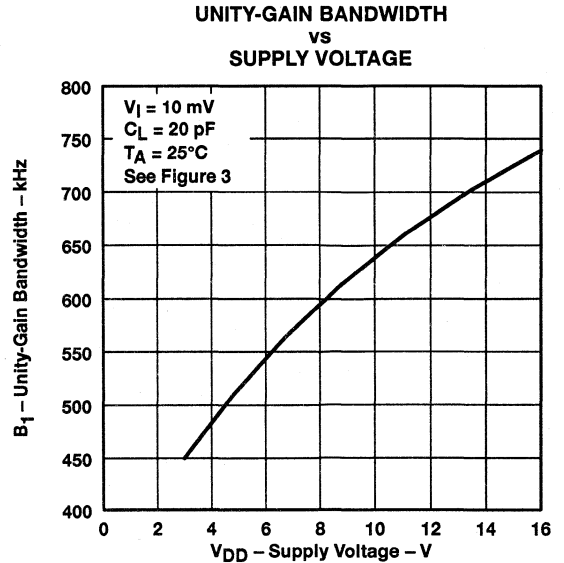


Figure 31

## LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

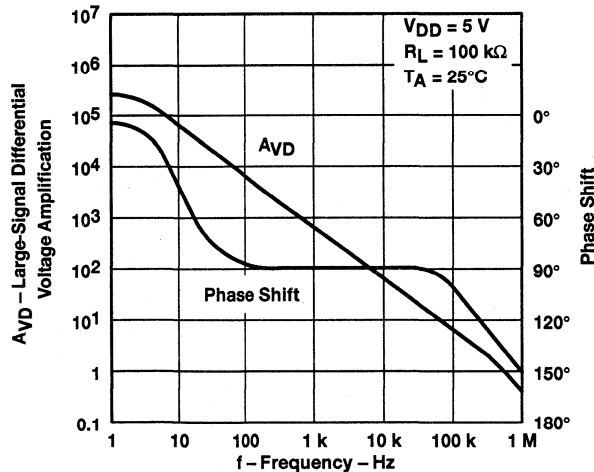


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS†

### LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

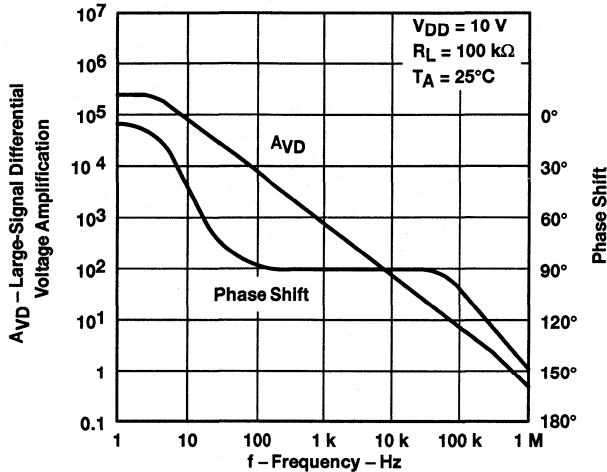


Figure 33

### PHASE MARGIN VS SUPPLY VOLTAGE

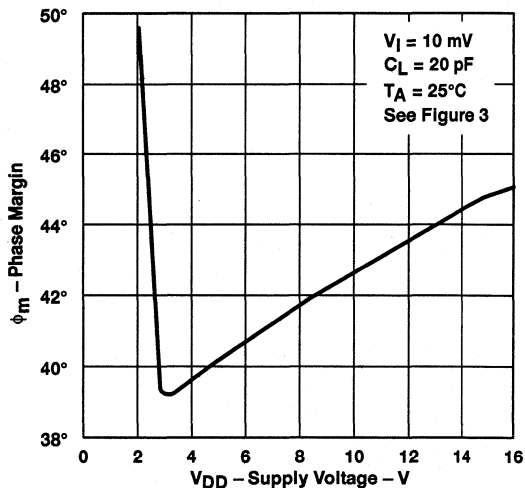


Figure 34

### PHASE MARGIN VS FREE-AIR TEMPERATURE

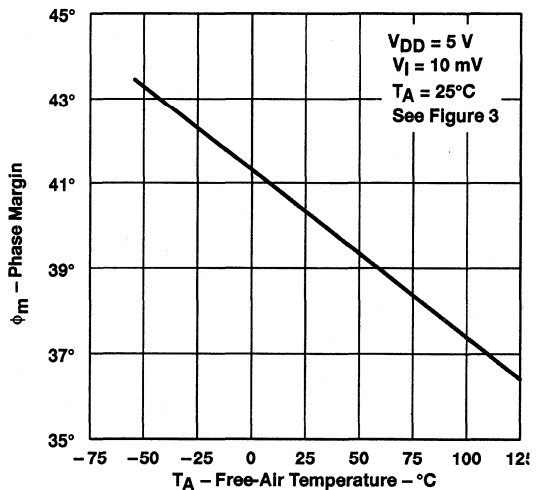


Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL CHARACTERISTICS

PHASE MARGIN  
vs  
CAPACITIVE LOAD

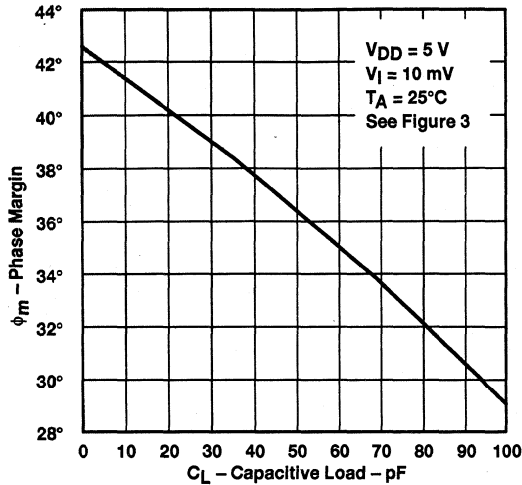


Figure 36

EQUIVALENT INPUT NOISE VOLTAGE  
vs  
FREQUENCY

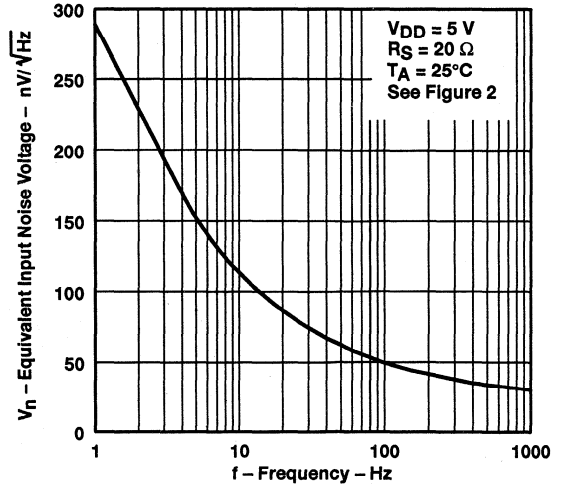


Figure 37

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

### single-supply operation

While the TLC27M4 and TLC27M9 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M4 and TLC27M9 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M4 and TLC27M9 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

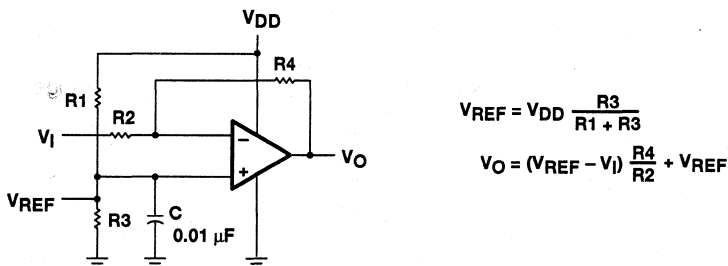


Figure 38. Inverting Amplifier With Voltage Reference



APPLICATION INFORMATION

single-supply operation (continued)

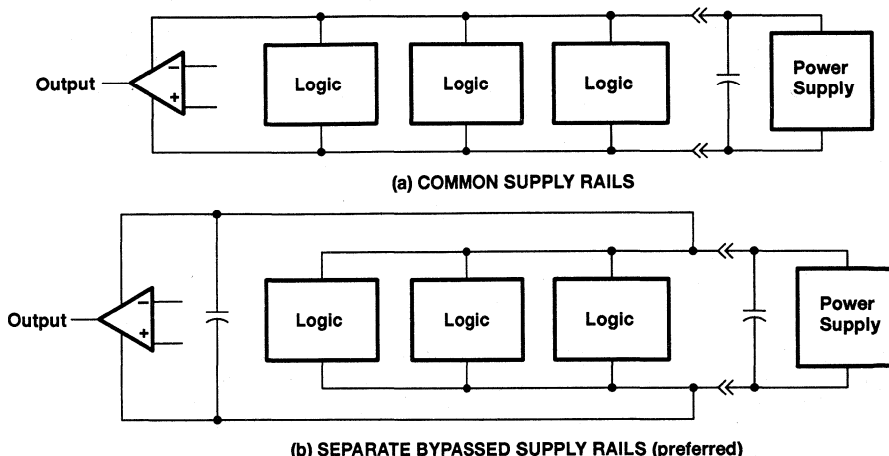


Figure 39. Common Versus Separate Supply Rails

input characteristics

The TLC27M4 and TLC27M9 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1\text{ V}$  at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.5\text{ V}$  at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M4 and TLC27M9 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically  $0.1\ \mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M4 and TLC27M9 are well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as unity-gain followers to avoid possible oscillation.

noise performance

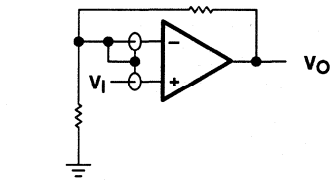
The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M4 and TLC27M9 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than  $50\ \text{k}\Omega$ , since bipolar devices exhibit greater noise currents.

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

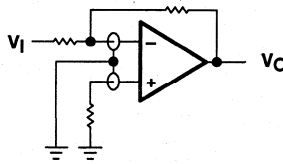
SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION

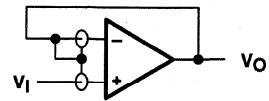
### noise performance (continued)



(a) NONINVERTING AMPLIFIER



(b) INVERTING AMPLIFIER



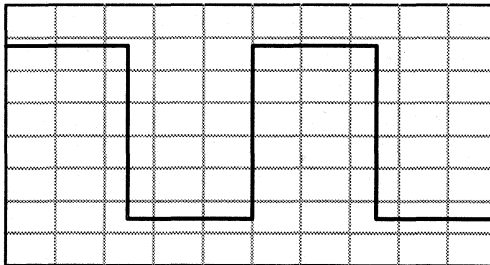
(c) UNITY-GAIN AMPLIFIER

Figure 40. Guard-Ring Schemes

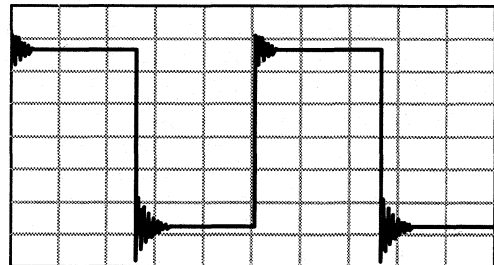
### output characteristics

The output stage of the TLC27M4 and TLC27M9 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

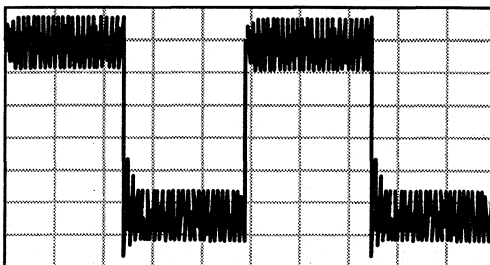
All operating characteristics of the TLC27M4 and TLC27M9 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.



(a)  $C_L = 20 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(b)  $C_L = 170 \text{ pF}$ ,  $R_L = \text{NO LOAD}$



(c)  $C_L = 190 \text{ pF}$ ,  $R_L = \text{NO LOAD}$

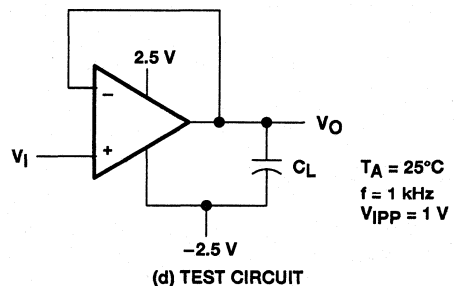


Figure 41. Effect of Capacitive Loads and Test Circuit

APPLICATION INFORMATION

output characteristics (continued)

Although the TLC27M4 and TLC27M9 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately  $60\ \Omega$  and  $180\ \Omega$ , depending on how hard the op amp input is driven. With very low values of  $R_P$ , a voltage offset from 0 V at the output occurs. Second, pullup resistor  $R_P$  acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

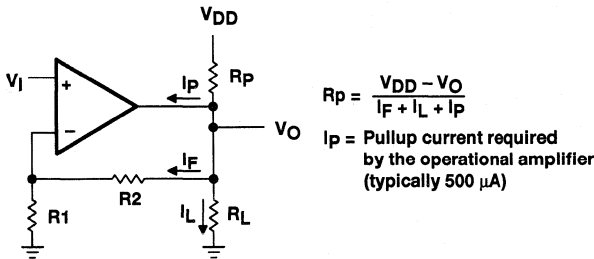


Figure 42. Resistive Pullup to Increase  $V_{OH}$

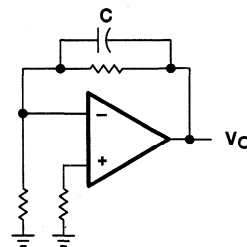


Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic discharge protection

The TLC27M4 and TLC27M9 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature-dependent and have the characteristics of a reverse-biased diode.

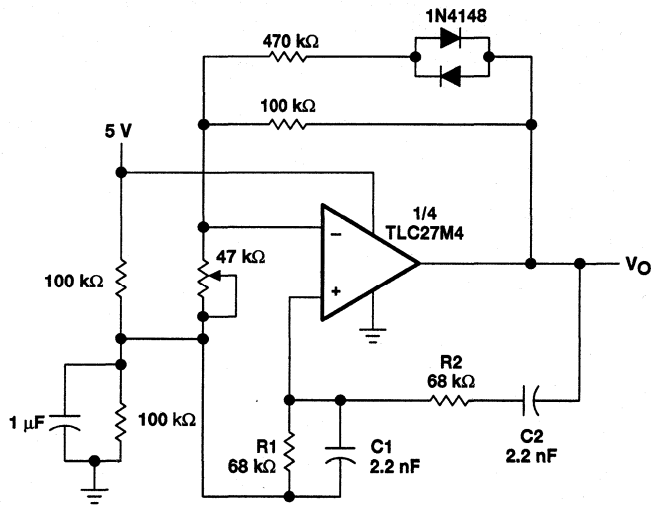
latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27M4 and TLC27M9 inputs and outputs were designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ( $0.1\ \mu\text{F}$  typical) located across the supply rails as close to the device as possible.

APPLICATION INFORMATION

latch-up (continued)

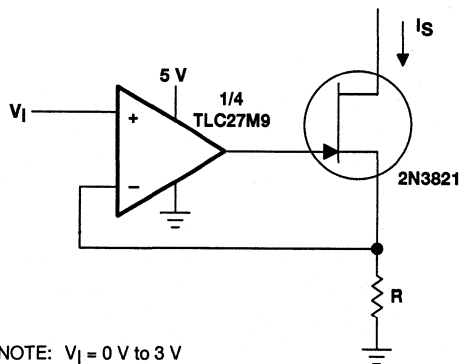
The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.



NOTE:  $V_{OPP} = 2\text{ V}$

$$f_0 = \frac{1}{2\pi \sqrt{R_1 R_2 C_1 C_2}}$$

Figure 44. Wien Oscillator



NOTE:  $V_I = 0\text{ V to } 3\text{ V}$

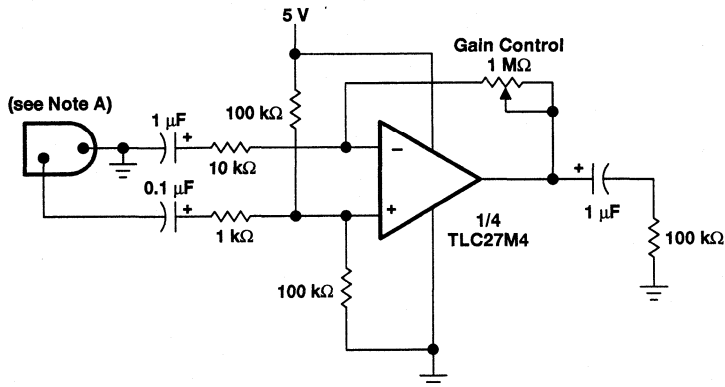
$$I_S = \frac{V_I}{R}$$

Figure 45. Precision Low-Current Sink

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 PRECISION QUAD OPERATIONAL AMPLIFIERS

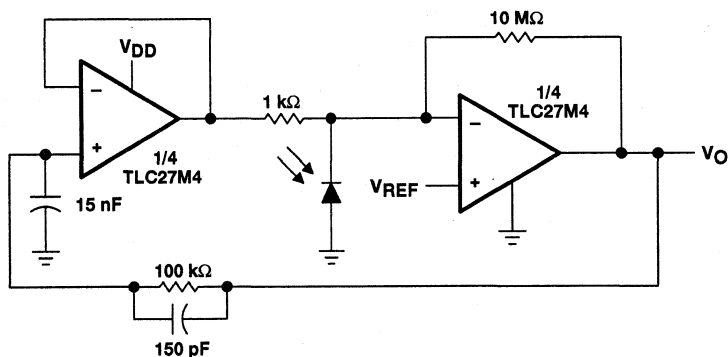
SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## APPLICATION INFORMATION



NOTE A: Low to medium impedance dynamic mike

Figure 46. Microphone Preamplifier



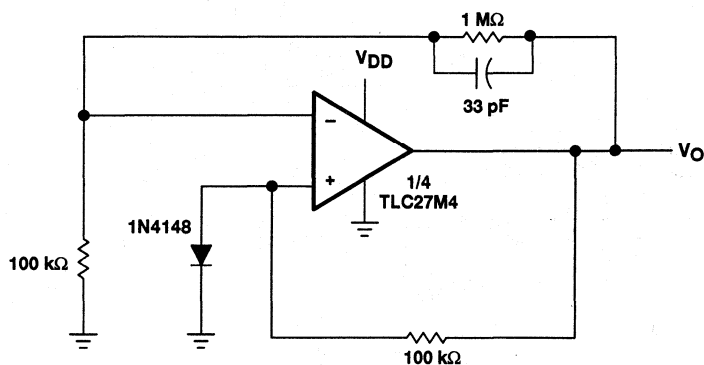
NOTE:  $V_{DD} = 4 \text{ V to } 15 \text{ V}$   
 $V_{REF} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

Figure 47. Photo-Diode Amplifier With Ambient Light Rejection

# TLC27M4, TLC27M4A, TLC27M4B, TLC27M4Y, TLC27M9 LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

SLOS093B – OCTOBER 1987 – REVISED AUGUST 1994

## TYPICAL APPLICATION DATA



NOTE:  $V_{DD} = 8 \text{ V to } 16 \text{ V}$   
 $V_O = 5 \text{ V, } 10 \text{ mA}$

Figure 48. Low-Power Voltage Regulator

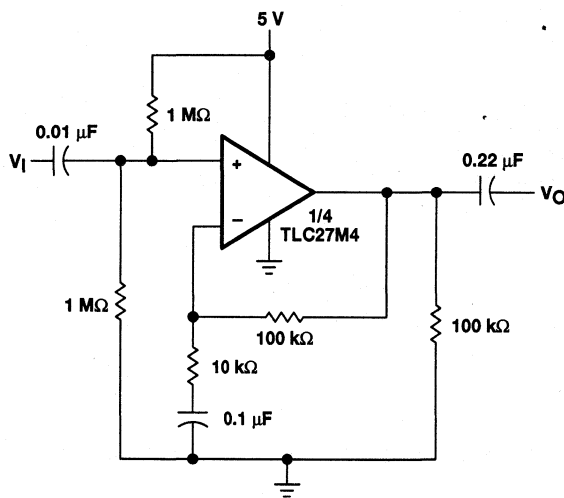


Figure 49. Single-Rail AC Amplifier

# TLC1078 LinCMOS™ $\mu$ POWER PRECISION DUAL OPERATIONAL AMPLIFIER

SLOS022C – AUGUST 1988 – REVISED AUGUST 1994

- Power Dissipation as Low as 10  $\mu$ W Typ Per Amplifier
- Operates on a Single Silver-Oxide Watch Battery,  $V_{DD} = 1.4$  V Min
- $V_{IO} \dots 450$   $\mu$ V Max in DIP and Small-Outline Package
- Input Offset Voltage Drift  $\dots 0.1$   $\mu$ V/Month Typ, Including the First 30 Days
- High-impedance LinCMOS™ Inputs  
 $I_{IB} = 0.6$  pA Typ
- High Open-Loop Gain  $\dots 800000$  Typ
- Output Drive Capability  $> 20$  mA
- Slew Rate  $\dots 47$  V/ms Typ
- Common-Mode Input Voltage Range Extends Below the Negative Rail
- Output Voltage Range Includes Negative Rail
- On-Chip ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel

## description

The TLC1078 operational amplifier offers ultra-low offset voltage, high gain, 110-kHz bandwidth, 47-V/ms slew rate, and just 150- $\mu$ W power dissipation per amplifier.

With a supply voltage of 1.4 V, common-mode input to the negative rail, and output swing to the negative rail, the TLC1078C is an ideal solution for low-voltage battery-operated systems. The 20-mA output drive capability means that the TLC1078 can easily drive small resistive and large capacitive loads when needed, while maintaining ultra-low standby power dissipation.

Since this device is functionally compatible as well as pin compatible with the TLC27L2 and TLC27L7, the TLC1078 easily upgrades existing designs that can benefit from its improved performance.

The TLC1078 incorporates internal ESD-protection circuits that will prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised when handling these devices as exposure to ESD may result in degradation of the device parametric performance. The TLC1078 design also inhibits latch-up of the device inputs and outputs even with surge currents as large as 100 mA.

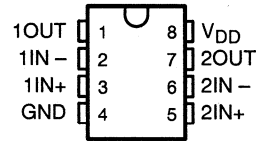
LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

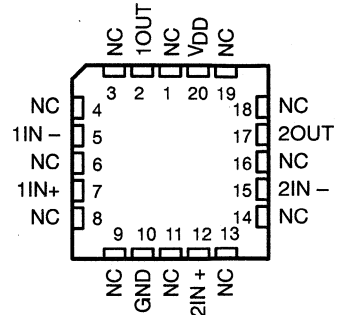
 **TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

D, JG, OR P PACKAGE  
(TOP VIEW)

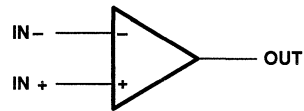


FK PACKAGE  
(TOP VIEW)

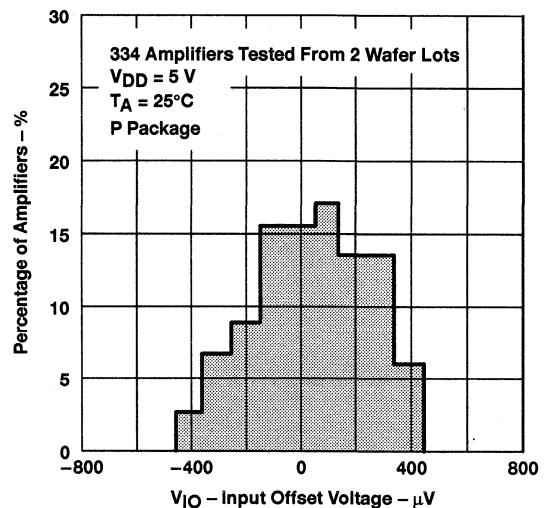


NC – No internal connection

## symbol (each amplifier)



DISTRIBUTION OF TLC1078  
INPUT OFFSET VOLTAGE



Copyright © 1994, Texas Instruments Incorporated

**TLC1078**  
**LinCMOS™  $\mu$ POWER PRECISION**  
**DUAL OPERATIONAL AMPLIFIER**  
 SLOS022C – AUGUST 1988 – REVISED AUGUST 1994

**description (continued)**

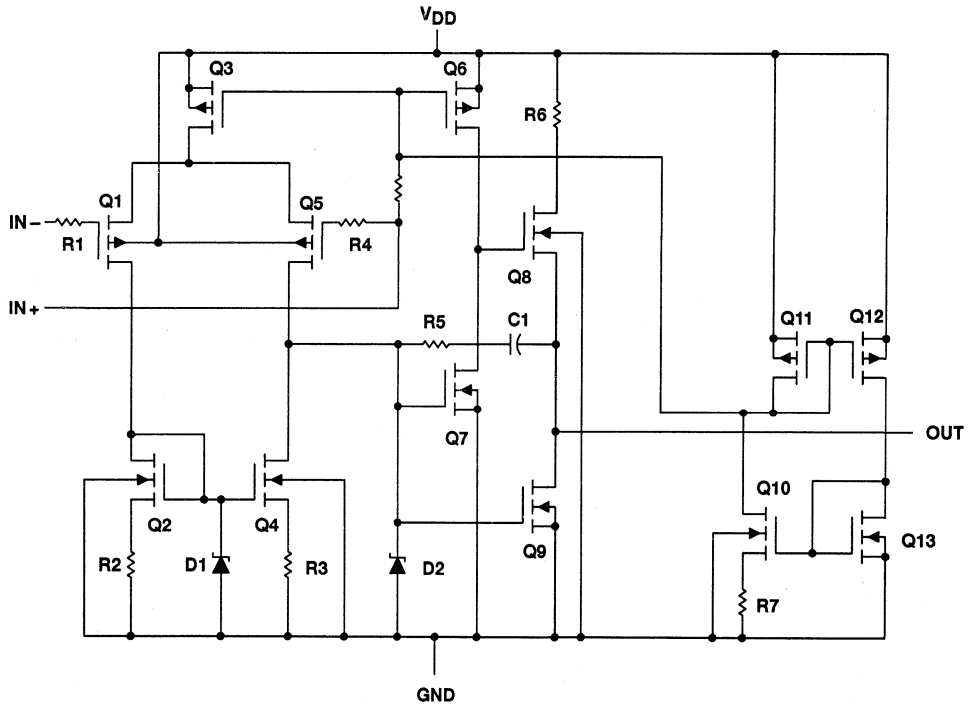
The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C. The wide range of packaging options includes small-outline and chip-carrier versions for high-density system applications.

**AVAILABLE OPTIONS**

T <sub>A</sub>	PACKAGE			
	SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	TLC1078CD	—	—	TLC1078CP
-40°C to 85°C	TLC1078ID	—	—	TLC1078IP
-55°C to 125°C	TLC1078MD	TLC1078MFK	TLC1078MJG	TLC1078MP

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC1078CDR).

**equivalent schematic (each amplifier)**



COMPONENT COUNT	
Transistors	38
Diodes	12
Resistors	16
Capacitors	2





**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$ (see Note 3)	45 mA
Duration of short-circuit at (or below) $T_A = 25^\circ\text{C}$ (see Note 3)	unlimited
Continuous total dissipation	see Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation ratings are not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		1.4	16	3	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	4	-0.2	4	0	4	V
	$V_{DD} = 10$ V	-0.2	9	-0.2	9	0	9	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C

**TLC1078**  
**LinCMOS™  $\mu$ POWER PRECISION**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS022C – AUGUST 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	TA†	TLC1078C						UNIT
			VDD = 5 V			VDD = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 $\Omega$ , V <sub>IC</sub> = 0, R <sub>I</sub> = 1 M $\Omega$	25°C	160 450		180 600		$\mu$ V	
			Full range	800		950			
$\alpha$ V <sub>IO</sub>	Temperature coefficient of input offset voltage		25°C to 70°C	1.1		1		$\mu$ V/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1		pA	
			70°C	7 300		7 300			
I <sub>IB</sub>	Input bias current (see Note 4)		25°C	0.6		0.7		pA	
			70°C	40 600		50 600			
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5		-0.2 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 M $\Omega$	25°C	3.2	4.1	8.2	8.9	V	
			0°C	3.2	4.1	8.2	8.9		
			70°C	3.2	4.2	8.2	8.9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0 25		0 25		mV	
			0°C	0 25		0 25			
			70°C	0 25		0 25			
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 M $\Omega$ , See Note 6	25°C	250	525	500	850	V/mV	
			0°C	250	680	500	1010		
			70°C	200	380	350	660		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	70	95	75	97	dB	
			0°C	70	95	75	97		
			70°C	70	95	75	97		
k <sub>SVR</sub>	Supply-voltage rejection ratio ( $\Delta$ V <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	75	98	75	98	dB	
			0°C	75	98	75	98		
			70°C	75	98	75	98		
I <sub>DD</sub>	Supply current (two amplifiers)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	20 34		29 46		$\mu$ A	
			0°C	24 42		36 66			
			70°C	16 28		22 40			

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**electrical characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC1078I				UNIT
			V <sub>DD</sub> = 5 V		V <sub>DD</sub> = 10 V		
			MIN	TYP	MAX	MIN	
V <sub>IO</sub> Input offset voltage	V <sub>O</sub> = 1.4 V, R <sub>S</sub> = 50 $\Omega$ , V <sub>IC</sub> = 0, R <sub>I</sub> = 1 M $\Omega$	25°C	160 450		180 600		$\mu$ V
		Full range	950		1100		
$\alpha$ V <sub>IO</sub> Temperature coefficient of input offset voltage	V <sub>IC</sub> = 0, R <sub>I</sub> = 1 M $\Omega$	25°C to 85°C	1.1		1		$\mu$ V/°C
I <sub>IO</sub> Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1		pA
		85°C	24	1000	26	1000	
I <sub>IB</sub> Input bias current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.6		0.7		pA
		85°C	200	2000	220	2000	
V <sub>ICR</sub> Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V
		Full range	-0.2 to 3.5		-0.2 to 8.5		V
V <sub>OH</sub> High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 M $\Omega$	25°C	3.2	4.1	8.2	8.9	V
		-40°C	3.2	4.1	8.2	8.9	
		85°C	3.2	4.2	8.2	8.9	
V <sub>OL</sub> Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0 25		0 25		mV
		-40°C	0 25		0 25		
		85°C	0 25		0 25		
A <sub>VD</sub> Large-signal differential voltage amplification	R <sub>L</sub> = 1 M $\Omega$ , See Note 6	25°C	250	525	500	850	V/mV
		-40°C	250	900	500	1550	
		85°C	150	300	250	585	
GMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	70	95	75	97	dB
		-40°C	70	95	75	97	
		85°C	70	95	75	97	
k <sub>SVR</sub> Supply-voltage rejection ratio ( $\Delta$ V <sub>DD</sub> / $\Delta$ V <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C	75	98	75	98	dB
		-40°C	75	98	75	98	
		85°C	75	98	75	98	
I <sub>DD</sub> Supply current (two amplifiers)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	20 34		29 46		$\mu$ A
		-40°C	31 54		50 86		
		85°C	15 26		20 36		

† Full range is -40°C to 80°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
 5. This range also applies to each input individually.  
 6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.

**TLC1078**  
**LinCMOS™  $\mu$ POWER PRECISION**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS022C – AUGUST 1988 – REVISED AUGUST 1994

**electrical characteristics at specified operating free-air temperature**

PARAMETER	TEST CONDITIONS	TA†	TLC1078M						UNIT	
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 $\Omega$ , R <sub>L</sub> = 1 M $\Omega$	25°C		160	450	180		600	$\mu$ V
		Full range		1250			1400			
$\alpha$ V <sub>IO</sub>	Temperature coefficient of input offset voltage		25°C to 125°C		1.4		1.4			$\mu$ V/°C
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> / 2, V <sub>IC</sub> = V <sub>DD</sub> / 2	25°C		0.1		0.1			pA
			125°C		1.4	15	1.8	15		nA
I <sub>IB</sub>	Input bias current (see Note 4)		25°C		0.6		0.7			pA
			125°C		9	35	10	35		nA
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C		0 to 4	-0.3 to 4.2	0 to 9	-0.3 to 9.2		V
			Full range		0 to 3.5		0 to 8.5			V
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 M $\Omega$	25°C		3.2	4.1	8.2	8.9		V
			-55°C		3.2	4.1	8.2	8.8		
			125°C		3.2	4.2	8.2	9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C		0		25	0	25	mV
			-55°C		0		25	0	25	
			125°C		0		25	0	25	
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 M $\Omega$ , See Note 6	25°C		250	525	500	850		V/mV
			-55°C		250	950	500	1750		
			125°C		35	200	75	380		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C		70	95	75	97		dB
			-55°C		70	95	75	97		
			125°C		70	85	75	91		
k <sub>SVR</sub>	Supply-voltage rejection ratio ( $\Delta$ V <sub>DD</sub> / $\Delta$ V <sub>IO</sub> )	V <sub>O</sub> = 1.4 V	25°C		75	98	75	98		dB
			-55°C		70	98	70	98		
			125°C		70	98	70	98		
I <sub>DD</sub>	Supply current (two amplifiers)	V <sub>O</sub> = V <sub>DD</sub> / 2, V <sub>IC</sub> = V <sub>DD</sub> / 2, No load	25°C		20	34	29	46		$\mu$ A
			-55°C		35	60	56	96		
			125°C		14	24	18	30		

† Full range is -55°C to 125°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1078C						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		0°C		35			51		
		70°C		27			38		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		0°C		100			125		
		70°C		65			90		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	25°C		34°			38°		
		0°C		36°			40°		
		70°C		30°			34°		

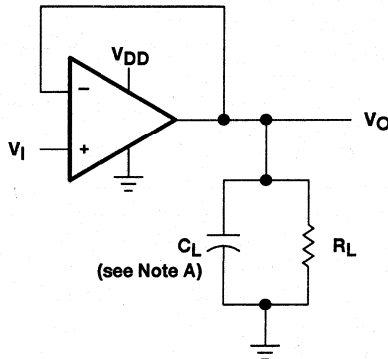
**operating characteristics at specified free-air temperature**

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1078I						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		-40°C		39			59		
		85°C		25			34		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		-40°C		130			155		
		85°C		55			80		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	25°C		34°			38°		
		-40°C		38°			40°		
		85°C		28°			32°		

**operating characteristics at specified free-air temperature**

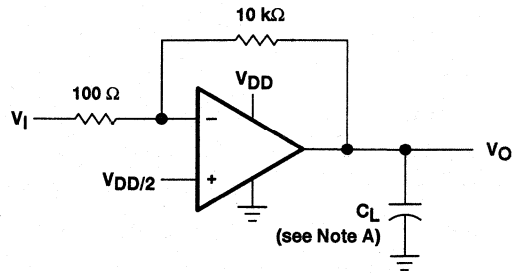
PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1078M						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		-55°C		41			63		
		125°C		20			27		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		-55°C		140			165		
		125°C		45			70		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	25°C		34°			38°		
		-55°C		39°			43°		
		125°C		25°			29°		

**PARAMETER MEASUREMENT INFORMATION**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 1. Slew-Rate Test Circuit**



**Figure 2. Unity-Gain Bandwidth and Phase-Margin Test Circuit**

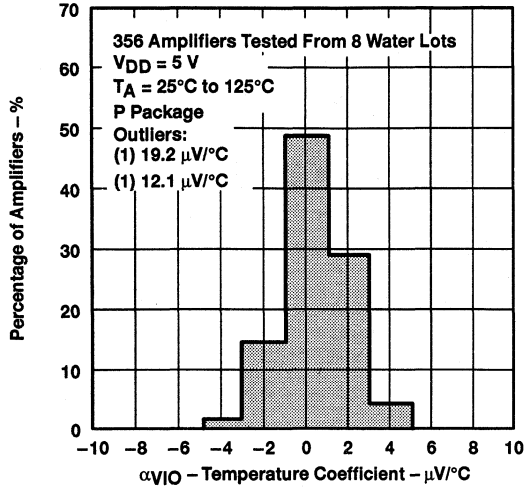
**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		<b>FIGURE</b>	
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	Distribution	3, 4
$V_{OH}$	High-level output voltage	vs High-level output current	5, 6
		vs Supply voltage	7
		vs Temperature	8
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	9, 10
		vs Differential input voltage	11
		vs Free-air temperature	12
		vs Low-level output current	13, 14
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	15
		vs Free-air temperature	16
		vs Frequency	27, 28
$I_{IB}$	Input bias current	vs Free-air temperature	17
$I_{IO}$	Input offset current	vs Free-air temperature	17
$V_{IC}$	Common-mode input voltage	vs Supply voltage	18
$I_{DD}$	Supply current	vs Supply voltage	19
		vs Free-air temperature	20
SR	Slew rate	vs Supply voltage	21
		vs Free-air temperature	22
	Normalized slew rate	vs Free-air temperature	23
$V_{OM}$	Maximum peak output voltage	vs Frequency	24
$B_1$	Unity-gain bandwidth	vs Free-air temperature	25
		vs Supply voltage	26
	Phase shift	vs Frequency	27, 28
$\phi_m$	Phase margin	vs Supply voltage	29
		vs Free-air temperature	30
		vs Load capacitance	31
$V_n$	Equivalent input noise voltage	vs Frequency	32

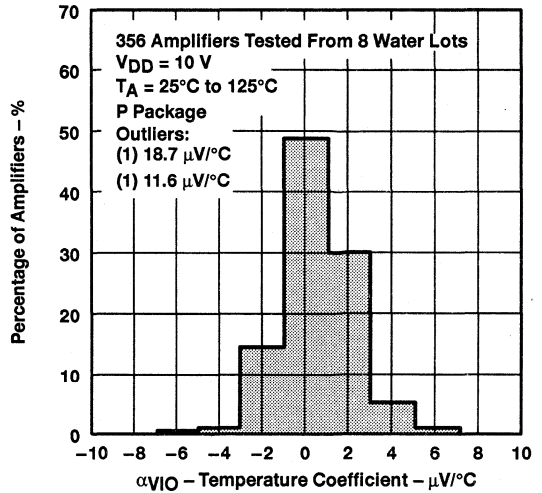
**TYPICAL CHARACTERISTICS**

**DISTRIBUTION OF TLC1078  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT**



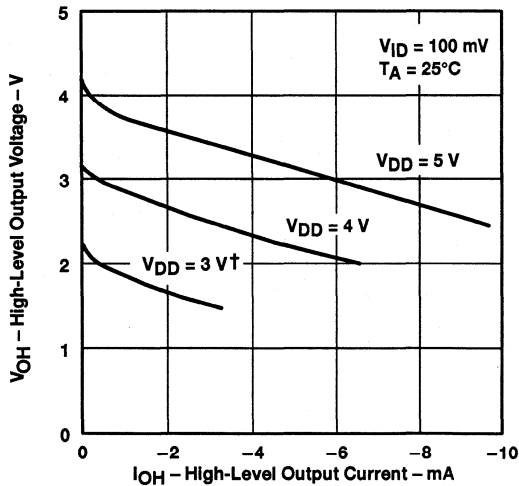
**Figure 3**

**DISTRIBUTION OF TLC1078  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT**



**Figure 4**

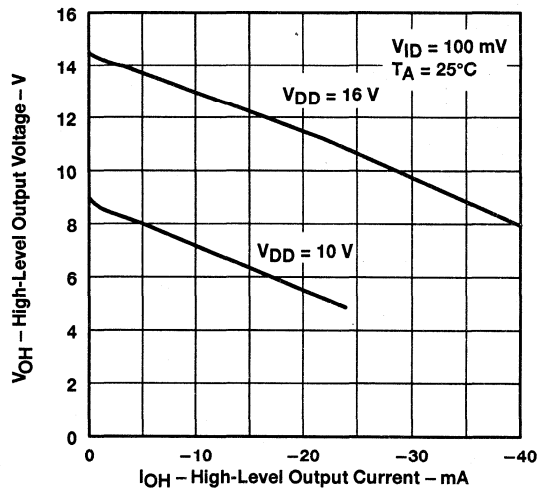
**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**



† The  $V_{DD} = 3\text{ V}$  curve does not apply to the TLC1078M.

**Figure 5**

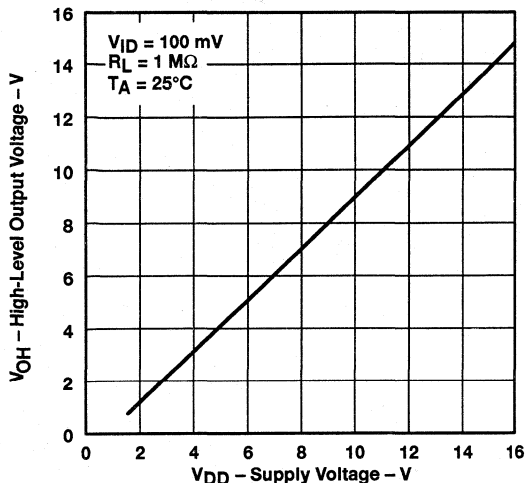
**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**



**Figure 6**

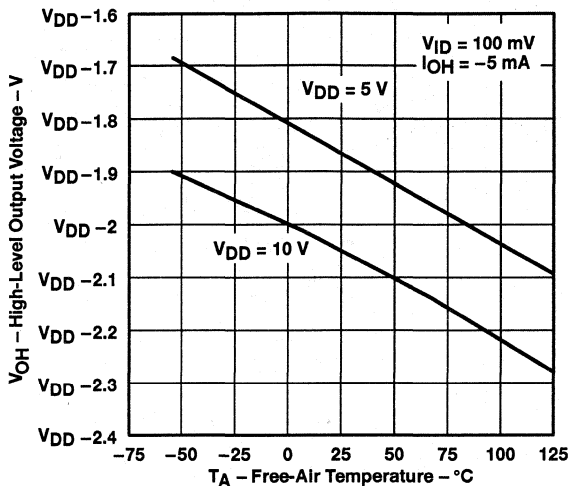
**TYPICAL CHARACTERISTICS†**

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE**



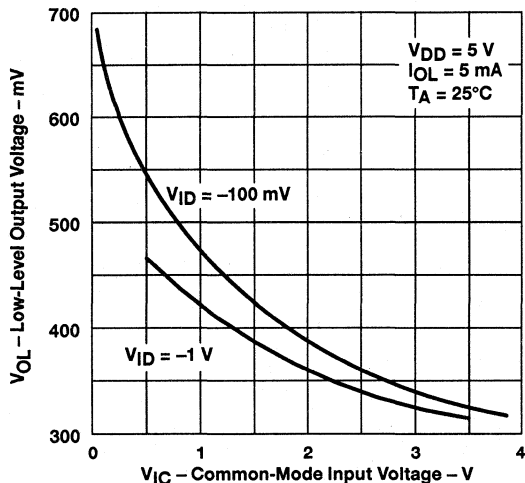
**Figure 7**

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE**



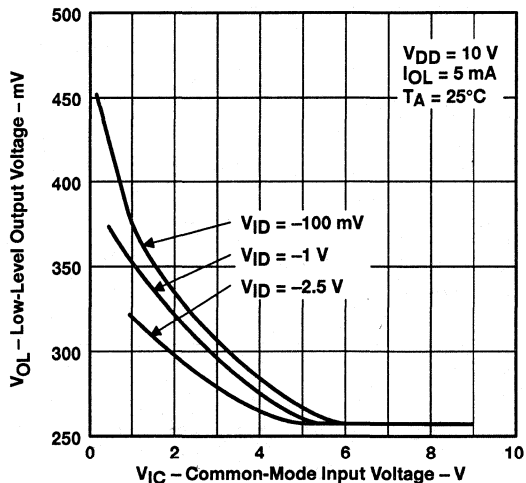
**Figure 8**

**LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE**



**Figure 9**

**LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE**



**Figure 10**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS†**

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
DIFFERENTIAL INPUT VOLTAGE**

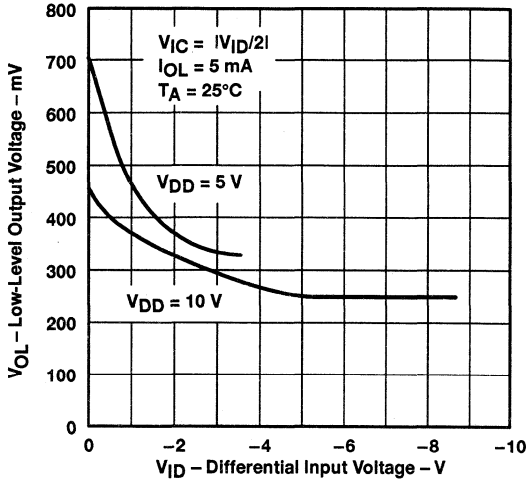


Figure 11

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

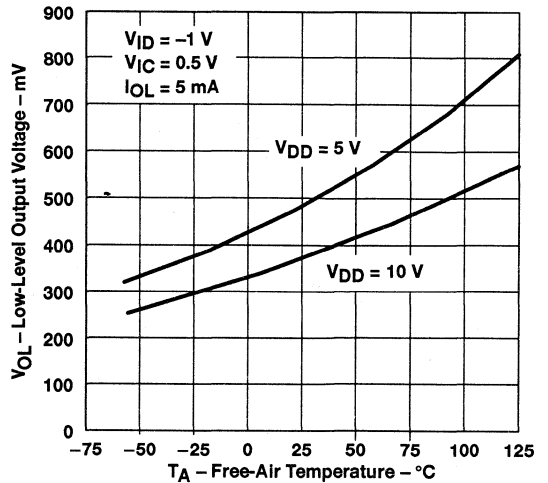


Figure 12

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

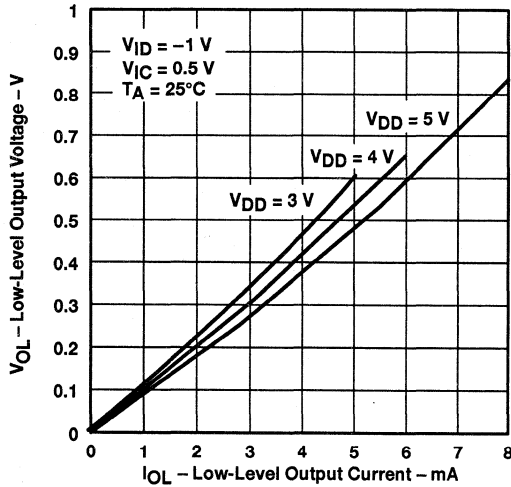


Figure 13

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
LOW-LEVEL OUTPUT CURRENT**

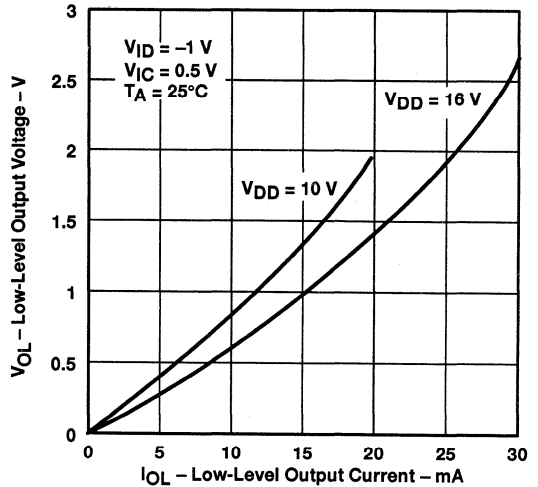
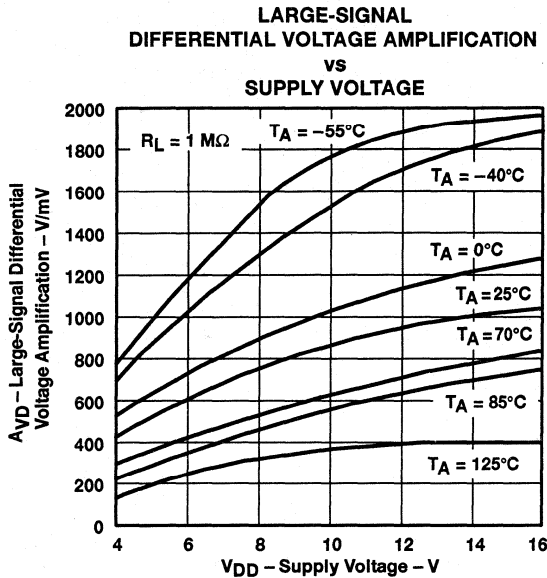


Figure 14

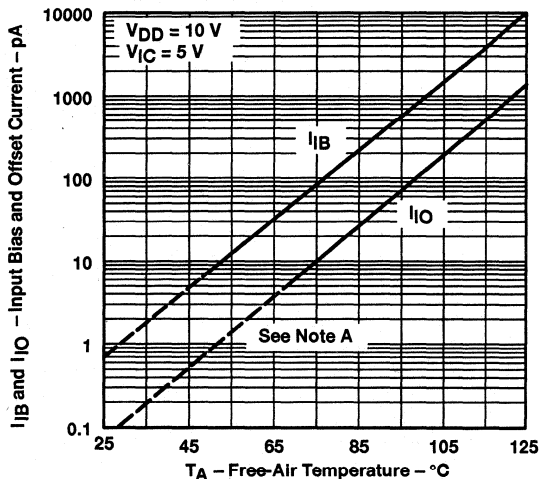
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**



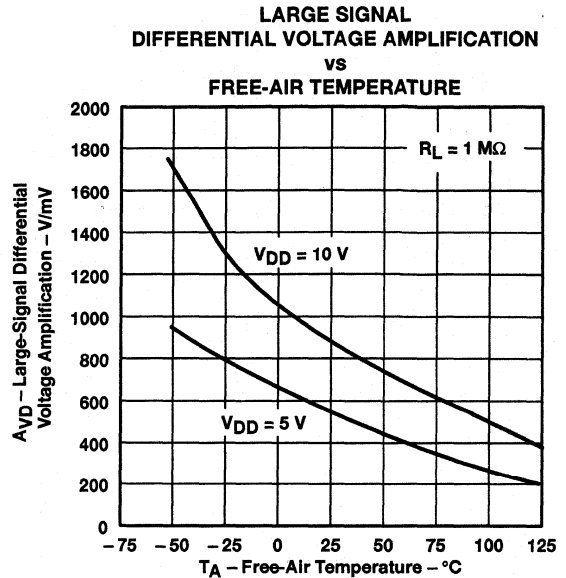
**Figure 15**

**INPUT BIAS AND OFFSET CURRENT  
vs  
FREE-AIR TEMPERATURE**



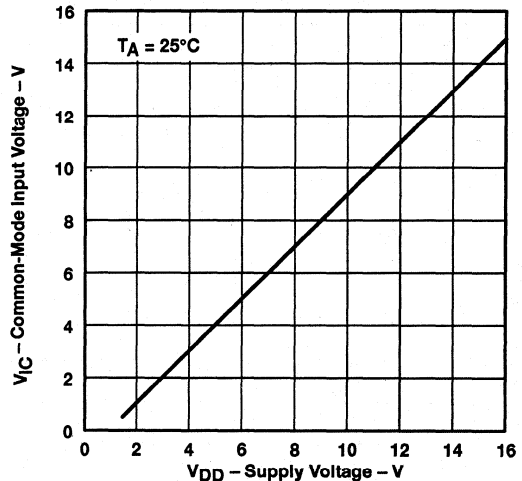
**Figure 17**

NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.



**Figure 16**

**COMMON-MODE INPUT VOLTAGE POSITIVE LIMIT  
vs  
SUPPLY VOLTAGE**



**Figure 18**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

**SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE**

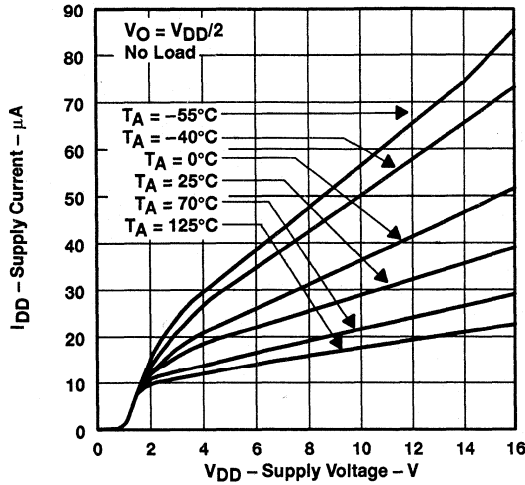


Figure 19

**SUPPLY CURRENT  
vs  
FREE-AIR TEMPERATURE**

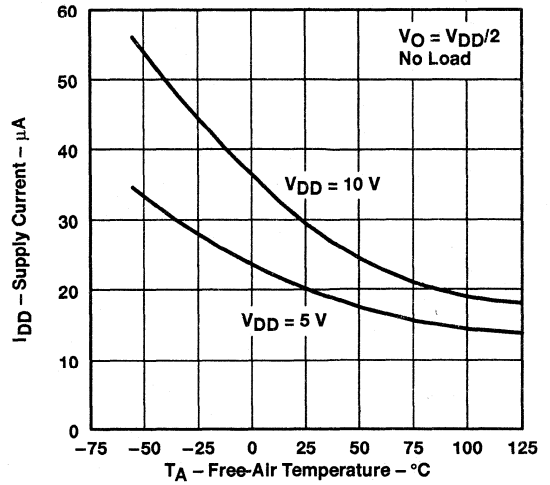


Figure 20

**SLEW RATE  
vs  
SUPPLY VOLTAGE**

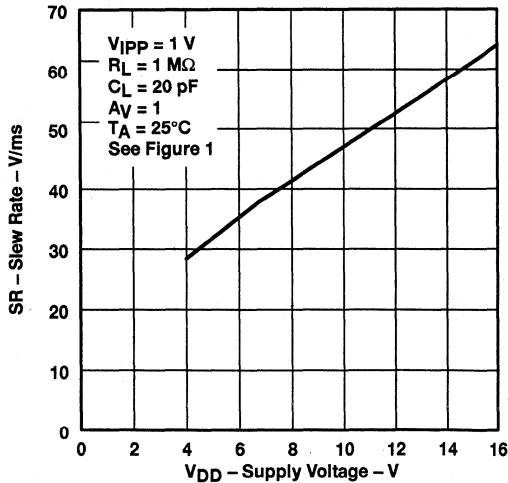


Figure 21

**SLEW RATE  
vs  
FREE-AIR TEMPERATURE**

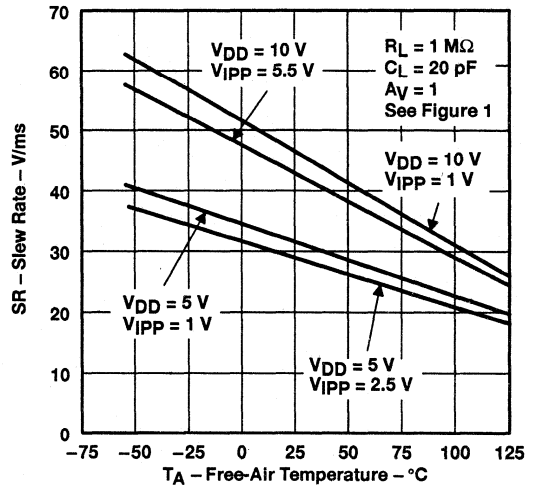
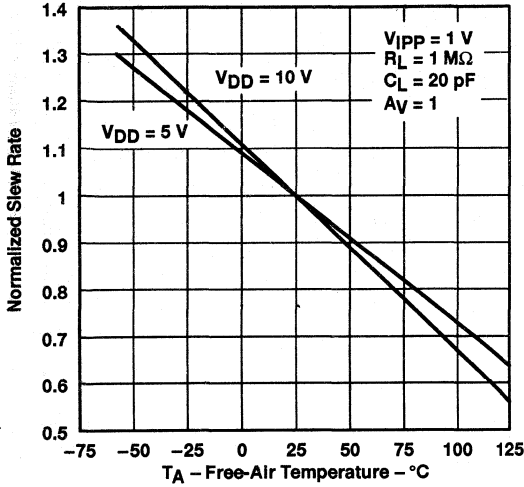


Figure 22

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

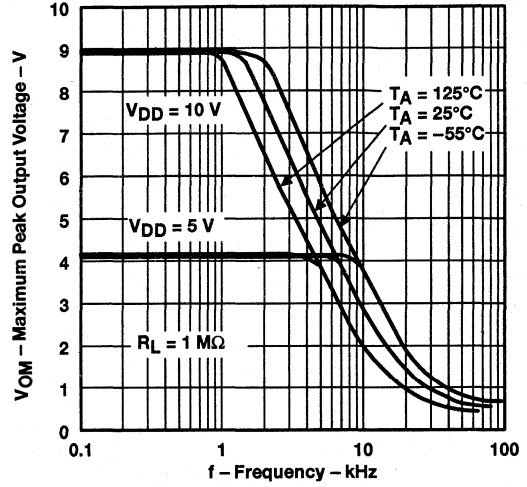
**TYPICAL CHARACTERISTICS†**

**NORMALIZED SLEW RATE  
 VS  
 FREE-AIR TEMPERATURE**



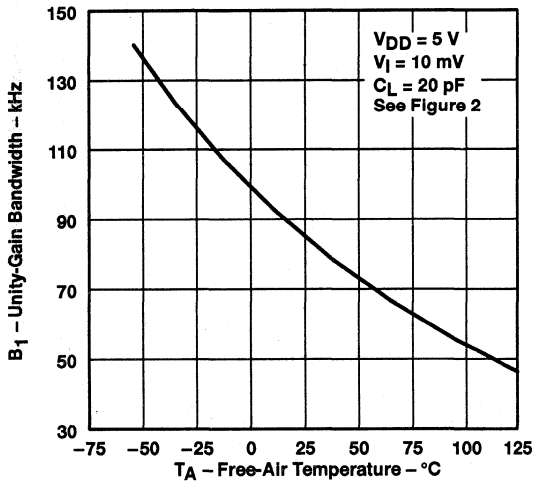
**Figure 23**

**MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 FREQUENCY**



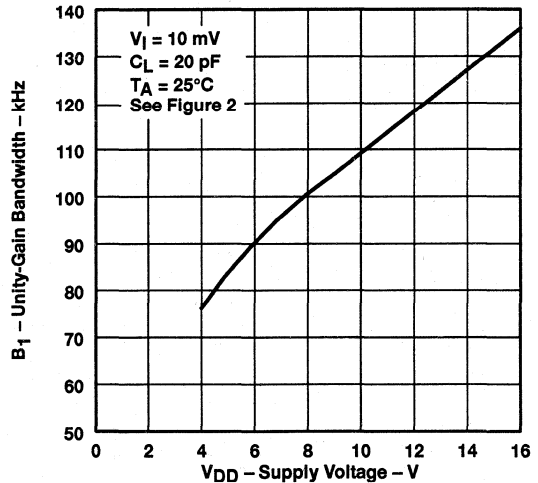
**Figure 24**

**UNITY-GAIN BANDWIDTH  
 VS  
 FREE-AIR TEMPERATURE**



**Figure 25**

**UNITY-GAIN BANDWIDTH  
 VS  
 SUPPLY VOLTAGE**



**Figure 26**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

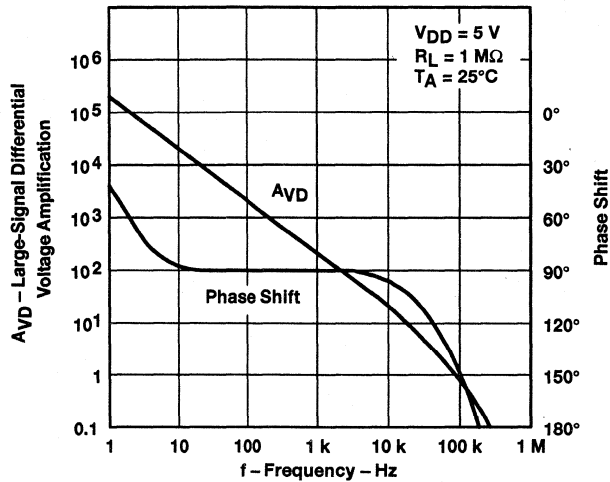


Figure 27

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

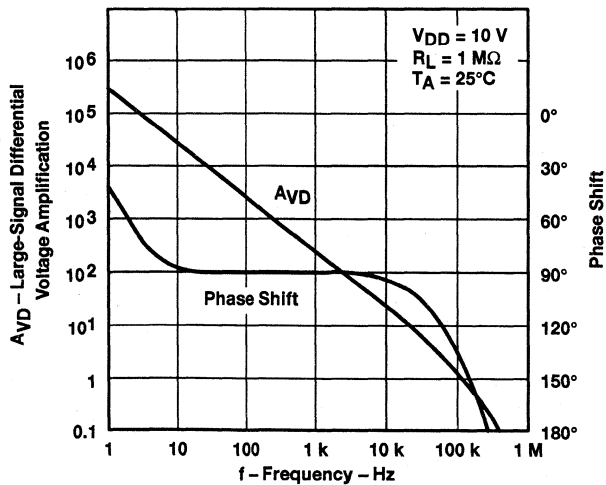


Figure 28

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

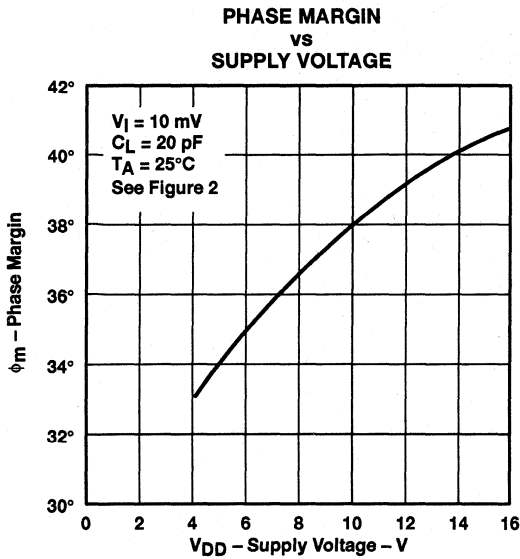


Figure 29

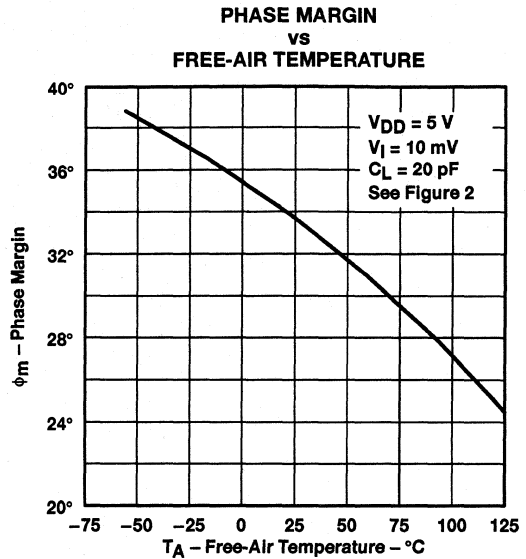


Figure 30

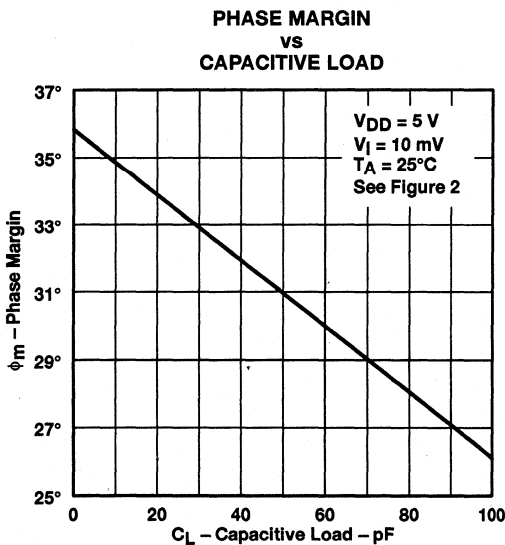


Figure 31

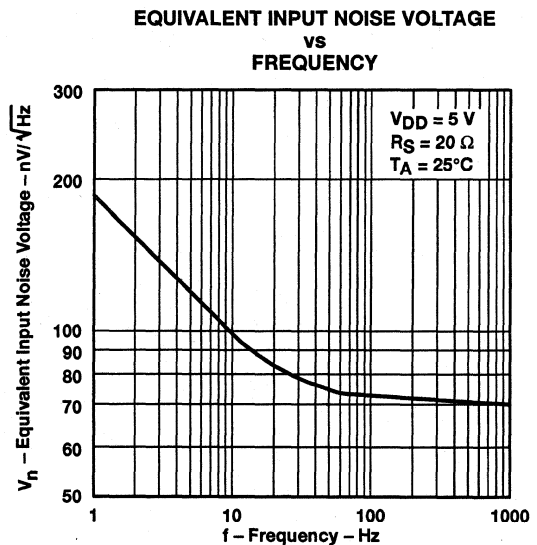


Figure 32

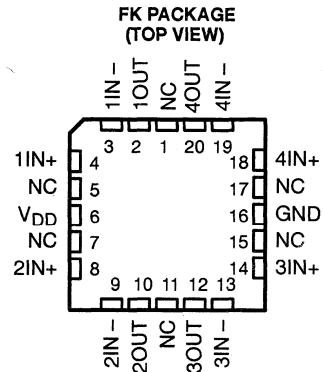
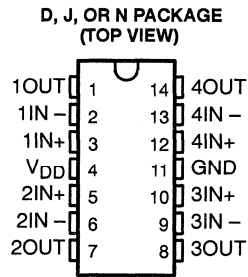
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC1079

## LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

- Power Dissipation as Low as 10  $\mu$ W Typ Per Amplifier
- Operates on a Single Silver-Oxide Watch Battery,  $V_{DD} = 1.4$  V Min
- $V_{IO} \dots 850$   $\mu$ V Max in DIP or Small-Outline Package
- Input Offset Voltage Drift  $\dots 0.1$   $\mu$ V/Month Typ, Including the First 30 Days
- High-Impedance LinCMOS™ Inputs  
 $I_{IB} = 0.6$  pA Typ
- High Open-Loop Gain  $\dots 800000$  Typ
- Output Drive Capability  $> 20$  mA
- Slew Rate  $\dots 47$  V/ms Typ
- Common-Mode Input Voltage Range Extends Below the Negative Rail
- Output Voltage Range Includes Negative Rail
- On-Chip ESD-Protection Circuitry
- 14-Pin Small-Outline Package Option Also Available in Tape and Reel



NC – No internal connection

### description

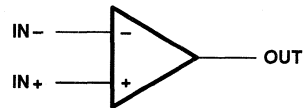
The TLC1079 operational amplifier offers ultra-low offset voltage, high gain, 110-kHz bandwidth, 47-V/ms slew rate, and just 150- $\mu$ W power dissipation per amplifier.

With a supply voltage of 1.4 V, common-mode input to the negative rail, and output swing to the negative rail, the TLC1079 is an ideal solution for low-voltage, battery-operated systems. The 20-mA output drive capability means that the TLC1079 can easily drive small resistive and large capacitive loads when needed, while maintaining ultra-low standby power dissipation.

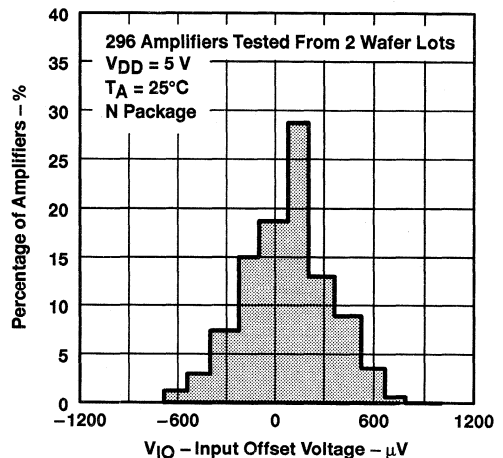
Since this device is functionally compatible as well as pin compatible with the TLC27L4 and TLC27L9, the TLC1079 easily upgrades existing designs that can benefit from its improved performance.

The TLC1079 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised when handling these devices as exposure to ESD may result in degradation of

### symbol (each amplifier)



### DISTRIBUTION OF TLC1079 INPUT OFFSET VOLTAGE



LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1992, Texas Instruments Incorporated

# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## description (continued)

the device parametric performance. The TLC1079 design also inhibits latch-up of the device inputs and outputs even with surge currents as large as 100 mA.

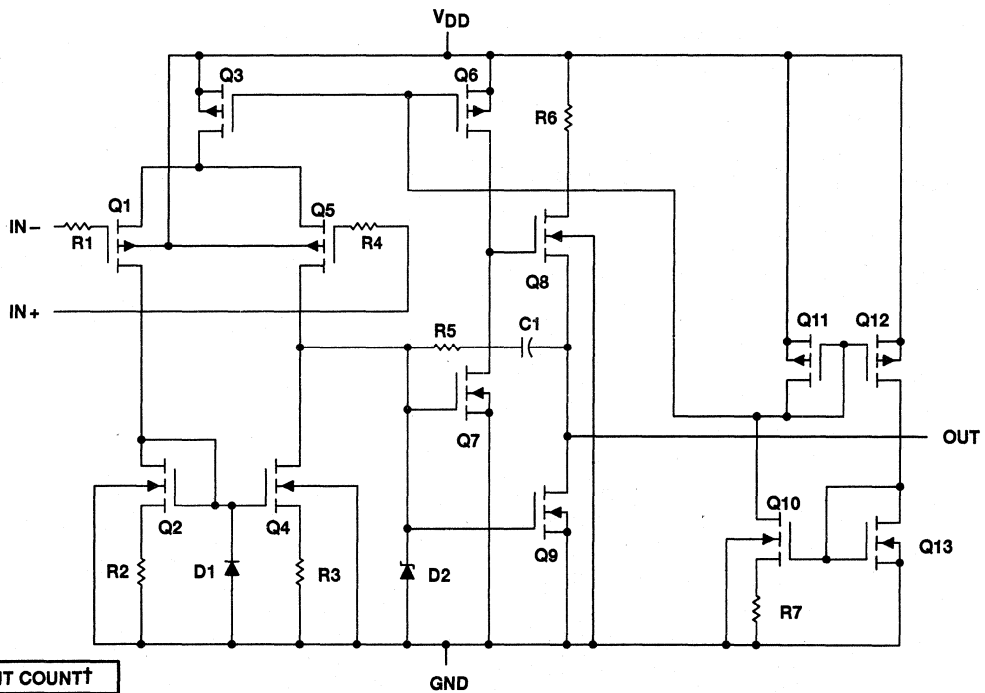
The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range -55°C to 125°C. The wide range of packaging options includes small-outline and chip-carrier versions for high-density system applications.

### AVAILABLE OPTIONS

T <sub>A</sub>	PACKAGE			
	SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (P)
0°C to 70°C	TLC1079CD	—	—	TLC1079CN
-40°C to 85°C	TLC1079ID	—	—	TLC1079IN
-55°C to 125°C	TLC1079MD	TLC1079MFK	TLC1079MJ	TLC1079MN

The D package is available taped and reeled. Add the suffix R to the device type (e.g., TLC1079CDR).

## equivalent schematic (each amplifier)



COMPONENT COUNT†	
Transistors	76
Diodes	24
Resistors	32
Capacitors	4

† Includes all four amplifiers, bias circuitry, and zener trim circuitry.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# TLC1079

## LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD}$ (see Note 1)	18 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 to $V_{DD}$
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$ (see Note 3)	45 mA
Duration of short-circuit current at (or below) $T_A = 25^\circ\text{C}$ (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at IN+ with respect to IN-.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation ratings are not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	230 mW

### recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD}$		1.4	16	3	16	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V	-0.2	4	-0.2	4	0	4	V
	$V_{DD} = 10$ V	-0.2	9	-0.2	9	0	9	
Operating free-air temperature, $T_A$		0	70	-40	85	-55	125	°C



# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC1079C						UNIT	
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω, R <sub>I</sub> = 1 MΩ	25°C		190	850	200		1150	μV
Full range			1200			1500				
αV <sub>IO</sub>	Temperature coefficient of input offset voltage	25°C to 70°C	1.1			1			μV/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C		0.1		0.1		pA	
70°C			7	300	7	300				
I <sub>IB</sub>	Input bias current (see Note 4)		25°C	0.6		0.7		pA		
70°C		40	600	50	600					
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)	25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V			
		Full range	-0.2 to 3.5		-0.2 to 8.5		V			
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 MΩ	25°C		3.2	4.1	8.2	8.9	V	
			0°C		3.2	4.1	8.2	8.9		
			70°C		3.2	4.2	8.2	8.9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C		0	25	0	25	mV	
			0°C		0	25	0	25		
			70°C		0	25	0	25		
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 MΩ, See Note 6	25°C		250	525	500	850	V/mV	
			0°C		250	700	500	1010		
			70°C		200	380	350	660		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C		70	95	75	97	dB	
			0°C		70	95	75	97		
			70°C		70	95	75	97		
k <sub>SVR</sub>	Supply-voltage rejection ratio (ΔV <sub>DD</sub> /ΔV <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C		75	98	75	98	dB	
			0°C		75	98	75	98		
			70°C		75	98	75	98		
I <sub>DD</sub>	Supply current (four amplifiers)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C		40	68	57	92	μA	
			0°C		48	84	72	132		
			70°C		31	56	44	80		

† Full range is 0°C to 70°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
5. This range also applies to each input individually.  
6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.



# TLC1079

## LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

### electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC1079I						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub>	Input offset voltage	V <sub>O</sub> = 1.4 V, V <sub>IC</sub> = 0, R <sub>S</sub> = 50 $\Omega$ , R <sub>I</sub> = 1 M $\Omega$	25°C	190 850		200 1150		$\mu$ V	
	Full range		1350		1650				
$\alpha$ V <sub>IO</sub>	Temperature coefficient of input offset voltage		25°C to 85°C	1.1		1		$\mu$ V/°C	
I <sub>IO</sub>	Input offset current (see Note 4)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2	25°C	0.1		0.1		pA	
			85°C	24	1000	26	1000		
I <sub>IB</sub>	Input bias current (see Note 4)		25°C	0.6		0.7		pA	
		85°C	200	2000	220	2000			
V <sub>ICR</sub>	Common-mode input voltage range (see Note 5)		25°C	-0.2 to 4	-0.3 to 4.2	-0.2 to 9	-0.3 to 9.2	V	
			Full range	-0.2 to 3.5		-0.2 to 8.5		V	
V <sub>OH</sub>	High-level output voltage	V <sub>ID</sub> = 100 mV, R <sub>L</sub> = 1 M $\Omega$	25°C	3.2	4.1	8.2	8.9	V	
			-40°C	3.2	4.1	8.2	8.9		
			85°C	3.2	4.2	8.2	8.9		
V <sub>OL</sub>	Low-level output voltage	V <sub>ID</sub> = -100 mV, I <sub>OL</sub> = 0	25°C	0 25		0 25		mV	
			-40°C	0 25		0 25			
			85°C	0 25		0 25			
A <sub>VD</sub>	Large-signal differential voltage amplification	R <sub>L</sub> = 1 M $\Omega$ , See Note 6	25°C	250	525	500	850	V/mV	
			-40°C	250	900	500	1550		
			85°C	150	330	250	585		
CMRR	Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub>	25°C	70	95	75	97	dB	
			-40°C	70	95	75	97		
			85°C	70	95	75	97		
k <sub>SVR</sub>	Supply-voltage rejection ratio ( $\Delta$ V <sub>DD</sub> / $\Delta$ V <sub>IO</sub> )	V <sub>DD</sub> = 5 V to 10 V, V <sub>O</sub> = 1.4 V	25°C	75	98	75	98	dB	
			-40°C	75	98	75	98		
			85°C	75	98	75	98		
I <sub>DD</sub>	Supply current (four amplifiers)	V <sub>O</sub> = V <sub>DD</sub> /2, V <sub>IC</sub> = V <sub>DD</sub> /2, No load	25°C	40 68		57 92		$\mu$ A	
			-40°C	62 108		98 172			
			85°C	29 52		40 72			

† Full range is -40°C to 85°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
 5. This range also applies to each input individually.  
 6. At V<sub>DD</sub> = 5 V, V<sub>O</sub> = 0.25 V to 2 V; at V<sub>DD</sub> = 10 V, V<sub>O</sub> = 1 V to 6 V.

# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC1079M						UNIT	
			$V_{DD} = 5\text{ V}$			$V_{DD} = 10\text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$	Input offset voltage	$V_O = 1.4\text{ V}$ , $V_{IC} = 0$ , $R_S = 50\ \Omega$ , $R_I = 1\text{ M}\Omega$	25°C		190	850		200	1150	$\mu\text{V}$
		Full range			1600			1900		
$\alpha V_{IO}$	Temperature coefficient of input offset voltage		25°C to 125°C		1.4			1.4		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$	Input offset current (see Note 4)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$	25°C		0.1			0.1		pA
			125°C		1.4	15		1.8	15	nA
$I_{IB}$	Input bias current (see Note 4)		25°C		0.6			0.7		pA
			125°C		9	35		10	35	nA
$V_{ICR}$	Common-mode input voltage range (see Note 5)		25°C	0 to 4	-0.3 to 4.2		0 to 9	-0.3 to 9.2		V
			Full range	0 to 3.5		0 to 8.5				V
$V_{OH}$	High-level output voltage	$V_{ID} = 100\text{ mV}$ , $R_L = 1\text{ M}\Omega$	25°C	3.2	4.1		8.2	8.9		V
			-55°C	3.2	4.1		8.2	8.9		
			125°C	3.2	4.2		8.2	9		
$V_{OL}$	Low-level output voltage	$V_{ID} = -100\text{ mV}$ , $I_{OL} = 0$	25°C		0	25		0	25	mV
			-55°C		0	25		0	25	
			125°C		0	25		0	25	
$AVD$	Large-signal differential voltage amplification	$R_L = 1\text{ M}\Omega$ , See Note 6	25°C	250	525		500	850		V/mV
			-55°C	250	950		500	1750		
			125°C	35	200		75	380		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	70	95		75	97		dB
			-55°C	70	95		75	97		
			125°C	70	85		75	91		
$kSVR$	Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 5\text{ V to }10\text{ V}$ , $V_O = 1.4\text{ V}$	25°C	75	98		75	98		dB
			-55°C	70	98		70	98		
			125°C	70	98		70	98		
$I_{DD}$	Supply current (four amplifiers)	$V_O = V_{DD}/2$ , $V_{IC} = V_{DD}/2$ , No load	25°C		40	68		57	92	$\mu\text{A}$
			-55°C		69	120		111	192	
			125°C		27	48		35	60	

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At  $V_{DD} = 5\text{ V}$ ,  $V_O = 0.25\text{ V to }2\text{ V}$ ; at  $V_{DD} = 10\text{ V}$ ,  $V_O = 1\text{ V to }6\text{ V}$ .



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# TLC1079

## LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

### operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1079C						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		0°C		35			51		
		70°C		27			38		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		0°C		100			125		
		70°C		65			90		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	25°C		34°			38°		
		0°C		36°			40°		
		70°C		30°			34°		

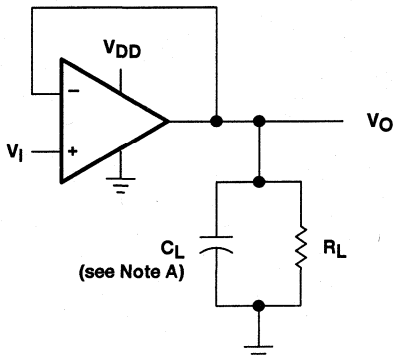
### operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1079I						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		-40°C		39			59		
		85°C		25			34		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		-40°C		130			155		
		85°C		55			80		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	25°C		34°			38°		
		-40°C		38°			42°		
		85°C		28°			32°		

### operating characteristics at specified free-air temperature

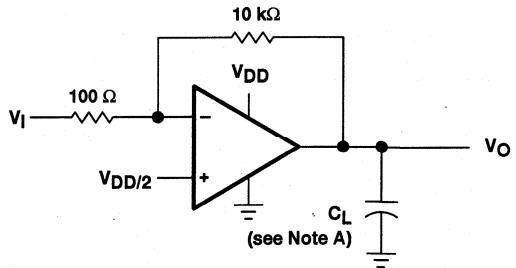
PARAMETER	TEST CONDITIONS	T <sub>A</sub>	TLC1079M						UNIT
			V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 1 M $\Omega$ , C <sub>L</sub> = 20 pF, V <sub>I</sub> PP = 1 V, See Figure 1	25°C		32			47	V/ms	
		-55°C		41			63		
		125°C		20			27		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, R <sub>S</sub> = 20 $\Omega$	25°C		68			68	nV/ $\sqrt$ Hz	
B <sub>1</sub> Unity-gain bandwidth	C <sub>L</sub> = 20 pF, See Figure 2	25°C		85			110	kHz	
		-55°C		140			165		
		125°C		45			70		
		25°C		34°			38°		
$\phi_m$ Phase margin at unity gain	C <sub>L</sub> = 20 pF, See Figure 2	-55°C		39°			43°		
		125°C		25°			29°		

**PARAMETER MEASUREMENT INFORMATION**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 1. Slew-Rate Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 2. Unity-Gain Bandwidth and Phase-Margin Test Circuit**

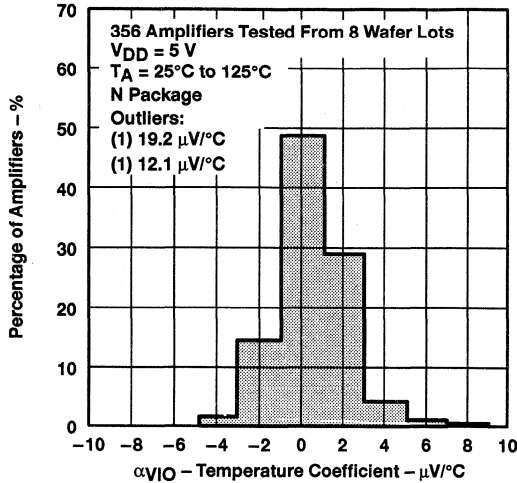
**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	Distribution	3, 4
$V_{OH}$	High-level output voltage	vs High-level output current	5, 6
		vs Supply voltage	7
		vs Free-air temperature	8
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	9, 10
		vs Differential input voltage	11
		vs Free-air temperature	12
		vs Low-level output current	13, 14
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	15
		vs Free-air temperature	16
		vs Frequency	27, 28
$I_{IB}$	Input bias current	vs Free-air temperature	17
$I_{IO}$	Input offset current	vs Free-air temperature	17
$V_{IC}$	Common-mode input voltage	vs Supply voltage	18
$I_{DD}$	Supply current	vs Supply voltage	19
		vs Free-air temperature	20
SR	Slew rate	vs Supply voltage	21
		vs Free-air temperature	22
		Normalized slew rate	vs Free-air temperature
$V_{OM}$	Maximum peak output voltage	vs Frequency	24
$B_1$	Unity-gain bandwidth	vs Free-air temperature	25
		vs Supply voltage	26
	Phase shift	vs Frequency	27, 28
$\phi_m$	Phase margin	vs Supply voltage	29
		vs Free-air temperature	30
		vs Load capacitance	31
$V_n$	Equivalent input noise voltage	vs Frequency	32

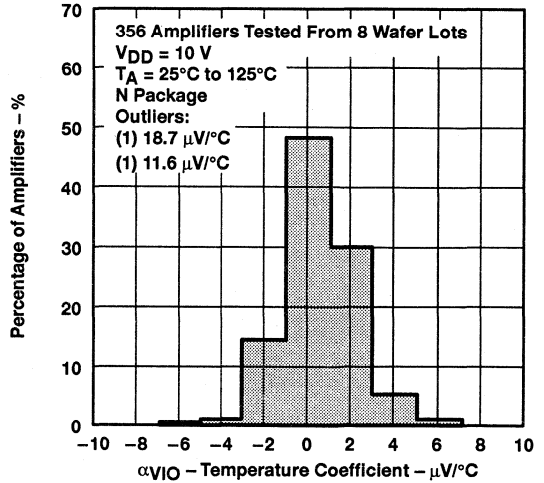
**TYPICAL CHARACTERISTICS**

**DISTRIBUTION OF TLC1079  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT**



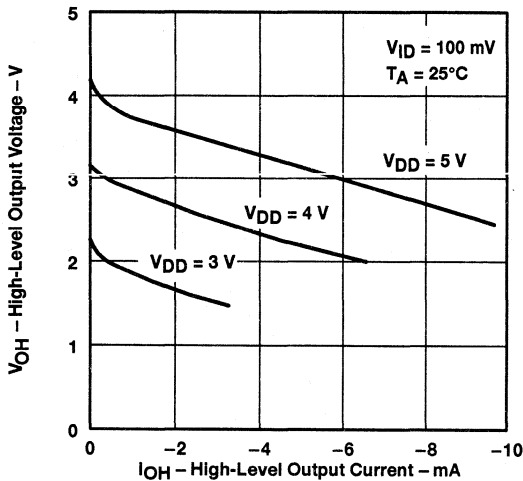
**Figure 3**

**DISTRIBUTION OF TLC1079  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT**



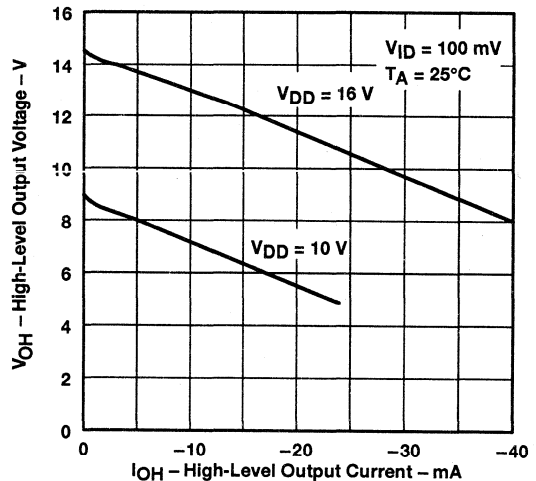
**Figure 4**

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**



**Figure 5**

**HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT**



**Figure 6**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## TYPICAL CHARACTERISTICS†

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE**

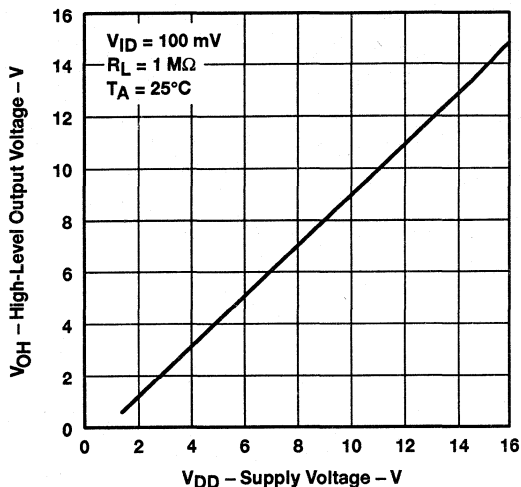


Figure 7

**HIGH-LEVEL OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**

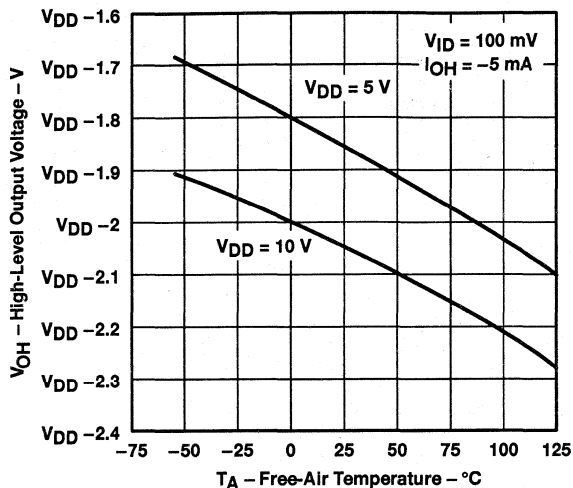


Figure 8

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE**

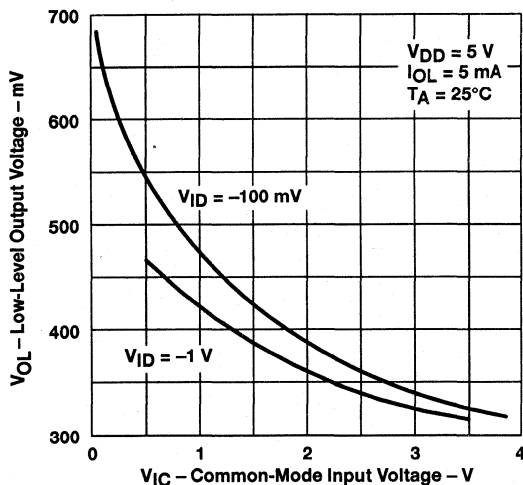


Figure 9

**LOW-LEVEL OUTPUT VOLTAGE  
vs  
COMMON-MODE INPUT VOLTAGE**

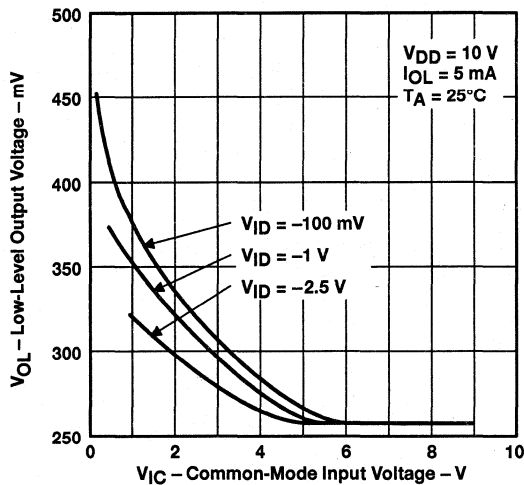


Figure 10

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



**TYPICAL CHARACTERISTICS†**

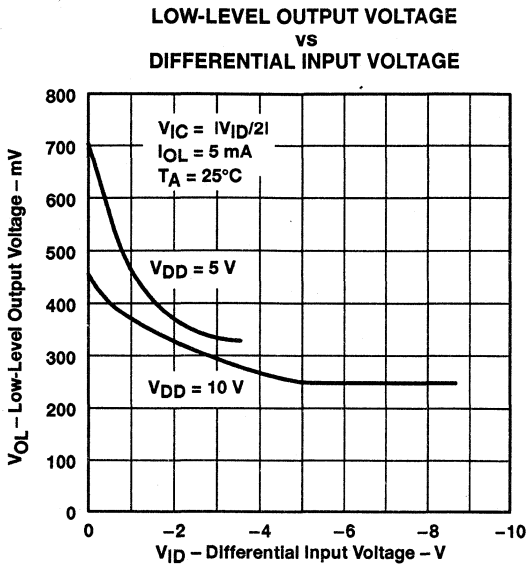


Figure 11

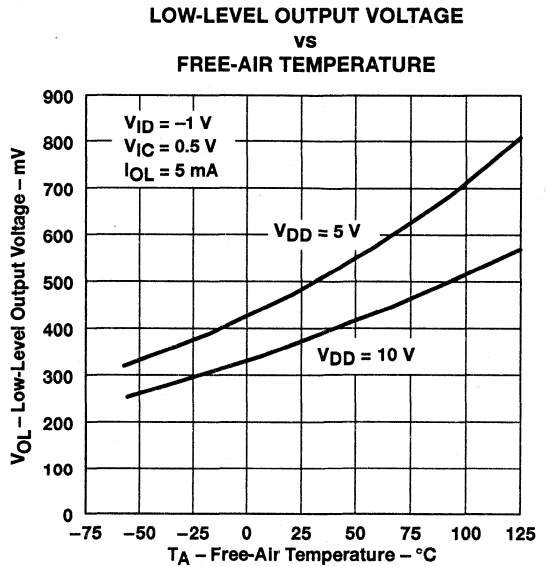


Figure 12

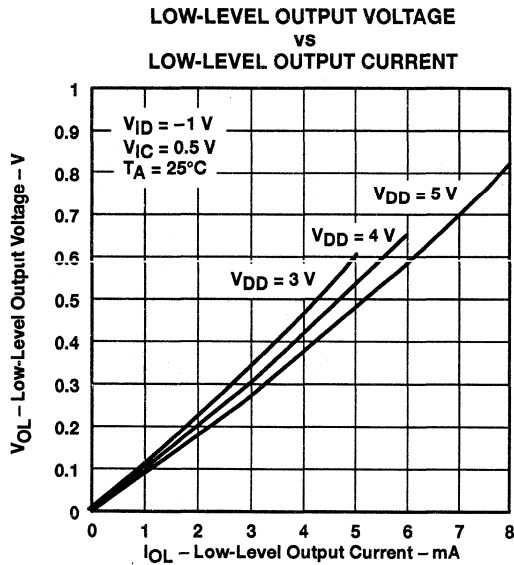


Figure 13

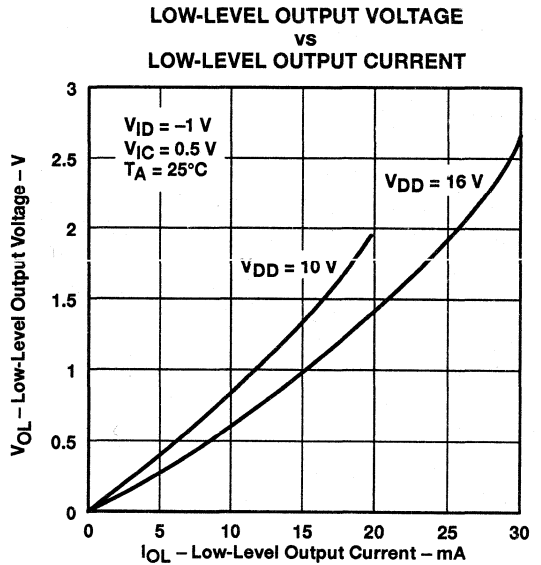
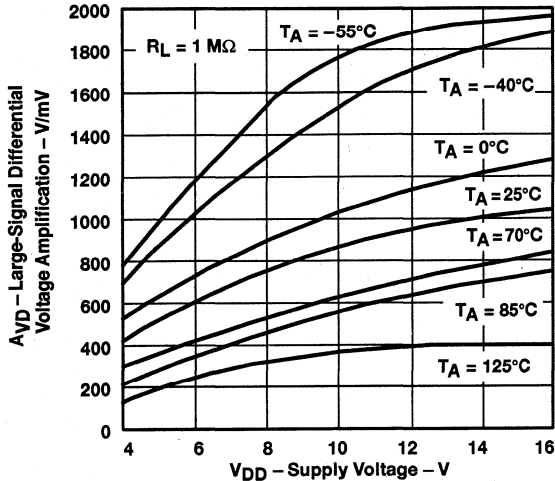


Figure 14

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

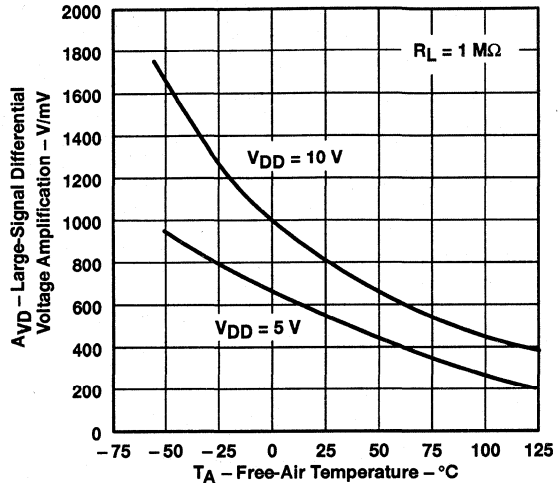
**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs SUPPLY VOLTAGE**



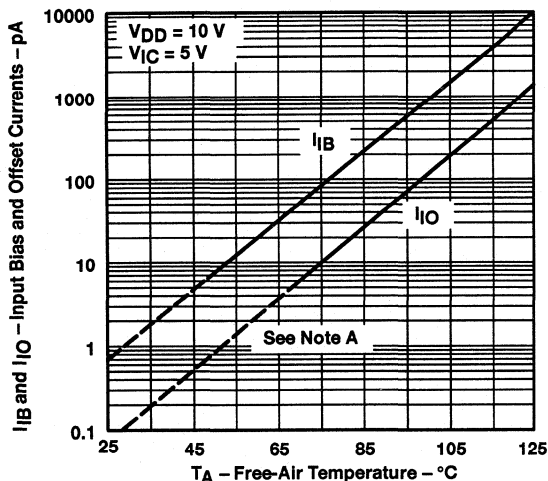
**Figure 15**

**LARGE SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs FREE-AIR TEMPERATURE**



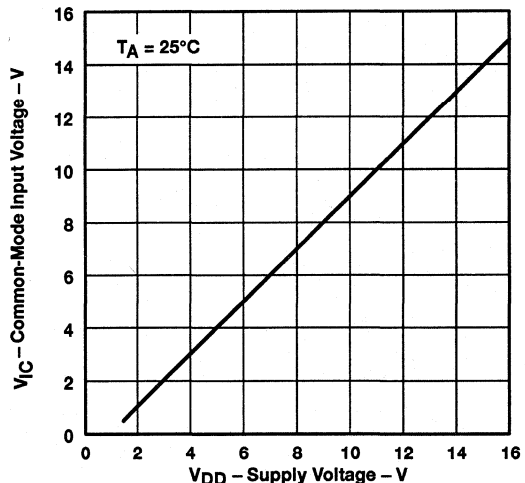
**Figure 16**

**INPUT BIAS AND OFFSET CURRENT vs FREE-AIR TEMPERATURE**



**Figure 17**

**COMMON-MODE INPUT VOLTAGE POSITIVE LIMIT vs SUPPLY VOLTAGE**



**Figure 18**

NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

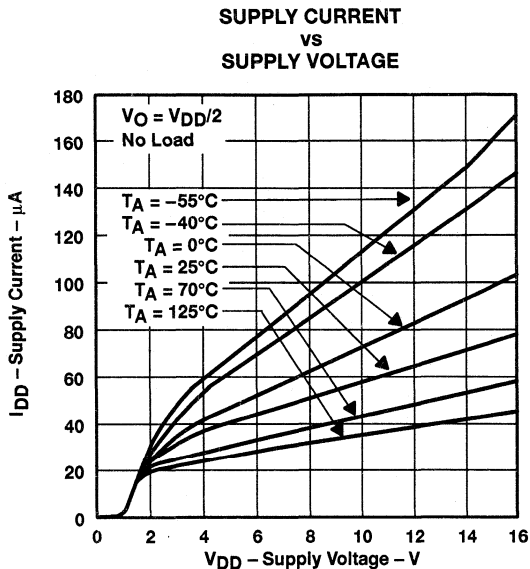


Figure 19

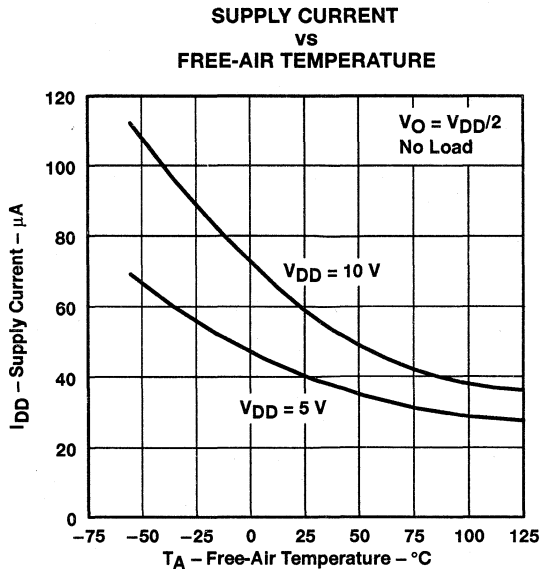


Figure 20

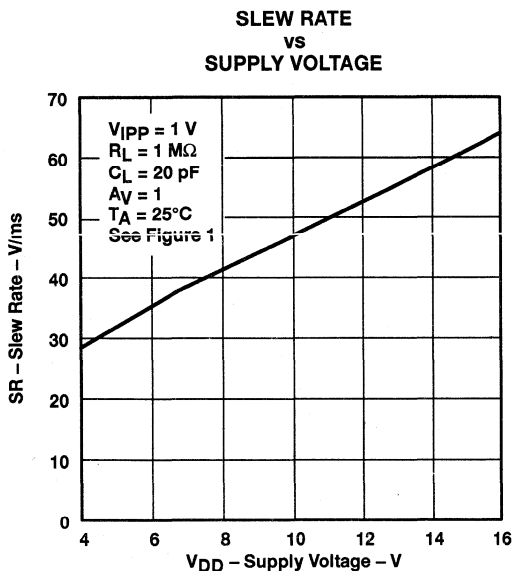


Figure 21

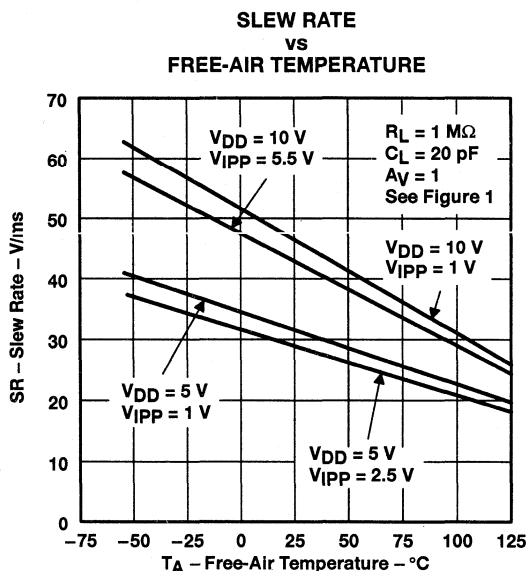


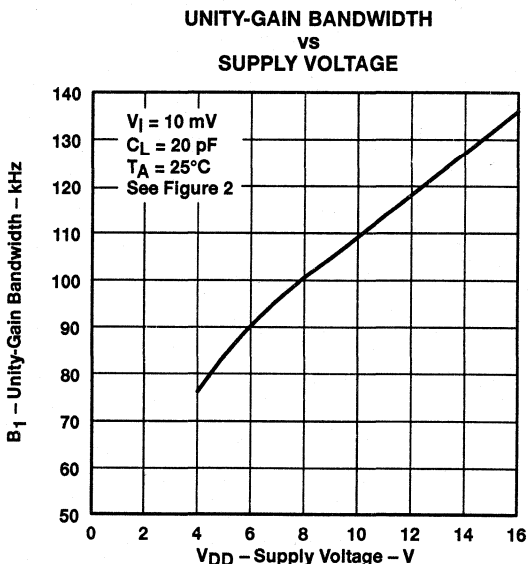
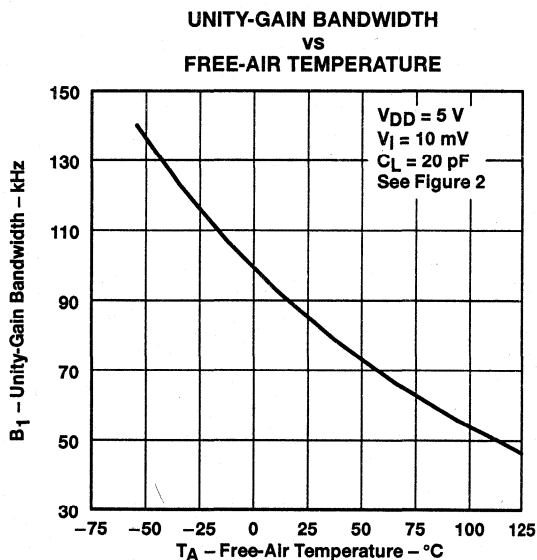
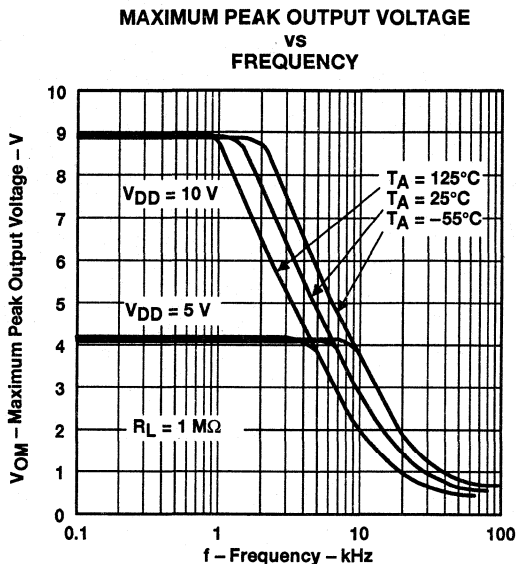
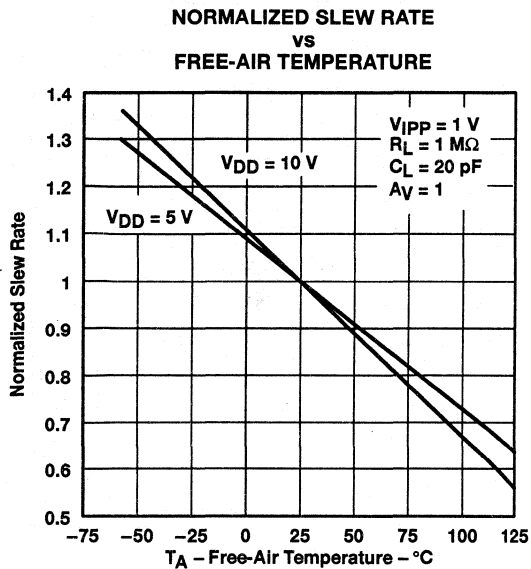
Figure 22

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY**

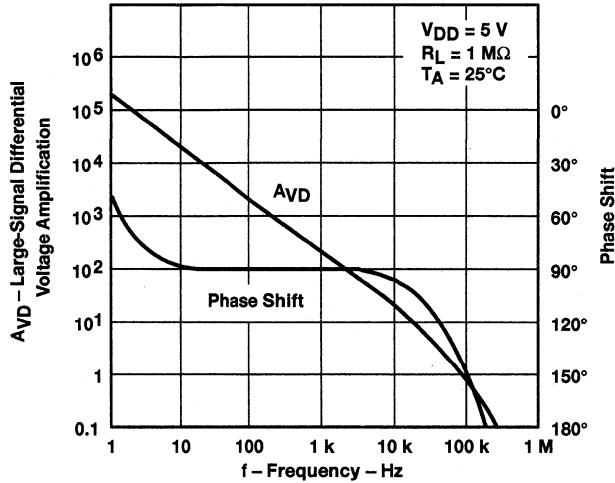


Figure 27

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY**

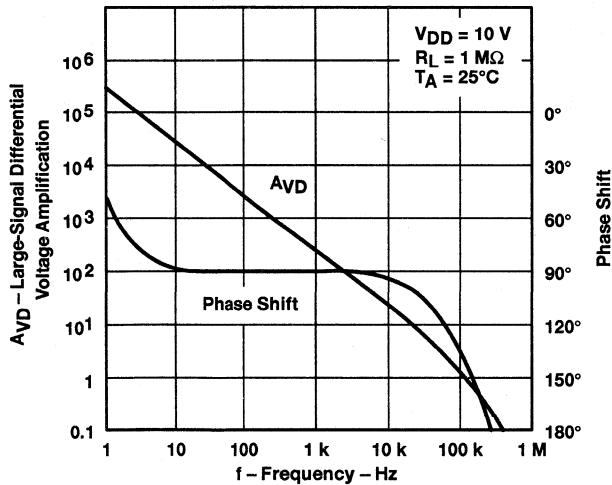


Figure 28

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC1079 LinCMOS™ $\mu$ POWER PRECISION QUAD OPERATIONAL AMPLIFIER

SLOS023B – AUGUST 1988 – REVISED DECEMBER 1992

## TYPICAL CHARACTERISTICS†

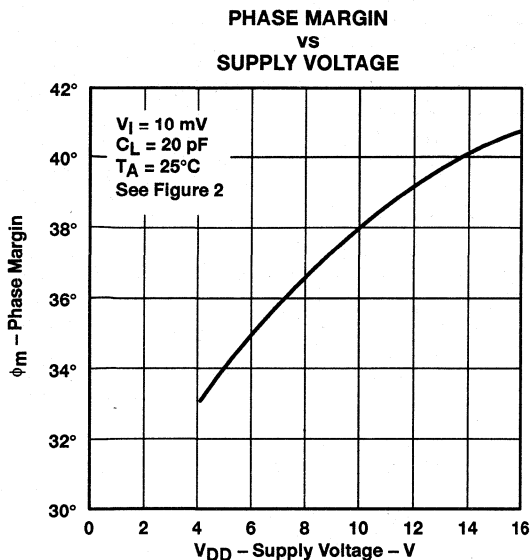


Figure 29

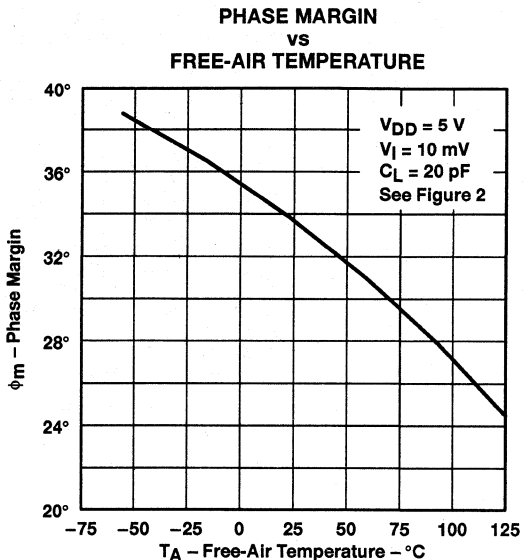


Figure 30

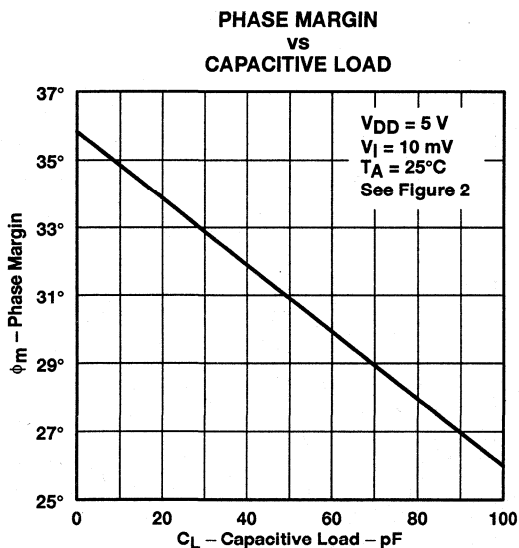


Figure 31

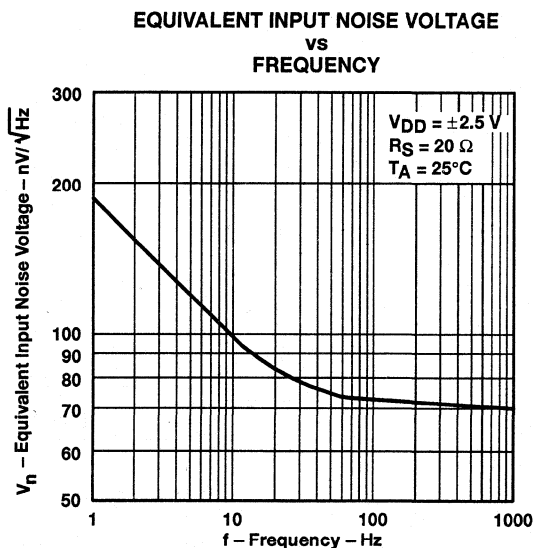


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

- **TLC2201B Is 100% Tested for Noise:**  
30 nV/√Hz Max at f = 10 Hz  
12 nV/√Hz Max at f = 1 kHz
- **Low Input Offset Voltage . . . 200 μV Max**
- **Excellent Offset Voltage Stability With Temperature . . . 0.5 μV/°C Typ**
- **Low Input Bias Current**  
1 pA at T<sub>A</sub> = 25°C
- **Fully Specified for Both Single-Supply and Split-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**

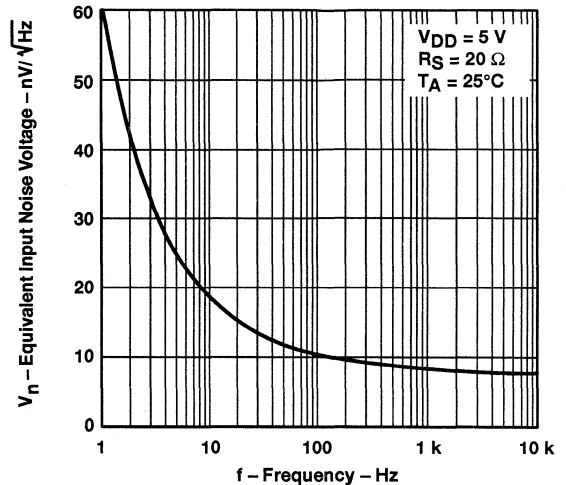
## description

The TLC2201, TLC2201A, TLC2201B, and TLC2201Y are precision, low-noise operational amplifiers using Texas Instruments Advanced LinCMOS™ process. These devices combine the noise performance of the lowest-noise JFET amplifiers with the dc precision available previously only in bipolar amplifiers. The Advanced LinCMOS™ process uses silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. In addition, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The combination of excellent dc and noise performance with a common-mode input voltage range that includes the negative rail makes these devices an ideal choice for high-impedance, low-level signal conditioning applications in either single-supply or split-supply configurations.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

TYPICAL EQUIVALENT  
INPUT NOISE VOLTAGE  
vs  
FREQUENCY



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	V <sub>n</sub> max f = 10 Hz AT 25°C	V <sub>n</sub> max f = 1 kHz AT 25°C	PACKAGED DEVICES				CHIP FORM (Y)
				SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	
0°C to 70°C	200 μV 200 μV 500 μV	35 nV/√Hz 30 nV/√Hz —	15 nV/√Hz 12 nV/√Hz —	TLC2201ACD TLC2201BCD TLC2201CD	—	—	TLC2201ACP TLC2201BCP TLC2201CP	TLC2201Y
–40°C to 85°C	200 μV 200 μV 500 μV	35 nV/√Hz 30 nV/√Hz —	15 nV/√Hz 12 nV/√Hz —	TLC2201AID TLC2201BID TLC2201ID	—	—	TLC2201AIP TLC2201BIP TLC2201IP	—
–55°C to 125°C	200 μV 200 μV 500 μV	35 nV/√Hz 30 nV/√Hz —	15 nV/√Hz 12 nV/√Hz —	TLC2201AMD TLC2201BMD TLC2201MD	TLC2201AMFK TLC2201BMFK TLC2201MFK	TLC2201AMJG TLC2201BMJG TLC2201MJG	TLC2201AMP TLC2201BMP TLC2201MP	—

The D packages are available taped and reeled. Add R suffix to device type (e.g., TLC2201BCDR). Chip-form versions are tested at 25°C only.

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

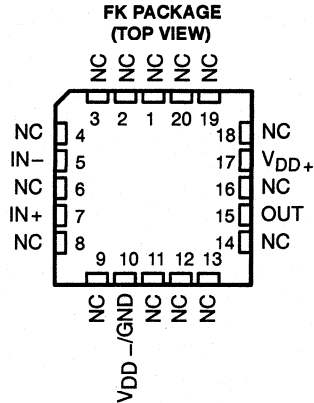
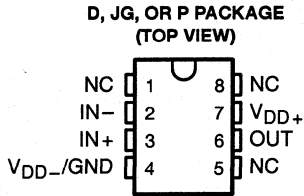
# TLC2201, TLC2201A, TLC2201B, TLC2201Y

## Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

### description (continued)

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.



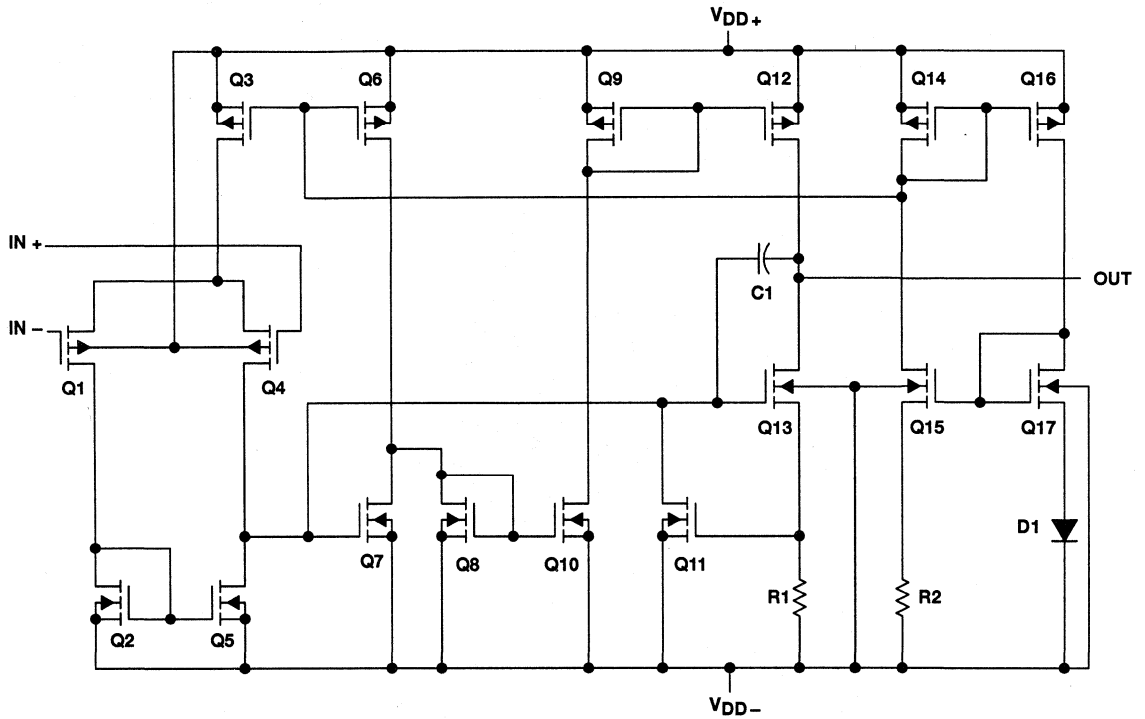
NC – No internal connection



TLC2201, TLC2201A, TLC2201B, TLC2201Y  
 Advanced LinCMOS™ LOW-NOISE PRECISION  
 OPERATIONAL AMPLIFIERS

SLOS021A - NOVEMBER 1988 - REVISED AUGUST 1994

equivalent schematic (each amplifier)



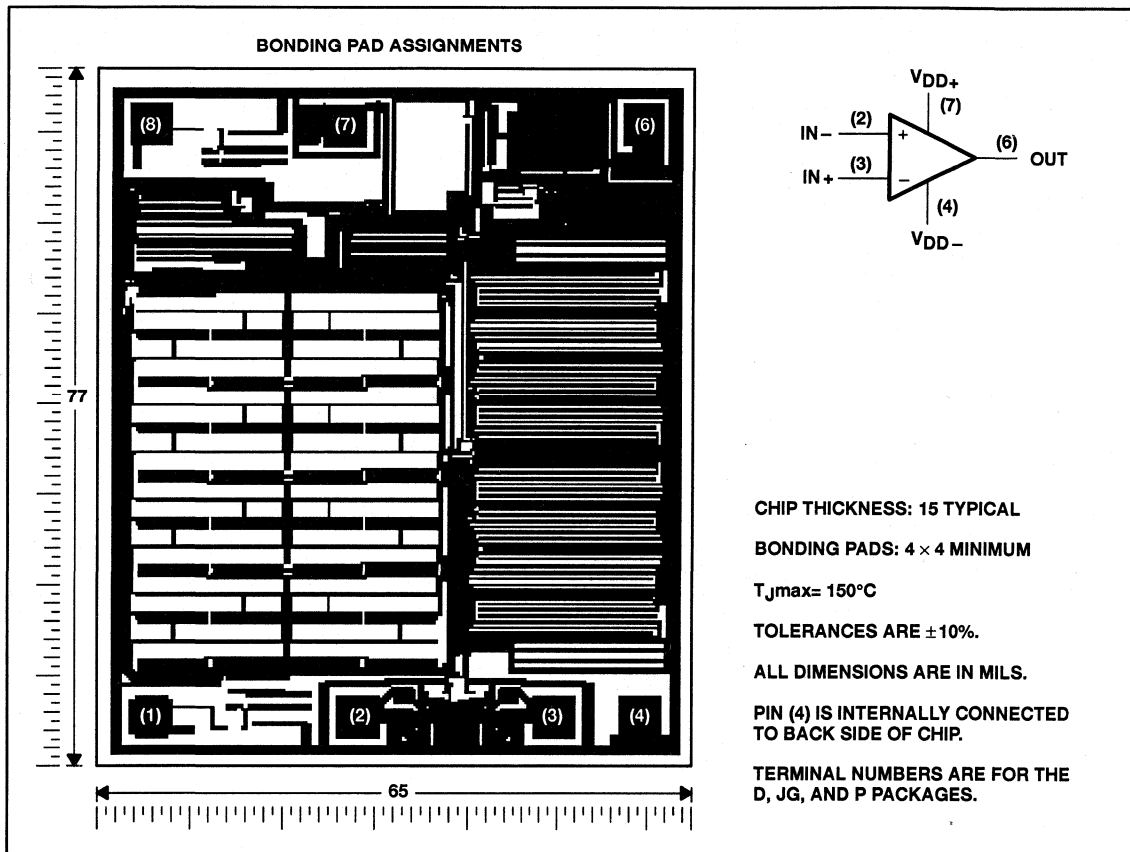
COMPONENT COUNT	
Transistors	17
Diodes	1
Resistors	2
Capacitors	1

# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

## TLC2201Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC2201C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding path. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

## absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 16$ V
Input voltage, $V_I$ (any input) (see Note 1)	$\pm 8$ V
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .
  2. Differential voltages are at IN+ with respect to IN-.
  3. The output can be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

## recommended operating conditions

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	$\pm 2.3$	$\pm 8$	$\pm 2.3$	$\pm 8$	$\pm 2.3$	$\pm 8$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, \quad R_S = 50 \Omega$	25°C		100	500	$\mu$ V
		Full range			600	
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu$ V/°C
		25°C		0.001	0.005	
Input offset voltage long-term drift (see Note 4)		25°C		0.5		pA
$I_{IO}$ Input offset current		Full range			100	
		$I_{IB}$ Input bias current	25°C		1	
Full range					100	
$V_{ICR}$ Common-mode input voltage range		$R_S = 50 \Omega$	Full range	-5 to 2.7		V
$V_{OM+}$ Maximum positive peak output voltage swing		$R_L = 10 \text{ k}\Omega$	25°C	4.7	4.8	V
	Full range		4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	25°C		-4.7	-4.9	V	
	Full range		-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4 \text{ V}, \quad R_L = 500 \text{ k}\Omega$	25°C	400	560	V/mV	
		Full range	300			
	$V_O = \pm 4 \text{ V}, \quad R_L = 10 \text{ k}\Omega$	25°C	90	100		
		Full range	70			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0, \quad R_S = 50 \Omega$	25°C	90	115	dB	
		Full range	85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3 \text{ V to } \pm 8 \text{ V}$	25°C	90	110	dB	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 0, \quad \text{No load}$	25°C		1.1 1.5	mA	
		Full range		1.5		

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3 \text{ V}, \quad R_L = 10 \text{ k}\Omega,$ $C_L = 100 \text{ pF}$	25°C	2	2.7	V/ $\mu$ s	
		Full range	1.5			
$V_n$ Equivalent input noise voltage	$f = 10 \text{ Hz}$	25°C		18	nV/ $\sqrt{\text{Hz}}$	
	$f = 1 \text{ kHz}$	25°C		8		
$V_N(PP)$ Peak-to-peak equivalent input noise voltage	$f = 0.1 \text{ to } 1 \text{ Hz}$	25°C		0.5	$\mu$ V	
	$f = 0.1 \text{ to } 10 \text{ Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	fA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10 \text{ kHz}, \quad R_L = 10 \text{ k}\Omega,$ $C_L = 100 \text{ pF}$	25°C		1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10 \text{ k}\Omega, \quad C_L = 100 \text{ pF}$	25°C		48°		

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A \uparrow$	TLC2201AC			TLC2201BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80		200	80		200	$\mu\text{V}$
		Full range	300			300			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005		$\mu\text{V}/\text{mo}$
		25°C	0.5			0.5			
$I_{IO}$ Input offset current		25°C	1			1			$\text{pA}$
$I_{IB}$ Input bias current	Full range	100			100				
	$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		-5 to 2.7		$\text{V}$	
25°C			4.7	4.8		4.7	4.8		
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	Full range	4.7			4.7			$\text{V}$
$V_{OM-}$ Maximum negative peak output voltage swing		25°C	-4.7	-4.9		-4.7	-4.9		
		Full range	±4.7			-4.7			$\text{V}$
$A_{VD}$ Large-signal differential voltage amplification		$V_O = \pm 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	400	560		400	560	
	Full range		300			300			
	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	90	100		90	100		
		Full range	70			70			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}, V_O = 0, R_S = 50\ \Omega$	25°C	90	115		90	115		$\text{dB}$
		Full range	85			85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3\ \text{V to } \pm 8\ \text{V}$	25°C	90	110		90	110		$\text{dB}$
		Full range	85			85			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.1		1.5	1.1		1.5	$\text{mA}$
		Full range	1.5			1.5			

## operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\ \text{V}$

PARAMETER	TEST CONDITIONS	$T_A \uparrow$	TLC2201AC			TLC2201BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7		2	2.7		$\text{V}/\mu\text{s}$
		Full range	1.5			1.5			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18		35	18		30	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C	8		15	8		12	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5			0.5			$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7			0.7			
$I_n$ Equivalent input noise current		25°C	0.6			0.6			$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9			1.9			$\text{MHz}$
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°			48°			

† Full range is 0°C to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	500	$\mu\text{V}$
		Full range			600	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5		pA
		Full range			100	
$I_{IB}$ Input bias current		25°C		1		pA
		Full range			100	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		V	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V	
		Full range	4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0 50	mV	
		Full range		50		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	V/mV	
		Full range	100			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	15			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}, V_O = 0, R_S = 50\ \Omega$	25°C	90	110	dB	
		Full range	85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD} \pm \Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110	dB	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1 1.5	mA	
		Full range		1.5		

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.3			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18	$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1\ \text{kHz}$	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C		0.5	$\mu\text{V}$	
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.8	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		45°		

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y

## Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

### electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AC			TLC2201BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		80	200		80	200	$\mu\text{V}$
		Full range			300			300	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		0.5			0.5	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0,$ $R_S = 50\ \Omega$	25°C		0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C		0.5			0.5		$\text{pA}$
		Full range			100			100	
$I_{IB}$ Input bias current		25°C		1			1		$\text{pA}$
	Full range			100			100		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			0 to 2.7		V	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	V	
		Full range	4.7			4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50		0	50	mV
		Full range			50			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V},$ $R_L = 500\ \text{k}\Omega$	25°C	150	315		150	315	V/mV	
		Full range	100			100			
	$V_O = 1\text{ V to }4\text{ V},$ $R_L = 10\ \text{k}\Omega$	25°C	25	55		25	55		
		Full range	15			15			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin},$ $V_O = 0,$ $R_S = 50\ \Omega$	25°C	90	110		90	110	dB	
		Full range	85			85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110		90	110	dB	
		Full range	85			85			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V},$ No load	25°C		1	1.5		1	1.5	mA
		Full range			1.5			1.5	

### operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AC			TLC2201BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V},$ $R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C	1.8	2.5		1.8	2.5	V/ $\mu\text{s}$	
		Full range	1.3			1.3			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	30	nV/ $\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15		8	12	
$V_N(\text{PP})$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C		0.5			0.5	$\mu\text{V}$	
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6	fA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz},$ $R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C		1.8			1.8	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C		45°			45°		

† Full range is 0°C to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	500	$\mu\text{V}$
		Full range			650	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005	$\mu\text{V}/\text{mo}$
		Full range				
$I_{IO}$ Input offset current		25°C		0.5		$\text{pA}$
$I_{IB}$ Input bias current	Full range			150		
	$V_{ICR}$ Common-mode input voltage range	25°C		1		$\text{pA}$
Full range				150		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V	
$V_{OM-}$ Maximum negative peak output voltage swing		Full range	4.7			
		25°C	-4.7	-4.9	V	
Full range		-4.7				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	V/mV	
		Full range	250			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100		
		Full range	65			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}, V_O = 0, R_S = 50\ \Omega$	25°C	90	115	dB	
		Full range	85			
kSVR Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	dB	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.1 1.5	mA	
		Full range		1.5		

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.4			
$V_n$ Equivalent input noise voltage	f = 10 Hz	25°C		18	$\text{nV}/\sqrt{\text{Hz}}$	
	f = 1 kHz	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	f = 0.1 to 1 Hz	25°C		0.5	$\mu\text{V}$	
	f = 0.1 to 10 Hz	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	f = 10 kHz, $R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°		

† Full range is -40°C to 85°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

## electrical characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AI			TLC2210BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	80 200			80 200			$\mu\text{V}$
		Full range	350			350			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.001	0.005		0.001	0.005		$\mu\text{V}/\text{mo}$
		25°C	0.5			0.5			
$I_{IO}$ Input offset current		25°C	1			1			pA
		Full range	150			150			
$I_{IB}$ Input bias current		25°C	1			1			pA
		Full range	150			150			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8		V
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing		25°C	-4.7	-4.9		-4.7	-4.9		V
		Full range	-4.7			-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560		400	560		V/mV
		Full range	250			250			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100		90	100		
		Full range	65			65			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}, V_O = 0, R_S = 50\ \Omega$	25°C	90	115		90	115		dB
		Full range	85			85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD} \pm / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110		90	110		dB
		Full range	85			85			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.1 1.5		1.1 1.5				mA
		Full range	1.5			1.5			

## operating characteristics at specified free-air temperature, $V_{DD} \pm \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AI			TLC2210BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7		2	2.7		$\text{V}/\mu\text{s}$
		Full range	1.4			1.4			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18 35		18 30				$\text{nV}/\sqrt{\text{Hz}}$
		25°C	8 15		8 12				
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5		0.5				$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7		0.7				
$I_n$ Equivalent input noise current		25°C	0.6		0.6				$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9			1.9			MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°			48°			

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100	500	$\mu\text{V}$	
		Full range	650			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range	0.5		$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5		$\text{pA}$	
		Full range	150			
$I_{IB}$ Input bias current		25°C	1		$\text{pA}$	
		Full range	150			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7	$\text{V}$		
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	$\text{V}$	
		Full range	4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C	0	50	$\text{mV}$	
		Full range	50			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	$\text{V}/\text{mV}$	
		Full range	100			
	$V_O = 1\ \text{V to } 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	15			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	90	110	$\text{dB}$	
		Full range	85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{V to } 16\ \text{V}$	25°C	90	110	$\text{dB}$	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}, \text{ No load}$	25°C	1	1.5	$\text{mA}$	
		Full range	1.5			

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\ \text{V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.2			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C	18		$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1\ \text{kHz}$	25°C	8			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5		$\mu\text{V}$	
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7			
$I_n$ Equivalent input noise current		25°C	0.6		$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8		$\text{MHz}$	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°			

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y

## Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

### electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AI			TLC2201BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	80	200		80	200	$\mu\text{A}$	
		Full range	350			350			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$	
		25°C	0.5			0.5			
$I_{IO}$ Input offset current		25°C	1			1			$\text{pA}$
		Full range	150			150			
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$
		Full range	150			150			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		0 to 2.7		$\text{V}$		
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	$\text{V}$	
		Full range	4.7			4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50		0	50	$\text{mV}$
		Full range	50			50			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		150	315	$\text{V}/\text{mV}$	
		Full range	100			100			
		25°C	25	55		25	55		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	90	110		90	110	$\text{dB}$	
		Full range	85			85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{V to } 16\ \text{V}$	25°C	90	110		90	110	$\text{dB}$	
		Full range	85			85			
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}, \text{ No load}$	25°C		1	1.5		1	1.5	$\text{mA}$
		Full range	1.5			1.5			

### operating characteristics at specified free-air temperature, $V_{DD} = 5\ \text{V}$

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AI			TLC2201BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5		1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.2			1.2			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	30	$\text{nV}/\sqrt{\text{Hz}}$
		25°C		8	15		8	12	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{ to } 1\ \text{Hz}$	25°C		0.5			0.5		$\mu\text{V}$
	$f = 0.1\ \text{ to } 10\ \text{Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6		$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.8			1.8		$\text{MHz}$
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		45°			45°		

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y

## Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A \dagger$	TLC2201M			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	500	$\mu\text{V}$
		Full range			700	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005	$\mu\text{V}/\text{mo}$
		Full range				
$I_{IO}$ Input offset current		25°C		0.5		$\text{pA}$
	Full range			500		
$I_{IB}$ Input bias current	25°C		1		$\text{pA}$	
	Full range			500		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V	
		Full range	4.7			
$V_{OM-}$ Maximum negative peak output voltage swing		25°C	-4.7	-4.9	V	
		Full range	-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	400	560	V/mV	
		Full range	200			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	90	100		
		Full range	45			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	90	115	dB	
		Full range	85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	dB	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.1	mA	
		Full range		1.5		

operating characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A \dagger$	TLC2201M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.3			
$V_n$ Equivalent input noise voltage	f = 10 Hz	25°C		18	$\text{nV}/\sqrt{\text{Hz}}$	
	f = 1 kHz	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	f = 0.1 to 1 Hz	25°C		0.5	$\mu\text{V}$	
	f = 0.1 to 10 Hz	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	f = 10 kHz, $R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9	MHz	
$\phi_m$ Phase margin	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°		

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



# TLC2201, TLC2201A, TLC2201B, TLC2201Y Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2201AM			TLC2210BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage	V <sub>IC</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	80		200	80		200	μV
		Full range				400		400	
α <sub>VIO</sub> Temperature coefficient of input offset voltage		Full range	0.5			0.5			μV/°C
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005		μV/mo
		Full range							
I <sub>IO</sub> Input offset current		25°C	0.5			0.5			pA
		Full range				500			
I <sub>IB</sub> Input bias current		25°C	1			1			pA
		Full range				500			
V <sub>ICR</sub> Common-mode input voltage range		R <sub>S</sub> = 50 Ω	Full range	-5 to 2.7		-5 to 2.7			
V <sub>OM+</sub> Maximum positive peak output voltage swing	R <sub>L</sub> = 10 kΩ	25°C	4.7	4.8		4.7	4.8		V
		Full range	4.7			4.7			
V <sub>OM-</sub> Maximum negative peak output voltage swing		25°C	-4.7	-4.9		-4.7	-4.9		V
		Full range	-4.7			-4.7			
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>O</sub> = ±4 V, R <sub>L</sub> = 500 kΩ	25°C	400	560		400	560		V/mV
		Full range	200			200			
	V <sub>O</sub> = ±4 V, R <sub>L</sub> = 10 kΩ	25°C	90	100		90	100		
		Full range	45			45			
CMRR Common-mode rejection ratio	V <sub>IC</sub> = V <sub>ICRmin</sub> , V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	90	115		90	115		dB
		Full range	85			85			
K <sub>SVR</sub> Supply voltage rejection ratio (ΔV <sub>DD±</sub> /ΔV <sub>IO</sub> )	V <sub>DD±</sub> = ±2.3 V to ±8 V	25°C	90	110		90	110		dB
		Full range	85			85			
I <sub>DD</sub> Supply current	V <sub>O</sub> = 0, No load	25°C	1.1		1.5	1.1		1.5	mA
		Full range				1.5			

† Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observable through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2201AM			TLC2201BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	2	2.7		2	2.7		V/μs
		Full range	1.3			1.3			
V <sub>n</sub> Equivalent input noise voltage (see Note 5)	f = 10 Hz	25°C		18	35		18	30	nV/√Hz
	f = 1 kHz	25°C		8	15		8	12	
V <sub>N(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 0.1 to 1 Hz	25°C		0.5			0.5		μV
	f = 0.1 to 10 Hz	25°C		0.7			0.7		
I <sub>n</sub> Equivalent input noise current		25°C		0.6			0.6		fA/√Hz
Gain-bandwidth product	f = 10 kHz, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		1.9			1.9		MHz
φ <sub>m</sub> Phase margin at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		48°			48°		

† Full range is -55°C to 125°C.

NOTE 5: This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201M			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100	500		$\mu\text{V}$
		Full range	700			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		25°C	0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005*		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			$\text{pA}$
		Full range	500			
$I_{IB}$ Input bias current		25°C	1			$\text{pA}$
		Full range	500			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			V
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		V
		Full range	4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C	0	50		mV
		Full range	50			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		V/mV
		Full range	75			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	10			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	90	110		dB
		Full range	85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110		dB
		Full range	85			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C	1	1.5		mA
		Full range	1.5			

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.5		$\text{V}/\mu\text{s}$
		Full range	1.1			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C	18			$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C	8			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C	0.5			$\mu\text{V}$
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C	0.7			
$I_n$ Equivalent input noise current		25°C	0.6			$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8			MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	45°			

\*On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AM			TLC2210BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80		200	80		200	$\mu\text{V}$
		Full range	400			400			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$
		Full range	500			500			
$I_{IB}$ Input bias current	25°C	1			1			$\text{pA}$	
	Full range	500			500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			0 to 2.7			V
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8		V
		Full range	4.7			4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C	0		50	0		50	V
		Full range	50			50			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		150	315		V/mV
		Full range	75			75			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		25	55		
		Full range	10			10			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, V_O = 0, R_S = 50\ \Omega$	25°C	90	110		90	110		dB
		Full range	85			85			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	90	110		90	110		dB
		Full range	85			85			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C	1.1		1.5	1.1		1.5	mA
		Full range	1.5			1.5			

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observable through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2201AM			TLC2201BM			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V},$ $R_L = 10\text{ k}\Omega,$ $C_L = 100\text{ pF}$	25°C	1.8	2.5		1.8	2.5	$\text{V}/\mu\text{s}$	
			Full range	1.1			1.1			
$V_n$	Equivalent input noise voltage (see Note 5)	$f = 10\text{ Hz}$	25°C		18	35		18	30	$\text{nV}/\sqrt{\text{Hz}}$
			25°C		8	15		8	12	
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$	25°C		0.5			0.5		$\mu\text{V}$
			25°C		0.7			0.7		
$I_n$	Equivalent input noise current		25°C		0.6			0.6	$\text{fA}/\sqrt{\text{Hz}}$	
	Gain-bandwidth product	$f = 10\text{ kHz},$ $R_L = 10\text{ k}\Omega,$ $C_L = 100\text{ pF}$	25°C		1.8			1.8	MHz	
$\phi_m$	Phase margin at unity gain	$R_L = 10\text{ k}\Omega,$ $C_L = 100\text{ pF}$	25°C		45°			45°		

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 5: This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2201Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 50\ \Omega$		100	500	$\mu\text{V}$
Input offset voltage long-term drift (see Note 4)			0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current				0.5	$\text{pA}$
$I_{IB}$ Input bias current				1	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	0 to 2.7			V
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	4.7	4.8		V
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 500\ \Omega$	25	55		V/mV
	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 10\ \Omega$	25	55		
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$ , $R_S = 50\ \Omega$ , $V_O = 0$ ,	90	110		dB
$K_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{to } 16\ \text{V}$	90	110		dB
$I_{DD}$ Supply current per amplifier	$V_O = 2.5\ \text{V}$ , No load		1	1.5	mA

**operating characteristics at  $V_{DD\pm} = \pm 5\ \text{V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2201Y			UNIT
		MIN	TYP	MAX	
SR Positive slew rate at unity gain	$V_O = \pm 0.5\ \text{to } 2.5\ \text{V}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$	1.8	2.5		V/ $\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		18		nV/ $\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$		0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$		0.7		
$I_n$ Equivalent input noise current			0.6		pA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.8		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		48°		

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



PARAMETER MEASUREMENT INFORMATION

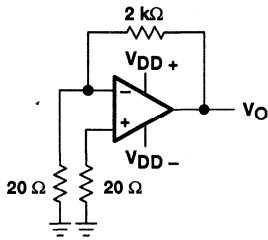
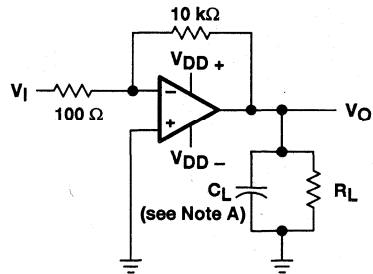
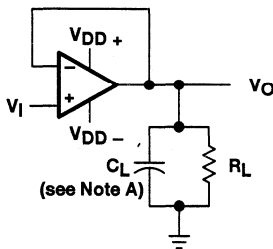


Figure 1. Noise-Voltage Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 2. Phase-Margin Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 3. Slew-Rate Test Circuit

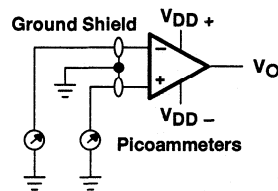


Figure 4. Input-Bias and Offset-Current Test Circuit

typical values

Typical values presented in this data sheet represent the median (50% point) of device parametric performance.

Input bias and offset current

At the picoamp bias current level typical of the TLC2201, TLC2201A, and TLC2201B, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket, and a second test measuring both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

noise

Texas Instruments offers automated production noise testing to meet individual applications requirements. Noise voltage at  $f = 10$  Hz and  $f = 1$  kHz is 100% tested on every TLC2201B device, while lot sample testing is performed on the TLC2201A. For other noise requirements, please contact the factory.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	5
$I_{IB}$	Input bias current	vs Common-mode voltage	6
		vs Free-air temperature	7
CMRR	Common-mode rejection ratio	vs Frequency	8
$V_{OM}$	Maximum peak output voltage	vs Output current	9
		vs Free-air temperature	10
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	11
$V_{OH}$	High-level output voltage	vs Frequency	12
		vs High-level output current	13
		vs Free-air temperature	14
$V_{OL}$	Low-level output voltage	vs Low-level output current	15
		vs Free-air temperature	16
AVD	Large-signal differential voltage amplification	vs Frequency	17
		vs Free-air temperature	18
$I_{OS}$	Short-circuit output current	vs Supply voltage	19
		vs Free-air temperature	20
$I_{DD}$	Supply current	vs Supply voltage	21
		vs Free-air temperature	22
SR	Slew rate	vs Supply voltage	23
		vs Free-air temperature	24
	Pulse response	Small signal	25, 26
		Large signal	27, 28
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	0.1 to 1 Hz	29
		0.1 to 10 Hz	30
	Gain-bandwidth product	vs Supply voltage	31
		vs Free-air temperature	32
$\phi_m$	Phase margin	vs Supply voltage	33
		vs Free-air temperature	34
	Phase shift	vs Frequency	17



TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TLC2201  
 INPUT OFFSET VOLTAGE

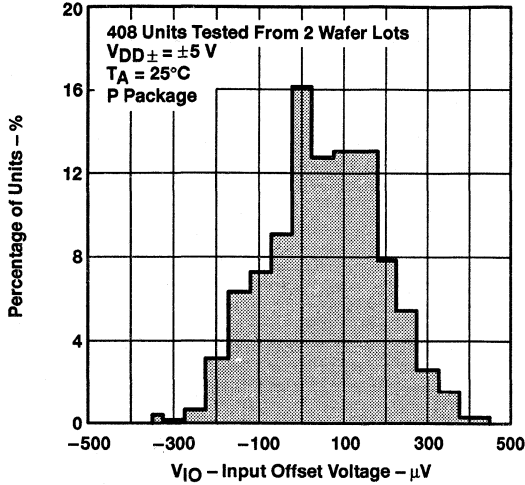


Figure 5

INPUT BIAS CURRENT  
 VS  
 COMMON-MODE INPUT VOLTAGE

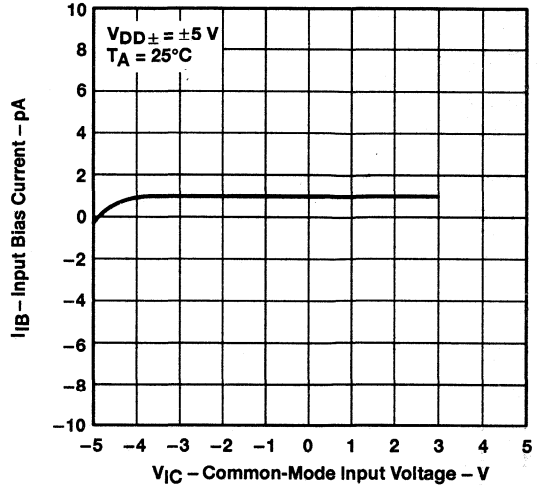


Figure 6

INPUT BIAS CURRENT  
 VS  
 FREE-AIR TEMPERATURE

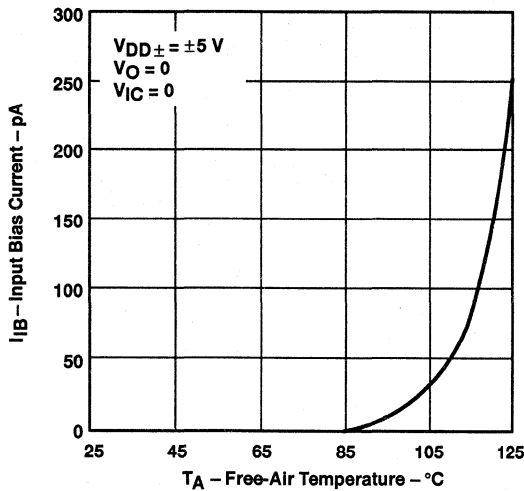


Figure 7

COMMON-MODE REJECTION RATIO  
 VS  
 FREQUENCY

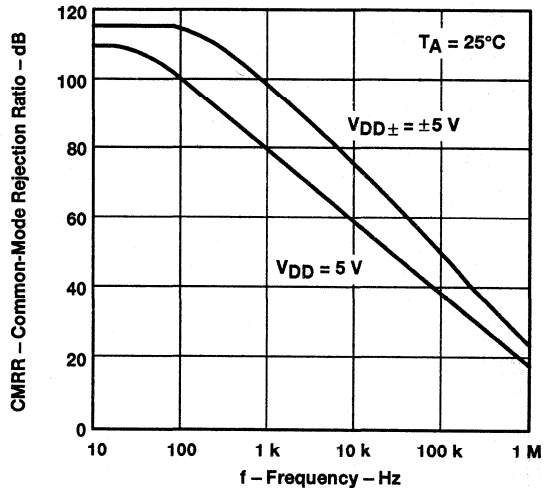


Figure 8

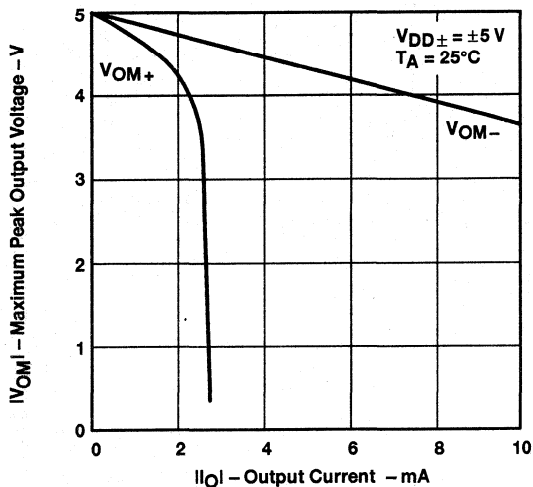
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

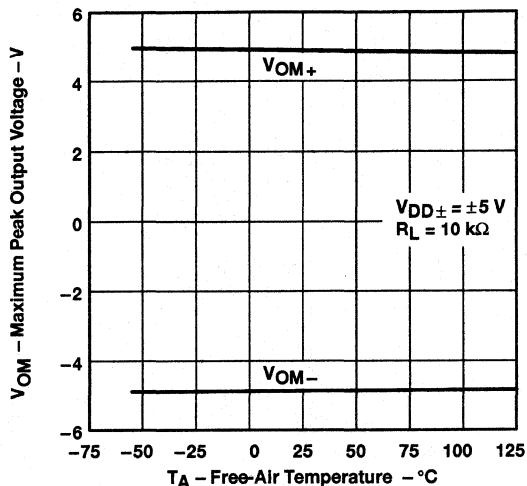
**TYPICAL CHARACTERISTICS†**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**OUTPUT CURRENT**



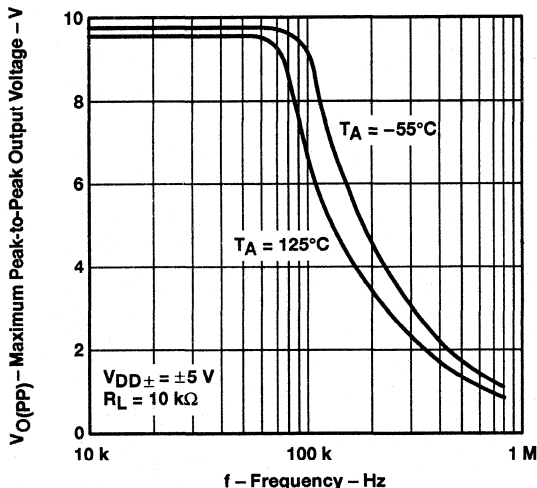
**Figure 9**

**MAXIMUM PEAK OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



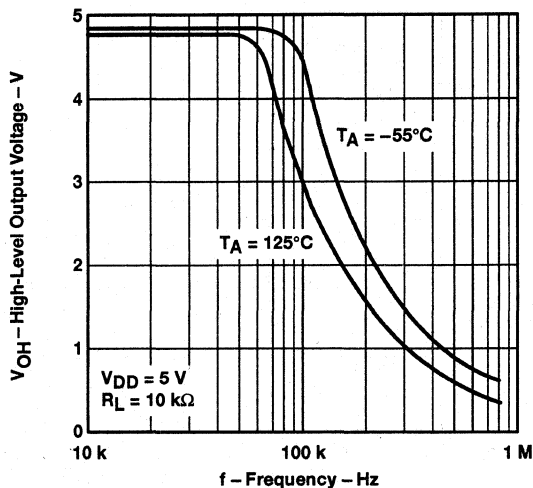
**Figure 10**

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 11**

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 12**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

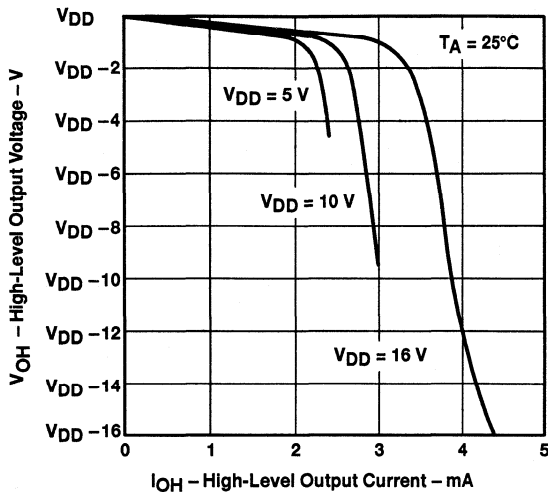


Figure 13

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

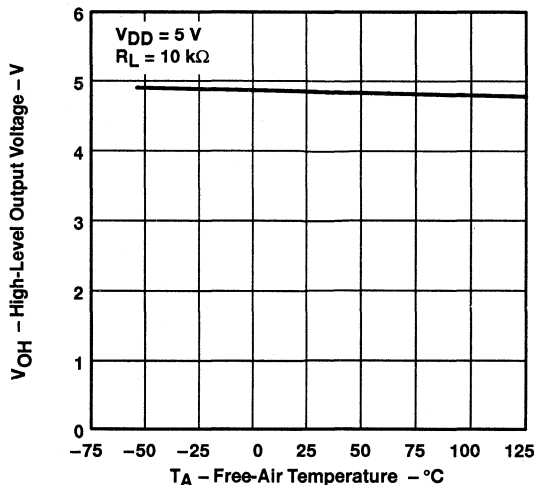


Figure 14

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

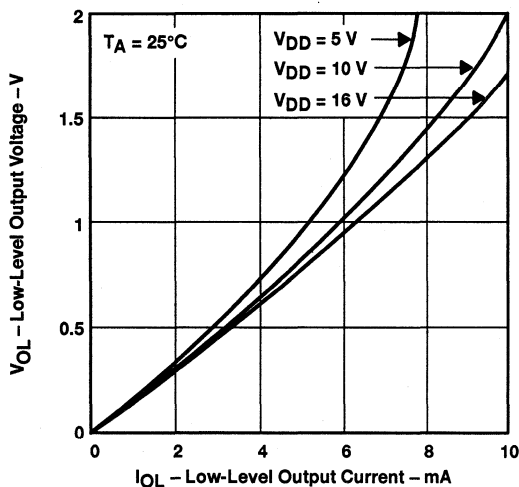


Figure 15

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

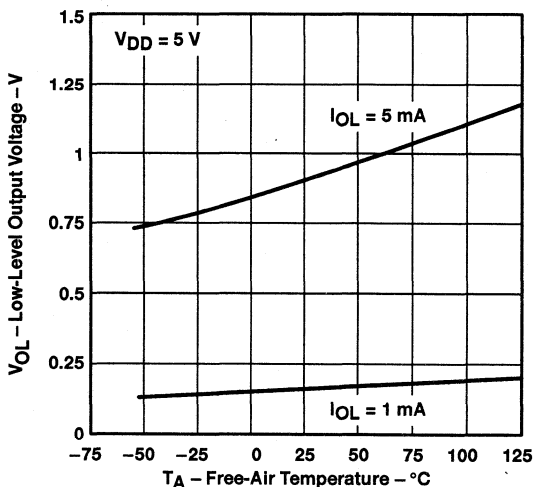


Figure 16

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT vs FREQUENCY**

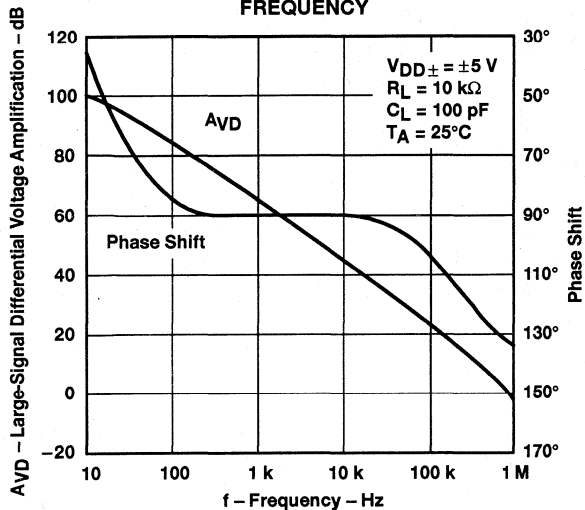


Figure 17

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs FREE-AIR TEMPERATURE**

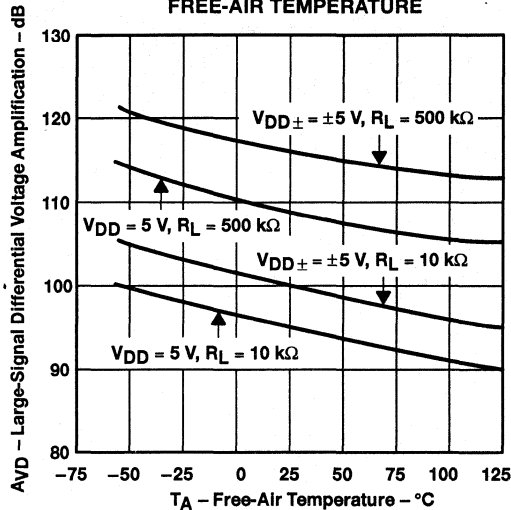


Figure 18

**SHORT-CIRCUIT OUTPUT CURRENT vs SUPPLY VOLTAGE**

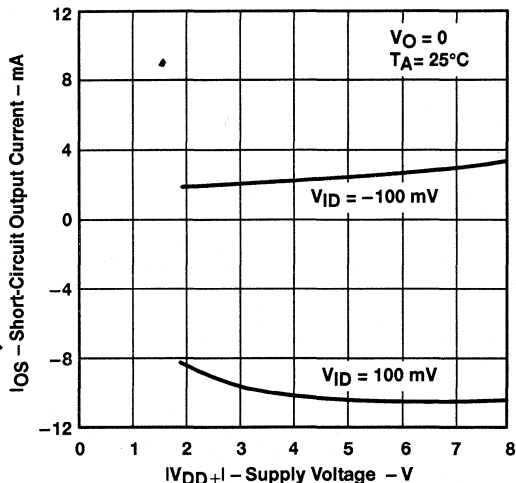


Figure 19

**SHORT-CIRCUIT OUTPUT CURRENT vs FREE-AIR TEMPERATURE**

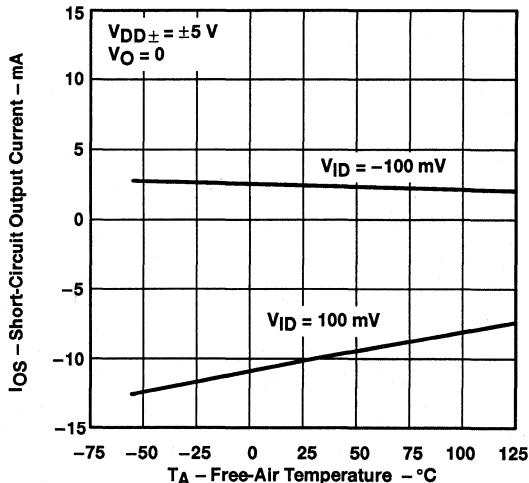


Figure 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





TLC2201, TLC2201A, TLC2201B, TLC2201Y  
 Advanced LinCMOS™ LOW-NOISE PRECISION  
 OPERATIONAL AMPLIFIERS

SLOS021A - NOVEMBER 1988 - REVISED AUGUST 1994

TYPICAL CHARACTERISTICS†

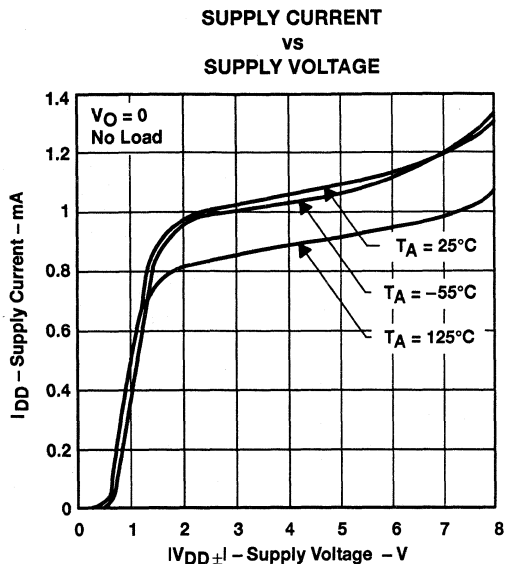


Figure 21

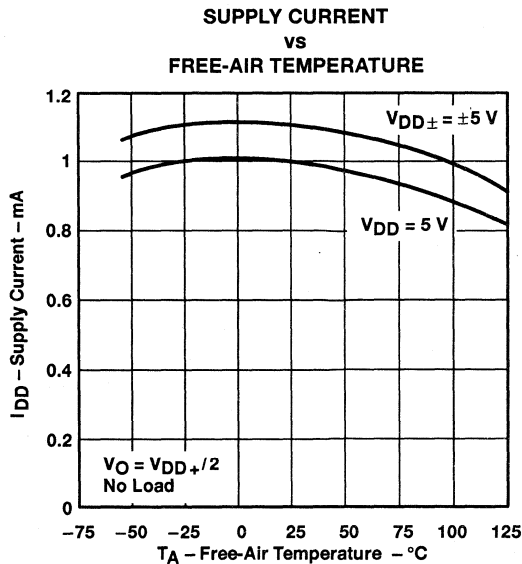


Figure 22

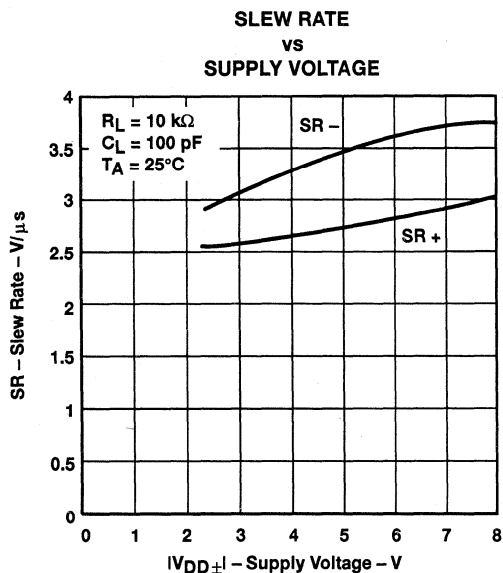


Figure 23

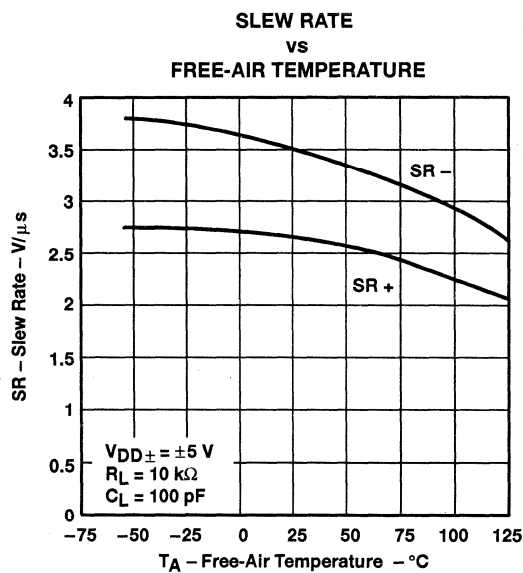


Figure 24

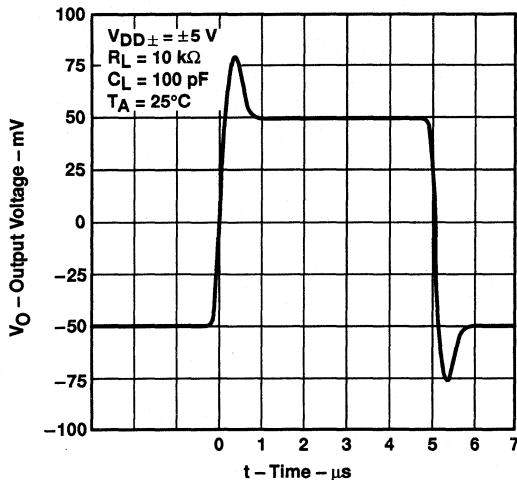
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

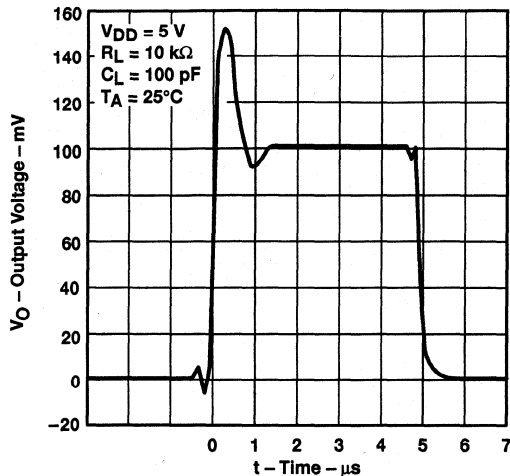
**TYPICAL CHARACTERISTICS**

**VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE**



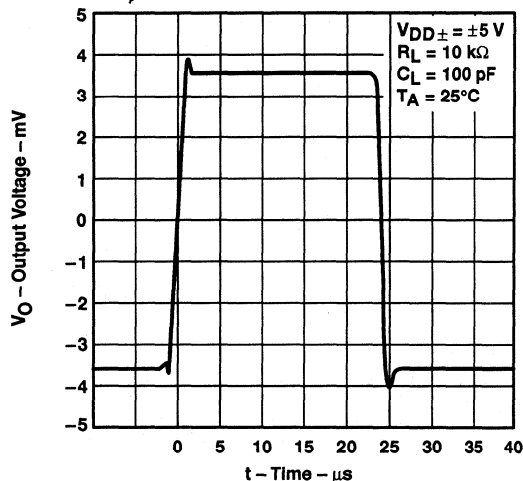
**Figure 25**

**VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE**



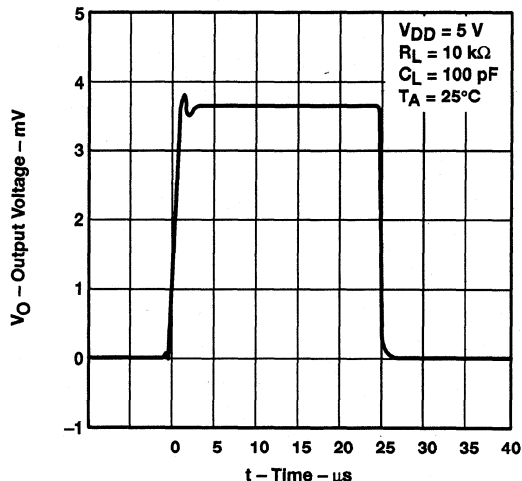
**Figure 26**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**



**Figure 27**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE**



**Figure 28**



TLC2201, TLC2201A, TLC2201B, TLC2201Y  
Advanced LinCMOS™ LOW-NOISE PRECISION  
OPERATIONAL AMPLIFIERS

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

TYPICAL CHARACTERISTICS†

PEAK-TO-PEAK EQUIVALENT  
INPUT NOISE VOLTAGE  
0.1 TO 1 Hz

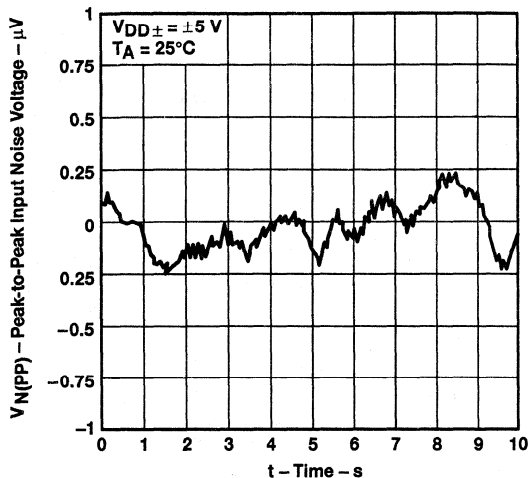


Figure 29

PEAK-TO-PEAK EQUIVALENT  
INPUT NOISE VOLTAGE  
0.1 TO 10 Hz

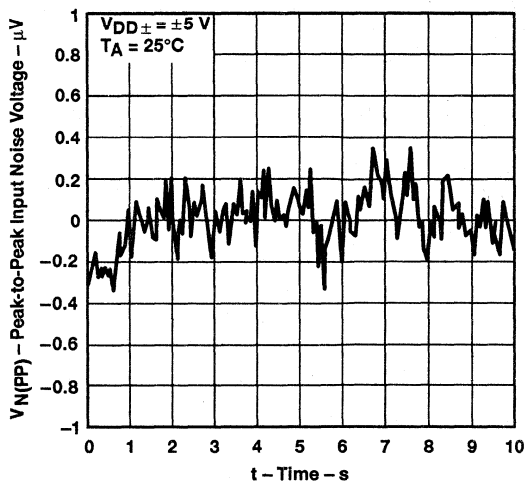


Figure 30

GAIN-BANDWIDTH PRODUCT  
vs  
SUPPLY VOLTAGE

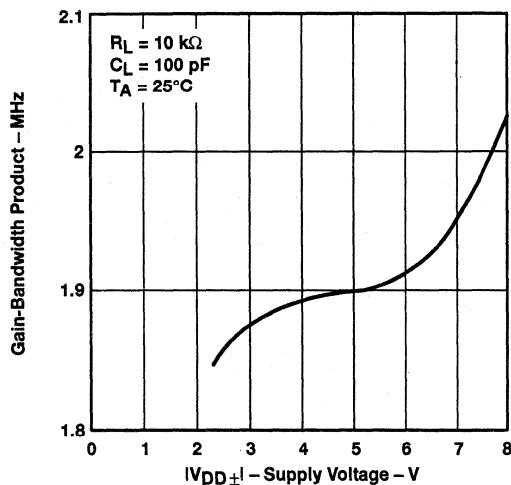


Figure 31

GAIN-BANDWIDTH PRODUCT  
vs  
FREE-AIR TEMPERATURE

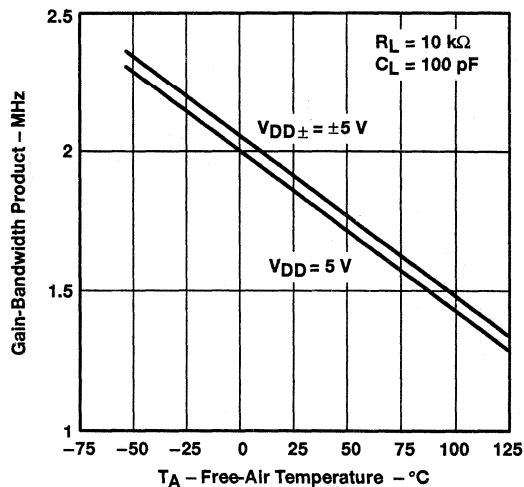


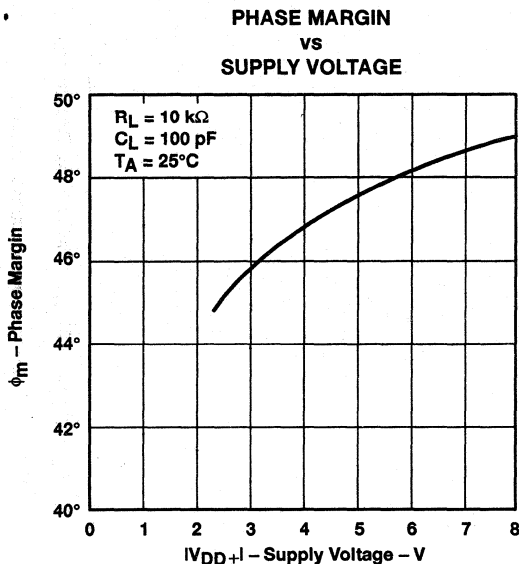
Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

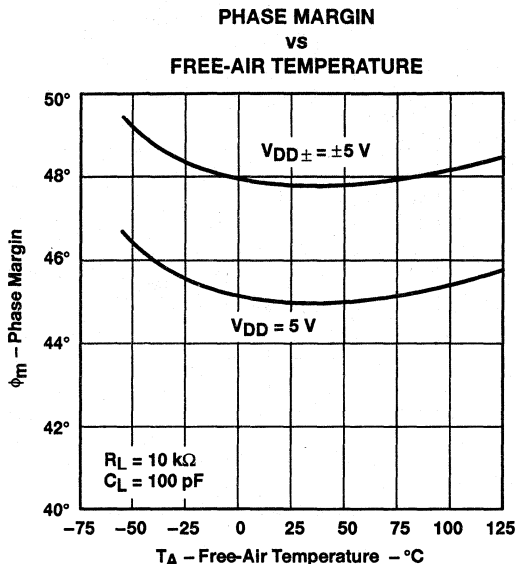
**TLC2201, TLC2201A, TLC2201B, TLC2201Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**

SLOS021A – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**



**Figure 33**



**Figure 34**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**APPLICATION INFORMATION**

**latch-up avoidance**

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC2201, TLC2201A, and TLC2201B inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques reducing the chance of latch-up should be used whenever possible. Internal protection diodes should not be forward biased in normal operation. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μF typical) located across the supply rails as close to the device as possible.

**electrostatic discharge protection**

These devices use internal ESD-protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.



# TLC2202, TLC2202A, TLC2202B, TLC2202Y Advanced LinCMOS™ LOW-NOISE PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS056A – MAY 1990 – REVISED AUGUST 1994

- **TLC2202B is 100% Tested for Noise**  
30 nV/√Hz Max at f = 10 Hz  
12 nV/√Hz Max at f = 1 kHz
- **Low Input Offset Voltage . . . 500 μV Max**
- **Excellent Offset Voltage Stability**  
With Temperature . . . 0.5 μV/°C Typ
- **Rail-to-Rail Output Swing**
- **Low Input Bias Current**  
1 pA Typ at T<sub>A</sub> = 25°C
- **Common-Mode Input Voltage Range**  
Includes the Negative Rail

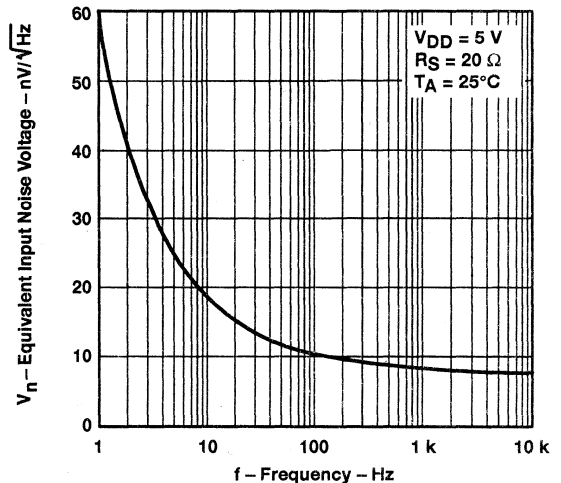
## description

The TLC2202, TLC2202A, TLC2202B, and TLC2202Y are precision, low-noise operational amplifiers using Texas Instruments Advanced LinCMOS™ process. These devices combine the noise performance of the lowest-noise JFET amplifiers with the dc precision available previously only in bipolar amplifiers. The Advanced LinCMOS™ process uses silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. In addition, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The combination of excellent dc and noise performance with a common-mode input voltage range that includes the negative rail makes these devices an ideal choice for high-impedance, low-level signal-conditioning applications in either single-supply or split-supply configurations.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the parametric performance.

TYPICAL EQUIVALENT  
INPUT NOISE VOLTAGE  
vs  
FREQUENCY



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	V <sub>n</sub> max f = 10 Hz AT 25°C	V <sub>n</sub> max f = 1 kHz AT 25°C	PACKAGED DEVICES				CHIP FORM (Y)
				SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	
0°C to 70°C	500 μV 500 μV 1 mV	30 nV/√Hz 35 nV/√Hz —	12 nV/√Hz 15 nV/√Hz —	TLC2202BCD TLC2202ACD TLC2202CD	— — —	— — —	TLC2202BCP TLC2202ACP TLC2202CP	TLC2202Y
–40°C to 85°C	500 μV 500 μV 1 mV	30 nV/√Hz 35 nV/√Hz —	12 nV/√Hz 15 nV/√Hz —	TLC2202BID TLC2202AID TLC2202ID	— — —	— — —	TLC2202BIP TLC2202AIP TLC2202IP	—
–55°C to 125°C	500 μV 500 μV 1 mV	30 nV/√Hz 35 nV/√Hz —	12 nV/√Hz 15 nV/√Hz —	TLC2202BMD TLC2202AMD TLC2202MD	TLC2202BMFK TLC2202AMFK TLC2202MFK	TLC2202BMJG TLC2202AMJG TLC2202MJG	TLC2202BMP TLC2202AMP TLC2202MP	—

The D packages are available taped and reeled. Add R suffix to device type (e.g. TLC2202BCDR). Chips are tested at 25°C.

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated  
On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

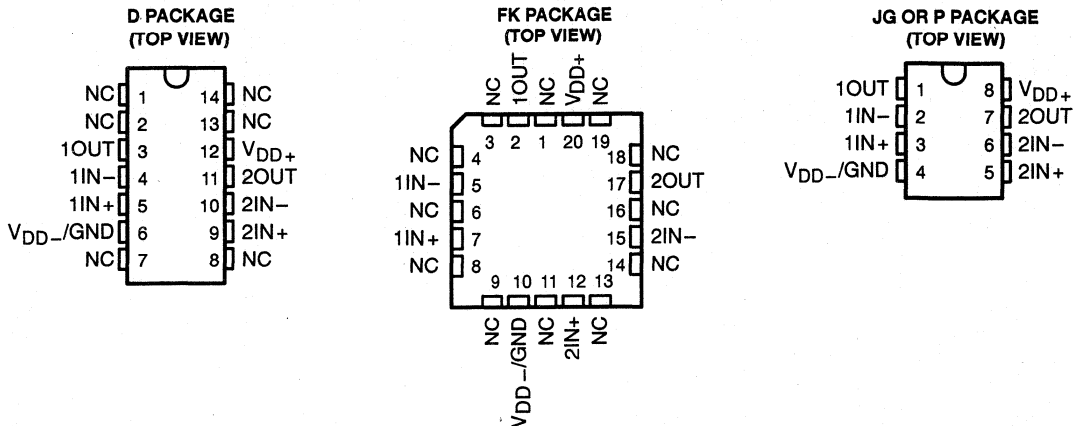
# TLC2202, TLC2202A, TLC2202B, TLC2202Y

## Advanced LinCMOS™ LOW-NOISE PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS056A – MAY 1990 – REVISED AUGUST 1994

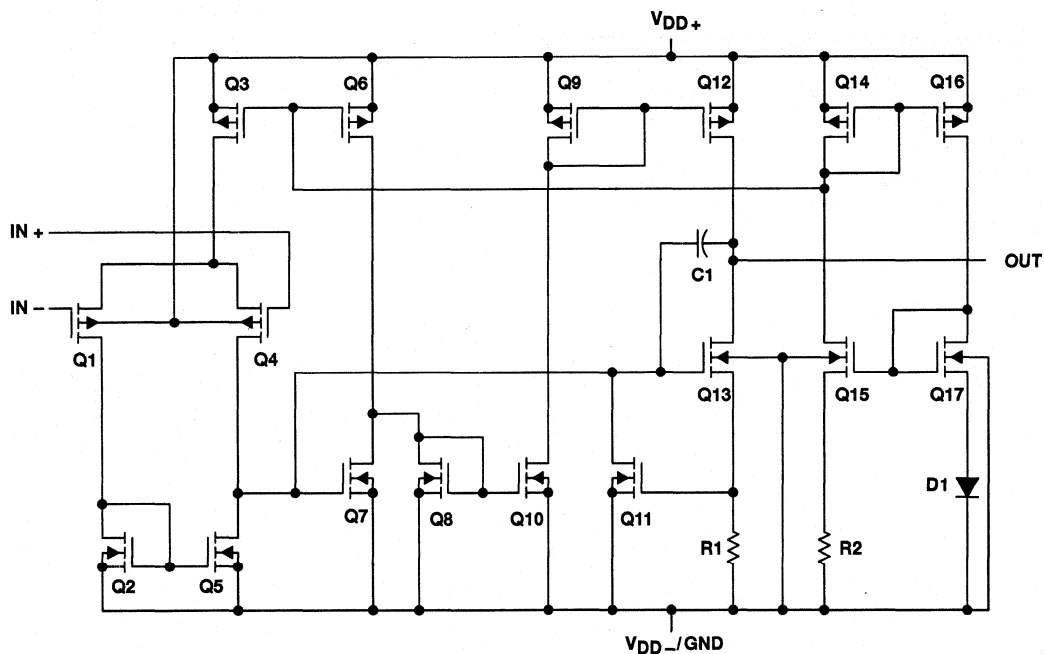
### description (continued)

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.



NC – No internal connection

### equivalent schematic (each amplifier)

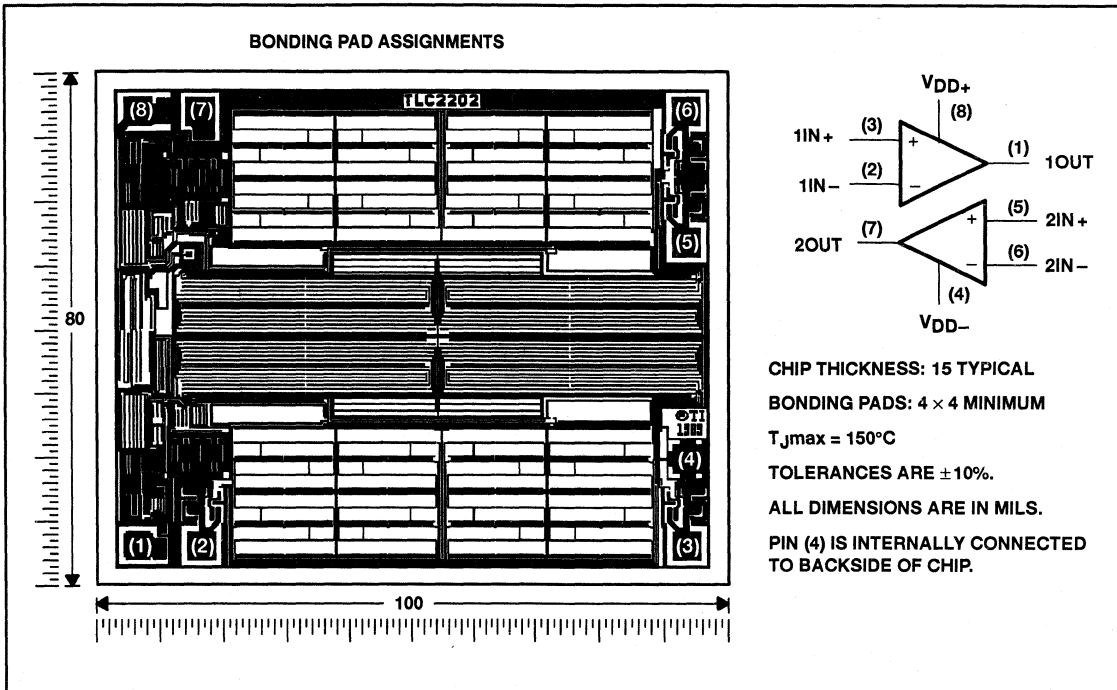


TLC2202, TLC2202A, TLC2202B, TLC2202Y  
 Advanced LinCMOS™ LOW-NOISE PRECISION  
 DUAL OPERATIONAL AMPLIFIERS

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**TLC2202Y chip formation**

This chip, when properly assembled, displays characteristics similar to the TLC2202C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	±16 V
Input voltage, $V_I$ (any input)	±8 V
Input current, $I_I$ (each input)	±5 mA
Output current, $I_O$ (each output)	±50 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values except differential voltages are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

**recommended operating conditions**

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	±2.3	±8	±2.3	±8	±2.3	±8	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C





**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise specified)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100	1000		$\mu\text{V}$
		Full range	1150			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range	0.5		$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			pA
		Full range	100			
$I_{IB}$ Input bias current		25°C	1			
		Full range	100			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		V
		Full range	4.7			
$V_{OM-}$ Maximum negative peak output voltage swing		25°C	-4.7	-4.9		
		Full range	-4.7			V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	300	560		V/mV
		Full range	200			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	50	100		
		Full range	25			
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50\ \Omega, V_{IC} = V_{ICR\text{min}}$	25°C	80	115		dB
		Full range	80			
KSVR Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	80	110		dB
		Full range	80			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.8	2.7		mA
		Full range	2.7			

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.7		$\text{V}/\mu\text{s}$
		Full range	1.3			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C	18			$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C	8			
$V_{N(\text{PP})}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C	0.5			$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C	0.7			
$I_n$ Equivalent input noise current		25°C	0.6			$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9			MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°			

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AC			TLC2202BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	80 500			80 500			$\mu\text{V}$
		Full range	650			650			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			0.5			$\text{pA}$
		Full range	100			100			
$I_{IB}$ Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1			1			$\text{pA}$
		Full range	100			100			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			-5 to 2.7		$\text{V}$	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	$\text{V}$	
Full range		4.7		4.7					
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	-4.7	-4.9		-4.7	-4.9	$\text{V}$	
		Full range	-4.7		-4.7				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	300	560		300	560	$\text{V}/\text{mV}$	
		Full range	200		200				
	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	50	100		50	100		
		Full range	25		25				
$\text{CMRR}$ Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\text{min}}, R_S = 50\ \Omega$	25°C	80	115		80	115	$\text{dB}$	
		Full range	80		80				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3\ \text{V}$ to $\pm 8\ \text{V}$	25°C	80	110		80	110	$\text{dB}$	
		Full range	80		80				
$I_{DD}$ Supply current	$V_O = 0, \text{No load}$	25°C	1.8 2.7		1.8 2.7		$\text{mA}$		
		Full range	2.7		2.7				

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\ \text{V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AC			TLC2202BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$\text{SR}$ Slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.7		1.8	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.3		1.3				
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C	18 35		18 30		$\text{nV}/\sqrt{\text{Hz}}$		
	$f = 1\ \text{kHz}$	25°C	8 15		8 12				
$V_{N(\text{PP})}$ Peak-to-peak equivalent input noise voltage	$f = 0.1$ to $1\ \text{Hz}$	25°C	0.5		0.5		$\mu\text{V}$		
	$f = 0.1$ to $10\ \text{Hz}$	25°C	0.7		0.7				
$I_n$ Equivalent input noise current		25°C	0.6		0.6		$\text{fA}/\sqrt{\text{Hz}}$		
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.9		1.9		$\text{MHz}$		
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	48°		48°				

† Full range is 0°C to 70°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.  
5. This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	100		1000	$\mu\text{V}$
		Full range			1150	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	Full range			100	$\text{pA}$
$I_{IB}$ Input bias current		25°C			1	
		Full range			100	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		V	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		V
		Full range	4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C	0		50	mV
		Full range			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		V/mV
		Full range	100			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	15			
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50\ \Omega, V_{IC} = V_{ICRmin}$	25°C	75	110		dB
		Full range	75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	80	110		dB
		Full range	80			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.7		2.6	mA
		Full range			2.6	

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.6	2.5		V/ $\mu\text{s}$
		Full range	1.1			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C			18	nV/ $\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C			8	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C			0.5	$\mu\text{V}$
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C			0.7	
$I_n$ Equivalent input noise current		25°C			0.6	fA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C			1.9	MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C			47°	

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AC			TLC2202BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		80	500		80	500	$\mu\text{V}$
		Full range			650			650	
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	Full range		0.5			0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		0.5			0.5		$\text{pA}$
		Full range			100			100	
$I_{IB}$ Input bias current		25°C		1			1		$\text{pA}$
		Full range			100			100	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			0 to 2.7		$\text{V}$	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	$\text{V}$	
		Full range	4.7			4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50		0	50	$\text{mV}$
		Full range			50			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		150	315	$\text{V}/\text{mV}$	
		Full range	100			100			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		25	55		
		Full range	15			15			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\text{min}}, R_S = 50\ \Omega$	25°C	75	110		75	110	$\text{dB}$	
		Full range	75			75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} \pm \Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	80	110		80	110	$\text{dB}$	
		Full range	80			80			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1.7	2.6		1.7	2.6	$\text{mA}$
		Full range			2.6			2.6	

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AC			TLC2202BC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.6	2.5		1.6	2.5	$\text{V}/\mu\text{s}$	
		Full range	1.1			1.1			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	30	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15		8	12	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C		0.5			0.5		$\mu\text{V}$
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9			1.9	$\text{MHz}$	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		47°			47°		

† Full range is 0°C to 70°C.

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.  
5. This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	1000	$\mu\text{V}$
		Full range			1200	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	Full range		0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	Full range			150	$\text{pA}$
$I_{IB}$ Input bias current		25°C			1	
		Full range			150	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		V
$V_{OM-}$ Maximum negative peak output voltage swing		Full range	4.7			
		25°C	-4.7	-4.9		V
Full range		-4.7				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	300	560		V/mV
		Full range	150			
	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	50	100		
		Full range	25			
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50\ \Omega, V_{IC} = V_{ICR\text{min}}$	25°C	80	115		dB
		Full range	80			
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = \pm 2.3\ \text{V to } \pm 8\ \text{V}$	25°C	80	110		dB
		Full range	80			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.8	2.7	mA
		Full range			2.7	

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\ \text{V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.7		V/ $\mu\text{s}$
		Full range	1.2			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18		nV/ $\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6		fA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°		

† Full range is -40°C to 85°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AI			TLC2202BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	80	500		80	500	$\mu\text{V}$	
		Full range			700		700		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			0.5			$\text{pA}$
		Full range			150			150	
$I_{IB}$ Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1			1			$\text{pA}$
Full range				150			150		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7			-5 to 2.7		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	V	
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	-4.7	-4.9		-4.7	-4.9	V	
		Full range	-4.7			-4.7			
$AVD$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	300	560		300	560	V/mV	
		Full range	150			150			
	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	50	100		50	100		
		Full range	25			25			
$CMRR$ Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50\ \Omega$	25°C	80	115		80	115	dB	
		Full range	80			80			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} \pm 2.3\ \text{V to } \pm 8\ \text{V}$	25°C	80	110		80	110	dB	
		Full range	80			80			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.8	2.7		1.8	2.7	mA
		Full range			2.7			2.7	

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\ \text{V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AI			TLC2202BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.8	2.7		1.8	2.7	$\text{V}/\mu\text{s}$	
		Full range	1.2			1.2			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	30	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15		8	12	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5			0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6		$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9			1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°			48°		

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202I			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	1000	$\mu\text{V}$
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range			1200	
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	Full range			150	$\text{pA}$
$I_{IB}$ Input bias current		25°C		1		
		Full range			150	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			V
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		V
		Full range		4.7		
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50	mV
		Full range			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		V/mV
		Full range		100		
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range		15		
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50\ \Omega, V_{IC} = V_{ICR\text{min}}$	25°C	75	110		dB
		Full range		75		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	80	110		dB
		Full range		80		
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C		1.7	2.6	mA
		Full range			2.6	

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.6	2.5		$\text{V}/\mu\text{s}$
		Full range		1		
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C		0.5		$\mu\text{V}$
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6		$\text{fA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		47°		

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AI			TLC2202BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		80	500		80	500	$\mu\text{V}$
		Full range			700			700	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	Full range		0.5			0.5	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005		0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		0.5			0.5	$\text{pA}$	
		Full range			150		150		
$I_{IB}$ Input bias current		25°C		1			1	$\text{pA}$	
		Full range			150		150		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			0 to 2.7		V	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C		4.7	4.8		4.7	4.8	V
		Full range		4.7			4.7		
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50		0	50	mV
		Full range			50			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C		150	315		150	315	V/mV
		Full range		100			100		
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C		25	55		25	55	
		Full range		15			15		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50\ \Omega$	25°C		75	110		75	110	dB
		Full range		75			75		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} \pm / \Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C		80	110		80	110	dB
		Full range		80			80		
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \rightarrow$ No load	25°C		1.7	2.6		1.7	2.6	mA
		Full range			2.6			2.6	

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AI			TLC2202BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.6	2.5		1.6	2.5	V/ $\mu\text{s}$
		Full range		1			1		
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		18	35		18	30	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15		8	12	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\ \text{Hz}$	25°C		0.5			0.5	$\mu\text{V}$	
	$f = 0.1\text{ to }10\ \text{Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9			1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		47°			47°		

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.





**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202M			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50 \Omega$	25°C		100	1000	$\mu V$
		Full range			1250	
$\alpha_{V_{IO}}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu V/^\circ C$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005*		$\mu V/mo$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50 \Omega$	Full range			500	$\mu A$
$I_{IB}$ Input bias current		25°C		1		
		Full range			500	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	Full range	-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 k\Omega$	25°C	4.7	4.8		V
		Full range	4.7			
$V_{OM-}$ Maximum negative peak output voltage swing		25°C	-4.7	-4.9		V
	Full range	-4.7				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1$ V to 4 V, $R_L = 500 k\Omega$	25°C	300	560	V/mV	
		Full range	100			
	$V_O = 1$ V to 4 V, $R_L = 10 k\Omega$	25°C	50	100		
		Full range	25			
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50 \Omega, V_{IC} = V_{ICRmin}$	25°C	80	115	dB	
		Full range	80			
KSVR Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = \pm 2.3$ V to $\pm 8$ V	25°C	80	110	dB	
		Full range	80			
$I_{DD}$ Supply current	$V_O = 0, \text{No load}$	25°C	1.8	2.7	mA	
		Full range		2.7		

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3$ V, $R_L = 10 k\Omega, C_L = 100$ pF	25°C	1.8	2.7	$V/\mu s$	
		Full range	1.1			
$V_n$ Equivalent input noise voltage	f = 10 Hz	25°C		18	$nV/\sqrt{Hz}$	
	f = 1 kHz	25°C		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	f = 0.1 to 1 Hz	25°C		0.5	$\mu V$	
	f = 0.1 to 10 Hz	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$fA/\sqrt{Hz}$	
Gain-bandwidth product	f = 10 kHz, $R_L = 10 k\Omega, C_L = 100$ pF	25°C		1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10 k\Omega, C_L = 100$ pF	25°C		48°		

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ C$  extrapolated to  $T_A = 25^\circ C$  using Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AM			TLC2202BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		80	500		80	500	$\mu$ V
		Full range			750			750	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50 \Omega$	Full range		0.5			0.5		$\mu$ V/°C
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005*		0.001	0.005*	$\mu$ V/mo
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50 \Omega$	25°C		0.5			0.5		pA
		Full range			500			500	
$I_{IB}$ Input bias current		25°C		1			1		pA
		Full range			500			500	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	Full range	-5 to 2.7			-5 to 2.7		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 \text{ k}\Omega$	25°C		4.7	4.8		4.7	4.8	V
		Full range		4.7			4.7		
$V_{OM-}$ Maximum negative peak output voltage swing		25°C		-4.7	-4.9		-4.7	-4.9	V
		Full range		-4.7			-4.7		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4 \text{ V}, R_L = 500 \text{ k}\Omega$	25°C		300	560		300	560	V/mV
		Full range		100			100		
	$V_O = \pm 4 \text{ V}, R_L = 10 \text{ k}\Omega$	25°C		50	100		50	100	
		Full range		25			25		
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50 \Omega$	25°C		80	115		80	115	dB
		Full range		80			80		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3 \text{ V to } \pm 8 \text{ V}$	25°C		80	110		80	110	dB
		Full range		80			80		
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.8	2.7		1.8	2.7	mA
		Full range			2.7			2.7	

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AM			TLC2202BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	1.8	2.7		1.8	2.7		V/ $\mu\text{s}$
		Full range	1.1			1.1			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\text{ Hz}$	25°C		18	35*		18	30*	nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	25°C		8	15*		8	12*	
$V_N(\text{PP})$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$	25°C		0.5			0.5		$\mu\text{V}$
	$f = 0.1\text{ to }10\text{ Hz}$	25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6		fA/ $\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		1.9			1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		48°			48°		

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 5: This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperatures,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202M			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	1000	$\mu\text{V}$
		Full range			1250	
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		Full range		0.5		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.001	0.005*	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	Full range			500	$\text{pA}$
$I_{IB}$ Input bias current		25°C		1		
		Full range			500	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7		$\text{V}$	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	$\text{V}$	
		Full range	4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C		0	50	$\text{mV}$
		Full range			50	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}, R_L = 500\ \text{k}\Omega$	25°C	150	315	$\text{V}/\text{mV}$	
		Full range	75			
	$V_O = 1\ \text{V to } 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	10			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}, R_S = 50\ \Omega$	25°C	75	110	$\text{dB}$	
		Full range	75			
KSVR Supply-voltage rejection ratio ( $\Delta V_{DD} \pm / \Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{V to } 16\ \text{V}$	25°C	80	110	$\text{dB}$	
		Full range	80			
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}, \text{ No load}$	25°C		1.7	2.6	$\text{mA}$
		Full range			2.6	

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\ \text{V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	1.6	2.5	$\text{V}/\mu\text{s}$	
		Full range	0.9			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18	$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1\ \text{kHz}$	25°C		8		
$V_{N(\text{PP})}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5	$\mu\text{V}$	
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9	$\text{MHz}$	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		47°		

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AM			TLC2202BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	80	500		80	500	$\mu\text{V}$	
		Full range		750		750			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	Full range	0.5			0.5			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005*		0.001	0.005*	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.5			0.5			pA
		Full range		500		500			
$I_{IB}$ Input bias current	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	1			1			pA
		Full range		500		500			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	0 to 2.7			0 to 2.7		V	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8		4.7	4.8	V	
		Full range	4.7			4.7			
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$	25°C	0 50			0 50			mV
		Full range	50			50			
AVD Large-signal differential voltage amplification	$V_O = 1\text{ V to }4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	150	315		150	315	V/mV	
		Full range	75			75			
	$V_O = 1\text{ V to }4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	25	55		25	55		
		Full range	10			10			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50\ \Omega$	25°C	75	110		75	110	dB	
		Full range	75			75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD} = 4.6\text{ V to }16\text{ V}$	25°C	80	110		80	110	dB	
		Full range	80			80			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C	1.7 2.6			1.7 2.6			mA
		Full range	2.6			2.6			

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2202AM			TLC2202BM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	1.6	2.5		1.6	2.5		V/ $\mu$ s
		Full range	0.9			1.1			
$V_n$ Equivalent input noise voltage (see Note 5)	$f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		18	35*		18	30*	nV/ $\sqrt{\text{Hz}}$
		25°C		8	15*		8	12*	
$V_N(PP)$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$ $f = 0.1\text{ to }10\text{ Hz}$	25°C		0.5			0.5		$\mu$ V
		25°C		0.7			0.7		
$I_n$ Equivalent input noise current		25°C		0.6			0.6	fA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		1.9			1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		47°			47°		

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$

NOTE 5: This parameter is tested on a sample basis for the TLC2202A and on all devices for the TLC2202B. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**electrical characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2202Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 50\ \Omega$		100	1000	$\mu\text{V}$
Input offset voltage long-term drift (see Note 4)			0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current				0.5	$\text{pA}$
$I_{IB}$ Input bias current				1	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$		0 to 2.7		$\text{V}$
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$		4.7	4.8	$\text{V}$
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$			0 50	$\text{mV}$
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 500\ \Omega$		150	315	$\text{V}/\text{mV}$
	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 10\ \Omega$		25	55	
CMRR Common-mode rejection ratio	$V_O = 0$ , $V_{ICR\text{min}}$ , $R_S = 50\ \Omega$		75	110	$\text{dB}$
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DCC}/\Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{ to } 16\ \text{V}$		80	110	$\text{dB}$
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}$ , No load			1.7 2.6	$\text{mA}$

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

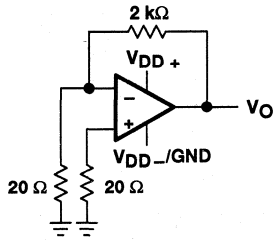
**operating characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2202Y			UNIT
		MIN	TYP	MAX	
SR Positive slew rate at unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$	1.6	2.5		$\text{V}/\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		18		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 10\ \text{kHz}$		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{ to } 1\ \text{Hz}$		0.5		$\mu\text{V}$
	$f = 0.1\ \text{ to } 10\ \text{Hz}$		0.7		
$I_n$ Equivalent input noise current			0.6		$\text{pA}/\sqrt{\text{Hz}}$
$B_1$ Gain-bandwidth product	$f = 10\ \text{kHz}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.9		$\text{MHz}$
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		47°		

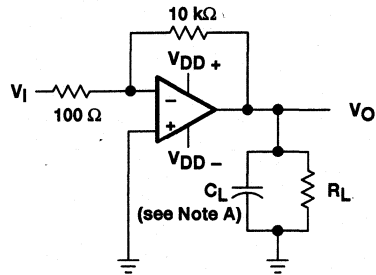
**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**PARAMETER MEASUREMENT INFORMATION**

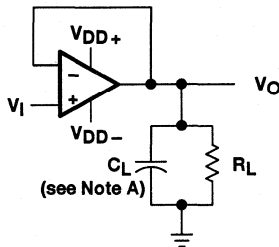


**Figure 1. Noise-Voltage Test Circuit**



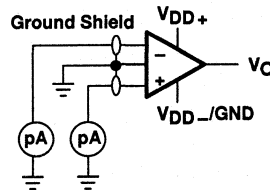
NOTE A:  $C_L$  includes fixture capacitance.

**Figure 2. Phase-Margin Test Circuit**



NOTE A:  $C_L$  includes fixture capacitance.

**Figure 3. Slew-Rate Test Circuit**



**Figure 4. Input-Bias and Offset-Current Test Circuit**

**typical values**

Typical values presented in this data sheet represent the median (50% point) of device parametric performance.

**input bias and offset current**

At the picoamp bias current level of the TLC2202, TLC2202A, and TLC2202B, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied but with no device in the socket. The device is then inserted in the socket, and a second test measuring both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

**noise**

Texas Instruments offers automated production noise testing to meet individual application requirements. Noise voltage at  $f = 10 \text{ Hz}$  and  $f = 1 \text{ kHz}$  is 100% tested on every TLC2201B device, while lot sample testing is performed on the TLC2202A. For other noise requirements, please contact the factory.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

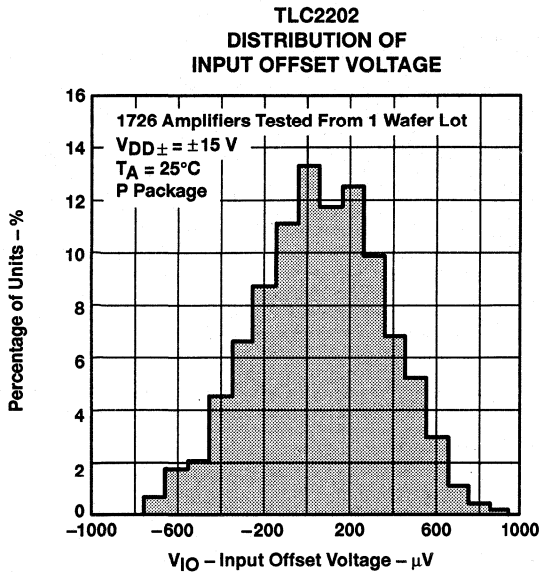
**Table of Graphs**

			FIGURE
$V_{IO}$	Input offset voltage	Distribution	5
$I_{IB}$	Input bias current	vs Common-mode input voltage	6
		vs Free-air temperature	7
$V_{OM}$	Maximum peak output voltage	vs Output current	8,9
		vs Free-air temperature	10
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	11
$V_{OH}$	High-level output voltage	vs Frequency	12
		vs Current	13
		vs Free-air temperature	14
$V_{OL}$	Low-level output voltage	vs Output current	15
		vs Free-air temperature	16
$A_{VD}$	Large-signal differential voltage amplification	vs Frequency	17
		vs Free-air temperature	18
$I_{OS}$	Short-circuit output current	vs Supply voltage	19
		vs Free-air temperature	20
$CMRR$	Common-mode rejection ratio	vs Frequency	21, 22
$I_{DD}$	Supply current	vs Supply voltage	23
		vs Free-air temperature	24
	Pulse response	Small signal	25, 26
		Large signal	27, 28
$SR$	Slew rate	vs Supply voltage	29
		vs Free-air temperature	30
	Noise voltage (referred to input)	0.1 to 1 Hz	31
		0.1 to 10 Hz	32
	Gain-bandwidth product	vs Supply voltage	33
		vs Free-air temperature	34
$\phi_m$	Phase margin	vs Supply voltage	35
		vs Free-air temperature	36
	Phase shift	vs Frequency	17

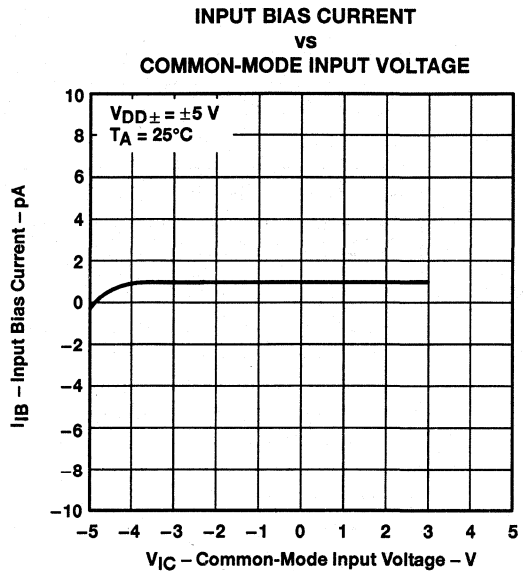
**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

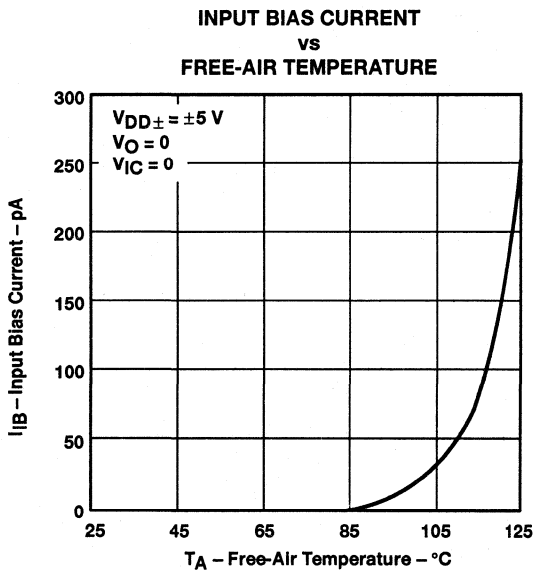
**TYPICAL CHARACTERISTICS†**



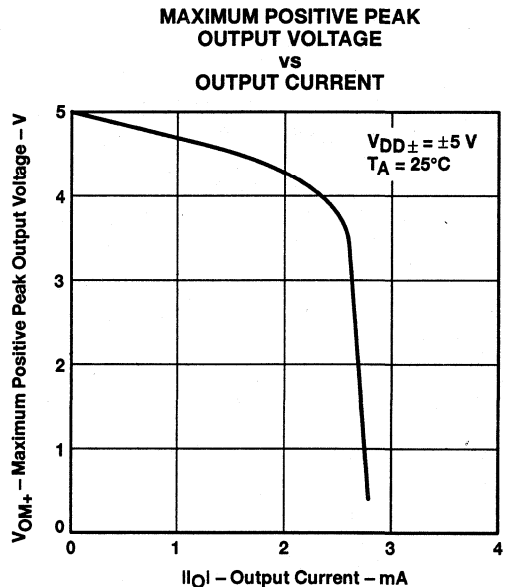
**Figure 5**



**Figure 6**



**Figure 7**



**Figure 8**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM NEGATIVE PEAK  
 OUTPUT VOLTAGE  
 VS  
 OUTPUT CURRENT

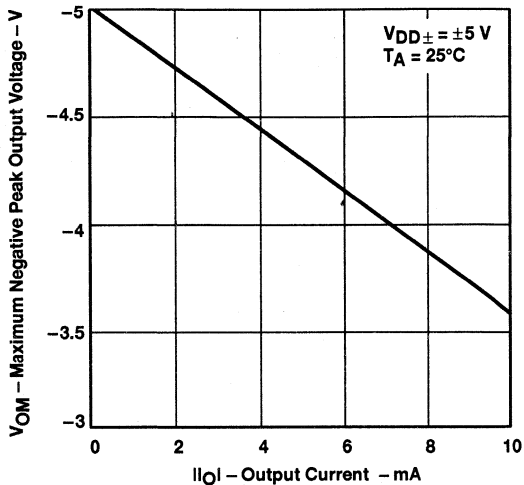


Figure 9

MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 FREE-AIR TEMPERATURE

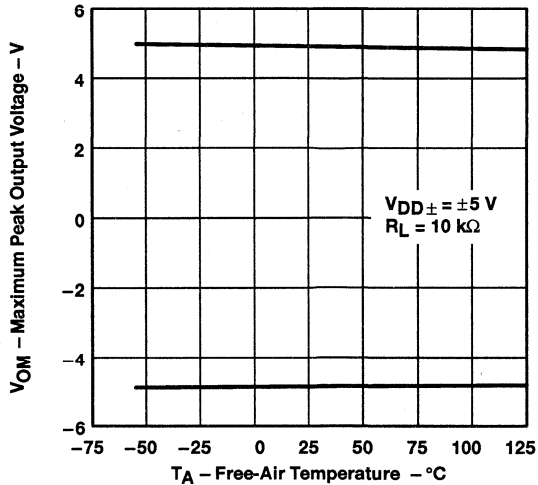


Figure 10

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
 VS  
 FREQUENCY

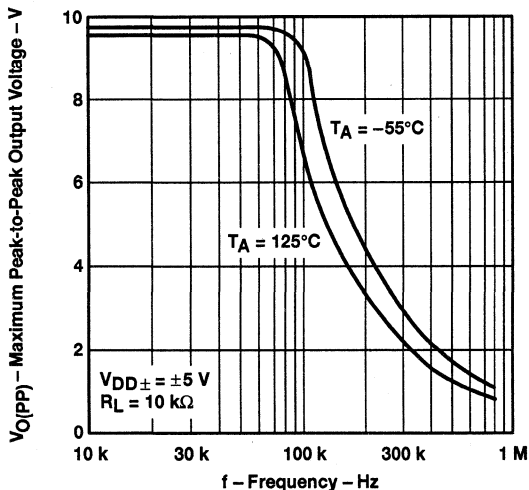


Figure 11

HIGH-LEVEL OUTPUT VOLTAGE  
 VS  
 FREQUENCY

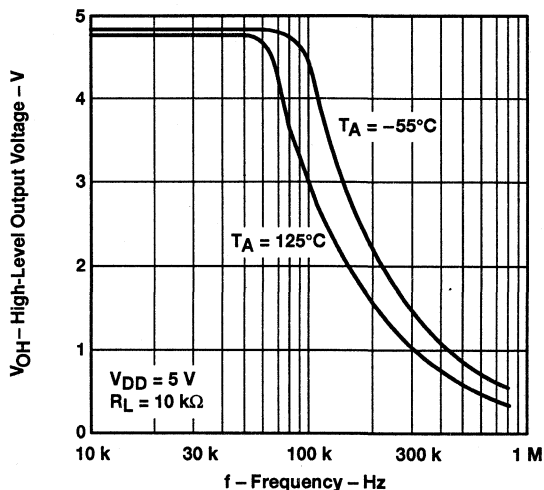


Figure 12

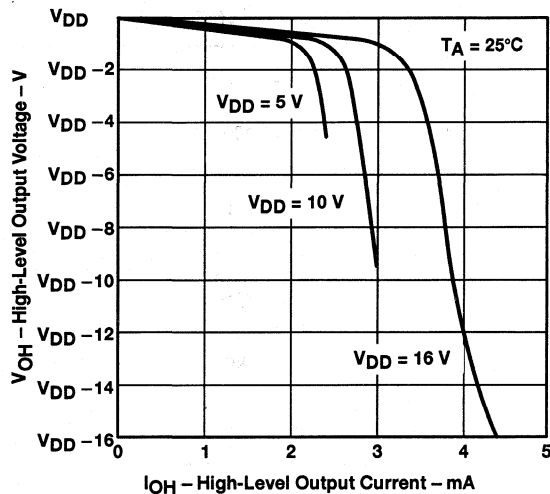
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

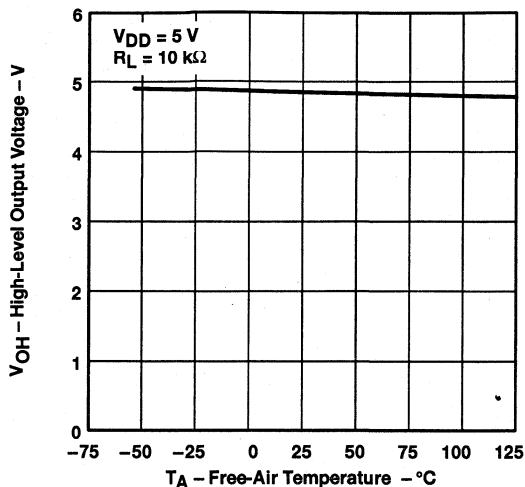
**TYPICAL CHARACTERISTICS†**

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**HIGH-LEVEL OUTPUT CURRENT**



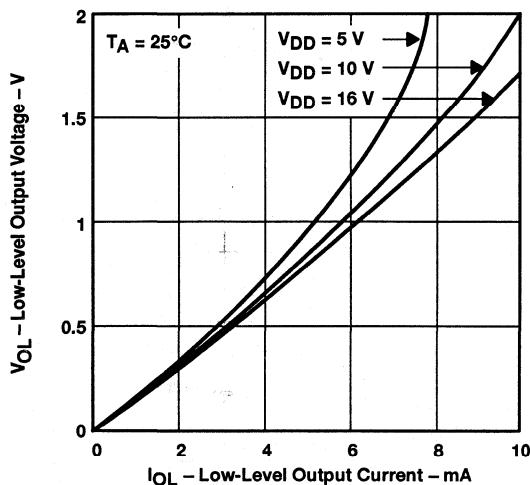
**Figure 13**

**HIGH-LEVEL OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



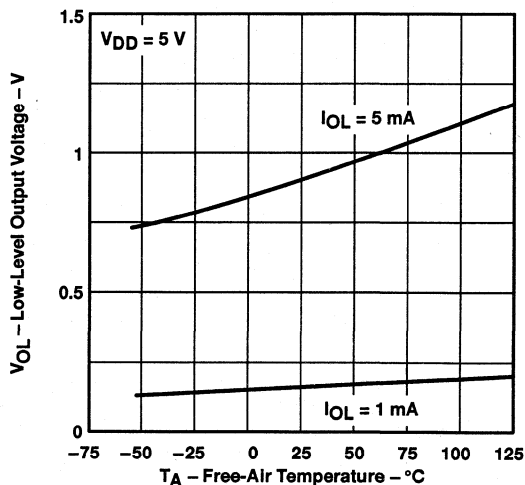
**Figure 14**

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**LOW-LEVEL OUTPUT CURRENT**



**Figure 15**

**LOW-LEVEL OUTPUT VOLTAGE**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 16**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 vs  
 FREQUENCY

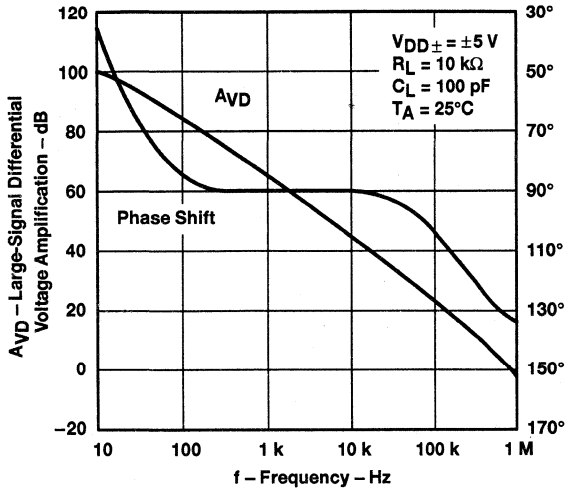


Figure 17

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

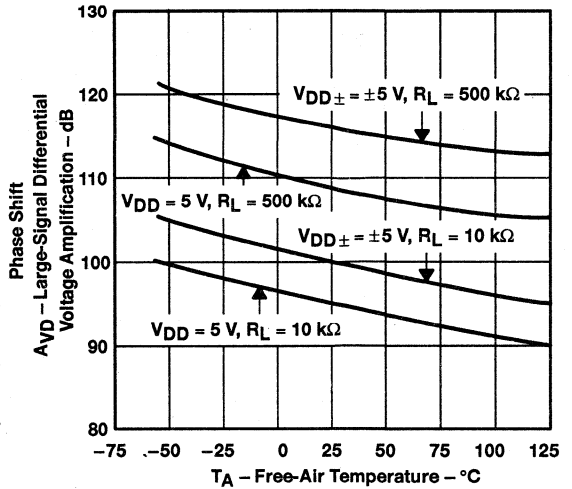


Figure 18

SHORT-CIRCUIT OUTPUT CURRENT  
 vs  
 SUPPLY VOLTAGE

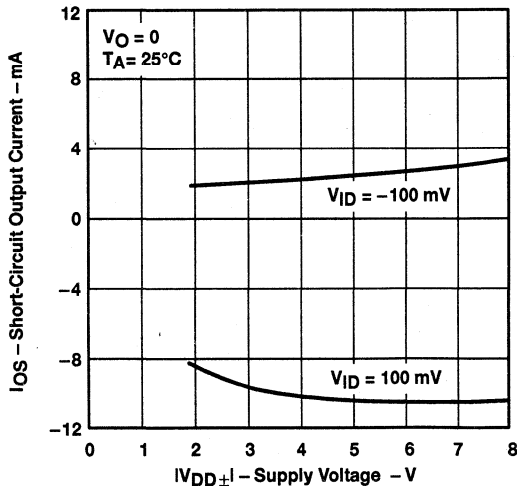


Figure 19

SHORT-CIRCUIT OUTPUT CURRENT  
 vs  
 FREE-AIR TEMPERATURE

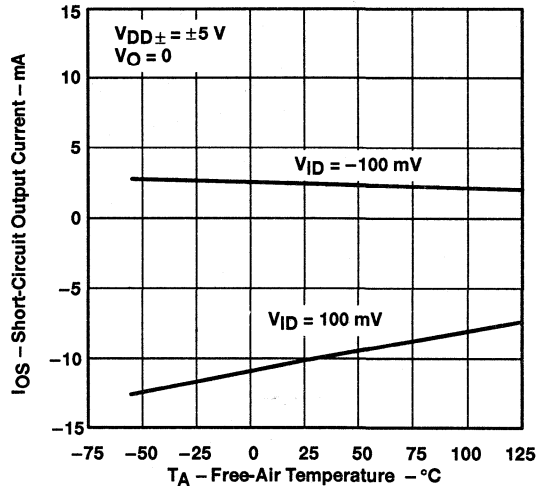


Figure 20

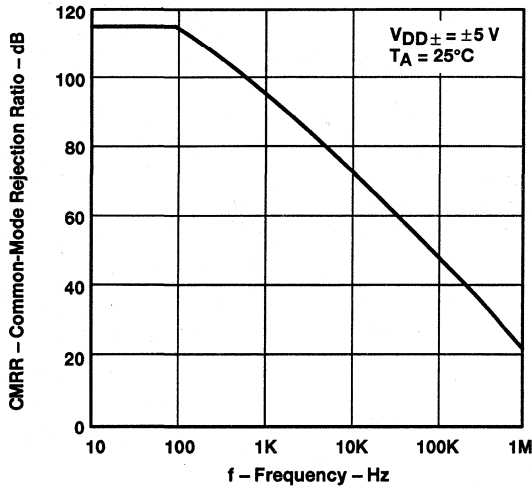
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

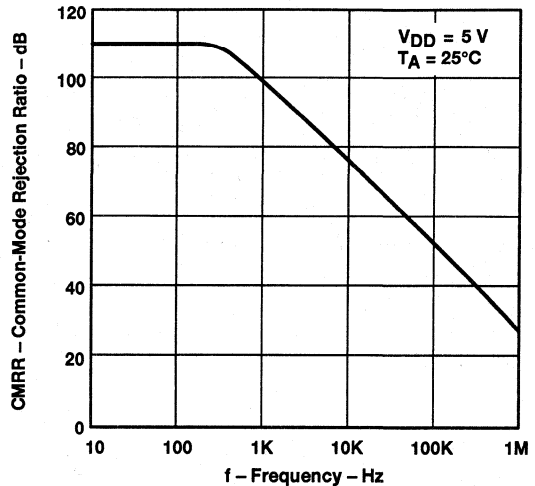
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREQUENCY**



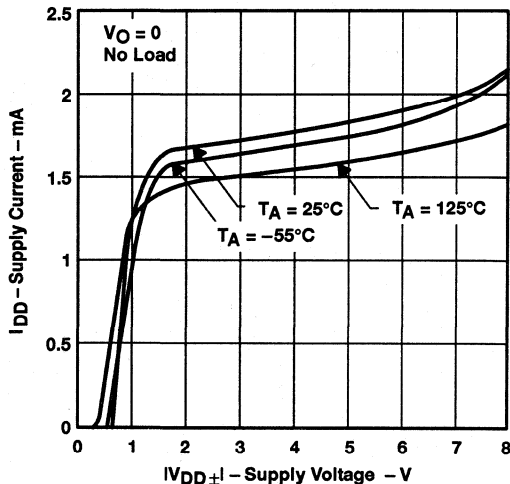
**Figure 21**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREQUENCY**



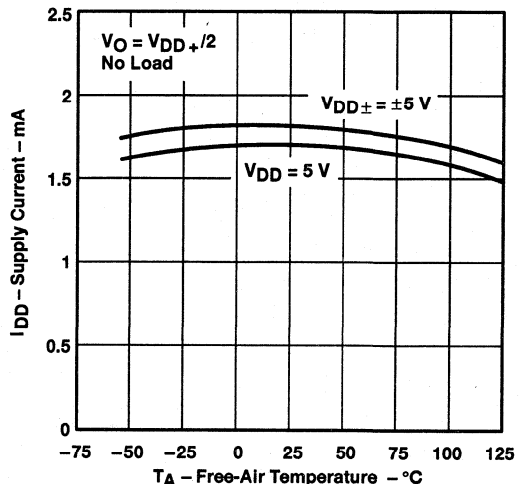
**Figure 22**

**SUPPLY CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 23**

**SUPPLY CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 24**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE

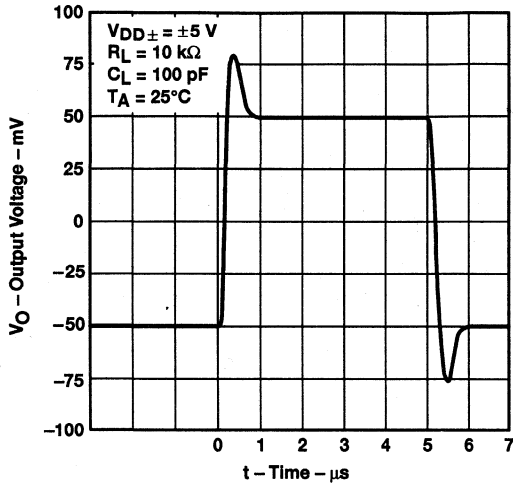


Figure 25

VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE

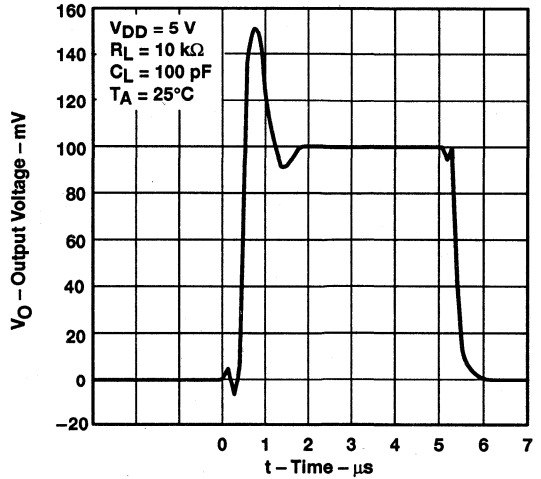


Figure 26

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE

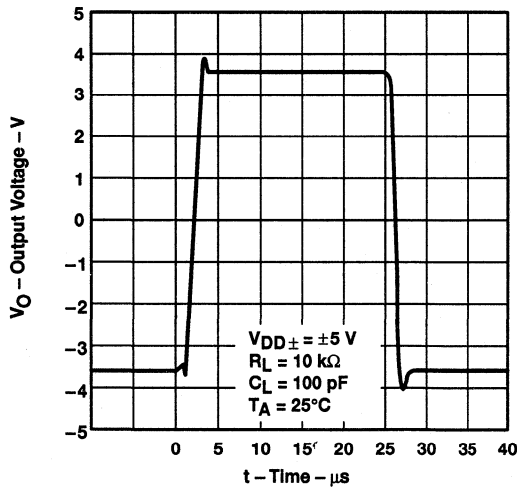


Figure 27

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE

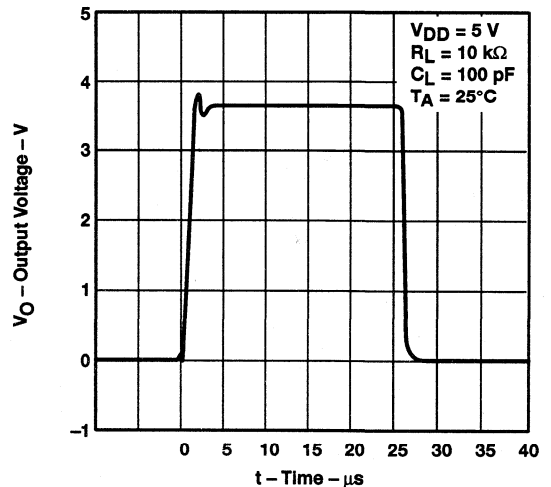


Figure 28

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

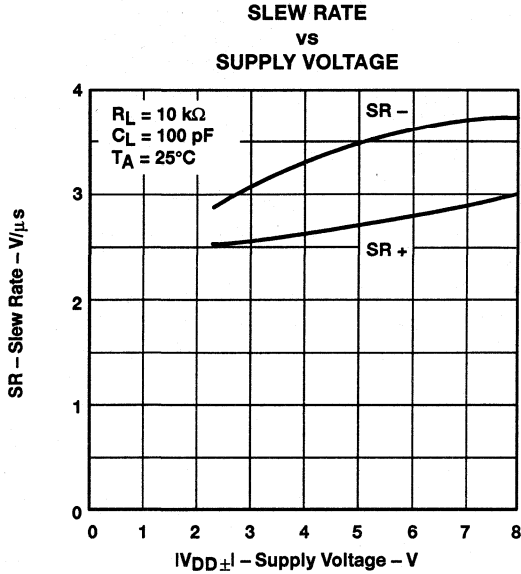


Figure 29

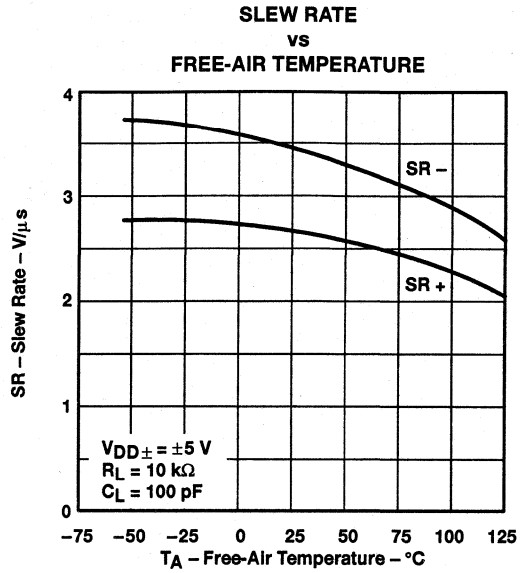


Figure 30

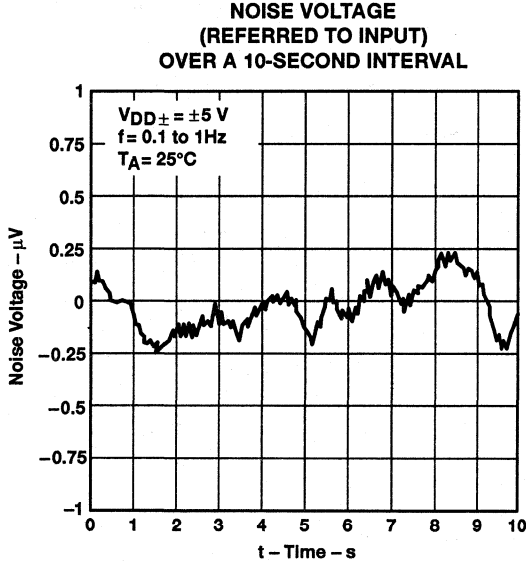


Figure 31

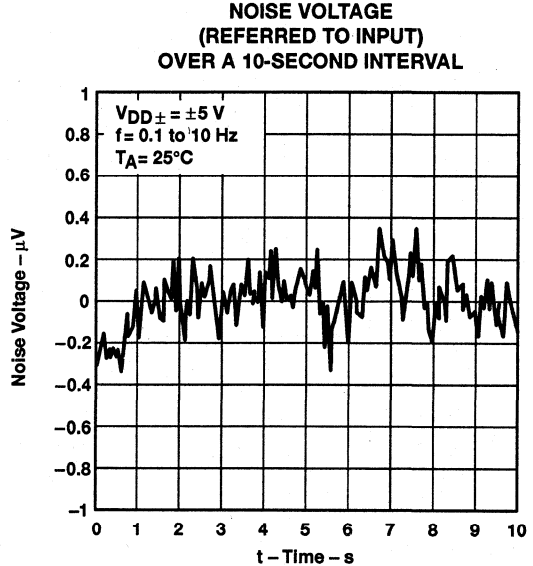


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

GAIN-BANDWIDTH PRODUCT  
 vs  
 SUPPLY VOLTAGE

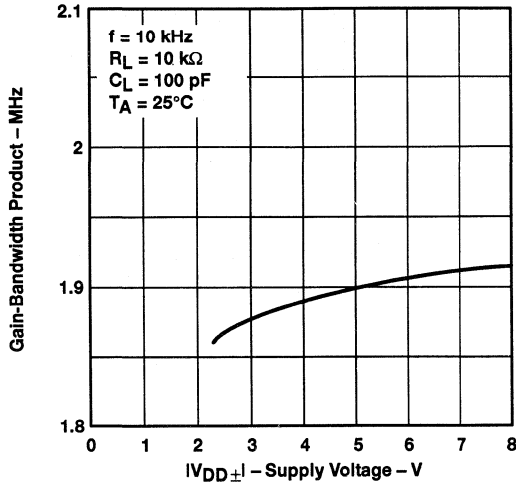


Figure 33

GAIN-BANDWIDTH PRODUCT  
 vs  
 FREE-AIR TEMPERATURE

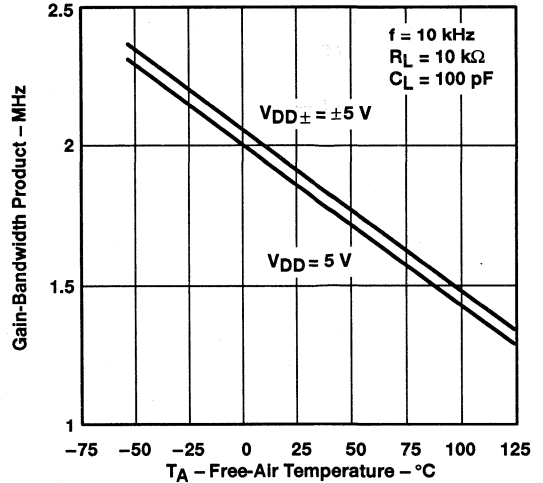


Figure 34

PHASE MARGIN  
 vs  
 SUPPLY VOLTAGE

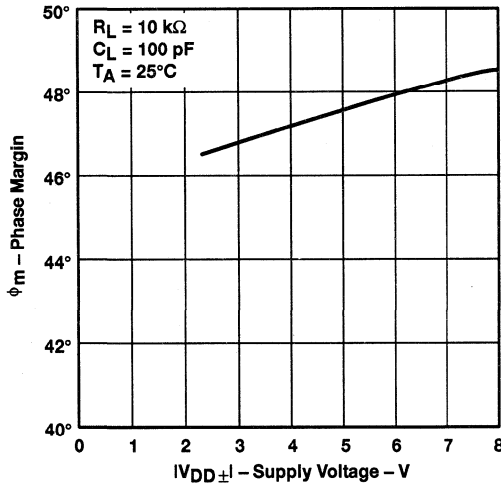


Figure 35

PHASE MARGIN  
 vs  
 FREE-AIR TEMPERATURE

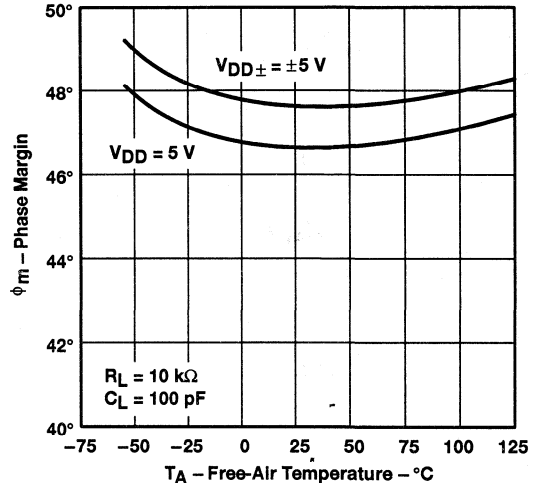


Figure 36

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2202, TLC2202A, TLC2202B, TLC2202Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS056A – MAY 1990 – REVISED AUGUST 1994

---

**APPLICATION INFORMATION**

**latch-up avoidance**

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC2202, TLC2202A, and TLC2202B inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques reducing the chance of latch-up should be used whenever possible. Internal protection diodes should not be forward biased in normal operation. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1  $\mu$ F typical) located across the supply rails as close to the device as possible.

**electrostatic discharge protection**

These devices use internal ESD-protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.



# TLC2262, TLC2262A, TLC2262Y Advanced LinCMOS™ RAIL-TO-RAIL DUAL OPERATIONAL AMPLIFIERS

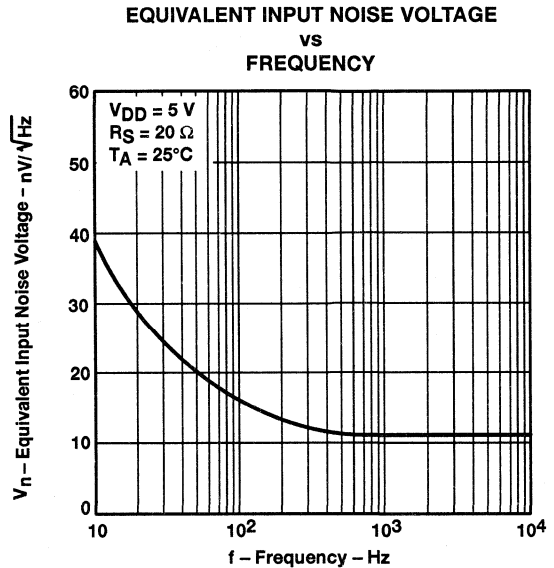
SLOS128A – AUGUST 1993 – REVISED APRIL 1994

- Output Swing includes Both Supply Rails
- Low Noise . . . 12 nV/√Hz Typ at f = 1 kHz
- Low Input Bias Current . . . 1 pA Typ
- Fully Specified for Both Single-Supply and Split-Supply Operation
- Low Power . . . 500 μA Max
- Common-Mode Input Voltage Range Includes Negative Rail
- Low Input Offset Voltage 950 μV Max at T<sub>A</sub> = 25°C (TLC2262A)
- Macromodel Included

## description

The TLC2262 and TLC2262A are dual operational amplifiers manufactured using Texas Instruments Advanced LinCMOS™ process. These devices exhibit rail-to-rail output performance while having better input offset voltage and lower power dissipation levels than existing CMOS operational amplifiers. In addition, the noise performance has been dramatically increased for this class of low-power CMOS amplifier. Figure 1 depicts the low level of voltage noise for this CMOS amplifier, which has only 200 μA (typical) of supply current per amplifier. Also, the common-mode input voltage range is wider than typical standard CMOS-type amplifiers. To take advantage of this improvement in performance and to make this device available for a wider range of applications, V<sub>ICR</sub> is specified with a larger maximum input offset voltage test limit of ±5 mV. The Advanced LinCMOS™ process uses a silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. This technology also makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The TLC2262 and TLC2262A, exhibiting high input impedance and low noise, are excellent for small-signal conditioning for high-impedance sources, such as piezoelectric transducers. Because of the low-power dissipation levels, these devices work well in hand-held monitoring and remote-sensing applications. In addition, the rail-to-rail output feature with single or split supplies makes these devices great choices when interfacing directly to ADCs. All of these features, combined with its temperature performance, make the TLC2262 family ideal for sonobuoys, remote pressure sensors, temperature control, active VR sensors, accelerometers, portable medical applications, hand-held metering, and many other applications.



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES			CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)	
0°C to 70°C	2.5 mV	TLC2262CD	TLC2262CP	TLC2262CPWLE	TLC2262Y
-40°C to 125°C	950 μV 2.5 mV	TLC2262AID	TLC2262AIP	TLC2262AIPWLE	
		TLC2262ID	TLC2262IP	—	

The D packages are available taped and reeled. Add R suffix to device type (e.g., TLC2262CDR). The PW package is available only left-end taped and reeled. Chips are tested at 25°C.

Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

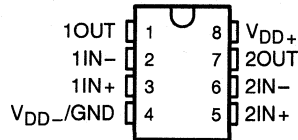
**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**description (continued)**

The device inputs and outputs are designed to withstand a 100-mA surge current without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance. Additional care should be exercised to prevent  $V_{DD+}$  supply line transients under powered conditions. Transients of greater than 20 V can trigger the ESD-protection structure inducing a low-impedance path to  $V_{DD-}/GND$ . Should this condition occur, the sustained current supplied to the device must be limited to 100 mA or less. Failure to do so could result in a latched condition and device failure.

**D, P, OR PW PACKAGE**  
**(TOP VIEW)**

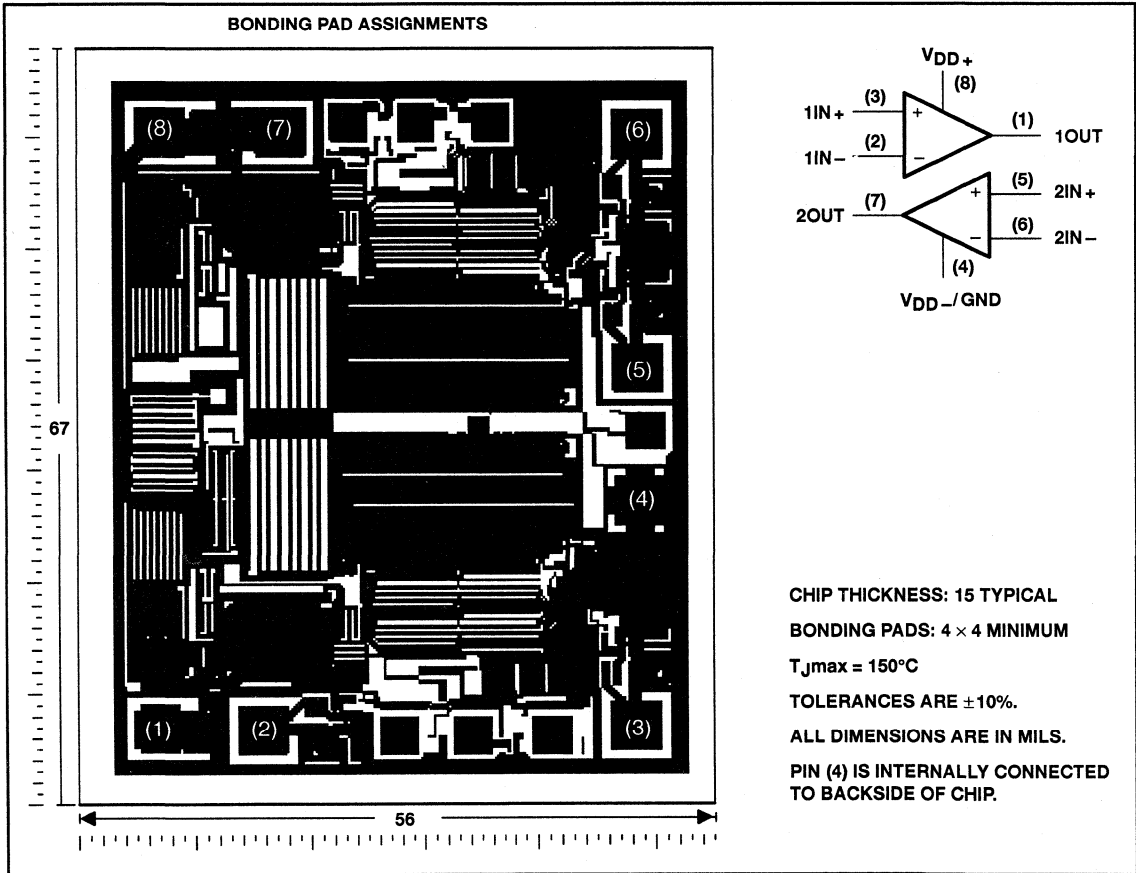


**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

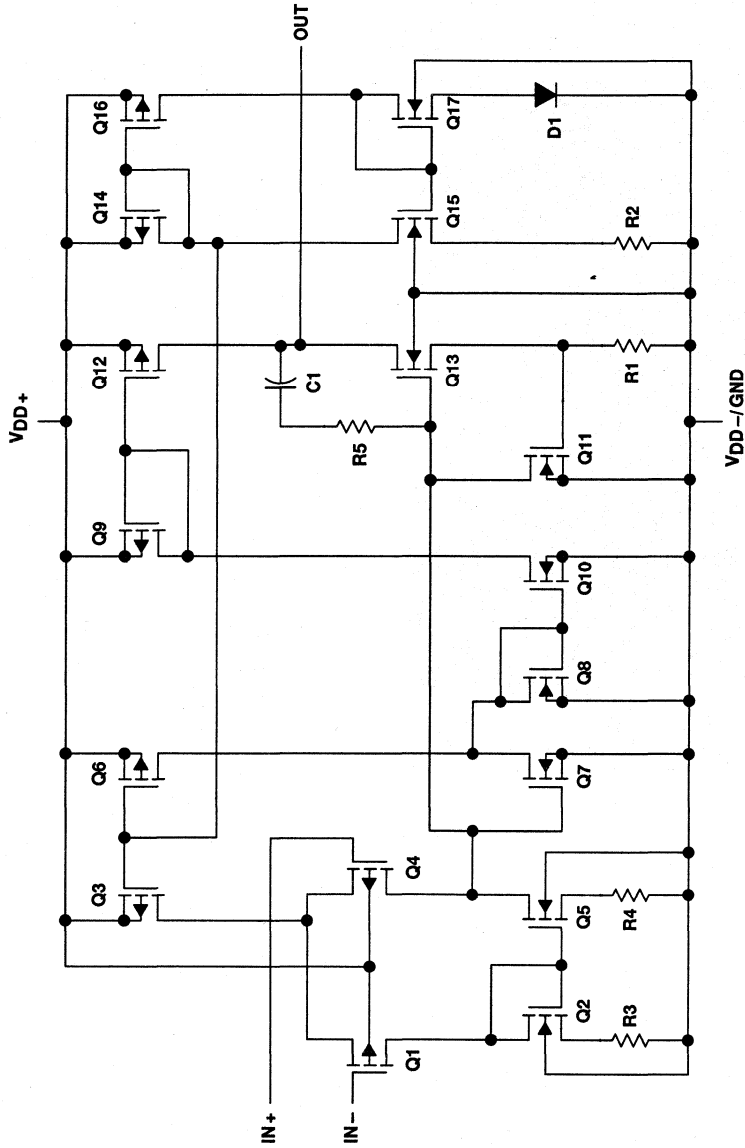
**TLC2262Y chip information**

This chip, when properly assembled, displays characteristics similar to the TLC2262C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. The chip may be mounted with conductive epoxy or a gold-silicon preform.



**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS128A – AUGUST 1993 – REVISED APRIL 1994

equivalent schematic (each amplifier)



COMPONENT COUNT†	
Transistors	38
Diodes	9
Resistors	26
Capacitors	3

† Includes both amplifiers and all ESD, bias, and trim circuitry

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 16$ V
Input voltage, $V_I$ (any input, see Note 1)	$\pm 8$ V
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$	$\pm 50$ mA
Total current into $V_{DD+}$	$\pm 50$ mA
Total current out of $V_{DD-}$	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 125°C
Storage temperature range	-65°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .
  2. Differential voltages are at  $IN+$  with respect to  $IN-$ . Excessive current flows if input is brought below  $V_{DD-} - 0.3$  V.
  3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW
PW	525 mW	4.2 mW/°C	336 mW	273 mW	105 mW

**recommended operating conditions**

	C SUFFIX		I SUFFIX		UNIT
	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	$\pm 2.2$	$\pm 8$	$\pm 2.2$	$\pm 8$	V
Input voltage range, $V_I$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Operating free-air temperature, $T_A$	0	70	-40	125	°C

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $V_O = 0,$ $V_{DD} \pm = \pm 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	300		2500	$\mu\text{V}$
		Full range	3000			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		25°C to 70°C	2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.003			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			$\text{pA}$
		Full range	100			
$I_{IB}$ Input bias current		25°C	1			$\text{pA}$
		Full range	100			
$V_{ICR}$ Common-mode input voltage range		$R_S = 50\ \Omega,$ $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2	$\text{V}$
			Full range	0 to 3.5		
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99		$\text{V}$	
		25°C	4.85	4.94		
	Full range	4.82				
	$I_{OH} = -200\ \mu\text{A}$	25°C	4.70	4.85		
Full range		4.60				
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 50\ \mu\text{A}$	25°C	0.01		$\text{V}$	
		25°C	0.09	0.15		
	Full range	0.15				
	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 500\ \mu\text{A}$	25°C	0.2			0.3
		Full range	0.3			
	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 1\text{ mA}$	25°C	0.7			1
		Full range	1.2			
	$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V},$ $V_O = 1\text{ V to } 4\text{ V}$	$R_L = 50\text{ k}\Omega$ ‡	25°C		80
Full range				55		
$R_L = 1\text{ M}\Omega$ ‡			25°C	550		
$r_{id}$ Differential input resistance		25°C	10 <sup>12</sup>		$\Omega$	
$r_i$ Common-mode input resistance		25°C	10 <sup>12</sup>		$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz},$ P package	25°C	8		$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 100\text{ kHz},$ $A_V = 10$	25°C	240		$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to } 2.7\text{ V},$ $V_O = 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	70	83	$\text{dB}$	
		Full range	70			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to } 16\text{ V},$ $V_{IC} = V_{DD}/2,$ No load	25°C	80	95	$\text{dB}$	
		Full range	80			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V},$ No load	25°C	400	500	$\mu\text{A}$	
		Full range	500			

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 1.5\text{ V to }3.5\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C	0.35	0.55		V/ $\mu\text{s}$
		Full range	0.3			
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C		40		nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	25°C		12		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$	25°C		0.7		$\mu\text{V}$
	$f = 0.1\text{ Hz to }10\text{ Hz}$	25°C		1.3		
$I_n$ Equivalent input noise current		25°C		0.6		fA/ $\sqrt{\text{Hz}}$
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}, f = 20\text{ kHz}, R_L = 50\text{ k}\Omega^\ddagger$	$A_V = 1$	0.017%			
		$A_V = 10$	0.03%			
Gain-bandwidth product	$f = 10\text{ kHz}, C_L = 100\text{ pF}^\ddagger, R_L = 50\text{ k}\Omega^\ddagger$	25°C		0.71		MHz
BOM Maximum output-swing bandwidth	$V_{O(PP)} = 2\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, A_V = 1, C_L = 100\text{ pF}^\ddagger$	25°C		185		kHz
Settling time	$A_V = -1, \text{ Step} = 0.5\text{ V to }2.5\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	To 0.1%	6.4			$\mu\text{s}$
		To 0.01%	14.1			
$\phi_m$ Phase margin at unity gain	$R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C		63°		
Gain margin	$R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C		14		dB

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A—AUGUST 1993—REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V (unless otherwise specified)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262C			UNIT		
			MIN	TYP	MAX			
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $R_S = 50 \Omega$ $V_O = 0,$	25°C	300	2500		$\mu V$		
		Full range		3000				
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 70°C		2		$\mu V/^\circ C$		
		Input offset voltage long-term drift (see Note 4)	25°C	0.003			$\mu V/mo$	
$I_{IO}$ Input offset current		25°C		0.5		pA		
		Full range			100			
$I_{IB}$ Input bias current		25°C		1		pA		
		Full range			100			
$V_{ICR}$ Common-mode input voltage range		$ V_{IO}  \leq 5$ mV, $R_S = 50 \Omega$	25°C	-5 to 4	-5.3 to 4.2		V	
			Full range	-5 to 3.5				
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20 \mu A$	25°C		4.99		V		
		Full range	4.85	4.94				
	$I_O = -100 \mu A$	25°C		4.7	4.85			
		Full range	4.6					
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0,$ $I_O = 50 \mu A$	25°C		-4.99		V		
		Full range	-4.85	-4.91				
	$V_{IC} = 0,$ $I_O = 500 \mu A$	25°C		-4.7	-4.8			
		Full range	-4.7					
	$V_{IC} = 0,$ $I_O = 1$ mA	25°C		-4	-4.3			
		Full range	-3.8					
	$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4$ V	$R_L = 50$ k $\Omega$	25°C	80		200	V/mV
				Full range	55			
$R_L = 1$ M $\Omega$			25°C		1000			
			Full range					
$r_{id}$ Differential input resistance		25°C		$10^{12}$	$\Omega$			
$r_i$ Common-mode input resistance		25°C		$10^{12}$	$\Omega$			
$c_i$ Common-mode input capacitance	$f = 10$ kHz, P package	25°C		8	pF			
$z_o$ Closed-loop output impedance	$f = 100$ kHz, $A_V = 10$	25°C		220	$\Omega$			
CMRR Common-mode rejection ratio	$V_{IC} = -5$ V to 2.7 V, $V_O = 0$ V, $R_S = 50 \Omega$	25°C	75	88	dB			
		Full range	75					
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD\pm} = 2.2$ V to $\pm 8$ V, $V_{IC} = 0,$ No load	25°C	80	95	dB			
		Full range	80					
$I_{DD}$ Supply current	$V_O = 0$ V, No load	25°C	425	500	$\mu A$			
		Full range		500				

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ C$  extrapolated to  $T_A = 25^\circ C$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2262C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	V <sub>O</sub> = ±1.9 V, C <sub>L</sub> = 100 pF R <sub>L</sub> = 50 kΩ	25°C	0.35	0.55		V/μs
		Full range	0.3			
V <sub>n</sub> Equivalent input noise voltage	f = 10 Hz	25°C	43		nV/√Hz	
	f = 1 kHz	25°C	12			
V <sub>N(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 1 Hz	25°C	0.8		μV	
	f = 0.1 Hz to 10 Hz	25°C	1.3			
I <sub>n</sub> Equivalent input noise current		25°C	0.6		fA/√Hz	
THD + N Total harmonic distortion pulse duration	V <sub>O</sub> = ±2.3 V, f = 20 kHz, R <sub>L</sub> = 50 kΩ	A <sub>V</sub> = 1	0.014%			
		A <sub>V</sub> = 10	0.024%			
Gain-bandwidth product	f = 10 kHz, C <sub>L</sub> = 100 pF R <sub>L</sub> = 50 kΩ	25°C	0.73		MHz	
BOM Maximum output-swing bandwidth	V <sub>O(PP)</sub> = 4.6 V, R <sub>L</sub> = 50 kΩ	A <sub>V</sub> = 1, C <sub>L</sub> = 100 pF	25°C	70		kHz
Settling time	A <sub>V</sub> = -1, Step = -2.3 V to 2.3 V, R <sub>L</sub> = 50 kΩ, C <sub>L</sub> = 100 pF	To 0.1%	25°C	7.1		μs
		To 0.01%		16.5		
φ <sub>m</sub> Phase margin at unity gain Gain margin	R <sub>L</sub> = 50 kΩ, C <sub>L</sub> = 100 pF	25°C	64°		dB	
		25°C	14			

† Full range is 0°C to 70°C.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262I			TLC2262AI			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$	
		Full range			3000			1500		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage	$V_{DD} \pm \pm 2.5\text{ V}$ , $V_{IC} = 0$ , $V_O = 0$ , $R_S = 50\ \Omega$	25°C to 85°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.003			0.003			$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$	
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$	
		Full range	500			500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2		V	
		Full range	0 to 3.5			0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99			4.99			V	
	$I_{OH} = -100\ \mu\text{A}$	25°C	4.85	4.94		4.85	4.94			
	$I_{OH} = -200\ \mu\text{A}$	25°C	4.7	4.85		4.7	4.85			
		Full range	4.5			4.5				
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$	25°C	0.01			0.01			V	
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$	25°C	0.09	0.15		0.09	0.15			
		Full range	0.15			0.15				
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 4\text{ mA}$	25°C	0.8	1		0.7	1			
Full range		1.2			1.2					
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 50\ \text{k}\Omega^\ddagger$	25°C	80	100		80	170		V/mV
		$R_L = 1\ \text{M}\Omega^\ddagger$	Full range	50			50			
			25°C	550			550			
$r_{id}$ Differential input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$	
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}$ , P package	25°C	8			8			$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz}$ , $A_V = 10$	25°C	240			240			$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$	25°C	70	83		70	83		dB	
		Full range	70			70				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load	25°C	80	95		80	95		dB	
		Full range	80			80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load	25°C	400	500		400	500		$\mu\text{A}$	
		Full range	500			500				

† Full range is  $-40^\circ\text{C}$  to  $125^\circ\text{C}$ .

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262I			TLC2262AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = 1.5\text{ V to }3.5\text{ V}$ , $C_L = 100\text{ pF}‡$ , $R_L = 50\text{ k}\Omega‡$	25°C	0.35	0.55		0.35	0.55	V/ $\mu\text{s}$	
		Full range	0.25			0.25			
$V_n$	Equivalent input noise voltage	f = 10 Hz	40			40			nV/ $\sqrt{\text{Hz}}$
		f = 1 kHz	12			12			
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 1 Hz	0.7			0.7			$\mu\text{V}$
		f = 0.1 Hz to 10 Hz	1.3			1.3			
$I_n$	Equivalent input noise current	25°C	0.6			0.6			fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}$ , f = 20 kHz, $R_L = 50\text{ k}\Omega‡$	$A_V = 1$	0.017%			0.017%			
		$A_V = 10$	0.03%			0.03%			
	Gain-bandwidth product f = 50 kHz, $C_L = 100\text{ pF}‡$	$R_L = 50\text{ k}\Omega‡$ , 25°C	0.82			0.82			MHz
$B_{OM}$	Maximum output-swing bandwidth $V_{O(PP)} = 2\text{ V}$ , $R_L = 50\text{ k}\Omega‡$	$A_V = 1$ , $C_L = 100\text{ pF}‡$ , 25°C	185			185			kHz
	Settling time $A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 50\text{ k}\Omega‡$ , $C_L = 100\text{ pF}‡$	To 0.1%	6.4			6.4			$\mu\text{s}$
		To 0.01%	14.1			14.1			
$\phi_m$	Phase margin at unity gain $R_L = 50\text{ k}\Omega‡$ , $C_L = 100\text{ pF}‡$	25°C	63°			63°			
		25°C	14			14			

† Full range is –40°C to 125°C.

‡ Referenced to 2.5 V

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2262I			TLC2262AI			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage		25°C	300 2500			300 950			$\mu\text{V}$	
		Full range	3000			1500				
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 85°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C	0.003			0.003			$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$	
		Full range	500			500				
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$	
		Full range	500			500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	25°C	-5 to 4	-5.3 to 4.2		-5 to 4	-5.3 to 4.2		V	
		Full range	-5 to 3.5			-5 to 3.5				
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C	4.99			4.99			V	
		25°C	4.85	4.94		4.85	4.94			
		Full range	4.82			4.82				
		25°C	4.7	4.85		4.7	4.85			
$V_{OM-}$ Maximum negative peak output voltage	$I_O = -200\ \mu\text{A}$	25°C	-4.99			-4.99			V	
		25°C	-4.85	-4.91		-4.85	-4.91			
		Full range	-4.85			-4.85				
		25°C	-4	-4.3		-4	-4.3			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 50\ \text{k}\Omega$	25°C	80 200			80 200			V/mV
			Full range	50			50			
		$R_L = 1\ \text{M}\Omega$	25°C	1000			1000			
			Full range	1000			1000			
$r_{id}$ Differential input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>			$\Omega$	
$r_i$ Common-mode input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>			$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, \text{ P package}$	25°C	8			8			$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz}, A_V = 10$	25°C	220			220			$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	75	88		75	88		dB	
		Full range	75			75				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} = 4.4\ \text{V to } 16\ \text{V}, V_{IC} = V_{DD}/2, \text{ No load}$	25°C	80	95		80	95		dB	
		Full range	80			80				
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}, \text{ No load}$	25°C	425	500		425	500		$\mu\text{A}$	
		Full range	500			500				

† Full range is  $-40^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS		T <sub>A</sub> †	TLC2262I			TLC2262AI			UNIT	
				MIN	TYP	MAX	MIN	TYP	MAX		
SR	Slew rate at unity gain	V <sub>O</sub> = ±1.9 V, C <sub>L</sub> = 100 pF	R <sub>L</sub> = 50 kΩ,	25°C	0.35	0.55		0.35	0.55	V/μs	
				Full range	0.25		0.25				
V <sub>n</sub>	Equivalent input noise voltage	f = 10 Hz	f = 1 kHz	25°C	43			43			nV/√Hz
				25°C	12			12			
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 1 Hz	f = 0.1 Hz to 10 Hz	25°C	0.8			0.8			μV
				25°C	1.3			1.3			
I <sub>n</sub>	Equivalent input noise current			25°C	0.6			0.6			fA√Hz
THD + N	Total harmonic distortion plus noise	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 50 kΩ, f = 20 kHz	A <sub>V</sub> = 1	25°C	0.014%			0.014%			
					A <sub>V</sub> = 10	0.024%			0.024%		
	Gain-bandwidth product	f = 10 kHz, C <sub>L</sub> = 100 pF	R <sub>L</sub> = 50 kΩ,	25°C	0.73			0.73			MHz
BOM	Maximum output-swing bandwidth	V <sub>O(PP)</sub> = 4.6 V, R <sub>L</sub> = 50 kΩ,	A <sub>V</sub> = 1, C <sub>L</sub> = 100 pF	25°C	70			70			kHz
	Settling time	A <sub>V</sub> = -1, Step = -2.3 V to 2.3 V, R <sub>L</sub> = 50 kΩ, C <sub>L</sub> = 100 pF	To 0.1%	25°C	7.1			7.1			μs
					To 0.01%	16.5			16.5		
φ <sub>m</sub>	Phase margin at unity gain	R <sub>L</sub> = 50 kΩ,	C <sub>L</sub> = 100 pF	25°C	64°			64°			
	Gain margin			25°C	14			14			

† Full range is -40°C to 125°C.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2262Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $V_O = 0$ , $V_{DD\pm} = \pm 2.5\text{ V}$ , $R_S = 50\ \Omega$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$ V_{IO}  \leq 5\text{ mV}$ , $R_S = 50\ \Omega$	0 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$		4.99		V
	$I_{OH} = -100\ \mu\text{A}$	4.85	4.94		
	$I_{OH} = -200\ \mu\text{A}$	4.7	4.85		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$		0.01		V
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$	0.09	0.15		
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 4\text{ mA}$		0.8	1	
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 50\ \text{k}\Omega^\dagger$	80	170	V/mV
		$R_L = 1\ \text{M}\Omega^\dagger$		550	
$r_{id}$ Differential input resistance			1012		$\Omega$
$r_i$ Common-mode input resistance			1012		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$		8		$\text{pF}$
$Z_O$ Closed-loop output impedance	$f = 100\text{ kHz}$ , $A_V = 10$		240		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$	70	83		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load	80	95		dB
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load		400	500	$\mu\text{A}$

† Referenced to 2.5 V





**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2262Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $V_O = 0$ $R_S = 50\ \Omega,$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$ V_{IO}  \leq 5\text{ mV},$ $R_S = 50\ \Omega$	-5 to 4	-5.3 to 4.2		V
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$		4.99		V
	$I_O = -100\ \mu\text{A}$	4.85	4.94		
	$I_O = -200\ \mu\text{A}$	4.7	4.85		
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0,$ $I_{OL} = 50\ \mu\text{A}$		-4.99		V
	$V_{IC} = 0,$ $I_{OL} = 500\ \mu\text{A}$	-4.85	-4.91		
	$V_{IC} = 0,$ $I_{OL} = 4\ \text{mA}$	-3.8	-4.1		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 50\ \text{k}\Omega$	80	200	V/mV
		$R_L = 1\ \text{M}\Omega$		1000	
$r_{id}$ Differential input resistance			$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}$		8		$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz},$ $A_V = 10$		220		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega$	75	88		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V},$ $V_{IC} = 0,$ No load	80	95		dB
$I_{DD}$ Supply current	$V_O = 0,$ No load	425	500		$\mu\text{A}$



**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution vs Common-mode input voltage	2,3 4,5
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	6,7
$I_{B}/I_{IO}$	Input bias and input offset currents	vs Free-air temperature	8
$V_I$	Input voltage range	vs Supply voltage vs Free-air temperature	9 10
$V_{OH}$	High-level output voltage	vs High-level output current	11
$V_{OL}$	Low-level output voltage	vs Low-level output current	12,13
$V_{OM+}$	Maximum positive peak output voltage	vs Output current	14
$V_{OM-}$	Maximum negative peak output voltage	vs Output current	15
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	16
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Free-air temperature	17 18
$V_O$	Output voltage	vs Differential input voltage	19,20
	Differential gain	vs Load resistance	21
$A_{VD}$	Large-signal differential voltage amplification	vs Frequency vs Free-air temperature	22,23 24,25
$z_o$	Output impedance	vs Frequency	26, 27
CMRR	Common-mode rejection ratio	vs Frequency vs Free-air temperature	28 29
kSVR	Supply-voltage rejection ratio	vs Frequency vs Free-air temperature	30, 31 32
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	33 34
SR	Slew rate	vs Load capacitance vs Free-air temperature	35 36
$V_O$	Large-signal pulse response	vs Time	37, 38, 39, 40
$V_O$	Small-signal pulse response	vs Time	41, 42, 43, 44
$V_n$	Equivalent input noise voltage	vs Frequency	45, 46
	Noise voltage (referred to input)	Over a 10-second period	47
	Integrated noise voltage	vs Frequency	48
THD + N	Total harmonic distortion plus noise	vs Frequency	49
	Gain-bandwidth product	vs Free-air temperature vs Supply voltage	50 51
$\phi_m$	Phase margin	vs Frequency vs Load capacitance	22, 23 52
$A_m$	Gain margin	vs Load capacitance	53
$B_1$	Unity-gain bandwidth	vs Load capacitance	54
	Overestimation of phase margin	vs Load capacitance	55



TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC2262  
 INPUT OFFSET VOLTAGE

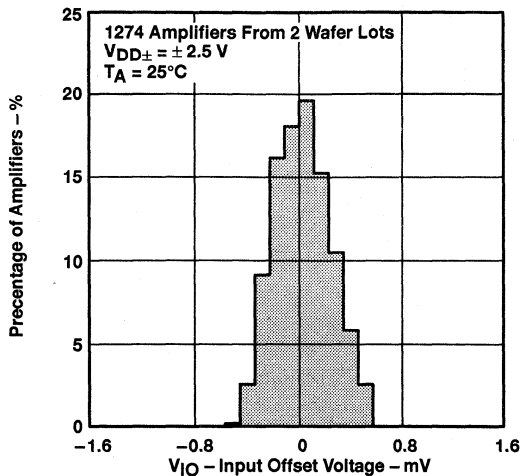


Figure 2

DISTRIBUTION OF TLC2262  
 INPUT OFFSET VOLTAGE

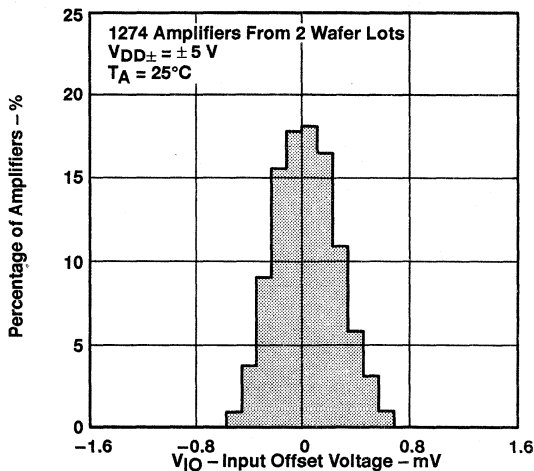


Figure 3

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

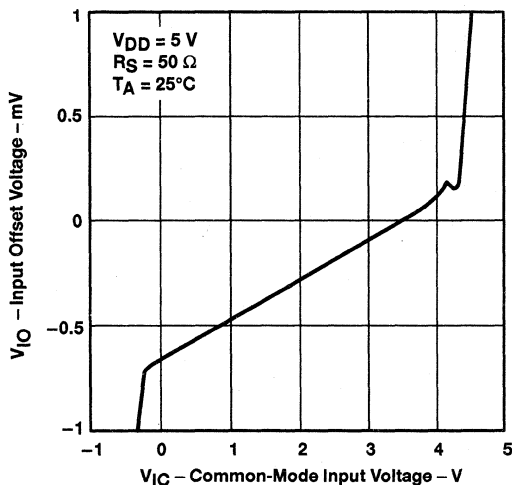


Figure 4

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

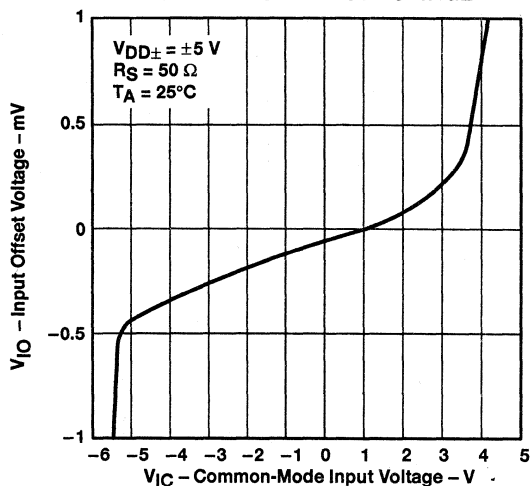


Figure 5

† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS†**

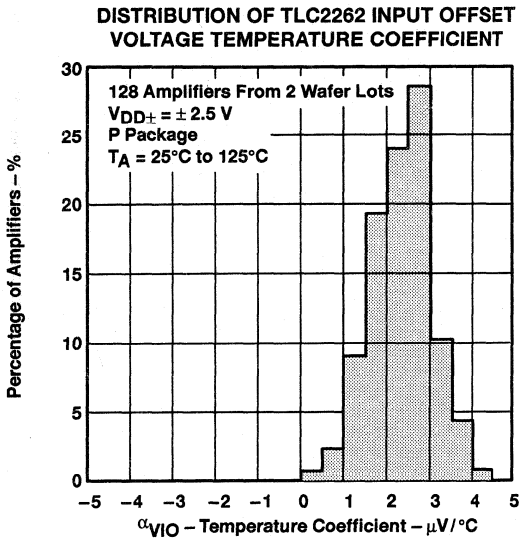


Figure 6

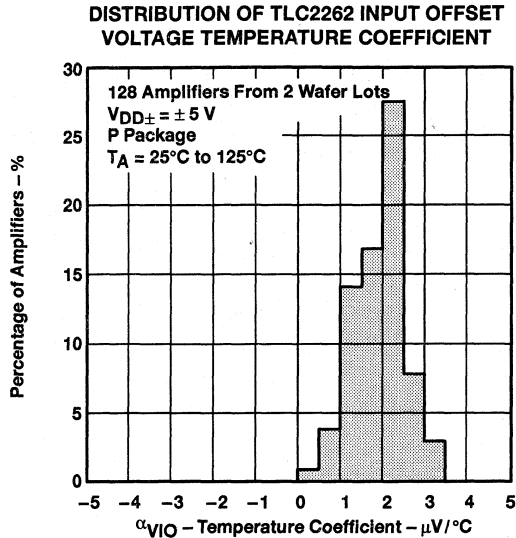


Figure 7

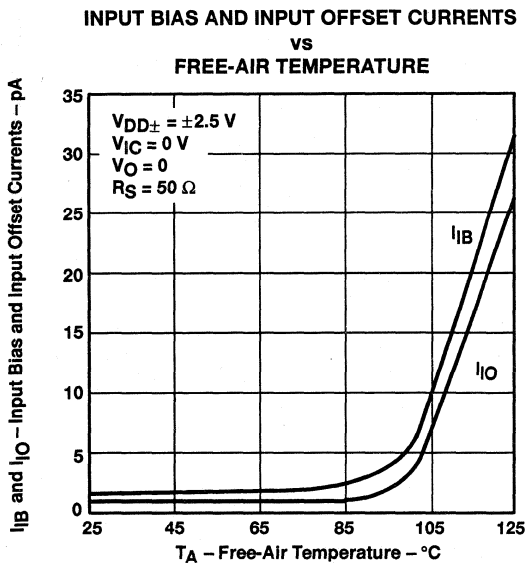


Figure 8

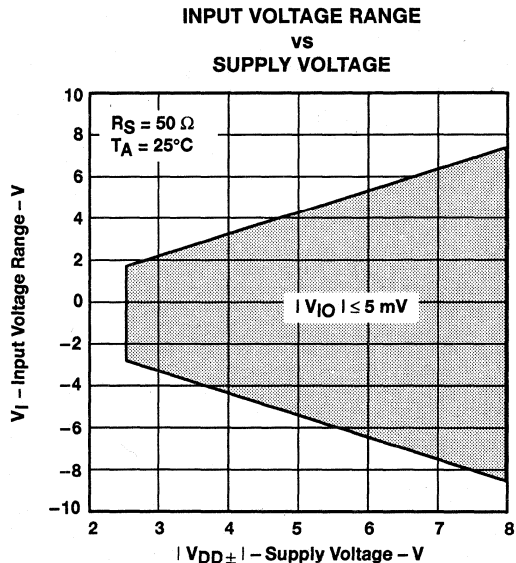


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

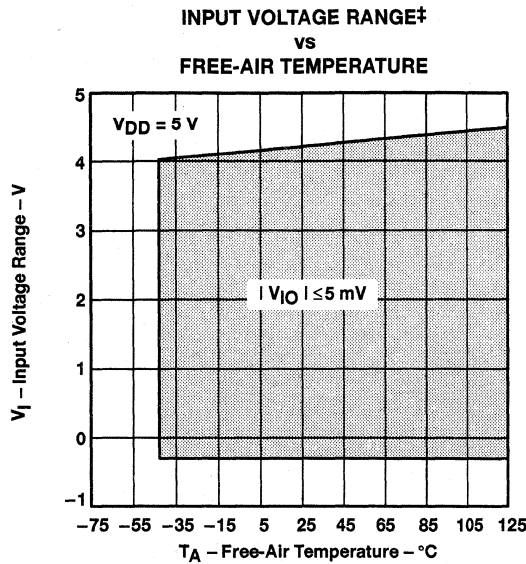


Figure 10

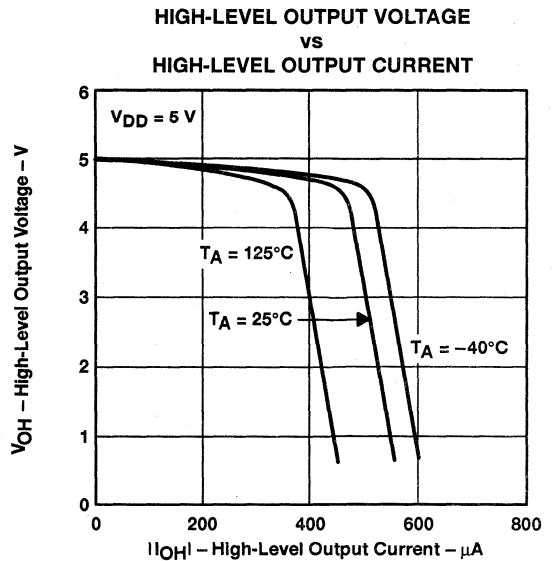


Figure 11

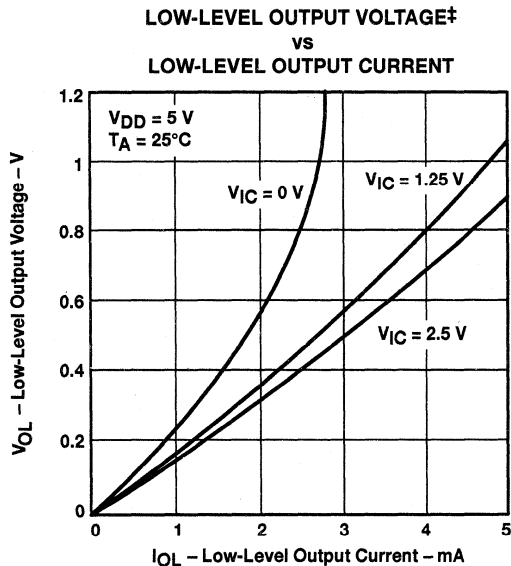


Figure 12

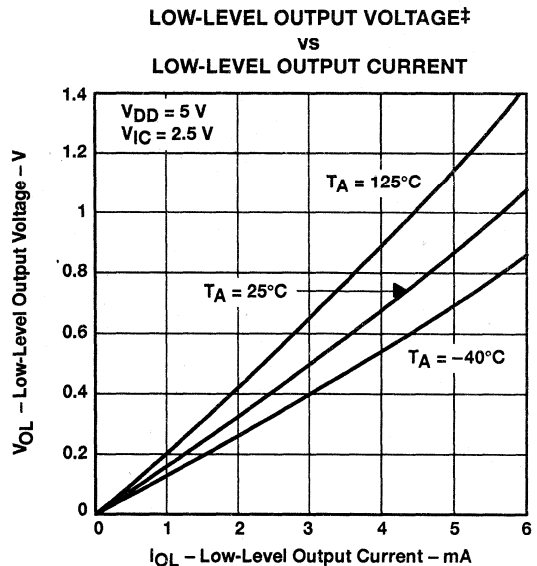


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

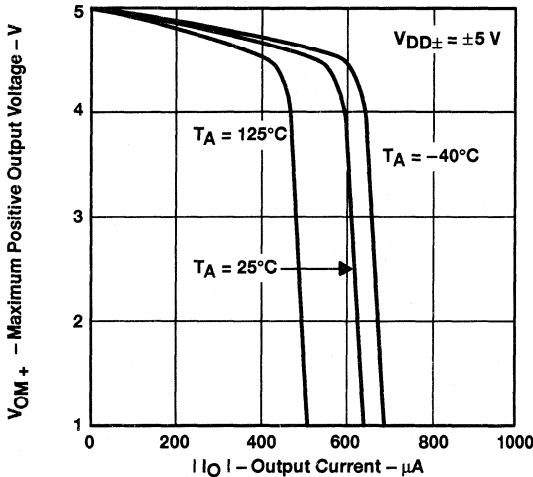
‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

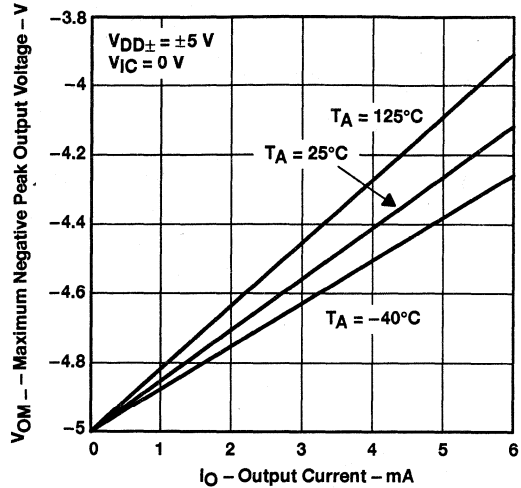
**TYPICAL CHARACTERISTICS**

**MAXIMUM POSITIVE OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT**



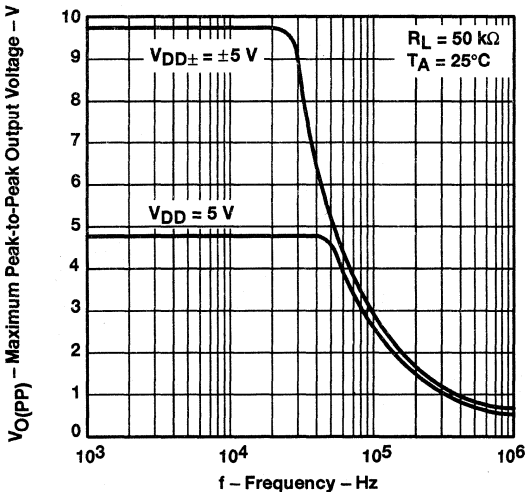
**Figure 14**

**MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE  
vs  
OUTPUT CURRENT**



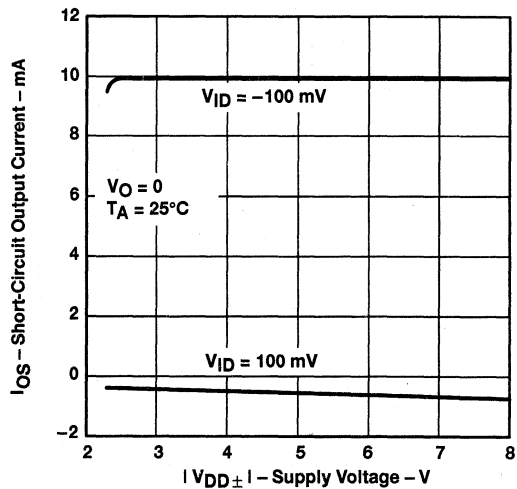
**Figure 15**

**MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE†  
vs  
FREQUENCY**



**Figure 16**

**SHORT-CIRCUIT OUTPUT CURRENT  
vs  
SUPPLY VOLTAGE**



**Figure 17**

† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS†

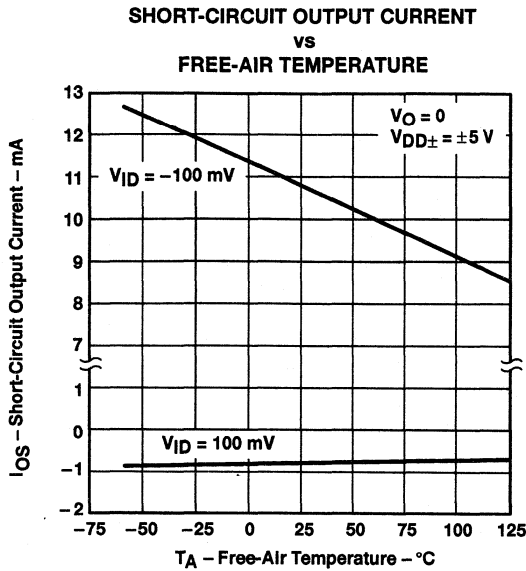


Figure 18

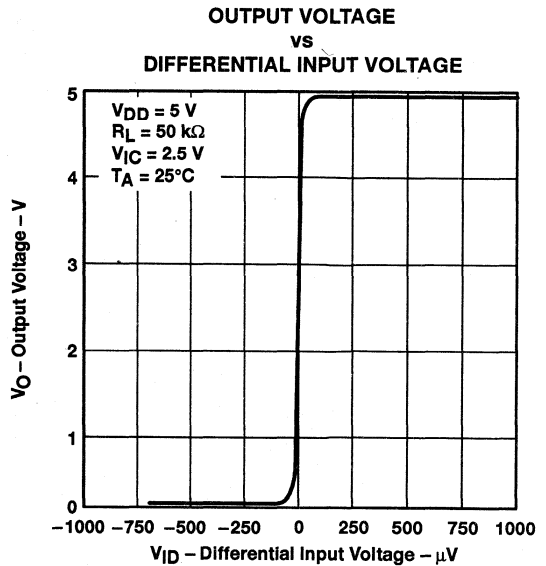


Figure 19

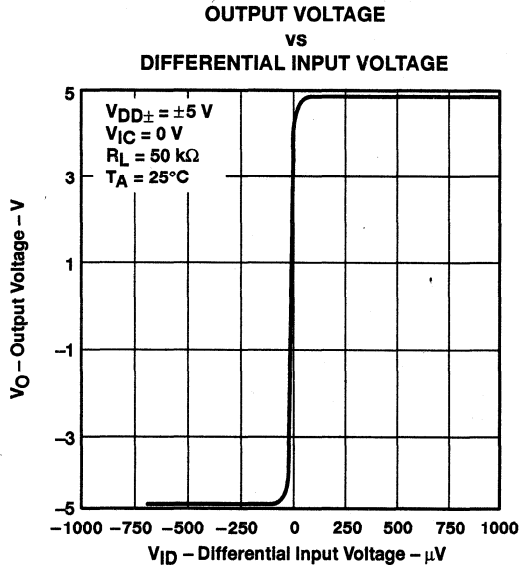


Figure 20

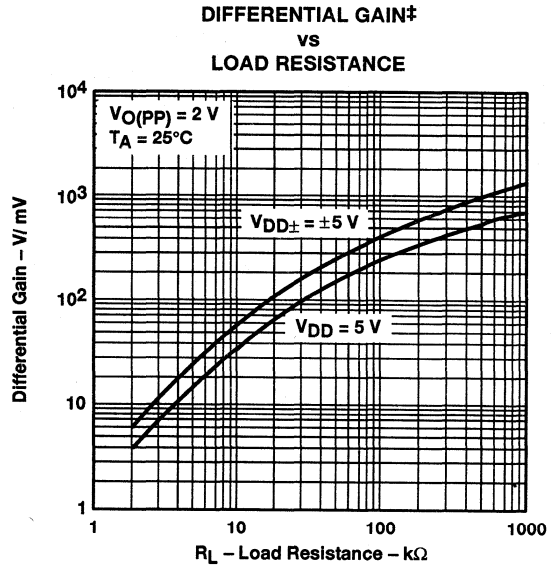


Figure 21

‡ For curves where  $V_{DD} = 5 \text{ V}$ , all loads are referenced to 2.5 V.

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

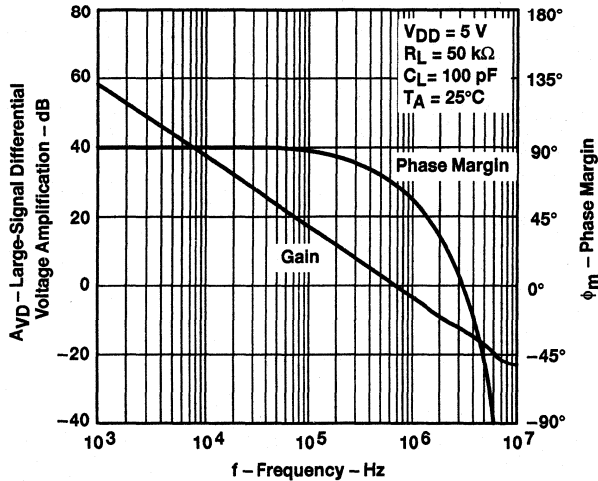
**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE†  
AMPLIFICATION AND PHASE MARGIN**

**vs  
FREQUENCY**

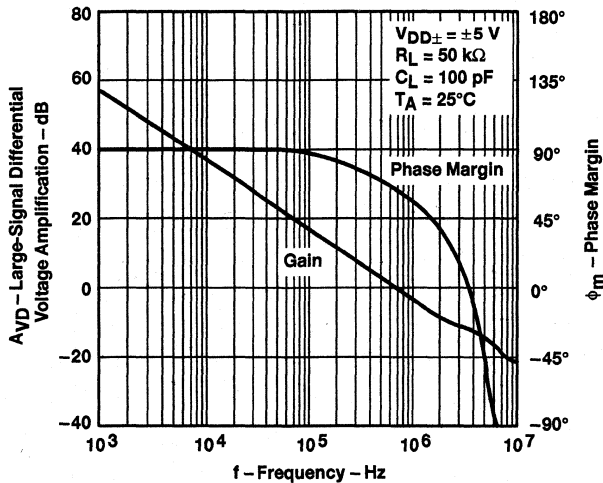


† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**Figure 22**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
AMPLIFICATION AND PHASE MARGIN**

**vs  
FREQUENCY**



**Figure 23**



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL‡  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

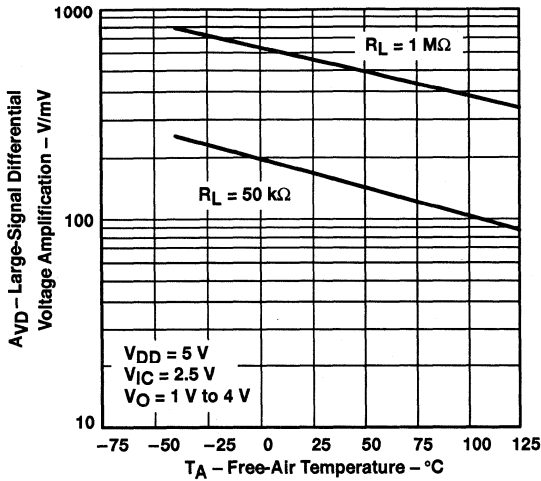


Figure 24

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

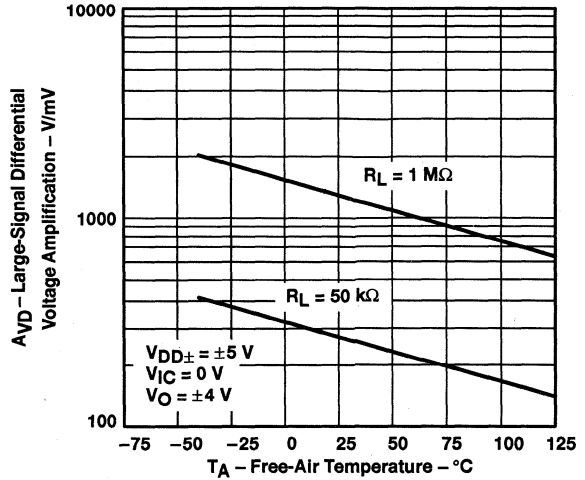


Figure 25

OUTPUT IMPEDANCE‡  
 VS  
 FREQUENCY

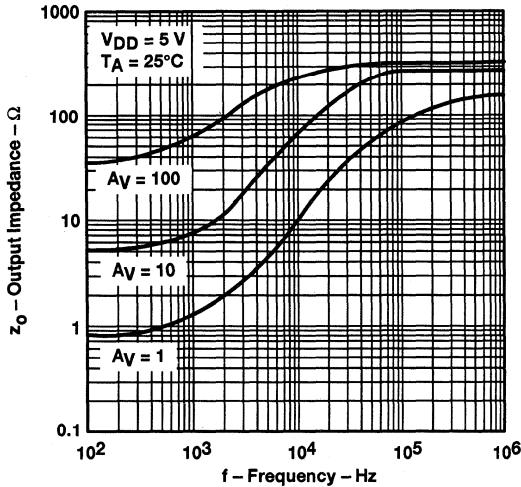


Figure 26

OUTPUT IMPEDANCE  
 VS  
 FREQUENCY

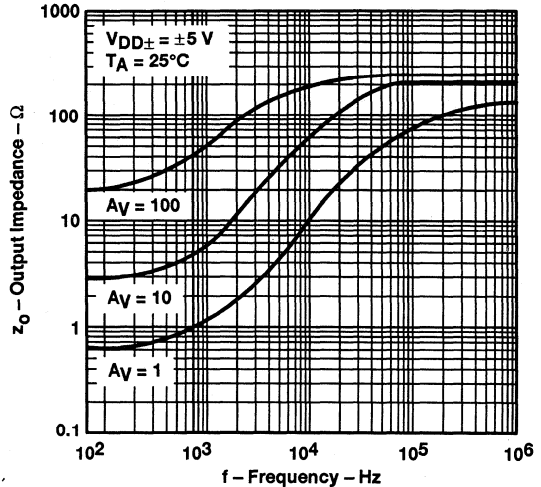


Figure 27

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

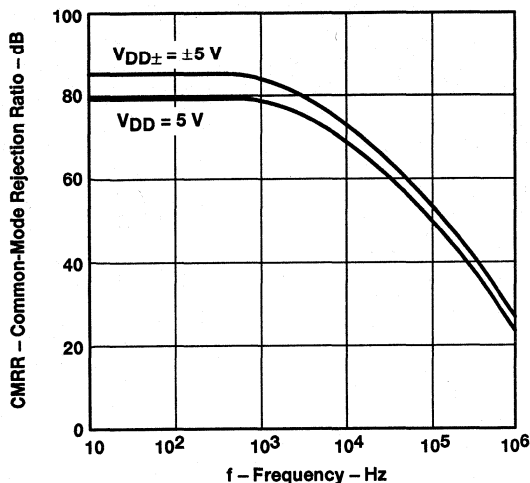
‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

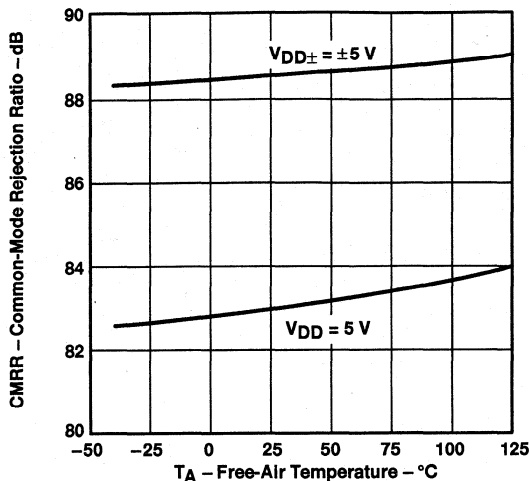
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO‡**  
**vs**  
**FREQUENCY**



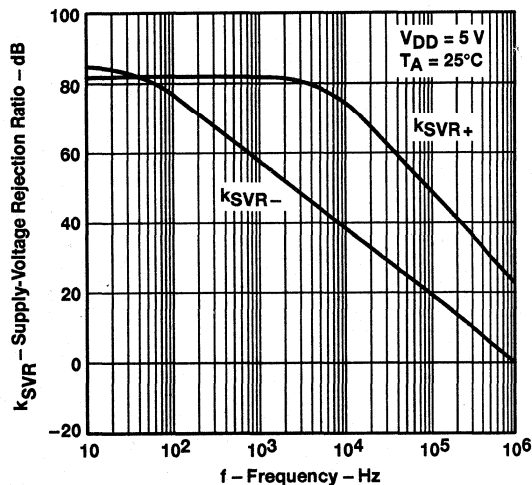
**Figure 28**

**COMMON-MODE REJECTION RATIO‡**  
**vs**  
**FREE-AIR TEMPERATURE**



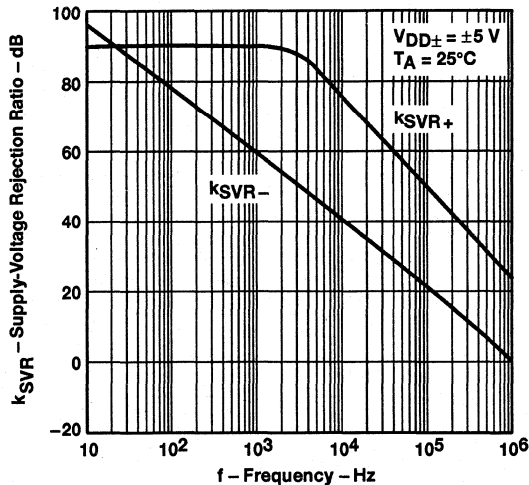
**Figure 29**

**SUPPLY-VOLTAGE REJECTION RATIO‡**  
**vs**  
**FREQUENCY**



**Figure 30**

**SUPPLY-VOLTAGE REJECTION RATIO‡**  
**vs**  
**FREQUENCY**



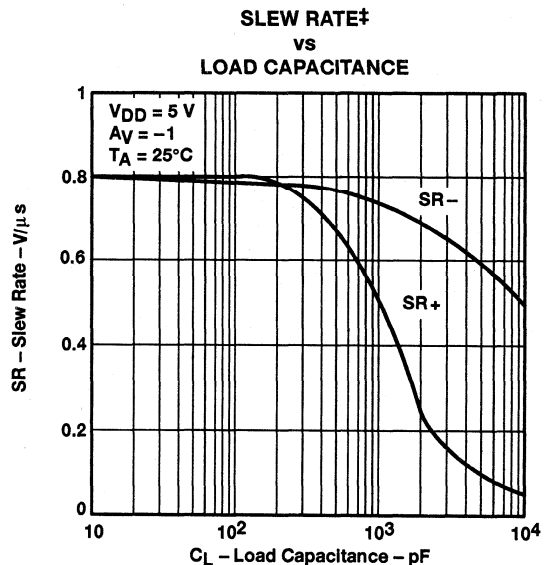
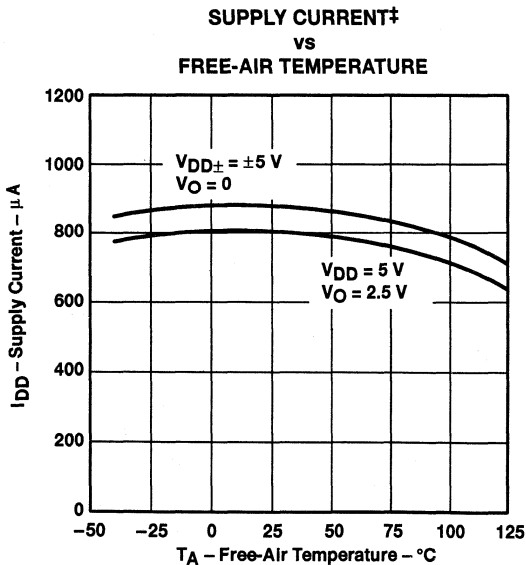
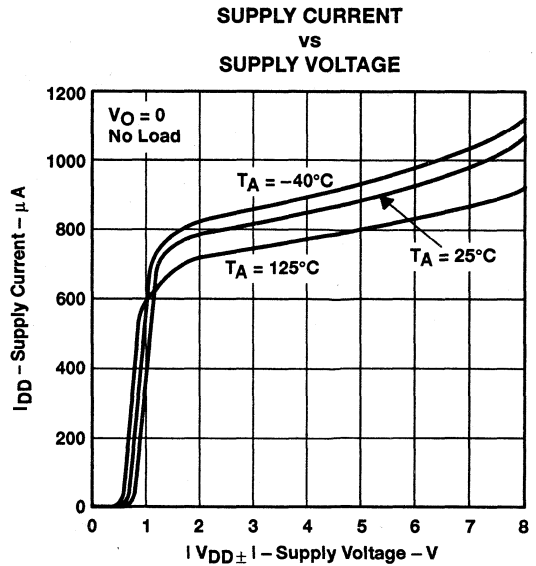
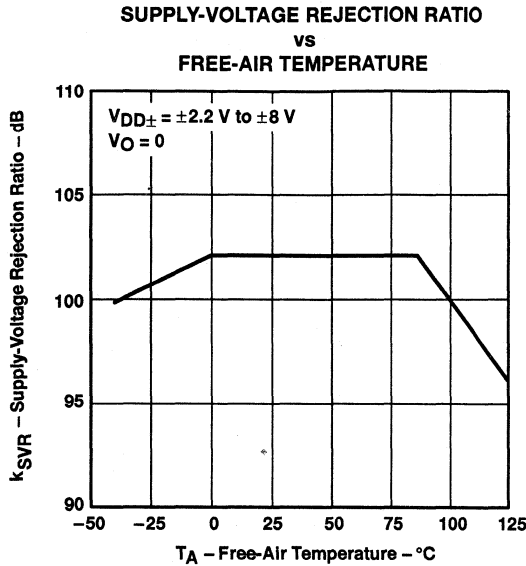
**Figure 31**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.



TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

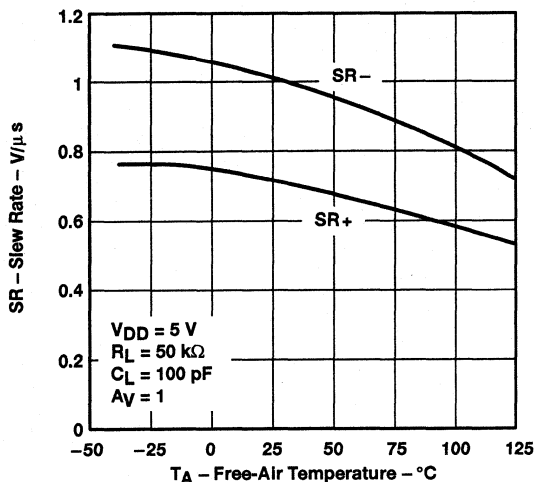
‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to  $2.5\text{ V}$ .

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

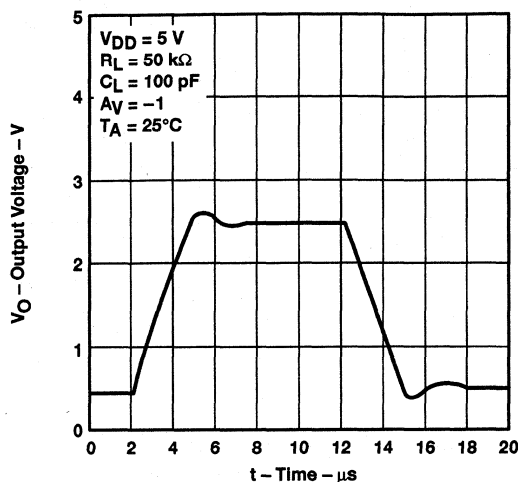
**TYPICAL CHARACTERISTICS†**

**SLEW RATE‡  
 vs  
 FREE-AIR TEMPERATURE**



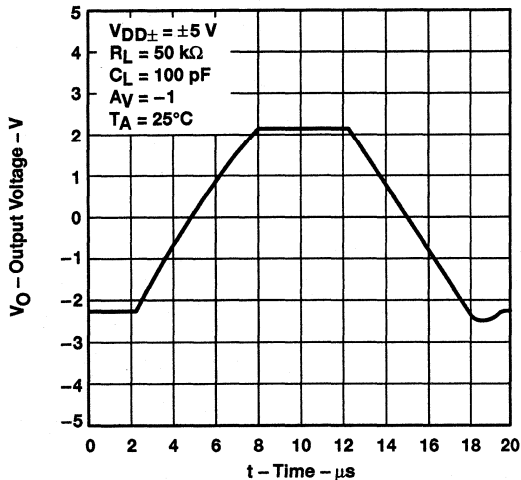
**Figure 36**

**INVERTING LARGE-SIGNAL PULSE  
 RESPONSE‡**



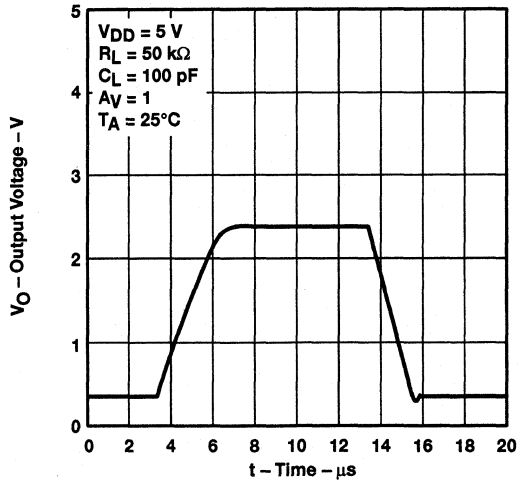
**Figure 37**

**INVERTING LARGE-SIGNAL PULSE  
 RESPONSE**



**Figure 38**

**VOLTAGE-FOLLOWER LARGE-SIGNAL  
 PULSE RESPONSE**



**Figure 39**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.



TYPICAL CHARACTERISTICS

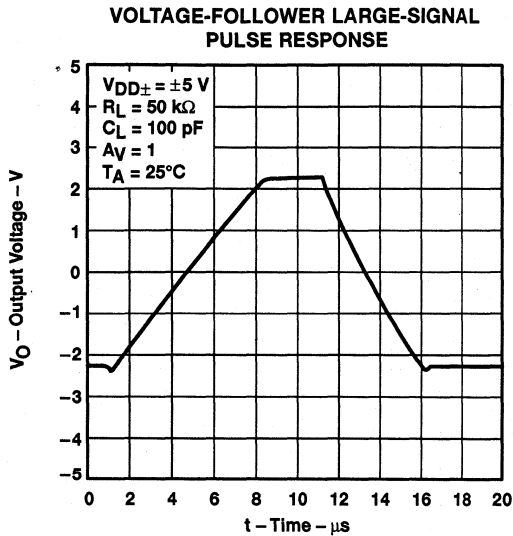


Figure 40

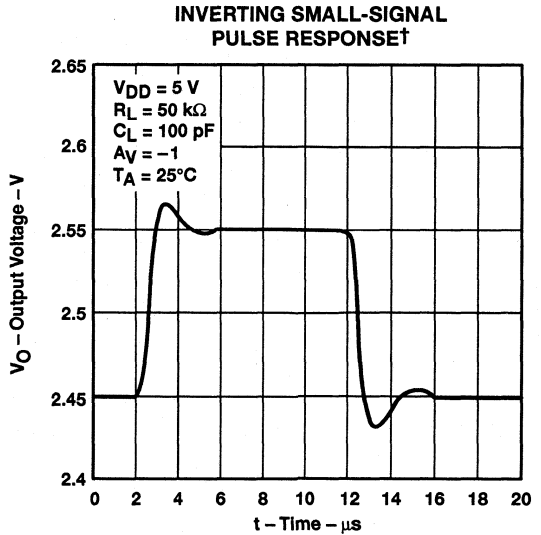


Figure 41

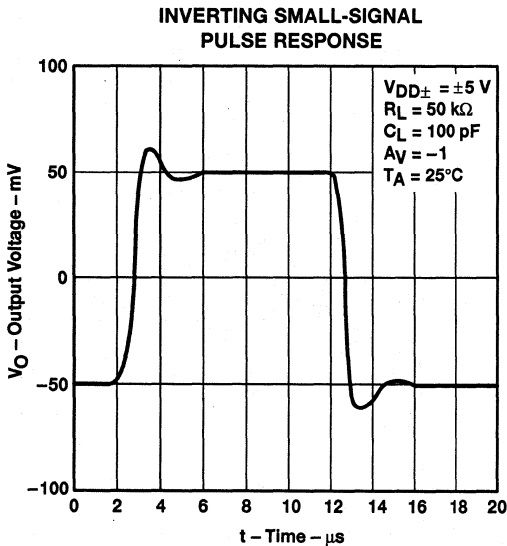


Figure 42

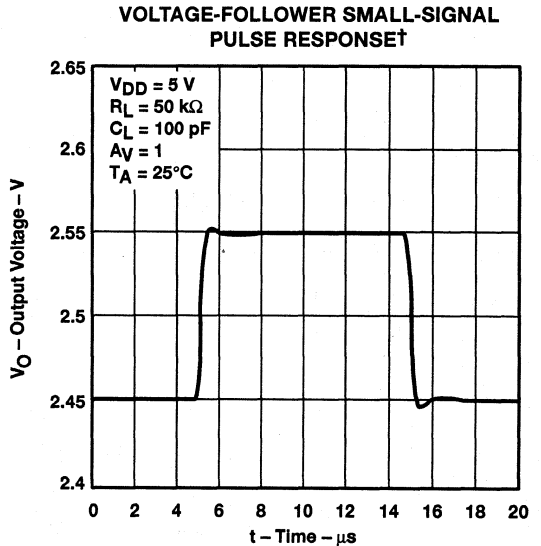


Figure 43

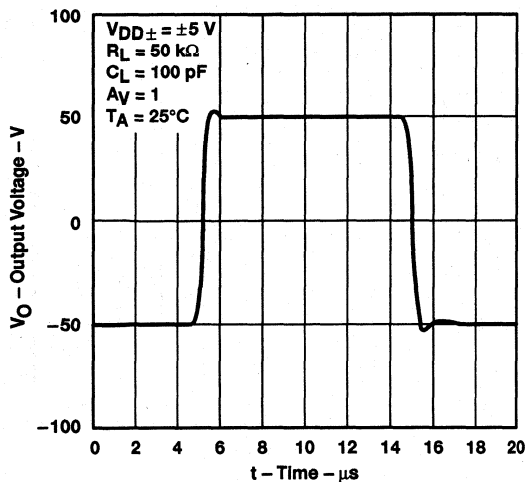
† For curves where  $V_{DD} = 5 \text{ V}$ , all loads are referenced to 2.5 V.

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

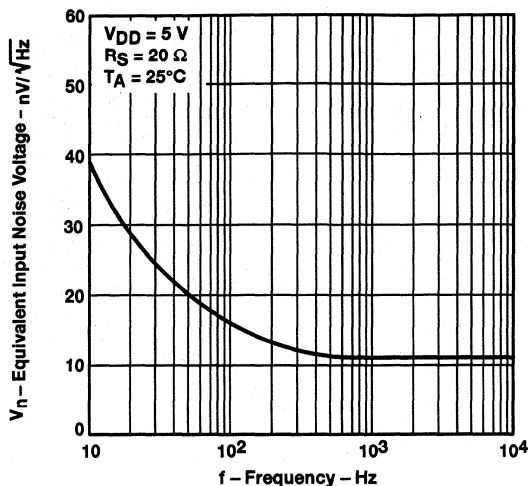
**TYPICAL CHARACTERISTICS**

**VOLTAGE-FOLLOWER SMALL-SIGNAL PULSE RESPONSE**



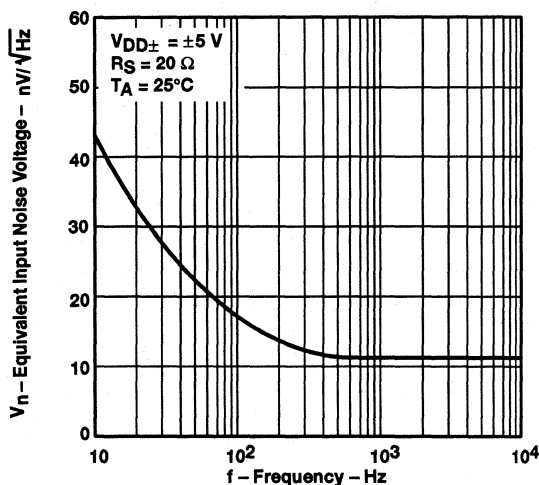
**Figure 44**

**EQUIVALENT INPUT NOISE VOLTAGE† vs FREQUENCY**



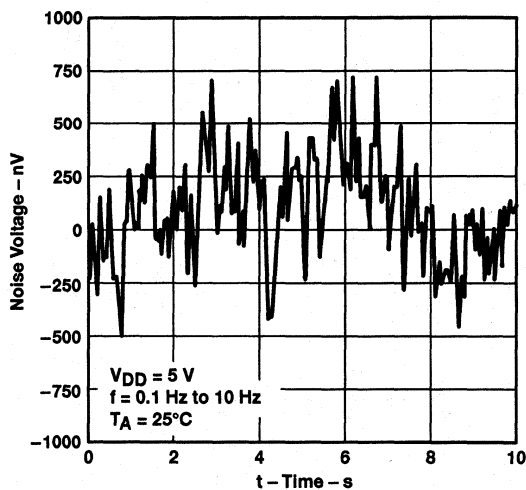
**Figure 45**

**EQUIVALENT INPUT NOISE VOLTAGE vs FREQUENCY**



**Figure 46**

**EQUIVALENT INPUT NOISE VOLTAGE OVER A 10-SECOND PERIOD†**



**Figure 47**

† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.



TYPICAL CHARACTERISTICS†

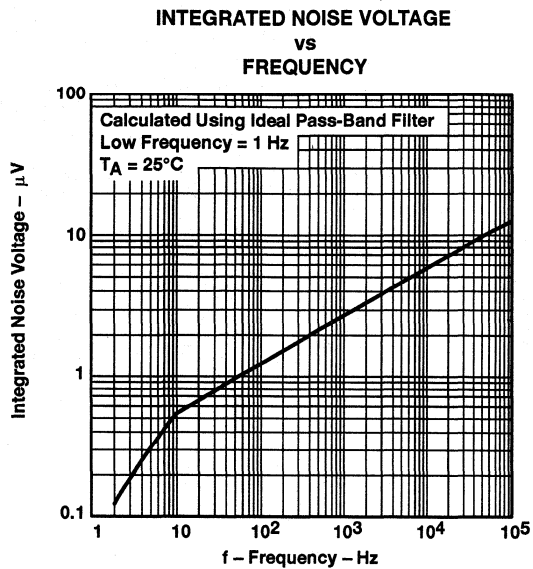


Figure 48

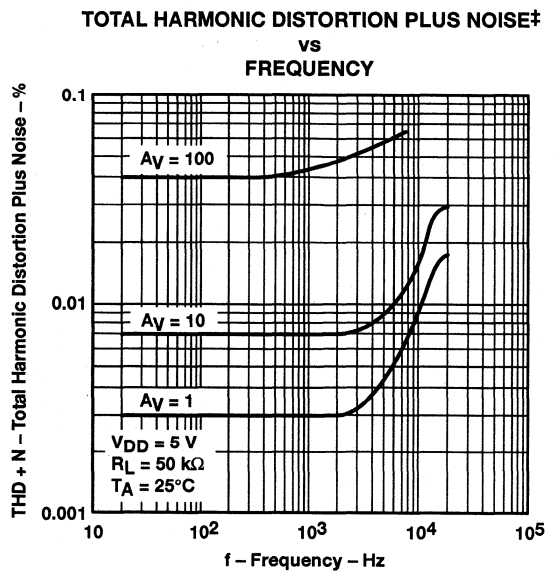


Figure 49

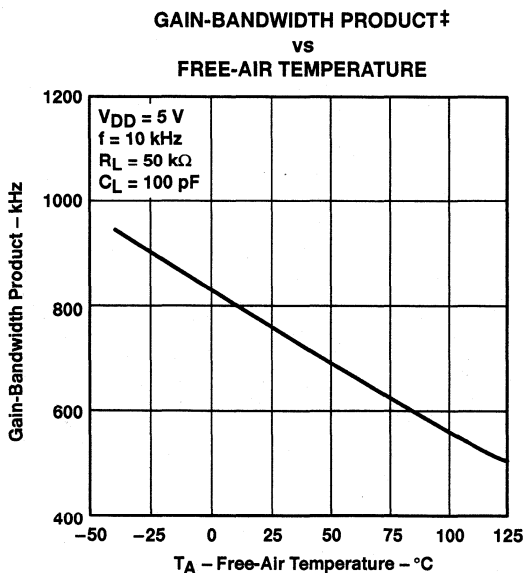


Figure 50

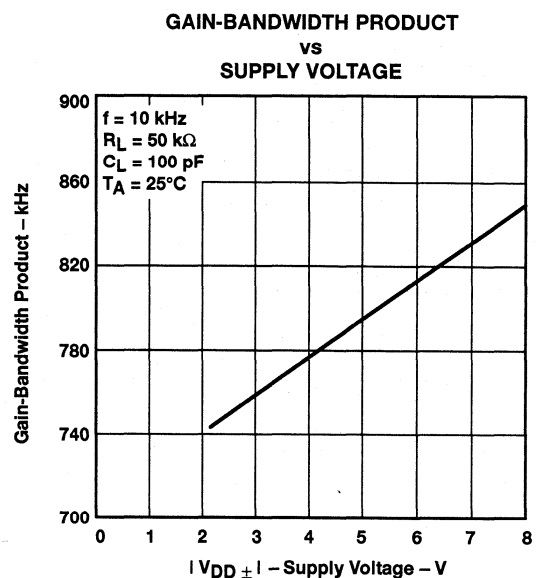
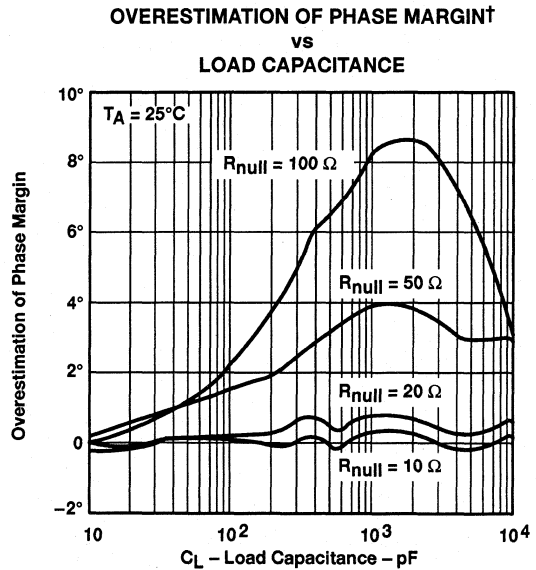
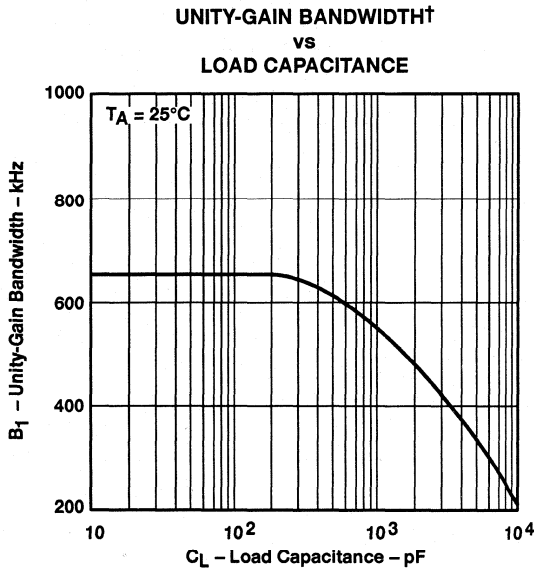
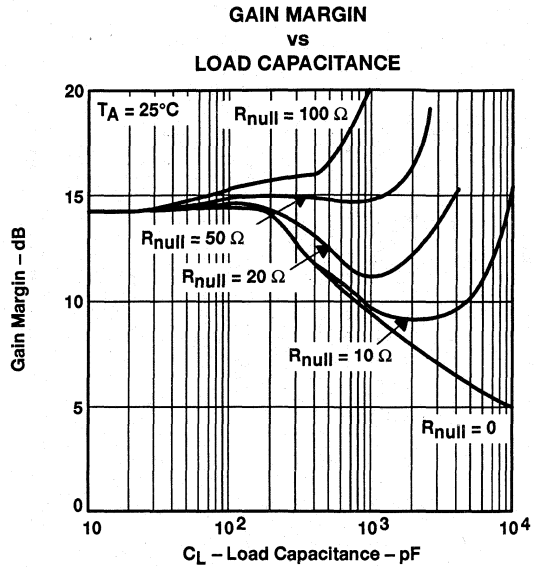
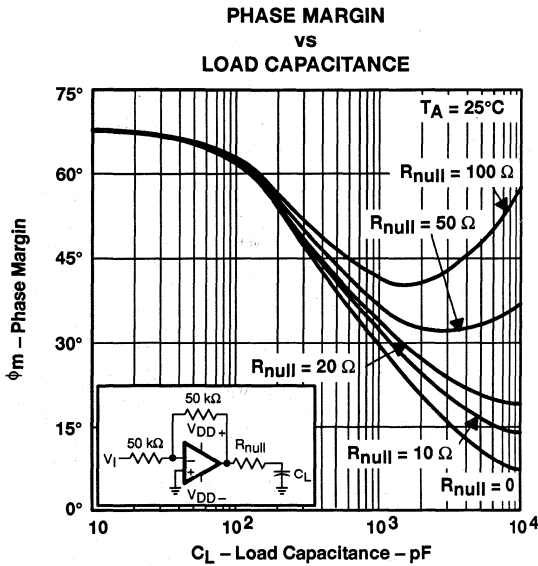


Figure 51

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS



† See application information



## APPLICATION INFORMATION

### loading considerations

The TLC2262 is a lower-power version of the TLC2272 with the appropriate design changes relative to the lower power level. The output drive performance to the negative rail for the TLC2262 is similar to the TLC2272 and is capable of driving several milliamperes.

The design topology used for the TLC2262 or the TLC2272 limits the drive to the positive rail to a value very close to the  $I_{DD}$  for the amplifier; thus, while the TLC2272 is capable of greater than 1-mA drive from the positive rail, the TLC2262 is capable of only a few 100 microamperes in proportion to the  $I_{DD}$  of the TLC2262. When designing with lower impedance loads (less than 50 k $\Omega$ ) with the TLC2262, the lower drive capability to the positive rail needs to be taken into consideration. Although the TLC2262 topology provides lower drive to the positive rail than other high-output-drive rail-to-rail operational amplifiers, it is a more stable topology.

### driving large capacitive loads

The TLC2262 is designed to drive larger capacitive loads than most CMOS operational amplifiers. Figure 52 and Figure 53 illustrate its ability to drive loads greater than 400 pF while maintaining good gain and phase margins ( $R_{null} = 0$ ).

A smaller series resistor ( $R_{null}$ ) at the output of the device (see Figure 56) improves the gain and phase margins when driving large capacitive loads. Figure 52 and Figure 53 show the effects of adding series resistances of 10  $\Omega$ , 20  $\Omega$ , 50  $\Omega$ , and 100  $\Omega$ . The addition of this series resistor has two effects: the first is that it adds a zero to the transfer function and the second is that it reduces the frequency of the pole associated with the output load in the transfer function.

The zero introduced to the transfer function is equal to the series resistance times the load capacitance. To calculate the improvement in phase margin, equation (1) can be used.

$$\Delta\theta_{m1} = \tan^{-1} \left( 2 \times \pi \times \text{UGBW} \times R_{null} \times C_L \right) \quad (1)$$

where :

- $\Delta\theta_{m1}$  = improvement in phase margin
- UGBW = unity-gain bandwidth frequency
- $R_{null}$  = output series resistance
- $C_L$  = load capacitance

The unity-gain bandwidth (UGBW) frequency decreases as the capacitive load increases (see Figure 54). To use equation (1), UGBW must be approximated from Figure 54.

Using equation (1) alone overestimates the improvement in phase margin, as illustrated in Figure 55. The overestimation is caused by the decrease in the frequency of the pole associated with the load, thus providing additional phase shift and reducing the overall improvement in phase margin. The pole associated with the load is reduced by the factor calculated in equation (2).

$$F = \frac{1}{1 + g_m \times R_{null}} \quad (2)$$

where :

- F = factor reducing frequency of pole
- $g_m$  = small-signal output transconductance (typically  $4.83 \times 10^{-3}$  mhos)
- $R_{null}$  = output series resistance

**TLC2262, TLC2262A, TLC2262Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

**APPLICATION INFORMATION**

**driving large capacitive loads (continued)**

For the TLC2262, the pole associated with the load is typically 6 MHz with 100-pF load capacitance. This value varies inversely with  $C_L$ : at  $C_L = 10$  pF, use 60 MHz, at  $C_L = 1000$  pF, use 600 kHz, and so on.

Reducing the pole associated with the load introduces phase shift, thereby reducing phase margin. This results in an error in the increase in phase margin expected by considering the zero alone [equation (1)]. Equation (3) approximates the reduction in phase margin due to the movement of the pole associated with the load. The result of this equation can be subtracted from the result of the equation in equation (1) to better approximate the improvement in phase margin.

$$\Delta\Theta_{m2} = \tan^{-1} \left[ \frac{UGBW}{(F \times P_2)} \right] - \tan^{-1} \left( \frac{UGBW}{P_2} \right) \quad (3)$$

where :

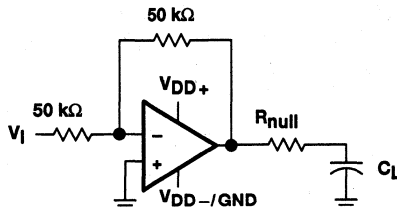
$\Delta\Theta_{m2}$  = reduction in phase margin

UGBW = unity-gain bandwidth frequency

F = factor from equation (2)

$P_2$  = unadjusted pole (60 MHz@ 10 pF, 6 MHz@ 100 pF, etc.)

Using these equations with Figure 54 and Figure 55 enables the designer to choose the appropriate output series resistance to optimize the design of circuits driving large capacitive loads.



**Figure 56. Series-Resistance Circuit**

# TLC2262, TLC2262A, TLC2262Y Advanced LinCMOS™ RAIL-TO-RAIL DUAL OPERATIONAL AMPLIFIERS

SLOS128A – AUGUST 1993 – REVISED APRIL 1994

## APPLICATION INFORMATION

### macromodel information

Macromodel information provided is derived using *PSpice™ Parts™* model generation software. The Boyle macromodel (see Note 5) and subcircuit in Figure 53 are generated using the TLC2262 typical electrical and operating characteristics at  $T_A = 25^\circ\text{C}$ . Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 5: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Intergrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).

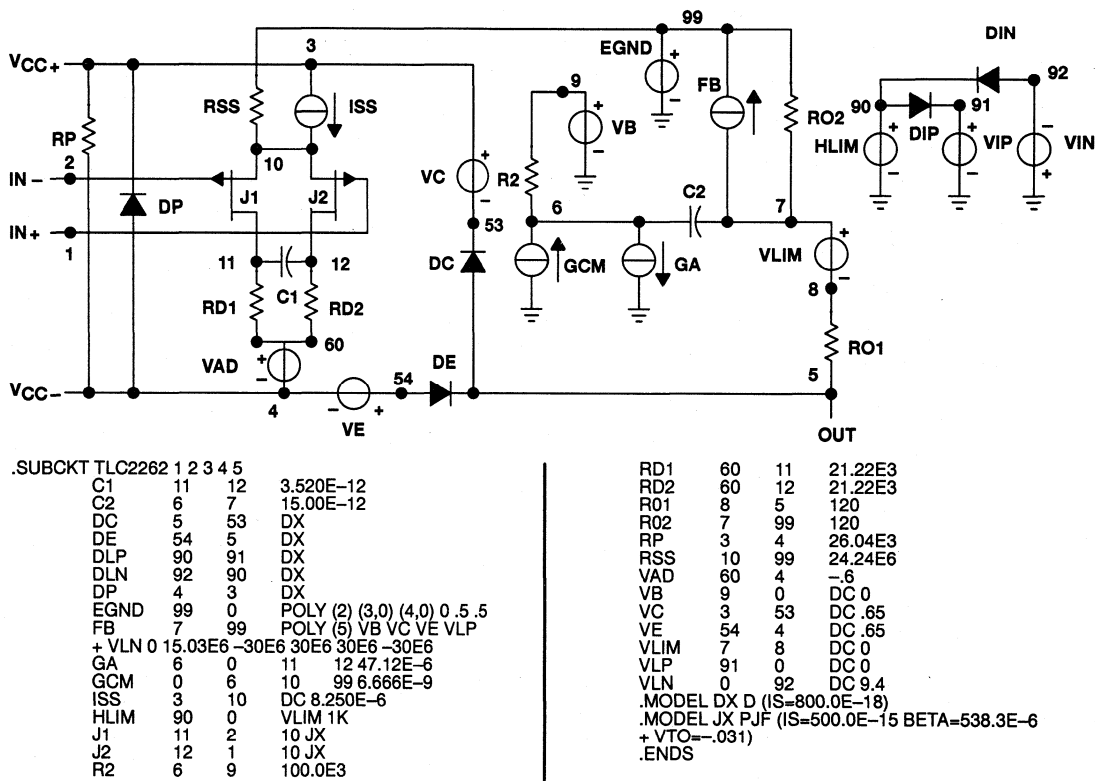


Figure 57. Boyle Macromodel and Subcircuit

*PSpice* and *Parts* are trademarks of MicroSim Corporation.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# TLC2264, TLC2264A, TLC2264Y Advanced LinCMOS™ RAIL-TO-RAIL QUAD OPERATIONAL AMPLIFIERS

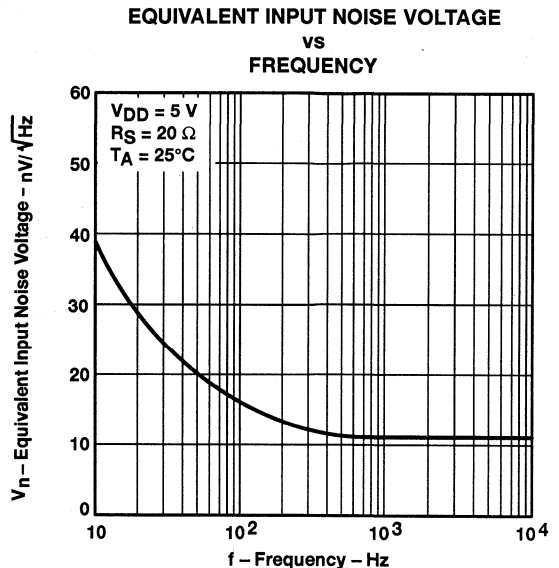
SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

- Output Swing Includes Both Supply Rails
- Low Noise . . . 12 nV/√Hz Typ at f = 1 kHz
- Low Input Bias Current . . . 1 pA Typ
- Fully Specified for Both Single-Supply and Split-Supply Operation
- Low Power . . . 1 mA Max
- Common-Mode Input Voltage Range Includes Negative Rail
- Low Input Offset Voltage  
950 μV Max at T<sub>A</sub> = 25°C (TLC2264A)
- Macromodel Included

## description

The TLC2264 and TLC2264A are quad operational amplifiers manufactured using Texas Instruments Advanced LinCMOS™ process. These devices exhibit rail-to-rail output performance while having better input offset voltage and lower power dissipation levels than existing CMOS operational amplifiers. In addition, the noise performance (see Figure 1) has been dramatically increased for this class of low-power CMOS amplifier. Figure 1 depicts the low level of voltage noise for this CMOS amplifier, which has only 200 μA (typical) of supply current per amplifier. Also, the common-mode input voltage range is wider than typical standard CMOS-type amplifiers. To take advantage of this improvement in performance and to make this device available for a wider range of applications, V<sub>ICR</sub> is specified with a larger maximum input offset voltage test limit of ±5 mV. The Advanced LinCMOS™ process uses a silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. This technology also makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The TLC2264 and TLC2264A, exhibiting high input impedance and low noise, are excellent for small-signal conditioning for high-impedance sources, such as piezoelectric transducers. Because of the low-power dissipation levels, these devices work well in hand-held monitoring and remote-sensing applications. In addition, the rail-to-rail output feature with single or split supplies makes these devices great choices when interfacing directly to analog-to-digital converters (ADCs). All of these features, combined with its temperature performance, make the TLC2264 family ideal for sonobuoys, remote pressure sensors, temperature control, active VR sensors, accelerometers, portable medical applications, hand-held metering, and many other applications.



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES			CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	2.5 mV	TLC2264CD	TLC2264CN	TLC2264CPWLE	TLC2264Y
-40°C to 125°C	950 μV 2.5 mV	TLC2264AID	TLC2264AIN	TLC2264AIPWLE	—
		TLC2264ID	TLC2264IN	—	

The D packages are available taped and reeled. Add R suffix to the device type (e.g., TLC2264CDR).  
The PW package is available only left-end taped and reeled. Chips are tested at 25°C.

Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA information is current as of publication date.  
Products conform to specifications per the terms of Texas Instruments  
standard warranty. Production processing does not necessarily include  
testing of all parameters.

**TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TLC2264, TLC2264A, TLC2264Y

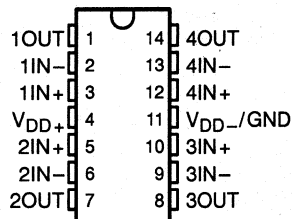
## Advanced LinCMOS™ RAIL-TO-RAIL QUAD OPERATIONAL AMPLIFIERS

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

### description (continued)

The device inputs and outputs are designed to withstand a 100-mA surge current without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Exercise care in handling these devices, as exposure to ESD may result in degradation of the device parametric performance. Additional care should be exercised to prevent  $V_{DD+}$  supply line transients under powered conditions. Transients of greater than 20 V can trigger the ESD-protection structure, inducing a low-impedance path to  $V_{DD-}/GND$ . Should this condition occur, the sustained current supplied to the device must be limited to 100 mA or less. Failure to do so could result in a latched condition and device failure.

D, N, OR PW PACKAGE  
(TOP VIEW)

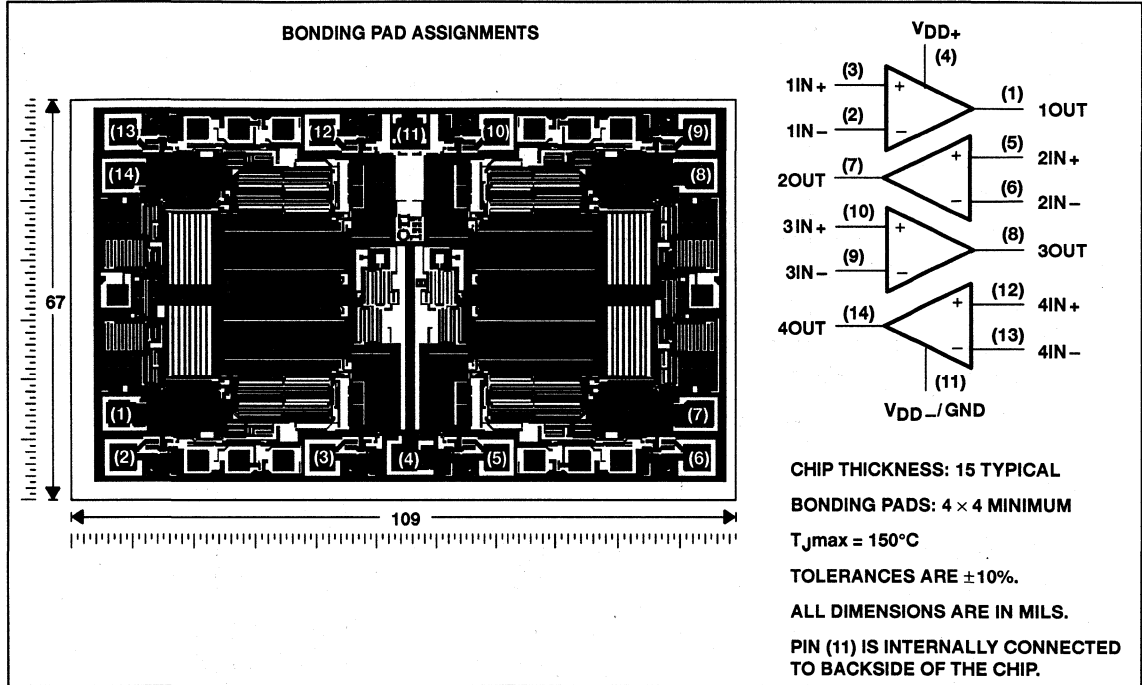


**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

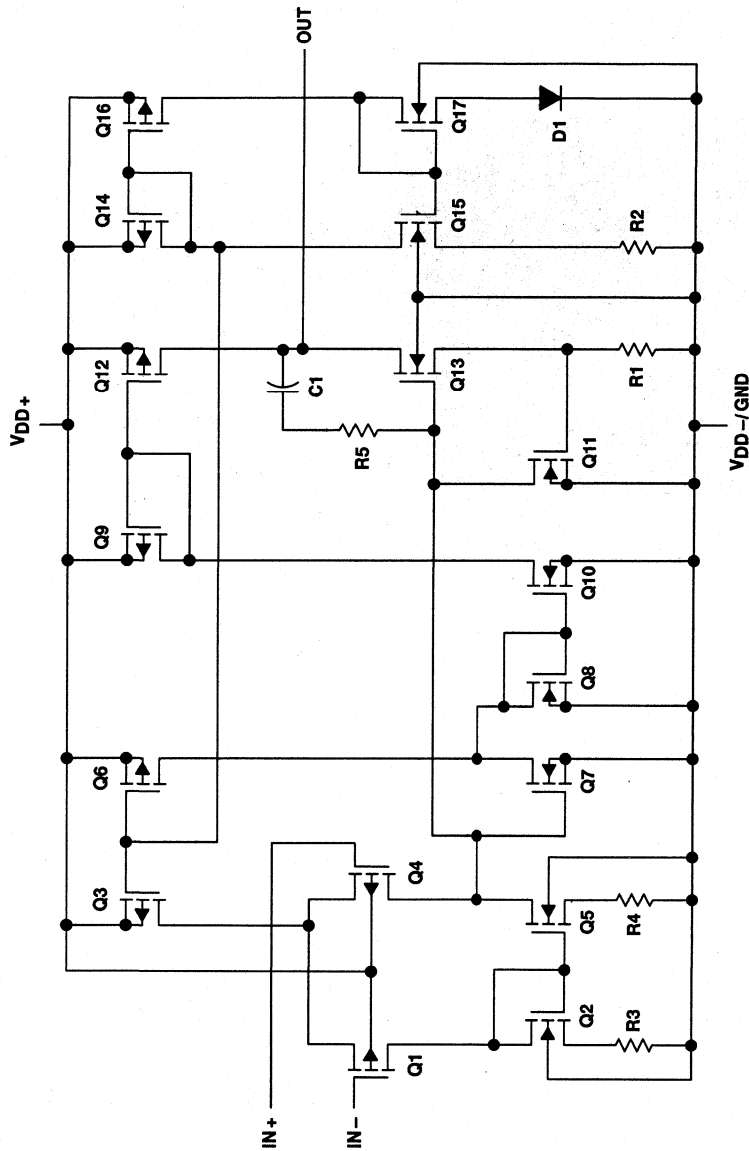
**TLC2264Y chip information**

This chip, when properly assembled, displays characteristics similar to the TLC2264C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. The chip may be mounted with conductive epoxy or a gold-silicon preform.



**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

equivalent schematic (each amplifier)



COMPONENT COUNT†	
Transistors	76
Diodes	18
Resistors	52
Capacitors	6

† Includes all amplifiers, ESD, bias, and trim circuitry



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	±16 V
Input voltage range, $V_I$ (any input, see Note 1)	$V_{DD-} - 0.3$ V to $V_{DD+}$
Input current, $I_I$ (each input)	±5 mA
Output current, $I_O$	±50 mA
Total current into $V_{DD+}$	±50 mA
Total current out of $V_{DD-}$	±50 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 125°C
Storage temperature range	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ . Excessive current flows if input is brought below  $V_{DD-} - 0.3$  V.  
 3. The output can be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	190 mW
N	1150 mW	9.2 mW/°C	736 mW	230 mW
PW	700 mW	5.6 mW/°C	448 mW	140 mW

**recommended operating conditions**

	C SUFFIX		I SUFFIX		UNIT
	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	±2.2	±8	±2.2	±8	V
Input voltage range, $V_I$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Operating free-air temperature, $T_A$	0	70	-40	125	°C

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $V_O = 0,$ $V_{DD} \pm = \pm 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C		300	2500	$\mu\text{V}$
		Full range			3000	
$\alpha_{V_{IO}}$ Temperature coefficient of input offset voltage		25°C to 70°C		2		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.003		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C		0.5		$\text{pA}$
		Full range			100	
$I_{IB}$ Input bias current		25°C		1		$\text{pA}$
		Full range			100	
$V_{ICR}$ Common-mode input voltage range		$R_S = 50\ \Omega,$ $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2	V
			Full range	0 to 3.5		
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C		4.99	V	
		Full range		4.85 4.94		
	$I_{OH} = -100\ \mu\text{A}$	25°C		4.82		
		Full range		4.70 4.85		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 50\ \mu\text{A}$	25°C		0.01	V	
		Full range		0.09 0.15		
	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 500\ \mu\text{A}$	25°C		0.15		
		Full range		0.2 0.3		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 1\text{ mA}$	25°C		0.3	V	
		Full range		0.7 1		
	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 4\text{ mA}$	25°C		1.2		
		Full range				
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V},$ $V_O = 1\text{ V to }4\text{ V}$	$R_L = 50\text{ k}\Omega^\ddagger$	25°C	80	170	V/mV
			Full range		55	
		$R_L = 1\text{ M}\Omega^\ddagger$	25°C		550	
$r_{id}$ Differential input resistance		25°C		$10^{12}$	$\Omega$	
$r_i$ Common-mode input resistance		25°C		$10^{12}$	$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz},$ N package	25°C		8	pF	
$z_o$ Closed-loop output impedance	$f = 100\text{ kHz},$ $A_V = 10$	25°C		240	$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V},$ $V_O = 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	70	83	dB	
		Full range		70		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V},$ $V_{IC} = V_{DD}/2,$ No load	25°C	80	95	dB	
		Full range		80		
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V},$ No load	25°C	0.8	1	mA	
		Full range		1		

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 1.4\text{ V to }2.6\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C	0.35	0.55		$\text{V}/\mu\text{s}$
		Full range	0.3			
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C	40		$\text{nV}/\sqrt{\text{Hz}}$	
	$f = 1\text{ kHz}$	25°C	12			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$	25°C	0.7		$\mu\text{V}$	
	$f = 0.1\text{ Hz to }10\text{ Hz}$	25°C	1.3			
$I_n$ Equivalent input noise current		25°C	0.6		$\text{fA}/\sqrt{\text{Hz}}$	
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}, f = 20\text{ kHz}, R_L = 50\text{ k}\Omega^\ddagger$	$A_V = 1$	0.017%			
		$A_V = 10$	0.03%			
Gain-bandwidth product	$f = 10\text{ kHz}, C_L = 100\text{ pF}^\ddagger, R_L = 50\text{ k}\Omega^\ddagger$	25°C	0.71		MHz	
BOM Maximum output-swing bandwidth	$V_{O(PP)} = 2\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, A_V = 1, C_L = 100\text{ pF}^\ddagger$	25°C	185		kHz	
$t_s$ Settling time	$A_V = -1, \text{Step} = 0.5\text{ V to }2.5\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	To 0.1%	6.4		$\mu\text{s}$	
		To 0.01%	14.1			
$\phi_m$ Phase margin at unity gain	$R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C	63°		dB	
		25°C	14			

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise specified)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264C			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	300	2500		$\mu\text{V}$
		Full range	3000			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 70°C	2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0,$ $R_S = 50\ \Omega$	25°C	0.003			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			$\text{pA}$
		Full range	100			
$I_{IB}$ Input bias current		25°C	1			$\text{pA}$
		Full range	100			
$V_{ICR}$ Common-mode input voltage range	$ V_{IO}  \leq 5\ \text{mV},$ $R_S = 50\ \Omega$	25°C	-5 to 4	-5.3 to 4.2		V
		Full range	-5 to 3.5			
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C	4.99			V
		25°C	4.85	4.94		
		Full range	4.82			
		25°C	4.7	4.85		
$V_{OM-}$ Maximum negative peak output voltage	$I_O = -200\ \mu\text{A}$	25°C	-4.99			V
		25°C	-4.85	-4.91		
		Full range	-4.85			
		25°C	-4.7	-4.8		
$V_{OM-}$ Maximum negative peak output voltage	$I_O = 1\ \text{mA}$	25°C	-4.7			V
		25°C	-4	-4.3		
		Full range	-3.8			
		25°C	-4	-4.3		
$V_{OM-}$ Maximum negative peak output voltage	$I_O = 4\ \text{mA}$	25°C	-4.99			V
		25°C	-4.85	-4.91		
		Full range	-4.85			
		25°C	-4.7	-4.8		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 50\ \text{k}\Omega$	25°C	80	200	$\text{V}/\text{mV}$
			Full range	55		
		$R_L = 1\ \text{M}\Omega$	25°C	1000		
$r_{id}$ Differential input resistance		25°C	$10^{12}$		$\Omega$	
$r_i$ Common-mode input resistance		25°C	$10^{12}$		$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz},$ N package	25°C	8		$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz},$ $A_V = 10$	25°C	220		$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega$	25°C	75	88		dB
		Full range	75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V},$ $V_{IC} = 0,$ No load	25°C	80	95		dB
		Full range	80			
$I_{DD}$ Supply current (four amplifiers)	$V_O = 0,$ No load	25°C	0.85	1		mA
		Full range	1			

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$ †	TLC2264C			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = \pm 1.9\text{ V}$ , $C_L = 100\text{ pF}$	$R_L = 50\text{ k}\Omega$	25°C	0.35	0.55	V/ $\mu\text{s}$	
			Full range	0.3			
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$		25°C	43		nV/ $\sqrt{\text{Hz}}$	
	$f = 1\text{ kHz}$		25°C	12			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$		25°C	0.8		$\mu\text{V}$	
	$f = 0.1\text{ Hz to }10\text{ Hz}$		25°C	1.3			
$I_n$ Equivalent input noise current			25°C	0.6		fA/ $\sqrt{\text{Hz}}$	
THD + N Total harmonic distortion pulse duration	$V_O = \pm 2.3\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 50\text{ k}\Omega$	$A_V = 1$	25°C	0.014%			
		$A_V = 10$		0.024%			
Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$	$R_L = 50\text{ k}\Omega$	25°C	0.73		MHz	
BOM Maximum output-swing bandwidth	$V_{O(PP)} = 4.6\text{ V}$ , $R_L = 50\text{ k}\Omega$	$A_V = 1$ , $C_L = 100\text{ pF}$	25°C	70		kHz	
$t_s$ Settling time	$A_V = -1$ , Step = $-2.3\text{ V to }2.3\text{ V}$ , $R_L = 50\text{ k}\Omega$ , $C_L = 100\text{ pF}$	To 0.1%	25°C	7.1		$\mu\text{s}$	
		To 0.01%		16.5			
$\phi_m$ Phase margin at unity gain	$R_L = 50\text{ k}\Omega$ , $C_L = 100\text{ pF}$		25°C	64°			
		Gain margin		25°C	14		dB

† Full range is 0°C to 70°C.

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264I			TLC2264AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$
		Full range			3000			1500	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 125°C		2			2		$\mu\text{V}/^\circ\text{C}$
		25°C		0.003			0.003		
Input offset voltage long-term drift (see Note 4)	$V_{DD} \pm \pm 2.5\text{ V}$ , $V_{IC} = 0$ , $V_O = 0$ , $R_S = 50\ \Omega$	25°C		0.5			0.5		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C		1		1			
$I_{IB}$ Input bias current		25°C		1		1			
		Full range		500		500			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2	V	
		Full range	0 to 3.5			0 to 3.5			
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$ $I_{OH} = -100\ \mu\text{A}$ $I_{OH} = -200\ \mu\text{A}$	25°C		4.99			4.99	V	
		25°C	4.85	4.94		4.85	4.94		
		Full range	4.82			4.82			
		25°C	4.7	4.85		4.7	4.85		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$ $V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$ $V_{IC} = 2.5\text{ V}$ , $I_{OL} = 4\text{ mA}$	25°C		0.01			0.01	V	
		25°C		0.09	0.15		0.09		0.15
		Full range			0.15				0.15
		25°C		0.8	1		0.7		1
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 50\text{ k}\Omega$ ‡	25°C	80	100		80	170	V/mV
			Full range	50			50		
		$R_L = 1\text{ M}\Omega$ ‡	25°C		550			550	
$r_{id}$ Differential input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$r_i$ Common-mode input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$ , N package	25°C		8			8	pF	
$z_o$ Closed-loop output impedance	$f = 100\text{ kHz}$ , $A_V = 10$	25°C		240			240	$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$	25°C		70	83		70	83	dB
		Full range		70			70		
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load	25°C		80	95		80	95	dB
		Full range		80			80		
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load	25°C		0.8	1		0.8	1	mA
		Full range			1			1	

† Full range is -40°C to 125°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264I			TLC2264AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = 1.4\text{ V to }2.6\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C	0.35	0.55		0.35	0.55		V/ $\mu\text{s}$
		Full range	0.25			0.25			
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$	40			40			nV/ $\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$	12			12			
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$	0.7			0.7			$\mu\text{V}$
		$f = 0.1\text{ Hz to }10\text{ Hz}$	1.3			1.3			
$I_n$	Equivalent input noise current	25°C	0.6			0.6			fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}, f = 20\text{ kHz}, R_L = 50\text{ k}\Omega^\ddagger$	$A_V = 1$	0.017%			0.017%			
		$A_V = 10$	0.03%			0.03%			
	Gain-bandwidth product $f = 50\text{ kHz}, C_L = 100\text{ pF}^\ddagger, R_L = 50\text{ k}\Omega^\ddagger$	25°C	0.71			0.71			MHz
BOM	Maximum output-swing bandwidth $V_{O(PP)} = 2\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, A_V = 1, C_L = 100\text{ pF}^\ddagger$	25°C	185			185			kHz
$t_s$	Settling time $A_V = -1, \text{ Step} = 0.5\text{ V to }2.5\text{ V}, R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	To 0.1%	6.4			6.4			$\mu\text{s}$
		To 0.01%	14.1			14.1			
$\phi_m$	Phase margin at unity gain $R_L = 50\text{ k}\Omega^\ddagger, C_L = 100\text{ pF}^\ddagger$	25°C	63°			63°			
		25°C	14			14			

† Full range is  $-40^\circ\text{C}$  to  $125^\circ\text{C}$ .

‡ Referenced to 2.5 V

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2264I			TLC2264AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$
		Full range			3000		1500		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 125°C		2			2	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C		0.003			0.003	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5			0.5	$\text{pA}$	
		Full range			500		500		
$I_{IB}$ Input bias current		25°C		1			1	$\text{pA}$	
		Full range			500		500		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	25°C	-5 to 4	-5.3 to 4.2		-5 to 4	-5.3 to 4.2	V	
		Full range	-5 to 3.5			-5 to 3.5			
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C		4.99			4.99	V	
	$I_O = -100\ \mu\text{A}$	25°C	4.85	4.94		4.85	4.94		
		Full range	4.82			4.82			
	$I_O = -200\ \mu\text{A}$	25°C	4.7	4.85		4.7	4.85		
		Full range	4.5			4.5			
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0, I_O = 50\ \mu\text{A}$	25°C		-4.99			-4.99	V	
	$V_{IC} = 0, I_O = 500\ \mu\text{A}$	25°C	-4.85	-4.91		-4.85	-4.91		
		Full range	-4.85			-4.85			
	$V_{IC} = 0, I_O = 4\ \text{mA}$	25°C	-4	-4.3		-4	-4.3		
		Full range	-3.8			-3.8			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 50\ \text{k}\Omega$	25°C	80	200		80	200	V/mV
			Full range	50			50		
		$R_L = 1\ \text{M}\Omega$	25°C		1000			1000	
$r_{id}$ Differential input resistance		25°C		$10^{12}$			$10^{12}$	$\Omega$	
$r_i$ Common-mode input resistance		25°C		$10^{12}$			$10^{12}$	$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, N$ package	25°C		8			8	$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz}, A_V = 10$	25°C		220			220	$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	75	88		75	88	dB	
		Full range	75			75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V}, V_{IC} = V_{DD\pm}/2, \text{ No load}$	25°C	80	95		80	95	dB	
		Full range	80			80			
$I_{DD}$ Supply current (four amplifiers)	$V_O = 0, \text{ No load}$	25°C		0.85	1		0.85	1	mA
		Full range			1		1		

† Full range is -40°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 500 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS		$T_A$ †	TLC2264I			TLC2264AI			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = \pm 1.9\text{ V}$ , $C_L = 100\text{ pF}$	$R_L = 50\text{ k}\Omega$	25°C	0.35	0.55		0.35	0.55	V/ $\mu\text{s}$
				Full range	0.25		0.25			
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$ $f = 1\text{ kHz}$		25°C	43		43		nV/ $\sqrt{\text{Hz}}$	
				25°C	12		12			
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$		25°C	0.8		0.8		$\mu\text{V}$	
				25°C	1.3		1.3			
$I_n$	Equivalent input noise current			25°C	0.6		0.6		fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise	$V_O = \pm 2.3\text{ V}$ , $R_L = 50\text{ k}\Omega$ , $f = 20\text{ kHz}$	$A_V = 1$ $A_V = 10$	25°C	0.014%		0.014%			
					0.024%		0.024%			
	Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$	$R_L = 50\text{ k}\Omega$	25°C	0.73		0.73		MHz	
BOM	Maximum output-swing bandwidth	$V_{O(PP)} = 4.6\text{ V}$ , $R_L = 50\text{ k}\Omega$	$A_V = 1$ , $C_L = 100\text{ pF}$	25°C	70		70		kHz	
$t_s$	Settling time	$A_V = -1$ , Step = $-2.3\text{ V to }2.3\text{ V}$ , $R_L = 50\text{ k}\Omega$ , $C_L = 100\text{ pF}$	$T_o = 0.1\%$ $T_o = 0.01\%$	25°C	7.1		7.1		$\mu\text{s}$	
					16.5		16.5			
$\phi_m$	Phase margin at unity gain	$R_L = 50\text{ k}\Omega$	$C_L = 100\text{ pF}$	25°C	64°		64°			
	Gain margin			25°C	14		14			dB

† Full range is  $-40^\circ\text{C}$  to  $125^\circ\text{C}$ .

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2264Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $V_O = 0$ , $V_{DD} = \pm 2.5\text{ V}$ , $R_S = 50\ \Omega$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$ V_{IO}  \leq 5\text{ mV}$ , $R_S = 50\ \Omega$	0 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$		4.99		V
	$I_{OH} = -100\ \mu\text{A}$		4.85	4.94	
	$I_{OH} = -200\ \mu\text{A}$		4.7	4.85	
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$		0.01		V
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$		0.09	0.15	
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 4\text{ mA}$		0.8	1	
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 50\text{ k}\Omega^\dagger$	80	170	V/mV
		$R_L = 1\text{ M}\Omega^\dagger$		550	
$r_{id}$ Differential input resistance			$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$		8		$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 100\text{ kHz}$ , $A_V = 10$		240		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$		70	83	$\text{dB}$
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load		80	95	$\text{dB}$
$I_{DD}$ Supply current (four amplifiers)	$V_O = 2.5\text{ V}$ , No load		0.8	1	$\text{mA}$

† Referenced to 2.5 V

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2264Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $V_O = 0$ $R_S = 50\ \Omega,$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$ V_{IO}  \leq 5\text{ mV},$ $R_S = 50\ \Omega$	-5 to 4	-5.3 to 4.2		V
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$		4.99		V
	$I_O = -100\ \mu\text{A}$	4.85	4.94		
	$I_O = -200\ \mu\text{A}$	4.7	4.85		
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0,$ $I_{OL} = 50\ \mu\text{A}$		-4.99		V
	$V_{IC} = 0,$ $I_{OL} = 500\ \mu\text{A}$	-4.85	-4.91		
	$V_{IC} = 0,$ $I_{OL} = 4\ \text{mA}$	-3.8	-4.1		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 50\ \text{k}\Omega$	80	200	V/mV
		$R_L = 1\ \text{M}\Omega$		1000	
$r_{id}$ Differential input resistance			10 <sup>12</sup>		$\Omega$
$r_i$ Common-mode input resistance			10 <sup>12</sup>		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}$		8		$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 100\ \text{kHz},$ $A_V = 10$		220		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V},$ $V_O = 0,$ $R_S = 50\ \Omega$	75	88		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V},$ $V_{IC} = 0,$ No load	80	95		dB
$I_{DD}$ Supply current (four amplifiers)	$V_O = 0,$ No load	0.85	1		$\text{mA}$

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution vs Common-mode input voltage	2,3 4,5
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	6,7
$I_{IB}/I_{IO}$	Input bias and input offset currents	vs Free-air temperature	8
$V_I$	Input voltage range	vs Supply voltage vs Free-air temperature	9 10
$V_{OH}$	High-level output voltage	vs High-level output current	11
$V_{OL}$	Low-level output voltage	vs Low-level output current	12,13
$V_{OM+}$	Maximum positive peak output voltage	vs Output current	14
$V_{OM-}$	Maximum negative peak output voltage	vs Output current	15
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	16
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Free-air temperature	17 18
$V_O$	Output voltage	vs Differential input voltage	19,20
	Differential gain	vs Load resistance	21
$A_{VD}$	Large-signal differential voltage amplification	vs Frequency vs Free-air temperature	22, 23 24, 25
$z_o$	Output impedance	vs Frequency	26, 27
$CMRR$	Common-mode rejection ratio	vs Frequency vs Free-air temperature	28 29
$k_{SVR}$	Supply-voltage rejection ratio	vs Frequency vs Free-air temperature	30, 31 32
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	33 34
$SR$	Slew rate	vs Load capacitance vs Free-air temperature	35 36
$V_O$	Large-signal pulse response	vs Time	37, 38, 39, 40
$V_O$	Small-signal pulse response	vs Time	41, 42, 43, 44
$V_n$	Equivalent input noise voltage	vs Frequency	45, 46
	Input noise voltage	Over a 10-second period	47
	Integrated noise voltage	vs Frequency	48
$THD + N$	Total harmonic distortion plus noise	vs Frequency	49
	Gain-bandwidth product	vs Free-air temperature vs Supply voltage	50 51
$\phi_m$	Phase margin	vs Frequency vs Load capacitance	22, 23 52
$A_m$	Gain margin	vs Load capacitance	53
$B_1$	Unity-gain bandwidth	vs Load capacitance	54
	Overestimation of phase margin	vs Load capacitance	55



TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC2264  
 INPUT OFFSET VOLTAGE

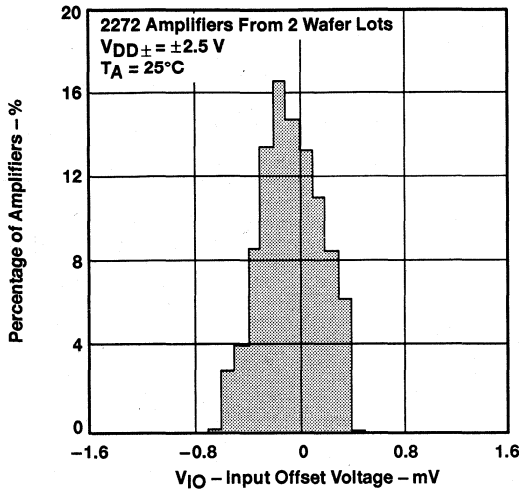


Figure 2

DISTRIBUTION OF TLC2264  
 INPUT OFFSET VOLTAGE

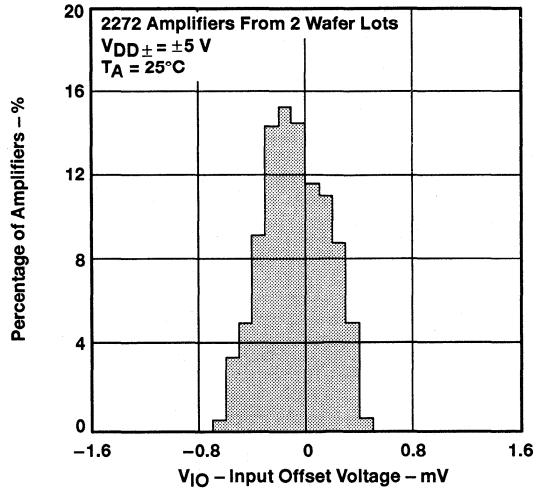


Figure 3

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

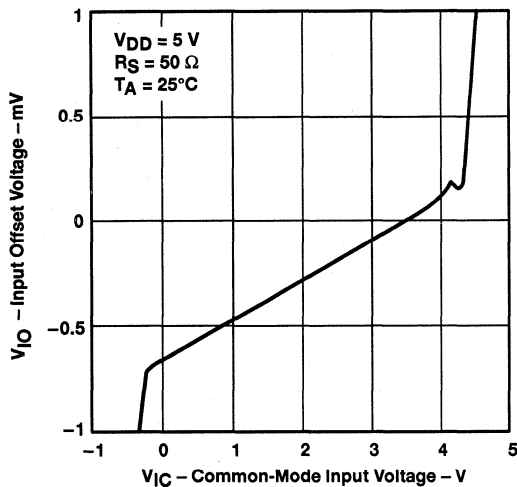


Figure 4

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE INPUT VOLTAGE

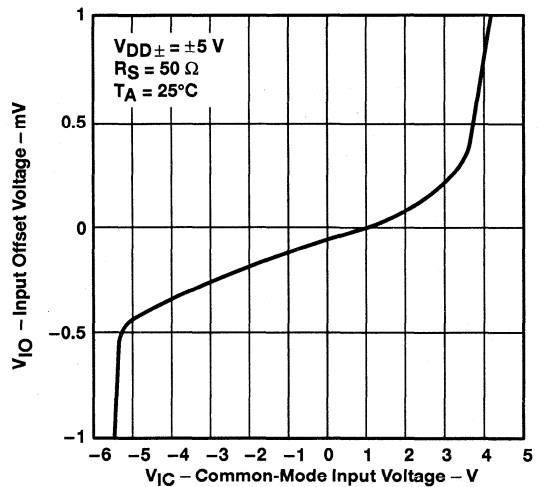


Figure 5

† For curves where  $V_{DD} = 5 \text{ V}$ , all loads are referenced to 2.5 V.

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS†**

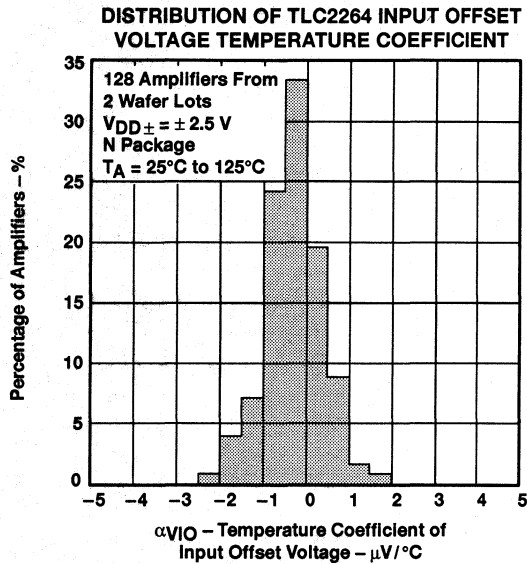


Figure 6

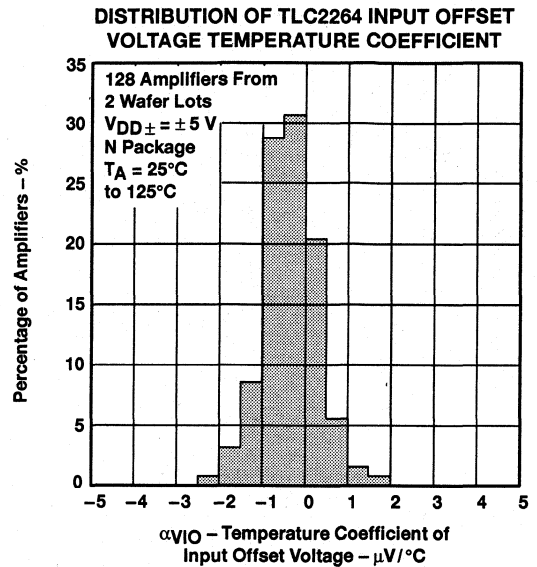


Figure 7

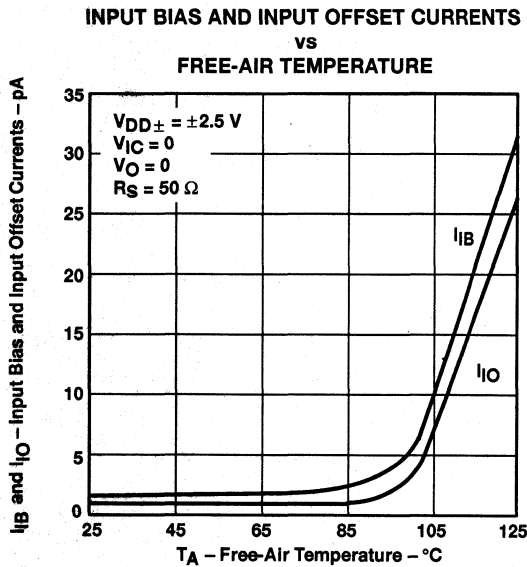


Figure 8

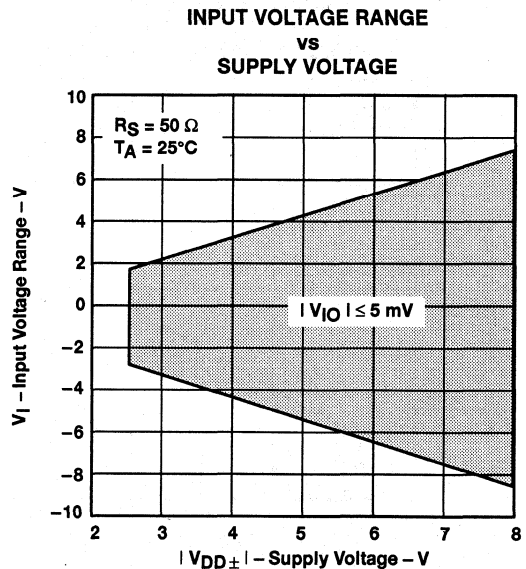


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

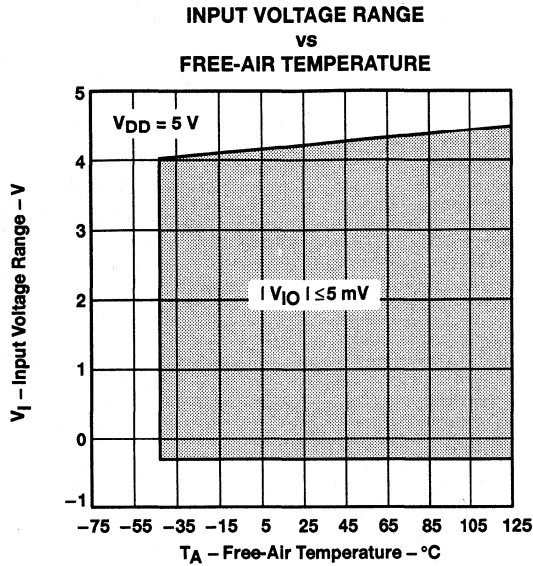


Figure 10

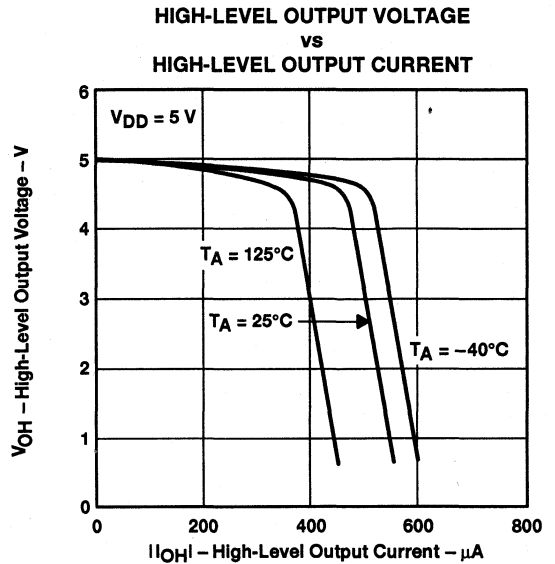


Figure 11

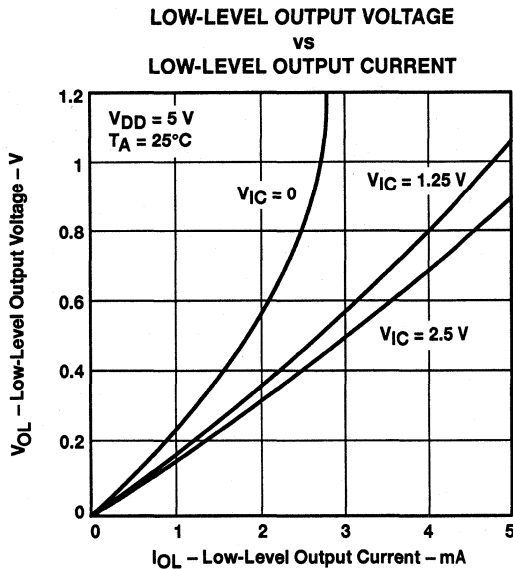


Figure 12

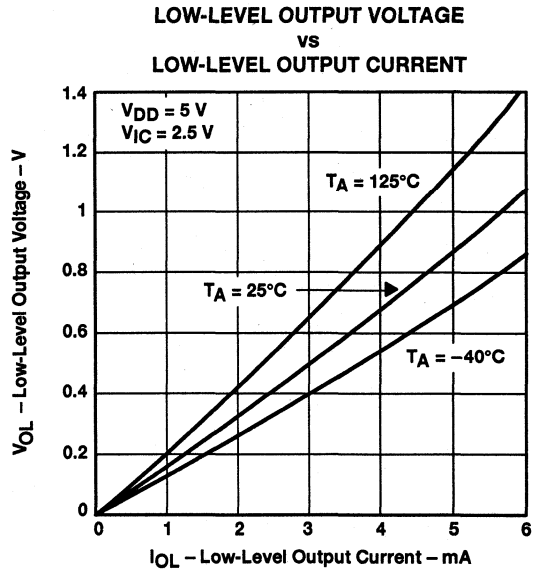


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS

MAXIMUM POSITIVE OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

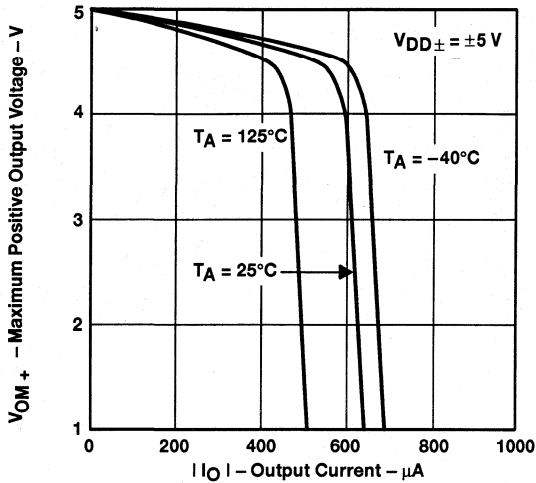


Figure 14

MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

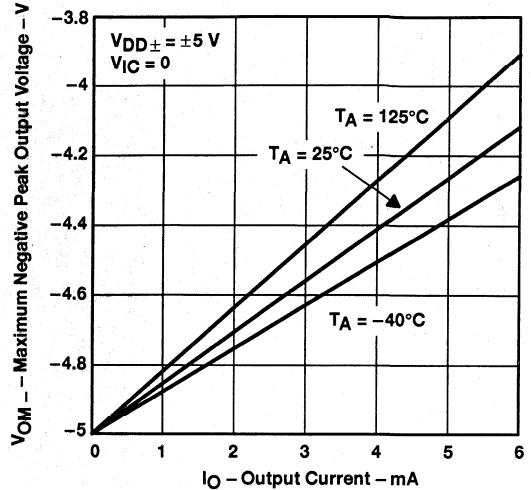
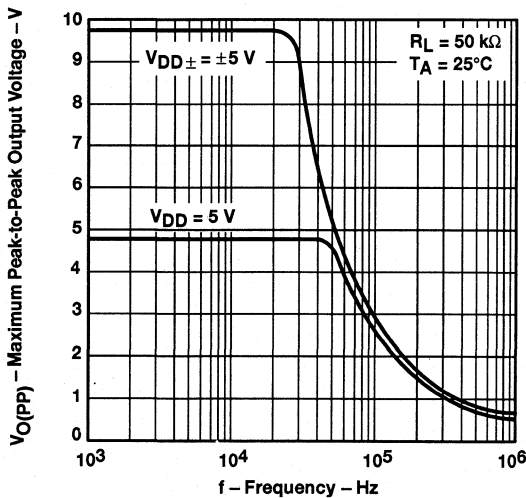


Figure 15

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE†  
 vs  
 FREQUENCY



† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

Figure 16

SHORT-CIRCUIT OUTPUT CURRENT  
 vs  
 SUPPLY VOLTAGE

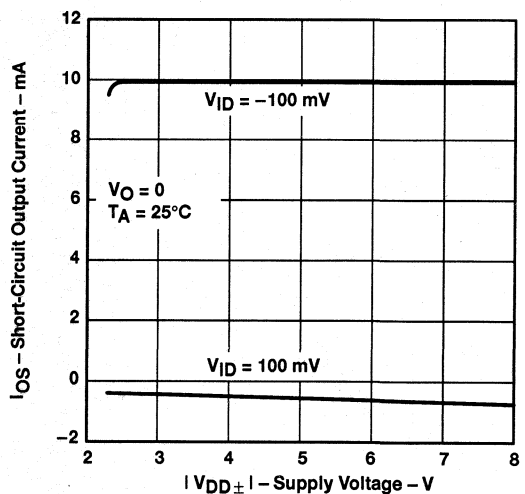


Figure 17



TYPICAL CHARACTERISTICS†

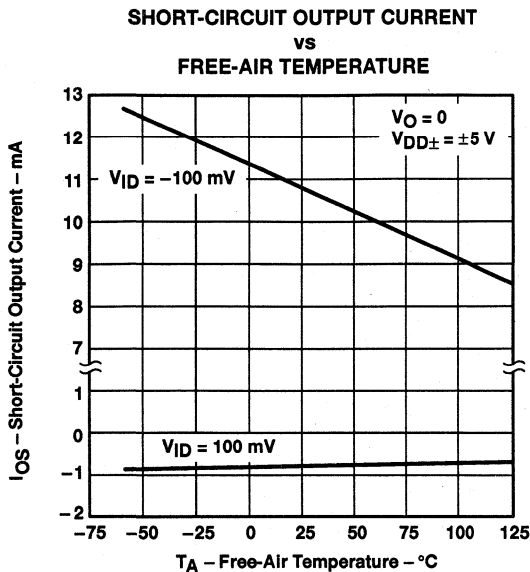


Figure 18

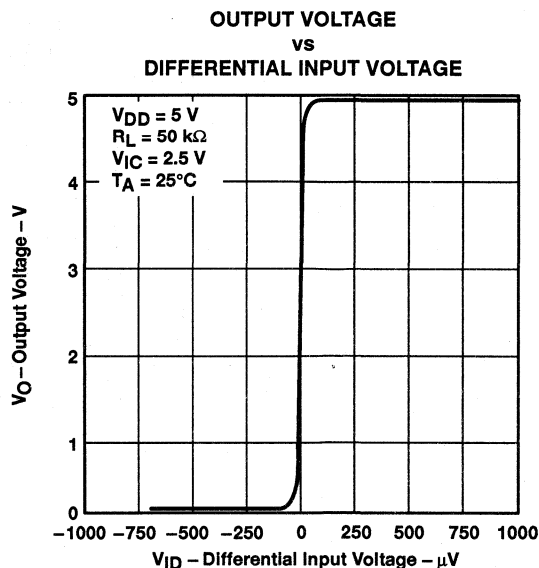


Figure 19

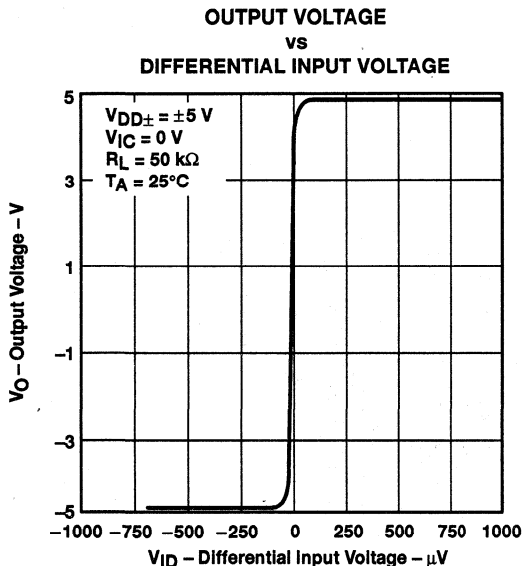


Figure 20

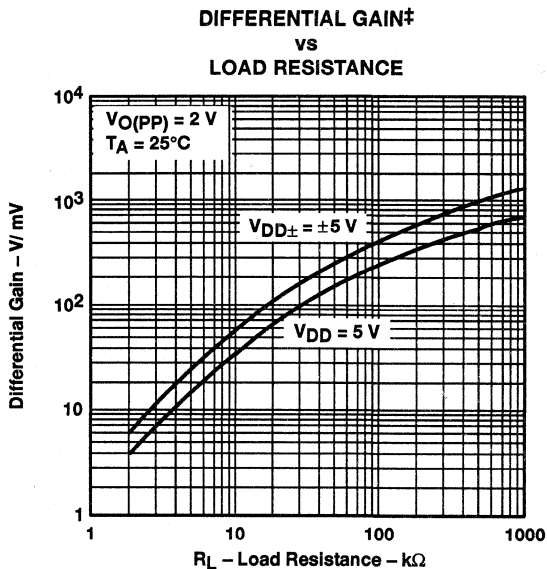


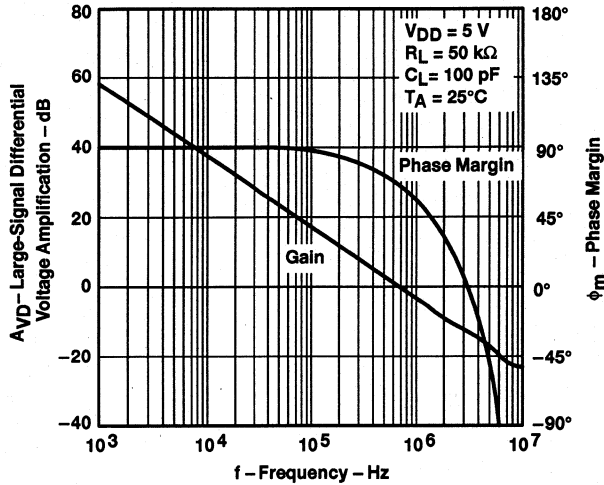
Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5 \text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE†  
 AMPLIFICATION AND PHASE MARGIN  
 vs  
 FREQUENCY



† For curves where V<sub>DD</sub> = 5 V, all loads are referenced to 2.5 V.

Figure 22

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE MARGIN  
 vs  
 FREQUENCY

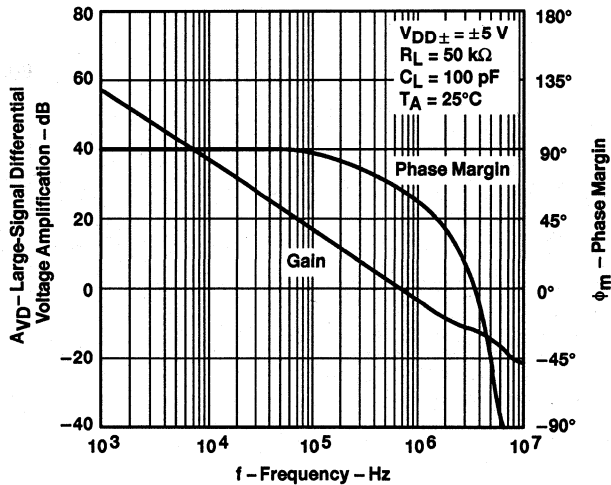


Figure 23

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

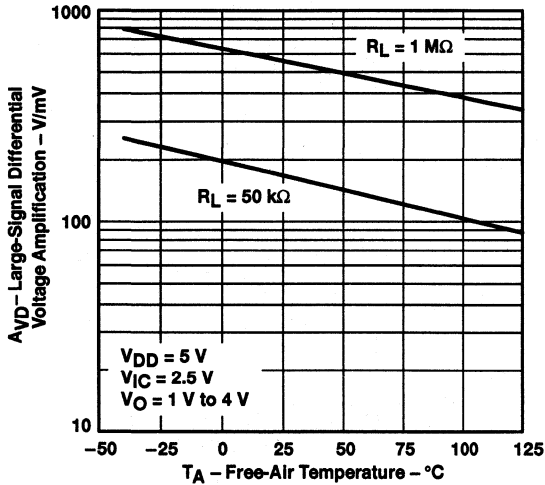


Figure 24

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

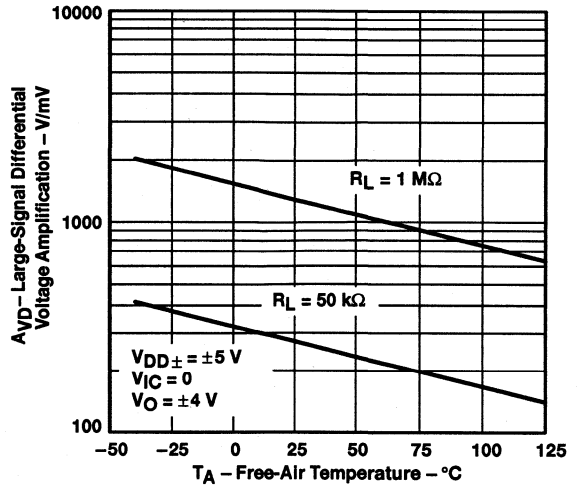


Figure 25

OUTPUT IMPEDANCE‡  
 vs  
 FREQUENCY

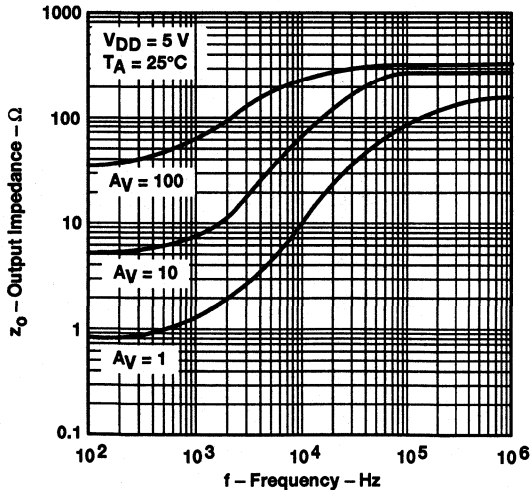


Figure 26

OUTPUT IMPEDANCE  
 vs  
 FREQUENCY

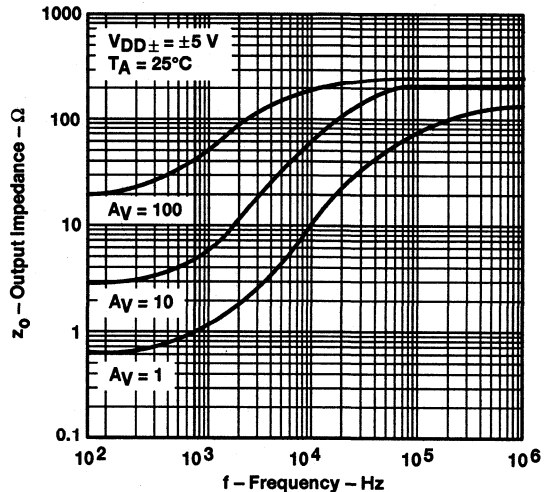


Figure 27

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

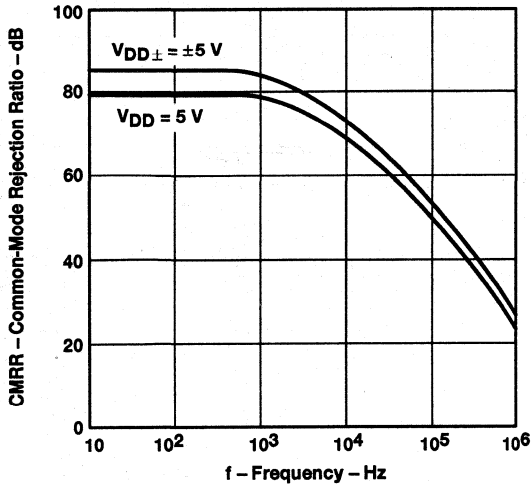
‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

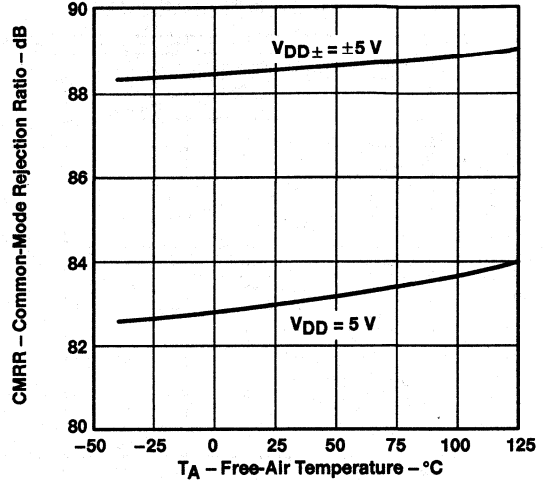
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO‡**  
**vs**  
**FREQUENCY**



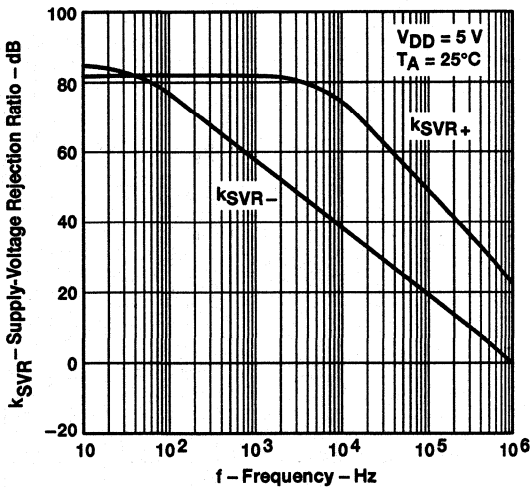
**Figure 28**

**COMMON-MODE REJECTION RATIO‡**  
**vs**  
**FREE-AIR TEMPERATURE**



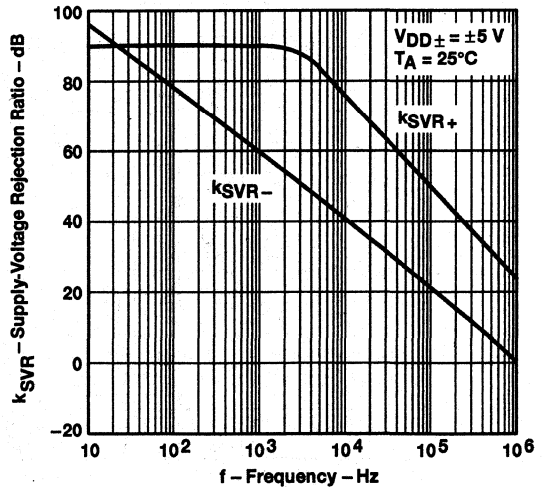
**Figure 29**

**SUPPLY-VOLTAGE REJECTION RATIO‡**  
**vs**  
**FREQUENCY**



**Figure 30**

**SUPPLY-VOLTAGE REJECTION RATIO**  
**vs**  
**FREQUENCY**

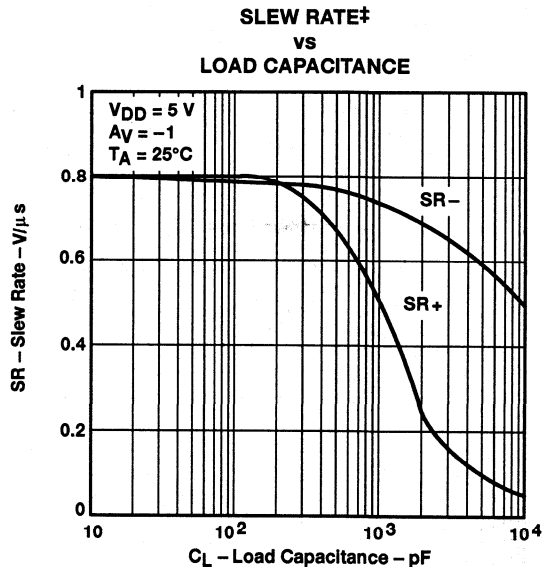
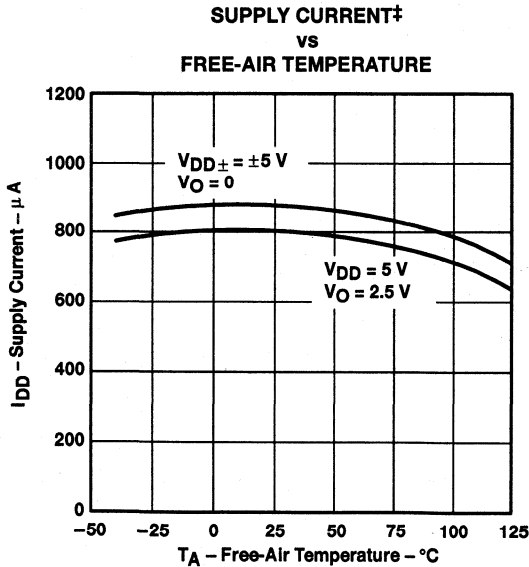
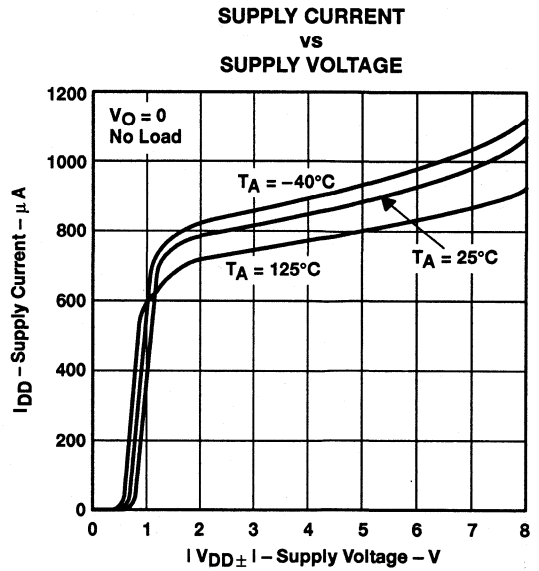
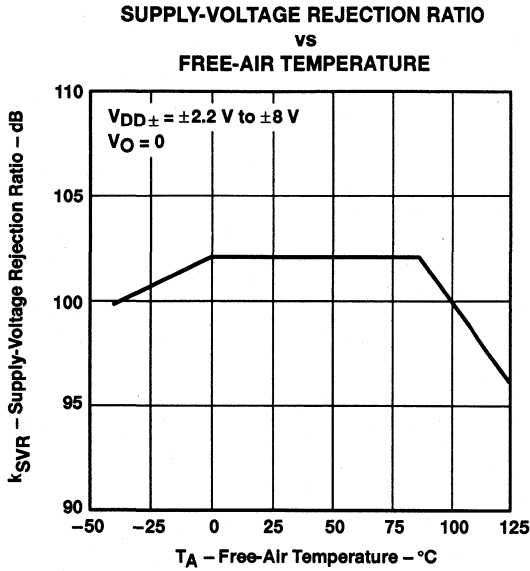


**Figure 31**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

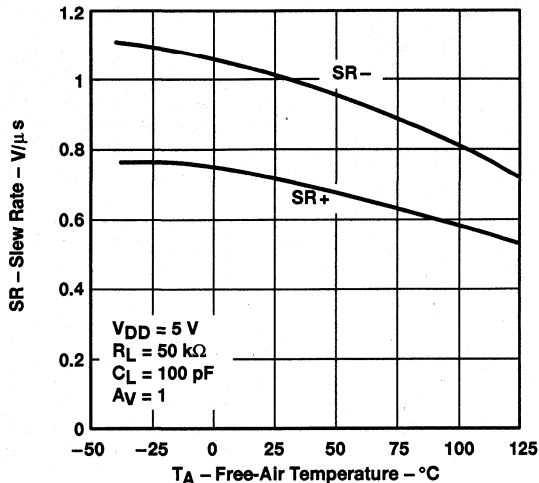
‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

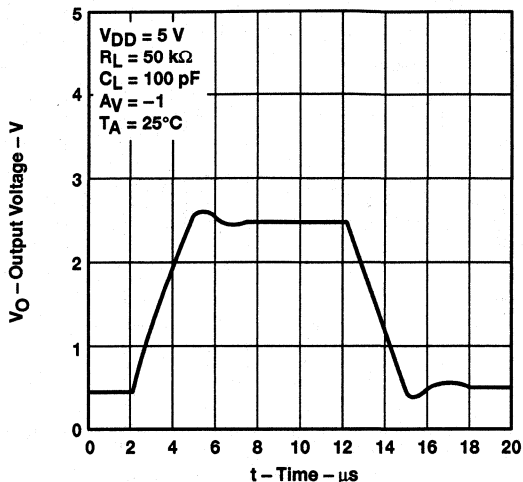
**TYPICAL CHARACTERISTICS†**

**SLEW RATE‡**  
**vs**  
**FREE-AIR TEMPERATURE**



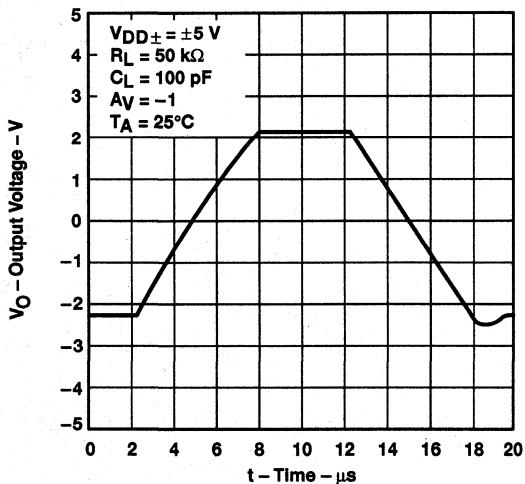
**Figure 36**

**INVERTING LARGE-SIGNAL PULSE RESPONSE‡**



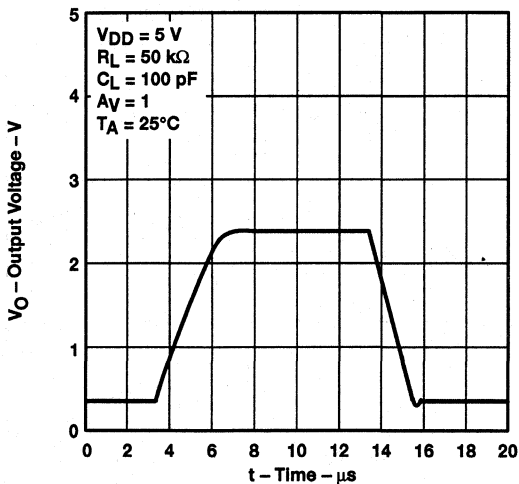
**Figure 37**

**INVERTING LARGE-SIGNAL PULSE RESPONSE**



**Figure 38**

**VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE**



**Figure 39**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER LARGE-SIGNAL  
 PULSE RESPONSE

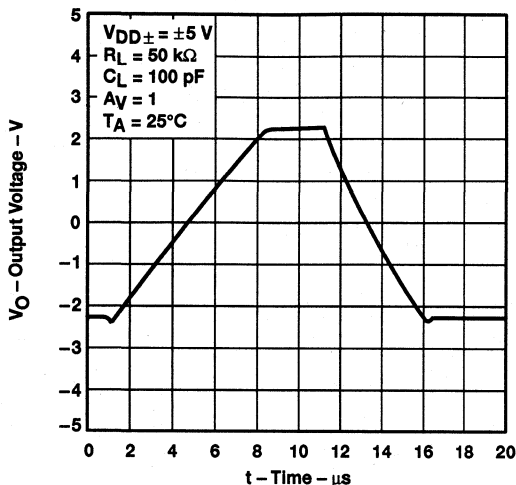


Figure 40

INVERTING SMALL-SIGNAL  
 PULSE RESPONSE†

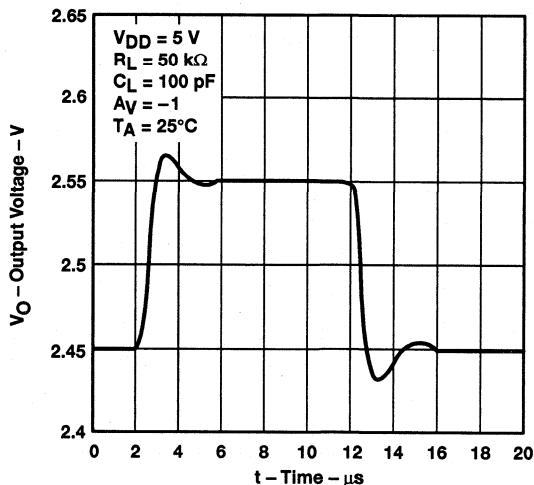


Figure 41

INVERTING SMALL-SIGNAL  
 PULSE RESPONSE

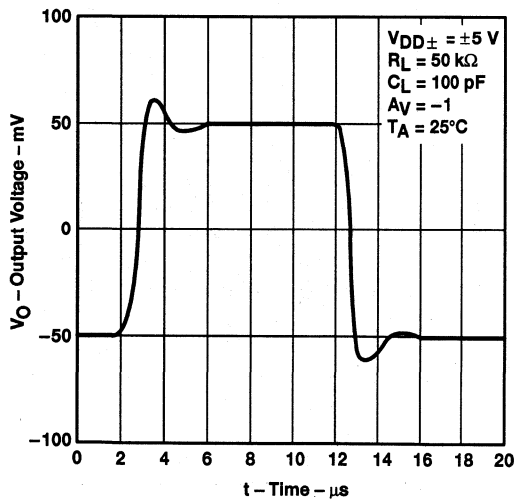


Figure 42

VOLTAGE-FOLLOWER SMALL-SIGNAL  
 PULSE RESPONSE†

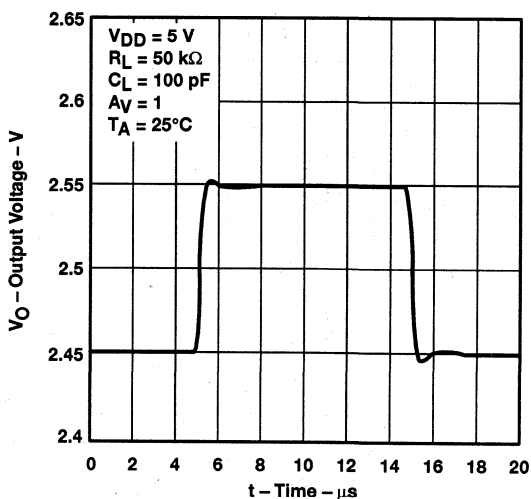


Figure 43

† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**TYPICAL CHARACTERISTICS**

**VOLTAGE-FOLLOWER SMALL-SIGNAL PULSE RESPONSE**

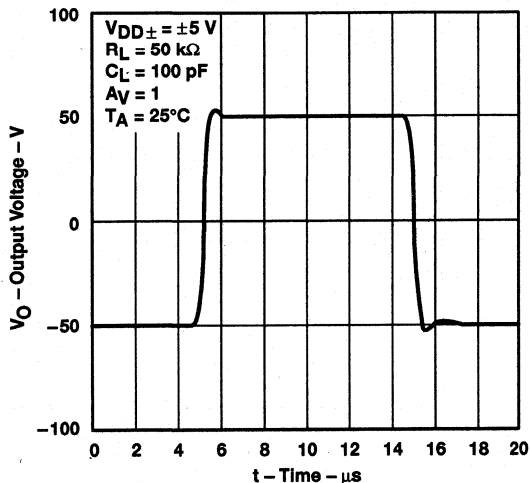


Figure 44

**EQUIVALENT INPUT NOISE VOLTAGE vs FREQUENCY**

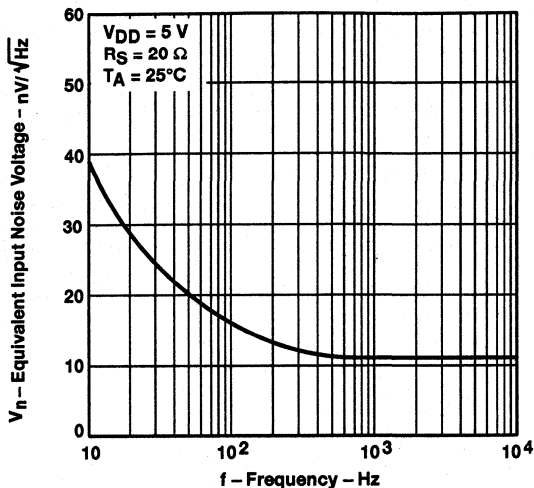


Figure 45

**EQUIVALENT INPUT NOISE VOLTAGE vs FREQUENCY**

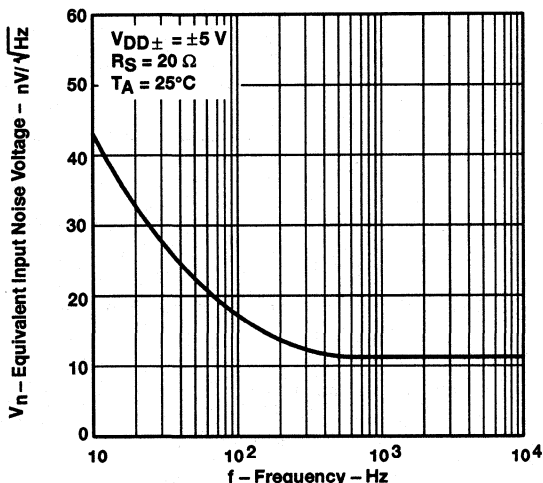


Figure 46

**INPUT NOISE VOLTAGE OVER A 10-SECOND PERIOD†**

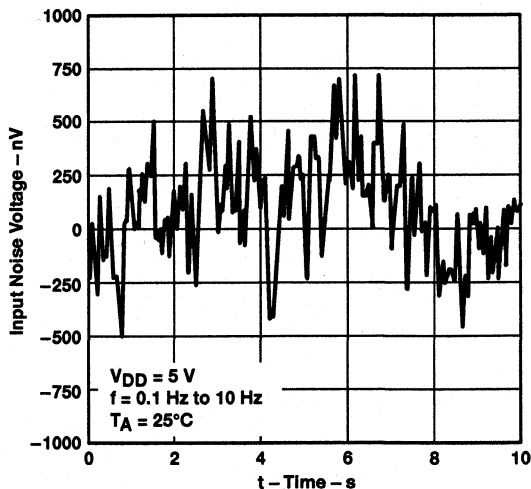


Figure 47

† For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.



TYPICAL CHARACTERISTICS†

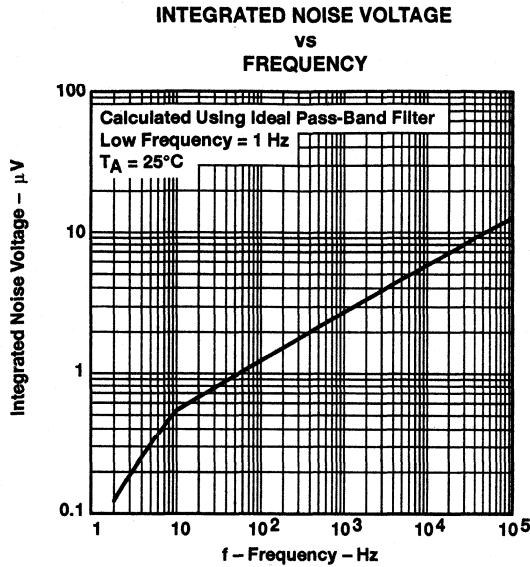


Figure 48

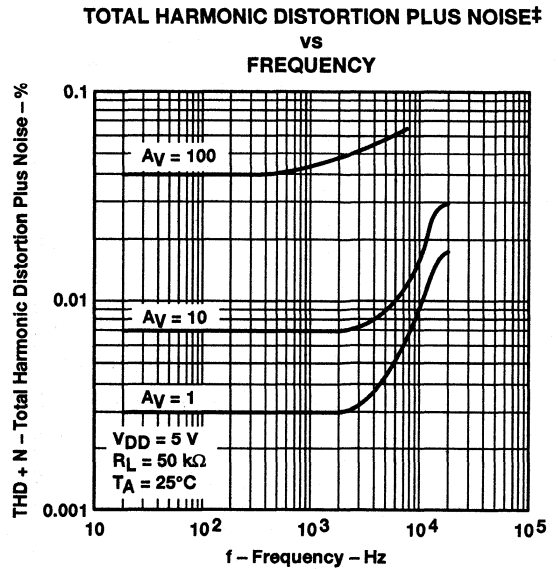


Figure 49

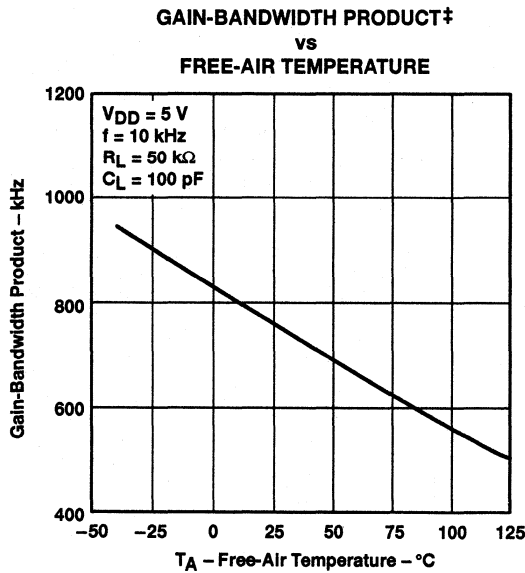


Figure 50

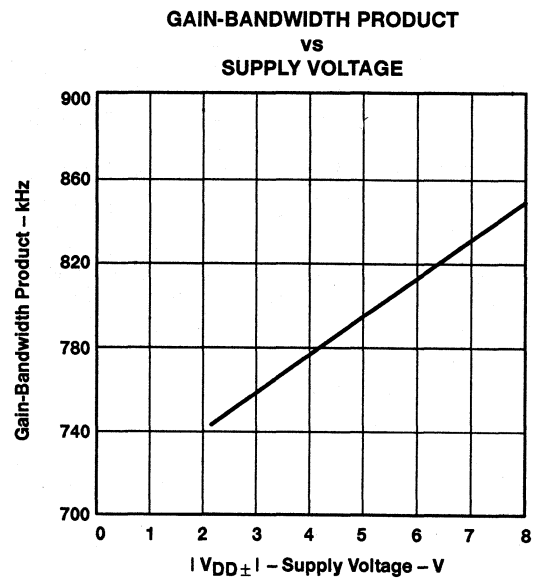


Figure 51

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

‡ For curves where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS

PHASE MARGIN  
 vs  
 LOAD CAPACITANCE

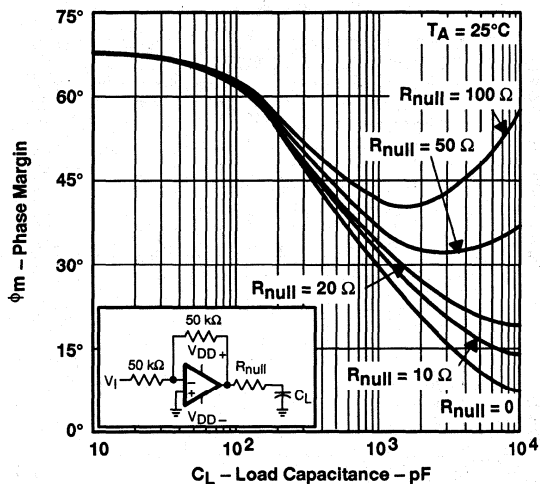


Figure 52

GAIN MARGIN  
 vs  
 LOAD CAPACITANCE

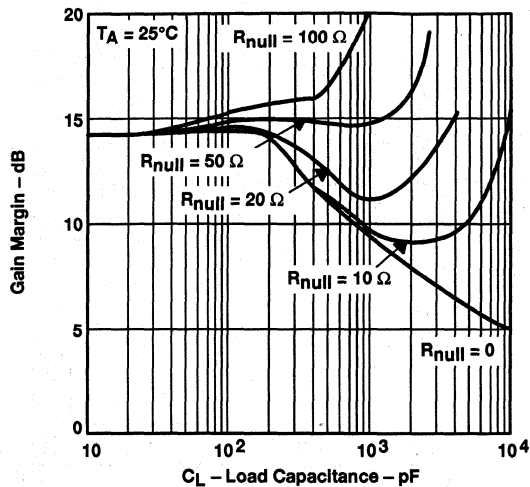


Figure 53

UNITY-GAIN BANDWIDTH†  
 vs  
 LOAD CAPACITANCE

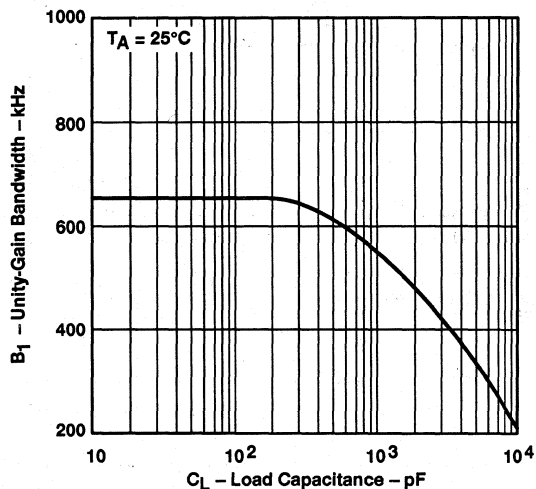


Figure 54

OVERESTIMATION OF PHASE MARGIN†  
 vs  
 LOAD CAPACITANCE

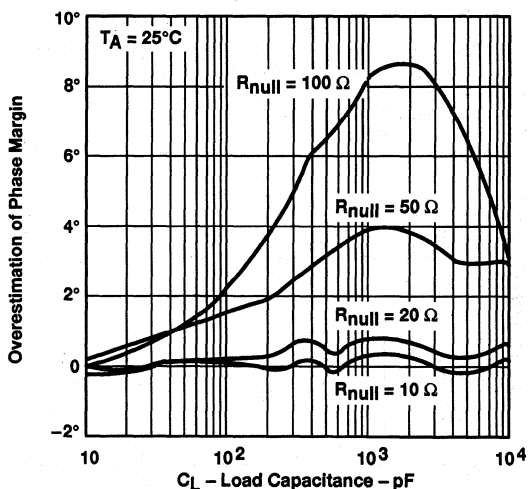


Figure 55

† See application information

## APPLICATION INFORMATION

### loading considerations

The TLC2264 is a low-power version of the TLC2274 with the appropriate design changes relative to the lower-power level. The output drive performance to the negative rail for the TLC2264 is similar to the TLC2274 and is capable of driving several milliamperes.

The design topology used for the TLC2264 or the TLC2274 limits the drive to the positive rail to a value very close to the  $I_{DD}$  for the amplifier; while the TLC2274 is capable of greater than 1-mA drive from the positive rail, the TLC2264 is capable of only a few hundred microamperes. When designing with lower impedance loads (less than 50 k $\Omega$ ) with the TLC2264, the lower drive capability to the positive rail needs to be taken into consideration. Although the TLC2264 topology provides lower drive to the positive rail than other high-output-drive rail-to-rail operational amplifiers, it is a more stable topology.

### driving large capacitive loads

The TLC2264 is designed to drive larger capacitive loads than most CMOS operational amplifiers. Figure 52 and Figure 53 illustrate its ability to drive loads greater than 400 pF while maintaining good gain and phase margins ( $R_{null} = 0$ ).

A smaller series resistor ( $R_{null}$ ) at the output of the device (see Figure 56) improves the gain and phase margins when driving large capacitive loads. Figure 52 and Figure 53 show the effects of adding series resistances of 10  $\Omega$ , 20  $\Omega$ , 50  $\Omega$ , and 100  $\Omega$ . The addition of this series resistor has two effects: the first is that it adds a zero to the transfer function, and the second is that it reduces the frequency of the pole associated with the output load in the transfer function.

The zero introduced to the transfer function is equal to the series resistance times the load capacitance. To calculate the improvement in phase margin, equation (1) can be used.

$$\Delta\theta_{m1} = \tan^{-1} \left( 2 \times \pi \times \text{UGBW} \times R_{null} \times C_L \right) \quad (1)$$

where:

- $\Delta\theta_{m1}$  = improvement in phase margin
- UGBW = unity-gain bandwidth frequency
- $R_{null}$  = output series resistance
- $C_L$  = load capacitance

The unity-gain-bandwidth (UGBW) frequency decreases as the capacitive load increases (see Figure 54). To use equation (1), UGBW must be approximated from Figure 54.

Using equation (1) alone overestimates the improvement in phase margin as illustrated in Figure 55. The overestimation is caused by the decrease in the frequency of the pole associated with the load, providing additional phase shift and reducing the overall improvement in phase margin. The pole associated with the load is reduced by the factor calculated in equation (2).

$$F = \frac{1}{1 + g_m \times R_{null}} \quad (2)$$

where:

- F = factor reducing frequency of pole
- $g_m$  = small-signal output transconductance (typically  $4.83 \times 10^{-3}$  mhos)
- $R_{null}$  = output series resistance

**TLC2264, TLC2264A, TLC2264Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS130A – DECEMBER 1993 – REVISED APRIL 1994

**APPLICATION INFORMATION**

**driving large capacitive loads (continued)**

For the TLC2264, the pole associated with the load is typically 6 MHz with 100-pF load capacitance. This value varies inversely with  $C_L$ : at  $C_L = 10$  pF, use 60 MHz, at  $C_L = 1000$  pF, use 600 kHz, and so on.

Reducing the pole associated with the load introduces phase shift, thereby reducing phase margin. This results in an error in the increase in phase margin expected by considering the zero alone [equation (1)]. Equation (3) approximates the reduction in phase margin due to the movement of the pole associated with the load. The result of this equation can be subtracted from the result of the equation in equation (1) to better approximate the improvement in phase margin.

$$\Delta\Theta_{m2} = \tan^{-1} \left[ \frac{UGBW}{(F \times P_2)} \right] - \tan^{-1} \left( \frac{UGBW}{P_2} \right) \quad (3)$$

Where:

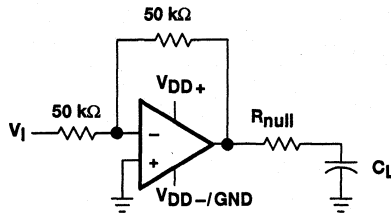
$\Delta\Theta_{m2}$  = reduction in phase margin

UGBW = unity-gain-bandwidth frequency

F = factor from equation (2)

$P_2$  = unadjusted pole (60 MHz@10 pF, 6 MHz@100 pF, etc.)

Using these equations with Figure 54 and Figure 55 enables the designer to choose the appropriate output series resistance to optimize the design of circuits driving large capacitive loads.



**Figure 56. Series-Resistance Circuit**

APPLICATION INFORMATION

macromodel information

Macromodel information provided is derived using *PSpice™ Parts™* model generation software. The Boyle macromodel (see Note 5) and subcircuit in Figure 57 are generated using the TLC2264 typical electrical and operating characteristics at  $T_A = 25^\circ\text{C}$ . Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 5: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).

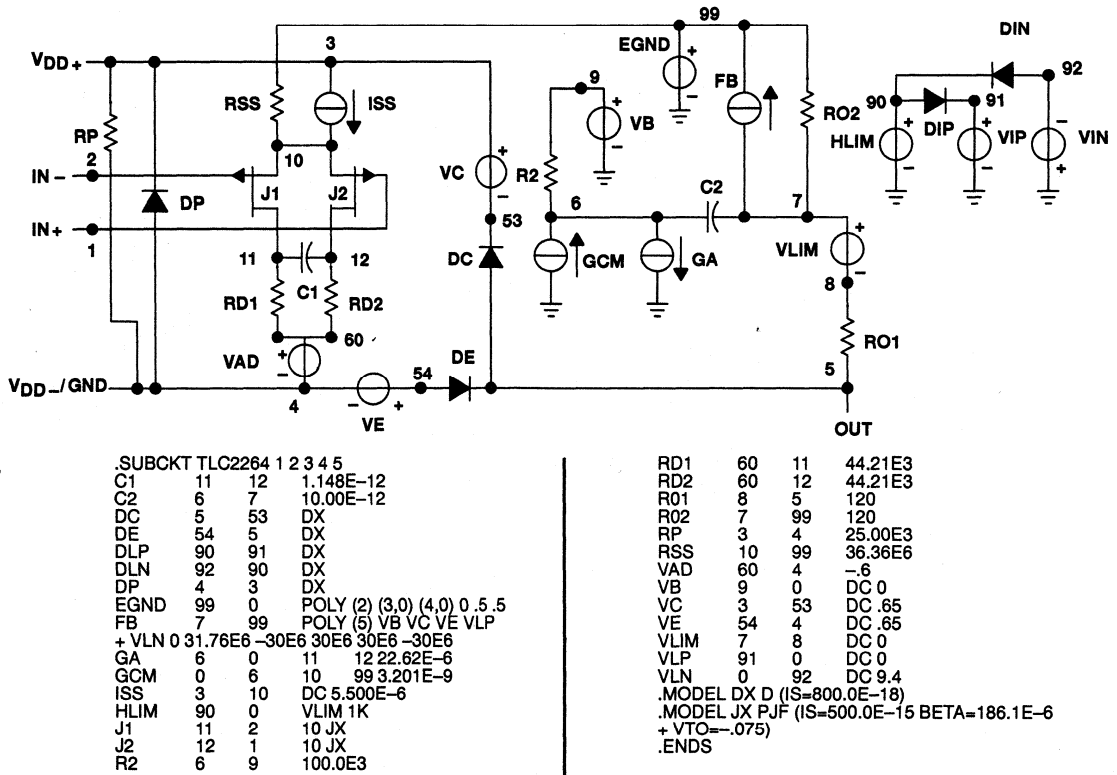


Figure 57. Boyle Macromodel and Subcircuit

*PSpice* and *Parts* are trademarks of MicroSim Corporation.



# TLC2272, TLC2272A, TLC2272Y Advanced LinCMOS™ RAIL-TO-RAIL DUAL OPERATIONAL AMPLIFIERS

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

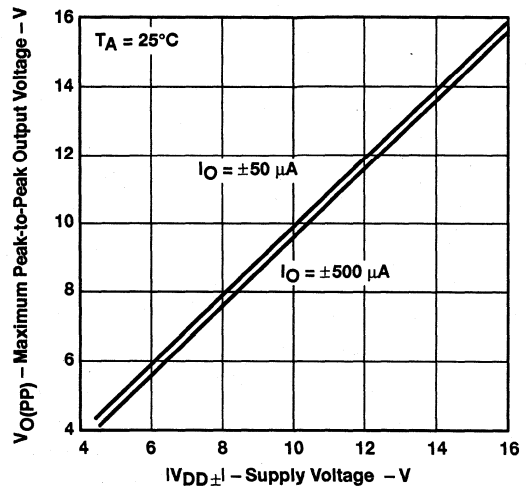
- Output Swing Includes Both Supply Rails
- Low Noise . . . 9 nV/√Hz Typ at f = 1 kHz
- Low Input Bias Current . . . 1 pA Typ
- Fully Specified for Both Single-Supply and Split-Supply Operation
- Common-Mode Input Voltage Range Includes Negative Rail
- High-Gain Bandwidth . . . 2 MHz Typ
- High Slew Rate . . . 3 V/μs Typ
- Low Input Offset Voltage  
950 μV Max at T<sub>A</sub> = 25°C
- Macromodel Included

## description

The TLC2272 and TLC2272A are dual rail-to-rail operational amplifiers manufactured using Texas Instruments Advanced LinCMOS™ process. These devices offer comparable ac performance while having better noise, input offset voltage, and power dissipation than existing CMOS operational amplifiers. In addition, the common-mode input voltage range is wider than typical standard CMOS type amplifiers. To take advantage of this improvement in performance, making this device available for a wider range of applications, V<sub>ICR</sub> is specified with a larger maximum input offset voltage test limit of ±5 mV. The Advanced LinCMOS™ process uses a silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. Also, this technology makes possible input impedance levels that meet or exceed levels offered by topgate JFET and expensive dielectric-isolated devices.

The TLC2272 and TLC2272A, exhibiting high input impedance and low noise, are excellent for small-signal conditioning for high-impedance sources, such as piezoelectric transducers. In addition, the rail-to-rail output feature with single or split supplies makes these devices great choices for inputs to ADCs in either the unipolar or bipolar mode of operation. This feature, combined with its temperature performance, makes the TLC2272 family ideal for sonobuoys, pressure sensors, temperature control, active VR sensors, accelerometers, and many other applications.

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
VS  
SUPPLY VOLTAGE



## AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max At 25°C	PACKAGED DEVICES			CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)	
0°C to 70°C	950 μV 2.5 mV	TLC2272ACD TLC2272CD	TLC2272ACP TLC2272CP	TLC2272CPWLE	TLC2272Y
-40°C to 85°C	950 μV 2.5 mV	TLC2272AID TLC2272ID	TLC2272AIP TLC2272IP	—	—
-55°C to 125°C	950 μV 2.5 mV	TLC2272AMD TLC2272MD	TLC2272AMP TLC2272MP	—	—

The D packages are available taped and reeled. Add R suffix to the device type (e.g., TLC2272CDR).  
The PW package is available only left-end taped and reeled. Chips are tested at 25°C.

Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

# TLC2272, TLC2272A, TLC2272Y

## Advanced LinCMOS™ RAIL-TO-RAIL

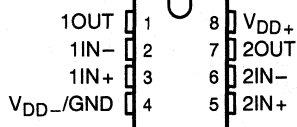
### DUAL OPERATIONAL AMPLIFIERS

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

#### description (continued)

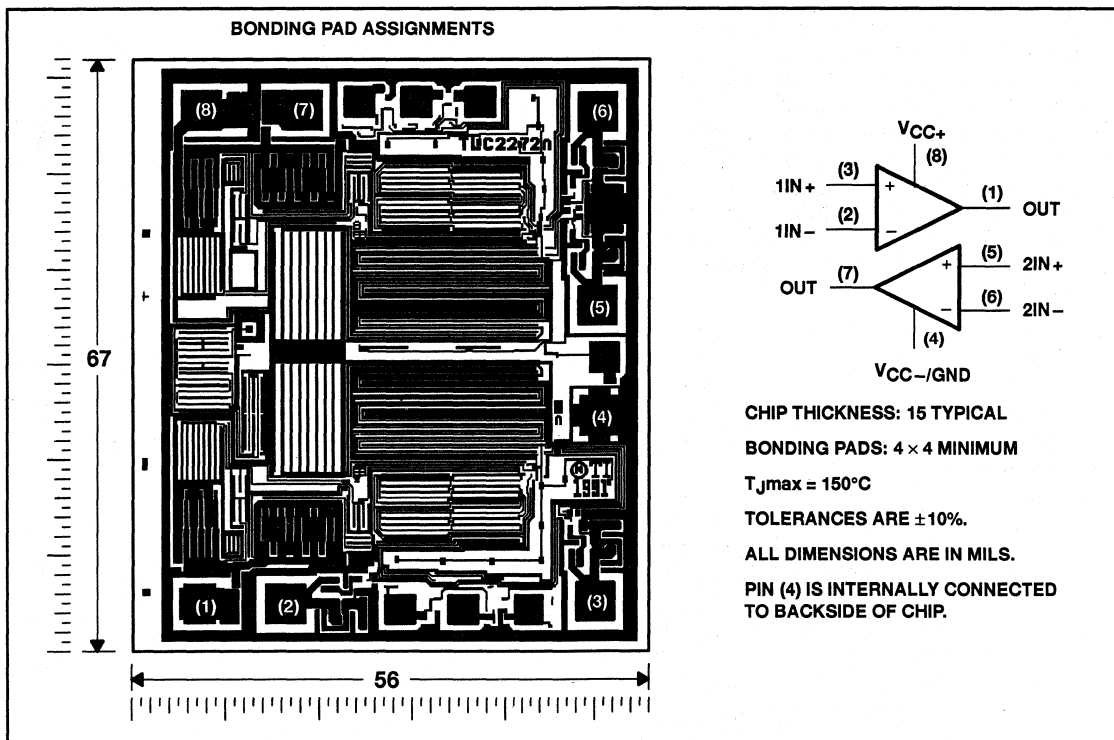
The device inputs and outputs are designed to withstand a 100-mA surge current without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

D, P, OR PW PACKAGE  
(TOP VIEW)



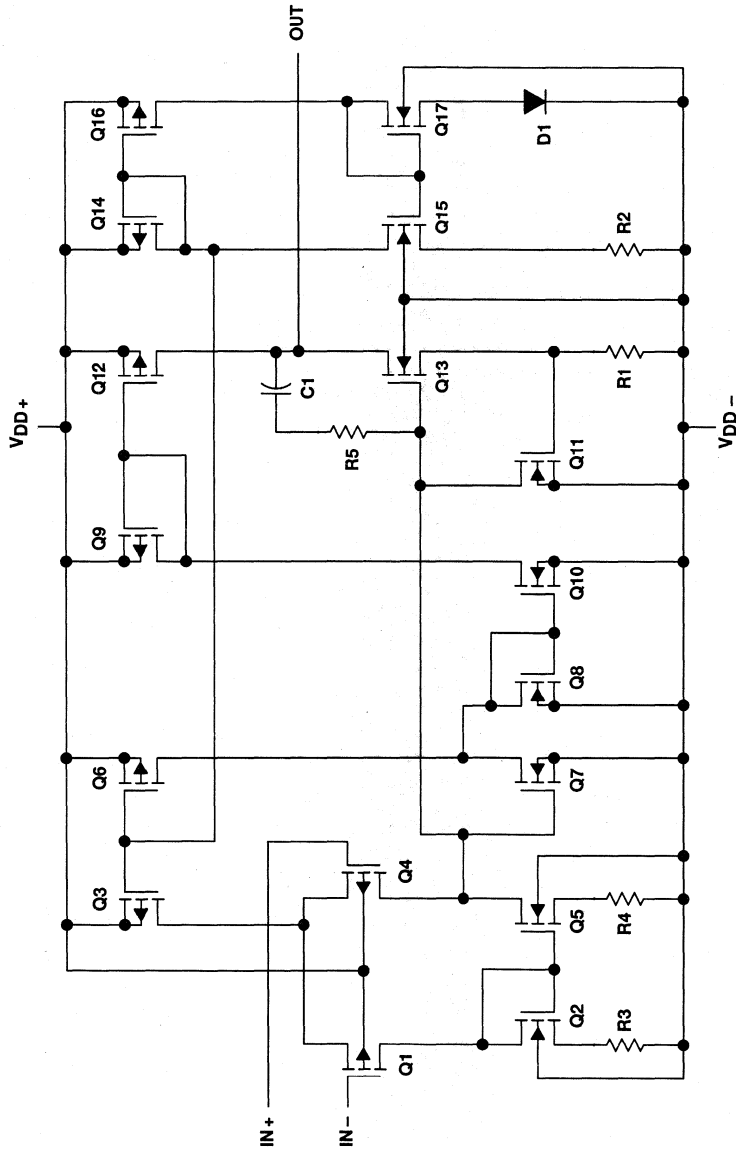
#### TLC2272Y chip information

These chips, when properly assembled, display characteristics similar to the TLC2272C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.





equivalent schematic (each amplifier)



COMPONENT COUNT†	
Transistors	38
Diodes	9
Resistors	26
Capacitors	3

† Includes both amplifiers and all ESD, bias, and trim circuitry

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	±16 V
Input voltage, $V_I$ (any input, see Note 1)	±8 V
Input current, $I_I$ (any input)	±5 mA
Output current, $I_O$	±50 mA
Total current into $V_{DD+}$	±50 mA
Total current out of $V_{DD-}$	±50 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ . Excessive current will flow if input is brought below  $V_{DD-} - 0.3$  V.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	337 mW	145 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW
PW	525 mW	4.2 mW/°C	336 mW	—	—

**recommended operating conditions**

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	±2.2	±8	±2.2	±8	±2.2	±8	V
Input voltage range, $V_I$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272C			TLC2272AC			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage		25°C	300 2500			300 950			$\mu\text{V}$	
		Full range	3000			1500				
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 70°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, V_{DD} \pm \pm 2.5\text{ V}, R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5			0.5			pA	
		Full range	100			100				
$I_{IB}$ Input bias current		25°C	1			1			pA	
		Full range	100			100				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2		V	
		Full range	0 to 3.5			0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99			4.99			V	
		25°C	4.85	4.93		4.85	4.93			
		Full range	4.85			4.85				
		25°C	4.25	4.65		4.25	4.65			
$I_{OH} = -1\text{ mA}$	25°C	4.25			4.25			V		
	Full range	4.25			4.25					
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}, I_{OL} = 50\ \mu\text{A}$	25°C	0.01			0.01			V	
		25°C	0.09	0.15		0.09	0.15			
		Full range	0.15			0.15				
		25°C	0.9	1.5		0.9	1.5			
$I_{OL} = 5\text{ mA}$	25°C	1.5			1.5			V		
	Full range	1.5			1.5					
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}, V_O = 1\text{ V to }4\text{ V}$	$R_L = 10\text{ k}\Omega$ ‡	25°C	15	35		15	35		V/mV
			Full range	15			15			
		$R_L = 1\text{ m}\Omega$ ‡	25°C	175			175			
$r_{id}$ Differential input resistance		25°C	1012			1012			$\Omega$	
$r_i$ Common-mode input resistance		25°C	1012			1012			$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}, \text{ P package}$	25°C	8			8			pF	
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}, A_V = 10$	25°C	140			140			$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}, V_O = 2.5\text{ V}, R_S = 50\ \Omega$	25°C	70	75		70	75		dB	
		Full range	70			70				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}, V_{IC} = V_{DD}/2, \text{ No load}$	25°C	80	95		80	95		dB	
		Full range	80			80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}, \text{ No load}$	25°C	2.2	3		2.2	3		mA	
		Full range	3			3				

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272C			TLC2272AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	2.3	3.6		2.3	3.6		V/ $\mu\text{s}$
		Full range	1.7			1.7			
$V_n$	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		nV/ $\sqrt{\text{Hz}}$
		25°C		9			9		
$V_{NPP}$	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
$I_n$	Equivalent input noise current	25°C		0.6			0.6		fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega^\ddagger$	25°C	$A_V = 1$	0.0013%		0.0013%			
			$A_V = 10$	0.004%		0.004%			
			$A_V = 100$	0.03%		0.03%			
	Gain-bandwidth product $f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C		2.18			2.18		MHz
BOM	Maximum output-swing bandwidth $V_{O(PP)} = 2\text{ V}$ , $A_V = 1$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C		1			1		MHz
$t_s$	Settling time $A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	To 0.1%	1.5		1.5		$\mu\text{s}$	
			To 0.01%	2.6		2.6			
$\phi_m$	Phase margin at unity gain $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C		50°			50°		
		25°C		10			10		dB

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise specified)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272C			TLC2272AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$
		Full range			3000			1500	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 70°C		2			2	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C		0.002			0.002	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5			0.5	$\text{pA}$	
$I_{IB}$ Input bias current		25°C		1			1		$\text{pA}$
		Full range			100			100	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	25°C	-5 to 4	-5.3 to 4.2		-5 to 4	-5.3 to 4.2	$\text{V}$	
		Full range	-5 to 3.5			-5 to 3.5			
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C		4.99			4.99	$\text{V}$	
		25°C	4.85	4.93		4.85	4.93		
		Full range	4.85			4.85			
		25°C	4.25	4.65		4.25	4.65		
$V_{OM-}$ Maximum negative peak output voltage	$I_O = -1\ \text{mA}$	25°C		-4.99			-4.99	$\text{V}$	
		25°C	-4.85	-4.91		-4.85	-4.91		
		Full range	-4.85			-4.85			
		25°C	-3.5	-4.1		-3.5	-4.1		
$V_{AVD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 10\ \text{k}\Omega$	25°C	25	50		25	50	$\text{V}/\text{mV}$
			Full range	25			25		
		$R_L = 1\ \text{m}\Omega$	25°C		300			300	
			Full range						
$r_{id}$ Differential input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$r_i$ Common-mode input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, \text{ P package}$	25°C		8			8	$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz}, A_v = 10$	25°C		130			130	$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{to}\ 2.7\ \text{V}, V_O = 0\ \text{V}, R_S = 50\ \Omega$	25°C	75	80		75	80	$\text{dB}$	
		Full range	75			75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD\pm} = 2.2\ \text{V to}\ \pm 8\ \text{V}, V_{IC} = 0, \text{ No load}$	25°C	80	95		80	95	$\text{dB}$	
		Full range	80			80			
$I_{DD}$ Supply current	$V_O = 0\ \text{V}, \text{ No load}$	25°C		2.4	3		2.4	3	$\text{mA}$
		Full range			3			3	

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5 \text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272C			TLC2272AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = \pm 2.3 \text{ V}$ , $C_L = 100 \text{ pF}$ , $R_L = 10 \text{ k}\Omega$	25°C	2.3	3.6		2.3	3.6	V/ $\mu\text{s}$	
		Full range	1.7			1.7			
$V_n$	Equivalent input noise voltage $f = 10 \text{ Hz}$ $f = 1 \text{ kHz}$	25°C		50			50	nV/ $\sqrt{\text{Hz}}$	
		25°C		9			9		
$V_{NPP}$	Peak-to-peak equivalent input noise voltage $f = 0.1 \text{ Hz to } 1 \text{ Hz}$ $f = 0.1 \text{ Hz to } 10 \text{ Hz}$	25°C		1			1	$\mu\text{V}$	
		25°C		1.4			1.4		
$I_n$	Equivalent input noise current	25°C		0.6			0.6	fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion pulse duration $V_O = \pm 2.3 \text{ V}$ , $f = 20 \text{ kHz}$ , $R_L = 10 \text{ k}\Omega$	25°C		$A_V = 1$		0.0011%		0.0011%	
				$A_V = 10$		0.004%		0.004%	
				$A_V = 100$		0.03%		0.03%	
	Gain-bandwidth product $f = 10 \text{ kHz}$ , $C_L = 100 \text{ pF}$ , $R_L = 10 \text{ k}\Omega$	25°C		2.25			2.25	MHz	
BOM	Maximum output-swing bandwidth $V_{O(PP)} = 4.6 \text{ V}$ , $R_L = 10 \text{ k}\Omega$ , $A_V = 1$ , $C_L = 100 \text{ pF}$	25°C		0.54			0.54	MHz	
$t_s$	Settling time $A_V = -1$ , Step = $-2.3 \text{ V to } 2.3 \text{ V}$ , $R_L = 10 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$	25°C		To 0.1%		1.5		1.5	$\mu\text{s}$
				To 0.01%		3.2		3.2	
$\phi_m$	Phase margin at unity gain $R_L = 10 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$	25°C		52°			52°		
		25°C		*	10			10	dB

† Full range is 0°C to 70°C.



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272I			TLC2272AI			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage		25°C	300	2500		300	950	$\mu\text{V}$		
		Full range			3000		1500			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 85°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0,$ $V_O = 0,$ $V_{DD} \pm \pm 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$	
	Full range	150			150					
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$	
		Full range	150			150				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,$ $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2	$\text{V}$		
		Full range	0 to 3.5			0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99			4.99			$\text{V}$	
		25°C	4.85	4.93		4.85	4.93			
		Full range	4.85			4.85				
		25°C	4.25	4.65		4.25	4.65			
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 50\ \mu\text{A}$	25°C	0.01			0.01			$\text{V}$	
		25°C	0.09	0.15		0.09	0.15			
		Full range	0.15			0.15				
		25°C	0.9	1.5		0.9	1.5			
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V},$ $V_O = 1\text{ V to }4\text{ V}$	$R_L = 10\ \text{k}\Omega^\ddagger$	25°C	15	35		15	35	$\text{V}/\text{mV}$	
			Full range	15			15			
		$R_L = 1\ \text{m}\Omega^\ddagger$	25°C	175			175			
			Full range	1.5			1.5			
$r_{id}$ Differential input resistance		25°C	1012			1012			$\Omega$	
$r_i$ Common-mode input resistance		25°C	1012			1012			$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz},$ P package	25°C	8			8			$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz},$ $A_V = 10$	25°C	140			140			$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V},$ $V_O = 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	70	75		70	75	dB		
		Full range	70			70				
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V},$ $V_{IC} = V_{DD}/2,$ No load	25°C	80	95		80	95	dB		
		Full range	80			80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V},$ No load	25°C	2.2	3		2.2	3	$\text{mA}$		
		Full range	3			3				

† Full range is -40°C to 85°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272I			TLC2272AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	2.3	3.6	2.3	3.6	V/ $\mu\text{s}$	
			Full range	1.7		1.7			
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50		50	nV/ $\sqrt{\text{Hz}}$	
			25°C		9		9		
$V_{NPP}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$	25°C		1		1	$\mu\text{V}$	
			25°C		1.4		1.4		
$I_n$	Equivalent input noise current		25°C		0.6		0.6	fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ ‡	25°C	$A_V = 1$	0.0013%		0.0013%		
				$A_V = 10$	0.004%		0.004%		
				$A_V = 100$	0.03%		0.03%		
	Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ ‡	25°C		2.18		2.18	MHz	
$B_{OM}$	Maximum output-swing bandwidth	$V_O(PP) = 2\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡	25°C		1		1	MHz	
$t_s$	Settling time	$A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	To 0.1%	1.5		1.5	$\mu\text{s}$	
				To 0.01%	2.6		2.6		
$\phi_m$	Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C		50°		50°		
			25°C		10		10		
	Gain margin		25°C		10		10	dB	

† Full range is -40°C to 85°C.

‡ Referenced to 2.5 V





**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272I			TLC2272AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$
		Full range		3000		1500			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 85°C		2		2		$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C		0.002		0.002		$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5		0.5		pA	
		Full range		150		150			
$I_{IB}$ Input bias current		25°C		1		1		pA	
		Full range		150		150			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	25°C	-5 to 4	-5.3 to 4.2	-5 to 4	-5.3 to 4.2		V	
		Full range	-5 to 3.5		-5 to 3.5				
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C		4.99		4.99		V	
		25°C	4.85	4.93	4.85	4.93			
		Full range	4.85		4.85				
		25°C	4.25	4.65	4.25	4.65			
$V_{OM-}$ Maximum negative peak output voltage	$I_O = -1\ \text{mA}$	25°C		-4.99		-4.99		V	
		25°C	-4.85	-4.91	-4.85	-4.91			
		Full range	-4.85		-4.85				
		25°C	-3.5	-4.1	-3.5	-4.1			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 10\ \text{k}\Omega$	25°C	25	50	25	50	V/mV	
			Full range	25		25			
		$R_L = 1\ \text{m}\Omega$	25°C	300		300			
			Full range	25		25			
$r_{id}$ Differential input resistance		25°C	$10^{12}$			$10^{12}$	$\Omega$		
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$	$\Omega$		
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, \text{ P package}$	25°C	8			8	pF		
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz}, A_V = 10$	25°C	130			130	$\Omega$		
CMRR Common-mode rejection ratio	$V_{IC} = 0\ \text{to}\ 2.7\ \text{V}, V_O = 2.5\ \text{V}, R_S = 50\ \Omega$	25°C	75	80	75	80	dB		
		Full range	75		75				
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD} \pm / \Delta V_{IO}$ )	$V_{DD} = 4.4\ \text{V}\ \text{to}\ 16\ \text{V}, V_{IC} = V_{DD}/2, \text{ No load}$	25°C	80	95	80	95	dB		
		Full range	80		80				
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}, \text{ No load}$	25°C	2.4	3	2.4	3	mA		
		Full range	3		3				

† Full range is -40°C to 85°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272I			TLC2272AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = \pm 2.3\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C	2.3	3.6		2.3	3.6		V/ $\mu\text{s}$
		Full range	1.7			1.7			
$V_n$	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		nV $\sqrt{\text{Hz}}$
		25°C		9			9		
$V_{NPP}$	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to } 1\text{ Hz}$ $f = 0.1\text{ Hz to } 10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
$I_n$	Equivalent input noise current	25°C		0.6			0.6		fA $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = \pm 2.3\text{ V}$ $R_L = 10\text{ k}\Omega$ , $f = 20\text{ kHz}$	25°C		$A_V = 1$		0.0011%		0.0011%	
				$A_V = 10$		0.004%		0.004%	
				$A_V = 100$		0.03%		0.03%	
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C		2.25			2.25		MHz
$B_{OM}$	Maximum output-swing bandwidth $V_{O(PP)} = 4.6\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$ , $A_V = 1$	25°C		0.54			0.54		MHz
$t_s$	Settling time $A_V = -1$ , Step = $-2.3\text{ V to } 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		To 0.1%		1.5		1.5	$\mu\text{s}$
				To 0.01%		3.2		3.2	
$\phi_m$	Phase margin at unity gain $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		52°			52°		
		25°C		10			10		dB

† Full range is  $-40^\circ\text{C to } 85^\circ\text{C}$ .



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2272M			TLC2272AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C		300	2500		300	950	$\mu\text{V}$
		Full range			3000			1500	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 125°C		2			2	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_{O} = 0,$ $V_{DD\pm} = \pm 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C		0.002			0.002	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5			0.5	$\text{pA}$	
		Full range			500		500		
$I_{IB}$ Input bias current		25°C		1			1	$\text{pA}$	
		Full range			500		500		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2	V	
		Full range	0 to 3.5			0 to 3.5			
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C		4.99			4.99	V	
	$I_{OH} = -200\ \mu\text{A}$	25°C	4.85	4.93		4.85	4.93		
	$I_{OH} = -1\text{ mA}$	Full range		4.85			4.85		
		25°C	4.25	4.65		4.25	4.65		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}, I_{OL} = 50\ \mu\text{A}$	25°C		0.01			0.01	V	
	$V_{IC} = 2.5\text{ V}, I_{OL} = 500\ \mu\text{A}$	25°C	0.09	0.15		0.09	0.15		
		Full range		0.15			0.15		
	$V_{IC} = 2.5\text{ V}, I_{OL} = 5\text{ mA}$	25°C	0.9	1.5		0.9	1.5		
Full range			1.5			1.5			
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}, V_{O} = 1\text{ V to }4\text{ V}$ $R_L = 10\text{ k}\Omega$ ‡ $R_L = 1\text{ m}\Omega$ ‡	25°C	10	35		10	35	V/mV	
		Full range		10			10		
		25°C		175			175		
$r_{id}$ Differential input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$r_i$ Common-mode input resistance		25°C		10 <sup>12</sup>			10 <sup>12</sup>	$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz},$ P package	25°C		8			8	pF	
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}, A_V = 10$	25°C		140			140	$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}, V_{O} = 2.5\text{ V}, R_S = 50\ \Omega$	25°C	70	75		70	75	dB	
		Full range		70			70		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}, V_{IC} = V_{DD}/2,$ No load	25°C	80	95		80	95	dB	
		Full range		80			80		
$I_{DD}$ Supply current	$V_{O} = 2.5\text{ V},$ No load	25°C		2.2	3		2.2	3	mA
		Full range			3			3	

† Full range is -55°C to 125°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	TA†	TLC2272M			TLC2272AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	2.3	3.6		2.3	3.6		V/ $\mu\text{s}$
		Full range	1.7			1.7			
Vn	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		nV/ $\sqrt{\text{Hz}}$
		25°C		9			9		
VNPP	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
In	Equivalent input noise current	25°C		0.6			0.6		fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega^\ddagger$	25°C		$A_V = 1$		0.0013%		0.0013%	
				$A_V = 10$		0.004%		0.004%	
				$A_V = 100$		0.03%		0.03%	
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}^\ddagger$	25°C			2.18			2.18	MHz
BOM	Maximum output-swing bandwidth $V_{O(PP)} = 2\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$	25°C			1			1	MHz
ts	Settling time $A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C		To 0.1%		1.5		1.5	$\mu\text{s}$
				To 0.01%		2.6		2.6	
$\phi_m$	Phase margin at unity gain Gain margin	25°C			50°			50°	
		25°C			10			10	dB

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

‡ Referenced to 2.5 V



**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2272M			TLC2272AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V <sub>IO</sub> Input offset voltage		25°C	300	2500	300	950	μV		
		Full range	3000			1500			
αV <sub>IO</sub> Temperature coefficient of input offset voltage		25°C to 125°C	2			2		μV/°C	
Input offset voltage long-term drift (see Note 4)	V <sub>IC</sub> = 0, V <sub>O</sub> = 0, R <sub>S</sub> = 50 Ω	25°C	0.002			0.002		μV/mo	
I <sub>IO</sub> Input offset current		25°C	0.5			0.5		pA	
		Full range	500			500			
I <sub>IB</sub> Input bias current		25°C	1			1		pA	
		Full range	500			500			
V <sub>ICR</sub> Common-mode input voltage range	R <sub>S</sub> = 50 Ω,  V <sub>IO</sub>   ≤ 5 mV	25°C	-5 to 4	-5.3 to 4.2	-5 to 4	-5.3 to 4.2	V		
		Full range	-5 to 3.5			-5 to 3.5			
V <sub>OM+</sub> Maximum positive peak output voltage	I <sub>O</sub> = -20 μA	25°C	4.99			4.99		V	
	I <sub>O</sub> = -200 μA	25°C	4.85	4.93	4.85	4.93			
	I <sub>O</sub> = -1 mA	Full range	4.85			4.85			
		25°C	4.25	4.65	4.25	4.65			
V <sub>OM-</sub> Maximum negative peak output voltage	V <sub>IC</sub> = 0, I <sub>O</sub> = 50 μA	25°C	-4.99			-4.99		V	
	V <sub>IC</sub> = 0, I <sub>O</sub> = 500 μA	25°C	-4.85	-4.91	-4.85	-4.91			
	V <sub>IC</sub> = 0, I <sub>O</sub> = 5 mA	Full range	-4.85			-4.85			
		25°C	-3.5	-4.1	-3.5	-4.1			
A <sub>VD</sub> Large-signal differential voltage amplification	V <sub>O</sub> = ±4 V	R <sub>L</sub> = 10 kΩ	25°C	20	50	20	50	V/mV	
			Full range	20			20		
		R <sub>L</sub> = 1 mΩ	25°C	300			300		
r <sub>id</sub> Differential input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>		Ω	
r <sub>i</sub> Common-mode input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>		Ω	
c <sub>i</sub> Common-mode input capacitance	f = 10 kHz, P package	25°C	8			8		pF	
z <sub>o</sub> Closed-loop output impedance	f = 1 MHz, A <sub>V</sub> = 10	25°C	130			130		Ω	
CMRR Common-mode rejection ratio	V <sub>IC</sub> = 0 to 2.7 V, V <sub>O</sub> = 2.5 V, R <sub>S</sub> = 50 Ω	25°C	75	80	75	80	dB		
		Full range	75			75			
k <sub>SVR</sub> Supply-voltage rejection ratio (ΔV <sub>DD</sub> ±/ΔV <sub>IO</sub> )	V <sub>DD</sub> = 4.4 V to 16 V, V <sub>IC</sub> = 0, No load	25°C	80	95	80	95	dB		
		Full range	80			80			
I <sub>DD</sub> Supply current	V <sub>O</sub> = 2.5 V, No load	25°C	2.4	3	2.4	3	mA		
		Full range	3			3			

† Full range is -55°C to 125°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**operating characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2272M			TLC2272AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = \pm 2.3\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C	2.3	3.6		2.3	3.6		V/ $\mu\text{s}$
		Full range	1.7			1.7			
V <sub>n</sub>	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		nV/ $\sqrt{\text{Hz}}$
		25°C		9			9		
V <sub>NPP</sub>	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
I <sub>n</sub>	Equivalent input noise current	25°C		0.6			0.6		fA/ $\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = \pm 2.3\text{ V}$ $R_L = 10\text{ k}\Omega$ , $f = 20\text{ kHz}$	25°C		A <sub>v</sub> = 1		0.0011%		0.0011%	
				A <sub>v</sub> = 10		0.004%		0.004%	
				A <sub>v</sub> = 100		0.03%		0.03%	
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C		2.25			2.25		MHz
BOM	Maximum output-swing bandwidth $V_O(\text{PP}) = 4.6\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $A_v = 1$ , $C_L = 100\text{ pF}$	25°C		0.54			0.54		MHz
t <sub>s</sub>	Settling time A <sub>v</sub> = -1, Step = -2.3 V to 2.3 V, $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	To 0.1%		1.5		1.5		$\mu\text{s}$
			To 0.01%		3.2		3.2		
$\phi_m$	Phase margin at unity gain $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		52°			52°		
	Gain margin	25°C		10			10		dB

† Full range is -55°C to 125°C.



**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2272Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $V_{O} = 0$ , $V_{DD} \pm = \pm 2.5\text{ V}$ , $R_S = 50\ \Omega$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	0 to 4	-0.3 to 4.2		$\text{V}$
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$		4.99		$\text{V}$
	$I_{OH} = -200\ \mu\text{A}$		4.85	4.93	
	$I_{OH} = -1\text{ mA}$		4.25	4.65	
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$		0.01		$\text{V}$
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$		0.09	0.15	
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 5\text{ mA}$		0.9	1.5	
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_{O} = 1\text{ V to } 4\text{ V}$	$R_L = 10\text{ k}\Omega^\dagger$	15	35	$\text{V/mV}$
		$R_L = 1\text{ M}\Omega^\dagger$		175	
$r_{id}$ Differential input resistance			$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$		8		$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$		140		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to } 2.7\text{ V}$ , $V_{O} = 2.5\text{ V}$ , $R_S = 50\ \Omega$		70	75	$\text{dB}$
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to } 16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load		80	95	$\text{dB}$
$I_{DD}$ Supply current	$V_{O} = 2.5\text{ V}$ , No load		2.2	3	$\text{mA}$

$^\dagger$  Referenced to 2.5 V

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	TLC2272Y			UNIT
			MIN	TYP	MAX	
$V_{IO}$	Input offset voltage	$V_{IC} = 0$ , $R_S = 50\ \Omega$ , $V_O = 0$		300	2500	$\mu\text{V}$
$I_{IO}$	Input offset current			0.5	100	$\text{pA}$
$I_{IB}$	Input bias current			1	100	$\text{pA}$
$V_{ICR}$	Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	-5 to 4	-5.3 to 4.2		V
$V_{OM+}$	Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$		4.99		V
		$I_O = -200\ \mu\text{A}$	4.85	4.93		
		$I_O = -1\text{ mA}$	4.25	4.65		
$V_{OM-}$	Maximum negative peak output voltage	$V_{IC} = 0$ , $I_{OL} = 50\ \mu\text{A}$		-4.99		V
		$V_{IC} = 0$ , $I_{OL} = 500\ \mu\text{A}$	-4.85	-4.91		
		$V_{IC} = 0$ , $I_{OL} = 5\text{ mA}$	-3.5	-4.1		
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}$	$R_L = 10\text{ k}\Omega$	25	50	V/mV
			$R_L = 1\text{ M}\Omega$		300	
$r_{id}$	Differential input resistance			$10^{12}$		$\Omega$
$r_i$	Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$	Common-mode input capacitance	$f = 10\text{ kHz}$		8		$\text{pF}$
$z_o$	Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$		130		$\Omega$
CMRR	Common-mode rejection ratio	$V_{IC} = -5\text{ V to } 2.7\text{ V}$ , $V_O = 0$ , $R_S = 50\ \Omega$	75	80		dB
kSVR	Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\text{ V to } \pm 8\text{ V}$ , $V_{IC} = 0$ , No load	80	95		dB
$I_{DD}$	Supply current	$V_O = 0$ , No load	2.4	3		$\text{mA}$



## TYPICAL CHARACTERISTICS

**Table of Graphs**

			FIGURE
$V_{IO}$	Input offset voltage	Distribution vs Common-mode voltage	1,2 3,4
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	5,6
$I_B/I_{IO}$	Input bias and input offset current	vs Free-air temperature	7
$V_I$	Input voltage range	vs Supply voltage vs Free-air temperature	8 9
$V_{OH}$	High-level output voltage	vs Output current	10
$V_{OL}$	Low-level output voltage	vs Output current	11,12
$V_{OM+}$	Maximum positive peak output voltage	vs Output current	13
$V_{OM-}$	Maximum negative peak output voltage	vs Output current	14
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	15
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Free-air temperature	16 17
$V_O$	Output voltage	vs Differential Input voltage	18,19
$A_{VD}$	Differential voltage amplification	vs Load resistance vs Frequency vs Free-air temperature	20 21, 22 23, 24
$z_o$	Output impedance	vs Frequency	25, 26
CMRR	Common-mode rejection ratio	vs Frequency vs Free-air temperature	27 28
ksVR	Supply-voltage rejection ratio	vs Frequency vs Free-air temperature	29, 30 31
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	32 33
SR	Slew rate	vs Load capacitance vs Free-air temperature	34 35
$V_O$	Large-signal pulse response	vs Time	36, 37, 38, 39
$V_O$	Small-signal pulse response	vs Time	40, 41, 42, 43
$V_n$	Equivalent input noise voltage	vs Frequency	44, 45
	Noise voltage (referred to input)	Over a 10-second period	46
	Integrated noise voltage	vs Frequency	47
THD + N	Total harmonic distortion plus noise	vs Frequency	48
	Gain-bandwidth product	vs Free-air temperature vs Supply voltage	49 50
$\phi_m$	Phase margin	vs Load capacitance vs Frequency	51 21, 22
	Gain margin	vs Load capacitance	52

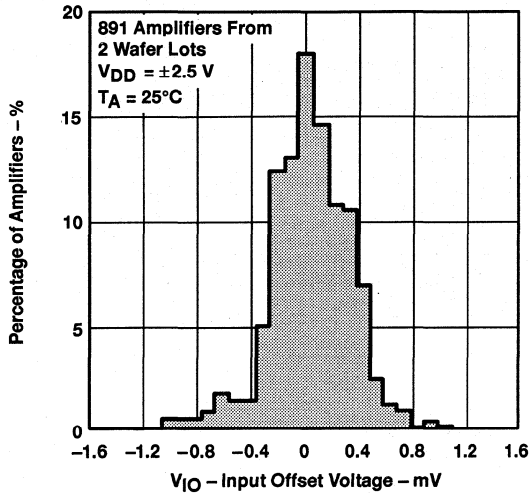
NOTE: For all graphs where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

**TLC2272, TLG2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

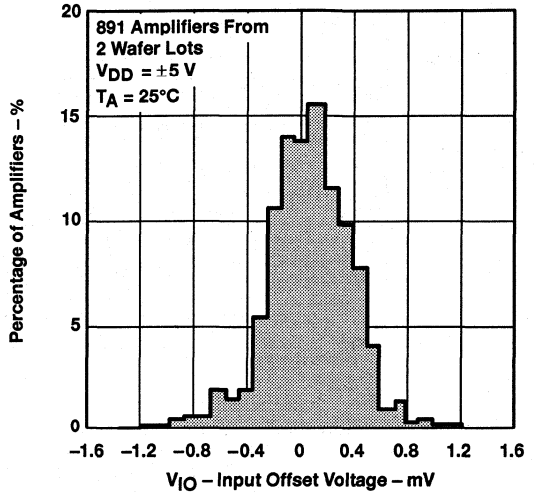
**TYPICAL CHARACTERISTICS**

**DISTRIBUTION OF TLC2272  
 INPUT OFFSET VOLTAGE**



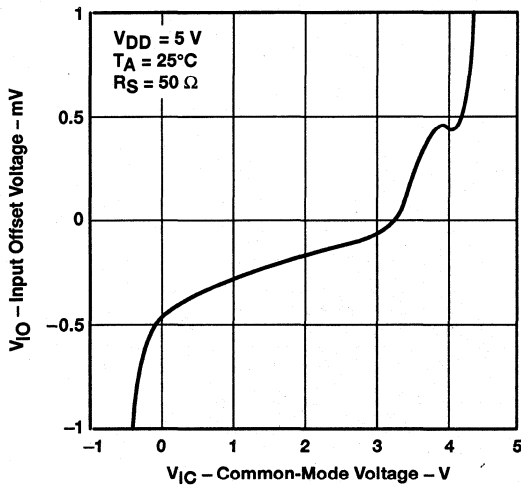
**Figure 1**

**DISTRIBUTION OF TLC2272  
 INPUT OFFSET VOLTAGE**



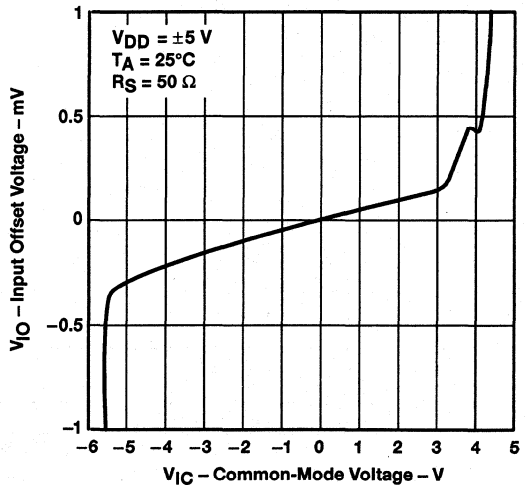
**Figure 2**

**INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE VOLTAGE**



**Figure 3**

**INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE VOLTAGE**



**Figure 4**

TYPICAL CHARACTERISTICS†

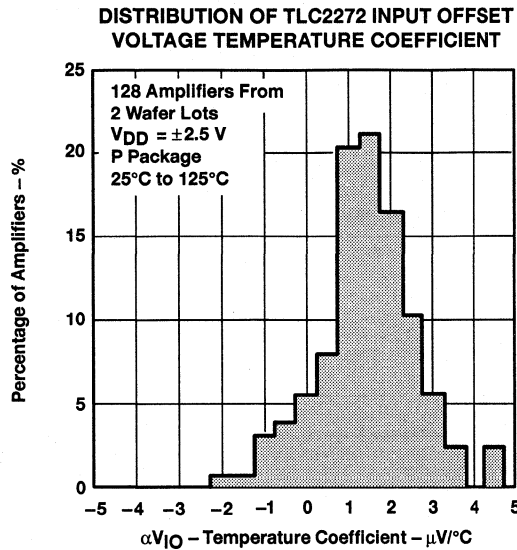


Figure 5

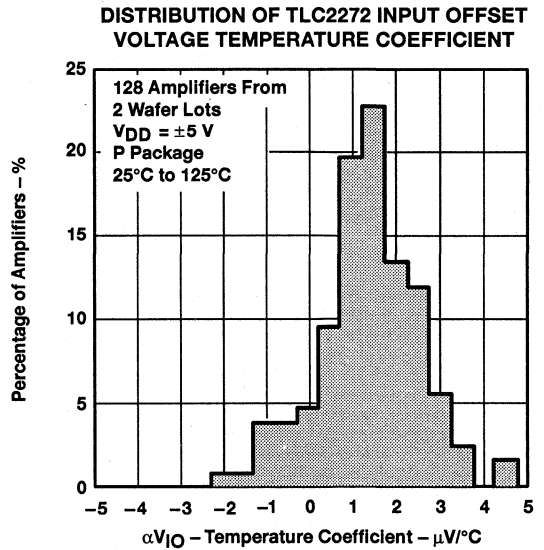


Figure 6

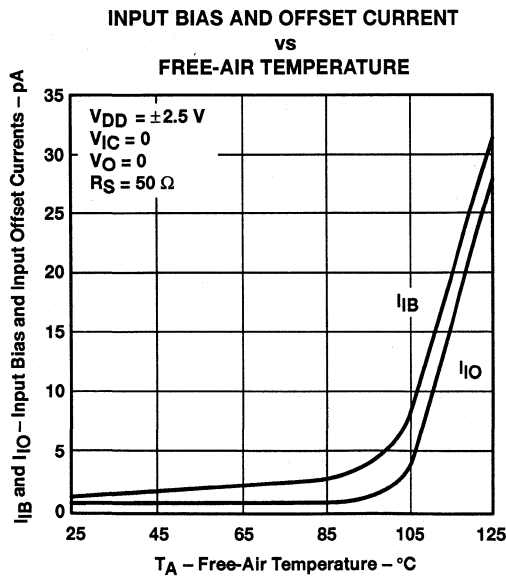


Figure 7

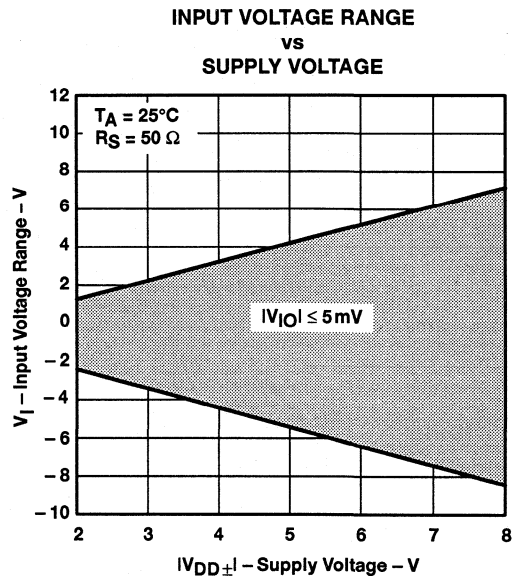


Figure 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

INPUT VOLTAGE RANGE  
 vs  
 FREE-AIR TEMPERATURE

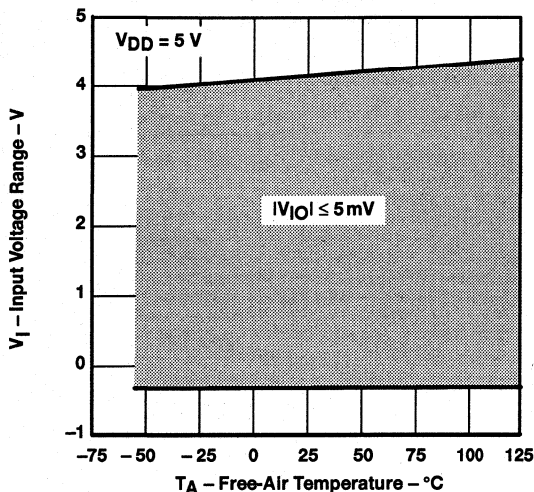


Figure 9

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

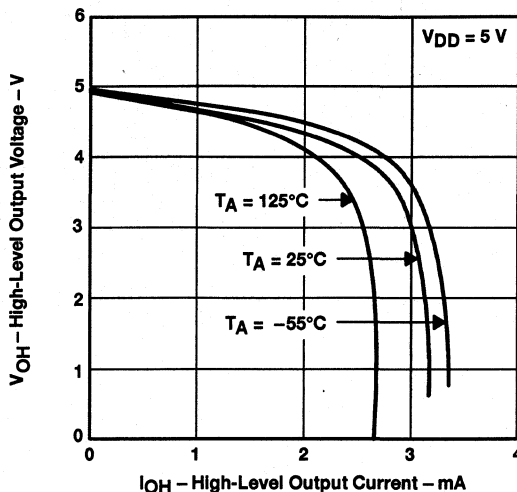


Figure 10

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

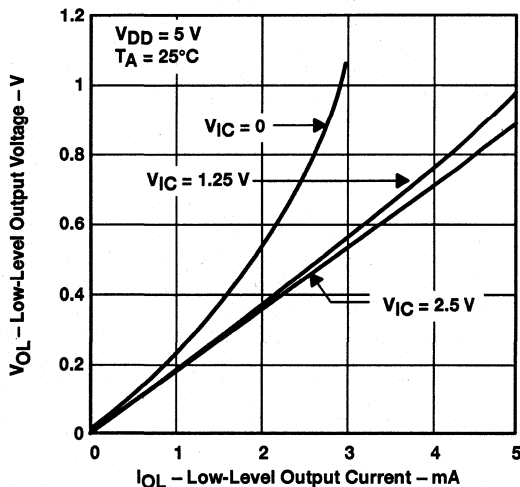


Figure 11

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

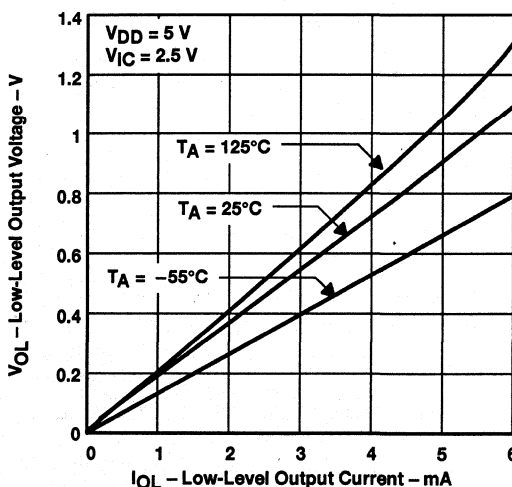


Figure 12

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

MAXIMUM POSITIVE PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

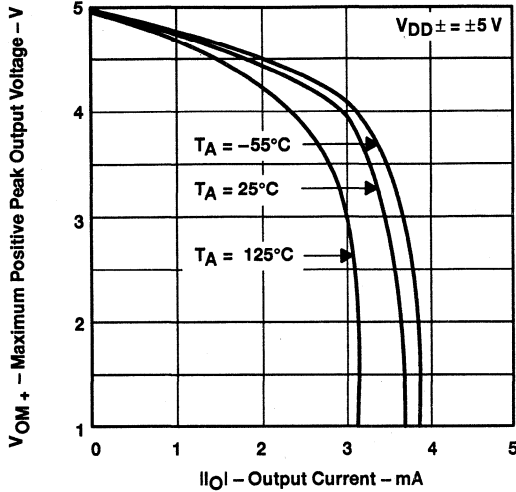


Figure 13

MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

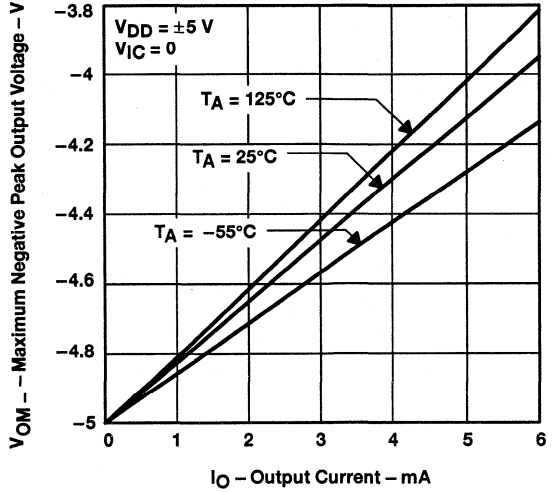


Figure 14

MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE  
 vs  
 FREQUENCY

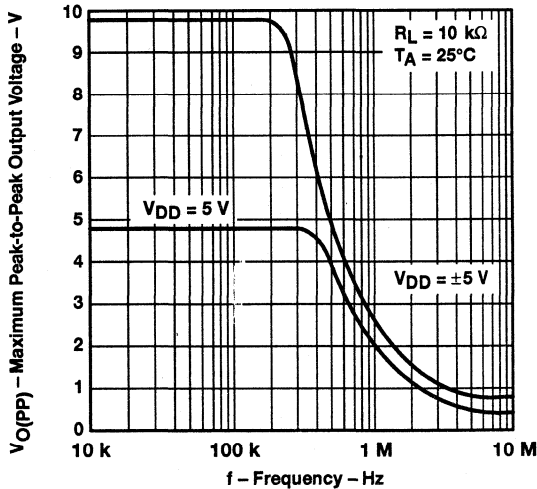


Figure 15

SHORT-CIRCUIT OUTPUT CURRENT  
 vs  
 SUPPLY VOLTAGE

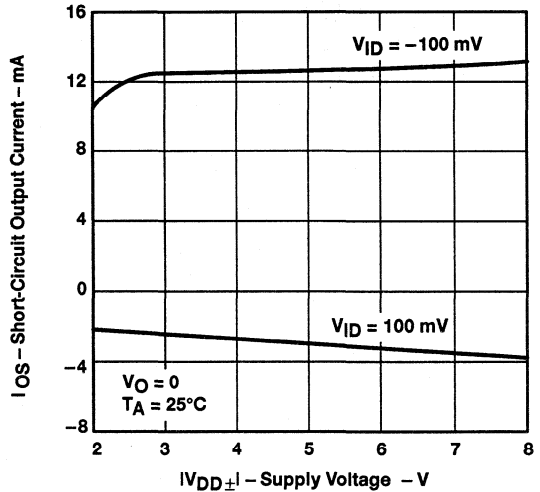
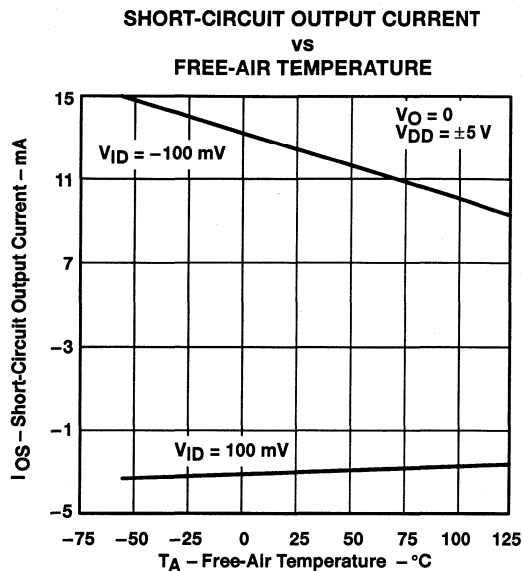


Figure 16

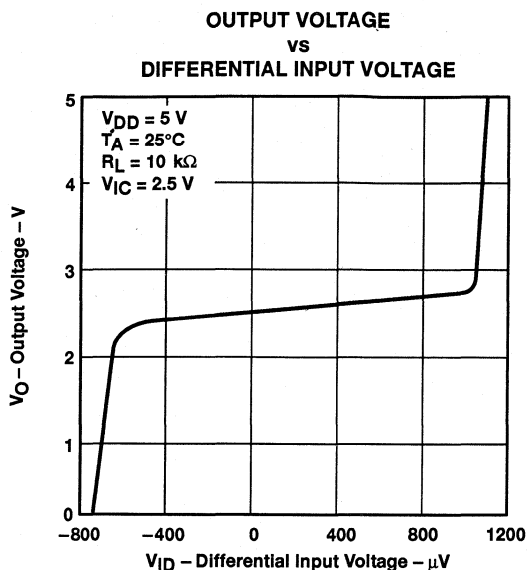
**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

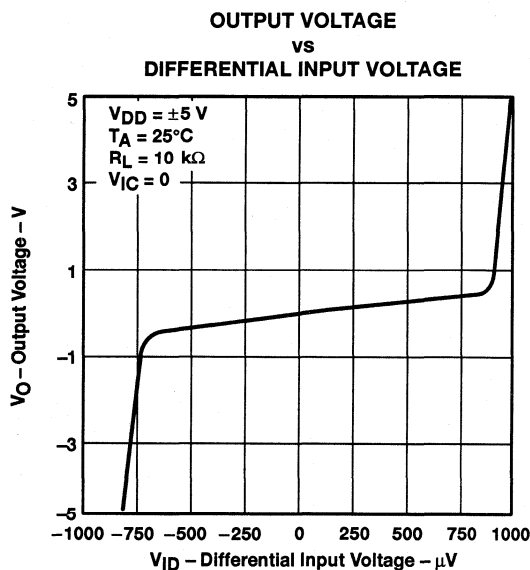
**TYPICAL CHARACTERISTICS†**



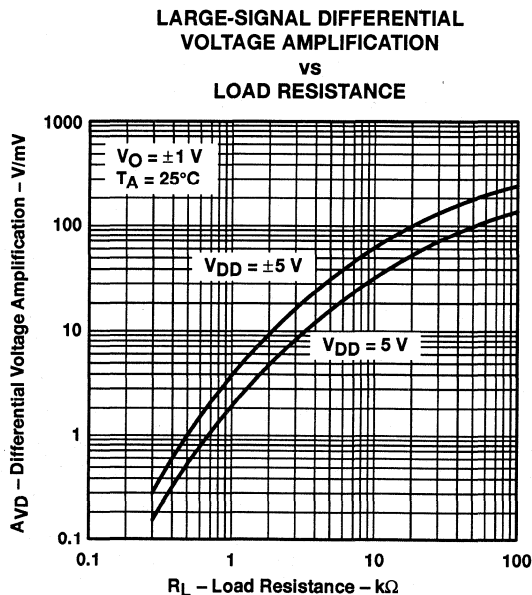
**Figure 17**



**Figure 18**



**Figure 19**



**Figure 20**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE MARGIN  
 vs  
 FREQUENCY

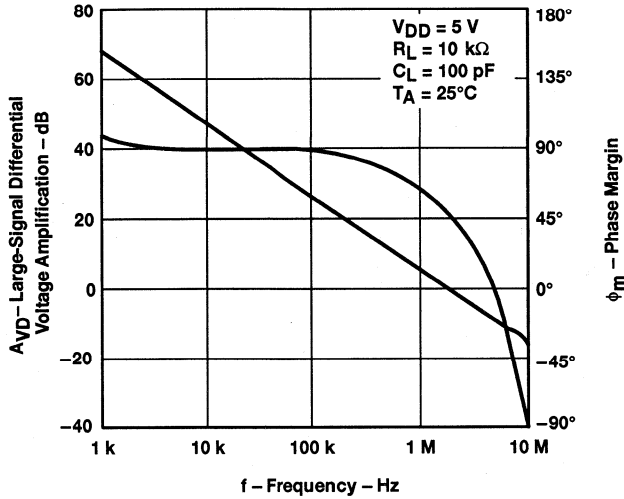


Figure 21

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE MARGIN  
 vs  
 FREQUENCY

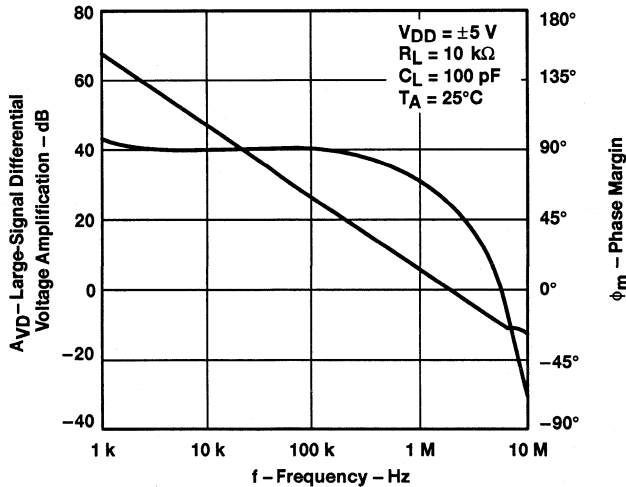


Figure 22

TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

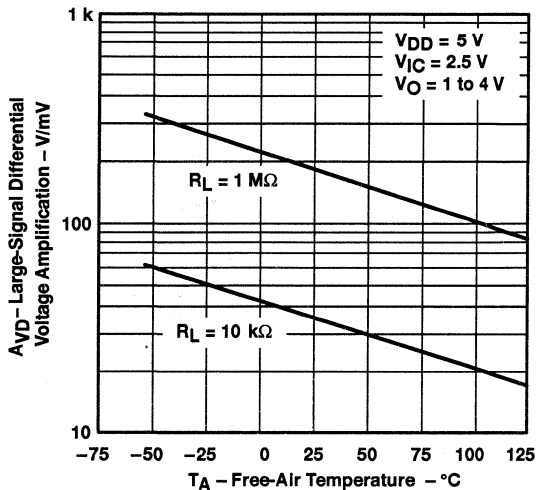


Figure 23

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

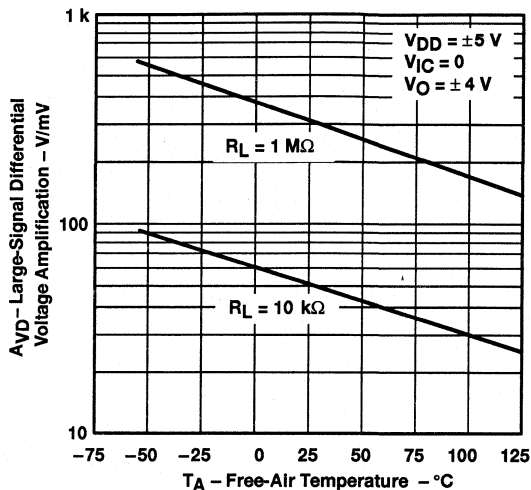


Figure 24

OUTPUT IMPEDANCE  
 vs  
 FREQUENCY

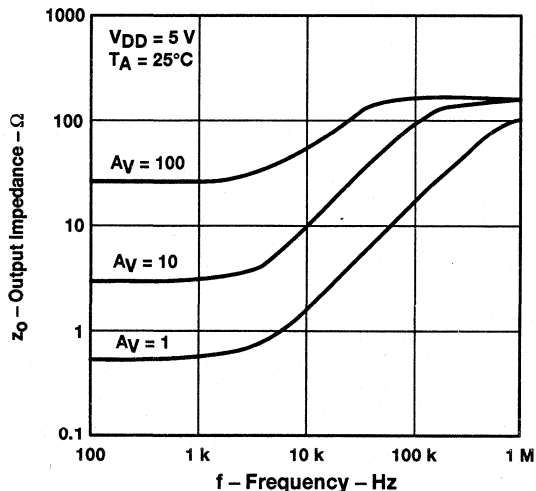


Figure 25

OUTPUT IMPEDANCE  
 vs  
 FREQUENCY

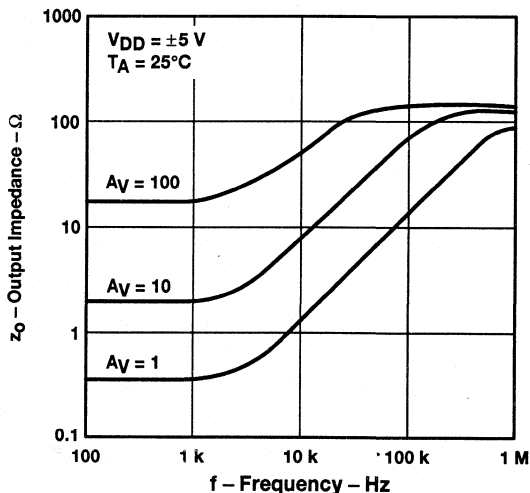


Figure 26

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

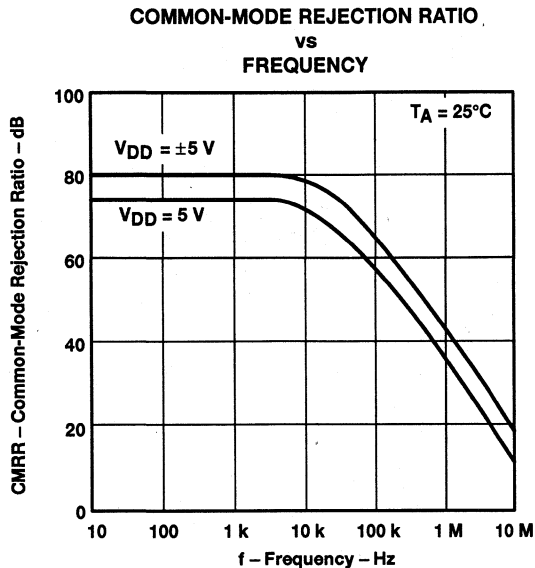


Figure 27

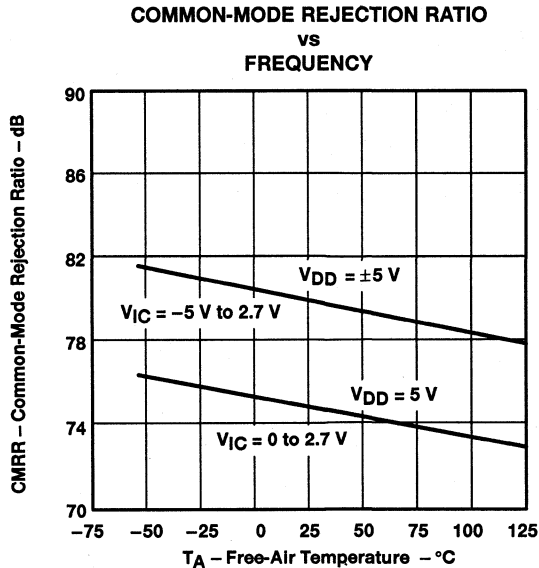


Figure 28

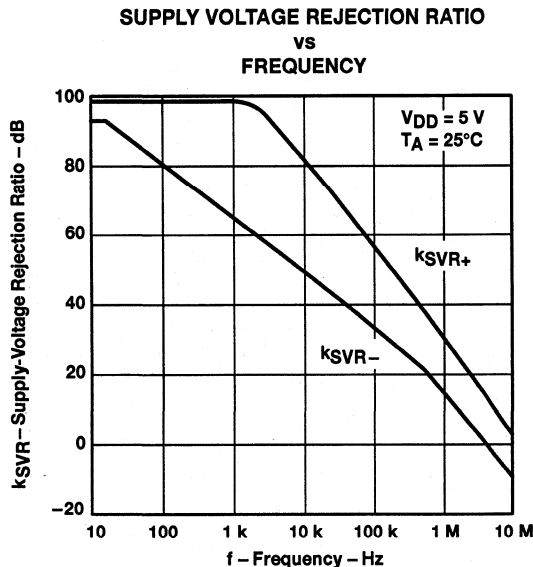


Figure 29

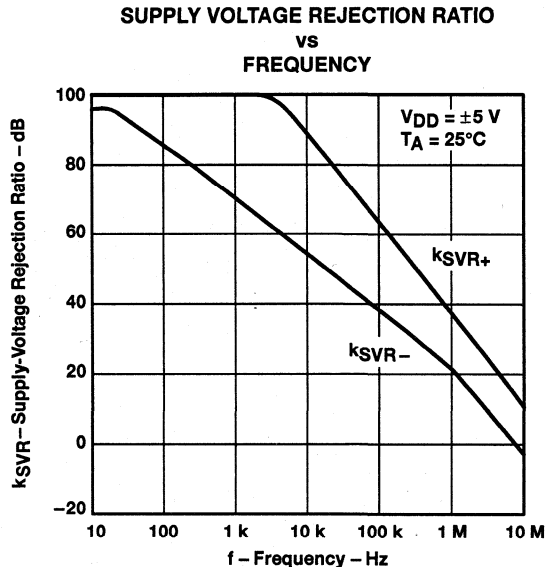


Figure 30

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

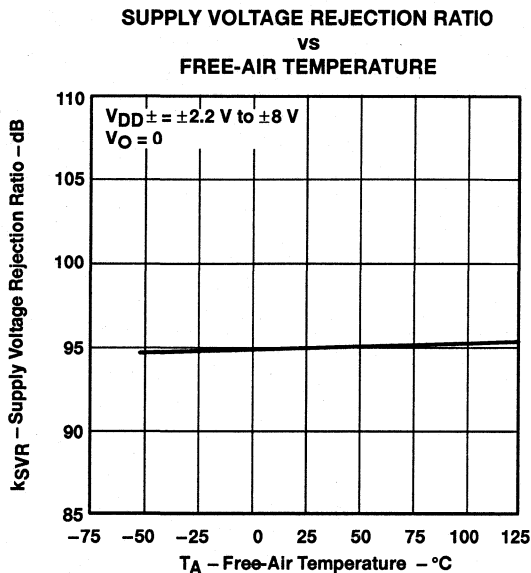


Figure 31

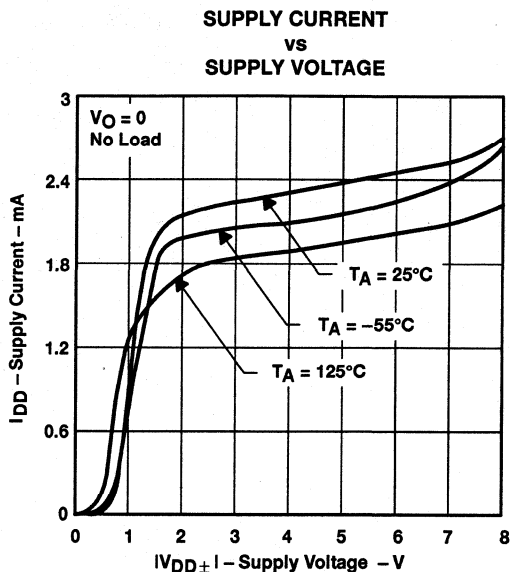


Figure 32

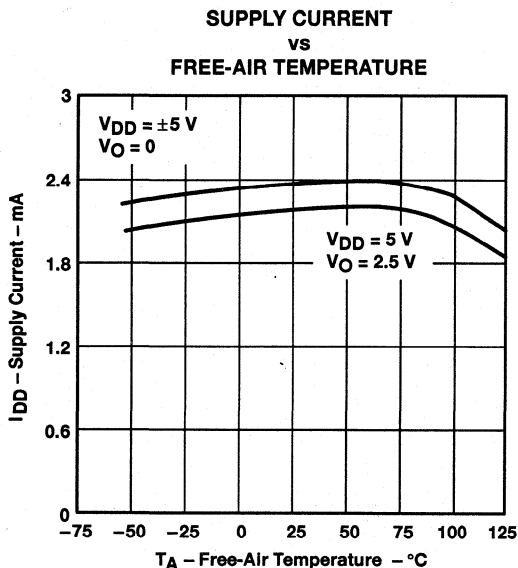


Figure 33

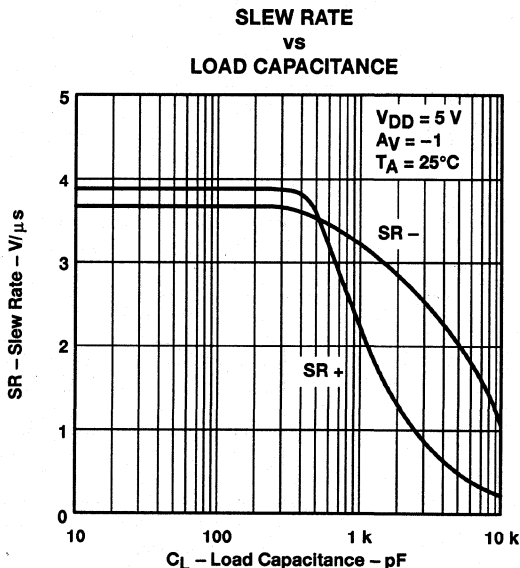
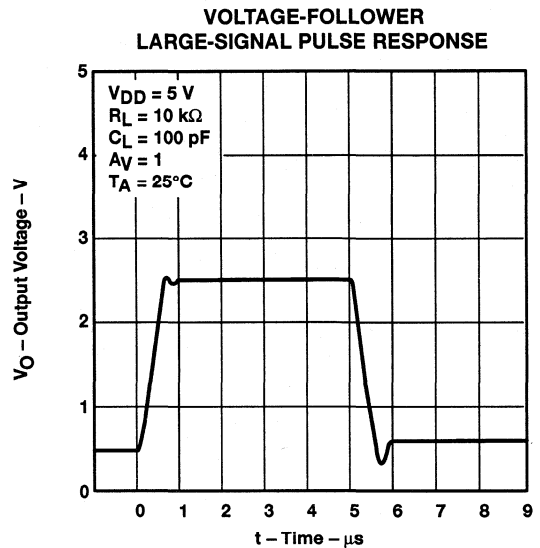
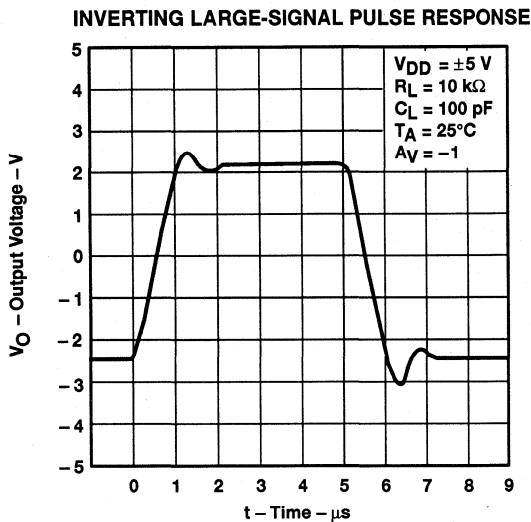
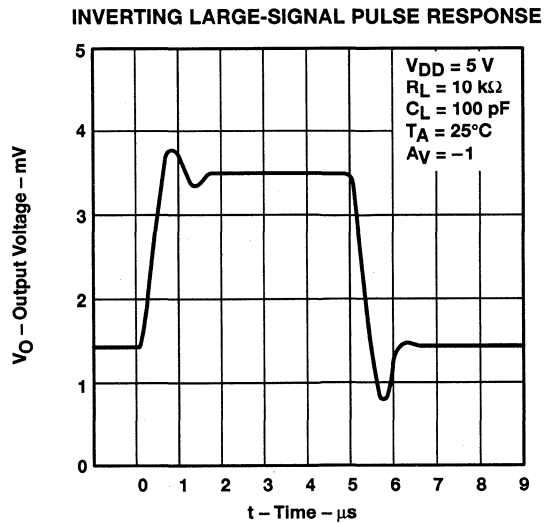
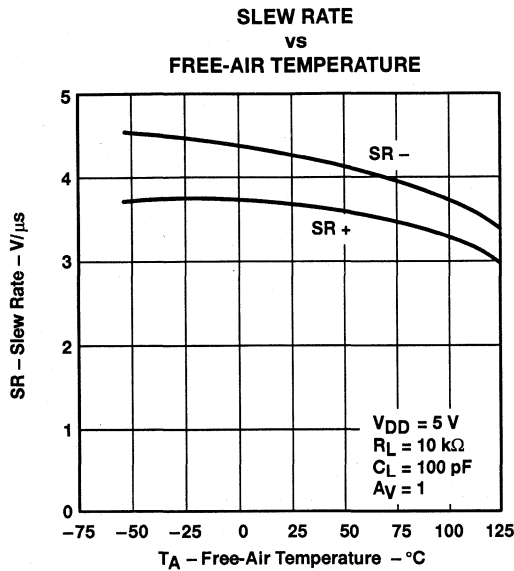


Figure 34

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

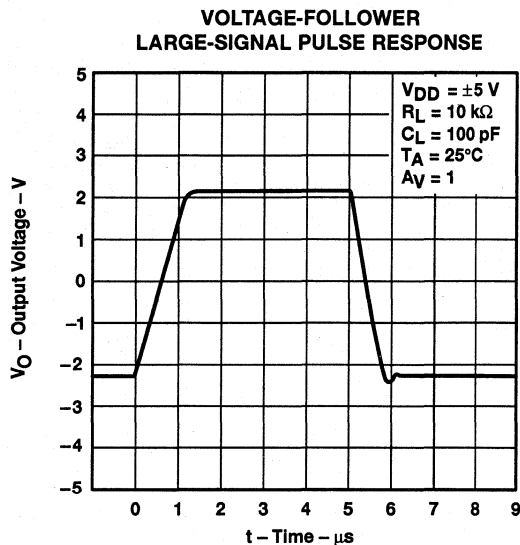


Figure 39

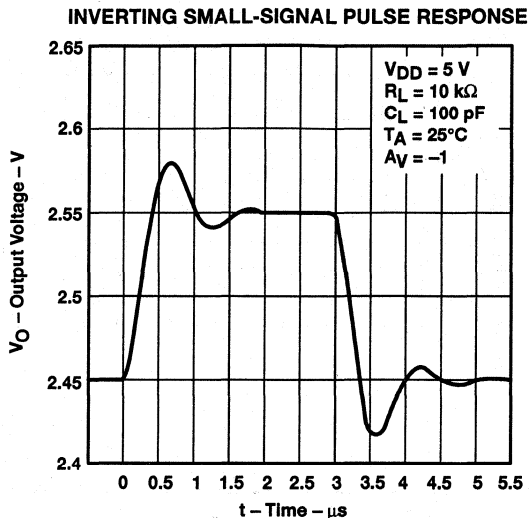


Figure 40

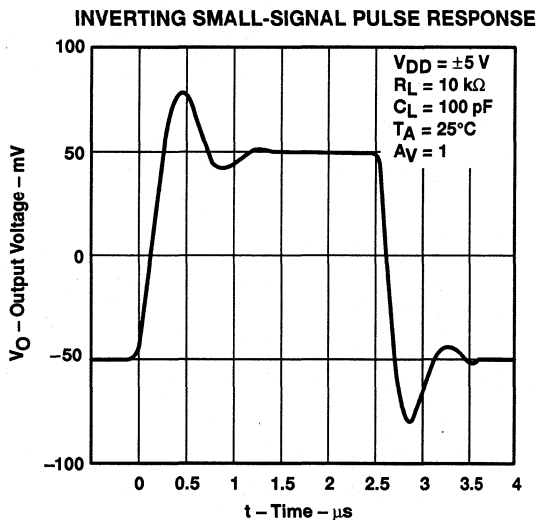


Figure 41

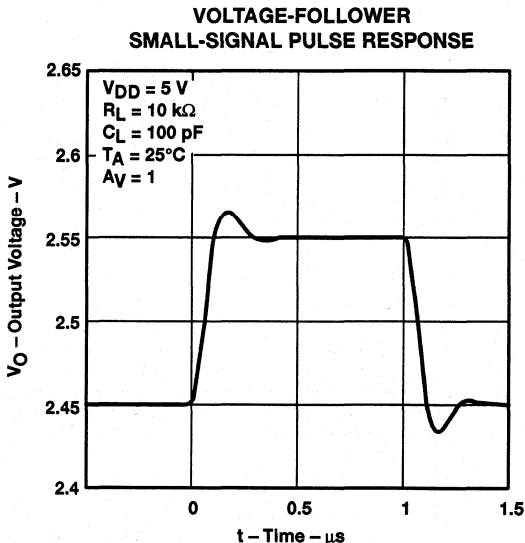


Figure 42

TYPICAL CHARACTERISTICS

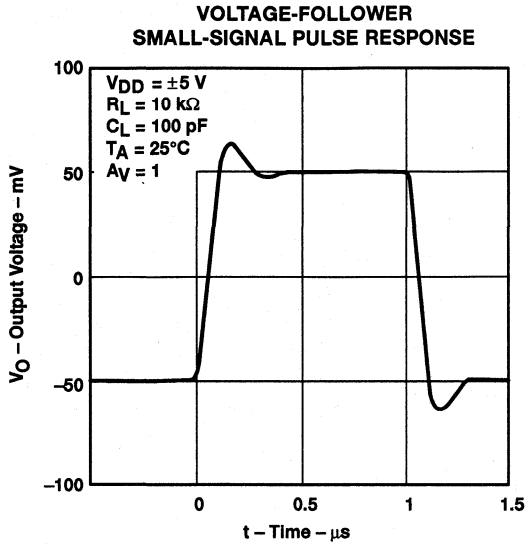


Figure 43

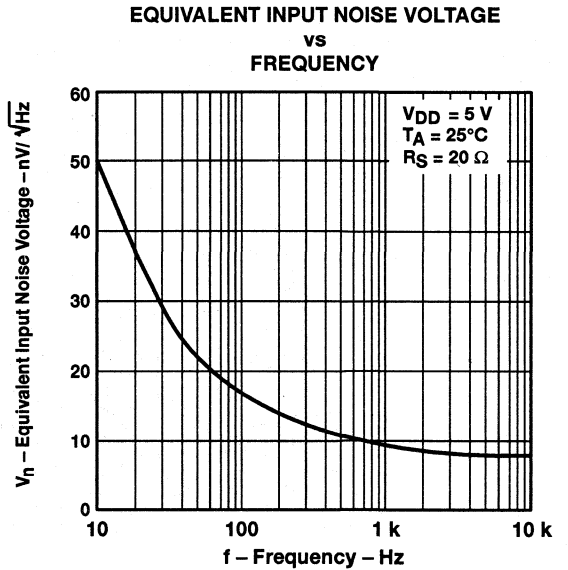


Figure 44

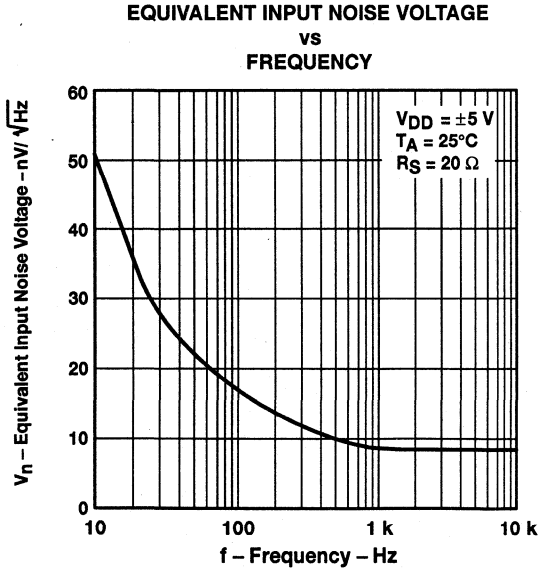


Figure 45

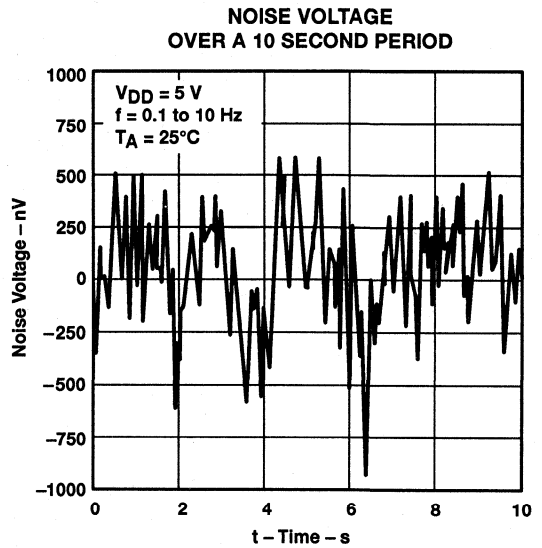


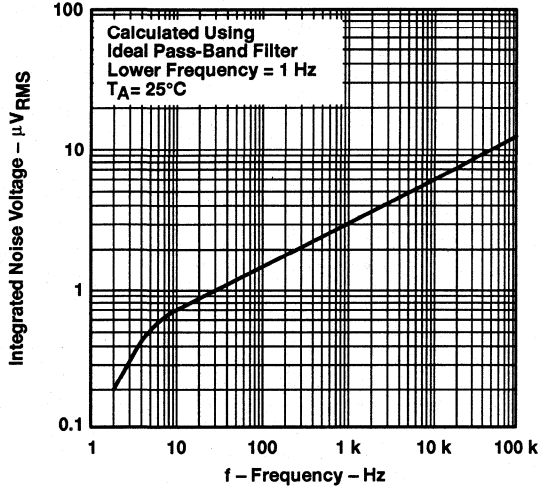
Figure 46

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

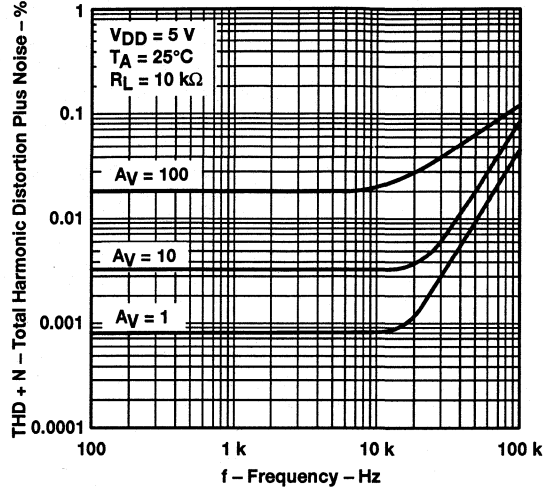
**TYPICAL CHARACTERISTICS†**

**INTEGRATED NOISE VOLTAGE  
 VS  
 FREQUENCY**



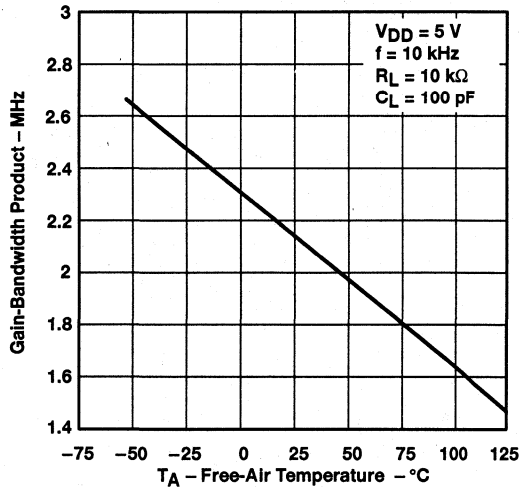
**Figure 47**

**TOTAL HARMONIC DISTORTION PLUS NOISE  
 VS  
 FREQUENCY**



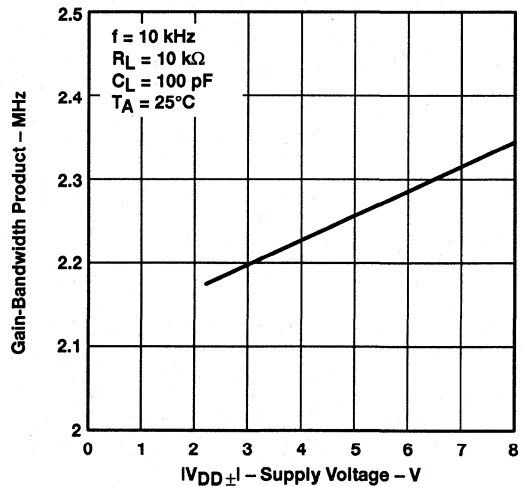
**Figure 48**

**GAIN-BANDWIDTH PRODUCT  
 VS  
 FREE-AIR TEMPERATURE**



**Figure 49**

**GAIN-BANDWIDTH PRODUCT  
 VS  
 FREE-AIR TEMPERATURE**



**Figure 50**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

PHASE MARGIN  
 vs  
 LOAD CAPACITANCE

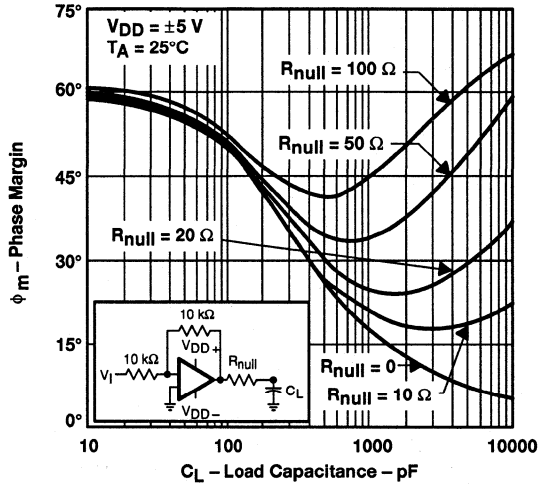


Figure 51

GAIN MARGIN  
 vs  
 LOAD CAPACITANCE

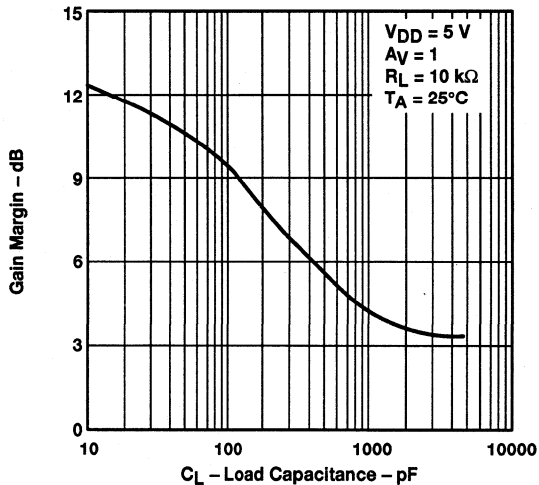


Figure 52

**TLC2272, TLC2272A, TLC2272Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS102C – NOVEMBER 1991 – REVISED APRIL 1994

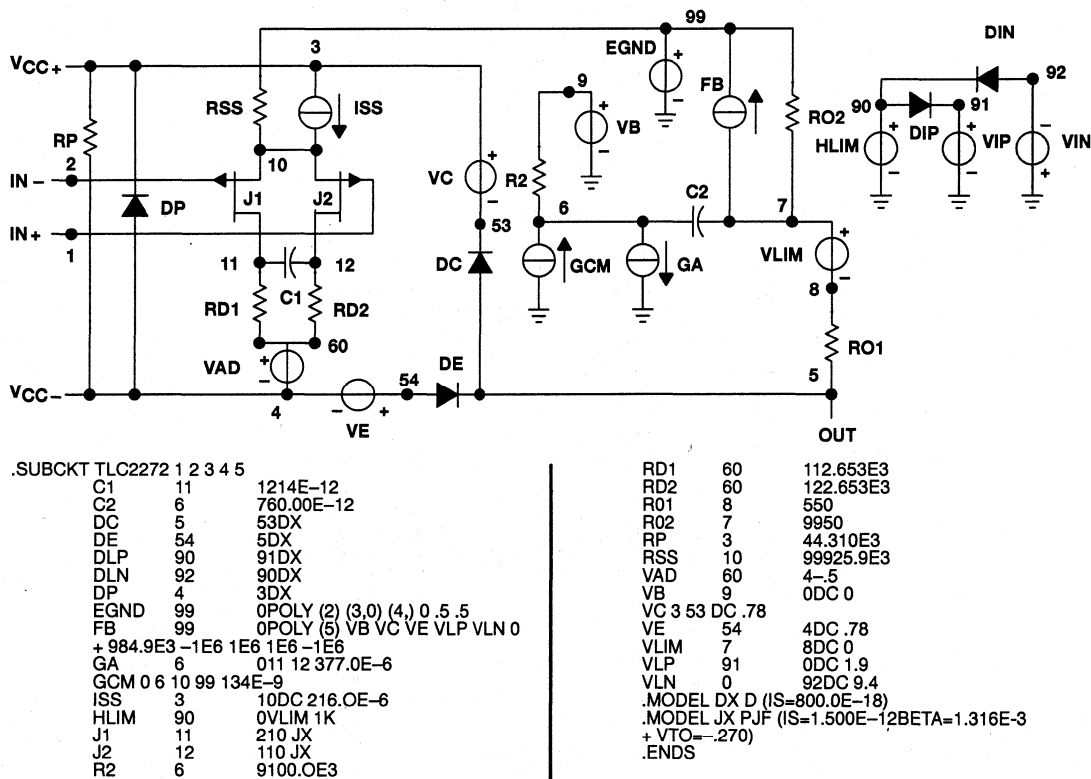
**APPLICATION INFORMATION**

**macromodel information**

Macromodel information provided was derived using *PSpice™ Parts™* model generation software. The Boyle macromodel (see Note 5) and subcircuit in Figure 53 were generated using the TLC2272 typical electrical and operating characteristics at  $T_A = 25^\circ\text{C}$ . Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 5: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).



**Figure 53. Boyle Macromodel and Subcircuit**

*PSpice* and *Parts* are registered trademarks of MicroSim Corporation.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# TLC2274, TLC2274A, TLC2274Y Advanced LinCMOS™ RAIL-TO-RAIL QUAD OPERATIONAL AMPLIFIERS

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

- Output Swing Includes Both Supply Rails
- Low Noise . . . 9 nV/√Hz Typ at f = 1 kHz
- Low Input Bias Current . . . 1 pA Typ
- Fully Specified for Both Single-Supply and Split-Supply Operation
- Common-Mode Input Voltage Range Includes Negative Rail
- High Unity-Gain Bandwidth  
2.18-MHz Typ Single Supply  
2.25-MHz Typ Split Supply
- High Slew Rate . . . 3.6 V/μs Typ
- Low Input Offset Voltage  
950 μV Max at T<sub>A</sub> = 25°C
- Macromodel Included

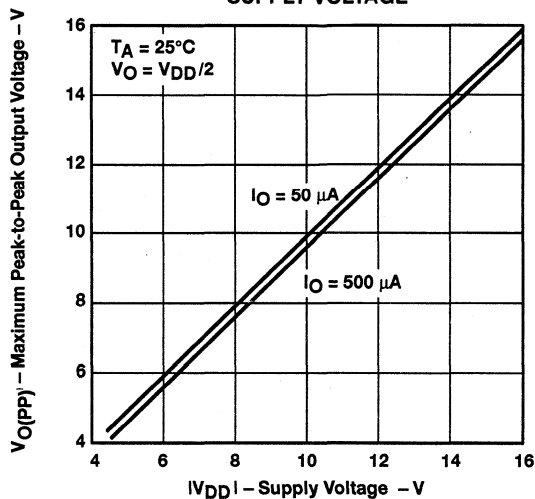
## description

The TLC2274 and TLC2274A are quad rail-to-rail operational amplifiers manufactured using Texas Instruments Advanced LinCMOS™ process. These devices offer comparable ac performance while having better noise, input offset voltage, and power dissipation than existing CMOS operational amplifiers. In addition, the common-mode input voltage range is wider than typical standard CMOS amplifiers. To take advantage of this improvement in performance and making this device available for a wider range of applications, V<sub>ICR</sub> is specified with a larger maximum input offset voltage test limit of ±5 mV. The Advanced LinCMOS™ process uses a silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. This technology also makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The TLC2274 and TLC2274A, exhibiting high input impedance and low noise, are excellent for small-signal conditioning for high-impedance sources such as piezoelectric transducers. In addition, the rail-to-rail output feature with single or split supply makes these devices great choices for inputs to ADCs in either the unipolar or bipolar mode of operation. This feature, combined with its temperature performance, makes the TLC2274 family ideal for sonobuoys, pressure sensors, temperature control, active VR sensors, accelerometers, and many other applications.

The device inputs and outputs are designed to withstand a 100-mA surge current without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

MAXIMUM PEAK-TO-PEAK  
OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE



Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

 **TEXAS  
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

2-957

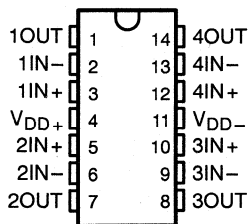
# TLC2274, TLC2274A, TLC2274Y

## Advanced LinCMOS™ RAIL-TO-RAIL

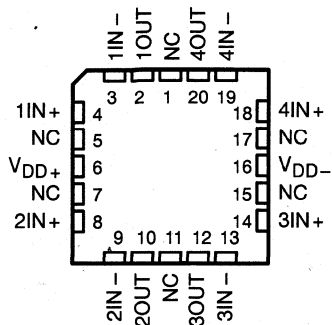
### QUAD OPERATIONAL AMPLIFIERS

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**D, J, N, OR PW PACKAGE  
(TOP VIEW)**



**FK PACKAGE  
(TOP VIEW)**



NC – No internal connection

#### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IOMax</sub> AT 25°C	PACKAGED DEVICES					CHIP FORM (Y)
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)	
0°C to 70°C	950 μV 2.5 mV	TLC2274ACD TLC2274CD	—	—	TLC2274ACN TLC2274CN	— TLC2274CPWLE	TLC2274Y
-40°C to 85°C	950 μV 2.5 mV	TLC2274AID TLC2274ID	—	—	TLC2274AIN TLC2274IN	— TLC2274IPWLE	—
-55°C to 125°C	950 μV 2.5 mV	TLC2274AMD TLC2274MD	TLC2274AMFK TLC2274MFK	TLC2274AMJ TLC2274MJ	TLC2274AMN TLC2274MN	—	—

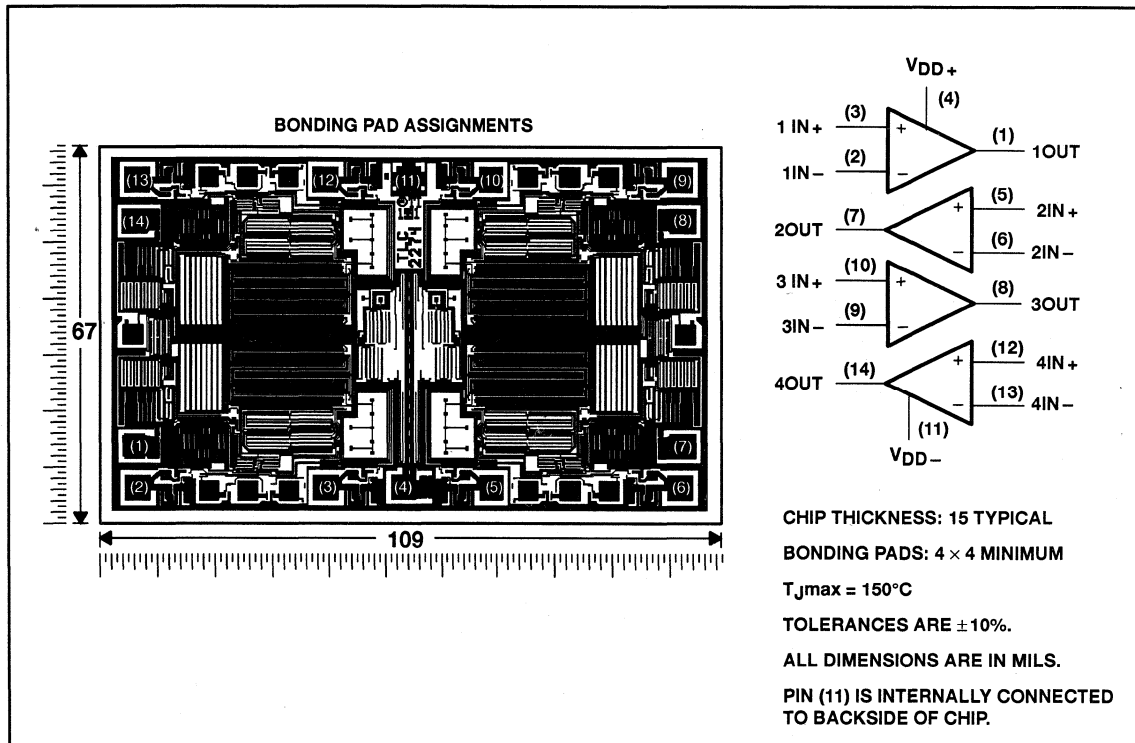
The D packages are available taped and reeled. Add R suffix to device type (e.g., TLE2274CDR). The PW package is available only left-end taped and reeled. Chips are tested at 25°C.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**TLC2274Y chip information**

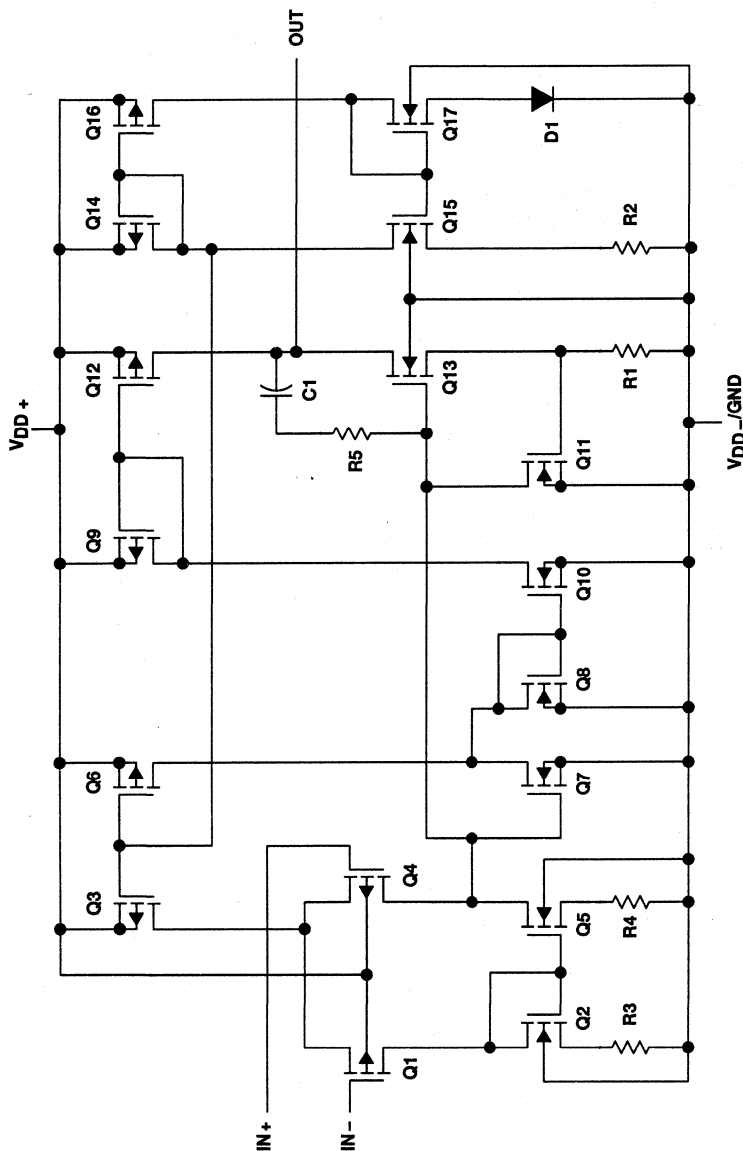
This chip, when properly assembled, displays characteristics similar to the TLC2274C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS108C – MARCH 1992 – REVISED AUGUST 1994

equivalent schematic



COMPONENT COUNT†	
Transistors	76
Diodes	18
Resistors	52
Capacitors	6

† Includes all four amplifiers and all ESD, bias, and trim circuitry.



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 16$ V
Input voltage, $V_I$ (any input) (see Note 1)	$\pm 8$ V
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$	$\pm 50$ mA
Total current into $V_{DD+}$	$\pm 50$ mA
Total current out of $V_{DD-}$	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at  $IN+$  with respect to  $IN-$ . Excessive current will flow if input is brought below  $V_{DD-} - 0.3$  V.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING	ABOVE $T_A = 25^\circ\text{C}$	POWER RATING	POWER RATING	POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	230 mW
PW	700 mW	5.6 mW/°C	448 mW	364 mW	—

**recommended operating conditions**

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	$\pm 2.2$	$\pm 8$	$\pm 2.2$	$\pm 8$	$\pm 2.2$	$\pm 8$	V
Input voltage, $V_I$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274C			TLC2274AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	300		2500	300		950	$\mu\text{V}$
		Full range	3000			1500			
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		25°C to 70°C	2			2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{DD} \pm \pm 2.5\text{ V}$ , $V_{IC} = 0$ , $V_O = 0$ , $R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$
		Full range	100			100			
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$
		Full range	100			100			
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2		V
		Full range	0 to 3.5		0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99		4.99				V
		Full range	4.85		4.85				
	$I_{OH} = -1\text{ mA}$	25°C	4.25		4.65				
		Full range	4.25		4.25				
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$	25°C	0.01		0.01				V
		Full range	0.09		0.15		0.09 0.15		
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$	25°C	0.9		1.5		0.9 1.5		
		Full range	1.5		1.5				
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 5\text{ mA}$	25°C	15		35		15 35		
Full range		15		15					
Full range		175		175					
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 10\text{ k}\Omega^\ddagger$	25°C	10 <sup>12</sup>		10 <sup>12</sup>		$\text{V}/\text{mV}$	
Full range			15		15				
		$R_L = 1\text{ m}\Omega^\ddagger$	25°C	175		175			
$r_{id}$ Differential input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		$\Omega$		
$r_i$ Common-mode input resistance		25°C	10 <sup>12</sup>		10 <sup>12</sup>		$\Omega$		
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$ , N package	25°C	8		8		$\text{pF}$		
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$	25°C	140		140		$\Omega$		
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$	25°C	70		75		70 75		dB
		Full range	70		70				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load	25°C	80		95		80 95		dB
		Full range	80		80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load	25°C	4.4		6		4.4 6		mA
		Full range	6		6				

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274C			TLC2274AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	2.3	3.6	2.3	3.6	V/ $\mu\text{s}$	
			Full range	1.7			1.7		
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C	50			50	nV/ $\sqrt{\text{Hz}}$	
			25°C	9			9		
$V_N(\text{PP})$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }10\text{ Hz}$	25°C	1			1	$\mu\text{V}$	
			25°C	1.4			1.4		
$I_n$	Equivalent input noise current		25°C	0.6			0.6	fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega^\ddagger$	25°C	$A_V = 1$	0.0013%			0.0013%	
				$A_V = 10$	0.004%			0.004%	
				$A_V = 100$	0.03%			0.03%	
	Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}^\ddagger$	$R_L = 10\text{ k}\Omega^\ddagger$	25°C	2.18			2.18	MHz
BOM	Maximum output-swinging bandwidth	$V_O(\text{PP}) = 2\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$	$A_V = 1$ , $C_L = 100\text{ pF}^\ddagger$	25°C	1			1	MHz
$t_s$	Settling time	$A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	To 0.1%	25°C	1.5			1.5	$\mu\text{s}$
			To 0.01%		2.6			2.6	
$\phi_m$	Phase margin at unity gain	$R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	50°			50°		
	Gain margin		25°C	10			10		dB

† Full range is 0°C to 70°C.

‡ Referenced to 2.5 V

# TLC2274, TLC2274A, TLC2274Y

## Advanced LinCMOS™ RAIL-TO-RAIL QUAD OPERATIONAL AMPLIFIERS

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274C			TLC2274AC			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C	300 2500			300 950			$\mu\text{V}$	
		Full range	3000			1500				
$\alpha V_{IO}$ Temperature coefficient of input offset voltage		25°C to 70°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.002			0.002			$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$	
		Full range	100			100				
$I_{IB}$ Input bias current	25°C	1			1			$\text{pA}$		
	Full range	100			100					
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\text{ mV}$	25°C	-5 to 4	-5.3 to 4.2	-5 to 4	-5.3 to 4.2			V	
		Full range	-5 to 3.5		-5 to 3.5					
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C	4.99		4.99				V	
		25°C	4.85	4.93	4.85	4.93				
	Full range	4.85		4.85						
	$I_O = -1\text{ mA}$	25°C	4.25	4.65	4.25	4.65				
Full range		4.25		4.25						
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0, I_O = 50\ \mu\text{A}$	25°C	-4.99		-4.99				V	
		25°C	-4.85	-4.91	-4.85	-4.91				
	Full range	-4.85		-4.85						
	$V_{IC} = 0, I_O = 5\text{ mA}$	25°C	-3.5	-4.1	-3.5	-4.1				
Full range		-3.5		-3.5						
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}$	$R_L = 10\text{ k}\Omega$	25°C	25	50	25	50			V/mV
			Full range	25		25				
		$R_L = 1\text{ M}\Omega$	25°C	300		300				
$r_{id}$ Differential input resistance		25°C	$10^{12}$			$10^{12}$		$\Omega$		
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$		$\Omega$		
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}, \text{ N package}$	25°C	8			8		pF		
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}, A_V = 10$	25°C	130			130		$\Omega$		
CMRR Common-mode rejection ratio	$V_{IC} = -5\text{ V to } 2.7\text{ V}, V_O = 0, R_S = 50\ \Omega$	25°C	75	80	75	80			dB	
		Full range	75		75					
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\text{ V to } \pm 8\text{ V}, V_{IC} = 0, \text{ No load}$	25°C	80	95	80	95			dB	
		Full range	80		80					
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	4.8	6	4.8	6			mA	
		Full range	6		6					

† Full range is 0°C to 70°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274C			TLC2274AC			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
SR	Slew rate at unity gain $V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	2.3	3.6		2.3	3.6	$\text{V}/\mu\text{s}$		
		Full range	1.7			1.7				
$V_n$	Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C			50			$\text{nV}/\sqrt{\text{Hz}}$	
		$f = 1\text{ Hz}$	25°C			9				
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ Hz to }1\text{ Hz}$	25°C			1			$\mu\text{V}$	
		$f = 0.1\text{ Hz to }10\text{ Hz}$	25°C			1.4				
$I_n$	Equivalent input noise current	25°C				0.6			$\text{fA}/\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise $V_O = \pm 2.3\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega$	$A_V = 1$	25°C			0.0011%				
		$A_V = 10$				0.004%				
		$A_V = 100$				0.03%				
	Gain-bandwidth product	$f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C			2.25			MHz	
BOM	Maximum output-swing bandwidth	$V_{O(PP)} = 4.6\text{ V}$ , $A_V = 1$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C			0.54			MHz	
$t_s$	Settling time	$A_V = -1$ , Step = $-2.3\text{ V to }2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	To 0.1%	25°C			1.5			$\mu\text{s}$
			To 0.01%				3.2			
$\phi_m$	Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C			52°				
	Gain margin		25°C			10				dB

† Full range is 0°C to 70°C.

# TLC2274, TLC2274A, TLC2274Y

## Advanced LinCMOS™ RAIL-TO-RAIL QUAD OPERATIONAL AMPLIFIERS

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274I			TLC2274AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $V_O = 0,$ $V_{DD} = \pm 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	300	2500		300	950		$\mu\text{V}$
		Full range			3000		1500		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 85°C	2			2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.002			0.002			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			0.5			pA
		Full range	150			150			
$I_{IB}$ Input bias current	25°C	1			1			pA	
	Full range	150			150				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,$ $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2	V	
		Full range	0 to 3.5		0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99		4.99		V		
		Full range	4.85		4.85				
	$I_{OH} = -1\text{ mA}$	25°C	4.25	4.65	4.25	4.65			
		Full range	4.25		4.25				
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 50\ \mu\text{A}$	25°C	0.01		0.01		V		
		Full range	0.09 0.15		0.09 0.15				
	$V_{IC} = 2.5\text{ V},$ $I_{OL} = 500\ \mu\text{A}$	25°C	0.09 0.15		0.09 0.15				
		Full range	0.15		0.15				
$V_{IC} = 2.5\text{ V},$ $I_{OL} = 5\text{ mA}$	25°C	0.9	1.5	0.9	1.5				
	Full range	1.5		1.5					
	$V_{IC} = 2.5\text{ V},$ $V_O = 1\text{ V to }4\text{ V}$	25°C	15	35	15	35	V/mV		
		Full range	15		15				
$R_L = 10\text{ k}\Omega$ ‡	25°C	175			175				
	$R_L = 1\text{ M}\Omega$ ‡	175			175				
$r_{id}$ Differential input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>			$\Omega$
$r_i$ Common-mode input resistance		25°C	10 <sup>12</sup>			10 <sup>12</sup>			$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz},$ N package	25°C	8			8			pF
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz},$ $A_V = 10$	25°C	140			140			$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V},$ $V_O = 2.5\text{ V},$ $R_S = 50\ \Omega$	25°C	70	75	70	75	dB		
		Full range	70		70				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V},$ $V_{IC} = V_{DD}/2,$ No load	25°C	80	95	80	95	dB		
		Full range	80		80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V},$ No load	25°C	4.4	6	4.4	6	mA		
		Full range	6		6				

† Full range is -40°C to 85°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
SLOS108C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274I			TLC2274AI			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
SR	Slew rate at unity gain $V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	2.3	3.6		2.3	3.6		V/ $\mu\text{s}$	
		Full range	1.7			1.7				
$V_n$	Equivalent input noise voltage	f = 10 Hz	50			50			nV/ $\sqrt{\text{Hz}}$	
		f = 1 kHz	9			9				
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 1 Hz	1			1			$\mu\text{V}$	
		f = 0.1 Hz to 10 Hz	1.4			1.4				
$I_n$	Equivalent input noise current	25°C	0.6			0.6			fA/ $\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}$ , f = 20 kHz, $R_L = 10\text{ k}\Omega^\ddagger$	25°C	$A_V = 1$	0.0013%			0.0013%			
			$A_V = 10$	0.004%			0.004%			
			$A_V = 100$	0.03%			0.03%			
	Gain-bandwidth product	f = 10 kHz, $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	2.18			2.18			MHz
BOM	Maximum output-swing bandwidth	$V_{O(PP)} = 2\text{ V}$ , $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	1			1			MHz
$t_s$	Settling time	$A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	To 0.1%	1.5			1.5			$\mu\text{s}$
			To 0.01%	2.6			2.6			
$\phi_m$	Phase margin at unity gain	$R_L = 10\text{ k}\Omega^\ddagger$ , $C_L = 100\text{ pF}^\ddagger$	25°C	50°			50°			
	Gain margin		25°C	10			10			

† Full range is – 40°C to 85°C.

‡ Referenced to 2.5 V

# TLC2274, TLC2274A, TLC2274Y

## Advanced LinCMOS™ RAIL-TO-RAIL

### QUAD OPERATIONAL AMPLIFIERS

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274I			TLC2274AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	300	2500		300	950	$\mu\text{V}$	
		Full range			3000		1500		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 85°C	2			2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$
		Full range			150		150		
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$
		Full range			150		150		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	25°C	-5 to 4	-5.3 to 4.2		-5 to 4	-5.3 to 4.2	V	
		Full range	-5 to 3.5			-5 to 3.5			
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C	4.99		4.99		V		
	$I_O = -200\ \mu\text{A}$	25°C	4.85	4.93	4.85	4.93			
		Full range	4.85		4.85				
	$I_O = -1\ \text{mA}$	25°C	4.25	4.65	4.25	4.65			
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0, I_O = 50\ \mu\text{A}$	25°C	-4.99		-4.99		V		
	$V_{IC} = 0, I_O = 500\ \mu\text{A}$	25°C	-4.85	-4.91	-4.85	-4.91			
		Full range	-4.85		-4.85				
	$V_{IC} = 0, I_O = 5\ \text{mA}$	25°C	-3.5	-4.1	-3.5	-4.1			
$AVD$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 10\ \text{k}\Omega$	25°C	25	50	25	50	V/mV	
			Full range	25		25			
		$R_L = 1\ \text{M}\Omega$	25°C	300			300		
$r_{id}$ Differential input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, \text{N package}$	25°C	8			8			$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz}, A_v = 10$	25°C	130			130			$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{to}\ 2.7\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	75	80		75	80	dB	
		Full range	75			75			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} = \pm 2.2\ \text{V to}\ \pm 8\ \text{V}, V_{IC} = 0, \text{No load}$	25°C	80	95		80	95	dB	
		Full range	80			80			
$I_{DD}$ Supply current	$V_O = 0, \text{No load}$	25°C	4.8	6		4.8	6	mA	
		Full range	6			6			

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274I			TLC2274AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = \pm 2.3\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C	2.3	3.6		2.3	3.6		$\text{V}/\mu\text{s}$
		Full range	1.7			1.7			
$V_n$	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		$\text{nV}/\sqrt{\text{Hz}}$
		25°C		9			9		
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to } 1\text{ Hz}$ $f = 0.1\text{ Hz to } 10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
$I_n$	Equivalent input noise current	25°C		0.6			0.6		$\text{fA}/\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $f = 20\text{ kHz}$	25°C	$A_V = 1$	0.0011%		0.0011%			
			$A_V = 10$	0.004%		0.004%			
			$A_V = 100$	0.03%		0.03%			
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C		2.25			2.25		MHz
$B_{OM}$	Maximum output-swing bandwidth $V_{O(PP)} = 4.6\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $A_V = 1$ , $C_L = 100\text{ pF}$	25°C		0.54			0.54		MHz
$t_s$	Settling time $A_V = -1$ , Step = $-2.3\text{ V to } 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	To 0.1%	1.5			1.5		$\mu\text{s}$
			To 0.01%	3.2			3.2		
$\phi_m$	Phase margin at unity gain $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		52°			52°		
		25°C		10			10		dB

† Full range is  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ .



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274M			TLC2274AM			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
$V_{IO}$ Input offset voltage		25°C	300 2500			300 950			$\mu\text{V}$	
		Full range	3000			1500				
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 125°C	2			2			$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)	$V_{DD} \pm = \pm 2.5\text{ V}$ , $V_{IC} = 0$ , $V_O = 0$ , $R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$	
		25°C	0.5			0.5				
$I_{IO}$ Input offset current		25°C	1			1			$\text{pA}$	
		Full range	500			500				
$I_{IB}$ Input bias current		25°C	1			1			$\text{pA}$	
		Full range	500			500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2		0 to 4	-0.3 to 4.2		V	
		Full range	0 to 3.5			0 to 3.5				
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.99			4.99			V	
		25°C	4.85	4.93		4.85	4.93			
		Full range	4.85			4.85				
		25°C	4.25	4.65		4.25	4.65			
$V_{OL}$ Low-level output voltage	$I_{OL} = -1\text{ mA}$	25°C	0.01			0.01			V	
		25°C	0.09	0.15		0.09	0.15			
		Full range	0.15			0.15				
		25°C	0.9	1.5		0.9	1.5			
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 10\text{ k}\Omega$ †	25°C	10	35		10	35		V/mV
			Full range	10			10			
		$R_L = 1\text{ M}\Omega$ †	25°C	175			175			
			Full range	1.5			1.5			
$r_{id}$ Differential input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$	
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$ , N package	25°C	8			8			$\text{pF}$	
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$	25°C	140			140			$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $V_O = 2.5\text{ V}$ , $R_S = 50\ \Omega$	25°C	70	75		70	75		dB	
		Full range	70			70				
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16$ , $V_{IC} = V_{DD}/2$ , No load	25°C	80	95		80	95		dB	
		Full range	80			80				
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load	25°C	4.4		6		4.4		mA	
		Full range	6			6				

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274M			TLC2274AM			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX		
SR	Slew rate at unity gain $V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	2.3	3.6		2.3	3.6		$\text{V}/\mu\text{s}$	
		Full range	1.7			1.7				
$V_n$	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C	50			50			$\text{nV}/\sqrt{\text{Hz}}$	
		25°C	9			9				
$V_N(\text{PP})$	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to }1\text{ Hz}$ $f = 0.1\text{ Hz to }10\text{ Hz}$	25°C	1			1			$\mu\text{V}$	
		25°C	1.4			1.4				
$I_n$	Equivalent input noise current	25°C	0.6			0.6			$\text{fA}/\sqrt{\text{Hz}}$	
THD + N	Total harmonic distortion plus noise $V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ ‡	25°C	$A_V = 1$	0.0013%			0.0013%			
			$A_V = 10$	0.004%			0.004%			
			$A_V = 100$	0.03%			0.03%			
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ ‡	$R_L = 10\text{ k}\Omega$ ‡	25°C	2.18			2.18			MHz
BOM	Maximum output-swing bandwidth $V_O(\text{PP}) = 2\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡	$A_V = 1$ , $C_L = 100\text{ pF}$ ‡	25°C	1			1			MHz
$t_s$	Settling time $A_V = -1$ , Step = $0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	To 0.1%	25°C	1.5			1.5			$\mu\text{s}$
		To 0.01%		2.6			2.6			
$\phi_m$	Phase margin at unity gain $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	50°			50°				
	Gain margin	25°C	10			10			dB	

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

‡ Referenced to  $2.5\text{ V}$

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274M			TLC2274AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage		25°C	300 2500			300 950			$\mu\text{V}$
		Full range	3000			1500			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 125°C	2			2			$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)	$V_{IC} = 0, V_O = 0, R_S = 50\ \Omega$	25°C	0.002			0.002			$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	0.5			0.5			$\text{pA}$
$I_{IB}$ Input bias current		Full range	500			500			
		25°C	1			1			$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,  V_{IO}  \leq 5\ \text{mV}$	Full range	500			500			
		25°C	1			1			
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$	25°C	4.99			4.99			$\text{V}$
		Full range	4.85			4.85			
$V_{OM-}$ Maximum negative peak output voltage	$I_O = -200\ \mu\text{A}$	25°C	4.85 4.93			4.85 4.93			
		Full range	4.85			4.85			
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0, I_O = -1\ \text{mA}$	25°C	4.25 4.65			4.25 4.65			
		Full range	4.25			4.25			
	$V_{IC} = 0, I_O = 50\ \mu\text{A}$	25°C	-4.99			-4.99			
		Full range	-4.85			-4.85			
$V_{IC} = 0, I_O = 500\ \mu\text{A}$	25°C	-4.85 -4.91			-4.85 -4.91				
	Full range	-4.85			-4.85				
$V_{IC} = 0, I_O = 5\ \text{mA}$	25°C	-3.5 -4.1			-3.5 -4.1				
	Full range	-3.5			-3.5				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 10\ \text{k}\Omega$	25°C	20 50		20 50		$\text{V}/\text{mV}$	
			Full range	20		20			
$r_{id}$ Differential input resistance		$R_L = 1\ \text{M}\Omega$	25°C	300		300			
			Full range	20		20			
$r_i$ Common-mode input resistance		25°C	$10^{12}$			$10^{12}$			$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}, N$ package	25°C	8			8			$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz}, A_v = 10$	25°C	130			130			$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V}$ $V_O = 0, R_S = 50\ \Omega$	25°C	75 80		75 80		$\text{dB}$		
		Full range	75		75				
KSVR Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V}$ $V_{IC} = 0, \text{No load}$	25°C	80 95		80 95		$\text{dB}$		
		Full range	80		80				
$I_{DD}$ Supply current	$V_O = 0, \text{No load}$	25°C	4.8 6		4.8 6		$\text{mA}$		
		Full range	6		6				

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.





**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
SLOS108C – MARCH 1992 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2274M			TLC2274AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate at unity gain $V_O = \pm 2.3\text{ V}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C	2.3	3.6		2.3	3.6		$\text{V}/\mu\text{s}$
		Full range	1.7			1.7			
$V_n$	Equivalent input noise voltage $f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C		50			50		$\text{nV}/\sqrt{\text{Hz}}$
		25°C		9			9		
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage $f = 0.1\text{ Hz to } 1\text{ Hz}$ $f = 0.1\text{ Hz to } 10\text{ Hz}$	25°C		1			1		$\mu\text{V}$
		25°C		1.4			1.4		
$I_n$	Equivalent input noise current	25°C		0.6			0.6		$\text{fA}/\sqrt{\text{Hz}}$
THD + N	Total harmonic distortion plus noise $V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $f = 20\text{ kHz}$	$A_V = 1$		0.0011%			0.0011%		
		$A_V = 10$	25°C		0.004%		0.004%		
		$A_V = 100$			0.03%		0.03%		
	Gain-bandwidth product $f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ , $R_L = 10\text{ k}\Omega$	25°C		2.25			2.25		MHz
BOM	Maximum output-swing bandwidth $V_{O(PP)} = 4.6\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $A_V = 1$ , $C_L = 100\text{ pF}$	25°C		0.54			0.54		MHz
$t_s$	Settling time $A_V = -1$ , Step = $-2.3\text{ V to } 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	To 0.1%	25°C		1.5			1.5	$\mu\text{s}$
		To 0.01%			3.2		3.2		
$\phi_m$	Phase margin at unit gain $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		52°			52°		
	Gain margin	25°C		10			10		dB

† Full range is  $-55^\circ\text{C}$  to  $125^\circ\text{C}$ .

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**  
 SLOS106C – MARCH 1992 – REVISED AUGUST 1994

electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC2274Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $V_O = 0$ , $V_{DD\pm} = \pm 2.5\text{ V}$ , $R_S = 50\ \Omega$		300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\mu\text{A}$
$I_{IB}$ Input bias current			1	100	$\mu\text{A}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	0 to 4	-0.3 .to 4.2		V
$V_{OH}$ High-level output voltage	$ V_{IO}  \leq 5\text{ mV}$		4.99		V
	$I_{OH} = -20\ \mu\text{A}$	4.85	4.93		
	$I_{OH} = -200\ \mu\text{A}$	4.25	4.65		
$V_{OL}$ Low-level output voltage	$I_{OL} = -1\text{ mA}$		0.01		V
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$	0.09	0.15		
	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$	0.9	1.5		
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 5\text{ mA}$	$R_L = 10\text{ k}\Omega^\dagger$	15	35	V/mV
		$R_L = 1\text{ M}\Omega^\dagger$		175	
$r_{id}$ Differential input resistance	$V_O = 1\text{ V to }4\text{ V}$		$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$		8		pF
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$		140		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $R_S = 50\ \Omega$ , $V_O = 2.5\text{ V}$	70	75		dB
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , No load, $V_{IC} = V_{DD}/2$	80	95		dB
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load		4.4	6	mA

† Referenced to 2.5 V



**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2274Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $R_S = 50\ \Omega$	$V_O = 0,$	300	2500	$\mu\text{V}$
$I_{IO}$ Input offset current			0.5	100	$\text{pA}$
$I_{IB}$ Input bias current			1	100	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,$	$ V_{IO}  \leq 5\ \text{mV}$	-5 to 4	-5.3 to 4.2	$\text{V}$
$V_{OM+}$ Maximum positive peak output voltage	$I_O = -20\ \mu\text{A}$		4.99		$\text{V}$
	$I_O = -200\ \mu\text{A}$		4.85	4.93	
	$I_O = -1\ \text{mA}$		4.25	4.65	
$V_{OM-}$ Maximum negative peak output voltage	$V_{IC} = 0,$ $I_{OL} = 50\ \mu\text{A}$		-4.99		$\text{V}$
	$V_{IC} = 0,$ $I_{OL} = 500\ \mu\text{A}$		-4.85	-4.91	
	$V_{IC} = 0,$ $I_{OL} = 5\ \text{mA}$		-3.5	-4.1	
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$	$R_L = 10\ \text{k}\Omega$	25	50	$\text{V/mV}$
		$R_L = 1\ \text{M}\Omega$	300		
$r_{id}$ Differential input resistance			$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz}$		8		$\text{pF}$
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz},$	$A_V = 10$	130		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = -5\ \text{V to } 2.7\ \text{V},$ $R_S = 50\ \Omega$	$V_O = 0,$	75	80	$\text{dB}$
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.2\ \text{V to } \pm 8\ \text{V},$	$V_{IC} = 0$	80	95	$\text{dB}$
$I_{DD}$ Supply current	$V_O = 0,$	No load	4.8	6	$\text{mA}$

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution vs Common-mode voltage	1, 2 3, 4
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	5, 6
$I_B/I_{IO}$	Input bias and offset currents	vs Free-air temperature	7
$V_I$	Input voltage range	vs Supply voltage vs Free-air temperature	8 9
$V_{OH}$	High-level output voltage	vs High-level output current	10
$V_{OL}$	Low-level output voltage	vs Low-level output current	11, 12
$V_{OM+}$	Maximum positive peak output voltage	vs Output current	13
$V_{OM-}$	Maximum negative peak output voltage	vs Output current	14
$V_{OM}$	Maximum output voltage	vs Frequency	15
$I_{OS}$	Short-circuit output current	vs Supply voltage vs Free-air temperature	16 17
$V_O$	Output voltage	vs Differential input voltage	18, 19
$A_{VD}$	Large-signal differential voltage amplification	vs Load resistance vs Frequency vs Free-air temperature	20 21, 22 23, 24
$z_o$	Output impedance	vs Frequency	25, 26
$CMRR$	Common-mode rejection ratio	vs Frequency vs Free-air temperature	27 28
$ks_{VR}$	Supply-voltage rejection ratio	vs Frequency vs Free-air temperature	29, 30 31
$I_{DD}$	Supply current	vs Supply voltage vs Free-air temperature	32 33
$SR$	Slew rate	vs Load capacitance vs Free-air temperature	34 35
	Large-signal pulse response	vs Time	36, 37, 38, 39
	Small-signal pulse response	vs Time	40, 41, 42, 43
$V_n$	Equivalent input noise voltage	vs Frequency	44, 45
	Noise voltage	Over a 10-second period	46
	Integrated noise voltage	vs Frequency	47
$THD + N$	Total harmonic distortion plus noise	vs Frequency	48
	Gain-bandwidth product	vs Free-air temperature vs Supply voltage	49 50
$\phi_m$	Phase margin	vs Frequency vs Load capacitance	21, 22 51
	Gain margin	vs Load capacitance	52

NOTE: For all graphs where  $V_{DD} = 5\text{ V}$ , all loads are referenced to 2.5 V.

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC2274  
 INPUT OFFSET VOLTAGE

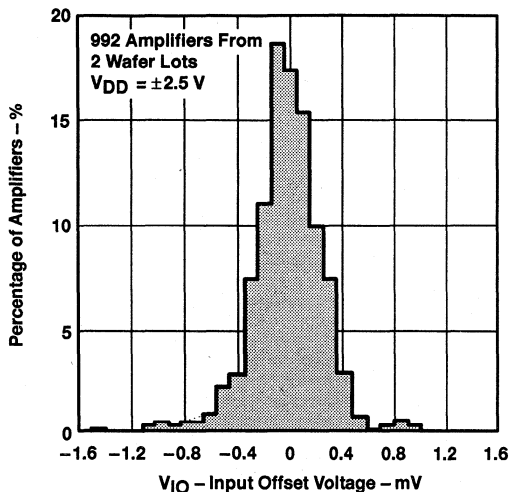


Figure 1

DISTRIBUTION OF TLC2274  
 INPUT OFFSET VOLTAGE

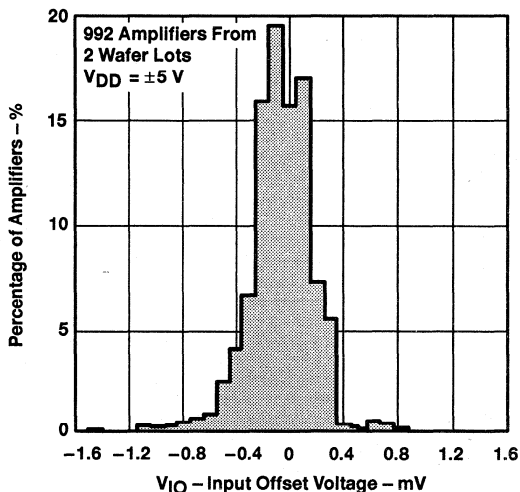


Figure 2

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE VOLTAGE

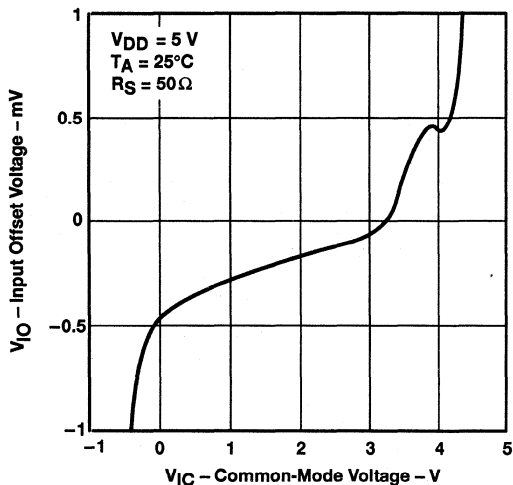


Figure 3

INPUT OFFSET VOLTAGE  
 vs  
 COMMON-MODE VOLTAGE

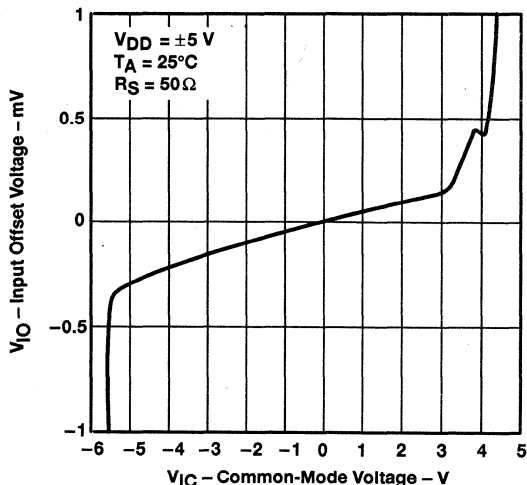


Figure 4

TYPICAL CHARACTERISTICS†

DISTRIBUTION OF TLC2274 INPUT OFFSET  
 VOLTAGE TEMPERATURE COEFFICIENT

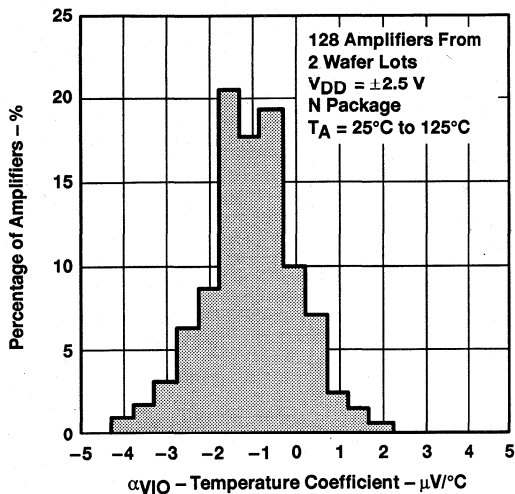


Figure 5

DISTRIBUTION OF TLC2274 INPUT OFFSET  
 VOLTAGE TEMPERATURE COEFFICIENT

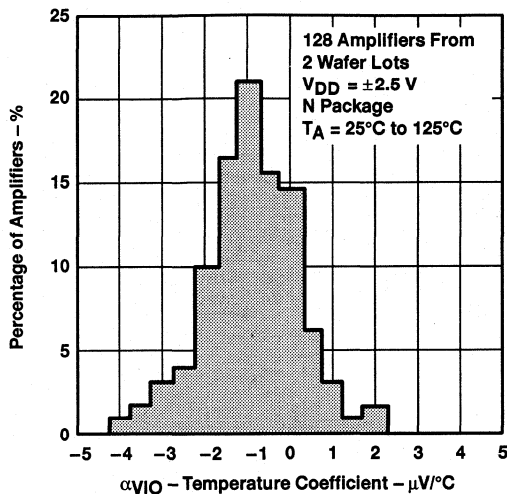


Figure 6

INPUT BIAS AND OFFSET CURRENTS  
 vs  
 FREE-AIR TEMPERATURE

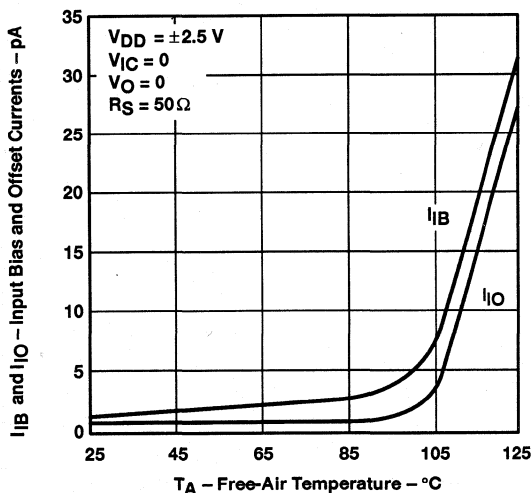


Figure 7

INPUT VOLTAGE RANGE  
 vs  
 SUPPLY VOLTAGE

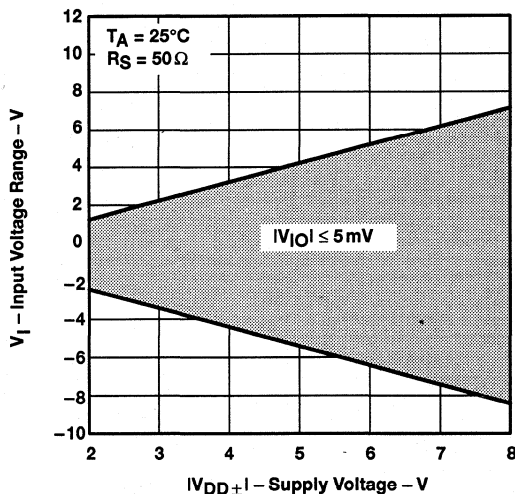
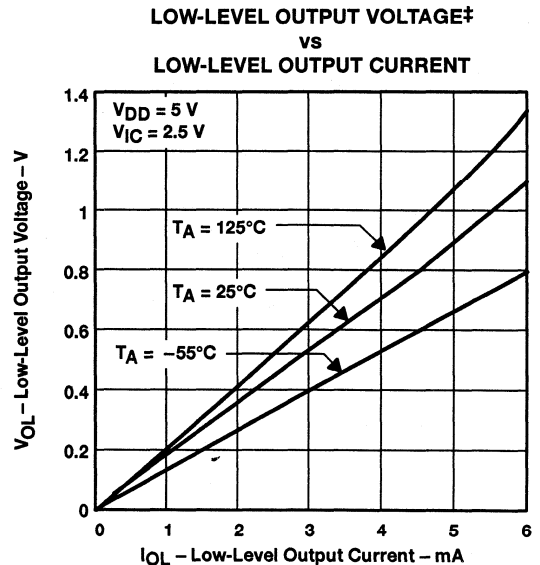
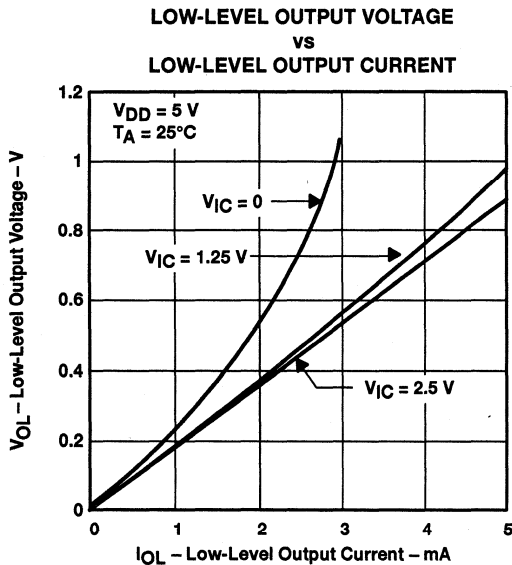
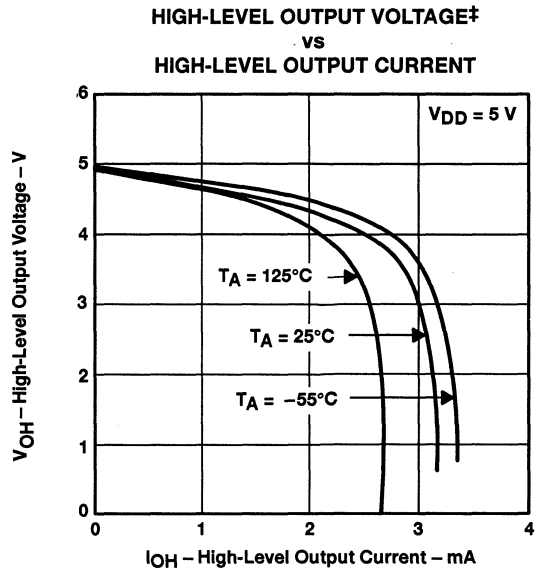
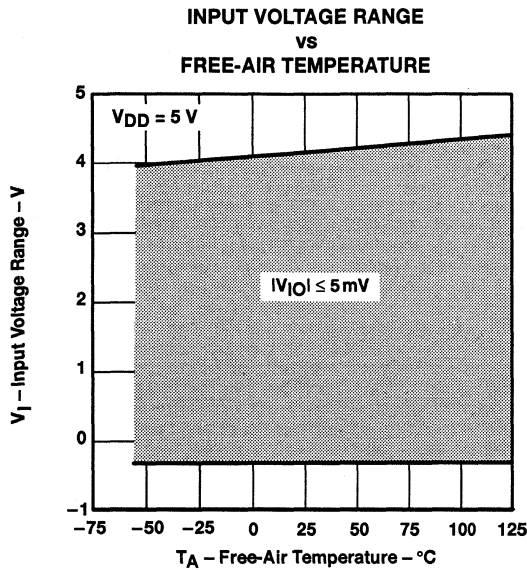


Figure 8

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



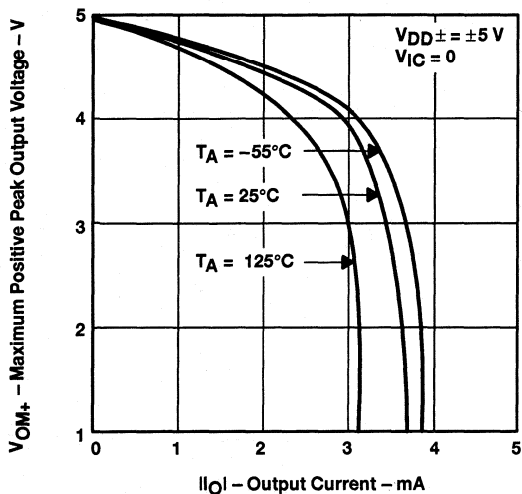
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.  
 ‡ The -55°C curve and the 125°C curve apply only to the M version.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

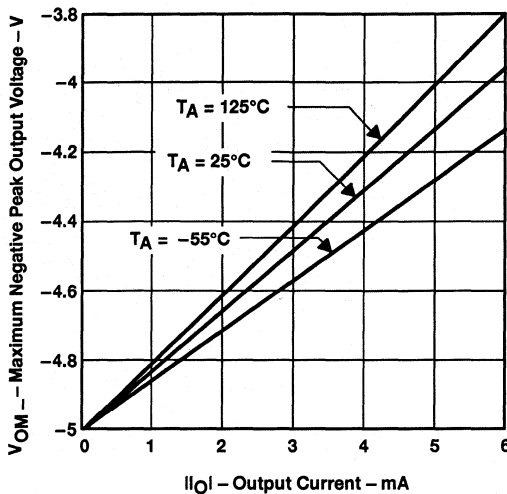
**TYPICAL CHARACTERISTICS**

**MAXIMUM POSITIVE PEAK OUTPUT VOLTAGE†**  
**vs**  
**OUTPUT CURRENT**



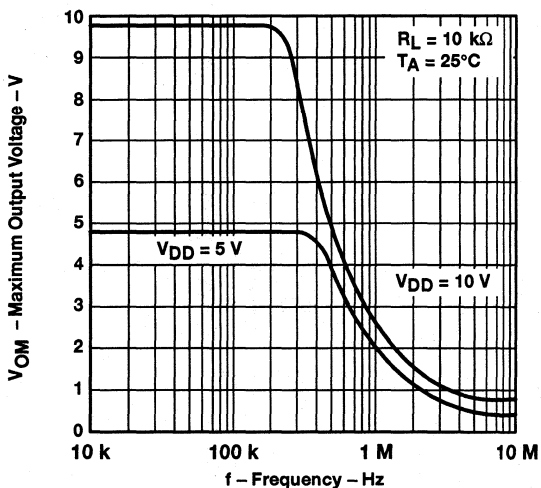
**Figure 13**

**MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE†**  
**vs**  
**OUTPUT CURRENT**



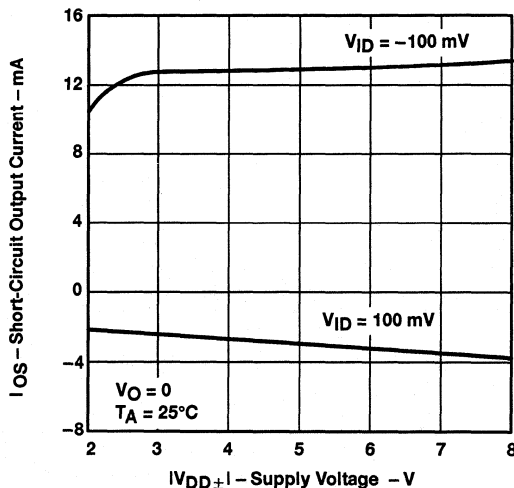
**Figure 14**

**MAXIMUM OUTPUT VOLTAGE**  
**vs**  
**FREQUENCY**



**Figure 15**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 16**

† The -55°C curve and the 125°C curve apply only to the M version.





TYPICAL CHARACTERISTICS†

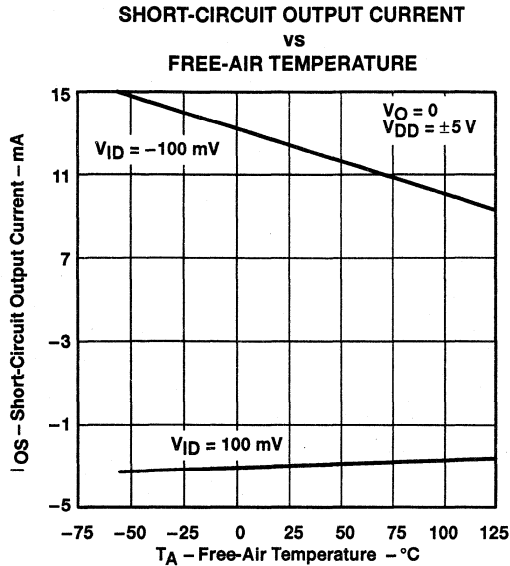


Figure 17

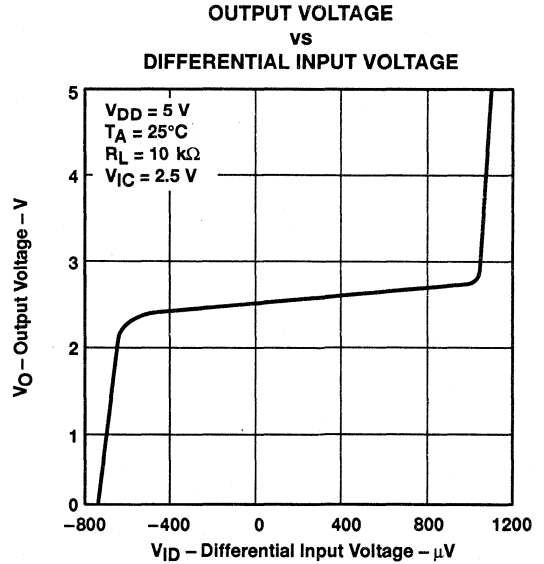


Figure 18

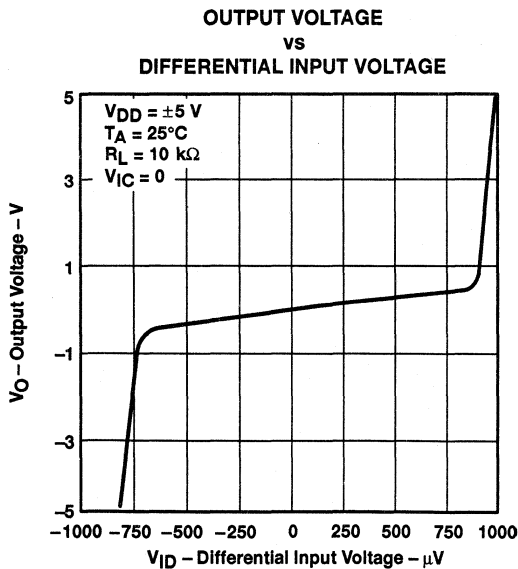


Figure 19

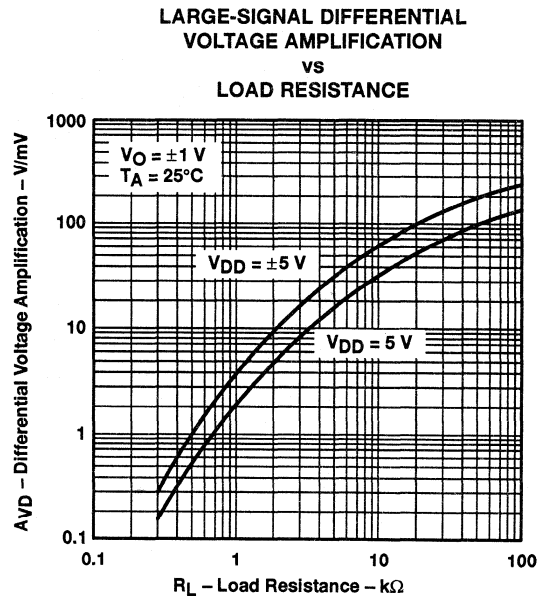


Figure 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE MARGIN vs FREQUENCY**

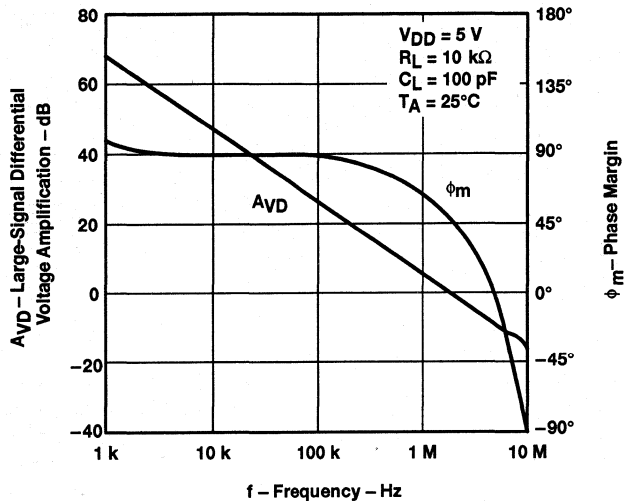


Figure 21

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE MARGIN vs FREQUENCY**

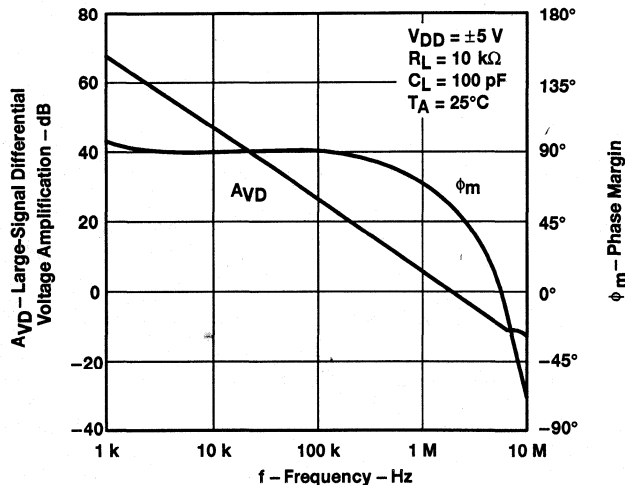


Figure 22



TYPICAL CHARACTERISTICS†

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

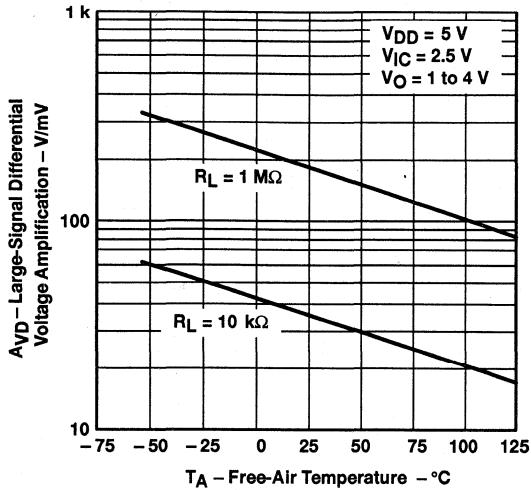


Figure 23

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

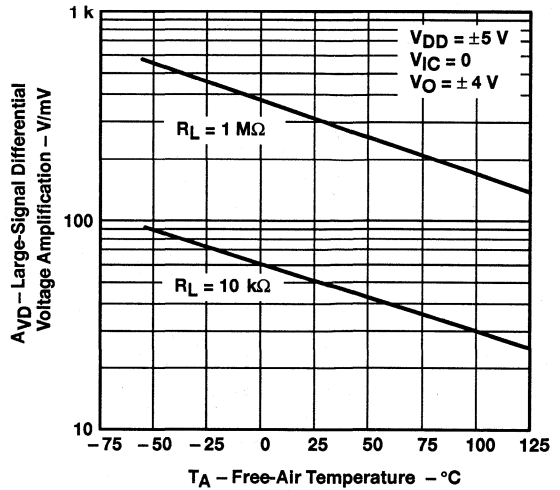


Figure 24

OUTPUT IMPEDANCE  
 VS  
 FREQUENCY

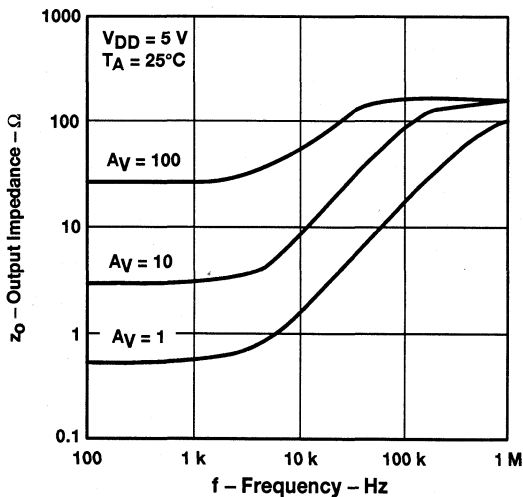


Figure 25

OUTPUT IMPEDANCE  
 VS  
 FREQUENCY

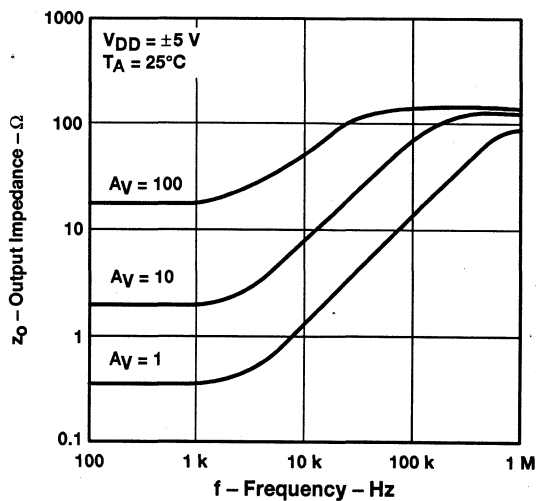


Figure 26

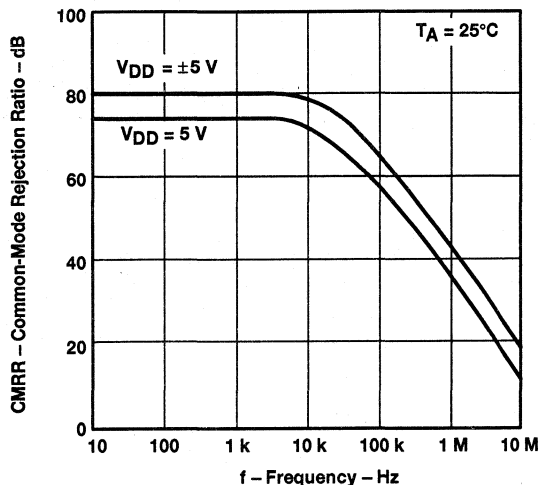
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

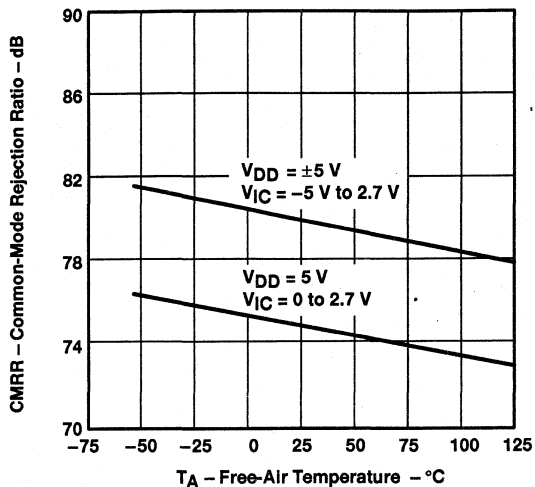
**TYPICAL CHARACTERISTICS†**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREQUENCY**



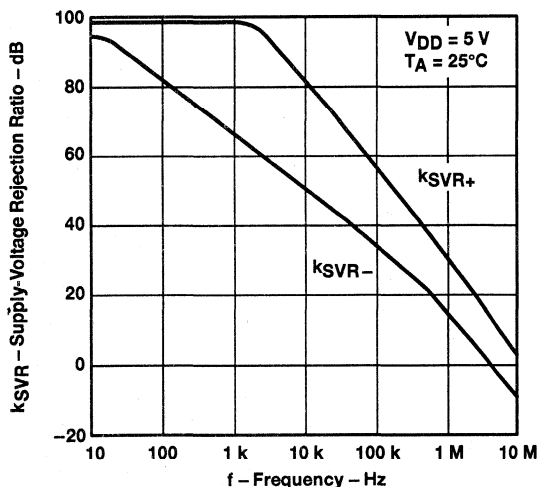
**Figure 27**

**COMMON-MODE REJECTION RATIO**  
**vs**  
**FREE-AIR TEMPERATURE**



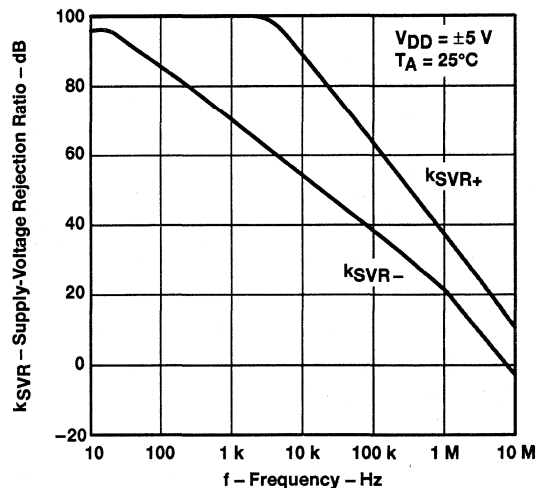
**Figure 28**

**SUPPLY-VOLTAGE REJECTION RATIO**  
**vs**  
**FREQUENCY**



**Figure 29**

**SUPPLY-VOLTAGE REJECTION RATIO**  
**vs**  
**FREQUENCY**

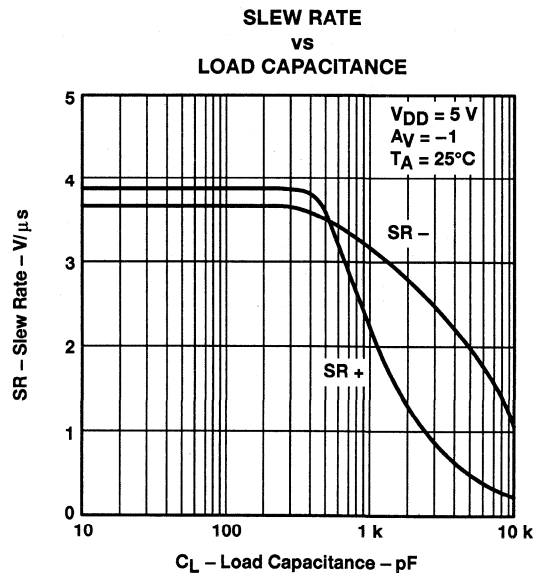
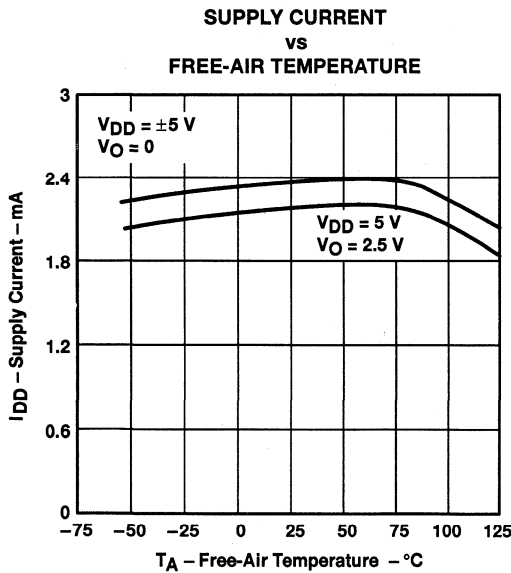
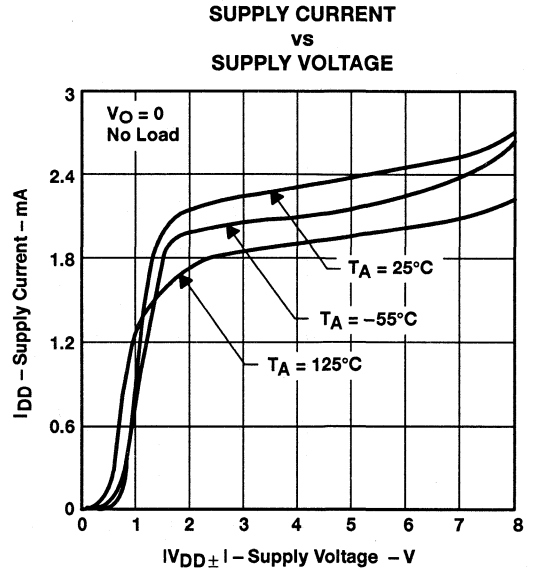
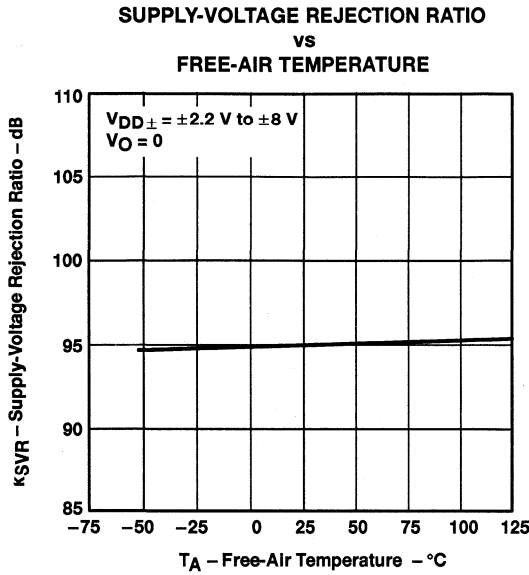


**Figure 30**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†



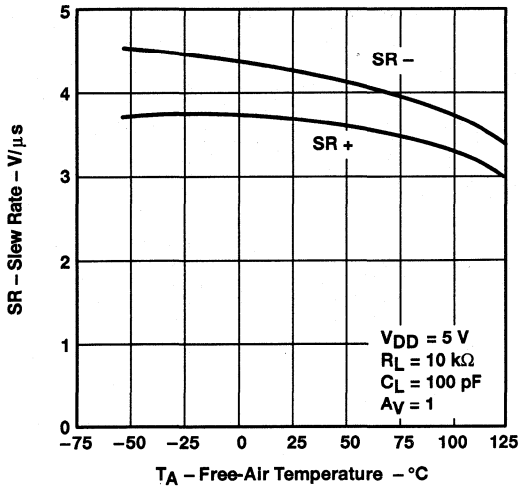
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

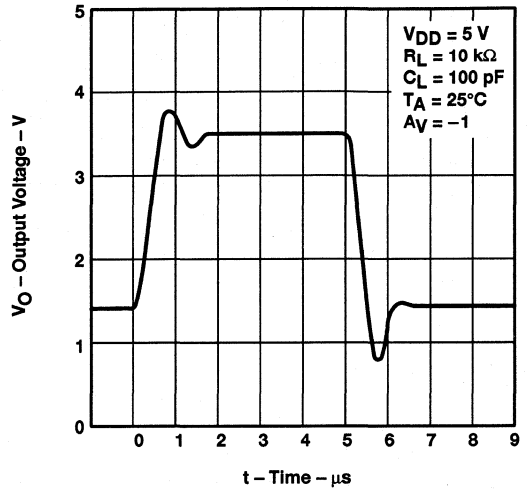
**TYPICAL CHARACTERISTICS†**

**SLEW RATE  
 vs  
 FREE-AIR TEMPERATURE**



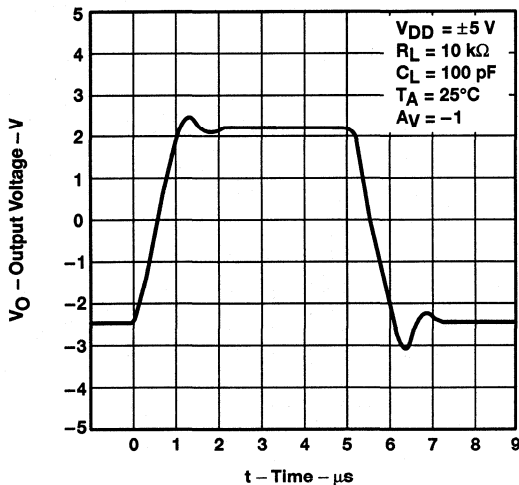
**Figure 35**

**INVERTING LARGE-SIGNAL PULSE RESPONSE**



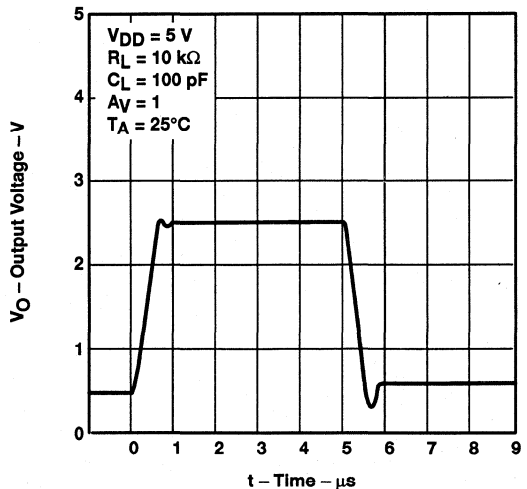
**Figure 36**

**INVERTING LARGE-SIGNAL PULSE RESPONSE**



**Figure 37**

**VOLTAGE-FOLLOWER  
 LARGE-SIGNAL PULSE RESPONSE**



**Figure 38**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

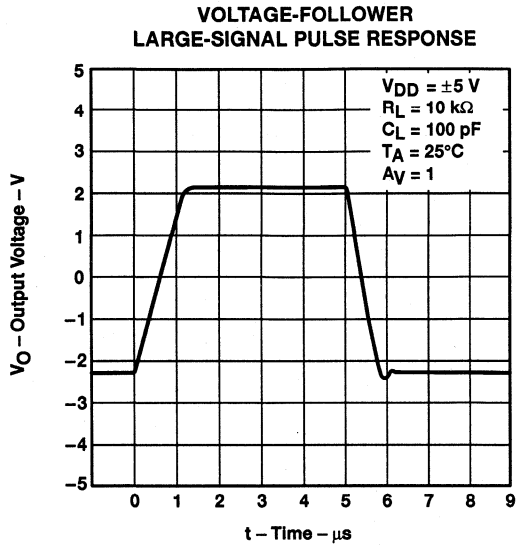


Figure 39

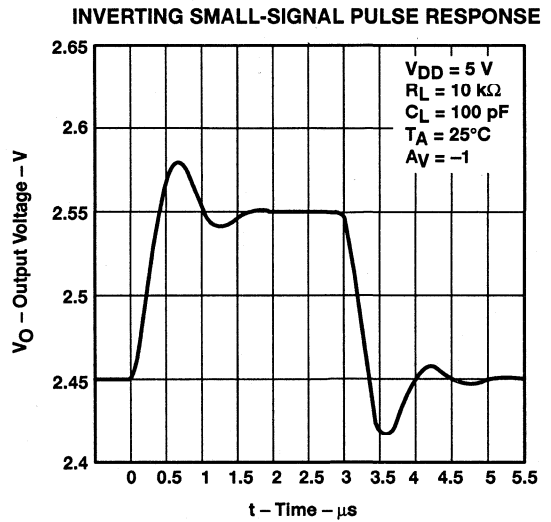


Figure 40

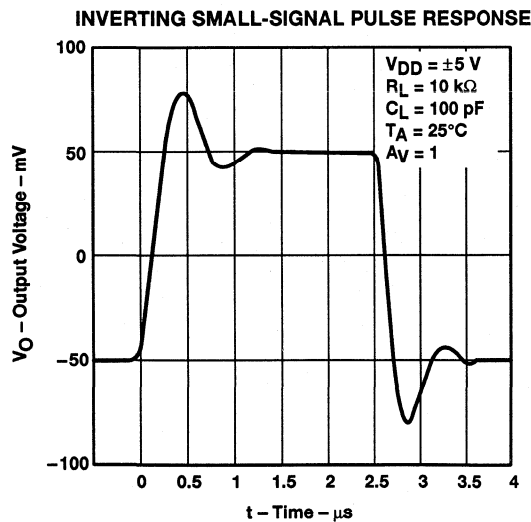


Figure 41

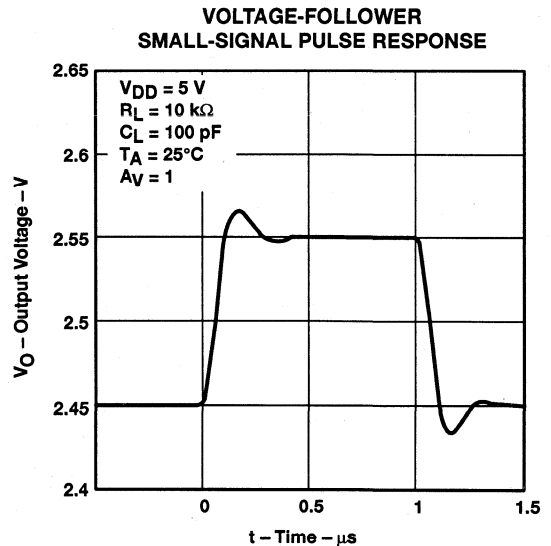


Figure 42

TYPICAL CHARACTERISTICS

VOLTAGE-FOLLOWER  
 SMALL-SIGNAL PULSE RESPONSE

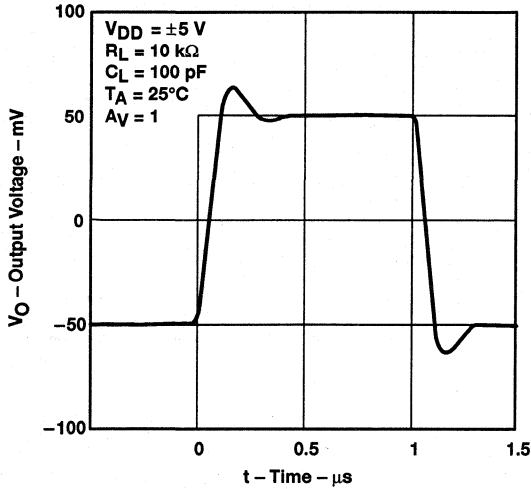


Figure 43

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

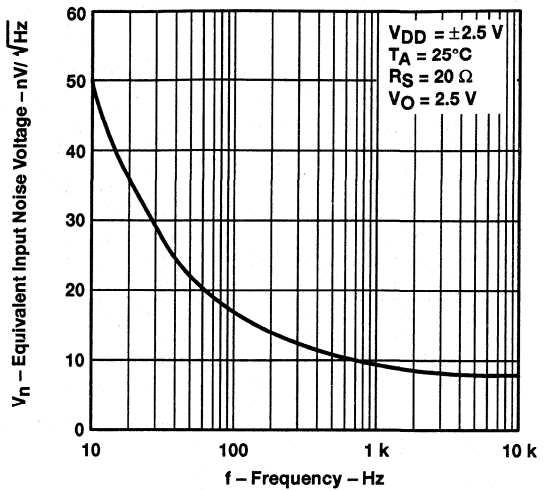


Figure 44

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

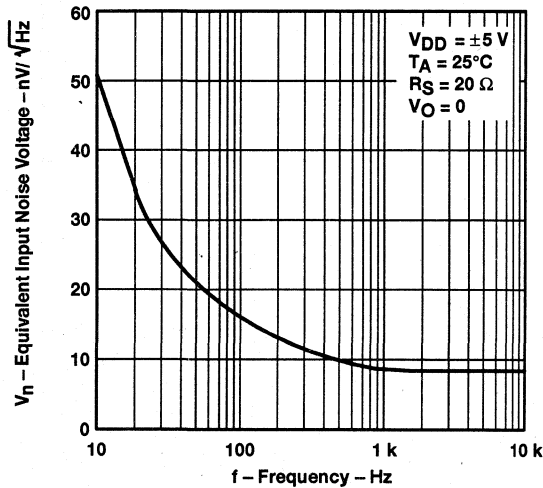


Figure 45

NOISE VOLTAGE  
 OVER A 10-SECOND PERIOD

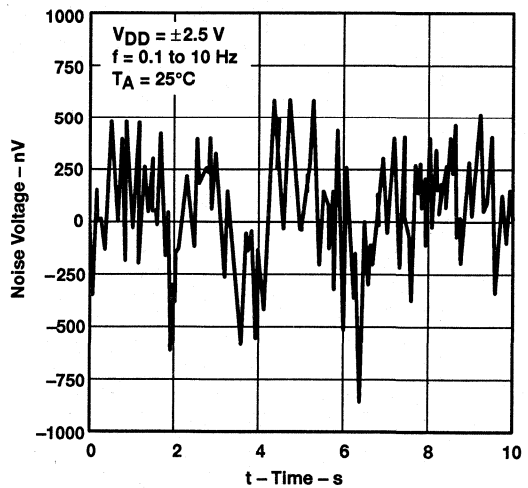


Figure 46



TYPICAL CHARACTERISTICS†

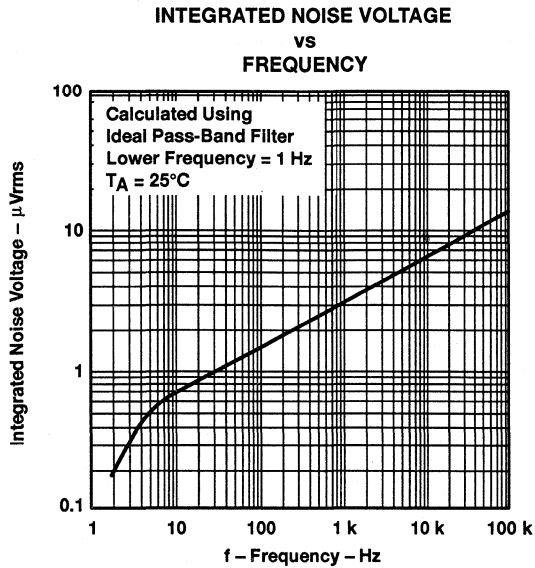


Figure 47

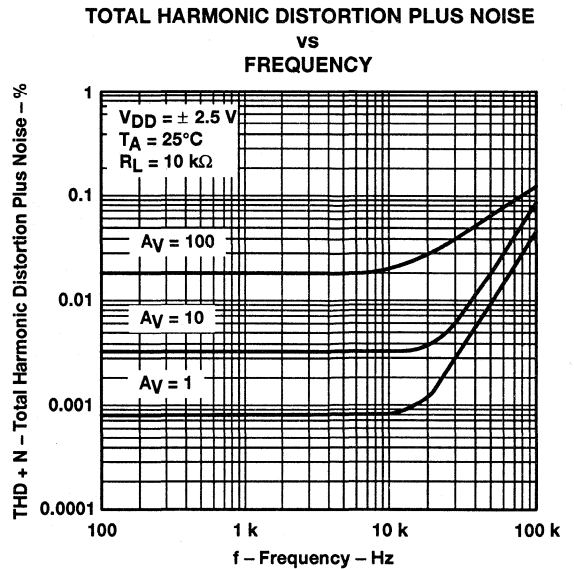


Figure 48

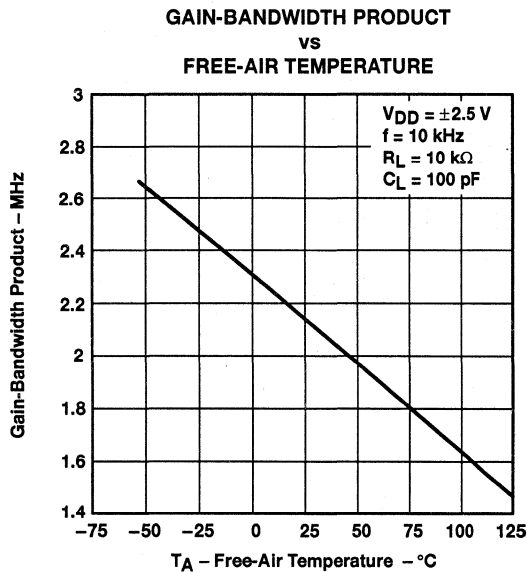


Figure 49

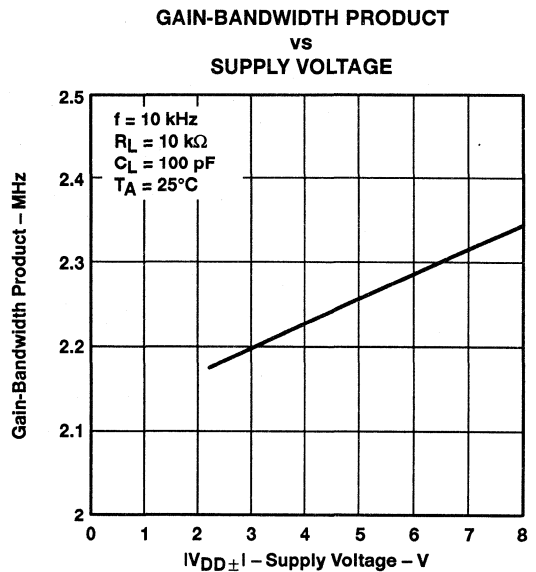


Figure 50

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2274, TLC2274A, TLC2274Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**QUAD OPERATIONAL AMPLIFIERS**

SLOS106C – MARCH 1992 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**PHASE MARGIN**  
**vs**  
**LOAD CAPACITANCE**

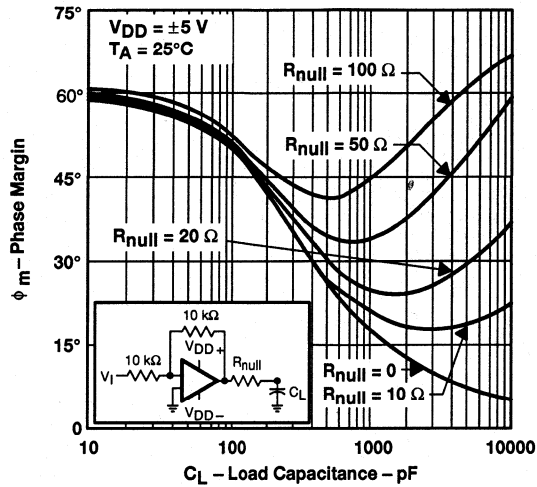


Figure 51

**GAIN MARGIN**  
**vs**  
**LOAD CAPACITANCE**

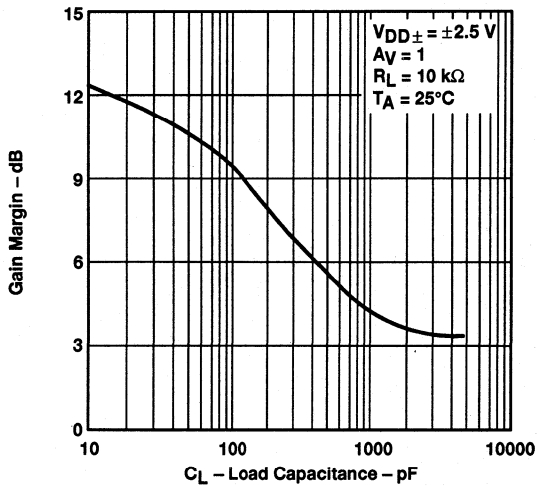


Figure 52







# TLC2652, TLC2652A, TLC2652Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

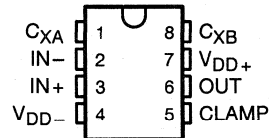
- Extremely Low Offset Voltage . . . 1  $\mu\text{V}$  Max
- Extremely Low Change on Offset Voltage With Temperature . . . 0.003  $\mu\text{V}/^\circ\text{C}$  Typ
- Low Input Offset Current  
500 pA Max at  $T_A = -55^\circ\text{C}$  to  $125^\circ\text{C}$
- $A_{\text{VD}}$  . . . 135 dB Min
- CMRR and  $k_{\text{SVR}}$  . . . 120 dB Min
- Single-Supply Operation
- Common-Mode Input Voltage Range Includes the Negative Rail
- No Noise Degradation With External Capacitors Connected to  $V_{\text{DD-}}$

### description

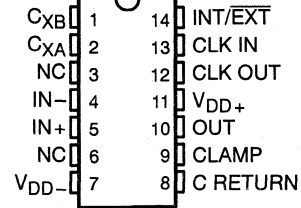
The TLC2652 and TLC2652A are high-precision chopper-stabilized operational amplifiers using Texas Instruments Advanced LinCMOS™ process. This process in conjunction with unique chopper-stabilization circuitry produces operational amplifiers whose performance matches or exceeds that of similar devices available today.

Chopper-stabilization techniques make possible extremely high dc precision by continuously nulling input offset voltage even during variation in temperature, time, common-mode voltage, and power supply voltage. In addition, low-frequency noise voltage is significantly reduced. This high precision, coupled with the extremely high input impedance of the CMOS input stage, makes the

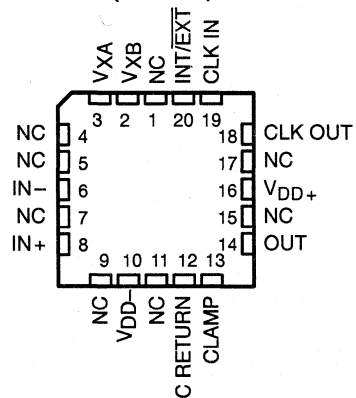
D008, JG, OR P PACKAGE  
(TOP VIEW)



D014, JG, OR N PACKAGE  
(TOP VIEW)



FK PACKAGE  
(TOP VIEW)



NC – No internal connection

### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IOMax</sub> AT 25°C	PACKAGED DEVICES							CHIP FORM (Y)
		8 PIN			14 PIN				
		SMALL OUTLINE (D008)	CERAMIC DIP (JG)	PLASTIC DIP (P)	SMALL OUTLINE (D014)	CERAMIC DIP (J)	PLASTIC DIP (N)	CHIP CARRIER (FK)	
0°C to 70°C	1 $\mu\text{V}$ 3 $\mu\text{V}$	TLC2652AC-8D TLC2652C-8D	— —	TLC2652ACP TLC2652CP	TLC2652AC-14D TLC2652C-14D	— —	TLC2652ACN TLC2652CN	— —	TLC2652Y
-40°C to 85°C	1 $\mu\text{V}$ 3 $\mu\text{V}$	TLC2652AI-8D TLC2652A-8D	— —	TLC2652AIP TLC2652IP	TLC2652AI-14D TLC2652I-14D	— —	TLC2652AIN TLC2652IN	— —	—
-55°C to 125°C	1 $\mu\text{V}$ 3 $\mu\text{V}$	TLC2652AM-8D TLC2652M-8D	TLC2652AMJG TLC2652MJG	TLC2652AMP TLC2652MP	TLC2652AM-14D TLC2652M-14D	TLC2652AMJ TLC2652MJ	TLC2652AMN TLC2652MN	TLC2652AMFK TLC2652MFK	—

The D008 and D014 packages are available taped and reeled. Add R suffix to the device type (e.g., TLC2652AC-8DR). Chips are tested at 25°C.

Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated

On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

# TLC2652, TLC2652A, TLC2652Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

### description (continued)

TLC2652 and TLC2652A are an ideal choice for low-level signal processing applications such as strain gauges, thermocouples, and other transducer amplifiers. For applications that require extremely low noise and higher usable bandwidth, use the TLC2654 or TLC2654A device, which has a chopping frequency of 10 kHz.

The TLC2652 and TLC2652A input common-mode range includes the negative rail, thereby providing superior performance in either single-supply or split-supply applications, even at power supply voltage levels as low as  $\pm 1.9$  V.

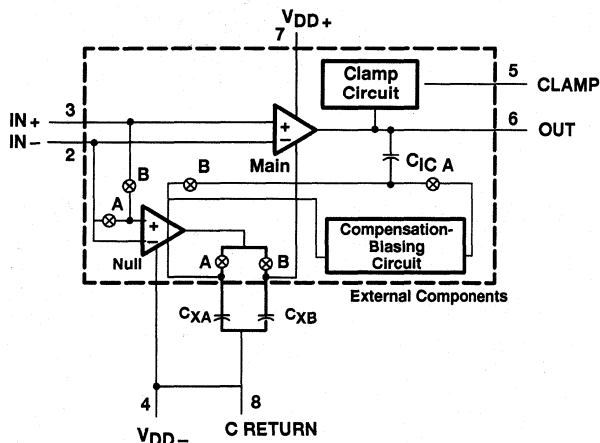
Two external capacitors are required for operation of the device; however, the on-chip chopper control circuitry is transparent to the user. On devices in the 14-pin and 20-pin packages, the control circuitry is made accessible to allow the user the option of controlling the clock frequency with an external frequency source. In addition, the clock threshold level of the TLC2652 and TLC2652A require no level shifting when used in the single-supply configuration with a normal CMOS or TTL clock input.

Innovative circuit techniques are used on the TLC2652 and TLC2652A to allow exceptionally fast overload recovery time. If desired, an output clamp pin is available to reduce the recovery time even further.

The device inputs and output are designed to withstand  $-100$ -mA surge currents without sustaining latch-up. Additionally the TLC2652 and TLC2652A incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

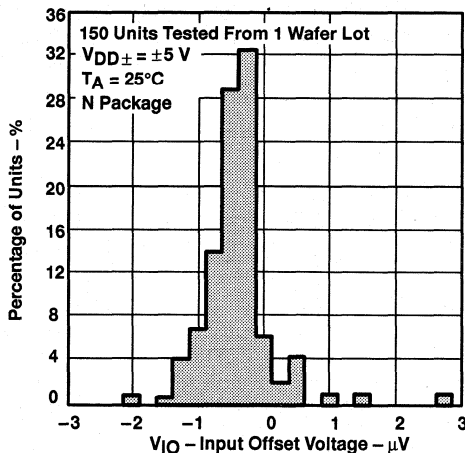
The C-suffix devices are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ . The I-suffix devices are characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The M-suffix devices are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

### functional block diagram



Pin numbers shown are for the D (14 pin), JG, and N packages.

DISTRIBUTION OF TLC2652 INPUT OFFSET VOLTAGE

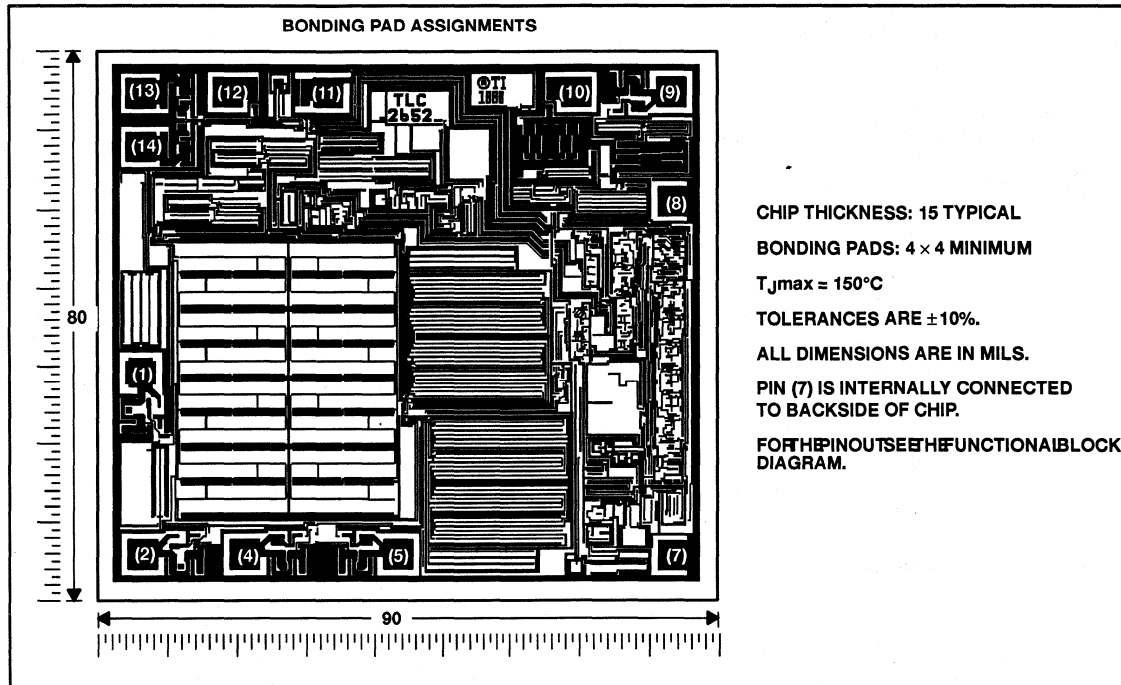


**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**TLC2652Y chip information**

This chip, when properly assembled, displays characteristics similar to the TLC2652C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TLC2652, TLC2652A, TLC2652Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage $V_{DD+}$ (see Note 1)	8 V
Supply voltage $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 16$ V
Input voltage, $V_I$ (any input, see Note 1)	$\pm 8$ V
Voltage range on CLK IN and INT/EXT	$V_{DD-}$ to $V_{DD+} + 5.2$ V
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Current into CLK IN and INT/EXT	$\pm 5$ mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .
2. Differential voltages are at  $IN+$  with respect to  $IN-$ .
3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D008	725 mV	5.8 mW/°C	464 mW	377 mW	145 mW
D014	950 mV	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mV	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mV	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mV	8.4 mW/°C	672 mW	546 mW	210 mW
N	1575 mV	12.6 mW/°C	1008 mW	819 mW	315 mW
P	1000 mV	8.0 mW/°C	640 mW	520 mW	200 mW

### recommended operating conditions

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	$\pm 1.9$	$\pm 8$	$\pm 1.9$	$\pm 8$	$\pm 1.9$	$\pm 8$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.9$	$V_{DD-}$	$V_{DD+} - 1.9$	$V_{DD-}$	$V_{DD+} - 1.9$	V
Clock input voltage	$V_{DD-}$	$V_{DD-} + 5$	$V_{DD-}$	$V_{DD-} + 5$	$V_{DD-}$	$V_{DD-} + 5$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C





**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2652C			TLC2652AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.6		3	0.5		1	$\mu\text{V}$
		Full range			4.35			2.35	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.003	0.03		0.003	0.03	$\mu\text{V}/^\circ\text{C}$	
Input offset voltage long-term drift (see Note 4)		25°C	0.003	0.06		0.003	0.02	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C	2			2			$\text{pA}$
		Full range			100			100	
$I_{IB}$ Input bias current		25°C	4			4			$\text{pA}$
		Full range			100			100	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 3.1			-5 to 3.1		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	25°C	4.7	4.8		4.7	4.8	V	
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	25°C	-4.7	-4.9		-4.7	-4.9	V	
		Full range	-4.7			-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$ , $R_L = 10\ \text{k}\Omega$	25°C	120	150		135	150	dB	
		Full range	120			130			
$f_{ch}$ Internal chopping frequency		25°C	450			450		Hz	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25			25		$\mu\text{A}$	
		Full range	25			25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C			100	100		$\text{pA}$	
		Full range			100	100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50\ \Omega$	25°C	120	140		120	140	dB	
		Full range	120			120			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 1.9\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	120	135		120	135	dB	
		Full range	120			120			
$I_{DD}$ Supply current		25°C	1.5	2.4		1.5	2.4	mA	
		Full range			2.5	2.5			

† Full range is 0° to 70°C.

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated at  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**operating characteristics specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2652C			TLC2652AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	$V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	2	2.8		2	2.8	V/ $\mu\text{s}$	
		Full range	1.5			1.5			
SR- Negative slew rate at unity gain		25°C	2.3	3.1		2.3	3.1	V/ $\mu\text{s}$	
		Full range	1.8			1.8			
$V_n$ Equivalent input noise voltage (see Note 6)	f = 10 Hz	25°C		94		94	140	nV/ $\sqrt{\text{Hz}}$	
	f = 1 kHz	25°C		23		23	35		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C		0.8		0.8		$\mu\text{V}$	
	f = 0 to 10 Hz	25°C		2.8		2.8			
$I_n$ Equivalent input noise current	f = 10 kHz	25°C		0.004		0.004		fA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	f = 10 kHz, $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		1.9		1.9		MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C		48°		48°			

† Full range is 0° to 70°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2652A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2652I			TLC2652AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, \quad R_S = 50\ \Omega$	25°C	0.6		3	0.5		1	$\mu\text{V}$
		Full range			4.95			2.95	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.003		0.03	0.003		0.03	$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.003		0.06	0.003		0.02	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C	2			2			$\text{pA}$
		Full range			150			150	
$I_{IB}$ Input bias current	25°C	4			4			$\text{pA}$	
	Full range			150			150		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	–5 to 3.1			–5 to 3.1		$\text{V}$	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega, \quad \text{See Note 5}$	25°C	4.7	4.8		4.7	4.8	$\text{V}$	
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega, \quad \text{See Note 5}$	25°C	–4.7	–4.9		–4.7	–4.9	$\text{V}$	
		Full range	–4.7			–4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, \quad R_L = 10\ \text{k}\Omega$	25°C	120	150		135	150	$\text{dB}$	
		Full range	120			125			
Internal chopping frequency		25°C	450			450			$\text{Hz}$
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C	25			25		$\mu\text{A}$	
		Full range	25			25			
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C			100			$\text{pA}$	
		Full range			100				
CMRR Common-mode rejection ratio	$V_O = 0, \quad V_{IC} = V_{ICRmin}, \quad R_S = 50\ \Omega$	25°C	120	140		120	140	$\text{dB}$	
		Full range	120			120			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 1.9\ \text{V to } \pm 8\ \text{V}, \quad V_O = 0, \quad R_S = 50\ \Omega$	25°C	120	135		120	135	$\text{dB}$	
		Full range	120			120			
$I_{DD}$ Supply current	$V_O = 0, \quad \text{No load}$	25°C	1.5		2.4	1.5		2.4	$\text{mA}$
		Full range			2.5			2.5	

† Full range is  $-40^\circ$  to  $85^\circ\text{C}$ .

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated at  $T_A = 25^\circ$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2652I			TLC2652AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	2	2.8		2	2.8		V/μs
		Full range	1.4			1.4			
SR- Negative slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	2.3	3.1		2.3	3.1		V/μs
		Full range	1.7			1.7			
V <sub>n</sub> Equivalent input noise voltage (see Note 6)	f = 10 Hz	25°C		94		94	140		nV/√Hz
	f = 1 kHz	25°C		23		23	35		
V <sub>N(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C		0.8		0.8			μV
	f = 0 to 10 Hz	25°C		2.8		2.8			
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C		0.004		0.004			pA/√Hz
Gain-bandwidth product	f = 10 kHz, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		1.9		1.9			MHz
φ <sub>m</sub> Phase margin at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		48°		48°			

† Full range is -40° to 85°C.

NOTE 6: This parameter is tested on a sample basis for the TLC2652A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A \dagger$	TLC2652M			TLC2652AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage (see Note 7)	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		0.6	3.5		0.5	3	$\mu\text{V}$
		Full range			10			8	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range		0.003	0.03*		0.003	0.03*	$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C		0.003	0.06*		0.003	0.02*	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current		25°C			2			2	$\text{pA}$
		Full range				500			
$I_{IB}$ Input bias current	25°C			4			4	$\text{pA}$	
	Full range				500				500
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range		-5 to 3.1			-5 to 3.1	$\text{V}$	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	25°C		4.7	4.8		4.7	4.8	$\text{V}$
		Full range		4.7			4.7		
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	25°C		-4.7	-4.9		-4.7	-4.9	$\text{V}$
		Full range		-4.7			-4.7		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$ , $R_L = 10\ \text{k}\Omega$	25°C		120	150		135	150	$\text{dB}$
		Full range		120			120		
$f_{ch}$ Internal chopping frequency		25°C		450			450	$\text{Hz}$	
Clamp on-state current	$R_L = 100\ \text{k}\Omega$	25°C		25			25	$\mu\text{A}$	
		Full range		25			25		
Clamp off-state current	$V_O = -4\ \text{V to } 4\ \text{V}$	25°C			100			100	$\text{pA}$
		Full range			500			500	
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50\ \Omega$	25°C		120	140		120	140	$\text{dB}$
		Full range		120			120		
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 1.9\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C		120	135		120	135	$\text{dB}$
		Full range		120			120		
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.5	2.4		1.5	2.4	$\text{mA}$
		Full range			2.5			2.5	

\* On products complaint to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is -55° to 125°C

- NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated at  $T_A = 25^\circ$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.  
5. Output clamp is not connected.  
7. This parameter is not production tested. Thermocouple effects preclude measurement of the actual  $V_{IO}$  of these devices in high speed automated testing.  $V_{IO}$  is measured to a limit determined by the test equipment capability at the temperature extremes. The test ensures that the stabilization circuitry is performing properly.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

operating characteristics specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A^\dagger$	TLC2652M TLC2652AM			UNIT
			MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	$V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	2	2.8	V/ $\mu\text{s}$	
		Full range	1.3			
SR- Negative slew rate at unity gain		25°C	2.3	3.1	V/ $\mu\text{s}$	
		Full range	1.6			
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C	94		nV/ $\sqrt{\text{Hz}}$	
	$f = 1\text{ kHz}$	25°C	23			
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0\text{ to }1\text{ Hz}$	25°C	0.8		$\mu\text{V}$	
	$f = 0\text{ to }10\text{ Hz}$	25°C	2.8			
$I_n$ Equivalent input noise current	$f = 1\text{ kHz}$	25°C	0.004		pA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	1.9		MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	48°			

$^\dagger$  Full range is  $-55^\circ$  to  $125^\circ\text{C}$ .



**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at  $V_{DD\pm} = \pm 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2652Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 50\ \Omega$		0.6	3	$\mu\text{V}$
Input offset voltage long-term drift (see Note 4)			0.003	0.006	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current				2	$\text{pA}$
$I_{IB}$ Input bias current				4	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	-5 to 3.1			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	4.7	4.8		V
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5	-4.7	-4.9		V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$ , $R_L = 10\ \text{k}\Omega$	120	150		dB
$f_{ch}$ Internal chopping frequency			450		Hz
Clamp on-state current	$R_L = 100\ \text{k}\Omega$		25		$\mu\text{A}$
Clamp off-state current	$V_O = -4\ \text{V}$ to $4\ \text{V}$			100	$\text{pA}$
CMRR Common-mode rejection ratio	$V_O = 0$ , $R_S = 50\ \Omega$ , $V_{IC} = V_{ICR\text{min}}$	120	140		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 1.9\ \text{V}$ to $\pm 8\ \text{V}$ , $R_S = 50\ \Omega$ , $V_O = 0$	120	135		dB
$I_{DD}$ Supply current	$V_O = 0$ , No load		1.5	2.4	$\text{mA}$

NOTES: 4. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated at  $T_A = 25^\circ$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

5. Output clamp is not connected.

**operating characteristics at  $V_{DD\pm} = \pm 5\ \text{V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2652Y			UNIT
		MIN	TYP	MAX	
$SR+$ Positive slew rate at unity gain	$V_O = \pm 2.3\ \text{V}$ , $R_L = 10\ \text{k}\Omega$	2	2.8		$\text{V}/\mu\text{s}$
$SR-$ Negative slew rate at unity gain	$C_L = 100\ \text{pF}$	2.3	3.1		$\text{V}/\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		94		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$		23		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0$ to $1\ \text{Hz}$		0.8		$\mu\text{V}$
	$f = 0$ to $10\ \text{Hz}$		2.8		
$I_n$ Equivalent input noise current	$f = 1\ \text{kHz}$				$\text{pA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		48°		

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Normalized input offset voltage	vs Chopping frequency	1
$I_{IB}$	Input bias current	vs Common-mode input voltage	2
		vs Chopping frequency	3
		vs Free-air temperature	4
$I_{IO}$	Input offset current	vs Chopping frequency	5
		vs Free-air temperature	6
	Clamp current	vs Output voltage	7
$V_{(OPP)}$	Maximum peak-to-peak output voltage	vs Frequency	8
$V_{OM}$	Maximum peak output voltage	vs Output current	9, 10
		vs Free-air temperature	11, 12
$AVD$	Large-signal differential voltage amplification	vs Frequency	13
		vs Free-air temperature	14
	Chopping frequency	vs Supply voltage	15
		vs Free-air temperature	16
$I_{DD}$	Supply current	vs Supply voltage	17
		vs Free-air temperature	18
$I_{OS}$	Short-circuit output current	vs Supply voltage	19
		vs Free-air temperature	20
$SR$	Slew rate	vs Supply voltage	21
		vs Free-air temperature	22
	Pulse response	Small-signal	23
		Large-signal	24
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	vs Chopping frequency	25, 26
$V_n$	Equivalent input noise voltage	vs Frequency	27
	Gain-bandwidth product	vs Supply voltage	28
		vs Free-air temperature	29
$\phi_m$	Phase margin	vs Supply voltage	30
		vs Free-air temperature	31
		vs Load capacitance	32
	Phase shift	vs Frequency	13



TYPICAL CHARACTERISTICS†

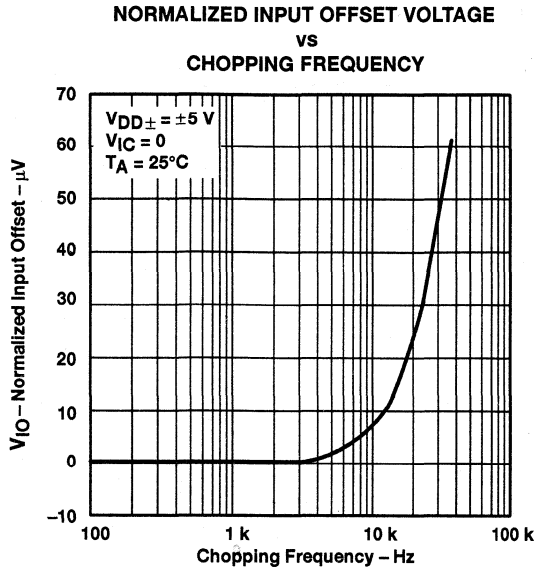


Figure 1

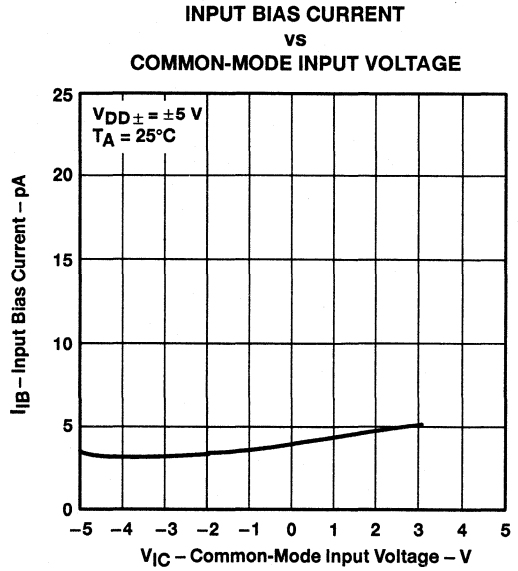


Figure 2

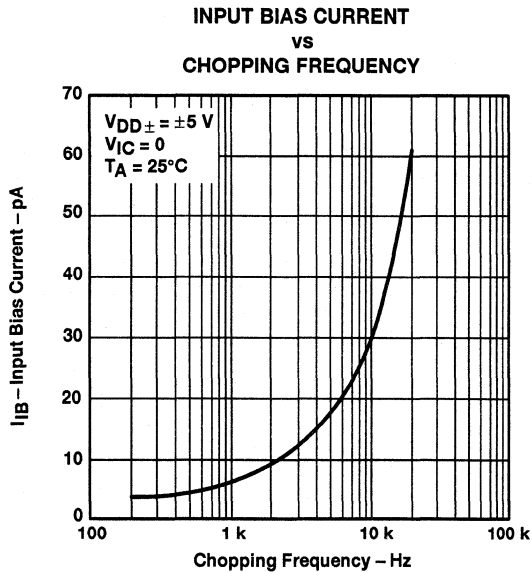


Figure 3

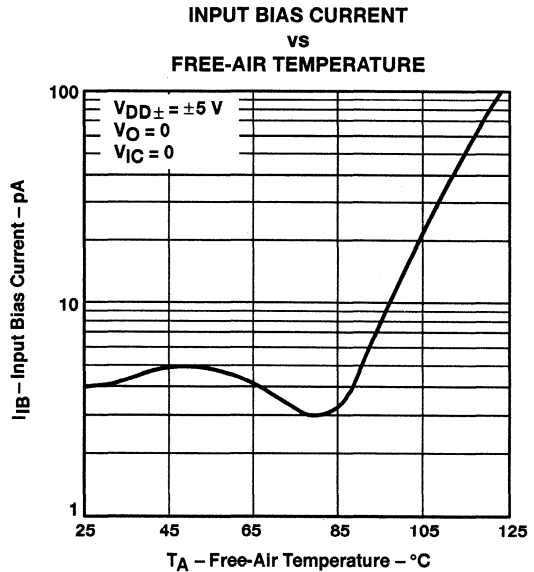


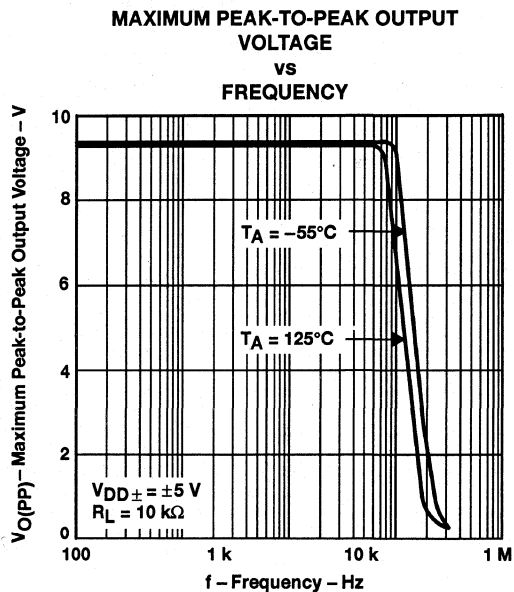
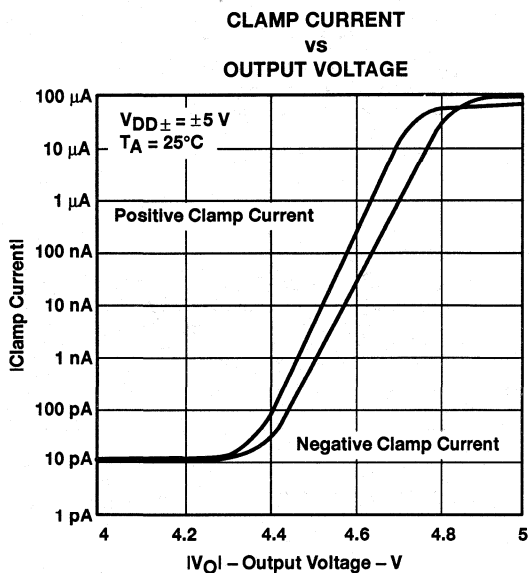
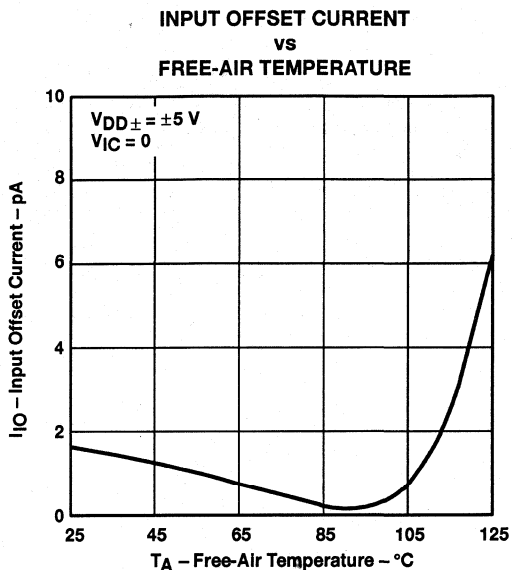
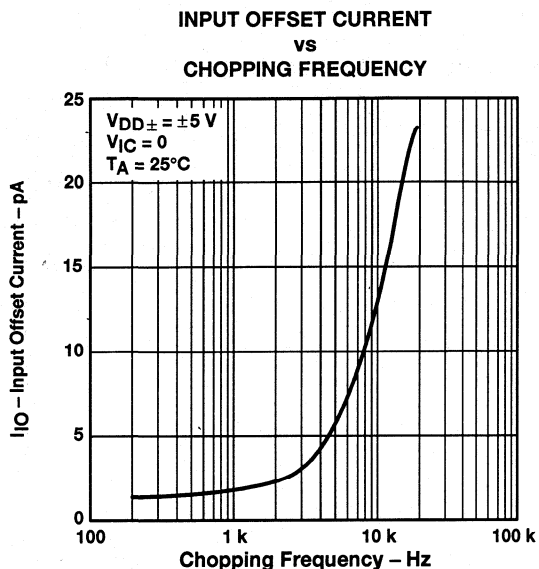
Figure 4

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

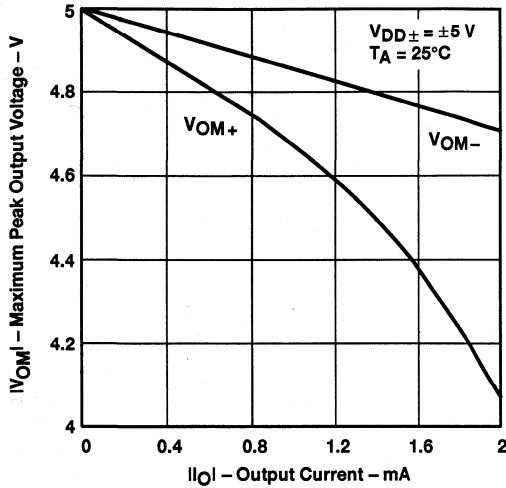


Figure 9

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 OUTPUT CURRENT

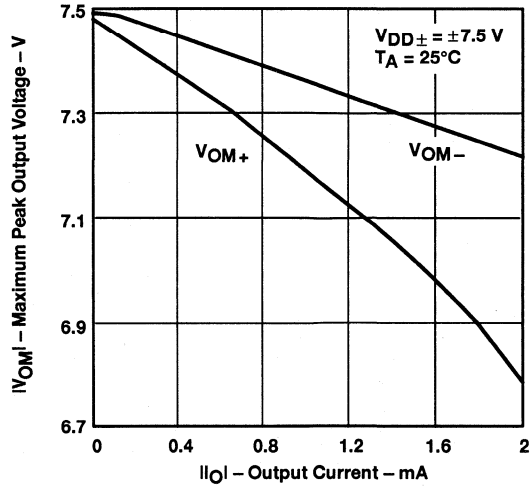


Figure 10

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

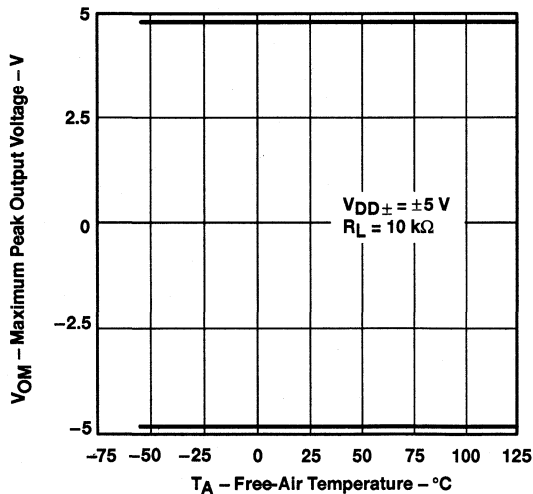


Figure 11

MAXIMUM PEAK OUTPUT VOLTAGE  
 vs  
 FREE-AIR TEMPERATURE

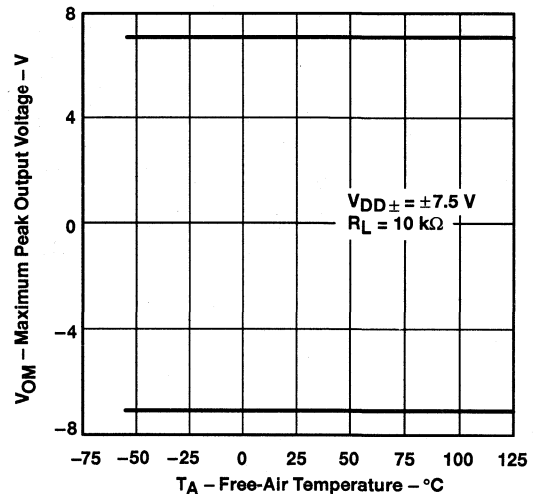


Figure 12

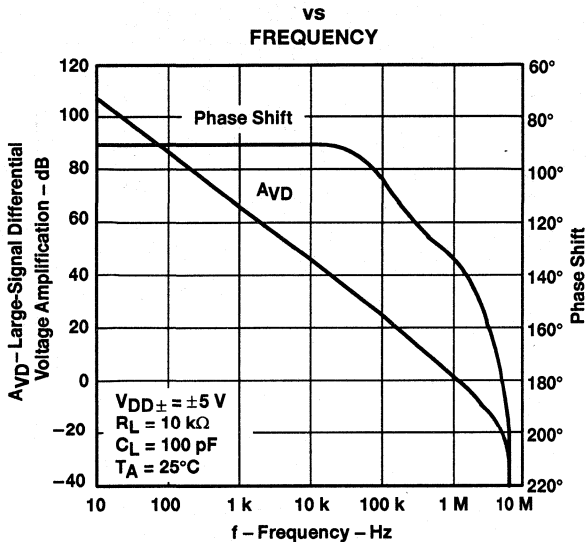
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

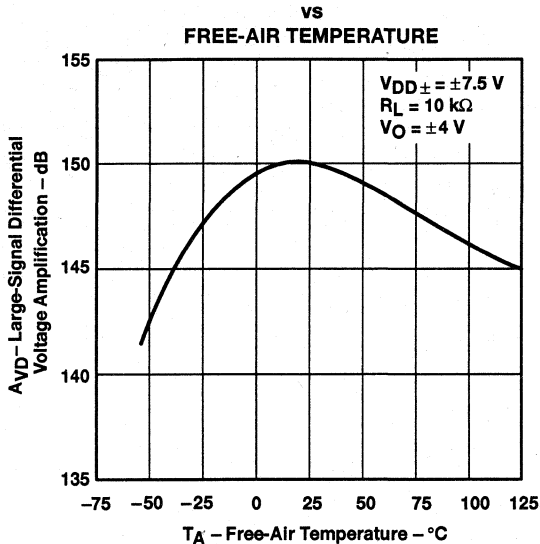
**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT**



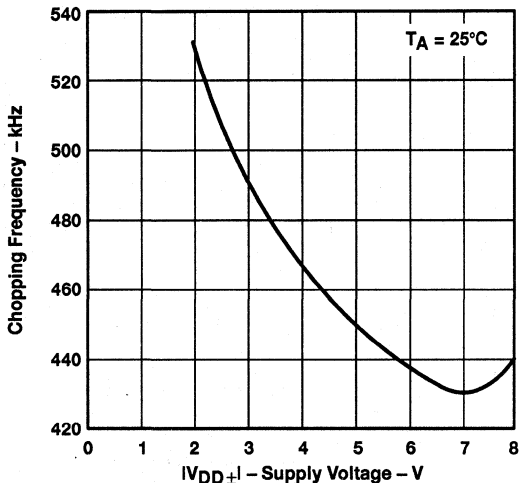
**Figure 13**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION**



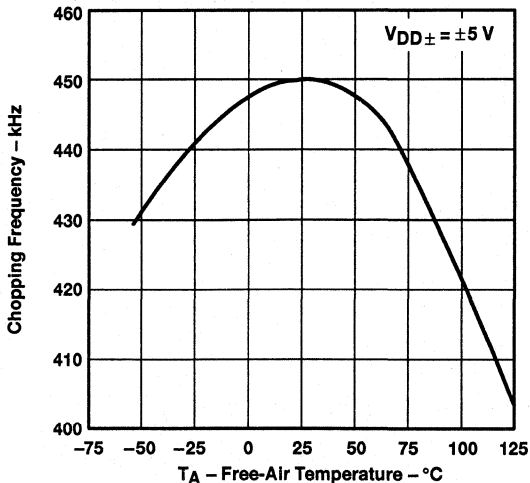
**Figure 14**

**CHOPPING FREQUENCY**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 15**

**CHOPPING FREQUENCY**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 16**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

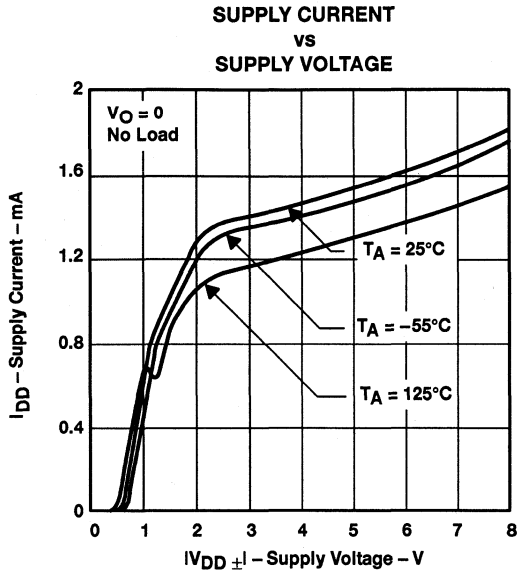


Figure 17

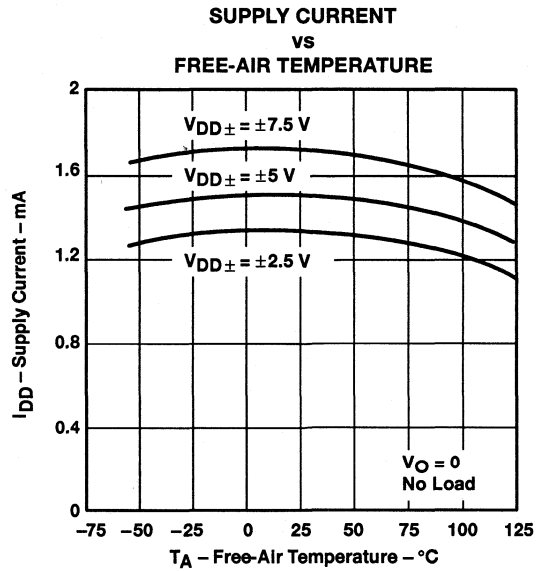


Figure 18

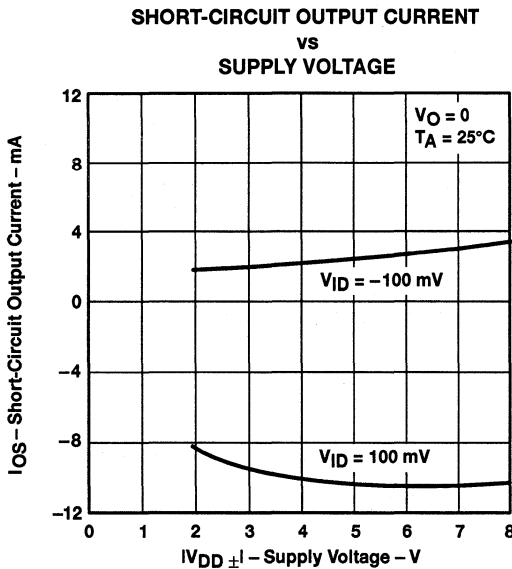


Figure 19

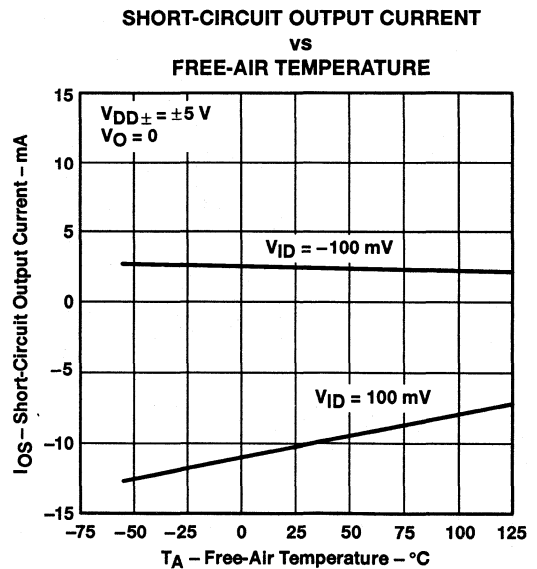


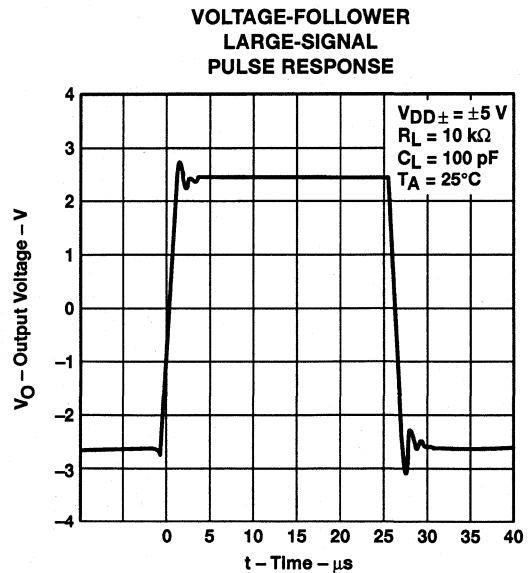
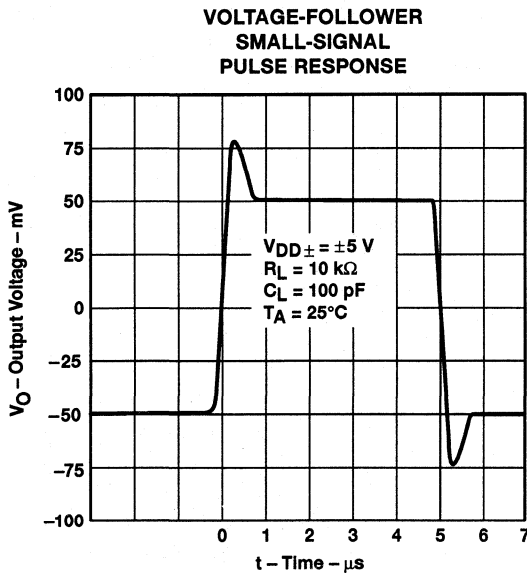
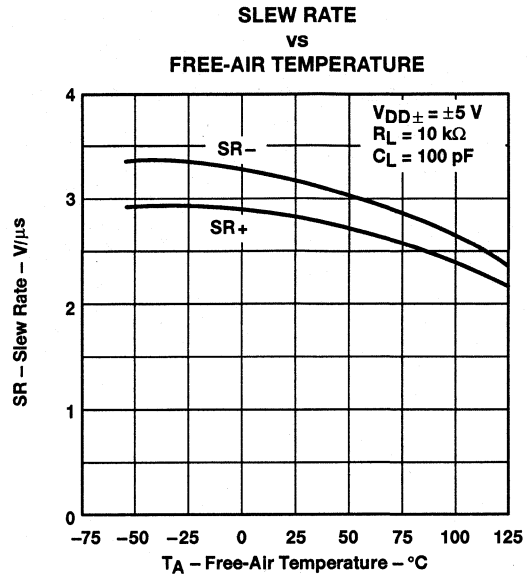
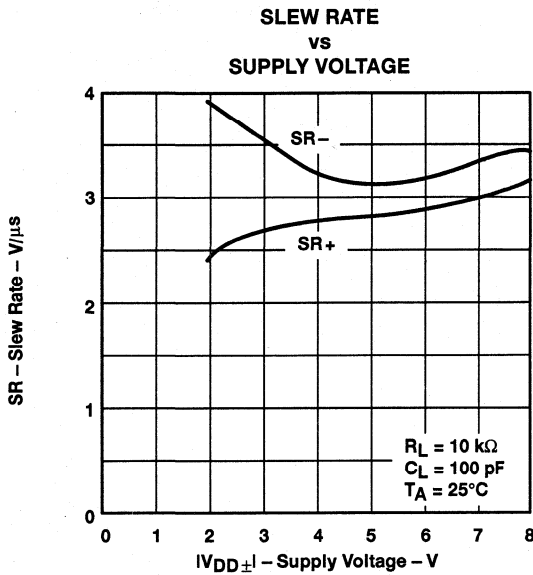
Figure 20

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

PEAK-TO-PEAK INPUT NOISE VOLTAGE  
 VS  
 CHOPPING FREQUENCY

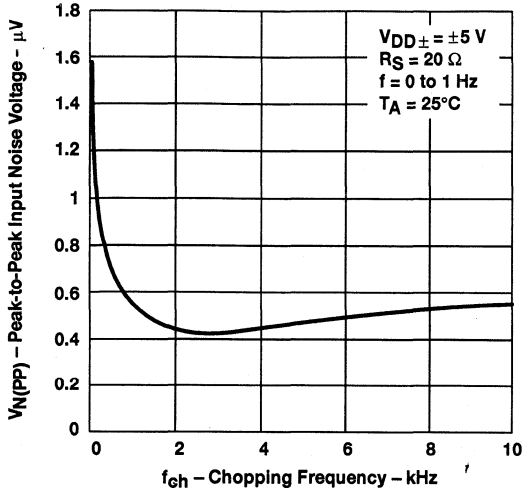


Figure 25

PEAK-TO-PEAK INPUT NOISE VOLTAGE  
 VS  
 CHOPPING FREQUENCY

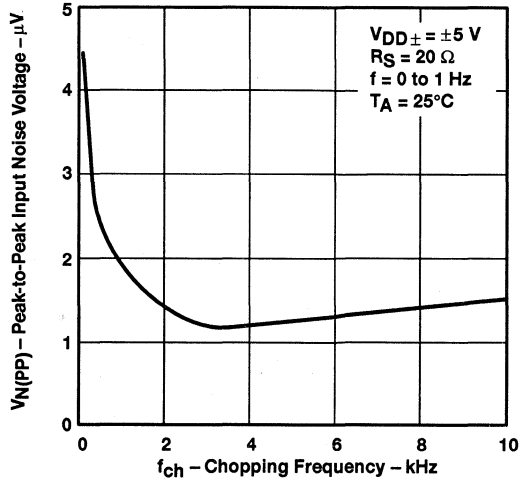


Figure 26

EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY

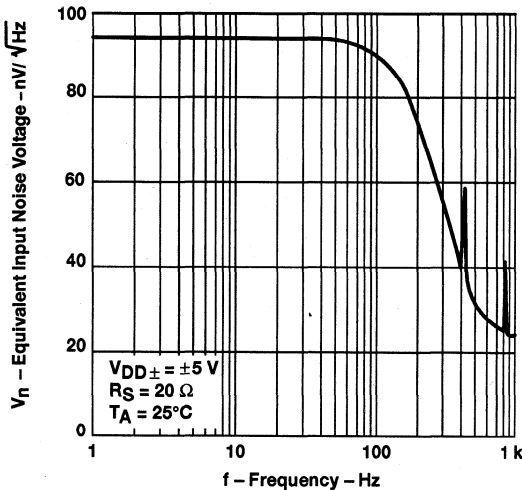


Figure 27

GAIN-BANDWIDTH PRODUCT  
 VS  
 SUPPLY VOLTAGE

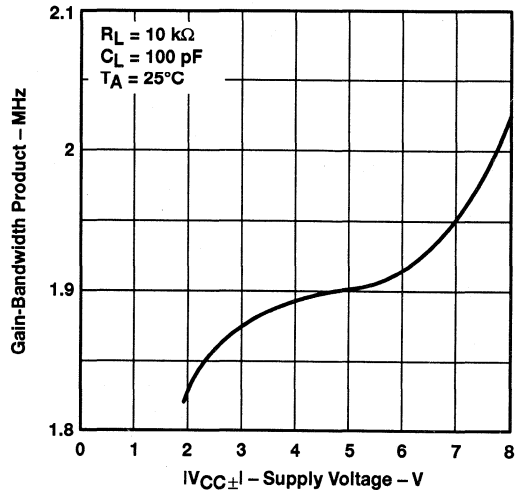


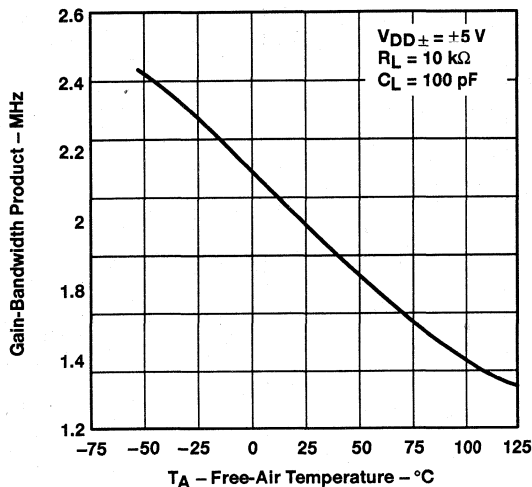
Figure 28

**TLC2652, TLC2652A, TLC2652Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

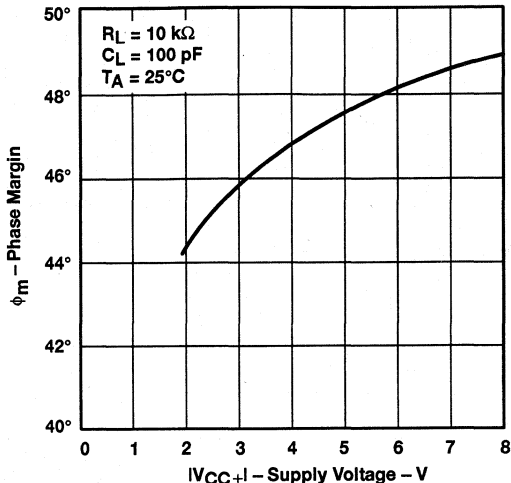
**TYPICAL CHARACTERISTICS†**

**GAIN-BANDWIDTH PRODUCT**  
**VS**  
**FREE-AIR TEMPERATURE**



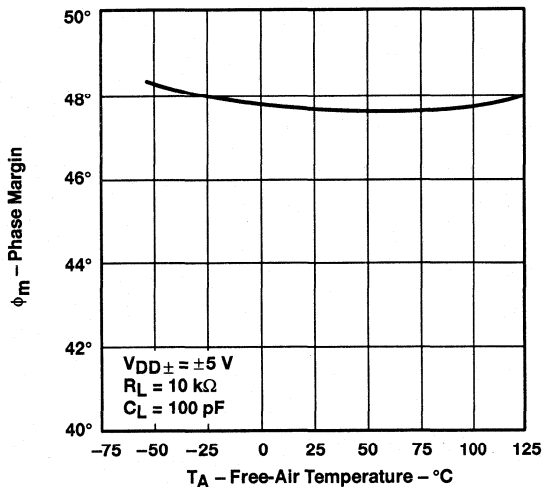
**Figure 29**

**PHASE MARGIN**  
**VS**  
**SUPPLY VOLTAGE**



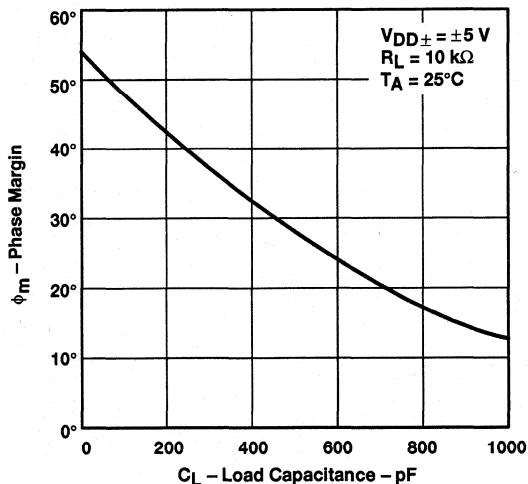
**Figure 30**

**PHASE MARGIN**  
**VS**  
**FREE-AIR TEMPERATURE**



**Figure 31**

**PHASE MARGIN**  
**VS**  
**LOAD CAPACITANCE**



**Figure 32**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





## APPLICATION INFORMATION

### capacitor selection and placement

The two important factors to consider when selecting external capacitors  $C_{XA}$  and  $C_{XB}$  are leakage and dielectric absorption. Both factors can cause system degradation, negating the performance advantages realized by using the TLC2652.

Degradation from capacitor leakage becomes more apparent with the increasing temperatures. Low-leakage capacitors and standoffs are recommended for operation at  $T_A = 125^\circ\text{C}$ . In addition, guard bands are recommended around the capacitor connections on both sides of the printed circuit board to alleviate problems caused by surface leakage on circuit boards.

Capacitors with high dielectric absorption tend to take several seconds to settle upon application of power, which directly affects input offset voltage. In applications where fast settling of input offset voltage is needed, it is recommended that high-quality film capacitors, such as mylar, polystyrene, or polypropylene, be used. In other applications, however, a ceramic or other low-grade capacitor can suffice.

Unlike many choppers available today, the TLC2652 is designed to function with values of  $C_{XA}$  and  $C_{XB}$  in the range of  $0.1\ \mu\text{F}$  to  $1\ \mu\text{F}$  without degradation to input offset voltage or input noise voltage. These capacitors should be located as close as possible to the  $C_{XA}$  and  $C_{XB}$  pins and returned to either  $V_{DD-}$  or C RETURN. On many choppers, connecting these capacitors to  $V_{DD-}$  causes degradation in noise performance. This problem is eliminated on the TLC2652.

### internal/external clock

The TLC2652 has an internal clock that sets the chopping frequency to a nominal value of 450 Hz. On 8-pin packages, the chopping frequency can only be controlled by the internal clock; however, on all 14-pin packages and the 20-pin FK package, the device chopping frequency can be set by the internal clock or controlled externally by use of the INT/EXT and CLK IN pins. To use the internal 450-Hz clock, no connection is necessary. If external clocking is desired, connect INT/EXT to  $V_{DD-}$  and the external clock to CLK IN. The external clock trip point is 2.5 V above the negative rail; however, CLK IN can be driven from the negative rail to 5 V above the negative rail. If this level is exceeded, damage could occur to the device unless the current into CLK IN is limited to  $\pm 5\ \text{mA}$ . When operating in the single-supply configuration, this feature allows the TLC2652 to be driven directly by 5-V TTL and CMOS logic. A divide-by-two frequency divider interfaces with CLK IN and sets the clock chopping frequency. The duty cycle of the external is not critical but should be kept between 30% and 60%.

### overload recovery/output clamp

When large differential input voltage conditions are applied to the TLC2652, the nulling loop attempts to prevent the output from saturating by driving  $C_{XA}$  and  $C_{XB}$  to internally-clamped voltage levels. Once the overdrive condition is removed, a period of time is required to allow the built-up charge to dissipate. This time period is defined as overload recovery time (see Figure 33). Typical overload recovery time for the TLC2652 is significantly faster than competitive products; however, if required, this time can be reduced further by use of internal clamp circuitry accessible through CLAMP if required.

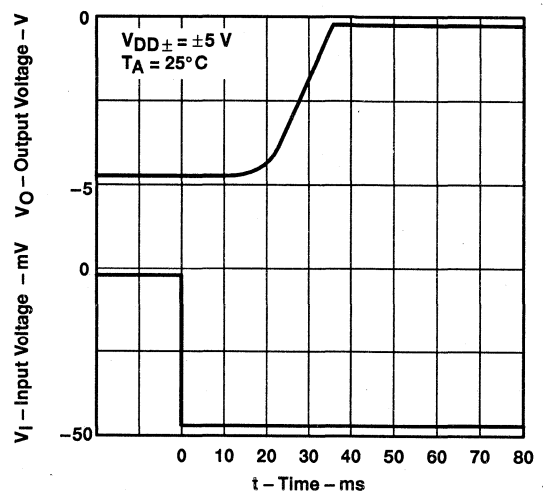


Figure 33. Overload Recovery

# TLC2652, TLC2652A, TLC2652Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS019B – SEPTEMBER 1988 – REVISED AUGUST 1994

---

### APPLICATION INFORMATION

#### overload recovery/output clamp (continued)

The clamp is a switch that is automatically activated when the output is approximately 1 V from either supply rail. When connected to the inverting input (in parallel with the closed-loop feedback resistor), the closed-loop gain is reduced, and the TLC2652 output is prevented from going into saturation. Since the output must source sink current through the switch (see Figure 7), the maximum output voltage swing is slightly reduced.

#### thermoelectric effects

To take advantage of the extremely low offset voltage drift of the TLC2652, care must be taken to compensate for the thermoelectric effects present when two dissimilar metals are brought into contact with each other (such as device leads being soldered to a printed circuit board). Dissimilar metal junctions can produce thermoelectric voltages in the range of several microvolts per degree Celsius (orders of magnitude greater than the  $0.01\text{-}\mu\text{V}/^\circ\text{C}$  typical of the TLC2652).

To help minimize thermoelectric effects, careful attention should be paid to component selection and circuit-board layout. Avoid the use of nonsoldered connections (such as sockets, relays, switches, etc.) in the input signal path. Cancel thermoelectric effects by duplicating the number of components and junctions in each device input. The use of low-thermoelectric-coefficient components, such as wire-wound resistors, is also beneficial.

#### latch-up avoidance

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC2652 inputs and output are designed to withstand  $\sim 100\text{-mA}$  surge currents without sustaining latch-up; however, techniques to reduce the chance of latch-up should be used whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ( $0.1\ \mu\text{F}$  typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the supply rails and is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor. The chance of latch-up occurring increases with increasing temperature and supply voltage.

#### electrostatic discharge protection

The TLC2652 incorporates internal ESD-protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

#### theory of operation

Chopper-stabilized operational amplifiers offer the best dc performance of any monolithic operational amplifier. This superior performance is the result of using two operational amplifiers, a main amplifier and a nulling amplifier, plus oscillator-controlled logic and two external capacitors to create a system that behaves as a single amplifier. With this approach, the TLC2652 achieves submicrovolt input offset voltage, submicrovolt noise voltage, and offset voltage variations with temperature in the  $\text{nV}/^\circ\text{C}$  range.

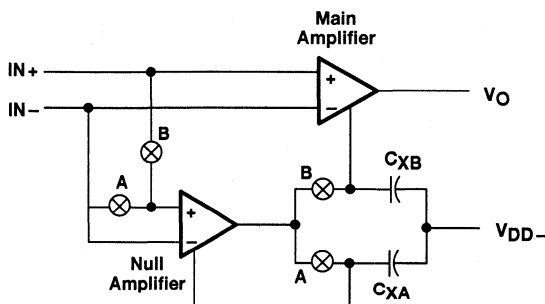
The TLC2652 on-chip control logic produces two dominant clock phases: a nulling phase and an amplifying phase. The term chopper-stabilized derives from the process of switching between these two clock phases. Figure 34 shows a simplified block diagram of the TLC2652. Switches A and B are make-before-break types.



## APPLICATION INFORMATION

### theory of operation (continued)

During the nulling phase, switch A is closed shorting the nulling amplifier inputs together and allowing the nulling amplifier to reduce its own input offset voltage by feeding its output signal back to an inverting input node. Simultaneously, external capacitor  $C_{XA}$  stores the nulling potential to allow the offset voltage of the amplifier to remain nulled during the amplifying phase.



**Figure 34. TLC2652 Simplified Block Diagram**

During the amplifying phase, switch B is closed connecting the output of the nulling amplifier to a noninverting input of the main amplifier. In this configuration, the input offset voltage of the main amplifier is nulled. Also, external capacitor  $C_{XB}$  stores the nulling potential to allow the offset voltage of the main amplifier to remain nulled during the next nulling phase.

This continuous chopping process allows offset voltage nulling during variations in time and temperature over the common-mode input voltage range and power supply range. In addition, because the low-frequency signal path is through both the null and main amplifiers, extremely high gain is achieved.

The low-frequency noise of a chopper amplifier depends on the magnitude of the component noise prior to chopping and the capability of the circuit to reduce this noise while chopping. The use of the Advanced LinCMOS process, with its low-noise analog MOS transistors and patent-pending input stage design, significantly reduces the input noise voltage.

The primary source of nonideal operation in chopper-stabilized amplifiers is error charge from the switches. As charge imbalance accumulates on critical nodes, input offset voltage can increase, especially with increasing chopping frequency. This problem has been significantly reduced in the TLC2652 by use of a patent-pending compensation circuit and the Advanced LinCMOS process.

The TLC2652 incorporates a feed-forward design that ensures continuous frequency response. Essentially, the gain magnitude of the nulling amplifier and compensation network crosses unity at the break frequency of the main amplifier. As a result, the high-frequency response of the system is the same as the frequency response of the main amplifier. This approach also ensures that the slewing characteristics remain the same during both the nulling and amplifying phases.



# TLC2654, TLC2654A, TLC2654Y

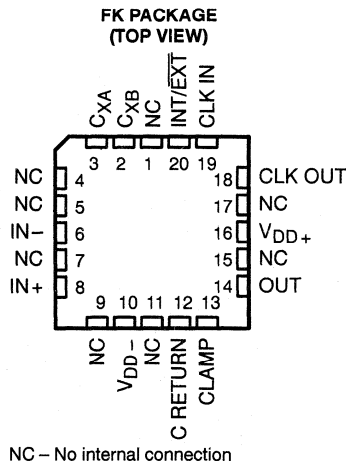
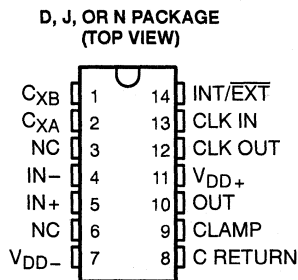
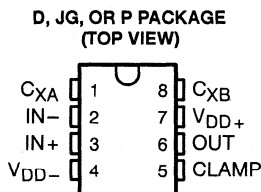
## Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

- **Input Noise Voltage**  
 0.5  $\mu\text{V}$  (Peak-to-Peak) Typ,  $f = 0$  to 1 Hz  
 1.5  $\mu\text{V}$  (Peak-to-Peak) Typ,  $f = 0$  to 10 Hz  
 47  $\text{nV}/\sqrt{\text{Hz}}$  Typ,  $f = 10$  Hz  
 13  $\text{nV}/\sqrt{\text{Hz}}$  Typ,  $f = 1$  kHz
- **High Chopping Frequency . . . 10 kHz Typ**
- **No Clock Noise Below 10 kHz**
- **No Intermodulation Error Below 5 kHz**
- **Low Input Offset Voltage**  
 10  $\mu\text{V}$  Max (TLC2654A)
- **Excellent Offset Voltage Stability With Temperature . . . 0.05  $\mu\text{V}/^\circ\text{C}$  Max**
- **$A_{\text{VD}}$  . . . 135 dB Min (TLC2654A)**
- **CMRR . . . 110 dB Min (TLC2654A)**
- **$k_{\text{SVR}}$  . . . 120 dB Min (TLC2654A)**
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**
- **No Noise Degradation With External Capacitors Connected to  $V_{\text{DD-}}$**

### description

The TLC2654 and TLC2654A are low-noise chopper-stabilized operational amplifiers using the Advanced LinCMOS™ process. Combining this process with chopper-stabilization circuitry makes excellent dc precision possible. In addition, circuit techniques are added that give the TLC2654 and TLC2654A noise performance unsurpassed by similar devices.



### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES							CHIP FORM (Y)
		8 PIN			14 PIN				
		SMALL OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)	SMALL OUTLINE (D)	CERAMIC DIP (J)	PLASTIC DIP (N)	CERAMIC DIP (FK)	
0°C to 70°C	10 $\mu\text{V}$ 20 mV	TLC2654AC-8D TLC2654C-8D	— —	TLC2654ACP TLC2654CP	TLC2654AC-14D TLC2654C-14D	— —	TLC2654ACN TLC2654CN	— —	TLC2654Y
-40°C to 85°C	10 $\mu\text{V}$ 20 $\mu\text{V}$	TLC2654AI-8D TLC2654I-8D	— —	TLC2654AIP TLC2654IP	TLC2654AI-14D TLC2654I-14D	— —	TLC2654AIN TLC2654IN	— —	—
-55°C to 125°C	10 $\mu\text{V}$ 20 $\mu\text{V}$	TLC2654AM-8D TLC2654M-8D	TLC2654AMJG TLC2654MJG	TLC2654AMP TLC2654MP	TLC2654AM-14D TLC2654M-14D	TLC2654AMJ TLC2654MJ	TLC2654AMN TLC2654MN	TLC2654AMFK TLC2654MFK	—

The 8-pin and 14-pin D packages are available taped and reeled. Add R suffix to device type (e.g., TLC2654AC-8DR).

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 1994, Texas Instruments Incorporated  
 On products compliant to MIL-STD-883, Class B, all parameters are tested unless otherwise noted. On all other products, production processing does not necessarily include testing of all parameters.

# TLC2654, TLC2654A, TLC2654Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

### description (continued)

Chopper-stabilization techniques provide for extremely high dc precision by continuously nulling input offset voltage even during variations in temperature, time, common-mode voltage, and power-supply voltage. The high chopping frequency of the TLC2654 and TLC2654A (see Figure 1) provides excellent noise performance in a frequency spectrum from near dc to 10 kHz. In addition, intermodulation or aliasing error is eliminated from frequencies up to 5 kHz.

This high dc precision and low noise, coupled with the extremely high input impedance of the CMOS input stage, makes the TLC2654 and TLC2654A ideal choices for a broad range of applications such as low-level, low-frequency thermocouple amplifiers and strain gauges and wide-bandwidth and subsonic circuits. For applications requiring even greater dc precision, use the TLC2652 or TLC2652A devices, which have a chopping frequency of 450 Hz.

The TLC2654 and TLC2654A common-mode input voltage range includes the negative rail, thereby providing superior performance in either single-supply or split-supply applications, even at power supply voltage levels as low as  $\pm 2.3$  V.

Two external capacitors are required to operate the device; however, the on-chip chopper-control circuitry is transparent to the user. On devices in the 14-pin and 20-pin packages, the control circuitry is accessible, allowing the user the option of controlling the clock frequency with an external frequency source. In addition, the clock threshold of the TLC2654 and TLC2654A requires no level shifting when used in the single-supply configuration with a normal CMOS or TTL clock input.

Innovative circuit techniques used on the TLC2654 and TLC2654A allow exceptionally fast overload recovery time. An output clamp pin is available to reduce the recovery time even further.

The device inputs and outputs are designed to withstand  $-100$ -mA surge currents without sustaining latch-up. In addition, the TLC2654 and TLC2654A incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, exercise care in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

The C-suffix devices are characterized for operation from  $0^{\circ}\text{C}$  to  $70^{\circ}\text{C}$ . The I-suffix devices are characterized for operation from  $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ . The M-suffix devices are characterized for operation over the full military temperature range of  $-55^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

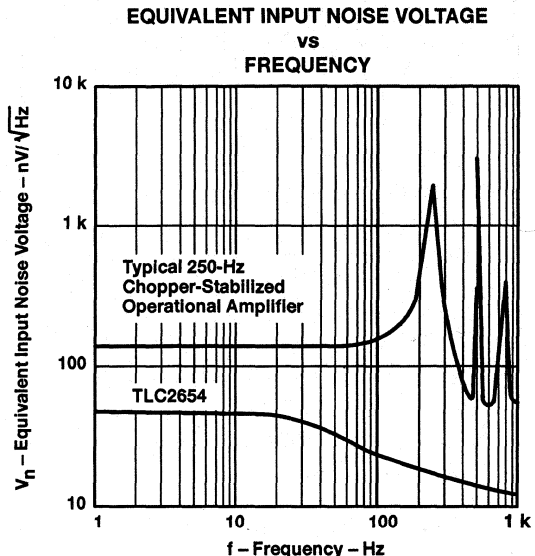


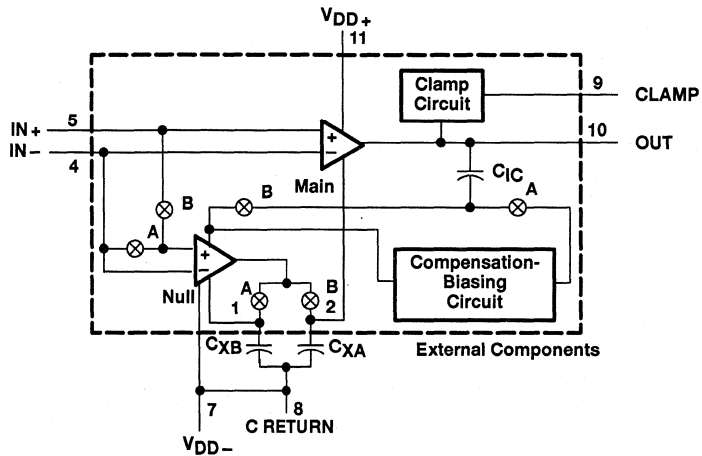
Figure 1

# TLC2654, TLC2654A, TLC2654Y

## Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D - NOVEMBER 1988 - REVISED AUGUST 1994

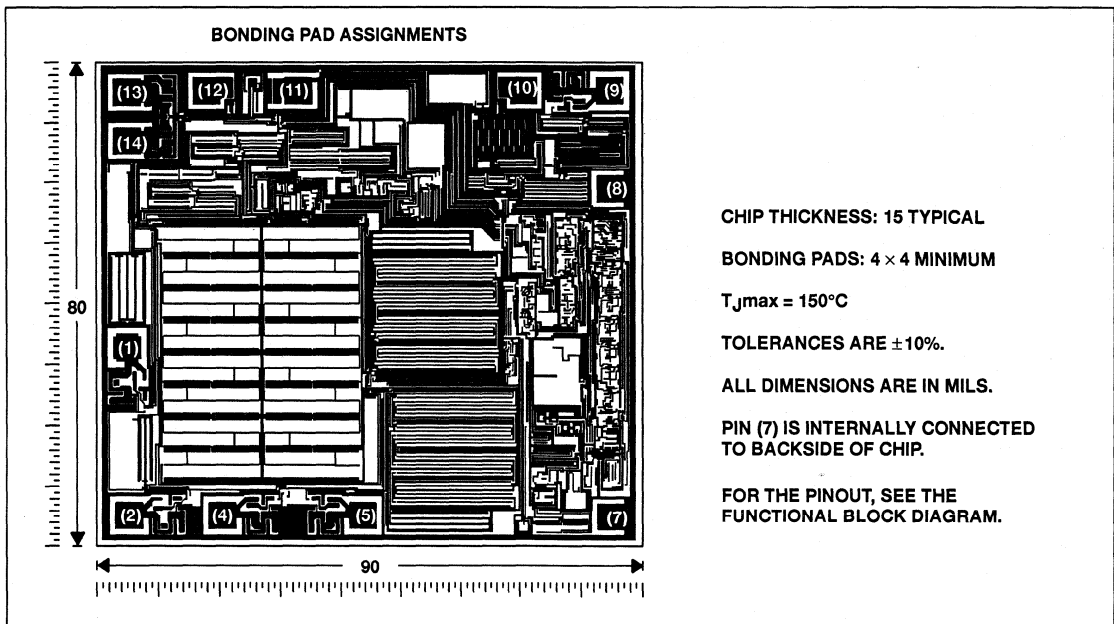
### functional block diagram



Pin numbers shown are for the D (14 pin), J, and N packages.

### TLC2654Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC2654C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



# TLC2654, TLC2654A, TLC2654Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	±16 V
Input voltage, $V_I$ (any input, see Note 1)	±8 V
Voltage range on CLK IN and INT/EXT	$V_{DD-}$ to $V_{DD-} + 5.2$ V
Input current, $I_I$ (each input)	±5 mA
Output current, $I_O$	±50 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Current into CLK IN and INT/EXT	±5 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J or JG package	300°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .
2. Differential voltages are at IN+ with respect to IN-.
3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 125^\circ\text{C}$
	POWER RATING		POWER RATING	POWER RATING	POWER RATING
D (8 pin)	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
D (14 pin)	950 mW	7.6 mW/°C	608 mW	494 mW	190 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	230 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

### recommended operating conditions

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, $V_{DD\pm}$	±2.3	±8	±2.3	±8	±2.3	±8	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	$V_{DD-}$	$V_{DD+} - 2.3$	V
Clock input voltage	$V_{DD-}$	$V_{DD-} + 5$	$V_{DD-}$	$V_{DD-} + 5$	$V_{DD-}$	$V_{DD-} + 5$	V
Operating free-air temperature, $T_A$	0	70	-40	85	-55	125	°C



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



# TLC2654, TLC2654A, TLC2654Y Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2654C			TLC2654AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage (see Note 4)	$V_{IC} = 0, R_S = 50 \Omega$	25°C	5 20			4 10			$\mu V$
		Full range	34			24			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.01 0.05			0.01 0.05			$\mu V/^\circ C$
Input offset voltage long-term drift (see Note 5)		25°C	0.003 0.06			0.003 0.02			$\mu V/mo$
$I_{IO}$ Input offset current		25°C	30			30			$pA$
		Full range	150			150			
$I_{IB}$ Input bias current	25°C	50			50			$pA$	
	Full range	150			150				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	Full range	-5 to 2.7			-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 k\Omega$ , See Note 6	25°C	4.7 4.8			4.7 4.8			V
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10 k\Omega$ , See Note 6	25°C	-4.7 -4.9			-4.7 -4.9			V
		Full range	-4.7			-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4$ V, $R_L = 10 k\Omega$	25°C	120 155			135 155			dB
		Full range	120			130			
Internal chopping frequency		25°C	10			10			kHz
Clamp on-state current	$R_L = 100 k\Omega$	25°C	25			25			$\mu A$
		Full range	25			25			
Clamp off-state current	$V_O = -4$ V to 4 V	25°C	100			100			$pA$
		Full range	100			100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50 \Omega$	25°C	105 125			110 125			dB
		Full range	105			110			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3$ V to $\pm 8$ V, $V_O = 0, R_S = 50 \Omega$	25°C	110 125			120 125			dB
		Full range	110			120			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.5 2.4			1.5 2.4			mA
		Full range	2.5			2.5			

† Full range is 0°C to 70°C.

NOTES: 4. This parameter is not production tested full range. Thermocouple effects preclude measurement of the actual  $V_{IO}$  of these devices in high-speed automated testing.  $V_{IO}$  is measured to a limit determined by the test equipment capability at the temperature extremes. The test ensures that the stabilization circuitry is performing properly.

5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ C$  extrapolated to  $T_A = 25^\circ$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

6. Output clamp is not connected.



**TLC2654. TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2654C			TLC2654AC			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	1.5	2		1.5	2		V/μs
		Full range	1.3			1.3			
SR- Negative slew rate at unity gain		25°C	2.3	3.7		2.3	3.7		V/μs
		Full range	1.7			1.7			
V <sub>n</sub> Equivalent input noise voltage (see Note 7)	f = 10 Hz	25°C		47		47	75	nV/√Hz	
	f = 1 kHz	25°C		13		13	20		
V <sub>N(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C		0.5		0.5		μV	
	f = 0 to 10 Hz	25°C		1.5		1.5			
I <sub>n</sub> Equivalent input noise current	f = 10 kHz	25°C		0.004		0.004		pA/√Hz	
Gain-bandwidth product	f = 10 kHz, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		1.9		1.9		MHz	
φ <sub>m</sub> Phase margin at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		48°		48°			

† Full range is 0°C to 70°C.

NOTE 7: This parameter is tested on a sample basis for the TLC2654A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2654I			TLC2654AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage (see Note 4)		25°C	5		20	4		10	$\mu$ V
		Full range			40			30	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.01		0.05	0.01		0.05	$\mu$ V/°C
Input offset voltage long-term drift (see Note 5)	$V_{IC} = 0, R_S = 50 \Omega$	25°C	0.003		0.06	0.003		0.02	$\mu$ V/mo
		Full range							
$I_{IO}$ Input offset current		25°C	30			30			pA
		Full range			200			200	
$I_{IB}$ Input bias current		25°C	50			50			pA
		Full range			200			200	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	Full range	-5 to 2.7			-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 \text{ k}\Omega$ , See Note 6	25°C	4.7	4.8		4.7	4.8		V
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10 \text{ k}\Omega$ , See Note 6	25°C	-4.7	-4.9		-4.7	-4.9		V
		Full range	-4.7			-4.7			
AVD Large-signal differential voltage amplification	$V_O = \pm 4$ V, $R_L = 10 \text{ k}\Omega$	25°C	120	155		135	155		dB
		Full range	120			125			
Internal chopping frequency		25°C	10			10		kHz	
Clamp on-state current	$R_L = 100 \text{ k}\Omega$	25°C	25			25		$\mu$ A	
		Full range	25			25			
Clamp off-state current	$V_O = -4$ V to 4 V	25°C			100	100		pA	
		Full range			100	100			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50 \Omega$	25°C	105	125		110	125		dB
		Full range	105			110			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3$ V to $\pm 8$ V, $V_O = 0, R_S = 50 \Omega$	25°C	110	125		120	125		dB
		Full range	110			120			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.5		2.4	1.5		2.4	mA
		Full range			2.5			2.5	

† Full range is -40°C to 85°C

- NOTES: 4. This parameter is not production tested full range. Thermocouple effects preclude measurement of the actual  $V_{IO}$  of these devices in high-speed automated testing.  $V_{IO}$  is measured to a limit determined by the test equipment capability at the temperature extremes. The test ensures that the stabilization circuitry is performing properly.
5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.
6. Output clamp is not connected.



**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$**

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2654I			TLC2654AI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR+	Positive slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	1.5	2		1.5	2	V/μs
			Full range	1.2			1.2		
SR-	Negative slew rate at unity gain	V <sub>O</sub> = ±2.3 V, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C	2.3	3.7		2.3	3.7	V/μs
			Full range	1.5			1.5		
V <sub>n</sub>	Equivalent input noise voltage (see Note 7)	f = 10 Hz	25°C		47		47	75	nV/√Hz
			25°C		13		13	20	
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage	f = 0 to 10 Hz	25°C		0.5		0.5		μV
			25°C		1.5		1.5		
I <sub>n</sub>	Equivalent input noise current	f = 10 kHz	25°C		0.004		0.004	pA/√Hz	
	Gain-bandwidth product	f = 10 kHz, R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		1.9		1.9	MHz	
φ <sub>m</sub>	Phase margin at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 100 pF	25°C		48°		48°		

† Full range is -40°C to 85°C.

NOTE 7: This parameter is tested on a sample basis for the TLC2654A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} \pm = \pm 5$  V (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2654M			TLC2654AM			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
$V_{IO}$ Input offset voltage (see Note 4)	$V_{IC} = 0, R_S = 50 \Omega$	25°C	5 20			4 10			$\mu V$
		Full range	50			40			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		Full range	0.01 0.05*			0.01 0.05*			$\mu V/^\circ C$
Input offset voltage long-term drift (see Note 5)		25°C	0.003 0.06*			0.003 0.02*			$\mu V/mo$
$I_{IO}$ Input offset current		25°C	30			30			$\mu A$
		Full range	500			500			
$I_{IB}$ Input bias current	25°C	50			50			$\mu A$	
	Full range	500			500				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50 \Omega$	Full range	-5 to 2.7			-5 to 2.7			V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10 k\Omega$ See Note 6	25°C	4.7 4.8			4.7 4.8			V
		Full range	4.7			4.7			
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10 k\Omega$ See Note 6	25°C	-4.7 -4.9			-4.7 -4.9			V
		Full range	-4.7			-4.7			
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4$ V, $R_L = 10 k\Omega$	25°C	120 155			135 155			dB
		Full range	120			120			
Internal chopping frequency		25°C	10			10			kHz
Clamp on-state current	$R_L = 100 k\Omega$	25°C	25			25			$\mu A$
		Full range	25			25			
Clamp off-state current	$V_O = -4$ V to 4 V	25°C	100			100			$\mu A$
		Full range	500			500			
CMRR Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICRmin}, R_S = 50 \Omega$	25°C	105 125			110 125			dB
		Full range	105			110			
KSVR Supply voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3$ V to $\pm 8$ V, $V_O = 0, R_S = 50 \Omega$	25°C	110 125			120 125			dB
		Full range	105			115			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C	1.5 2.4			1.5 2.4			mA
		Full range	2.5			2.5			

\* On products compliant to MIL-STD-883, Class B, this parameter is not production tested.

† Full range is  $-55^\circ C$  to  $125^\circ C$ .

- NOTES: 4. This parameter is not production tested full range. Thermocouple effects preclude measurement of the actual  $V_{IO}$  of these devices in high-speed automated testing.  $V_{IO}$  is measured to a limit determined by the test equipment capability at the temperature extremes. The test ensures that the stabilization circuitry is performing properly.
5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ C$  extrapolated to  $T_A = 25^\circ C$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.
6. Output clamp is not connected.

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	T <sub>A</sub> †	TLC2654M TLC2654AM			UNIT
			MIN	TYP	MAX	
SR+ Positive slew rate at unity gain	$V_O = \pm 2.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	1.5	2	V/ $\mu\text{s}$	
		Full range	1.1			
SR- Negative slew rate at unity gain		25°C	2.3	3.7	V/ $\mu\text{s}$	
		Full range	1.3			
V <sub>n</sub> Equivalent input noise voltage	f = 10 Hz	25°C	47		nV/ $\sqrt{\text{Hz}}$	
	f = 1 kHz	25°C	13			
V <sub>N(PP)</sub> Peak-to-peak equivalent input noise voltage	f = 0 to 1 Hz	25°C	0.5		$\mu\text{V}$	
	f = 0 to 10 Hz	25°C	1.5			
I <sub>n</sub> Equivalent input noise current	f = 1 kHz	25°C	0.004		pA/ $\sqrt{\text{Hz}}$	
Gain-bandwidth product	f = 10 kHz, $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	1.9		MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$	25°C	48°			

† Full range is -55°C to 125°C.



# TLC2654, TLC2654A, TLC2654Y

## Advanced LinCMOS™ LOW-NOISE CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

### electrical characteristics, $V_{DD} \pm = \pm 5\text{ V}$ , $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC2654Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage (see Note 4)	$V_{IC} = 0$ , $R_S = 50\ \Omega$		5	20	$\mu\text{V}$
Input offset voltage long-term drift (see Note 5)			0.003	0.06	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current				30	$\text{pA}$
$I_{IB}$ Input bias current			50	$\text{pA}$	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$		-5 to 2.7		V
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5		4.7	4.8	V
$V_{OM-}$ Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$ , See Note 5		-4.7	-4.9	V
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}$ , $R_L = 10\ \text{k}\Omega$		120	155	dB
$f_{ch}$ Internal chopping frequency			10		Hz
Clamp on-state current	$R_L = 100\ \text{k}\Omega$		25		$\mu\text{A}$
Clamp off-state current	$V_O = -4\ \text{V}$ to $4\ \text{V}$			100	$\text{pA}$
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}$ , $V_O = 0$ , $R_S = 50\ \Omega$		105	125	dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} \pm = \pm 2.3\ \text{V}$ to $\pm 8\ \text{V}$ , $V_O = 0$ , $R_S = 50\ \Omega$		110	125	dB
$I_{DD}$ Supply current	$V_O = 0$ , No load		1.5	2.4	$\text{mA}$

- NOTES: 4. This parameter is not production tested full range. Thermocouple effects preclude measurement of the actual  $V_{IO}$  of these devices in high-speed automated testing.  $V_{IO}$  is measured to a limit determined by the test equipment capability at the temperature extremes. The test ensures that the stabilization circuitry is performing properly.
5. Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

### operating characteristics, $V_{DD} \pm = \pm 5\ \text{V}$ , $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC2654Y			UNIT
		MIN	TYP	MAX	
$SR+$ Positive slew rate at unity gain	$V_O = \pm 2.3\ \text{V}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.5	2	$\text{V}/\mu\text{s}$
$SR-$ Negative slew rate at unity gain			2.3	3.7	$\text{V}/\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		47		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$		13		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0$ to $1\ \text{Hz}$		0.5		$\mu\text{V}$
	$f = 0$ to $10\ \text{Hz}$		1.5		
$I_n$ Equivalent input noise current	$f = 1\ \text{kHz}$		0.004		$\text{pA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.9		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		48°		

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

		<b>FIGURE</b>	
$V_{IO}$	Input offset voltage	Distribution	1
	Normalized input offset voltage	vs Chopping frequency	2
$I_{IO}$	Input offset current	vs Chopping frequency	3
		vs Free-air temperature	4
$I_{IB}$	Input bias current	vs Common-mode input voltage	5
		vs Chopping frequency	6
		vs Free-air temperature	7
	Clamp current	vs Output voltage	8
$V_{OM}$	Maximum peak output voltage swing	vs Output current	9
		vs Free-air temperature	10
$V_{O(PP)}$	Maximum peak-to-peak output voltage swing	vs Frequency	11
$CMRR$	Common-mode rejection ratio	vs Frequency	12
$AVD$	Large-signal differential voltage amplification	vs Frequency	13
		vs Free-air temperature	14
	Chopping frequency	vs Supply voltage	15
		vs Free-air temperature	16
$I_{DD}$	Supply current	vs Supply voltage	17
		vs Free-air temperature	18
$I_{OS}$	Short-circuit output current	vs Supply voltage	19
		vs Free-air temperature	20
$SR$	Slew rate	vs Supply voltage	21
		vs Free-air temperature	22
	Pulse response	Small signal	23
		Large signal	24
$V_{N(PP)}$	Peak-to-peak input noise voltage	vs Chopping frequency	25, 26
$V_n$	Equivalent input noise voltage	vs Frequency	27
$k_{SVR}$	Supply voltage rejection ratio	vs Frequency	28
		Gain-bandwidth product	29
		vs Supply voltage	30
		vs Free-air temperature	30
$\phi_m$	Phase margin	vs Supply voltage	31
		vs Load capacitance	32
	Phase shift	vs Frequency	13





TYPICAL CHARACTERISTICS†

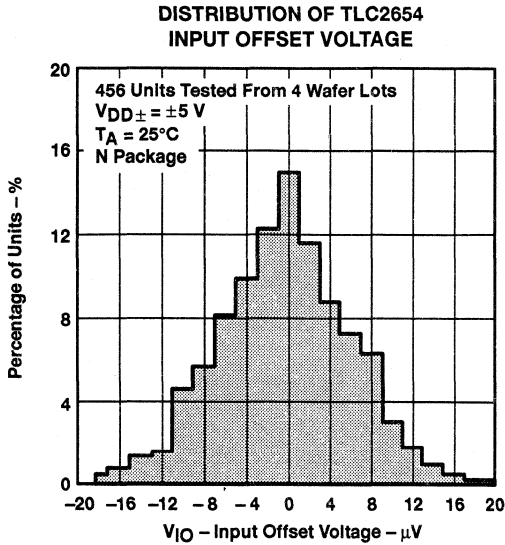


Figure 2

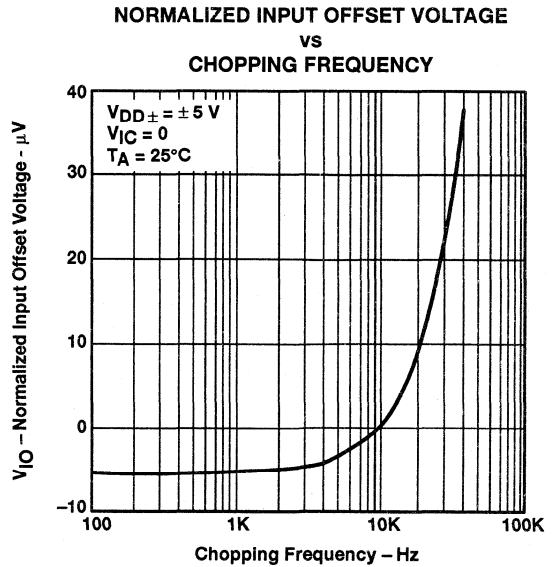


Figure 3

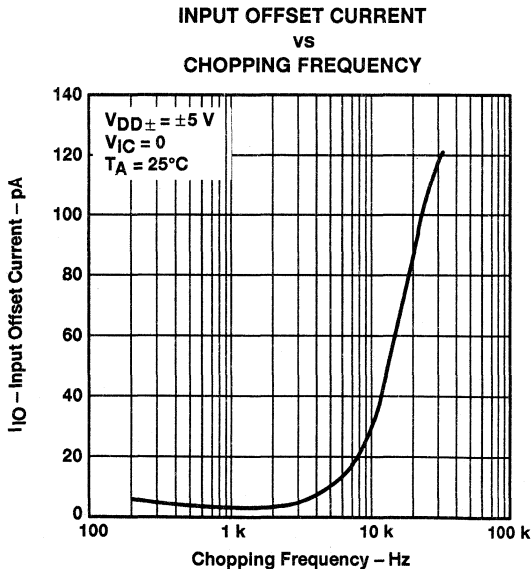


Figure 4

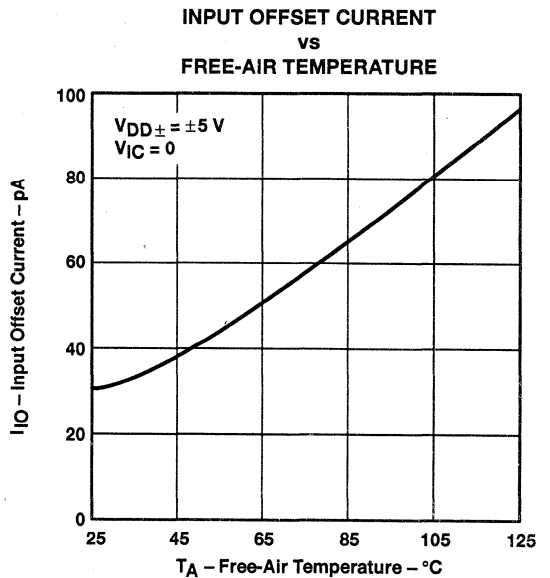


Figure 5

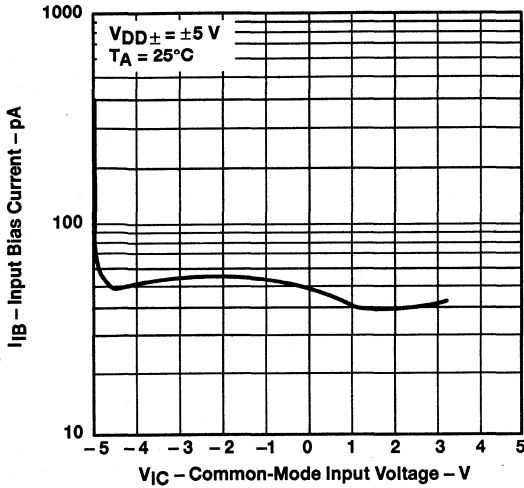
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

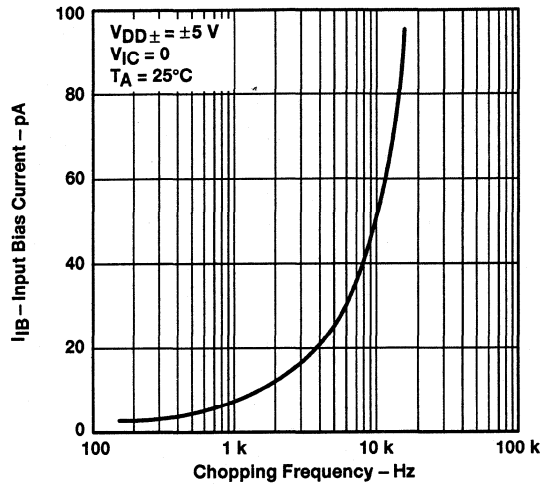
**TYPICAL CHARACTERISTICS†**

**INPUT BIAS CURRENT**  
**vs**  
**COMMON-MODE INPUT VOLTAGE**



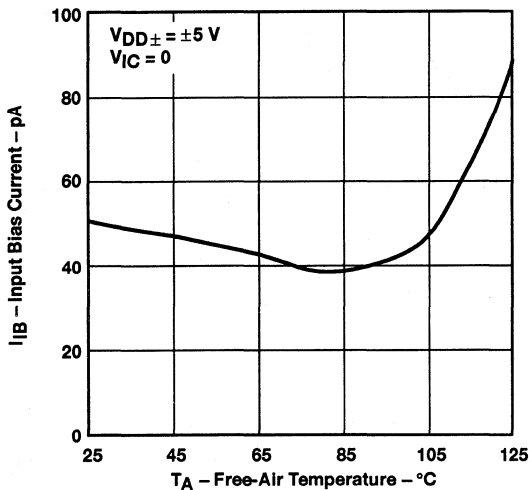
**Figure 6**

**INPUT BIAS CURRENT**  
**vs**  
**CHOPPING FREQUENCY**



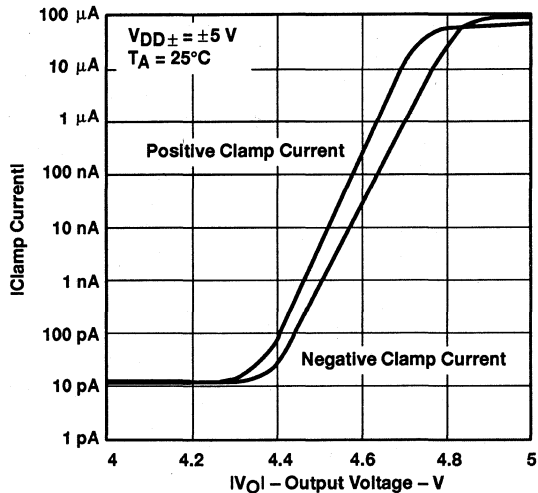
**Figure 7**

**INPUT BIAS CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



**Figure 8**

**CLAMP CURRENT**  
**vs**  
**OUTPUT VOLTAGE**



**Figure 9**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

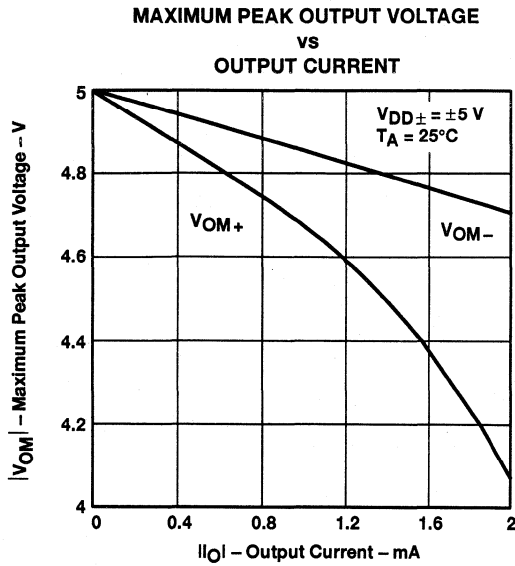


Figure 10

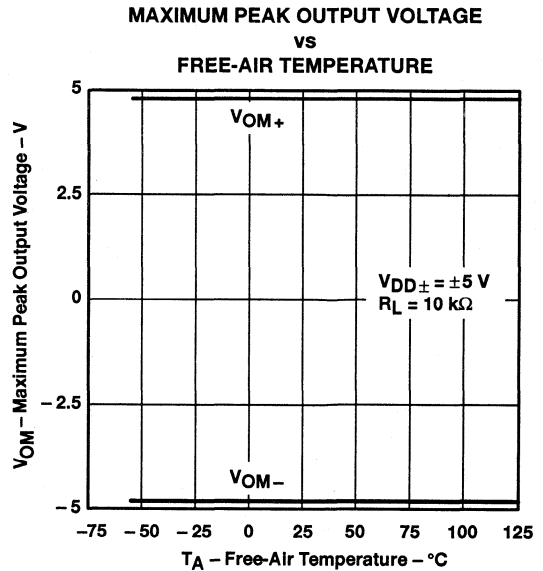


Figure 11

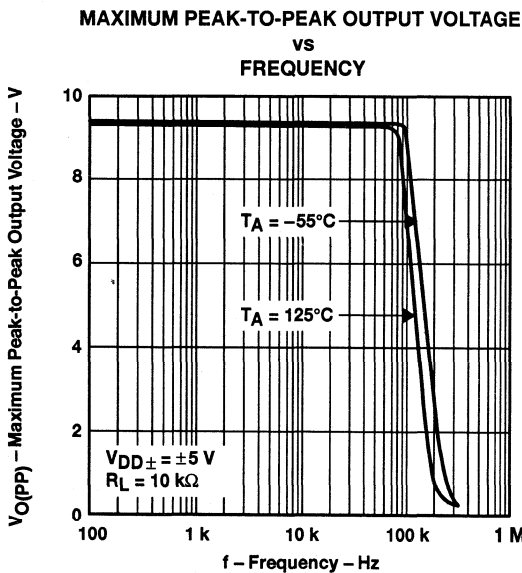


Figure 12

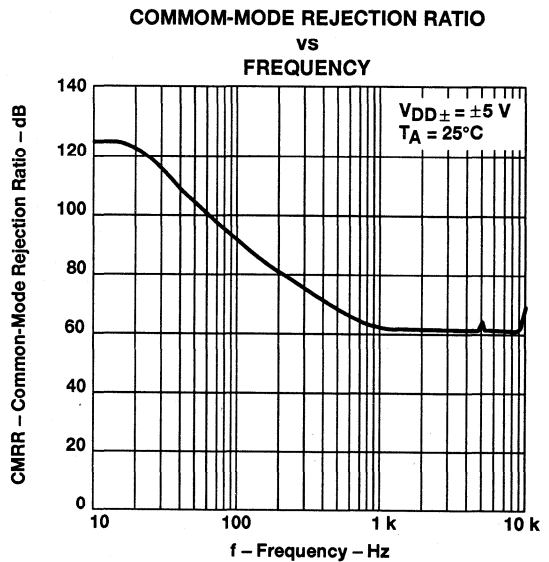


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT vs FREQUENCY**

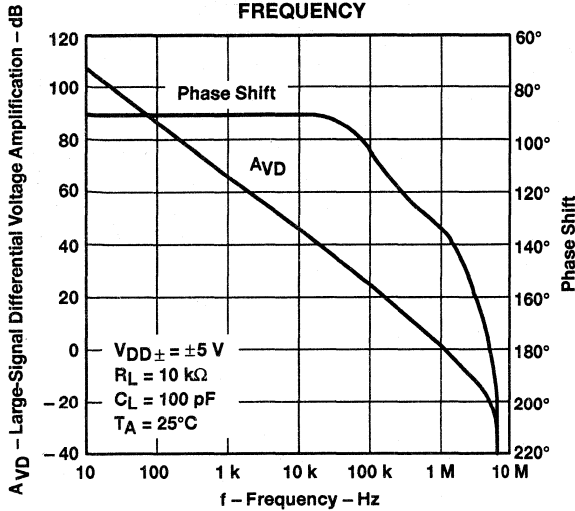


Figure 14

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs FREE-AIR TEMPERATURE**

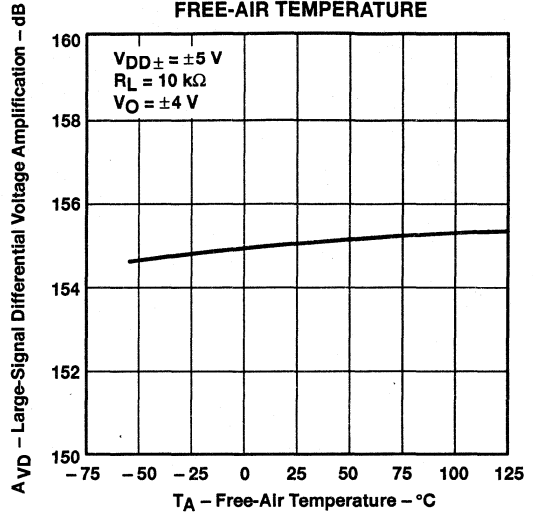


Figure 15

**CHOPPING FREQUENCY vs SUPPLY VOLTAGE**

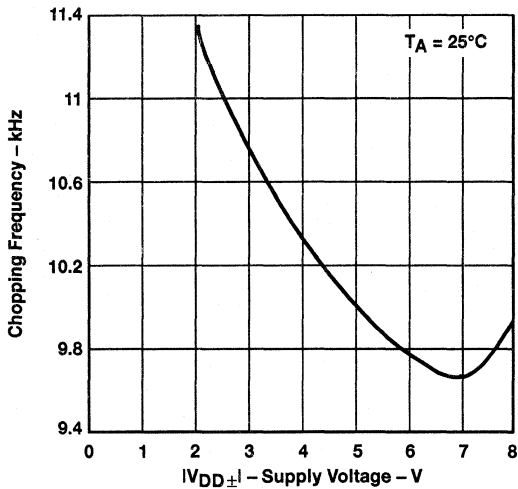


Figure 16

**CHOPPING FREQUENCY vs FREE-AIR TEMPERATURE**

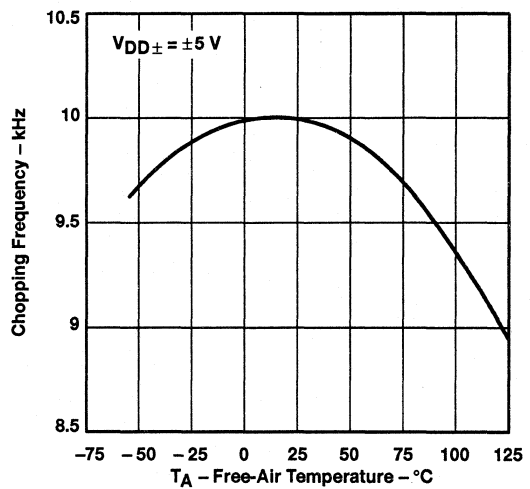


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS†

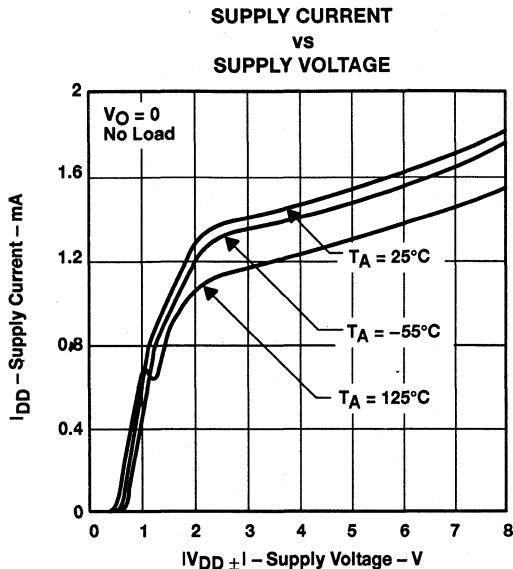


Figure 18

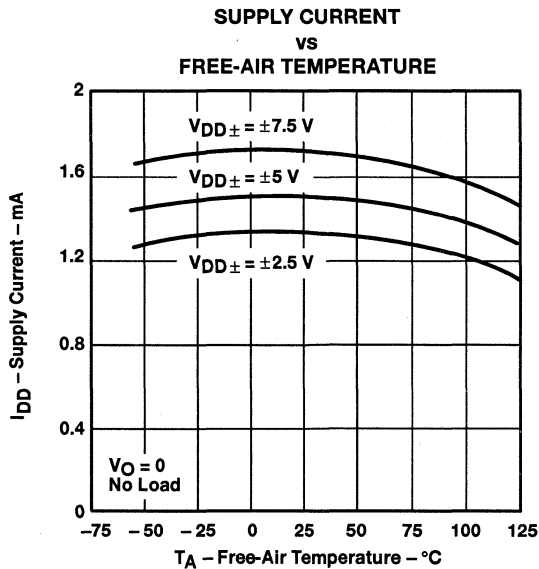


Figure 19

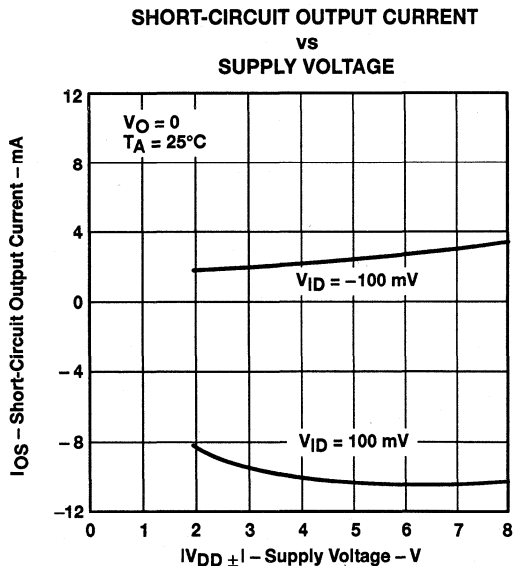


Figure 20

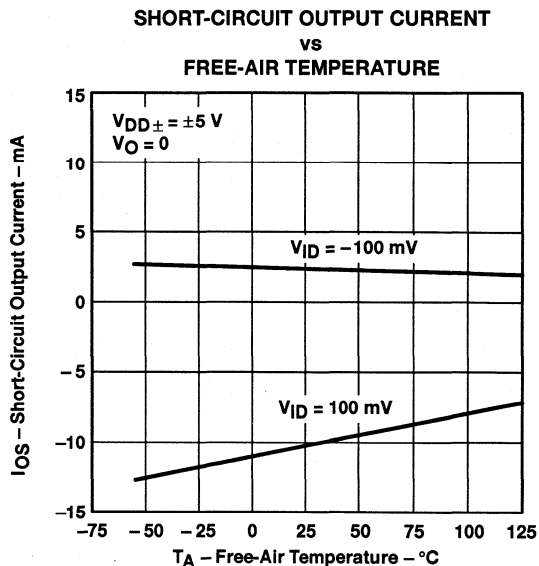


Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS†**

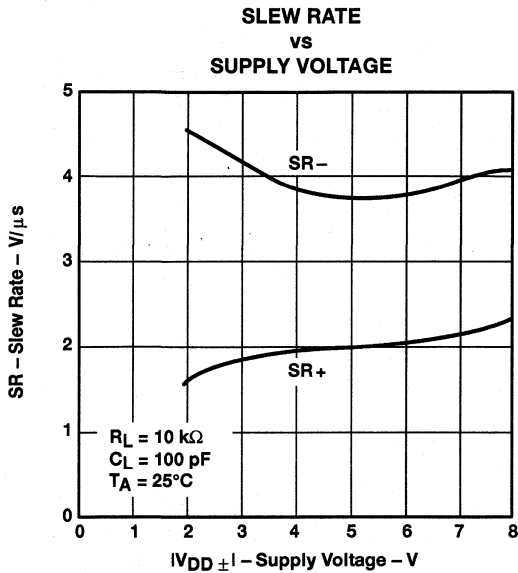


Figure 22

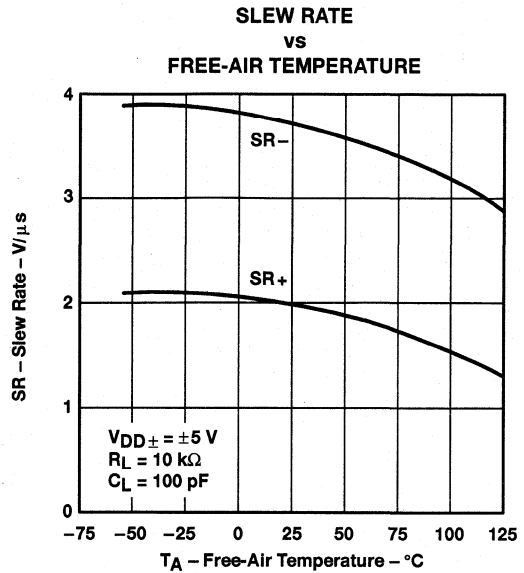


Figure 23

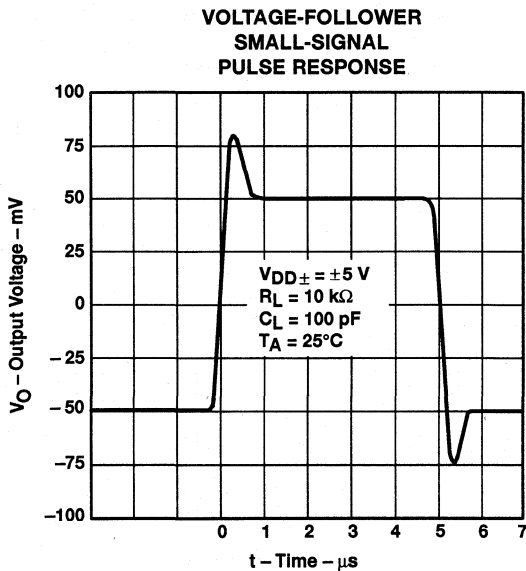


Figure 24

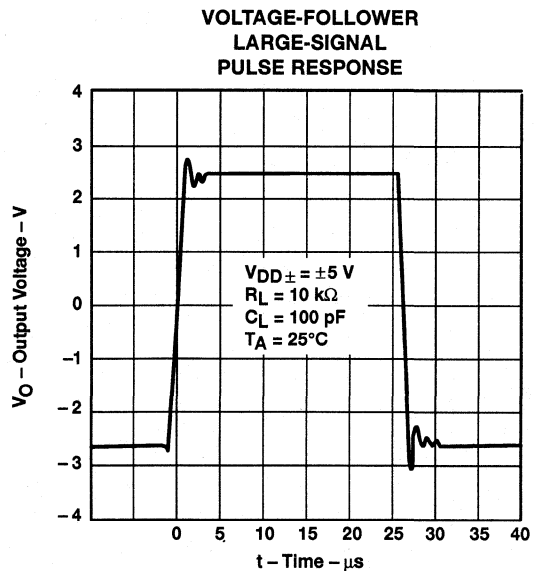


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

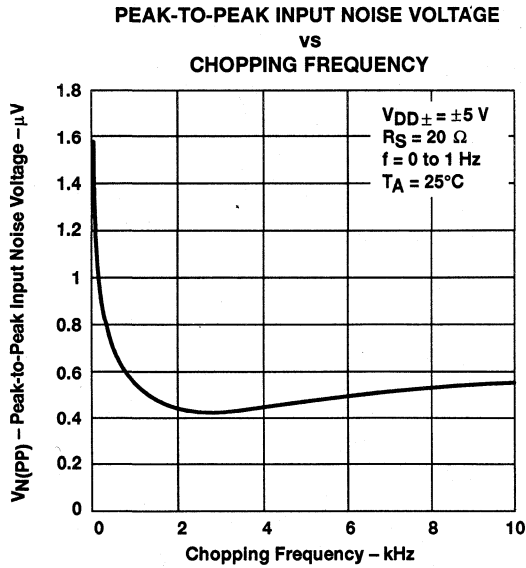


Figure 26

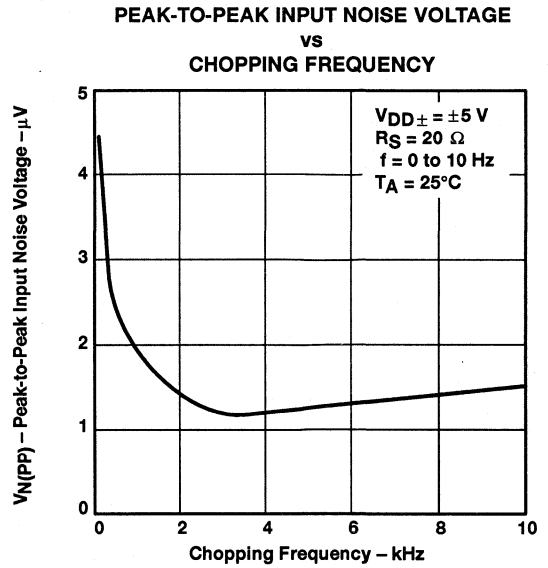


Figure 27

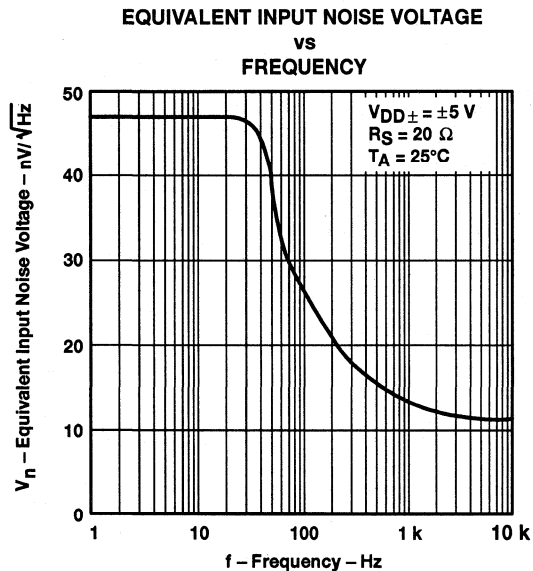


Figure 28

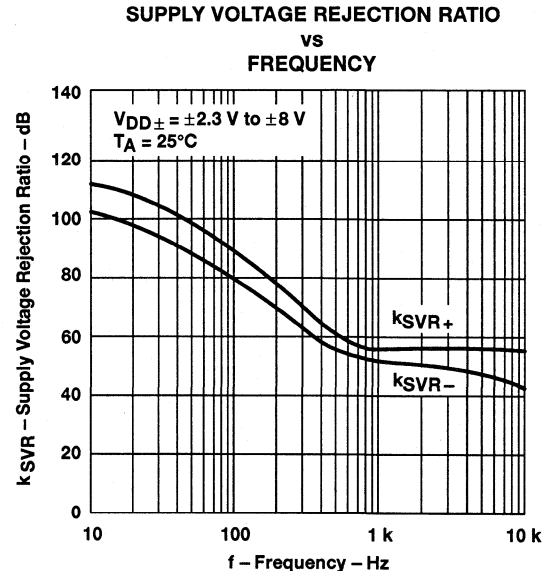


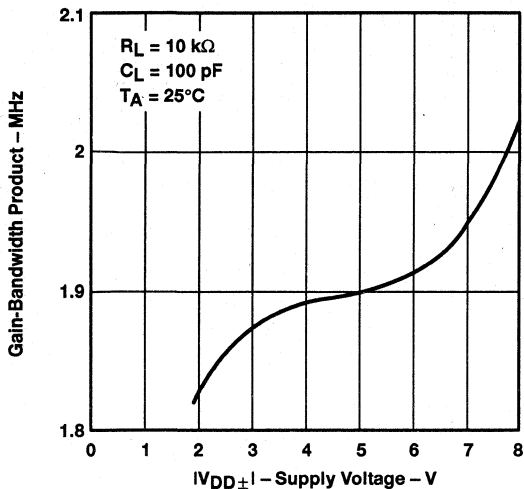
Figure 29

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

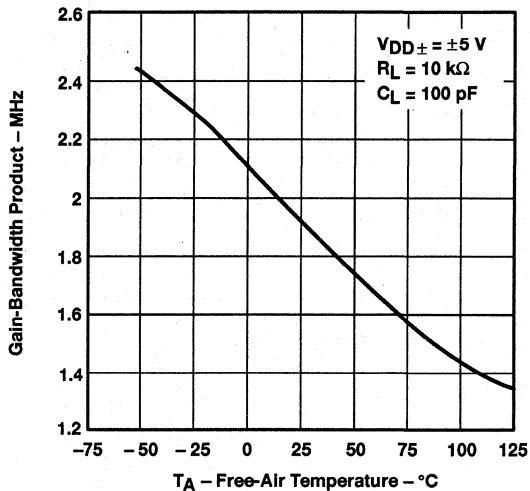
**TYPICAL CHARACTERISTICS†**

**GAIN-BANDWIDTH PRODUCT**  
**vs**  
**SUPPLY VOLTAGE**



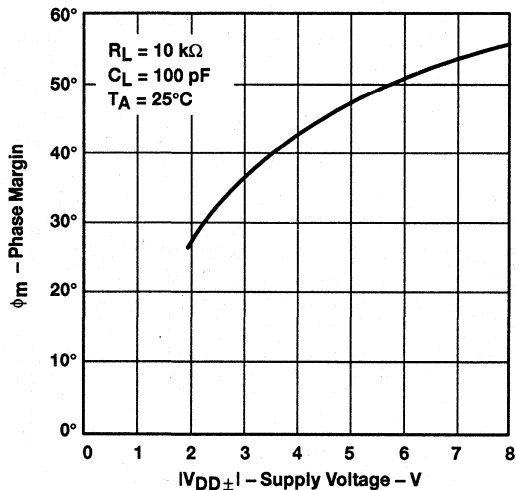
**Figure 30**

**GAIN-BANDWIDTH PRODUCT**  
**vs**  
**FREE-AIR TEMPERATURE**



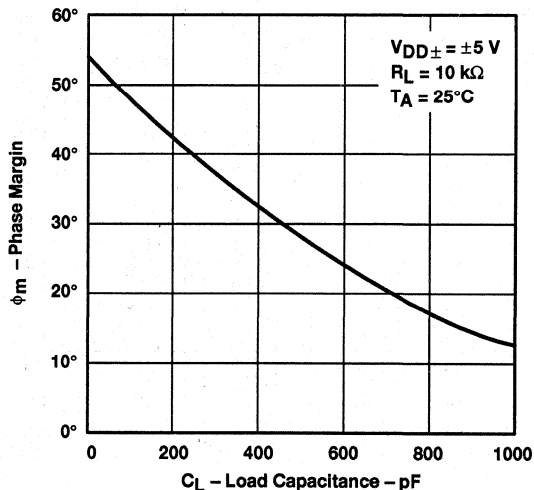
**Figure 31**

**PHASE MARGIN**  
**vs**  
**SUPPLY VOLTAGE**



**Figure 32**

**PHASE MARGIN**  
**vs**  
**LOAD CAPACITANCE**



**Figure 33**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.





## APPLICATION INFORMATION

### capacitor selection and placement

Leakage and dielectric absorption are the two important factors to consider when selecting external capacitors  $C_{XA}$  and  $C_{XB}$ . Both factors can cause system degradation, negating the performance advantages realized by using the TLC2654.

Degradation from capacitor leakage becomes more apparent with increasing temperatures. Low-leakage capacitors and standoffs are recommended for operation at  $T_A = 125^\circ\text{C}$ . In addition, guard bands are recommended around the capacitor connections on both sides of the printed-circuit board to alleviate problems caused by surface leakage on circuit boards.

Capacitors with high dielectric absorption tend to take several seconds to settle upon application of power, which directly affects input offset voltage. In applications needing fast settling of input voltage, high-quality film capacitors such as mylar, polystyrene, or polypropylene should be used. In other applications, a ceramic or other low-grade capacitor can suffice.

Unlike many choppers available today, the TLC2654 is designed to function with values of  $C_{XA}$  and  $C_{XB}$  in the range of  $0.1\ \mu\text{F}$  to  $1\ \mu\text{F}$  without degradation to input offset voltage or input noise voltage. These capacitors should be located as close as possible to  $C_{XA}$  and  $C_{XB}$  and return to either  $V_{DD-}$  or C RETURN. On many choppers, connecting these capacitors to  $V_{DD-}$  causes degradation in noise performance; this problem is eliminated on the TLC2654.

### internal/external clock

The TLC2654 has an internal clock that sets the chopping frequency to a nominal value of 10 kHz. On 8-pin packages, the chopping frequency can only be controlled by the internal clock; however, on all 14-pin packages and the 20-pin FK package the device chopping frequency can be set by the internal clock or controlled externally by use of the INT/EXT and CLK IN. To use the internal 10-kHz clock, no connection is necessary. If external clocking is desired, connect INT/EXT to  $V_{DD-}$  and the external clock to CLK IN. The external clock trip point is 2.5 V above the negative rail; however, CLK IN can be driven from the negative rail to 5 V above the negative rail. This allows the TLC2654 to be driven directly by 5-V TTL and CMOS logic when operating in the single-supply configuration. If this 5-V level is exceeded, damage could occur to the device unless the current into CLK IN is limited to  $\pm 5\ \text{mA}$ . A divide-by-two frequency divider interfaces with CLK IN and sets the chopping frequency appearing on CLK OUT.

### overload recovery/output clamp

When large differential-input-voltage conditions are applied to the TLC2654, the nulling loop attempts to prevent the output from saturating by driving  $C_{XA}$  and  $C_{XB}$  to internally-clamped voltage levels. Once the overdrive condition is removed, a period of time is required to allow the built-up charge to dissipate. This time period is defined as overload recovery time (see Figure 34). Typical overload recovery time for the TLC2654 is significantly faster than competitive products; however, this time can be reduced further by use of internal clamp circuitry accessible through CLAMP if required.

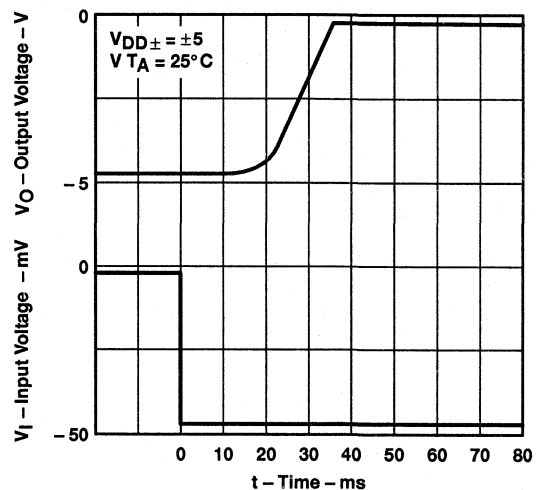


Figure 34. Overload Recovery

# TLC2654, TLC2654A, TLC2654Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

---

### APPLICATION INFORMATION

#### overload recovery/output clamp (continued)

The clamp is a switch that is automatically activated when the output is approximately 1 V from either supply rail. When connected to the inverting input (in parallel with the closed-loop feedback resistor), the closed-loop gain is reduced and the TLC2654 output is prevented from going into saturation. Since the output must source or sink current through the switch (see Figure 9), the maximum output voltage swing is slightly reduced.

#### thermoelectric effects

To take advantage of the extremely low offset voltage temperature coefficient of the TLC2654, care must be taken to compensate for the thermoelectric effects present when two dissimilar metals are brought into contact with each other (such as device leads being soldered to a printed-circuit board). It is not uncommon for dissimilar metal junctions to produce thermoelectric voltages in the range of several microvolts per degree Celsius (orders of magnitude greater than the 0.01  $\mu\text{V}/^\circ\text{C}$  typical of the TLC2654).

To help minimize thermoelectric effects, pay careful attention to component selection and circuit-board layout. Avoid the use of nonsoldered connections (such as sockets, relays, switches, etc.) in the input signal path. Cancel thermoelectric effects by duplicating the number of components and junctions in each device input. The use of low-thermoelectric-coefficient components, such as wire-wound resistors, is also beneficial.

#### latch-up avoidance

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC2654 inputs and outputs are designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up; however, techniques to reduce the chance of latch-up should be used whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be stunted by the use of decoupling capacitors (0.1  $\mu\text{F}$  typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the supply rails and is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor. The chance of latch-up occurring increases with increasing temperature and supply voltage.

#### electrostatic-discharge protection

The TLC2654 incorporates internal ESD-protection circuits that prevent functional failures at voltages at or below 2000 V. Care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

#### theory of operation

Chopper-stabilized operational amplifiers offer the best dc performance of any monolithic operational amplifier. This superior performance is the result of using two operational amplifiers – a main amplifier and a nulling amplifier – plus oscillator-controlled logic and two external capacitors to create a system that behaves as a single amplifier. With this approach, the TLC2654 achieves submicrovolt input offset voltage, submicrovolt noise voltage, and offset voltage variations with temperature in the  $\text{nV}/^\circ\text{C}$  range.

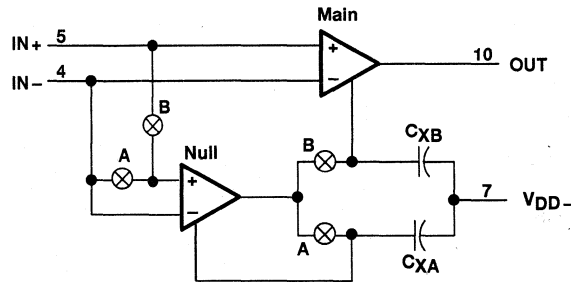
The TLC2654 on-chip control logic produces two dominant clock phases: a nulling phase and an amplifying phase. The term chopper-stabilized derives from the process of switching between these two clock phases. Figure 35 shows a simplified block diagram of the TLC2654. Switches A and B are make-before-break types.



## APPLICATION INFORMATION

### theory of operation (continued)

During the nulling phase, switch A is closed, shorting the nulling amplifier inputs together and allowing the nulling amplifier to reduce its own input offset voltage by feeding its output signal back to an inverting input node. Simultaneously, external capacitor  $C_{XA}$  stores the nulling potential to allow the offset voltage of the amplifier to remain nulled during the amplifying phase.



Pin numbers shown are for the D (14 pin), J, and N packages.

**Figure 35. TLC2654 Simplified Block Diagram**

During the amplifying phase, switch B is closed, connecting the output of the nulling amplifier to a noninverting input of the main amplifier. In this configuration, the input offset voltage of the main amplifier is nulled. Also, external capacitor  $C_{XB}$  stores the nulling potential to allow the offset voltage of the main amplifier to remain nulled during the next nulling phase.

This continuous chopping process allows offset voltage nulling during variations in time and temperature and over the common-mode input voltage range and power supply range. In addition, because the low-frequency signal path is through both the null and main amplifiers, extremely high gain is achieved.

The low-frequency noise of a chopper amplifier depends on the magnitude of the component noise prior to chopping and the capability of the circuit to reduce this noise while chopping. The use of the Advanced LinCMOS process, with its low-noise analog MOS transistors and patent-pending input stage design, significantly reduces the input noise voltage.

The primary source of nonideal operation in chopper-stabilized amplifiers is error charge from the switches. As charge imbalance accumulates on critical nodes, input offset voltage can increase especially with increasing chopping frequency. This problem has been significantly reduced in the TLC2654 by use of a patent-pending compensation circuit and the Advanced LinCMOS process.

The TLC2654 incorporates a feed-forward design that ensures continuous frequency response. Essentially, the gain magnitude of the nulling amplifier and compensation network crosses unity at the break frequency of the main amplifier. As a result, the high-frequency response of the system is the same as the frequency response of the main amplifier. This approach also ensures that the slewing characteristics remain the same during both the nulling and amplifying phases.

The primary limitation on ac performance is the chopping frequency. As the input signal frequency approaches the chopper's clock frequency, intermodulation (or aliasing) errors result from the mixing of these frequencies. To avoid these error signals, the input frequency must be less than half the clock frequency. Most choppers available today limit the internal chopping frequency to less than 500 Hz in order to eliminate errors due to the charge imbalancing phenomenon mentioned previously. However, to avoid intermodulation errors on a 500-Hz chopper, the input signal frequency must be limited to less than 250 Hz.

**TLC2654, TLC2654A, TLC2654Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS020D – NOVEMBER 1988 – REVISED AUGUST 1994

**APPLICATION INFORMATION**

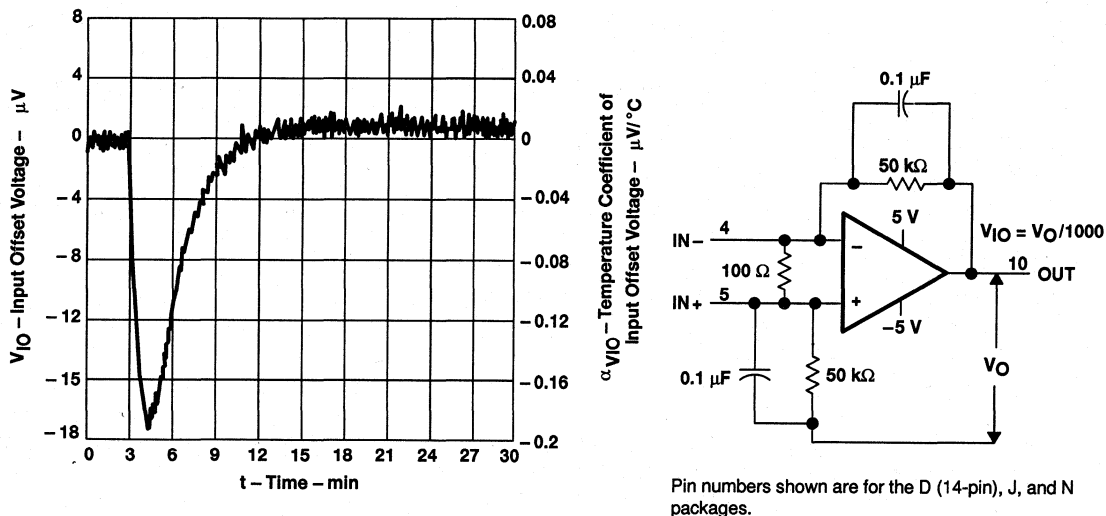
**theory of operation (continued)**

The TLC2654 removes this restriction on ac performance by using a 10-kHz internal clock frequency. This high chopping frequency allows amplification of input signals up to 5 kHz without errors due to intermodulation and greatly reduces low-frequency noise.

**THERMAL INFORMATION**

**temperature coefficient of input offset voltage**

Figure 36 shows the effects of package-included thermal EMF. The TLC2654 can null only the offset voltage within its nulling loop. There are metal-to-metal junctions outside the nulling loop (bonding wires, solder joints, etc.) that produce EMF. In Figure 36, a TLC2654 packaged in a 14-pin plastic package (N package) was placed in an oven at 25°C at t = 0, biased up, and allowed to stabilize. At t = 3 min, the oven was turned on and allowed to rise in temperature to 125°C. As evidenced by the curve, the overall change in input offset voltage with temperature is less than the specified maximum limit of 0.05  $\mu\text{V}/^\circ\text{C}$ .



**Figure 36. Effects of Package-Induced Thermal EMF**

# TLC2801Z, TLC2801Y

## Advanced LinCMOS™ LOW-NOISE PRECISION OPERATIONAL AMPLIFIERS

SLOS116A – JULY 1992 – REVISED AUGUST 1994

- **Low Input Noise Voltage:**  
35 nV/ $\sqrt{\text{Hz}}$  Max at  $f = 10 \text{ Hz}$   
15 nV/ $\sqrt{\text{Hz}}$  Max at  $f = 1 \text{ kHz}$
- **Low Input Offset Voltage:**  
500  $\mu\text{V}$  Max at  $T_A = 25^\circ\text{C}$   
1.5 mV Max at  $T_A = \text{Full Range}$
- **Excellent Offset Voltage Stability With Temperature . . . 4  $\mu\text{V}/^\circ\text{C}$  Typ**
- **Low Input Bias Current:**  
1 pA Typ at  $T_A = 25^\circ\text{C}$   
250 pA Typ at  $T_A = 150^\circ\text{C}$
- **Specified for Both Single-Supply and Split-Supply Operation**
- **Common-Mode Input Voltage Range Includes the Negative Rail**

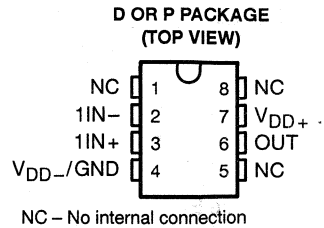
### description

The TLC2801 is a precision, low-noise operational amplifier manufactured using Texas Instruments Advanced LinCMOS™ process. The TLC2801 combines the noise performance of the lowest-noise JFET amplifiers with the dc precision available previously only in bipolar amplifiers. The Advanced LinCMOS™ process uses silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. In addition, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

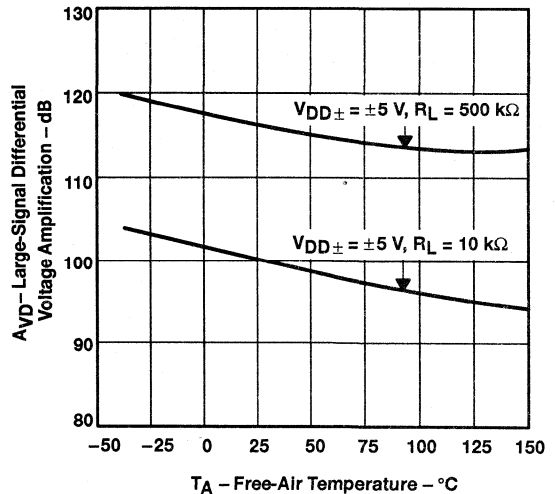
The combination of excellent dc and noise performance with a common-mode input voltage range that includes the negative rail makes the TLC2801 an ideal choice for high-impedance, low-level signal conditioning applications in either single-supply or split-supply configurations.

The device inputs and output are designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in degradation of the device parametric performance.

The TLC2801 is characterized for operation over the temperature range of  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .



### LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION VS FREE-AIR TEMPERATURE



### AVAILABLE OPTIONS

$T_A$	$V_{IOmax}$ AT $150^\circ\text{C}$	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
$-40^\circ\text{C}$ to $150^\circ\text{C}$	1.5 mV	TLC2801ZD	TLC2801ZP	TLC2801Y

The D packages are available taped and reeled. Add R suffix to the device type when ordering (e.g., TLC2801ZDR).

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



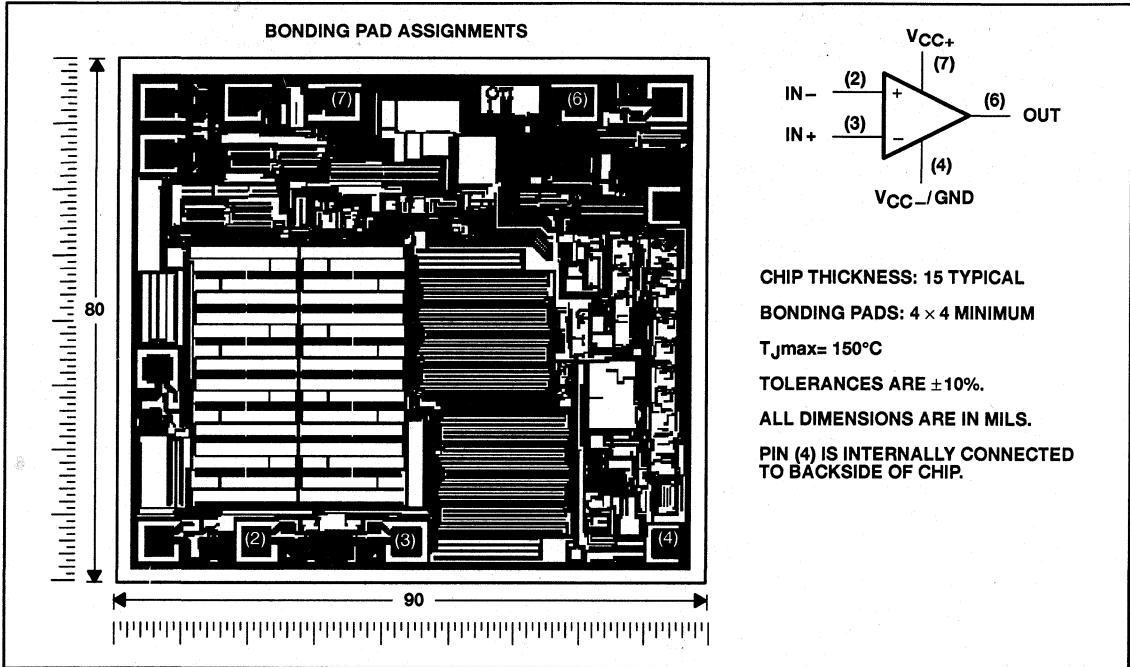
Copyright © 1994, Texas Instruments Incorporated

# TLC2801Z, TLC2801Y Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

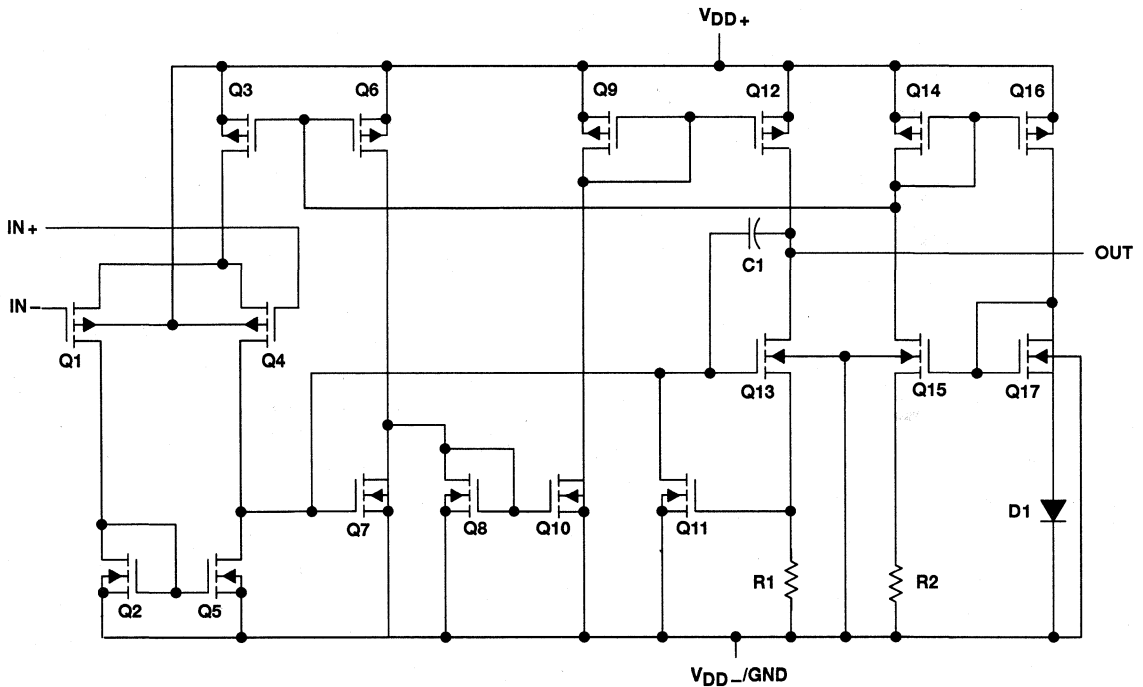
SLOS116A – JULY 1992 – REVISED AUGUST 1994

## TLC2801Y chip information

This chip, properly assembled, displays characteristics similar to the TLC2801. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



**equivalent schematic**



**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ (see Note 1)	8 V
Supply voltage, $V_{DD-}$ (see Note 1)	-8 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm 16$ V
Input voltage range, $V_I$ (any input, see Note 1)	$\pm 8$ V
Input current, $I_I$ (each input)	$\pm 5$ mA
Output current, $I_O$	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Operating free-air temperature range, $T_A$	-40°C to 150°C
Storage temperature range	-65°C to 175°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at the noninverting input with respect to the inverting point.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD\pm}$	$\pm 2.3$	$\pm 8$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 2.3$	V
Operating free-air temperature, $T_A$	-40	150	°C

# TLC2801Z, TLC2801Y

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIERS

SLOS116A – JULY 1992 – REVISED AUGUST 1994

### electrical characteristics at specified free-air temperature, $V_{DD\pm} = \pm 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2801Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C		100	500	$\mu\text{V}$
		Full range			1500	
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		-55°C to 150°C		4		$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)		25°C	0.001	0.005	$\mu\text{V}/\text{mo}$	
$I_{IO}$ Input offset current		25°C		0.5		pA
		Full range			3	nA
$I_{IB}$ Input bias current	25°C		1		pA	
	Full range			30	nA	
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		V	
$V_{OM+}$ Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	V	
$V_{OM-}$ Maximum negative peak output voltage swing		Full range	4.5			
		25°C	-4.7	-4.9	V	
Full range		-4.5				
$A_{VD}$ Large-signal differential voltage amplification	$V_O = \pm 4\text{ V}, R_L = 500\ \text{k}\Omega$	25°C	300	460	V/mV	
		Full range	100			
	$V_O = \pm 4\text{ V}, R_L = 10\ \text{k}\Omega$	25°C	50	100		
		Full range	15			
CMRR Common-mode rejection ratio	$V_O = 0, R_S = 50\ \Omega, V_{IC} = V_{ICRmin}$	25°C	90	115	dB	
		Full range	85			
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD\pm} = \pm 2.3\text{ V to } \pm 8\text{ V}$	25°C	90	110	dB	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 0, \text{ No load}$	25°C		1.1 1.5	mA	
		Full range		1.5		

### operating characteristics at specified free-air temperature, $V_{DD\pm} = \pm 5\text{ V}$

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2801Z			UNIT
			MIN	TYP	MAX	
SR Slew rate unity gain	$V_O = \pm 2.3\text{ V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2	2.7		$\text{V}/\mu\text{s}$
		Full range	1			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18	35	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$			8	15	
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$			0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9	MHz	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°		

† Full range is -40°C to 150°C.

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265



**TLC2801Z, TLC2801Y**  
**Advanced LinCMOS™ LOW-NOISE PRECISION**  
**OPERATIONAL AMPLIFIERS**  
 SLOS116A – JULY 1992 – REVISED AUGUST 1994

**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2801Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0, \quad R_S = 50\ \Omega$	25°C		100	500	$\mu\text{V}$
		Full range		1500		
$\alpha_{VIO}$ Temperature coefficient of input offset voltage			25°C		4	$\mu\text{V}/^\circ\text{C}$
Input offset voltage long-term drift (see Note 4)			25°C	0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current			25°C		0.5	$\text{pA}$
			Full range		3	
$I_{IB}$ Input bias current		25°C		1	$\text{pA}$	
		Full range		30		
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 2.7		$\text{V}$	
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	25°C	4.7	4.8	$\text{V}$	
		Full range	4.4			
$V_{OL}$ Maximum low-level output voltage		25°C		0	50	$\text{mV}$
		Full range		50		
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V},$ $R_L = 500\ \text{k}\Omega$	25°C	150	315	$\text{V/mV}$	
		Full range	50			
	$V_O = 1\ \text{V to } 4\ \text{V},$ $R_L = 10\ \text{k}\Omega$	25°C	25	55		
		Full range	5			
$\text{CMRR}$ Common-mode rejection ratio	$V_O = 0, \quad V_{IC} = V_{ICR\text{min}},$ $R_S = 50\ \Omega$	25°C	90	110	$\text{dB}$	
		Full range	85			
$\text{kSVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{V to } 16\ \text{V}$	25°C	90	110	$\text{dB}$	
		Full range	85			
$I_{DD}$ Supply current	$V_O = 0, \quad \text{No load}$	25°C		1.1	1.5	$\text{mA}$
		Full range		1.5		

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2801Z			UNIT
			MIN	TYP	MAX	
$\text{SR}$ Slew rate unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V},$ $R_L = 10\ \text{k}\Omega, \quad C_L = 100\ \text{pF}$	25°C	1.8	2.5	$\text{V}/\mu\text{s}$	
		Full range	0.8			
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$	25°C		18	35	$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$	25°C		8	15	
$V_{N(\text{PP})}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$	25°C		0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$	25°C		0.7		
$I_n$ Equivalent input noise current		25°C		0.6	$\text{fA}/\sqrt{\text{Hz}}$	
Gain-bandwidth product	$f = 10\ \text{kHz}, \quad R_L = 10\ \text{k}\Omega,$ $C_L = 100\ \text{pF}$	25°C		1.8	$\text{MHz}$	
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, \quad C_L = 100\ \text{pF}$	25°C		45°		

† Full range is  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2801Z, TLC2801Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

SLOS116A – JULY 1992 – REVISED AUGUST 1994

**electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2801Z			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0$ , $R_S = 50\ \Omega$		100	500	$\mu\text{V}$
Input offset voltage long-term drift (see Note 4)			0.001	0.005	$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current				0.5	$\text{pA}$
$I_{IB}$ Input bias current				1	$\text{pA}$
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$	0 to 2.7			V
$V_{OH}$ Maximum high-level output voltage	$R_L = 10\ \text{k}\Omega$	4.7	4.8		V
$V_{OL}$ Maximum low-level output voltage	$I_O = 0$		0	50	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 500\ \text{k}\Omega$	150	315		V/mV
	$V_O = 1\ \text{V to } 4\ \text{V}$ , $R_L = 10\ \text{k}\Omega$	25	55		
CMRR Common-mode rejection ratio	$V_O = 0$ , $R_S = 50\ \Omega$ , $V_{IC} = V_{ICR\text{min}}$	90	110		dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.6\ \text{V to } 16\ \text{V}$	90	110		dB
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V}$ , No load		1	1.5	mA

**operating characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2801Z			UNIT
		MIN	TYP	MAX	
SR Positive slew rate at unity gain	$V_O = 0.5\ \text{V to } 2.5\ \text{V}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$	1.8	2.5		V/ $\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\ \text{Hz}$		18		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\ \text{kHz}$		8		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\ \text{to } 1\ \text{Hz}$		0.5		$\mu\text{V}$
	$f = 0.1\ \text{to } 10\ \text{Hz}$		0.7		
$I_n$ Equivalent input noise current			0.6		$\text{pA}/\sqrt{\text{Hz}}$
Gain-bandwidth product	$f = 10\ \text{kHz}$ , $R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		1.8		MHz
$\phi_m$ Phase margin at unity gain	$R_L = 10\ \text{k}\Omega$ , $C_L = 100\ \text{pF}$		45°		

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.



PARAMETER MEASUREMENT INFORMATION

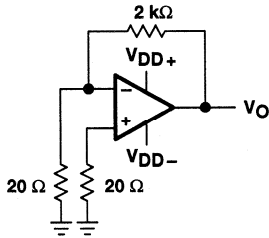
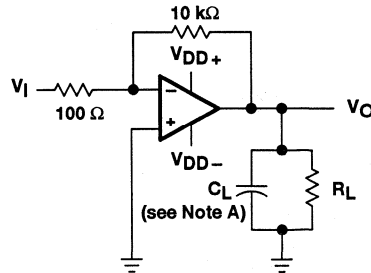
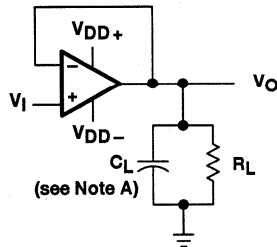


Figure 1. Noise-Voltage Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 2. Phase-Margin Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 3. Slew-Rate Test Circuit

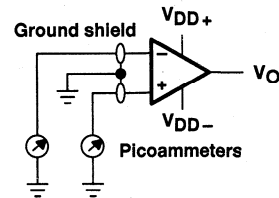


Figure 4. Input-Bias and Offset-Current Test Circuit

typical values

Typical values as presented in this data sheet represents the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp bias-current level typical of the TLC2801, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test socket leakages can easily exceed the actual device bias currents. To measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltage applied but with no device in the socket. The device is then inserted in the socket and a second test measuring both the socket leakage and the device input bias current is performed. The two measurements are then subtracted algebraically to determine the bias current of the device.

**TLC2801Z, TLC2801Y**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIERS**

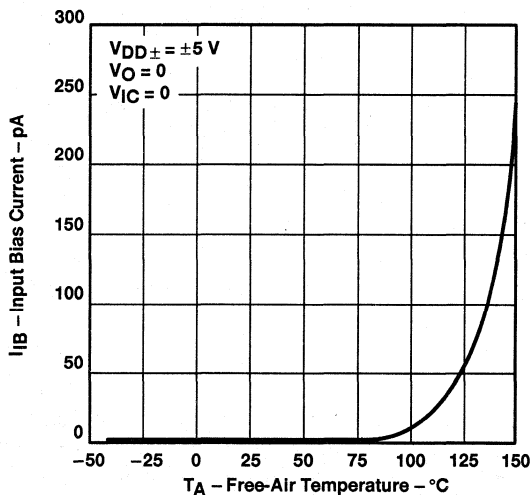
SLOS116A – JULY 1992 – REVISED AUGUST 1994

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

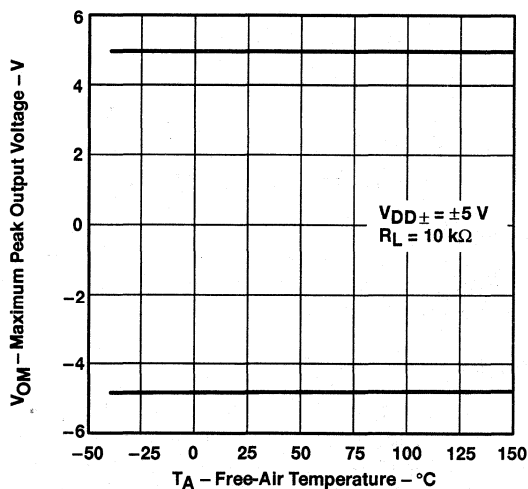
			FIGURE
$I_{IB}$	Input bias current	vs Free-air temperature	5
$V_{OM}$	Maximum peak output voltage	vs Free-air temperature	6
$V_{OH}$	High-level output voltage	vs Free-air temperature	7
$V_{OL}$	Low-level output voltage	vs Free-air temperature	8
$A_{VD}$	Differential voltage amplification	vs Free-air temperature	9
$I_{OS}$	Short-circuit output current	vs Free-air temperature	10
$I_{DD}$	Supply current	vs Free-air temperature	11
SR	Slew rate	vs Free-air temperature	12
	Gain-bandwidth product	vs Free-air temperature	13

**INPUT BIAS CURRENT  
vs  
FREE-AIR TEMPERATURE**



**Figure 5**

**MAXIMUM PEAK OUTPUT VOLTAGE  
vs  
FREE-AIR TEMPERATURE**



**Figure 6**

TYPICAL CHARACTERISTICS

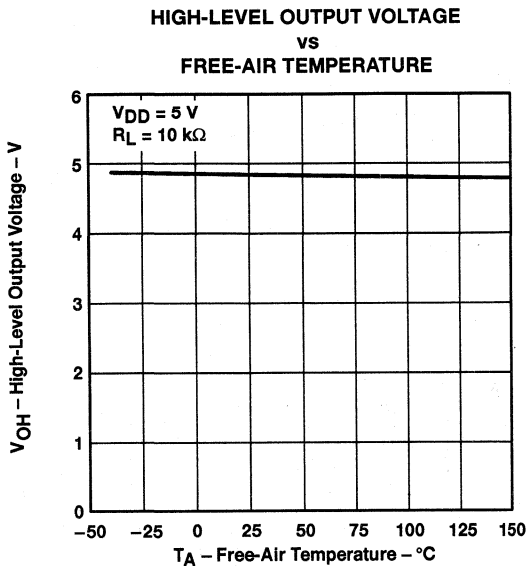


Figure 7

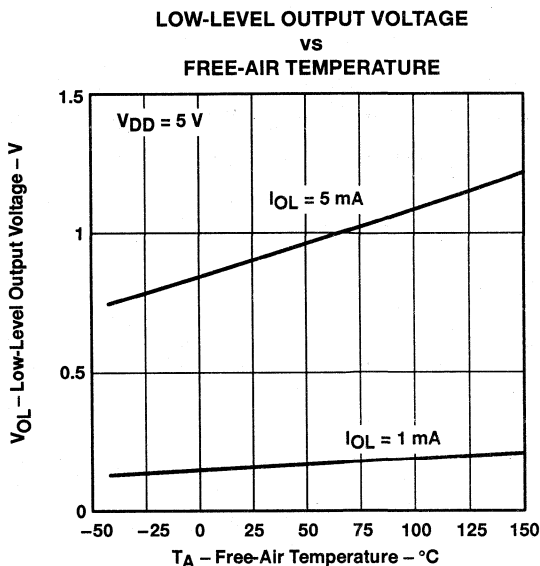


Figure 8

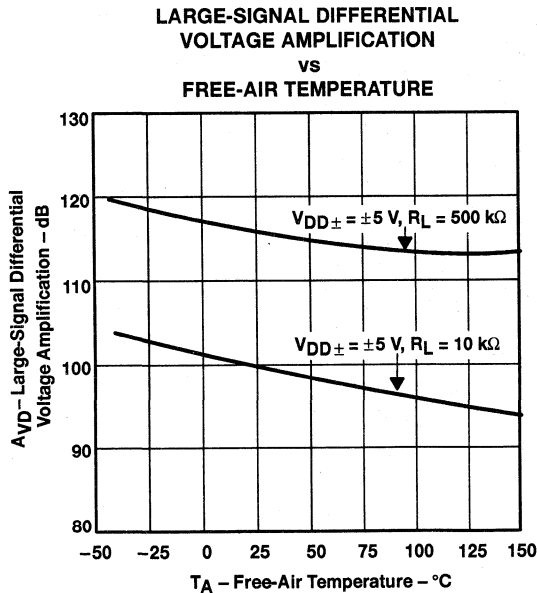


Figure 9

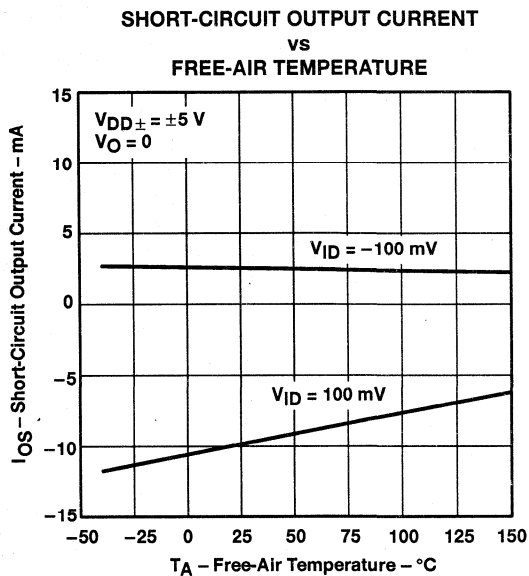
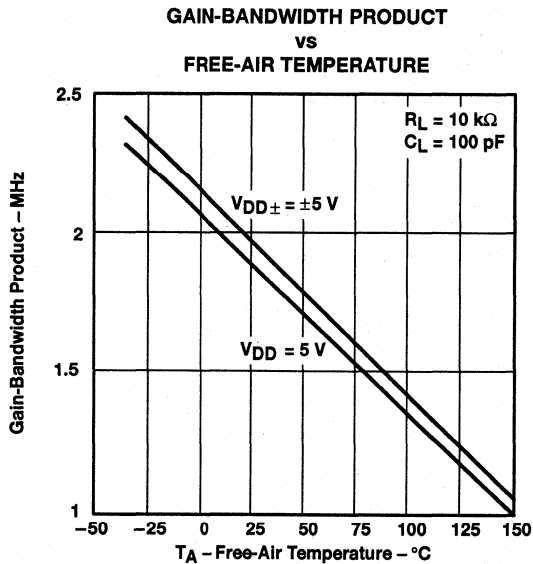
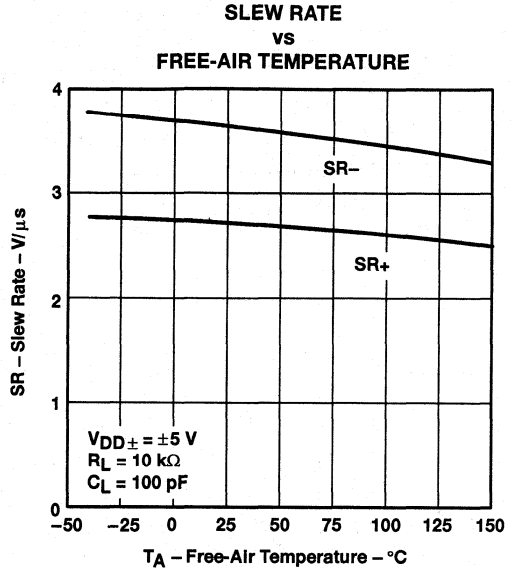
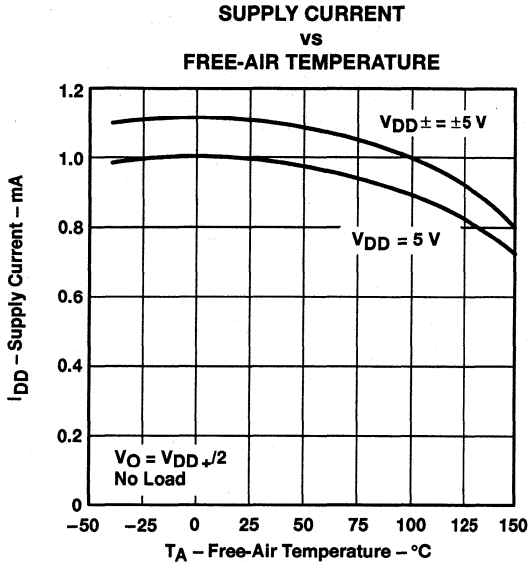


Figure 10

TYPICAL CHARACTERISTICS



# TLC2810Z, TLC2810Y LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

SLOS120A – AUGUST 1993 – REVISED AUGUST 1994

- **Trimmed Input Offset Voltage:**  
10 mV Max at 25°C,  $V_{DD} = 5\text{ V}$
- **Input Offset Voltage Drift Typically**  
0.1  $\mu\text{V}/\text{Month}$ , Including the First 30 Days
- **Wide Range of Supply Voltages Over Specified Temperature Range:**  
–40°C to 150°C . . . 4 V to 16 V
- **Single-Supply Operation**
- **Common-Mode Input Voltage Range Extends to the Negative Rail**
- **Low Noise . . . 25 nV/ $\sqrt{\text{Hz}}$  Typ at  $f = 1\text{ kHz}$**
- **Output Voltage Range Includes Negative Rail**
- **High Input Impedance . . .  $10^{12}\ \Omega$  Typ**
- **ESD-Protection Circuitry**
- **Small-Outline Package Option Also Available in Tape and Reel**
- **Designed-In Latch-Up Immunity**

## description

The TLC2810Z dual operational amplifiers combine low offset voltage drift with high input impedance, low noise, and speeds approaching that of general-purpose JFET devices. In addition, the use of Texas Instruments silicon-gate LinCMOS technology assures offset stability that greatly exceeds the stability available with conventional metal-gate processes.

The high input impedance, low bias current, and high slew rate make the TLC2810Z ideal for applications that have previously been reserved for JFET and NFET products. These advantages, in combination with an upper operating temperature of 150°C, make the TLC2810Z an ideal choice for precision, extremely high-temperature applications.

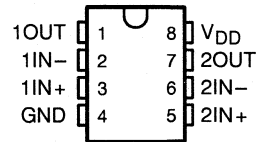
In general, many features associated with bipolar technology are available on the TLC2810Z without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are designed easily with the TLC2810Z.

The TLC2810Z package options include a small-outline version for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up at 25°C. The TLC2810Z incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD 883C, Method 3015.2. However, care should be exercised in handling the TLC2810Z as exposure to ESD may result in the degradation of the device parametric performance. Additional care should be exercised to prevent  $V_{DD}$  supply line transients under power conditions. Transients of greater than 20 V can trigger the ESD-protection structure, inducing a low-impedance path to GND. Should this condition occur, the sustained current supplied to the device must be limited to 100 mA or less. Failure to do so can result in a latched condition and device failure.

The TLC2810Z is characterized for operation over the extended temperature range from –40°C to 150°C.

D OR P PACKAGE  
(TOP VIEW)



### AVAILABLE OPTIONS

T <sub>A</sub>	PACKAGED DEVICES		CHIP FORM (Y)
	SMALL OUTLINE (D) <sup>†</sup>	PLASTIC DIP (P)	
–40°C to 150°C	TLC2810ZD	TLC2810ZP	TLC2810Y

<sup>†</sup> The D packages are available taped and reeled. Add R suffix to the device type when ordering (e.g., TLC2810ZDR).

LinCMOS is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

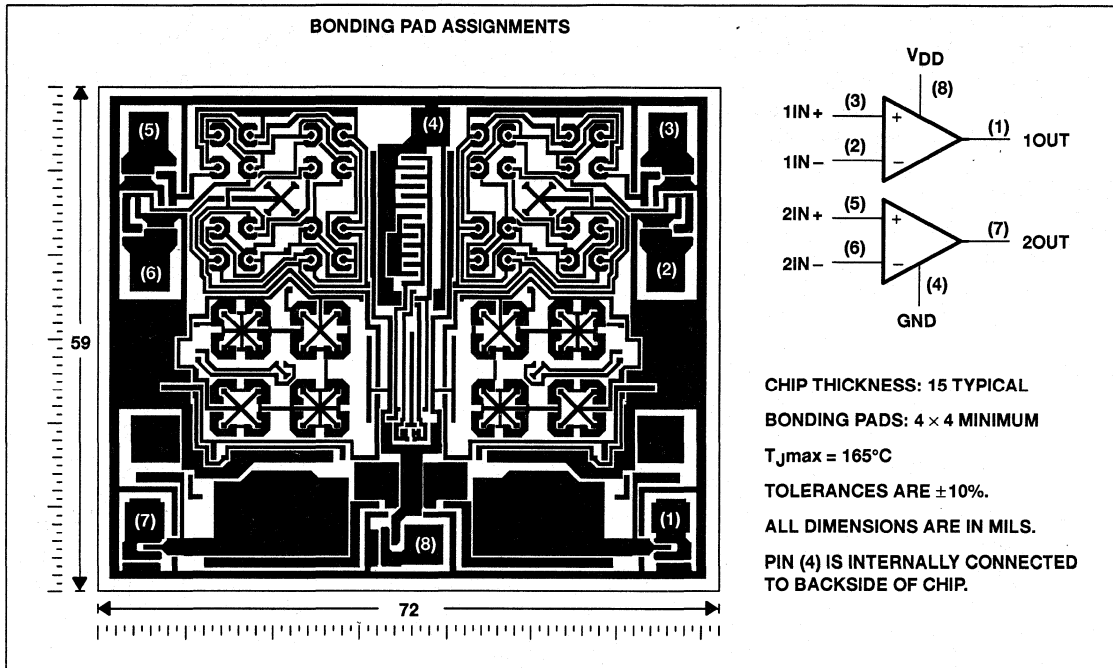
Copyright © 1994, Texas Instruments Incorporated

**TLC2810Z, TLC2810Y**  
**LinCMOS™ PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS120A—AUGUST 1993—REVISED AUGUST 1994

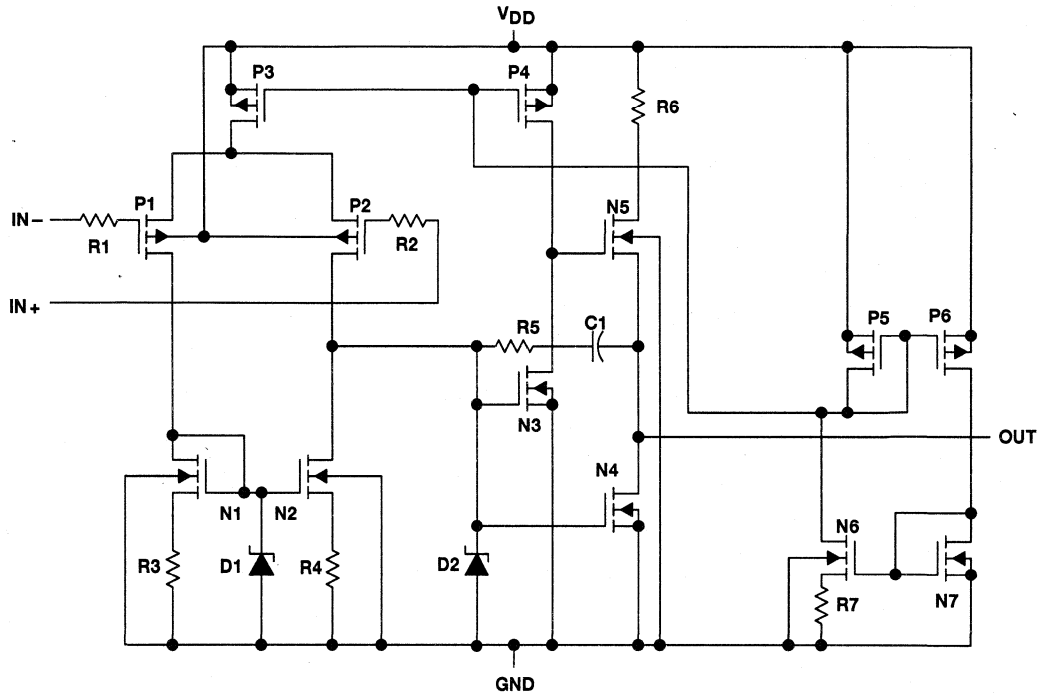
**TLC2810Y chip information**

This chip, when properly assembled, displays characteristics similar to the TLC2810Z. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.





equivalent schematic (each amplifier)



COMPONENT COUNT†	
Transistors	26
Diodes	4
Resistors	14
Capacitors	2

† Includes both amplifiers

**TLC2810Z, TLC2810Y**  
**LinCMOS™ PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**

SLOS120A – AUGUST 1993 – REVISED AUGUST 1994

**absolute maximum ratings over operating free-air temperature (unless otherwise noted)†**

Supply voltage, $V_{DD}$ (see Note 1)	16 V
Differential input voltage, $V_{ID}$ (see Note 2)	$\pm V_{DD}$
Input voltage range, $V_I$ (any input)	-0.3 V to $V_{DD}$
Input current, $I_I$	$\pm 2$ mA
Output current, $I_O$ (each output)	$\pm 30$ mA
Total current into $V_{DD}$	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) $T_A = 25^\circ\text{C}$ (see Note 3)	unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, $T_A$	$-40^\circ\text{C}$ to $150^\circ\text{C}$
Storage temperature range	$-65^\circ\text{C}$ to $165^\circ\text{C}$
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	$260^\circ\text{C}$

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.  
 2. Differential voltages are at IN+ with respect to IN-.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application selection).

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 105^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING	$T_A = 150^\circ\text{C}$ POWER RATING
D	812 mW	5.8 mW/°C	551 mW	348 mW	232 mW	87 mW
P	1120 mW	8.0 mW/°C	760 mW	480 mW	320 mW	120 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD}$	4	16	V
Common-mode input voltage, $V_{IC}$	$V_{DD} = 5$ V,	$T_A = 25^\circ\text{C}$	-0.2 3.5 V
Input voltage, $V_I$	$V_{DD} = 5$ V		-0.2 3.5 V
Operating free-air temperature, $T_A$	-40	150	°C



**electrical characteristics,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2810Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_O = 1\text{ V}, V_{IC} = 1\text{ V},$ $R_S = 50\ \Omega, R_L = 10\text{ k}\Omega$	25°C		1.8	10	mV
		Full range			12	
$\alpha_{VIO}$ Average temperature coefficient of input offset voltage		25°C to 150°C		3.5		$\mu\text{V}/^\circ\text{C}$
$I_{IO}$ Input offset current (see Note 4)	$V_{IC} = 1\text{ V}, V_O = 1\text{ V},$	25°C		2.4	100	pA
		150°C		5.2	30	nA
$I_{IB}$ Input bias current (see Note 4)	$V_{IC} = 1\text{ V}, V_O = 1\text{ V}$	25°C		7	100	pA
		150°C		50	150	nA
$V_{ICR}$ Common-mode input voltage range (see Note 5)	$R_S = 50\ \Omega$	25°C	-0.2 to 4	-0.3 to 4.2		V
		Full range	-0.2 to 3.8			V
$V_{OH}$ High-level output voltage	$V_{IC} = 1\text{ V},$ $V_{ID} = 100\text{ mV},$ $I_{OH} = -1\text{ mA}$	25°C	3.2	3.8		V
		Full range	3			
$V_{OL}$ Low-level output voltage	$V_{IC} = 1\text{ V},$ $V_{ID} = -100\text{ mV},$ $I_{OL} = 1\text{ mA}$	25°C		80	150	mV
		Full range			190	
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 1\text{ V},$ $V_O = 0.25\text{ V to }2\text{ V},$ $R_L = 10\text{ k}\Omega$	25°C	5	25		V/mV
		Full range	4			
CMRR Common-mode rejection ratio	$V_O = 1\text{ V},$ $V_{IC} = V_{ICRmin},$ $R_S = 50\ \Omega$	25°C	65	90		dB
		Full range	60			
kSVR Supply-voltage rejection ratio ( $\Delta V_{DD} / \Delta V_{IO}$ )	$V_{DD} = 4\text{ V to }16\text{ V},$ $V_O = 1\text{ V}, V_{IC} = 1\text{ V},$ $R_S = 50\ \Omega$	25°C	65	75		dB
		Full range	60			
$I_{DD}$ Supply current	$V_O = 1\text{ V}, V_{IC} = 1\text{ V},$ No load	25°C		1	3.2	mA
		Full range			4.4	

† Full range is -40°C to 150°C.

- NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
 5. This range also applies to each input individually.

**TLC2810Z, TLC2810Y**  
**LinCMOS™ PRECISION**  
**DUAL OPERATIONAL AMPLIFIERS**  
 SLOS120A – AUGUST 1993 – REVISED AUGUST 1994

**operating characteristics,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS		T <sub>A</sub>	TLC2810Z			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	R <sub>L</sub> = 10 kΩ, C <sub>L</sub> = 20 pF, See Figure 26	V <sub>I(PP)</sub> = 1 V	25°C		3.6	V/μs	
			150°C		2.8		
		V <sub>I(PP)</sub> = 2.5 V	25°C		2.2		
			150°C		2.1		
V <sub>n</sub> Equivalent input noise voltage	f = 1 kHz, See Figure 27	R <sub>S</sub> = 20 Ω	25°C		25	nV/√Hz	
B <sub>OM</sub> Maximum output-swing bandwidth	V <sub>O</sub> = V <sub>OH</sub> , R <sub>L</sub> = 10 kΩ	C <sub>L</sub> = 20 pF, See Figure 26	25°C		320	kHz	
			150°C		200		
B <sub>1</sub> Unity-gain bandwidth	V <sub>I</sub> = 10 mV, See Figure 28	C <sub>L</sub> = 20 pF	25°C		1.7	MHz	
			150°C		0.8		
φ <sub>m</sub> Phase margin	V <sub>I</sub> = 10 mV, C <sub>L</sub> = 20 pF	f = B <sub>1</sub> , See Figure 28	25°C		46°		
			150°C		40°		



**electrical characteristics at  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2810Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 1\text{ V}$ , $R_S = 50\ \Omega$			10	mV
$I_{IO}$ Input offset current (see Note 4)					
$I_{IB}$ Input bias current (see Note 4)					
$V_{ICR}$ Common-mode input voltage range (see Note 5)	$R_S = 50\ \Omega$			-0.2 to 4	V
$V_{OH}$ High-level output voltage	$V_{IC} = 1\text{ V}$ , $I_{OH} = -1\text{ mA}$	$V_{ID} = 100\text{ mV}$ ,		3.2	V
$V_{OL}$ Low-level output voltage	$V_{IC} = 1\text{ V}$ , $I_{OL} = 1\text{ mA}$	$V_{ID} = -100\text{ mV}$ ,		150	mV
$A_{VD}$ Large-signal differential voltage amplification	$V_O = 0.25\text{ V to } 2\text{ V}$ , $V_{IC} = 1\text{ V}$	$R_L = 10\text{ k}\Omega$ .		5	V/mV
CMRR Common-mode rejection ratio	$V_O = 1\text{ V}$ , $R_S = 50\ \Omega$	$V_{IC} = V_{ICRmin}$ ,		65	dB
$k_{SVR}$ Supply-voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4\text{ V to } 16\text{ V}$ , $V_O = 1\text{ V}$ ,	$V_{IC} = 1\text{ V}$ , $R_S = 50\ \Omega$		65	dB
$I_{DD}$ Supply current	$V_O = 1\text{ V}$ , No load	$V_{IC} = 1\text{ V}$ ,		3.2	mA

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.  
 5. This range also applies to each input individually.

**operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2810Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$ , $C_L = 20\text{ pF}$ , See Figure 26	$V_I(PP) = 1\text{ V}$	3.6		V/ $\mu\text{s}$
		$V_I(PP) = 2.5\text{ V}$	2.9		
$V_n$ Equivalent input noise voltage	$f = 1\text{ kHz}$ , See Figure 27	$R_S = 20\ \Omega$ ,		25	nV/ $\sqrt{\text{Hz}}$
$B_{OM}$ Maximum output-swing bandwidth	$V_O = V_{OH}$ , $R_L = 10\text{ k}\Omega$ ,	$C_L = 20\text{ pF}$ , See Figure 26		320	kHz
$B_1$ Unity-gain bandwidth	$V_I = 10\text{ mV}$ , See Figure 28	$C_L = 20\text{ pF}$ ,		1.7	MHz
$\phi_m$ Phase margin	$V_I = 10\text{ mV}$ , $C_L = 20\text{ pF}$ ,	$f = B_1$ , See Figure 28		46°	

**TYPICAL CHARACTERISTICS**

**Table of Graphs**

			<b>FIGURE</b>
$V_{IO}$	Input offset voltage	Distribution	1
$\alpha V_{IO}$	Input offset voltage temperature coefficient	Distribution	2
$V_{OH}$	High-level output voltage	vs Output current	3
		vs Supply voltage	4
		vs Free-air temperature	5
$V_{OL}$	Low-level output voltage	vs Common-mode input voltage	6
		vs Differential input voltage	7
		vs Free-air temperature	8
		vs Low-level output current	9
$A_{VD}$	Large-signal differential voltage amplification	vs Supply voltage	10
		vs Free-air temperature	11
		vs Frequency	21
$I_{IB}/I_{IO}$	Input bias and offset current	vs Free-air temperature	12
$V_{IC}$	Common-mode input voltage	vs Supply voltage	13
$I_{DD}$	Supply current	vs Supply voltage	14
		vs Free-air temperature	15
SR	Slew rate	vs Supply voltage	16
		vs Free-air temperature	17
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	18
$B_1$	Gain-bandwidth product	vs Free-air temperature	19
		vs Supply voltage	20
$\phi_m$	Phase margin	vs Supply voltage	22
		vs Free-air temperature	23
		vs Load capacitance	24
$V_n$	Equivalent input noise voltage	vs Frequency	25
		Phase shift	vs Frequency

TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC2810Z  
 INPUT OFFSET VOLTAGE

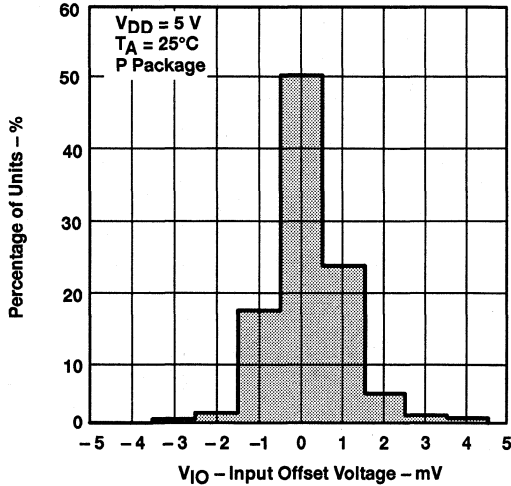


Figure 1

DISTRIBUTION OF TLC2810Z  
 INPUT OFFSET VOLTAGE  
 TEMPERATURE COEFFICIENT

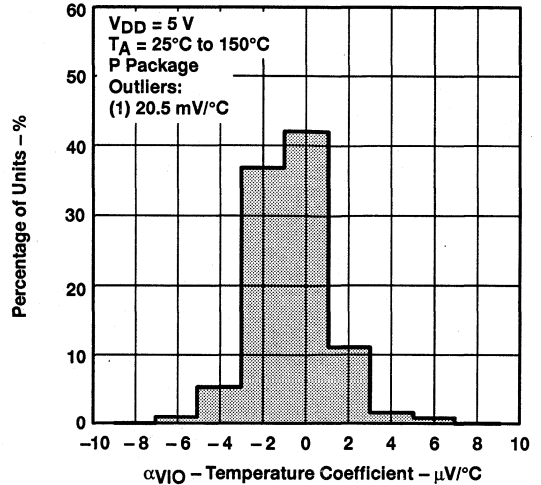


Figure 2

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 HIGH-LEVEL OUTPUT CURRENT

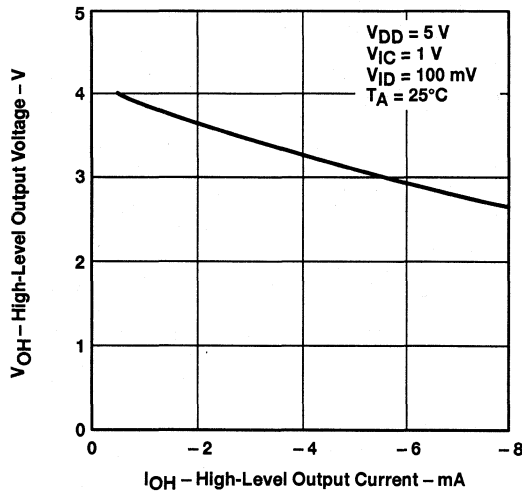


Figure 3

HIGH-LEVEL OUTPUT VOLTAGE  
 vs  
 SUPPLY VOLTAGE

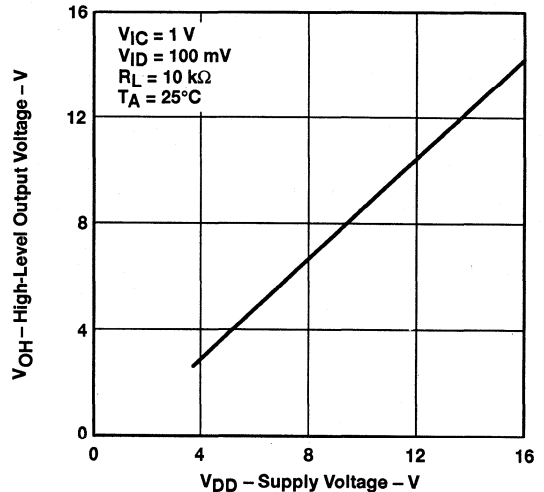
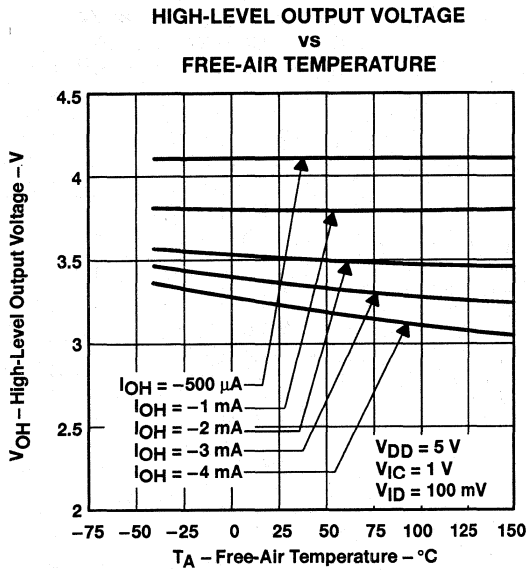
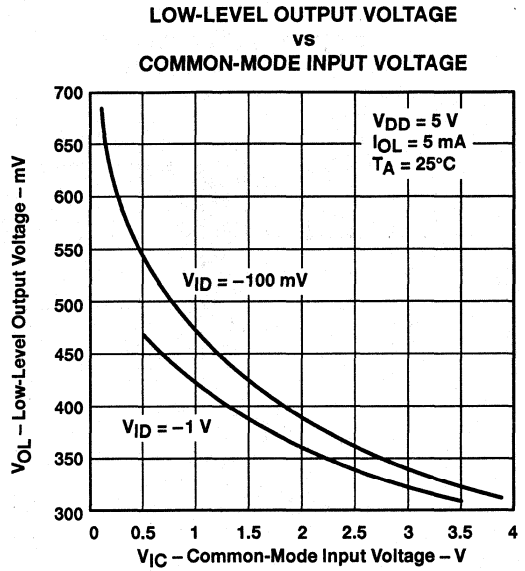


Figure 4

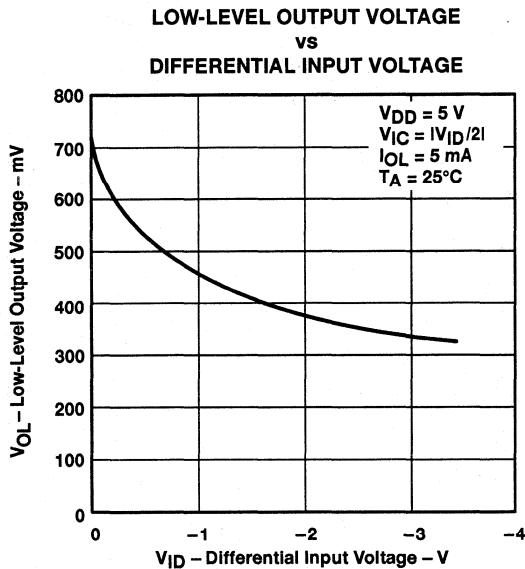
**TYPICAL CHARACTERISTICS**



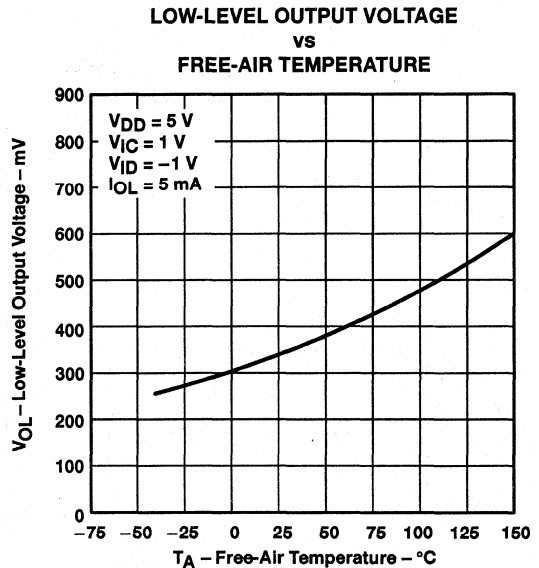
**Figure 5**



**Figure 6**



**Figure 7**



**Figure 8**



TYPICAL CHARACTERISTICS

LOW-LEVEL OUTPUT VOLTAGE  
 vs  
 LOW-LEVEL OUTPUT CURRENT

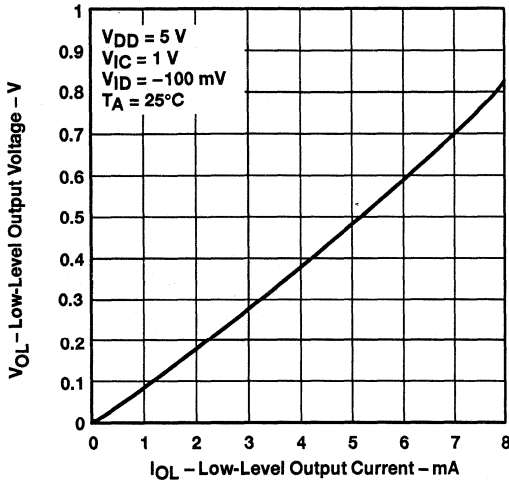


Figure 9

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 SUPPLY VOLTAGE

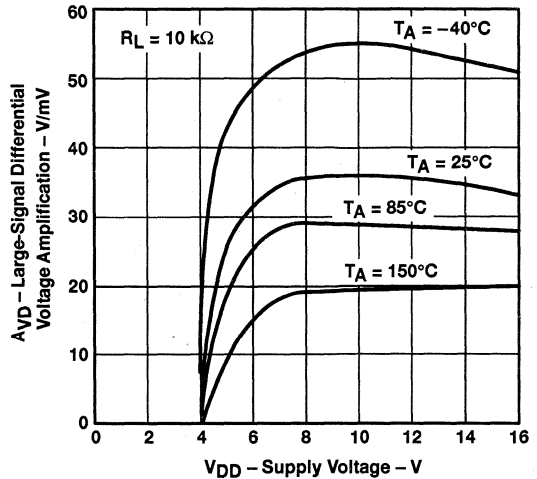


Figure 10

LARGE-SIGNAL  
 DIFFERENTIAL VOLTAGE AMPLIFICATION  
 vs  
 FREE-AIR TEMPERATURE

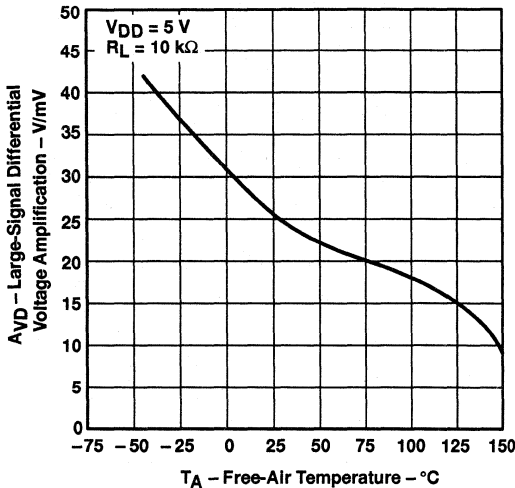


Figure 11

INPUT BIAS CURRENT AND INPUT  
 OFFSET CURRENT  
 vs  
 FREE-AIR TEMPERATURE

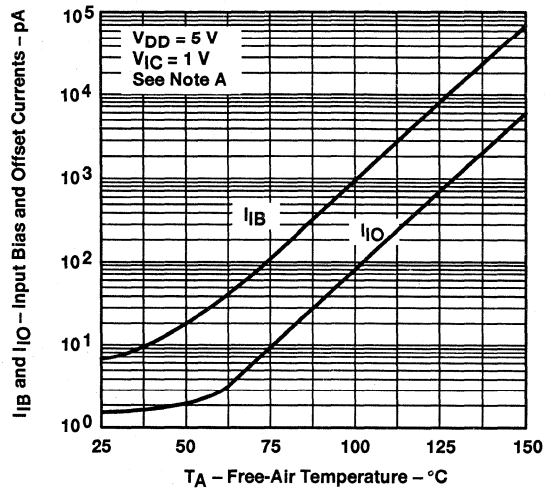


Figure 12

NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

TYPICAL CHARACTERISTICS

COMMON-MODE INPUT VOLTAGE  
 POSITIVE LIMIT  
 vs  
 SUPPLY VOLTAGE

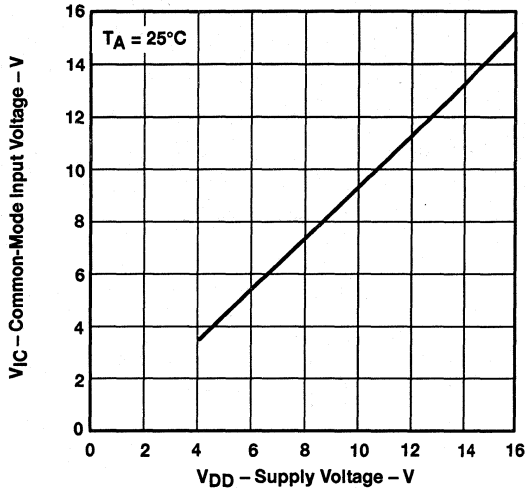


Figure 13

SUPPLY CURRENT  
 vs  
 SUPPLY VOLTAGE

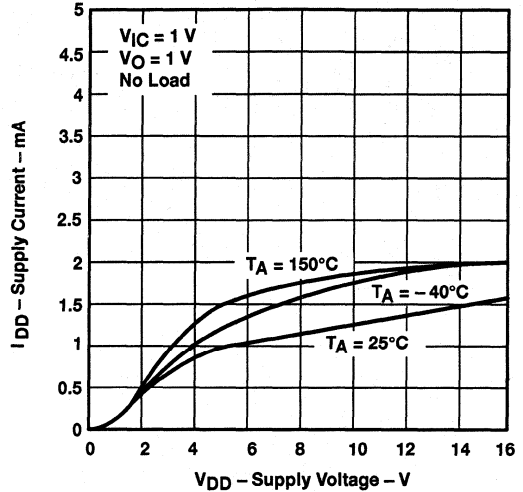


Figure 14

SUPPLY CURRENT  
 vs  
 FREE-AIR TEMPERATURE

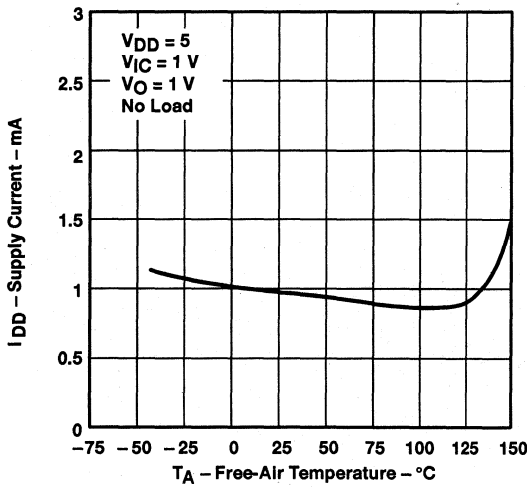


Figure 15

SLEW RATE  
 vs  
 SUPPLY VOLTAGE

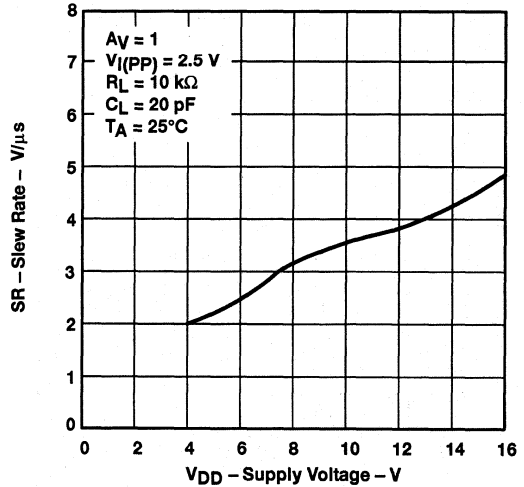
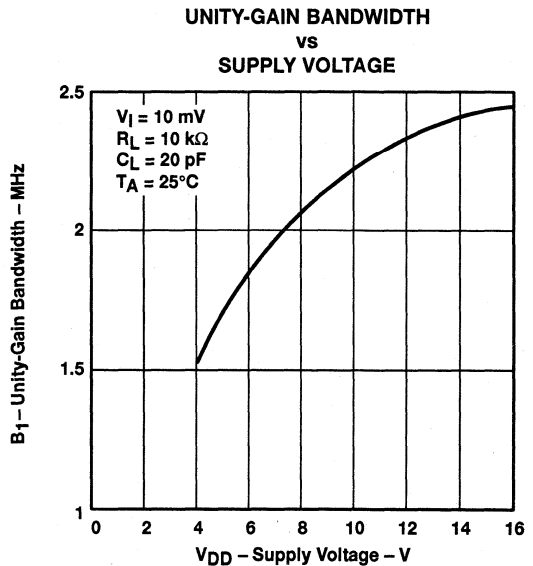
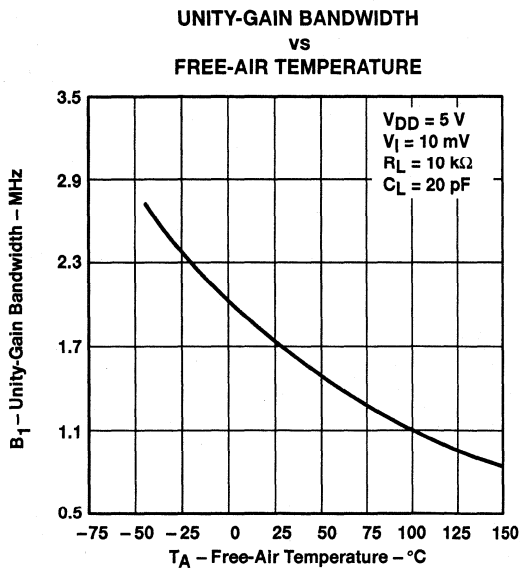
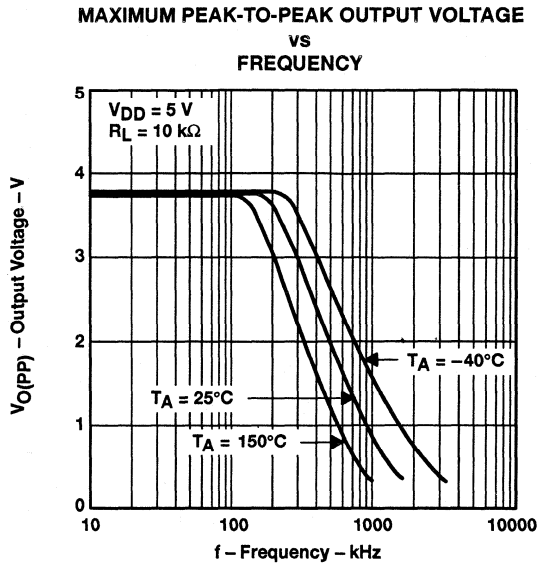
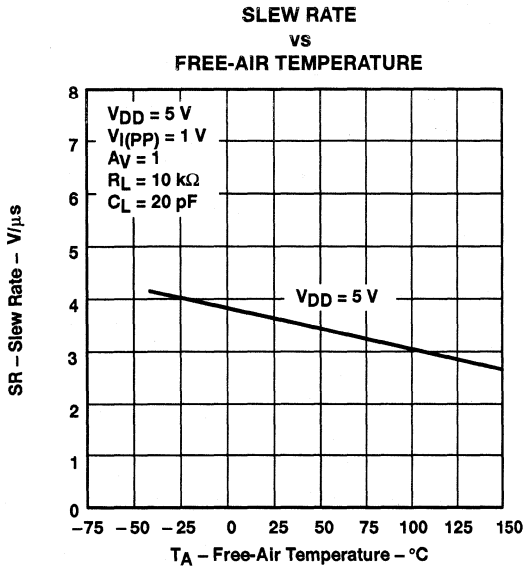


Figure 16

TYPICAL CHARACTERISTICS



TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE  
 AMPLIFICATION AND PHASE SHIFT  
 VS  
 FREQUENCY

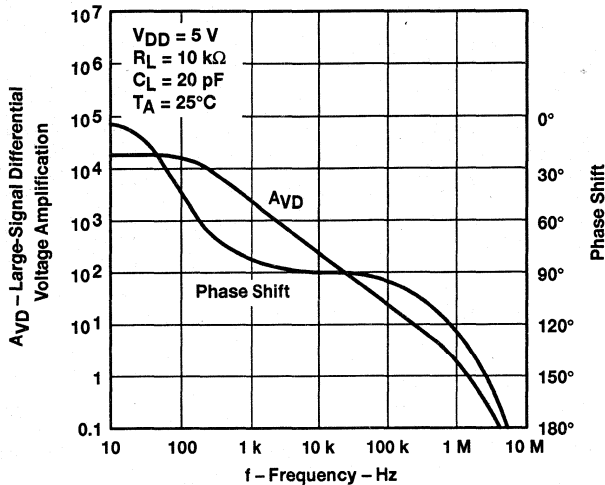


Figure 21

PHASE MARGIN  
 VS  
 SUPPLY VOLTAGE

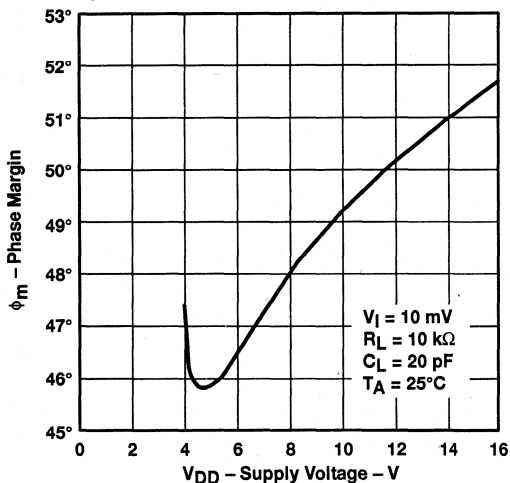


Figure 22

PHASE MARGIN  
 VS  
 FREE-AIR TEMPERATURE

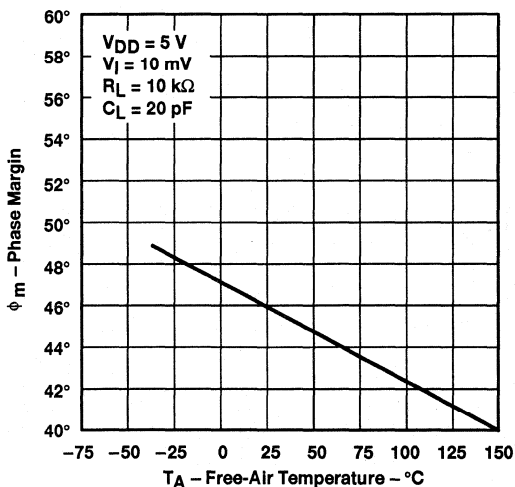


Figure 23

TYPICAL CHARACTERISTICS

PHASE MARGIN  
 vs  
 LOAD CAPACITANCE

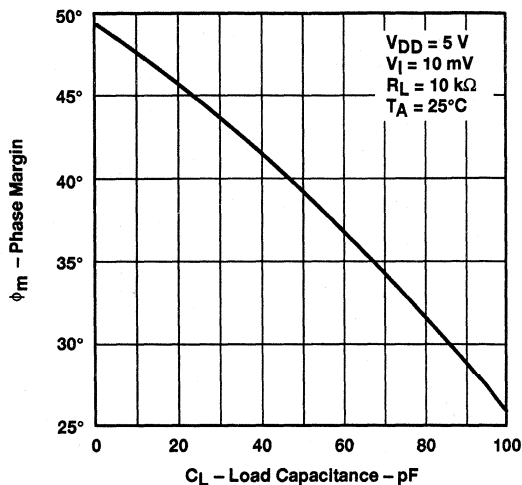


Figure 24

EQUIVALENT INPUT NOISE VOLTAGE  
 vs  
 FREQUENCY

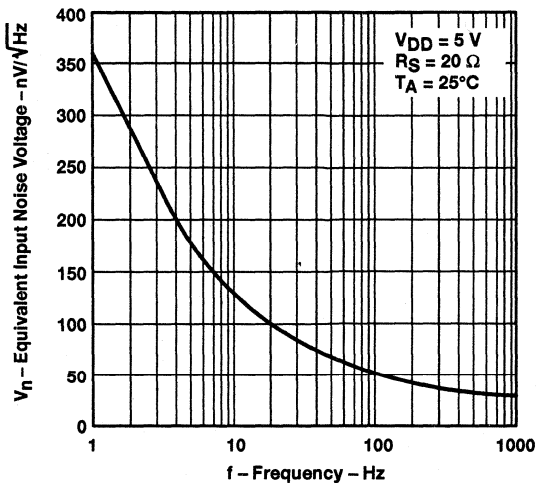
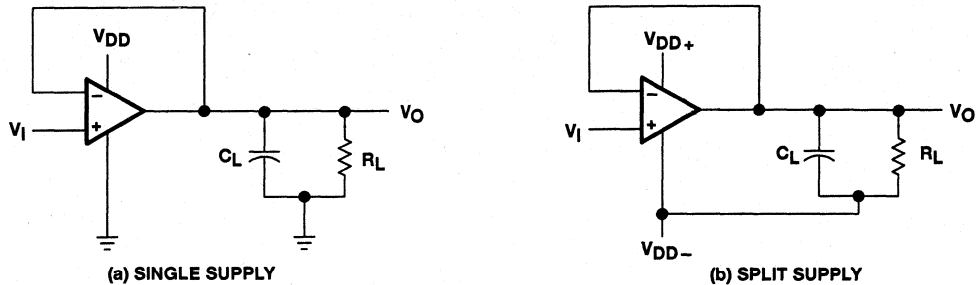


Figure 25

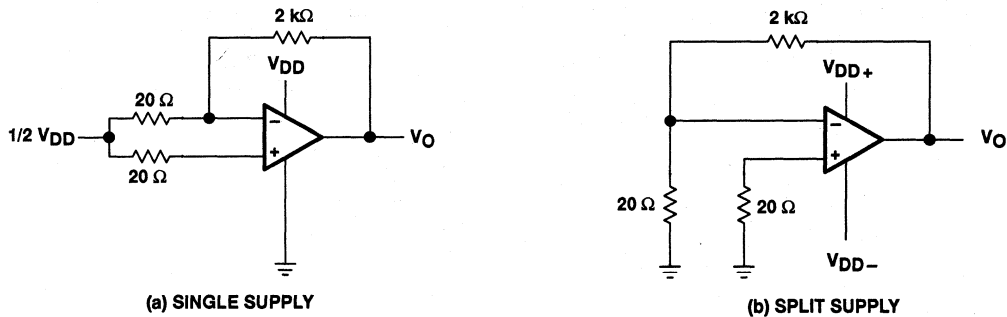
**PARAMETER MEASUREMENT INFORMATION**

**single-supply versus split-supply test circuits**

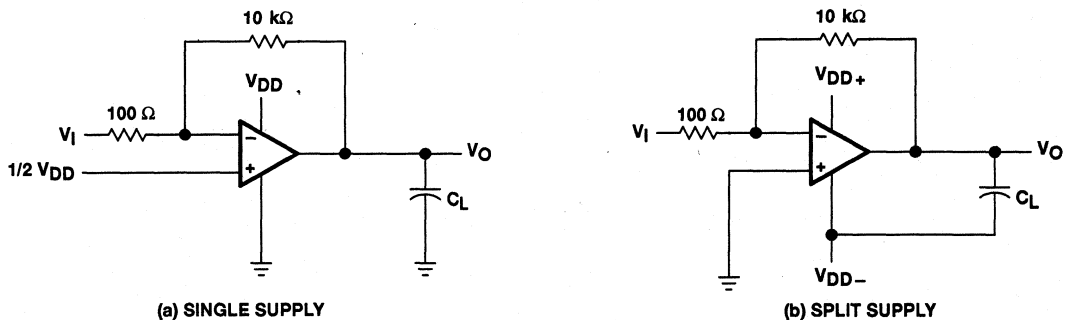
Because the TLC2810Z is optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply and split-supply test circuits is shown below. The use of either circuit gives the same result.



**Figure 26. Unity-Gain Amplifier**



**Figure 27. Noise-Test Circuit**



**Figure 28. Gain-of-100 Inverting Amplifier**

## PARAMETER MEASUREMENT INFORMATION

### input bias current

Because of the high input impedance of the TLC2810Z operational amplifier, attempts to measure the input bias current can result in erroneous readings. The bias current at normal ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 29). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into a test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

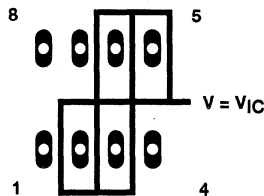


Figure 29. Isolation Metal Around Device Inputs  
(P package)

### low-level output voltage

To obtain low-supply-voltage operation, some compromise is necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to the Typical Characteristics of this data sheet.

### input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance that can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

### full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal

## PARAMETER MEASUREMENT INFORMATION

### full-power response (continued)

input signal until the maximum frequency above which the output contains significant distortion is found. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 26. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 30). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

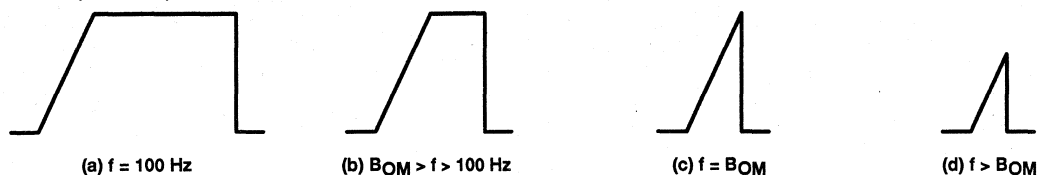


Figure 30. Full-Power-Response Output Signal

### test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices; hence, CMOS devices require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced power supply levels and lower temperatures.

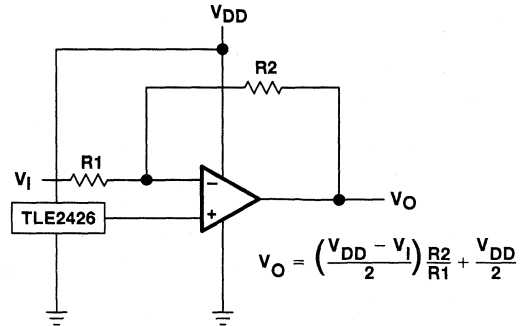


**APPLICATION INFORMATION**

**single-supply operation**

While the TLC2810Z performs well using dual-power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 4 V, thus allowing operation with supply levels commonly available for TTL and CMOS.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. This virtual ground can be generated using two large resistors, but a preferred technique is to use a virtual ground generator such as the TLE2426 (see Figure 31). The TLE2426 supplies an accurate voltage equal to  $V_{DD}/2$ , while consuming very little power and is suitable for supply voltages of greater than 4 V.

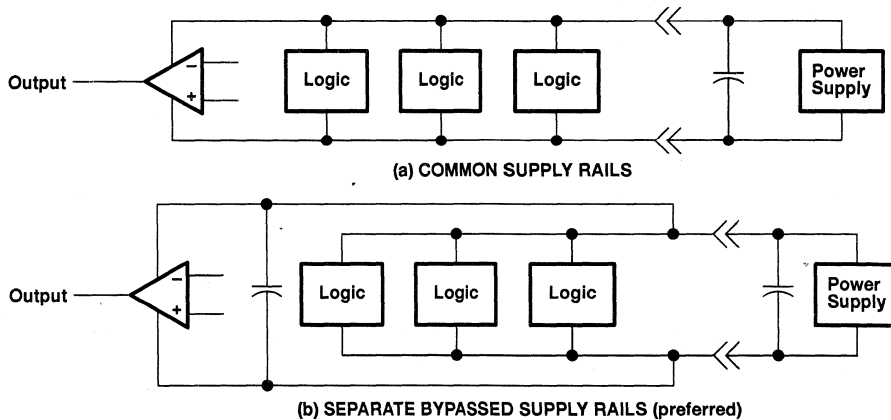


$$V_O = \left( \frac{V_{DD} - V_I}{2} \right) \frac{R_2}{R_1} + \frac{V_{DD}}{2}$$

**Figure 31. Inverting Amplifier With Voltage Reference**

The TLC2810Z works well in conjunction with digital logic. However, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 32). Otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate. However, RC decoupling may be necessary in high-frequency applications.



**Figure 32. Common Versus Separate Supply Rails**

**APPLICATION INFORMATION**

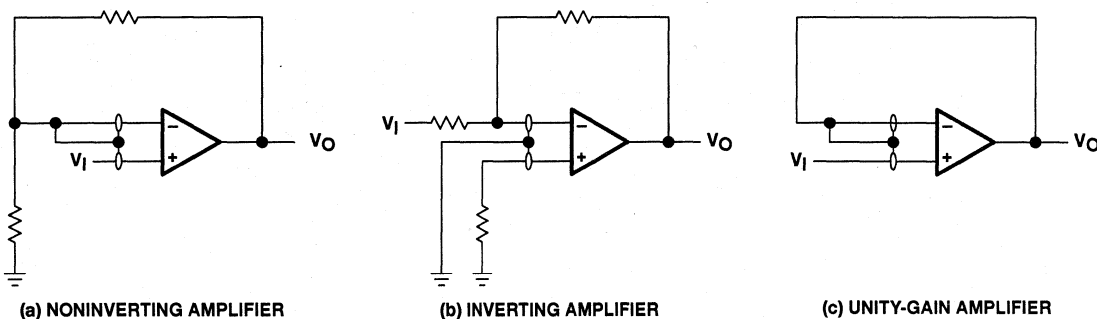
**input characteristics**

The TLC2810Z is specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. The lower range limit includes the negative rail, while the upper range limit is specified at  $V_{DD} - 1$  V at  $T_A = 25^\circ\text{C}$  and at  $V_{DD} - 1.2$  V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design give the TLC2810Z very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically 0.1  $\mu\text{V}/\text{month}$ , including the first month of operation.

Because of the extremely high input impedance and resulting low-bias current requirements, the TLC2810Z is well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias-current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 29 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 33).

Unused amplifiers should be connected as grounded voltage followers to avoid possible oscillation.



**Figure 33. Guard-Ring Schemes**

**noise performance**

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC2810Z results in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k $\Omega$  since bipolar devices exhibit greater noise currents.

## APPLICATION INFORMATION

### feedback

Operational amplifier circuits nearly always employ feedback and, since feedback is the first prerequisite for oscillation, a little caution is appropriate. Most oscillation problems result from driving capacitive loads and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 34). The value of this capacitor is optimized empirically.

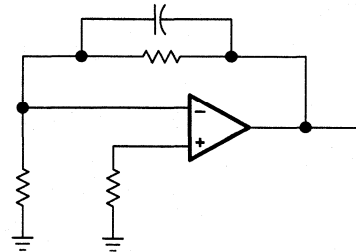


Figure 34. Compensation for Input Capacitance

### electrostatic discharge protection

The TLC2810Z incorporates an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

### latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC2810Z inputs and outputs are designed to withstand  $-100\text{-mA}$  surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not by design be forward biased. Applied input and output voltages should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ( $0.1\ \mu\text{F}$  typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

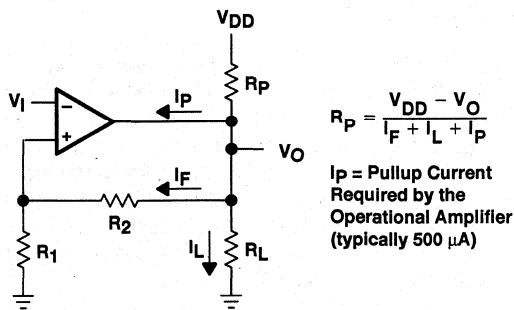
**APPLICATION INFORMATION**

**output characteristics**

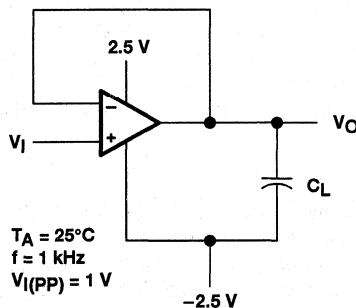
The output stage of the TLC2810Z is designed to sink and source relatively high amounts of current (see Typical Characteristics). If the output is subjected to a short-circuit condition, this high-current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

Although the TLC2810Z possesses excellent high-level output voltage and current capability, methods are available for boosting this capability if needed. The simplest method involves the use of a pullup resistor ( $R_P$ ) connected from the output to the positive supply rail (see Figure 35). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor, N4 (see equivalent schematic), must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60  $\Omega$  and 180  $\Omega$ , depending on how hard the operational amplifier input is driven. With very low values of  $R_P$ , a voltage offset from 0 V at the output occurs. Secondly, pullup resistor  $R_P$  acts as a drain load to N4, and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

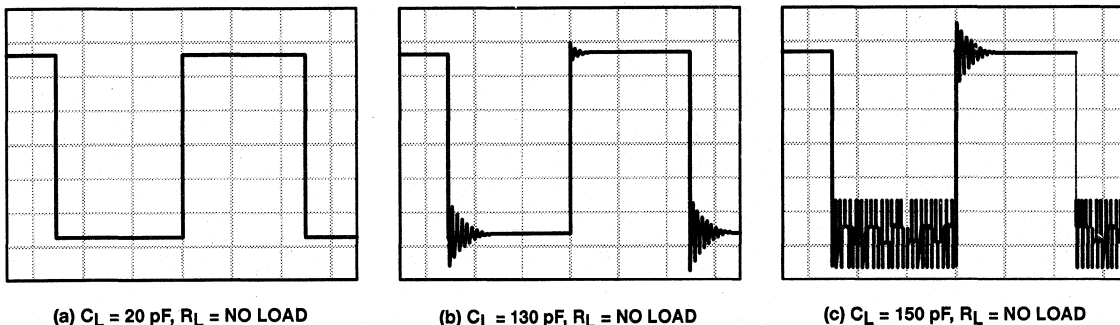
All operating characteristics of the TLC2810Z are measured using a 20-pF load. The devices can drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 37). In many cases, adding some compensation in the form of a series resistor in the feedback loop alleviates the problem.



**Figure 35. Resistive Pullup to Increase  $V_{OH}$**



**Figure 36. Test Circuit for Output Characteristics**



**Figure 37. Effect of Capacitive Loads**

# TLC2872Z, TLC2872Y

## Advanced LinCMOS™ RAIL-TO-RAIL DUAL OPERATIONAL AMPLIFIERS

SLOS117 – OCTOBER 1992

- **Free-Air Operating Temperature**  
–40°C to 150°C
- **Output Swing Includes Both Supply Rails**
- **Low Noise . . . 9 nV/√Hz Typ at f = 1 kHz**
- **Low Input Bias Current . . . 1 pA Typ**
- **Common-Mode Input Voltage Range Includes Negative Rail**
- **High Unity-Gain Bandwidth . . . 2.2 MHz Typ**
- **High Slew Rate . . . 3.6 V/μs Typ**
- **Low Input Offset Voltage**  
300 μV Typ at T<sub>A</sub> = 25°C
- **Macromodel Included**

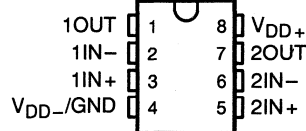
### description

The TLC2872Z is a dual rail-to-rail output operational amplifier manufactured using Texas Instruments Advanced LinCMOS™ process. These devices offer comparable ac performance while having better noise, input offset voltage and power dissipation than existing CMOS operational amplifiers. In addition, the common-mode input voltage range is wider than typical standard CMOS type amplifiers. To take advantage of this improvement in performance, making this device available for a wider range of applications, V<sub>ICR</sub> is specified with a larger maximum input offset voltage test limit of ±5 mV. The Advanced LinCMOS™ process uses a silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. Also, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

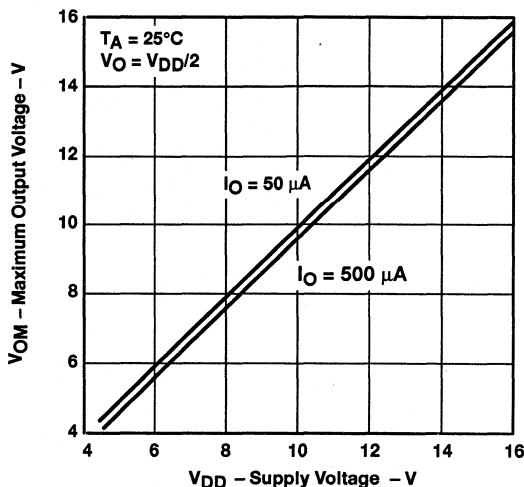
The TLC2872Z, manufactured using Texas Instruments high-temperature process flow, allows extended temperature operation up to 150°C in a plastic package. This adds extra reliability at the extended temperature and reduces the need for expensive hermetically sealed ceramic packages.

The TLC2872Z, which exhibits high input impedance and low noise, is excellent for small signal conditioning of high impedance sources, such as piezoelectric transducers. In addition, the rail-to-rail output feature with single or split supplies makes this device a great choice for inputs to ADCs in either the unipolar or bipolar mode of operation. This feature, combined with its temperature performance, makes the TLC2872Z ideal for sonobuoys, pressure sensors, temperature controls, active VR sensors, accelerometers, and many other applications.

D OR P PACKAGE  
(TOP VIEW)



MAXIMUM OUTPUT VOLTAGE  
vs  
SUPPLY VOLTAGE



### AVAILABLE OPTIONS

T <sub>A</sub>	V <sub>IO</sub> max AT 25°C	PACKAGED DEVICES		CHIP FORM (Y)
		SMALL OUTLINE (D)	PLASTIC DIP (P)	
–40°C to 150°C	2.5 mV	TLC2872ZD	TLC2872ZP	TLC2872Y

The D packages are available taped and reeled. Add R suffix to device type (e.g., TLC2872DR).

Advanced LinCMOS™ is a trademark of Texas Instruments Incorporated.

PRODUCTION DATA Information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



Copyright © 1992, Texas Instruments Incorporated

**TLC2872Z, TLC2872Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIERS**

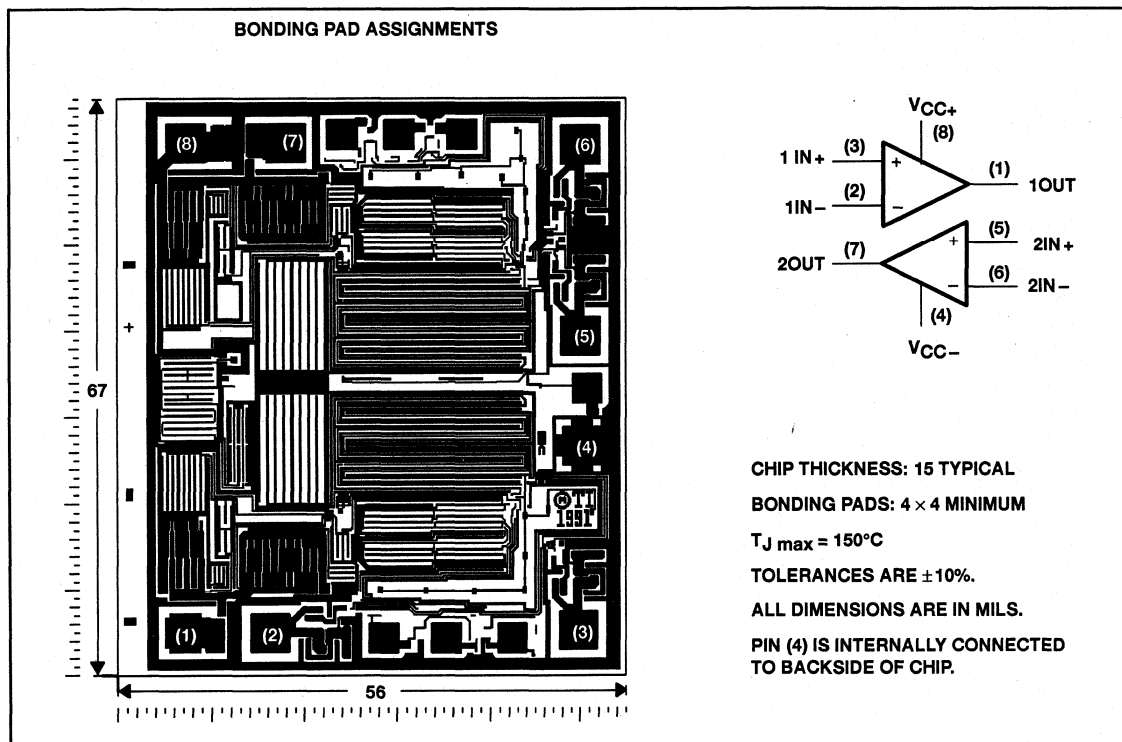
SLOS117 – OCTOBER 1992

**description (continued)**

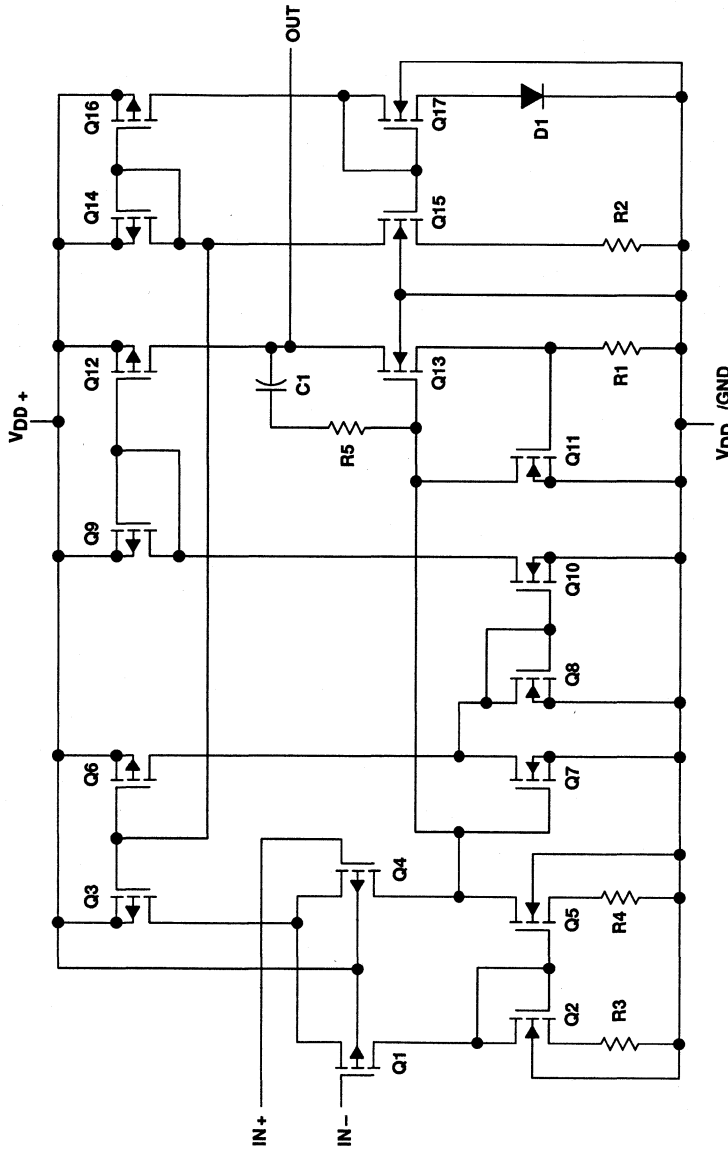
The inputs and outputs of this device are designed to withstand 100-mA surge current without sustaining latch-up. In addition, internal ESD-protection circuits prevent functional failures up to 2000 V. The device is characterized for operation over the extended (Z) temperature range of  $-40^{\circ}\text{C}$  to  $150^{\circ}\text{C}$ .

**TLC2872Y chip information**

This chip, when properly assembled, displays characteristics similar to TLC2872Z. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



equivalent schematic (each amplifier)



COMPONENT	COUNT†
Transistors	38
Diodes	9
Resistors	26
Capacitors	3

† Includes both amplifiers and all ESD, bias, and trim circuitry.

**TLC2872Z, TLC2872Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS117 – OCTOBER 1992

**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†**

Supply voltage, $V_{DD+}$ .....	8 V
Supply voltage, $V_{DD-}$ .....	-8 V
Differential input voltage, $V_{ID}$ (see Note 1) .....	$\pm 16$ V
Input voltage range, $V_I$ (any input, see Note 2) .....	$\pm 8$ V
Input current, $I_I$ (each input) .....	$\pm 5$ mA
Output current, $I_O$ .....	$\pm 50$ mA
Total current into $V_{DD+}$ .....	$\pm 50$ mA
Total current out of $V_{DD-}$ .....	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3) .....	unlimited
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ .....	-40°C to 150°C
Storage temperature range .....	-65°C to 165°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .....	260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. Differential voltages are at  $IN+$  with respect to  $IN-$ . Excessive current will flow if input is brought below  $V_{DD-} - 0.3$  V.  
 2. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

**DISSIPATION RATING TABLE**

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 105^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING	$T_A = 150^\circ\text{C}$ POWER RATING
D	812 mW	5.8 mW/°C	551 mW	348 mW	232 mW	87 mW
P	1120 mW	8 mW/°C	760 mW	480 mW	320 mW	120 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD\pm}$	$\pm 2.2$	$\pm 8$	V
Input voltage range, $V_I$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.5$	V
Operating free-air temperature, $T_A$	-40	150	°C





**electrical characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2872Z			UNIT
			MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{DD} = \pm 2.5\text{ V}$ , $V_O = 0$ , $V_{IC} = 0$ , $R_S = 50\ \Omega$	25°C	300		2500	$\mu\text{V}$
		Full range	3000			
$\alpha_{VIO}$ Temperature coefficient of input offset voltage		25°C to 150°C	2			$\mu\text{V}/^\circ\text{C}$
		25°C	0.002			
Input offset voltage long-term drift (see Note 4)		25°C	0.0005			nA
$I_{IO}$ Input offset current		Full range	3			
	$I_{IB}$ Input bias current	25°C	0.001			nA
Full range		5				
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega$ , $ V_{IO}  \leq 5\text{ mV}$	25°C	0 to 4	-0.3 to 4.2	V	
		Full range	0 to 3.5			
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	25°C	4.95	4.99	V	
		25°C	4.85	4.93		
		Full range	4.75			
		$I_{OH} = -1\text{ mA}$	25°C	4.25		4.65
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 50\ \mu\text{A}$	25°C	0.01	0.02	V	
		25°C	0.09	0.15		
		Full range	0.2			
		$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 500\ \mu\text{A}$	25°C	0.9		1.5
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\text{ V}$ , $I_{OL} = 5\text{ mA}$	25°C	0.01		V	
		25°C	0.09			
		Full range	0.2			
		Full range	2			
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$ , $V_O = 1\text{ V to }4\text{ V}$	$R_L = 10\text{ k}\Omega$ ‡	25°C	15	35	V/mV
			Full range	10		
		$R_L = 1\text{ M}\Omega$ ‡	25°C	175		
$r_{id}$ Differential input resistance		25°C	10 <sup>12</sup>		$\Omega$	
$r_i$ Common-mode input resistance		25°C	10 <sup>12</sup>		$\Omega$	
$c_i$ Common-mode input capacitance	$f = 10\text{ kHz}$ , P package	25°C	8		pF	
$z_o$ Closed-loop output impedance	$f = 1\text{ MHz}$ , $A_V = 10$	25°C	140		$\Omega$	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$ , $R_S = 50\ \Omega$ , $V_O = 2.5\text{ V}$	25°C	70	75	dB	
		Full range	70			
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD}/\Delta V_{IO}$ )	$V_{DD} = 4.4\text{ V to }16\text{ V}$ , $V_{IC} = V_{DD}/2$ , No load	25°C	80	95	dB	
		Full range	80			
$I_{DD}$ Supply current	$V_O = 2.5\text{ V}$ , No load	25°C	2.2	3	mA	
		Full range	3			

† Full range is -40°C to 150°C.

‡ Referenced to 2.5 V

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

**TLC2872Z, TLC2872Y**  
**Advanced LinCMOST™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS117 – OCTOBER 1992

**operating characteristics at specified free-air temperature,  $V_{DD} = 5\text{ V}$**

PARAMETER	TEST CONDITIONS	$T_A$ †	TLC2872Z			UNIT	
			MIN	TYP	MAX		
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	2.3	3.6		V/ $\mu\text{s}$	
		Full range	1.1				
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$ $f = 1\text{ kHz}$	25°C	50			nV/ $\sqrt{\text{Hz}}$	
		25°C	9				
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$ $f = 0.1\text{ to }10\text{ Hz}$	25°C	1			$\mu\text{V}$	
		25°C	1.4				
$I_n$ Equivalent input noise current		25°C	0.6			fA/ $\sqrt{\text{Hz}}$	
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega$ ‡	25°C	$A_V = 1$	0.0013%			
			$A_V = 10$	0.004%			
			$A_V = 100$	0.03%			
Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}$ ‡	$R_L = 10\text{ k}\Omega$ ‡, 25°C	2.18			MHz	
$B_{OM}$ Maximum output-swing bandwidth	$V_{O(PP)} = 2\text{ V}$ , $R_L = 10\text{ k}\Omega$ ‡	$A_V = 1$ , $C_L = 100\text{ pF}$ ‡	25°C	1			MHz
Settling time	$A_V = -1$ , Step = 0.5 V to 2.5 V, $R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	To 0.1%	25°C	1.5		$\mu\text{s}$	
		To 0.01%		2.6			
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega$ ‡, $C_L = 100\text{ pF}$ ‡	25°C	50°				
		25°C	10				dB

† Full range is  $-40^\circ\text{C}$  to  $150^\circ\text{C}$ .

‡ Referenced to 2.5 V

**TLC2872Z, TLC1872Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS117 – OCTOBER 1992

**electrical characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)**

PARAMETER	TEST CONDITIONS	TLC2872Y			UNIT
		MIN	TYP	MAX	
$V_{IO}$ Input offset voltage	$V_{IC} = 0,$ $R_S = 50\ \Omega$ $V_O = 0,$		300	2500	$\mu\text{V}$
Input offset voltage long-term drift (see Note 4)			0.002		$\mu\text{V}/\text{mo}$
$I_{IO}$ Input offset current			0.0005		nA
$I_{IB}$ Input bias current			0.001		nA
$V_{ICR}$ Common-mode input voltage range	$R_S = 50\ \Omega,$ $ V_{IO}  \leq 5\text{ mV}$	0 to 4	-0.3 to 4.2		V
$V_{OH}$ High-level output voltage	$I_{OH} = -20\ \mu\text{A}$	4.95	4.99		V
	$I_{OH} = -200\ \mu\text{A}$	4.85	4.93		
	$I_{OH} = -1\ \text{mA}$	4.25	4.65		
$V_{OL}$ Low-level output voltage	$V_{IC} = 2.5\ \text{V},$ $I_{OL} = 50\ \mu\text{A}$		0.01	0.02	V
	$V_{IC} = 2.5\ \text{V},$ $I_{OL} = 500\ \mu\text{A}$		0.09	0.15	
	$V_{IC} = 2.5\ \text{V},$ $I_{OL} = 5\ \text{mA}$		0.9	1.5	
$A_{VD}$ Large-signal differential voltage amplification	$V_{IC} = 2.5\ \text{V},$ $V_O = 1\ \text{V to } 4\ \text{V}$	$R_L = 10\ \text{k}\Omega^\dagger$	15	35	V/mV
		$R_L = 1\ \text{M}\Omega^\dagger$		175	
$r_{id}$ Differential input resistance			$10^{12}$		$\Omega$
$r_i$ Common-mode input resistance			$10^{12}$		$\Omega$
$c_i$ Common-mode input capacitance	$f = 10\ \text{kHz},$ P package		8		pF
$z_o$ Closed-loop output impedance	$f = 1\ \text{MHz},$ $A_V = 10$		140		$\Omega$
CMRR Common-mode rejection ratio	$V_{IC} = 0\ \text{to } 2.7\ \text{V},$ $R_S = 50\ \Omega$ $V_O = 2.5\ \text{V},$	70	75		dB
$k_{SVR}$ Supply voltage rejection ratio ( $\Delta V_{DD\pm}/\Delta V_{IO}$ )	$V_{DD} = 4.4\ \text{V to } 16\ \text{V},$ No load $V_{IC} = V_{DD}/2,$	80	95		dB
$I_{DD}$ Supply current	$V_O = 2.5\ \text{V},$ No load	2.2	3		mA

† Referenced to 2.5 V

**TLC2872Z, TLC2872Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS117 – OCTOBER 1992

**operating characteristics,  $V_{DD} = 5\text{ V}$ ,  $T_A = 25^\circ\text{C}$**

PARAMETER	TEST CONDITIONS	TLC2872Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\dagger$ , $C_L = 100\text{ pF}^\dagger$	2.3	3.6		V/ $\mu\text{s}$
$V_n$ Equivalent input noise voltage	$f = 10\text{ Hz}$		50		nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$		9		
$V_N(\text{PP})$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$		1		$\mu\text{V}$
	$f = 0.1\text{ to }10\text{ Hz}$		1.4		
$I_n$ Equivalent input noise current			0.6		fA/ $\sqrt{\text{Hz}}$
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}$ , $f = 20\text{ kHz}$ , $R_L = 10\text{ k}\Omega^\dagger$	$A_V = 1$	0.0013%		
		$A_V = 10$	0.004%		
		$A_V = 100$	0.03%		
Gain-bandwidth product	$f = 10\text{ kHz}$ , $C_L = 100\text{ pF}^\dagger$	$R_L = 10\text{ k}\Omega^\dagger$	2.18		MHz
BOM Maximum output-swing bandwidth	$V_O(\text{PP}) = 4.6\text{ V}$ , $R_L = 10\text{ k}\Omega^\dagger$	$A_V = 1$ , $C_L = 100\text{ pF}^\dagger$	1		MHz
Settling time	$A_V = -1$ , Step = $0.5\text{ V to }2.5\text{ V}$ , $R_L = 10\text{ k}\Omega^\dagger$ , $C_L = 100\text{ pF}^\dagger$	To 0.1%	1.5		$\mu\text{s}$
		To 0.01%	2.6		
$\phi_m$ Phase margin at unity gain	$R_L = 10\text{ k}\Omega^\dagger$ , $C_L = 100\text{ pF}^\dagger$	50°			
Gain margin		10			dB

$^\dagger$  Referenced to 2.5 V



**TYPICAL CHARACTERISTICS**

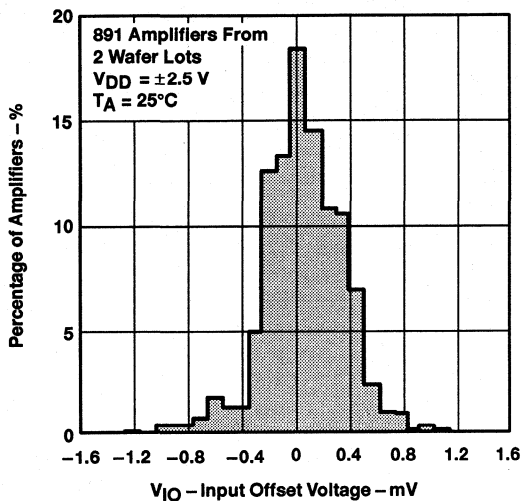
**Table of Graphs**

			FIGURE
$V_{IO}$	Input offset voltage	Distribution	1
$\alpha_{VIO}$	Input offset voltage temperature coefficient	Distribution	2
$I_{IB}/I_{IO}$	Input bias and offset currents	vs Free-air temperature	3
$V_I$	Input voltage range	vs Free-air temperature	4
$V_{OH}$	High-level output voltage	vs Output current	5
$V_{OL}$	Low-level output voltage	vs Output current	6, 7
$V_{OM}$	Maximum output voltage	vs Frequency	8
$I_{OS}$	Short-circuit output current	vs Supply voltage	9
		vs Free-air temperature	10
$A_{VD}$	Large-signal differential voltage amplification	vs Load resistance	11
		vs Frequency	12
		vs Free-air temperature	13
$I_{DD}$	Supply current	vs Supply voltage	14
		vs Free-air temperature	15
SR	Slew rate	vs Load capacitance	16
		vs Free-air temperature	17
$\phi_m$	Phase margin	vs Frequency	12
		vs Load capacitance	18
	Gain margin	vs Load capacitance	19

NOTE: All loads are referenced to 2.5 V.

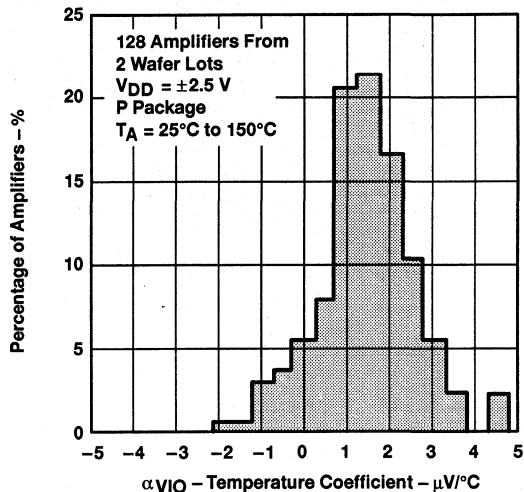
**TYPICAL CHARACTERISTICS**

**DISTRIBUTION OF TLC2872Z  
 INPUT OFFSET VOLTAGE**



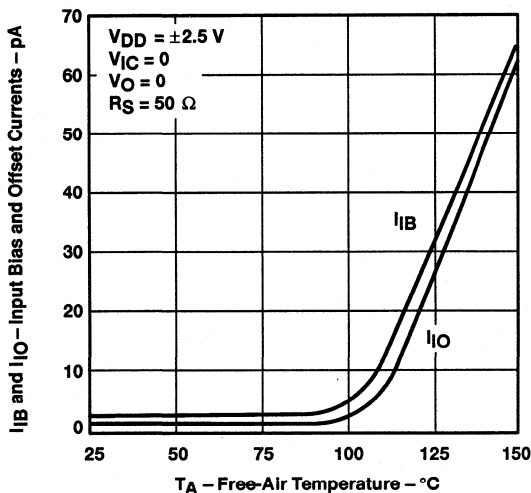
**Figure 1**

**DISTRIBUTION OF TLC2872Z  
 TEMPERATURE COEFFICIENT**



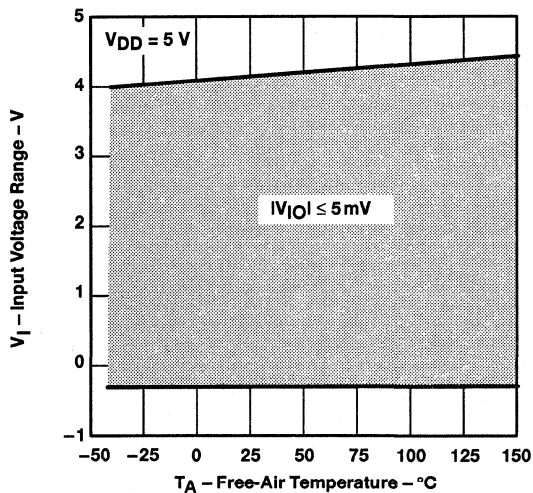
**Figure 2**

**INPUT BIAS AND OFFSET CURRENTS  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 3**

**INPUT VOLTAGE RANGE  
 vs  
 FREE-AIR TEMPERATURE**



**Figure 4**

TYPICAL CHARACTERISTICS†

HIGH-LEVEL OUTPUT VOLTAGE  
 VS  
 HIGH-LEVEL OUTPUT CURRENT

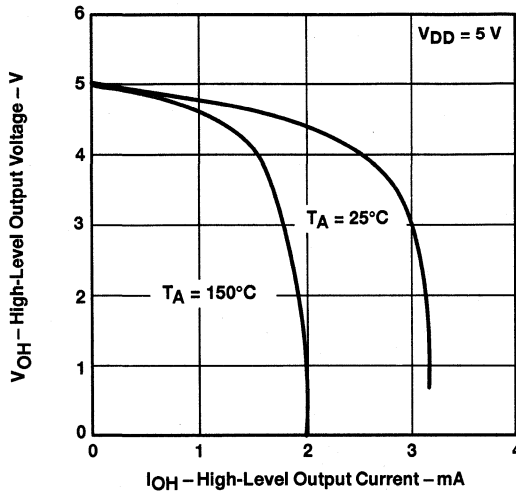


Figure 5

LOW-LEVEL OUTPUT VOLTAGE  
 VS  
 LOW-LEVEL OUTPUT CURRENT

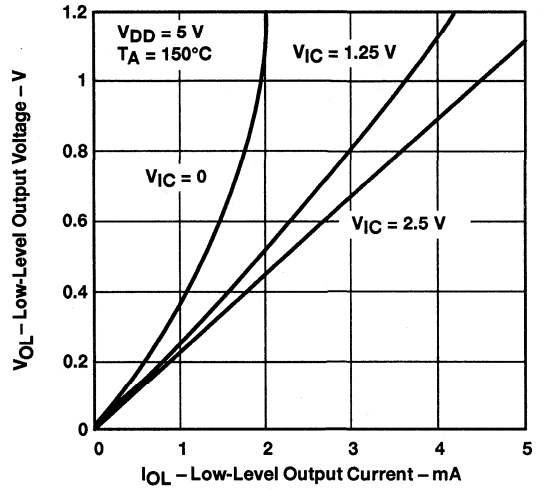


Figure 6

LOW-LEVEL OUTPUT VOLTAGE  
 VS  
 LOW-LEVEL OUTPUT CURRENT

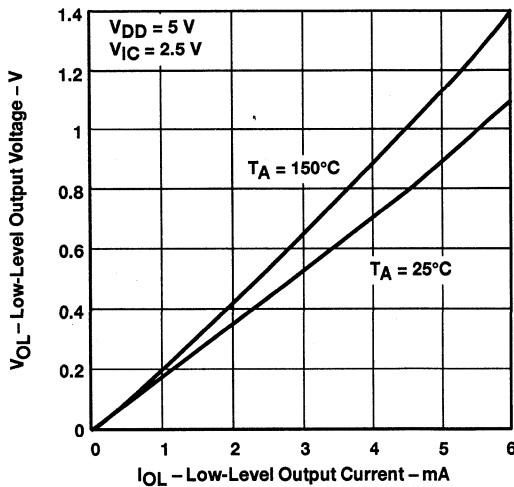


Figure 7

MAXIMUM OUTPUT VOLTAGE  
 VS  
 FREQUENCY

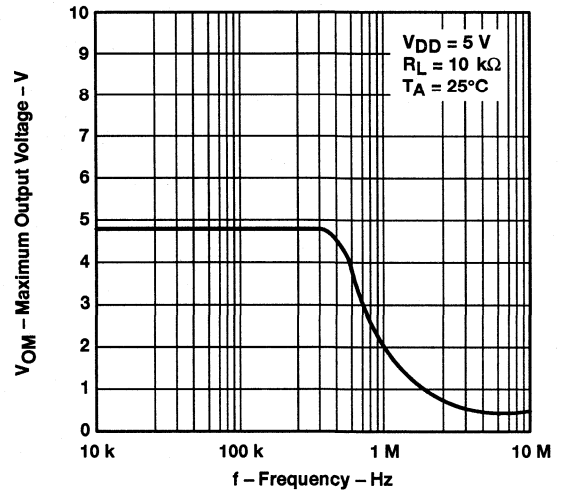


Figure 8

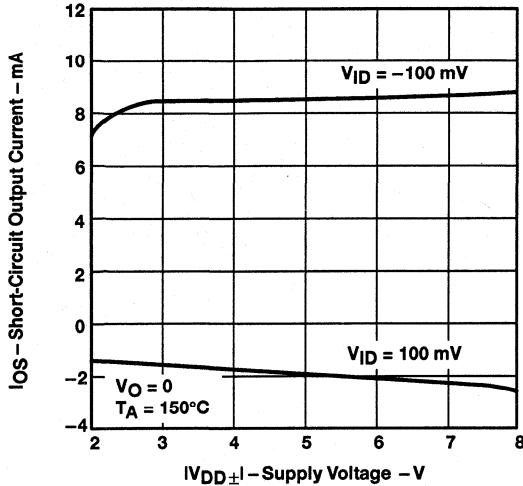
† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

**TLC2872Z, TLC2872Y**  
**Advanced LinCMOS™ RAIL-TO-RAIL**  
**DUAL OPERATIONAL AMPLIFIER**

SLOS117 – OCTOBER 1992

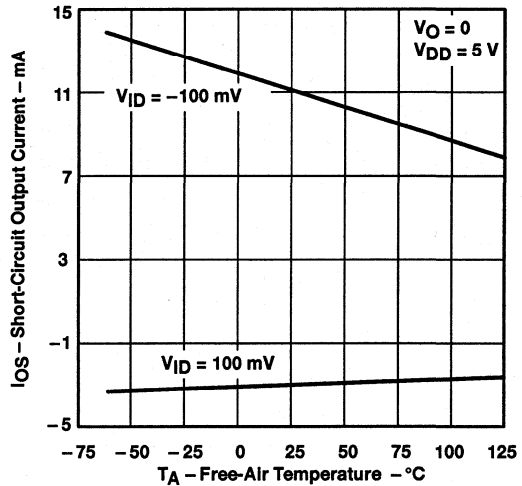
**TYPICAL CHARACTERISTICS†**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**SUPPLY VOLTAGE**



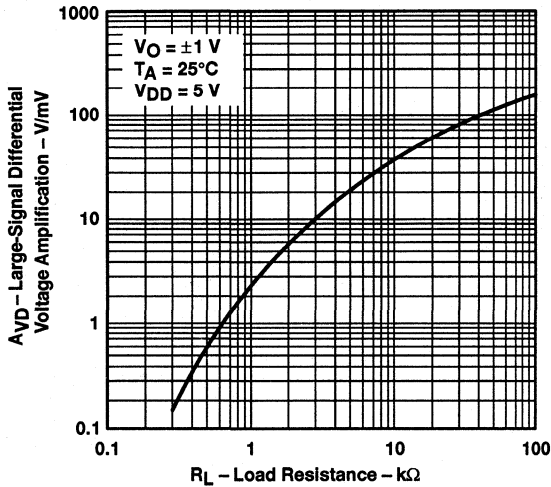
**Figure 9**

**SHORT-CIRCUIT OUTPUT CURRENT**  
**vs**  
**FREE-AIR TEMPERATURE**



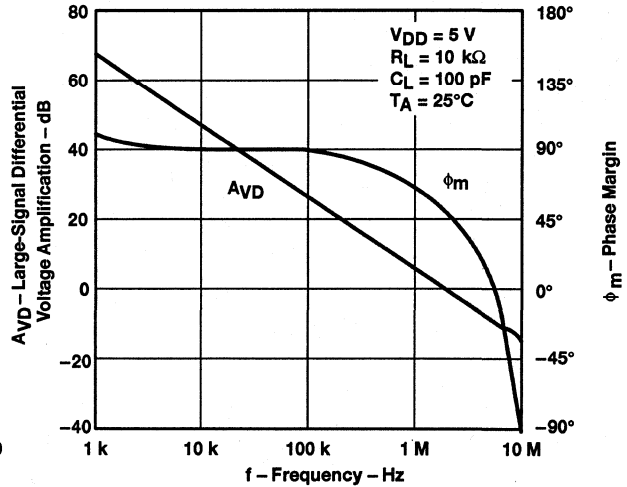
**Figure 10**

**LARGE-SIGNAL DIFFERENTIAL**  
**VOLTAGE AMPLIFICATION**  
**vs**  
**LOAD RESISTANCE**



**Figure 11**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE**  
**AMPLIFICATION and PHASE MARGIN**  
**vs**  
**FREQUENCY**



**Figure 12**

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL  
 VOLTAGE AMPLIFICATION  
 VS  
 FREE-AIR TEMPERATURE

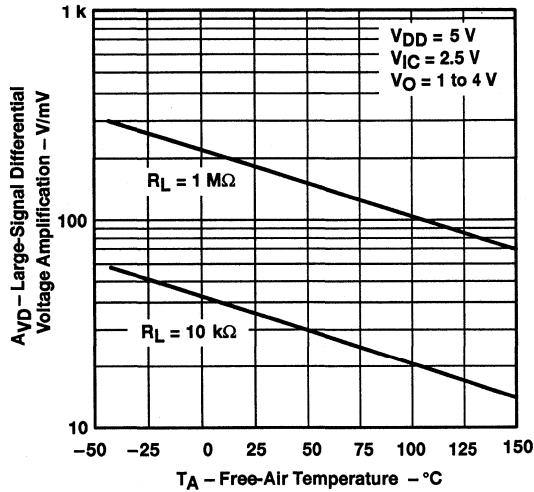


Figure 13

SUPPLY CURRENT  
 VS  
 SUPPLY VOLTAGE

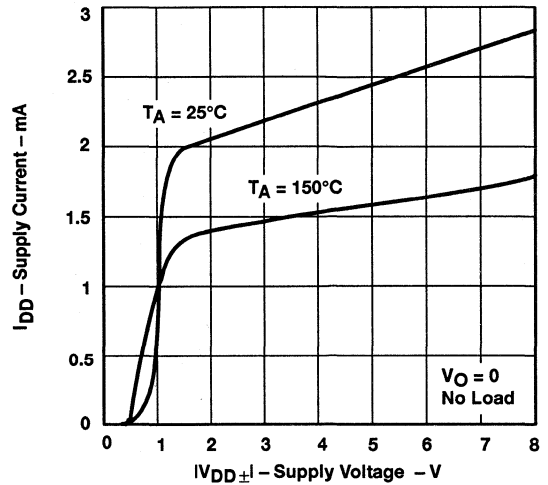


Figure 14

SUPPLY CURRENT  
 VS  
 FREE-AIR TEMPERATURE

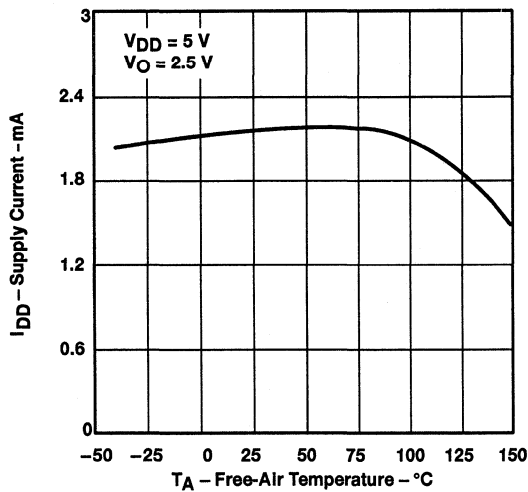


Figure 15

SLEW RATE  
 VS  
 LOAD CAPACITANCE

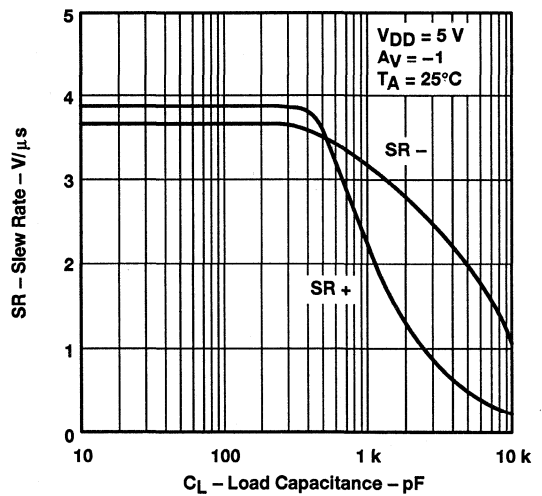


Figure 16

TYPICAL CHARACTERISTICS†

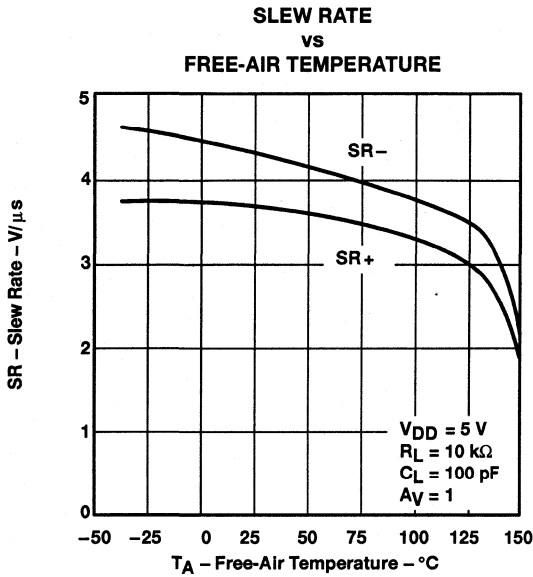


Figure 17

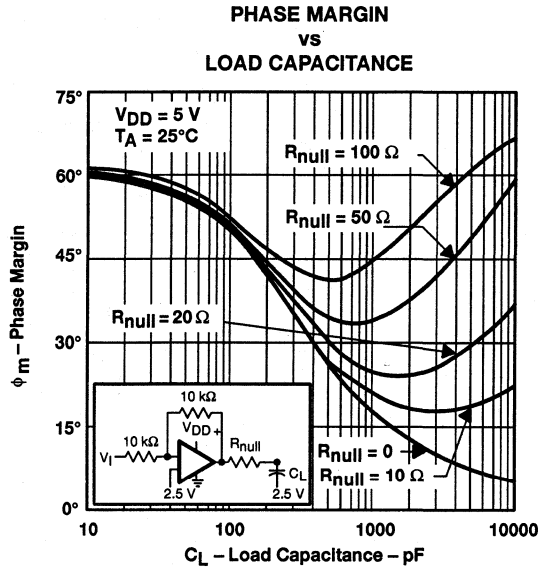


Figure 18

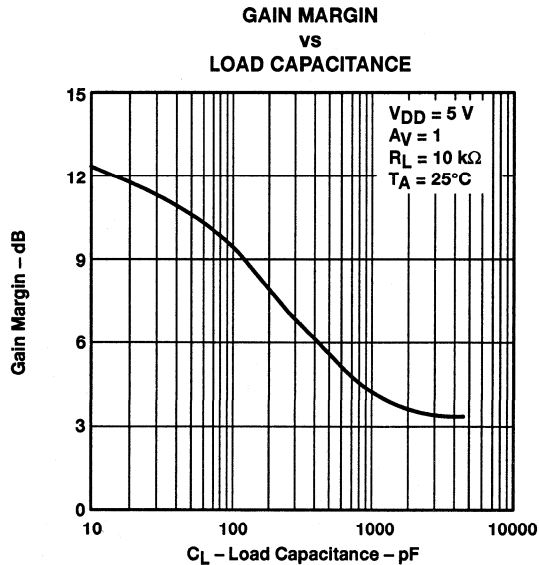


Figure 19

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

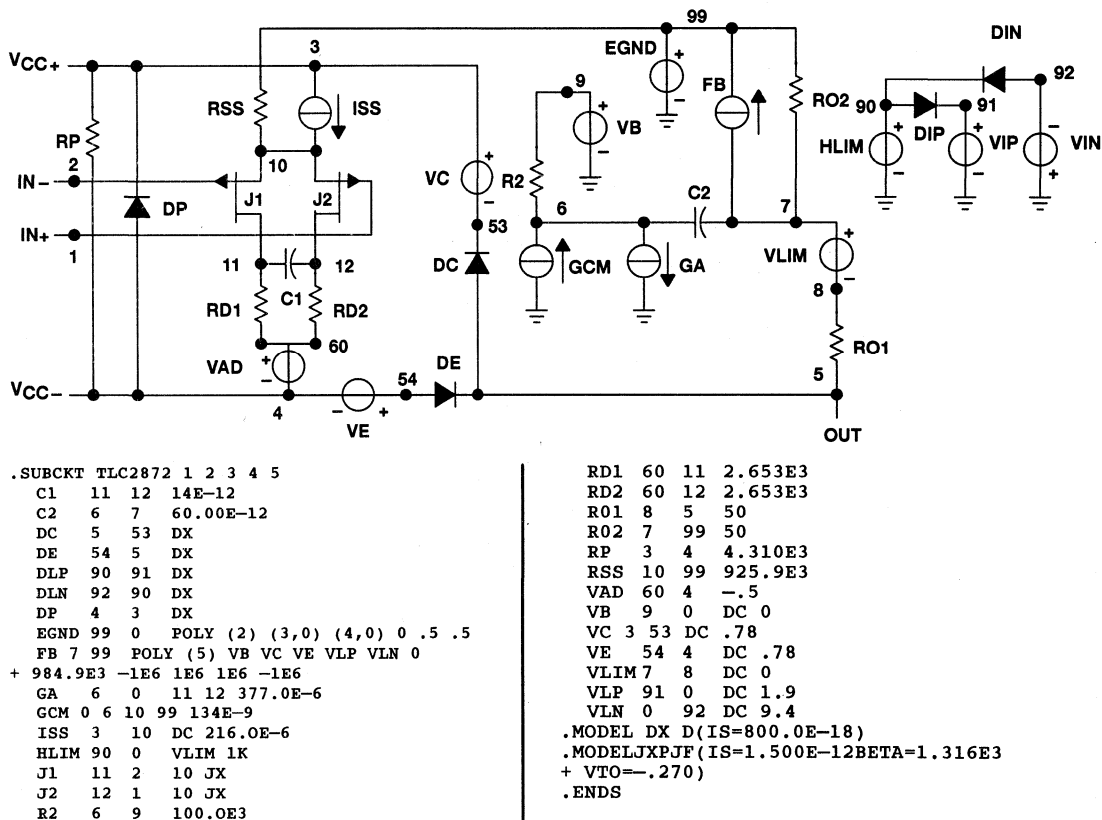
## APPLICATION INFORMATION

### macromodel information

Macromodel information provided was derived using *PSpice™ Parts™* model generation software. The Boyle macromodel (see Note 5) and subcircuit in Figure 20 were generated using the TLC2872Z typical electrical and operating characteristics at  $T_A = 25^\circ\text{C}$ . Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 5: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).



**Figure 20. Boyle Macromodel and Subcircuit**

*PSpice* and *Parts* are trademarks of MicroSim Corporation.

Macromodels, simulation models, or other models provided by TI, directly or indirectly, are not warranted by TI as fully representing all of the specification and operating characteristics of the semiconductor product to which the model relates.





<b>General Information (Volume A)</b>	<b>1</b>
<b>Operational Amplifiers</b>	<b>2</b>
<b>Mechanical Data (Volume A)</b>	<b>3</b>
<b>General Information (Volume B)</b>	<b>4</b>
<b>Operational Amplifiers (continued)</b>	<b>5</b>
<b>Comparators</b>	<b>6</b>
<b>Mechanical Data (Volume B)</b>	<b>7</b>

# Contents

	Page
Ordering Instructions .....	3-3
Mechanical Data .....	3-5

# ORDERING INSTRUCTIONS

Electrical characteristics presented in this data book, unless otherwise noted, apply for the circuit type(s) listed in the page heading regardless of package. The availability of a circuit function in a particular package is denoted by an alphabetical reference above the pin-connection diagram(s). These alphabetical references refer to mechanical outline drawings shown in this section.

Factory orders for circuits described in this data book should include a four-part type number as shown in the following example.

**Example:**                      **TLE**        **2022**                      **PW**                      **LE**

**Prefix** \_\_\_\_\_

**MUST CONTAIN TWO OR THREE LETTERS**

- SN ..... TI Special Functions or Interface Products
- TL, TLE ..... TI Linear Products
- TLC ..... TI Linear Silicon-Gate CMOS Products

**STANDARD SECOND-SOURCE PREFIXES**

- AD ..... Analog Devices
- ADC, LF, LM, LP, or MP ..... National
- LT or LTC ..... Linear Technology
- MC ..... Motorola
- NE, SA, or SE ..... Signetics
- OP ..... PMI
- RC, RM, or RV ..... Raytheon
- uA ..... Fairchild/National
- UC ..... Unitrode

**Unique Circuit Description Including Temperature Range** \_\_\_\_\_

**MUST CONTAIN TWO OR MORE CHARACTERS**

(from individual data sheets)

- Examples:    10                      34070  
                   592                      1451AC  
                   7757                      2217-285

**Package** \_\_\_\_\_

**MUST CONTAIN ONE OR TWO LETTERS**

- D, DB, DW, FK, FN, J, JD, JG, KC, KV, LP, LPF, N, NE, NS, NT, NW, P, PK, PW, U, W
- (from pin-connection diagrams on individual data sheet)

**Available Taped and Reeled or Left-Ended Taped and Reeled** \_\_\_\_\_

- R – Available Taped and Reeled
- LE – Available Only Left-Ended Taped and Reeled



# ORDERING INSTRUCTIONS

---

Circuits are shipped in one of the carriers below. Unless a specific method of shipment is specified by the customer (with possible additional costs), circuits will be shipped via the most practical carrier.

Dual-In-Line (J, JD, JG, N, NT, NS, NW, P)

- A-Channel Antistatic or Conductive Plastic Tubing

Shrink Small Outline (DB)

- Tape and Reel
- Thin Shrink Small Outline (PW)
- Tape and Reel

Plug-In (LP)

- Plastic Bag
- Tape and Reel

Small Outline (D, DW)

- Tape and Reel
- Antistatic or Conductive Plastic Tubing

Chip Carriers (FK, FN)

- Antistatic or Conductive Plastic Tubing

Flat (U, W)

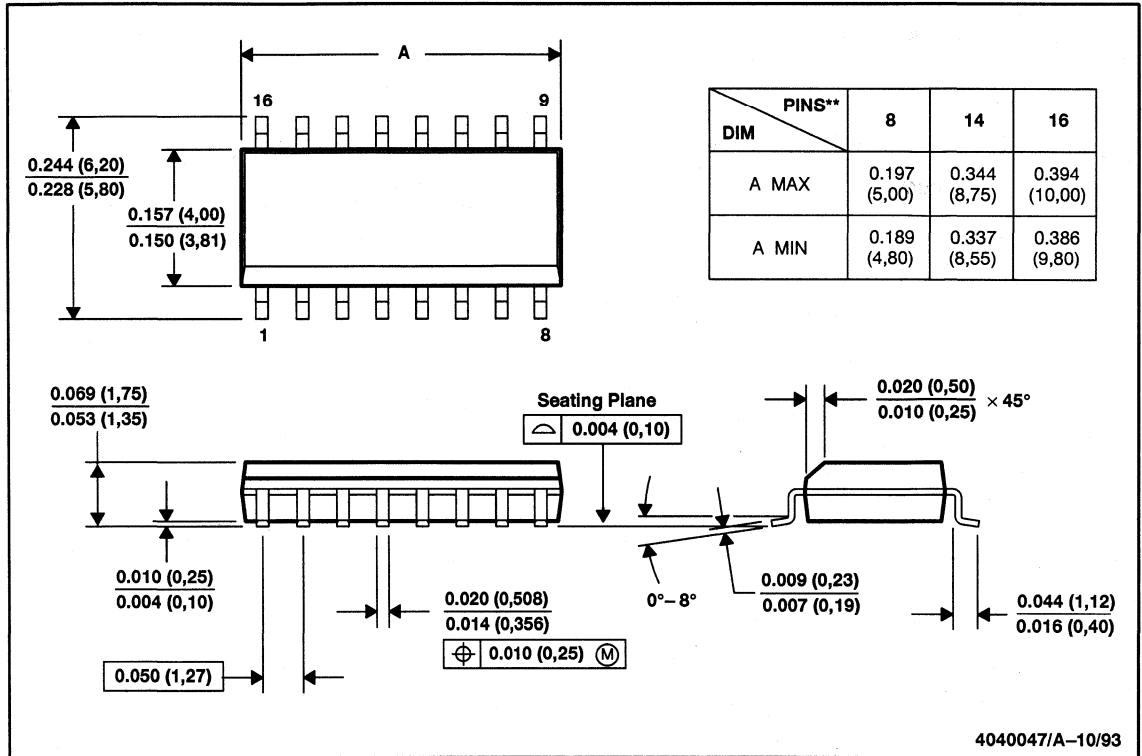
- Milton Ross Carriers



D/R-PDSO-G\*\*

PLASTIC NARROW-BODY SMALL-OUTLINE PACKAGE

16 PIN SHOWN



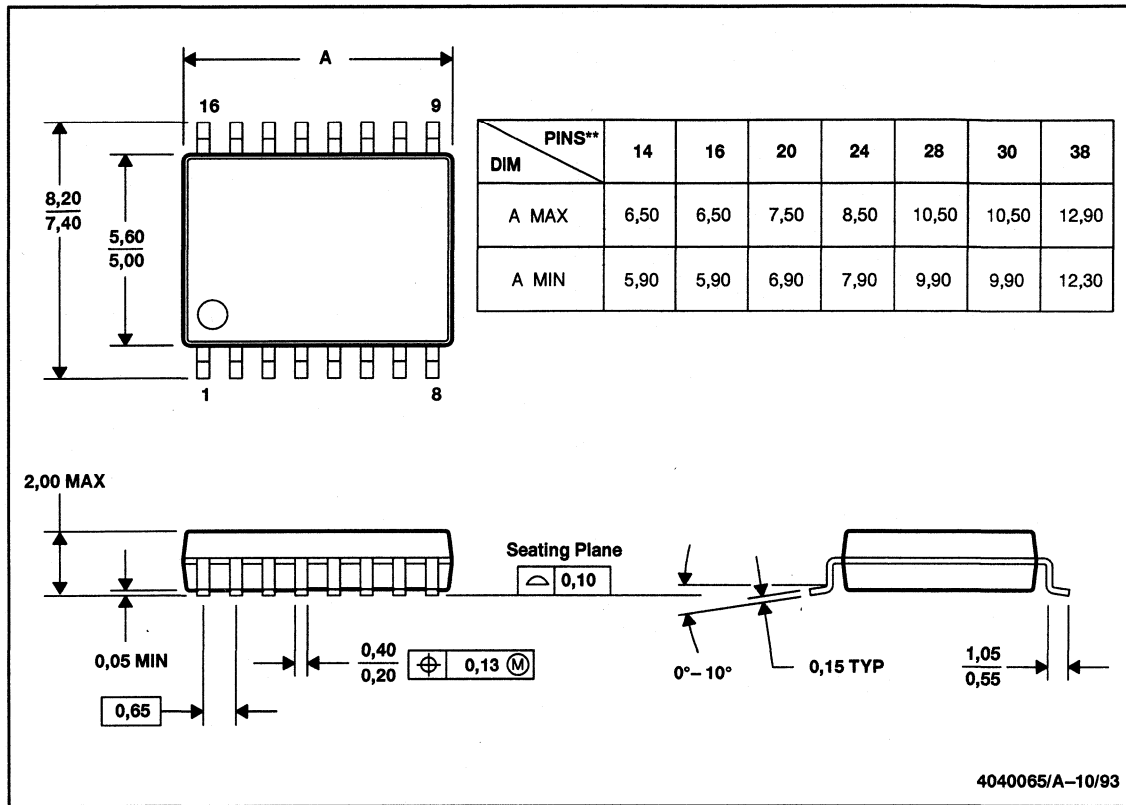
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Body dimensions do not include mold flash or protrusion.  
 D. Mold protrusion shall not exceed 0.006 (0,15).

# MECHANICAL DATA

DB/R-PDSO-G\*\*

PLASTIC SMALL-OUTLINE PACKAGE

16 PIN SHOWN

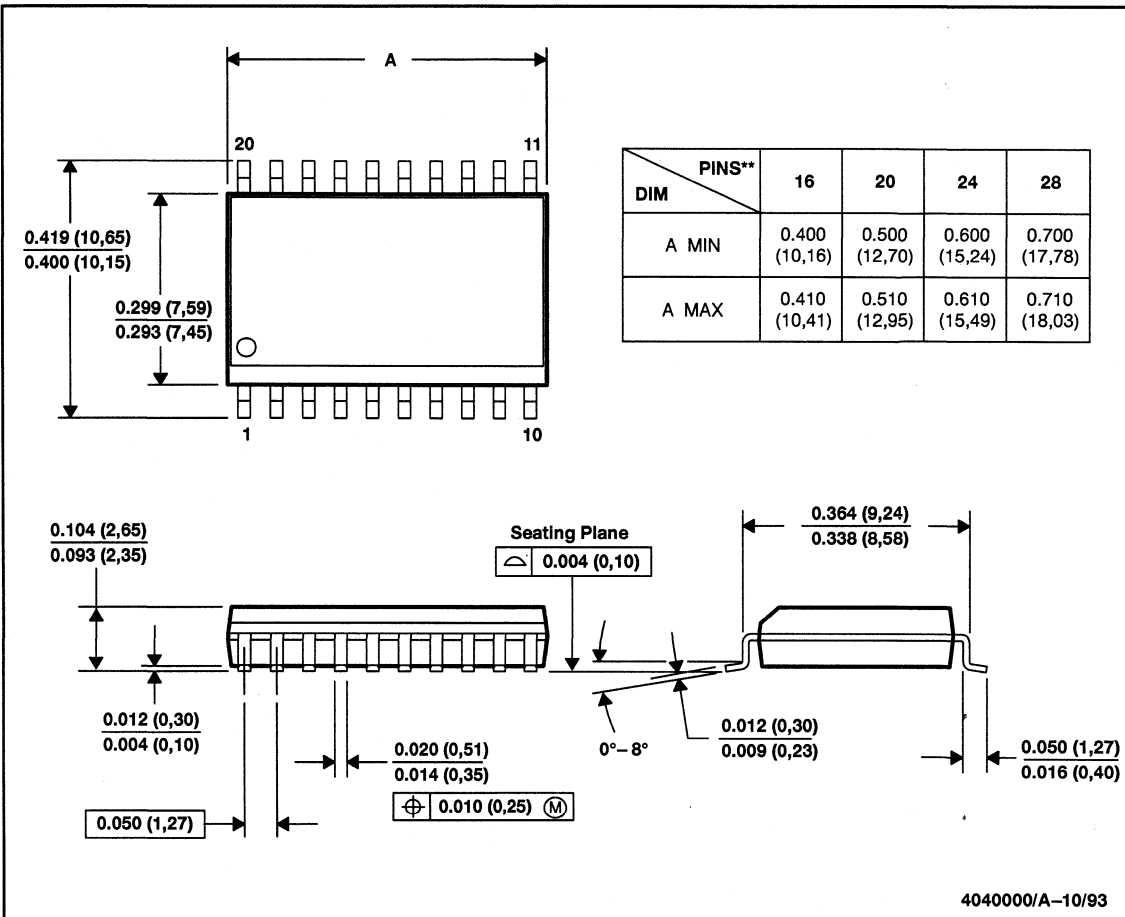


- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

DW/R-PDSO-G\*\*

PLASTIC WIDE-BODY SMALL-OUTLINE PACKAGE

20 PIN SHOWN



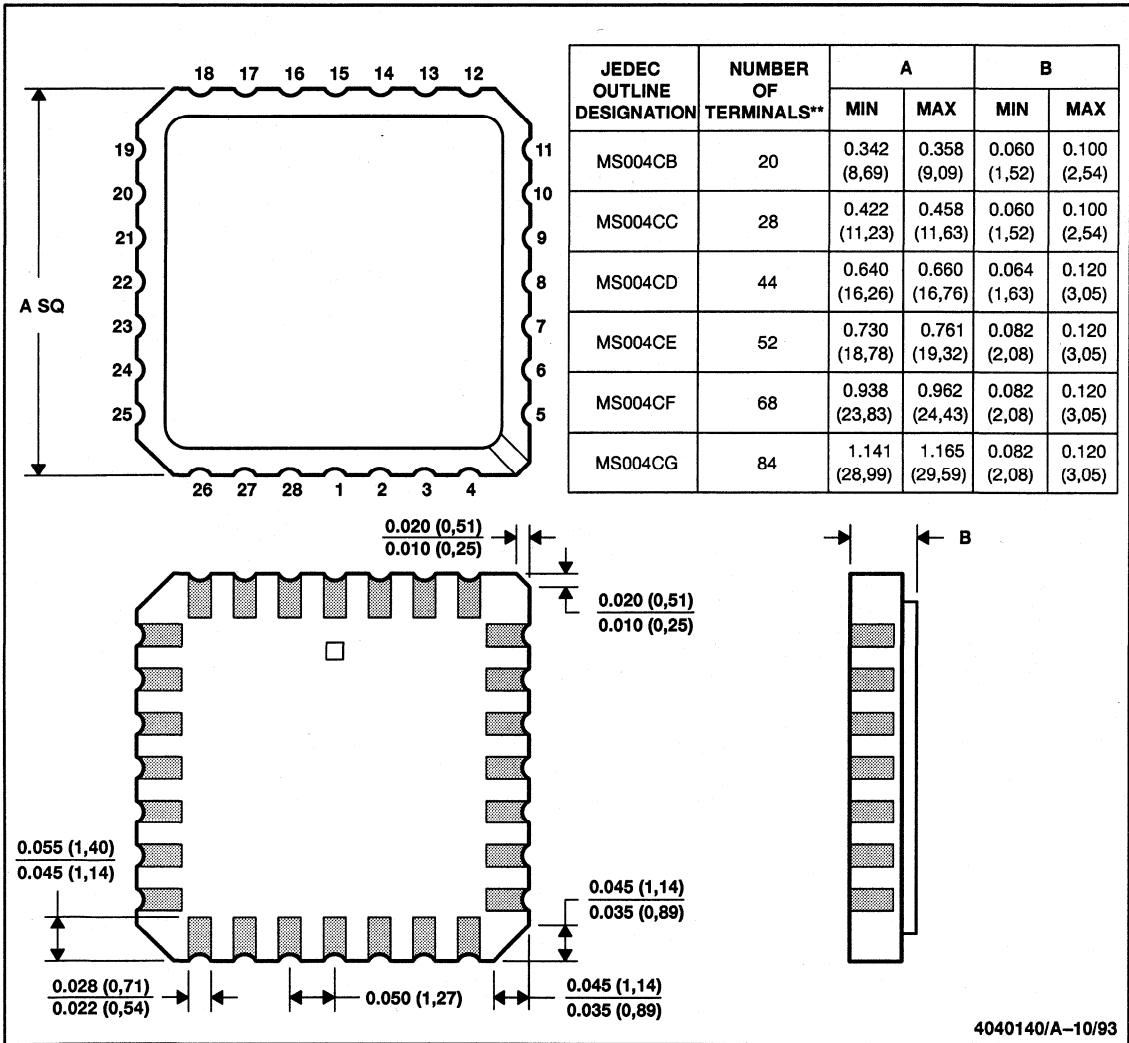
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).

# MECHANICAL DATA

FK/S-CQCC-N\*\*

LEADLESS CERAMIC CHIP CARRIER

28 TERMINAL SHOWN

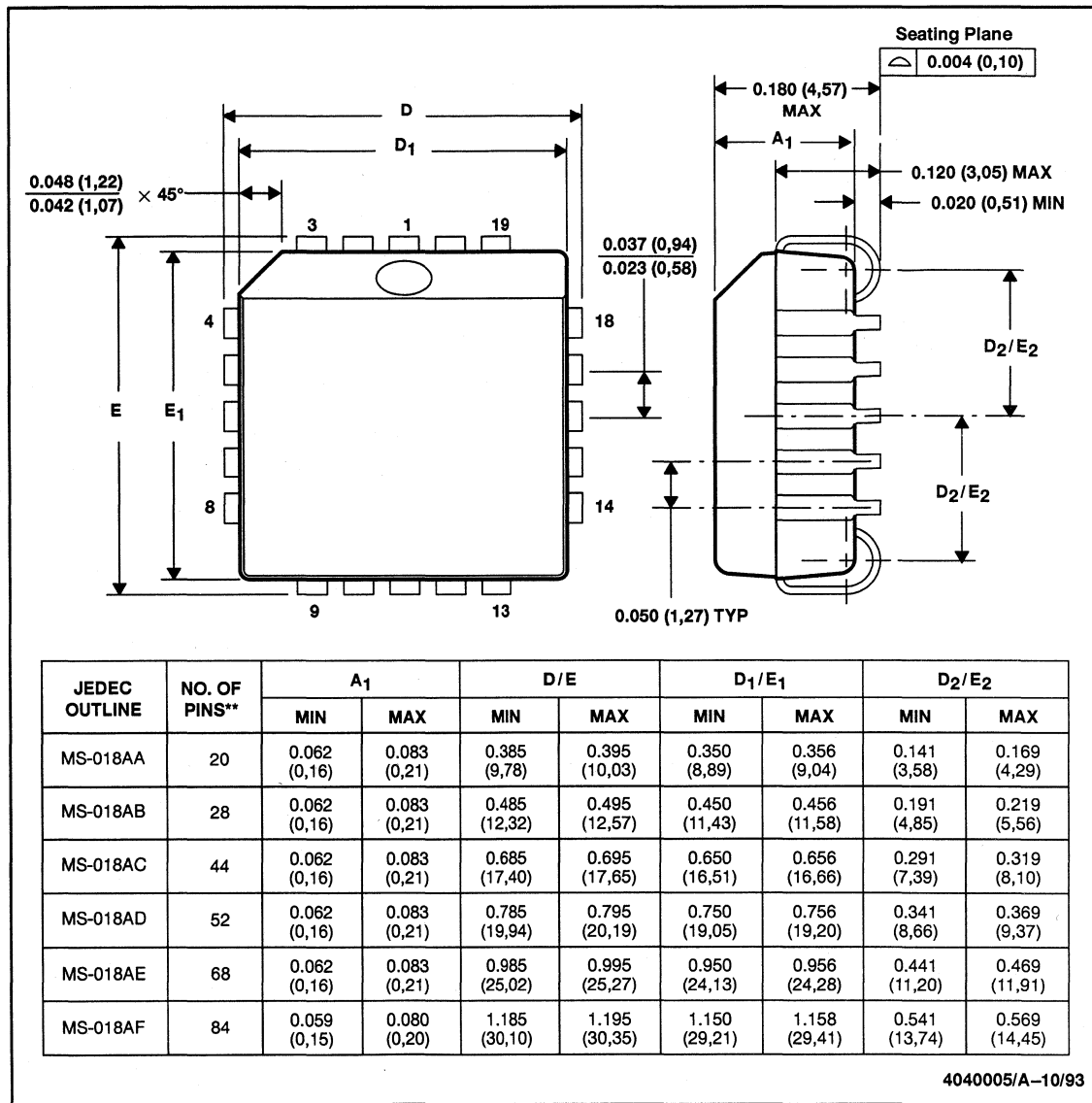


- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a metal lid.  
 D. The terminals will be gold plated.

FN/S-PQCC-J\*\*

PLASTIC J-LEADED CHIP CARRIER

20 PIN SHOWN



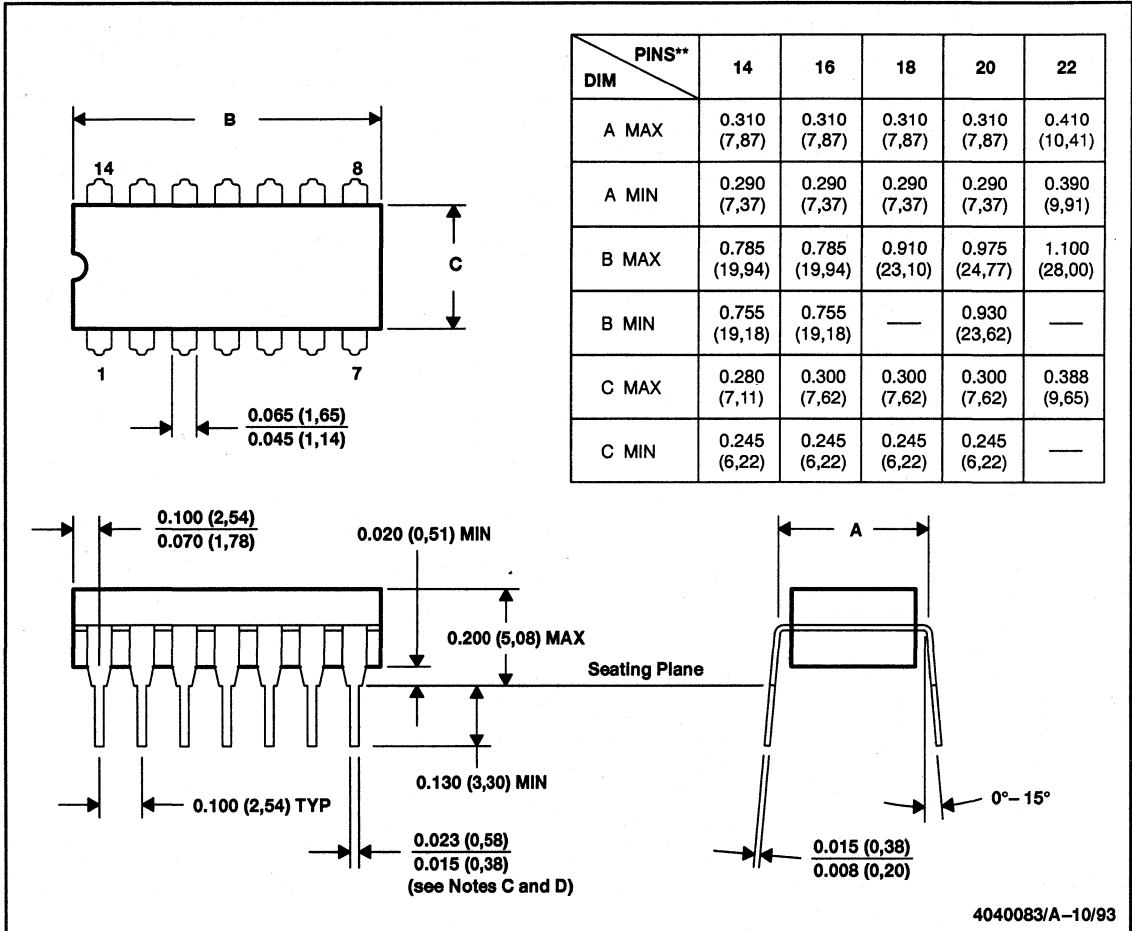
NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Falls within JEDEC MS-018

# MECHANICAL DATA

J/R-GDIP-T\*\*

CERAMIC DUAL-IN-LINE PACKAGE

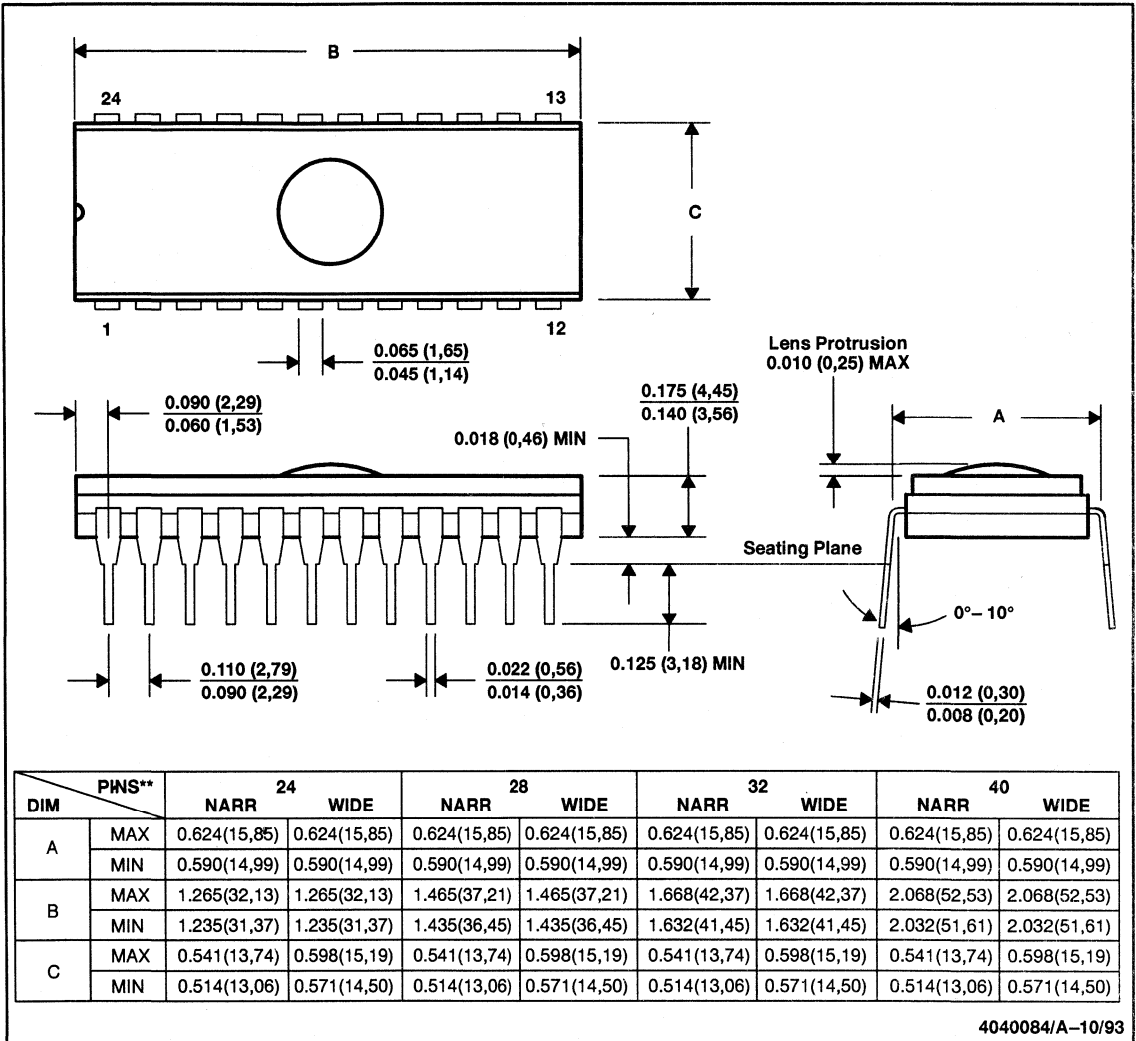
14 PIN SHOWN



- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.  
 D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.

J/R-CDIP-T\*\*  
24 PIN SHOWN

CERAMIC DUAL-IN-LINE PACKAGE



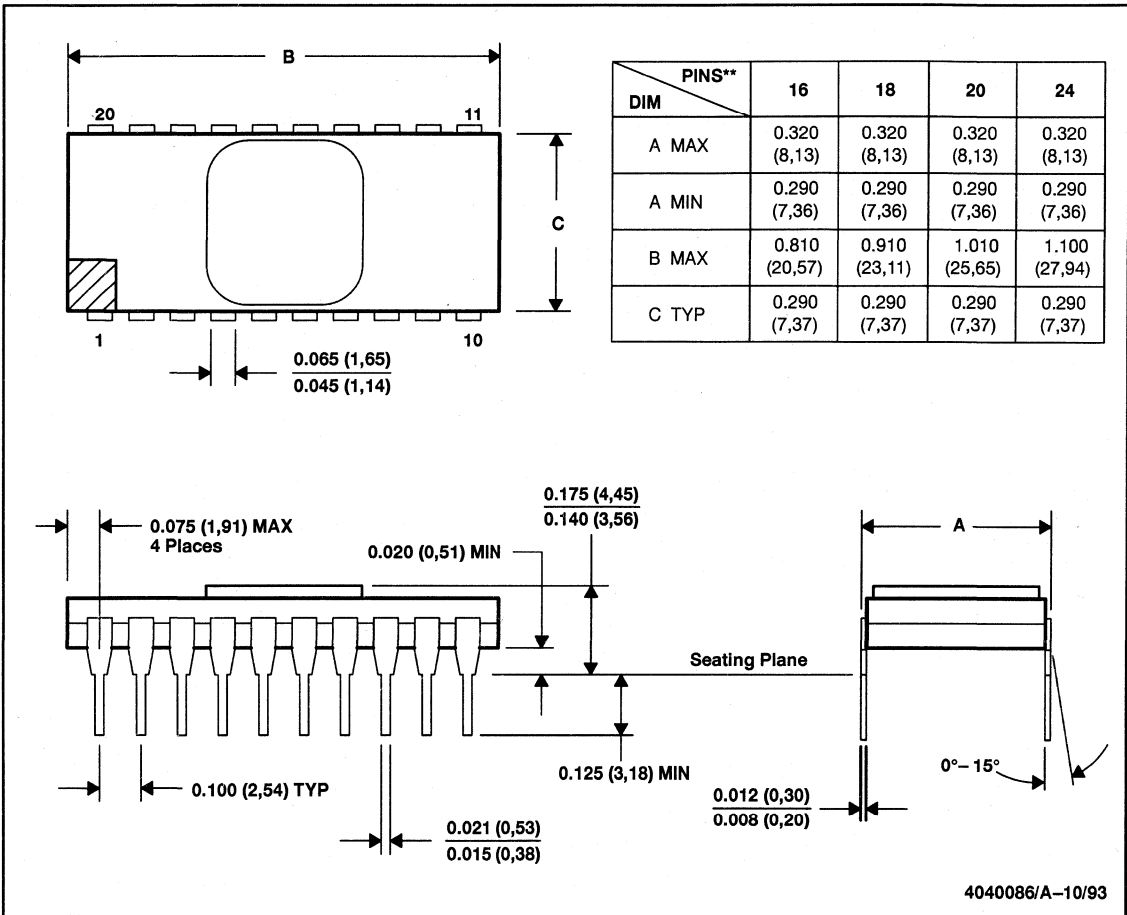
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.  
 D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.

# MECHANICAL DATA

JD/R-CDIP-T\*\*

CERAMIC SIDE-BRAZE DUAL-IN-LINE PACKAGE

20 PIN SHOWN

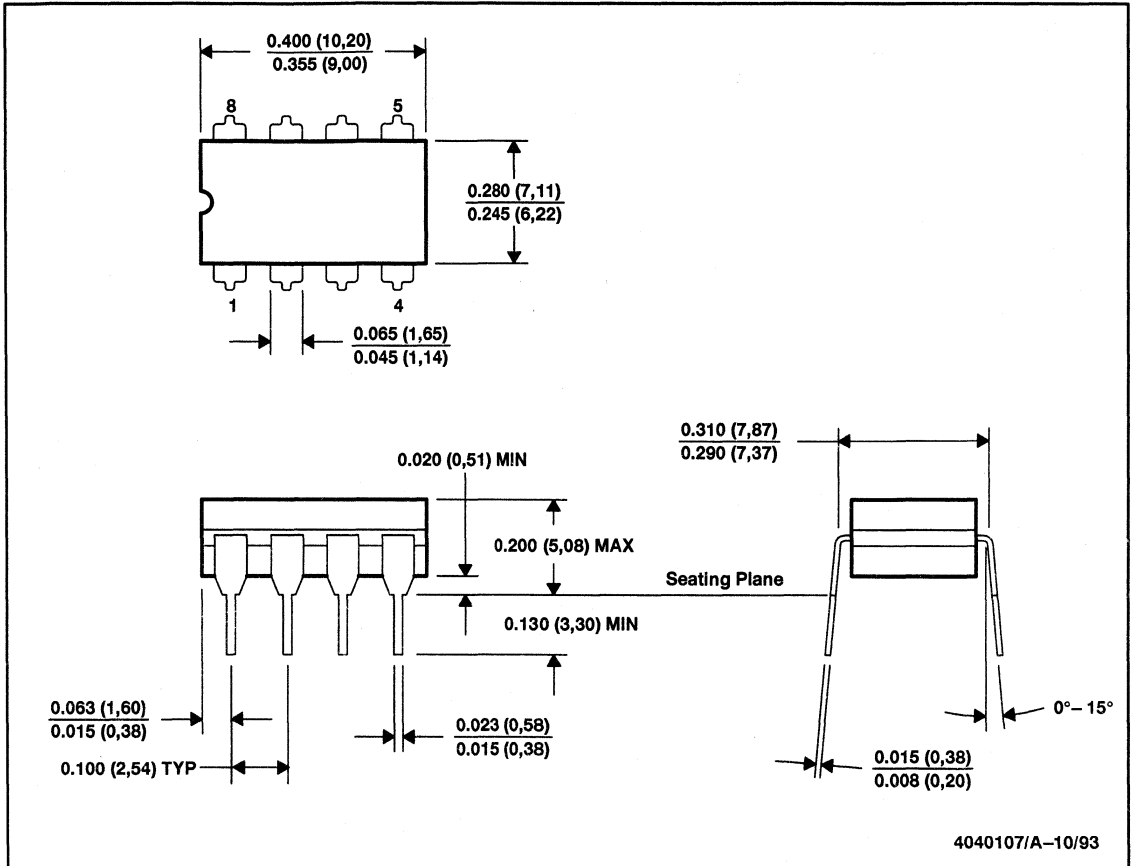


- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a metal lid.  
 D. The pins will be gold plated.



JG/R-GDIP-T8

CERAMIC DUAL-IN-LINE PACKAGE



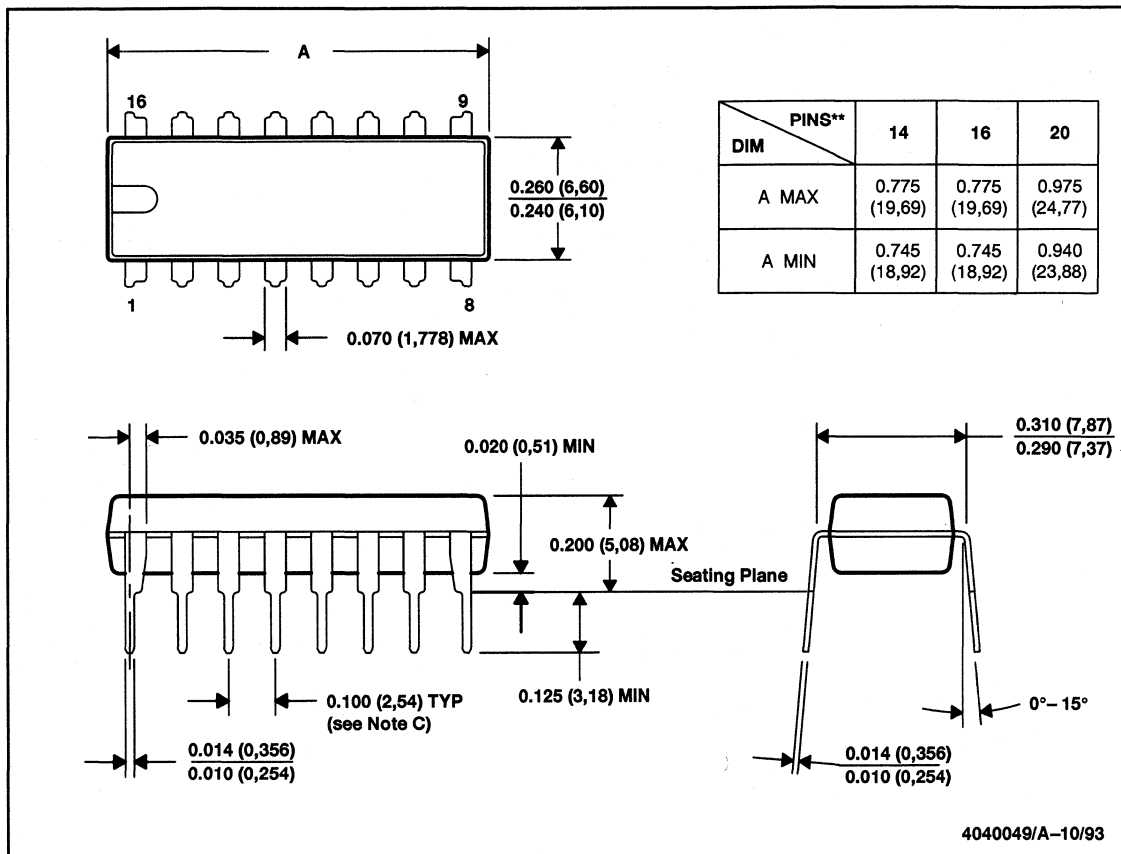
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.  
 D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.

# MECHANICAL DATA

N/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

16 PIN SHOWN



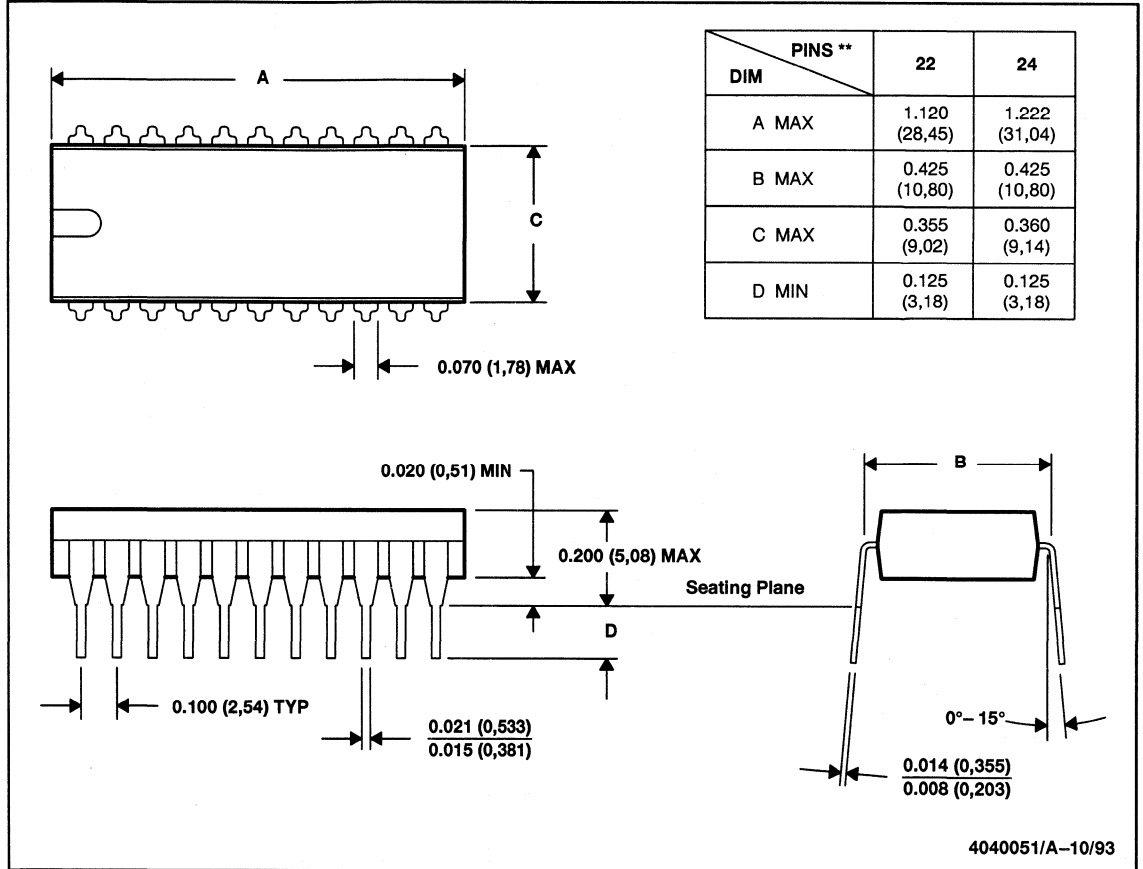
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Each lead centerline is located within 0.010 (0,254) of its true longitudinal position.

4040049/A-10/93

N/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

22 PIN SHOWN



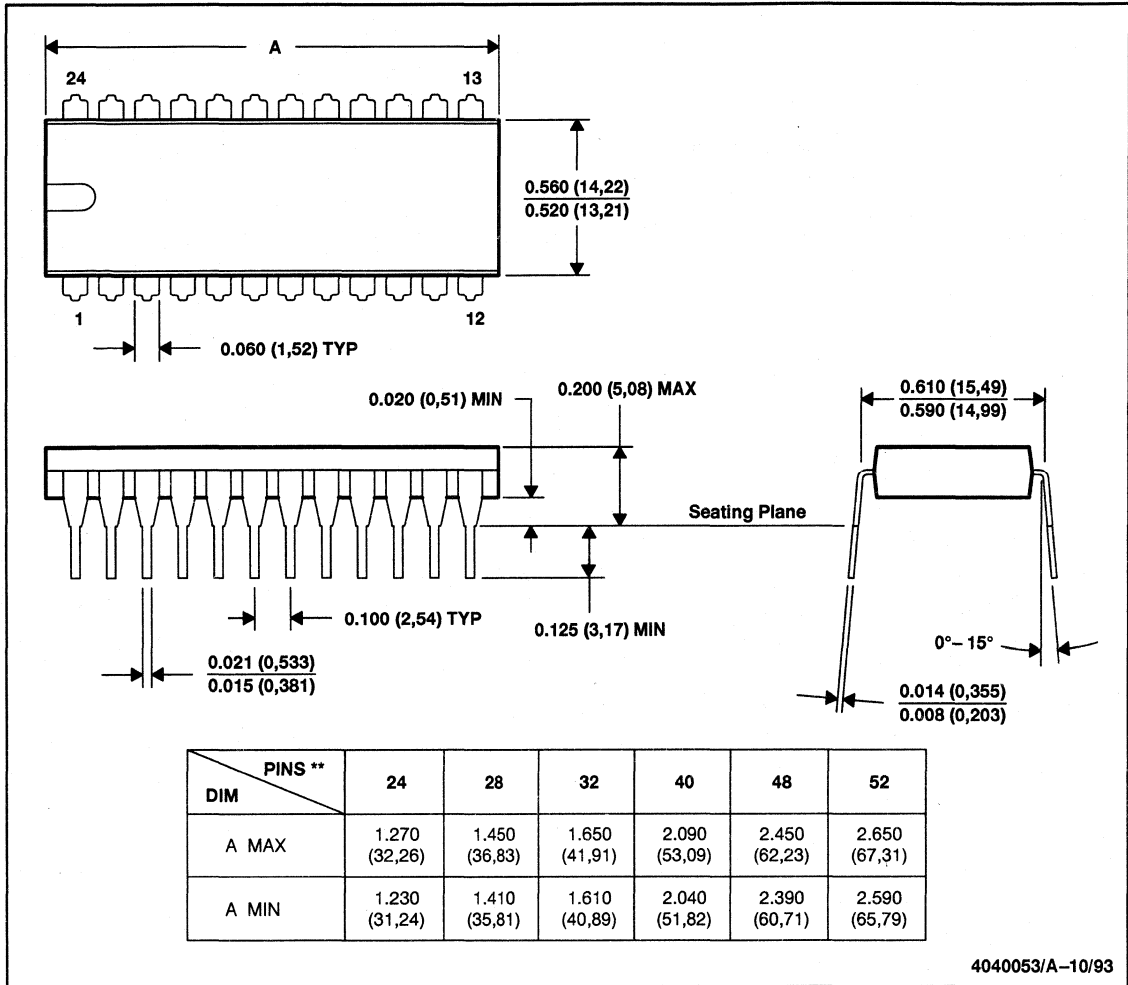
NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.

# MECHANICAL DATA

N/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

24 PIN SHOWN

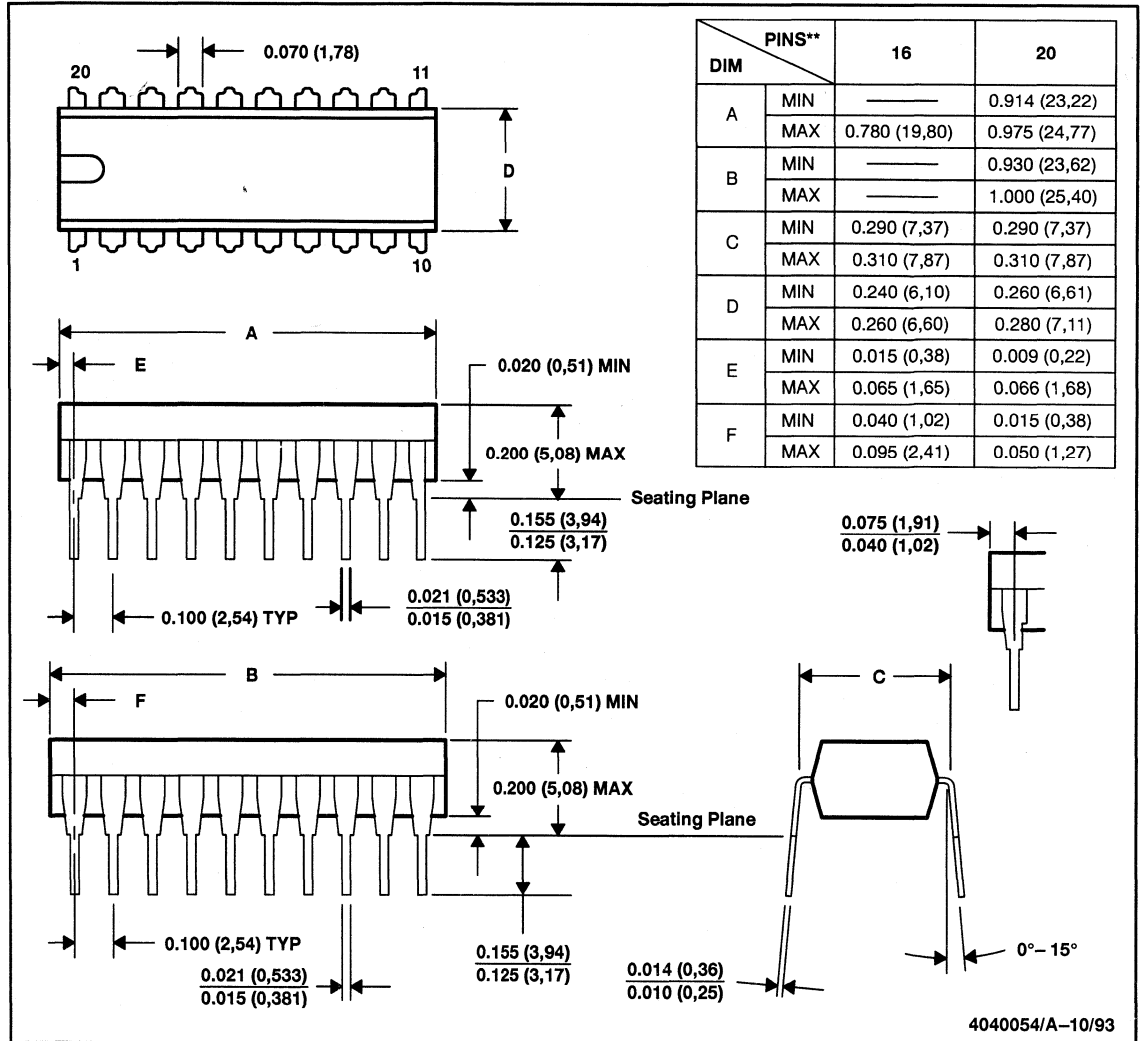


NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.

NE/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

20 PIN SHOWN



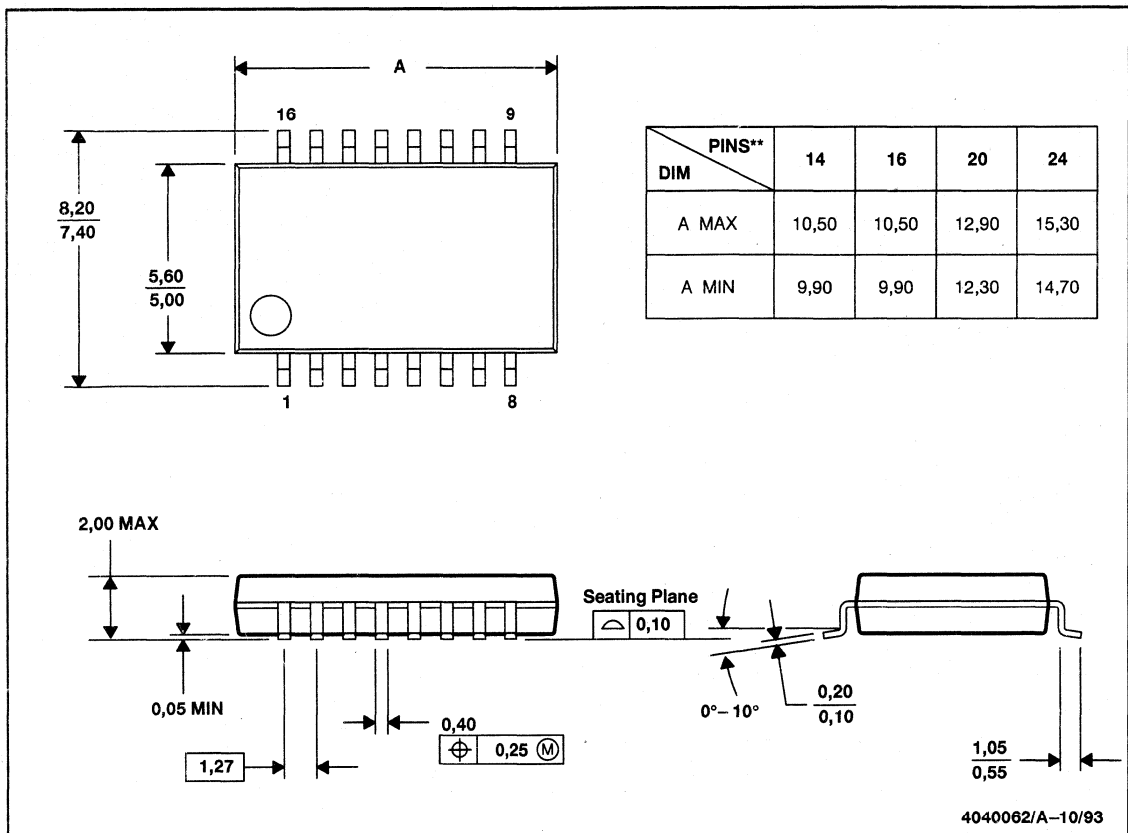
NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.

# MECHANICAL DATA

NS/R-PDSO-G\*\*

PLASTIC SMALL-OUTLINE PACKAGE

16 PIN SHOWN



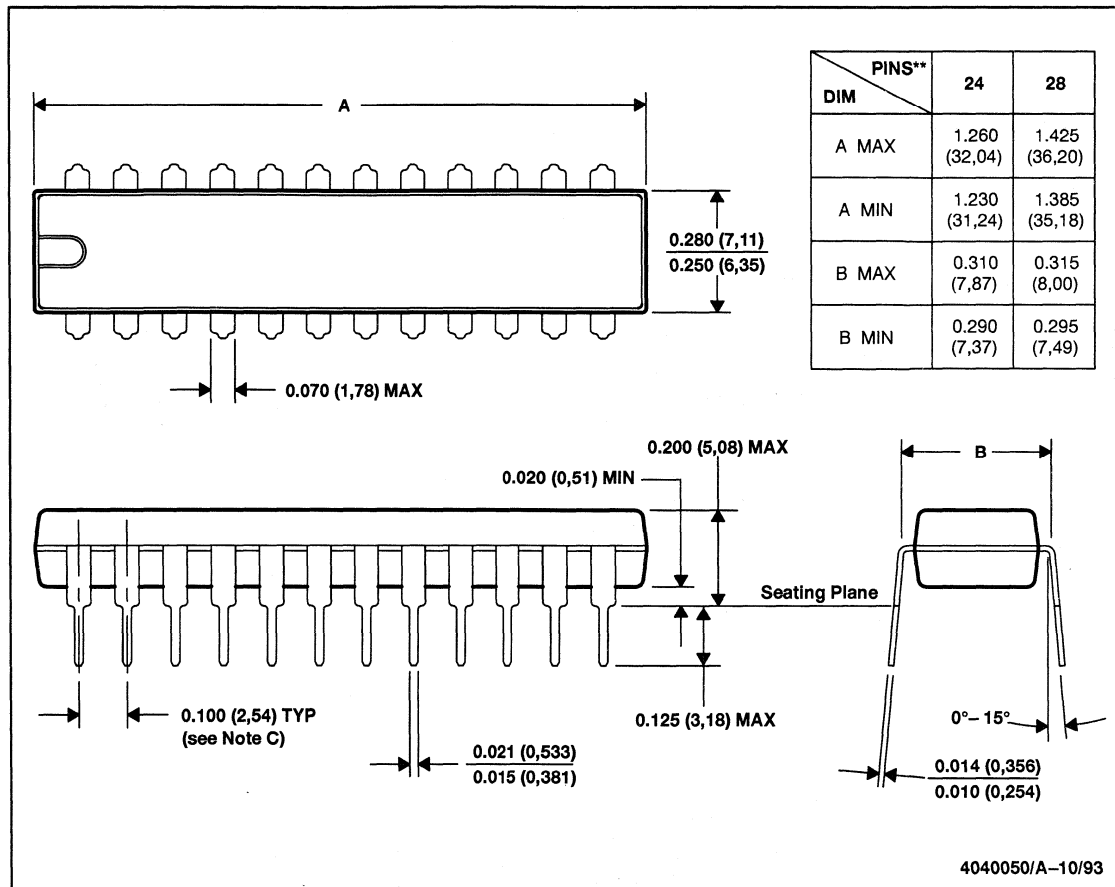
4040062/A-10/93

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

NT/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

24 PIN SHOWN



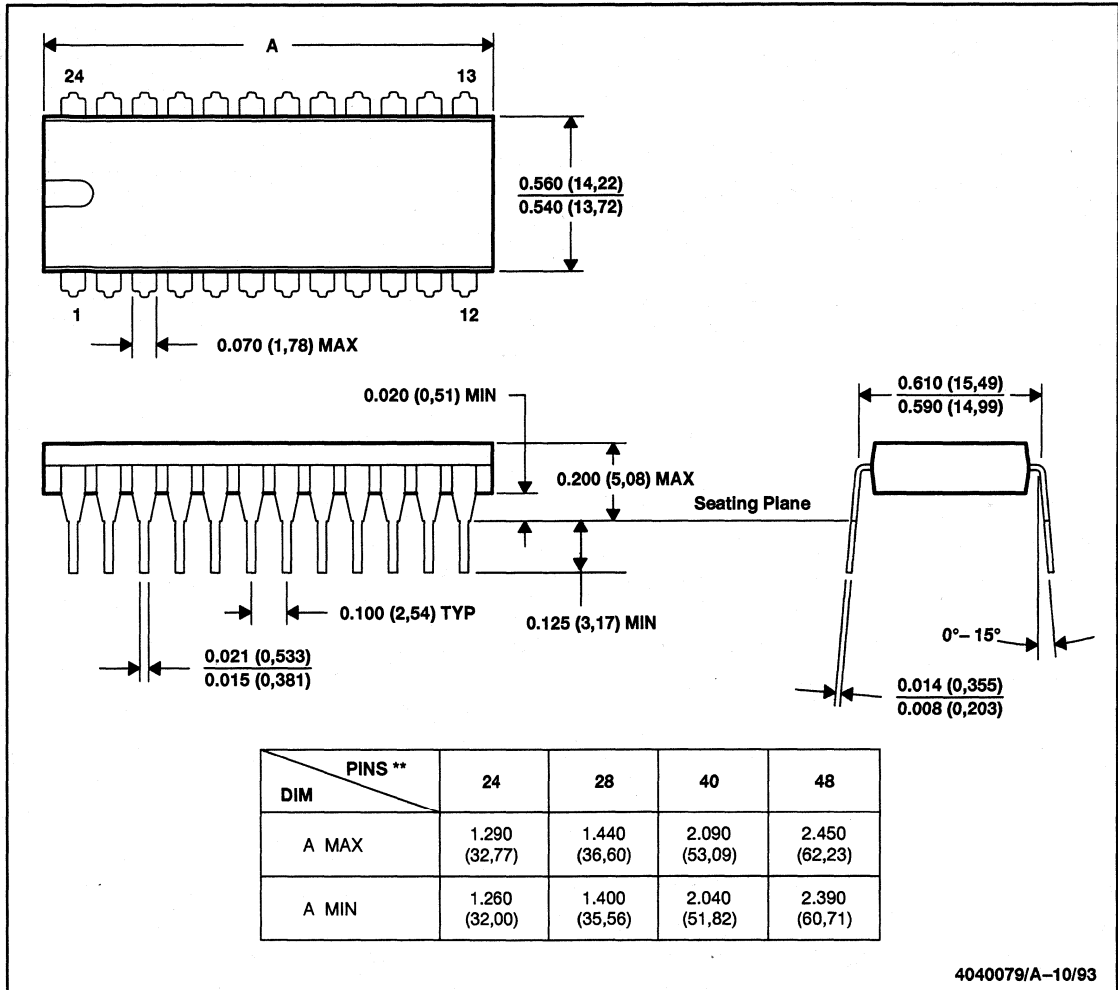
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Each lead centerline is located within 0.010 (0,254) of its true longitudinal position.

# MECHANICAL DATA

NW/R-PDIP-T\*\*

PLASTIC DUAL-IN-LINE PACKAGE

24 PIN SHOWN



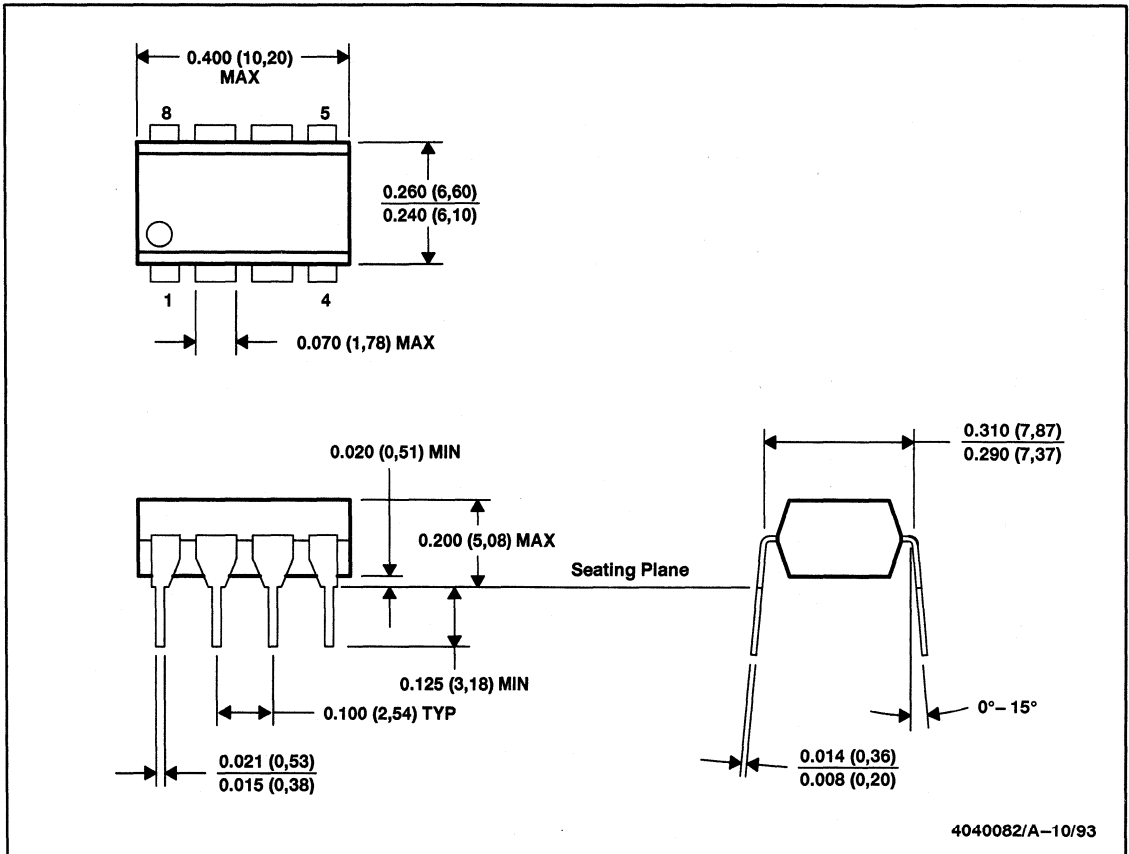
4040079/A-10/93

NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.



P/R-PDIP-T8

PLASTIC DUAL-IN-LINE PACKAGE



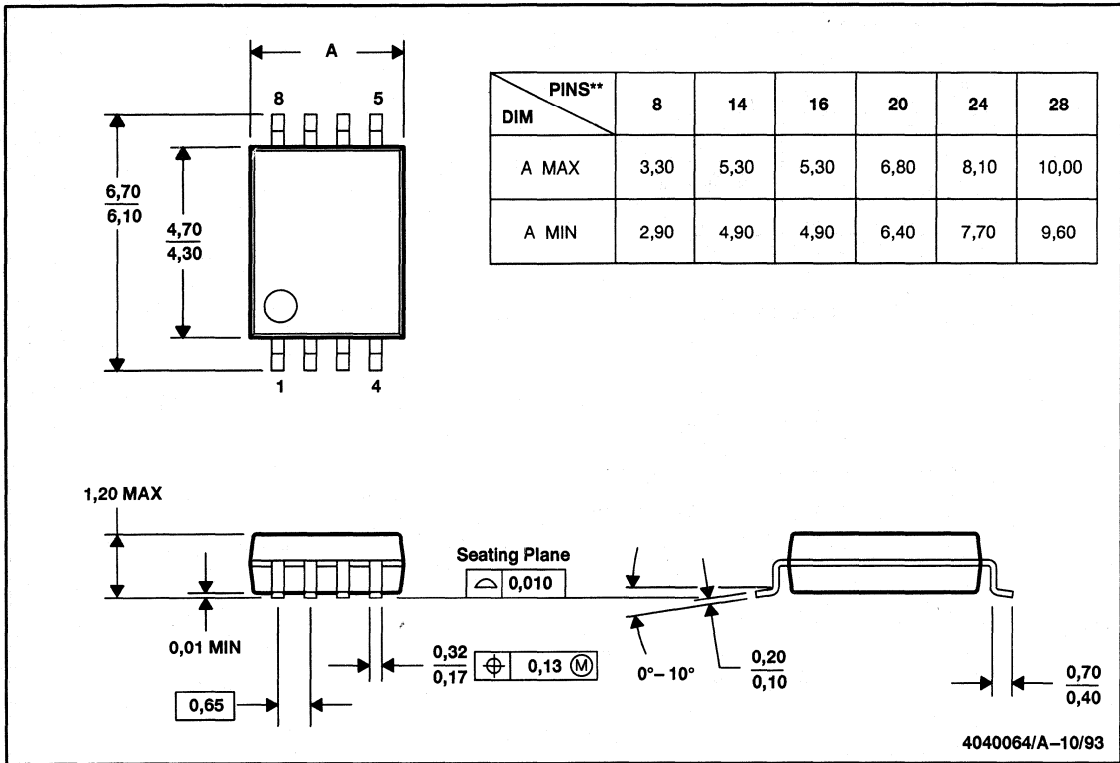
- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.

# MECHANICAL DATA

PW/R-PDSO-G\*\*

PLASTIC THIN SHRINK SMALL-OUTLINE PACKAGE

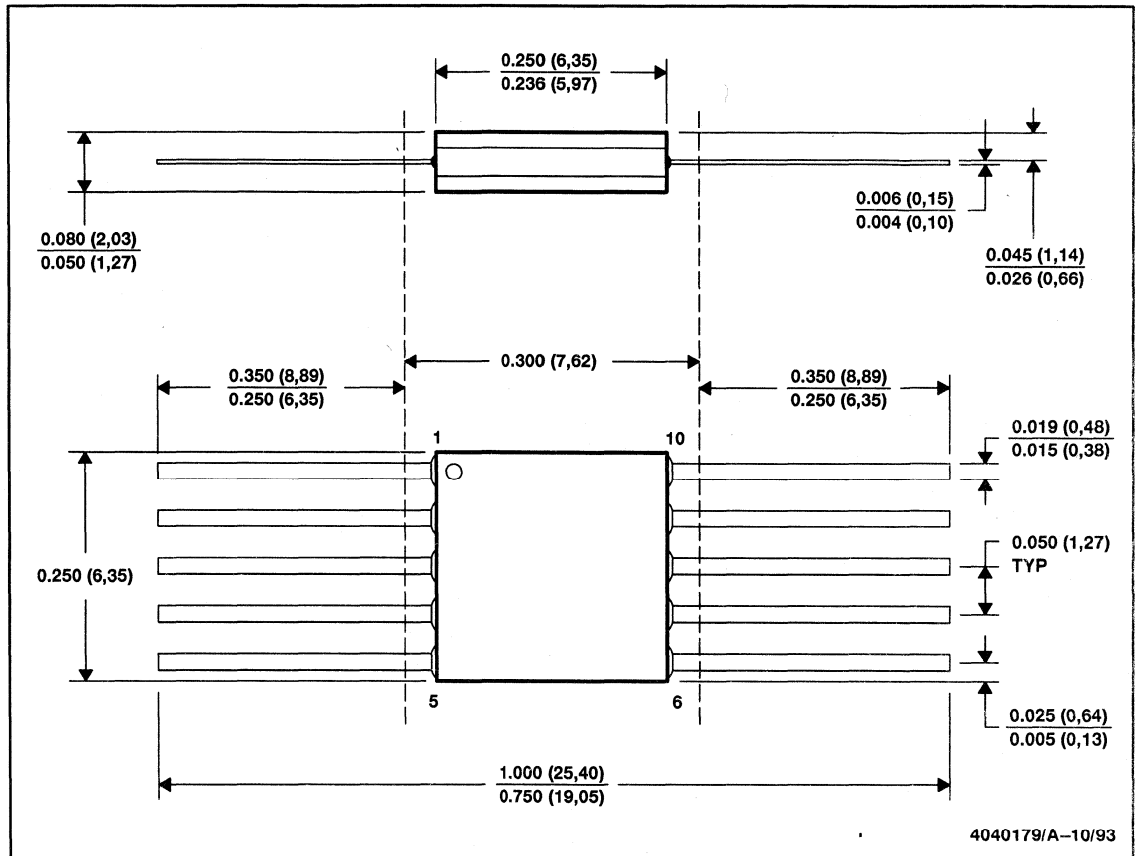
8 PIN SHOWN



- NOTES: A. All linear dimensions are in millimeters.  
 B. This drawing is subject to change without notice.  
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

U/S-GDFP-F10

CERAMIC FLATPACK



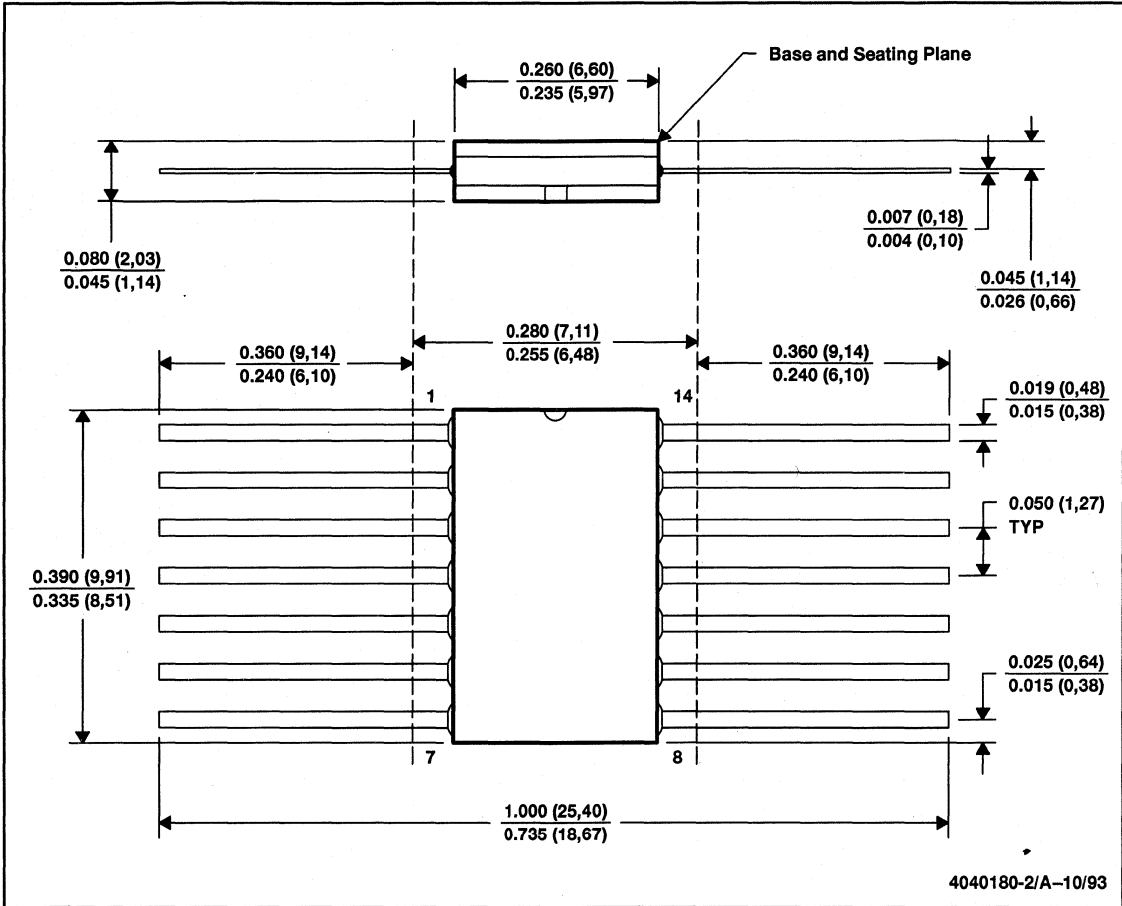
4040179/A-10/93

- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. This package can be hermetically sealed with a ceramic lid using glass frit.  
 D. Falls within JEDEC MO-004AE

# MECHANICAL DATA

W/R-GDFP-F14

CERAMIC FLATPACK



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. This package can be hermetically sealed with a ceramic lid using glass frit.
  - D. Index point is provided on cap for terminal identification only.

# TI Worldwide Sales Offices

**ALABAMA:** Huntsville: 4970 Corporate Drive, NW Suite 125H, Huntsville, AL 35805-6230, (205) 630-0114.

**ARIZONA:** Phoenix: 2525 E. Camelback, Suite 500, Phoenix, AZ 85016, (602) 224-7800.

**CALIFORNIA:** Irvine: 1920 Main Street, Suite 900, Irvine, CA 92714, (714) 860-1200;

**San Diego:** 5625 Ruffin Road, Suite 100, San Jose, CA 92123, (619) 278-9600;

**San Jose:** 2825 North First Street, Suite 200, San Jose, CA 95134, (408) 894-9000;

**Woodland Hills:** 21550 Oxnard Street, Suite 700, Woodland Hills, CA 91367, (818) 704-8100.

**COLORADO:** Aurora: 1400 S. Potomac Street, Suite 101, Aurora, CO 80012, (303) 368-8000.

**CONNECTICUT:** Wallingford: 1062 Barnes Industrial Park Road, Suite 303, Wallingford, CT 06492, (203) 265-3807.

**FLORIDA:** Orlando: 370 S. North Lake Boulevard, Suite 1008, Altamonte Springs, FL 32701, (407) 260-2116;

**Fort Lauderdale:** Hillsboro Center, Suite 110, 600 W. Hillsboro Boulevard, Deerfield Beach, FL 33441, (305) 425-7820; **Tampa:** 4803 George Road, Suite 390, Tampa, FL 33634-6234, (813) 882-0017.

**GEORGIA:** Atlanta: 5515 Spalding Drive, Norcross, GA 30092-2560, (404) 662-7967.

**ILLINOIS:** Arlington Heights: 515 West Algonquin, Arlington Heights, IL 60005, (708) 640-2925.

**INDIANA:** Indianapolis: 550 Congressional Drive, Suite 100, Carmel, IN 46032, (317) 573-6400;

**Fort Wayne:** 103 Airport North Office Park, Fort Wayne, IN 46825, (219) 489-3860.

**KANSAS:** Kansas City: 7300 College Boulevard, Lighton Plaza, Suite 150, Overland Park, KS 66210, (913) 451-4511.

**MARYLAND:** Columbia: 8815 Centre Park Drive, Suite 100, Columbia, MD 21045, (410) 964-2003.

**MASSACHUSETTS:** Boston: Bay Colony Corporate Center, 950 Winter Street, Suite 2800, Waltham, MA 02154, (617) 895-9100.

**MICHIGAN:** Detroit: 33737 W. 12 Mile Road, Farmington Hills, MI 48331, (313) 553-1500.

**MINNESOTA:** Minneapolis: 11000 W. 78th Street, Suite 100, Eden Prairie, MN 55344, (612) 828-9300.

**NEW JERSEY:** Edison: 399 Thornall Street, Edison, NJ 08837-2236, (908) 906-0033.

**NEW MEXICO:** Albuquerque: 3916 Juan Tabo Place NE, Suite 22, Albuquerque, NM 87111, (505) 345-2555.

**NEW YORK:** East Syracuse: 5015 Campuswood Drive, East Syracuse, NY 13057, (315) 463-9291;

**Poughkeepsie:** 300 Westage Business Center, Suite 250, Fishkill, NY 12524, (914) 897-2900;

**Long Island:** 48 South Service Road, Suite 100, Melville, NY 11747, (516) 454-6601;

**Rochester:** 2851 Clover Street, Pittsford, NY 14534, (716) 385-6700.

**NORTH CAROLINA:** Charlotte: 8 Woodlawn Green, Suite 100, Charlotte, NC 28217, (704) 522-5487; **Raleigh:** Highwoods Tower 1, 3200 Beach Leaf Court, Suite 206, Raleigh, NC 27604, (919) 876-2725.

**OHIO:** Cleveland: 23775 Commerce Park Road, Beachwood, OH 44122-5875, (216) 765-7528;

**Dayton:** 4035 Colonel Glenn Highway, Suite 310, Beavercreek, OH 45431-1601, (513) 427-6200.

**OREGON:** Portland: 6700 S.W. 105th Street, Suite 110, Beaverton, OR 97005, (503) 643-6758.

**PENNSYLVANIA:** Philadelphia: 600 W. Germantown Pike, Suite 200, Plymouth Meeting, PA 19482, (215) 825-9500.

**PUERTO RICO:** Hato Rey: 615 Mercantil Plaza Building, Suite 505, Hato Rey, PR 00919, (809) 753-8700.

**TEXAS:** Austin: 12501 Research Boulevard, Austin, TX 78759, (512) 250-6769;

**Dallas:** 7839 Churchill Way, Dallas, TX 75251, (214) 917-1264; **Houston:** 9301 Southwest Freeway, Commerce Park, Suite 360, Houston, TX 77074, (713) 778-6592;

**Midland:** FM 1788 & I-20, Midland, TX 79711-0448, (915) 561-7137.

**UTAH:** Salt Lake City: 2180 South 1300 East, Suite 335, Salt Lake City, UT 54106, (801) 466-8973.

**WISCONSIN:** Milwaukee: 20825 Swenson Street, Suite 900, Waukesha WI 53186, (414) 798-1001.

**CANADA:** Ottawa: 303 Moodie Drive, Suite 1200, Mallon Centre, Nepean, Ontario, Canada K2H 9R4, (613) 725-3201; **Toronto:** 280 Centre Street East, Richmond Hill, Ontario, Canada L4C 1B1, (416) 884-9181; **Montreal:** 9460 Trans Canada Highway, St. Laurent, Quebec, Canada H4S 1R7, (514) 335-8392.

**MEXICO:** Texas Instruments de Mexico S.A. de C.V., Xola 613, Modulo 1-2, Colina del Valle, 03100 Mexico, D.F., 5-639-9740.

**AUSTRALIA (& NEW ZEALAND):** Texas Instruments Australia Ltd., 6-10 Talavera Road, North Ryde (Sydney), New South Wales, Australia 2113, 2-878-9000; 14th Floor, 380 Street, Kilda Road, Melbourne, Victoria, Australia 3000, 3-695-1211.

**BELGIUM:** Texas Instruments Belgium S.A./N.V., Avenue Jules Bordetlaan 11, 1140 Brussels, Belgium, (02) 242 30 80.

**BRAZIL:** Texas Instruments Electronicos do Brasil Ltda., Av. Eng. Luiz Carlos Berrini, 1461, 11 andar, 04571-903, Sao Paulo, SP, Brazil, (11) 535-5133.

**DENMARK:** Texas Instruments A/S, Borupvang 2D, 2750 Ballerup, Denmark, (44) 68 74 00.

**FINLAND:** Texas Instruments OY, Teknikantie 12, 02150 Espoo, Finland, (0) 43 54 20 33.

**FRANCE:** Texas Instruments France, 8-10 Avenue Morane-Saulnier, B.P. 67, 78141 Velizy-Villacoublay Cedex, France, (1) 30 70 10 01.

**GERMANY:** Texas Instruments Deutschland GmbH., Haggertystraße 1, 85358 Freising, Germany, (0816 1) 80-0; Kirchroser Straße 2, 30659 Hannover, Germany, (0511) 90 49 60; Maybachstraße 11, 73760 Ostfildern, Germany, (0711) 34 03 0.

**HONG KONG:** Texas Instruments Hong Kong Ltd., 8th Floor, World Shipping Centre, 7 Canton Road, Kowloon, Hong Kong, 737-0338.

**HUNGARY:** Texas Instruments Representation, Budaörsi u.50, 3rd floor, 1112 Budapest, Hungary, (1) 269 8310.

**INDIA:** Texas Instruments India Private Ltd., AL-Aabeb, 150/1 Infantry Road, Bangalore 560 001, India, (91-80) 226-9007.

**IRELAND:** Texas Instruments Ireland Ltd., 7/8 Harcourt Street, Dublin 2, Ireland, (01) 475 52 33.

**ITALY:** Texas Instruments Italia S.p.A., Centro Direzionale Colleoni, Palazzo Perseo-Via Paracelso 12, 20041 Agrate Brianza (MI), Italy, (039) 63 221; Via Castello della Magliana, 38, 00148 Roma, Italy (06) 657 26 51.

**JAPAN:** Texas Instruments Japan Ltd., Aoyama Fuji Building 3-6-12 Kita-Aoyama Minato-ku, Tokyo, Japan 107, 03-498-12111; MS Shibaura Building 9F, 4-13-23 Shibaura, Minato-ku, Tokyo, Japan 108, 03-769-8700; Nissho-Iwai Building 5F, 2-5-8 Imabashi, Chuo-ku, Osaka, Japan 541, 06-204-1881; Dai-ri Toyota Building Nishi-kan 7F, 4-10-27 Meieki, Nakamura-ku, Nagoya, Japan 450, 052-583-8691; Kanazawa Oyama-cho Daichi Seimei Building 6F, 3-10 Oyama-cho, Kanazawa-shi, Ishikawa, Japan 920, 0762-23-5471; Matsumoto Showa Building 6F, 1-2-11 Fukushi, Matsumoto-shi, Nagano, Japan 390, 0263-33-1060; Daiichi Olympic Tachikawa Building 6F, 1-25-12, Akebono-cho, Tachikawa-shi, Tokyo, Japan 190, 0425-27-6760; Yokohama Business Park East Tower 10F, 134 Goudo-cho, Hodogaya-ku, Yokohama-shi, Kanagawa, Japan 240, 045-338-1220; Nihon Seimei Kyoto Yasaka Building 5F, 643-2, Higashi-Shiokoji-cho, Higashi-iru, Nishinotoh-in, Shiokoji-cho, Shimogyo-ku, Kyoto, Japan 600, 075-341-7713; Sumitomo Seimei Kumagaya Building 8F, 2-44 Yayoi, Kumagaya-shi, Saitama, Japan 360, 0485-22-440; 4262, Aza Takao, Oaza Kawasaki, Hiji-Machi, Havami-Gun, Oita, Japan 879-15, 0977-73-1557.

**KOREA:** Texas Instruments Korea Ltd., 28th Floor, Trade Tower, 159-1, Samsung-Dong, Kangnam-ku Seoul, Korea, 2-551-2800.

**MALAYSIA:** Texas Instruments Malaysia, SDN. BHD., Lot 36.1 #Box 93, Menara Maybank, 100 Jalan Tun Perak, 50050 Kuala Lumpur, Malaysia, 50-3-230-6001.

**NORWAY:** Texas Instruments Norge A/S, P.B. 106, Brn Sveien 3, 0513 Oslo 5, Norway, (02) 264 75 70.

**PEOPLE'S REPUBLIC OF CHINA:** Texas Instruments China Inc., Beijing Representative Office, 7-05 CITIC Building, 19 Jianguomenwai Dajie, Beijing, China, 500-2255, Ext. 3750.

**PHILIPPINES:** Texas Instruments Asia Ltd., Philippines Branch, 14th Floor, Ba-Lepanto Building, 8747 Paseo de Roxas, 1226 Makati, Metro Manila, Philippines, 2-817-6031.

**PORTUGAL:** Texas Instruments Equipamento Electronico (Portugal) LDA., Eng. Frederico Ulricho, 2650 Moreira Da Maia, 4470 Maia, Portugal (2) 948 10 03.

**SINGAPORE (& INDONESIA, THAILAND):** Texas Instruments Singapore (PTE) Ltd., 990 Bendemeer Road, Singapore 1233, (65) 390-7100.

**SPAIN:** Texas Instruments España S.A., c/Gobelas 43, 28023, Madrid, Spain, (1) 372 80 51; Parc Technologic Del Valles, 08290 Cerdanyola, Barcelona, Spain, (3) 31 791 80.

**SWEDEN:** Texas Instruments International Trade Corporation (Sverigefilialen), Box 30, 164 83, Isafjordsgatan 7, Kista, Sweden, (08) 752 58 00.

**SWITZERLAND:** Texas Instruments Switzerland AG, Fliedstrasse 6, CH-8953 Dietikon, Switzerland, (01) 744 2811.

**TAIWAN:** Texas Instruments Taiwan Limited, Taipei Branch, 23th Floor, Sec. 2, Tun Hua S. Road, Taipei 106, Taiwan, Republic of China, (2) 378-6800.

**UNITED KINGDOM:** Texas Instruments Ltd., Manton Lane, Bedford, England, MK41 7PA, (0234) 270 111.



A0294